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FEATURES
Low noise
Voltage noise: 2.3 nV/ $\sqrt{\text{Hz}}$
Current noise: 2 pA/ $\sqrt{\text{Hz}}$
Wide bandwidth
Small signal: 235 MHz (VGAx); 80 MHz (output amplifier)

Large signal: 80 MHz (1 V p-p)

Gain range
0 to 24 dB (input to VGA output)
6 to 30 dB (input to differential output)
Gain scaling: 20 dB/V

DC-coupled
Single-ended input and differential output
Supplies: ± 2.5 V to ± 5 V

Low power: 140 mW per channel at ± 3.3 V

APPLICATIONS
Multichannel data acquisition
Positron emission tomography
Gain trim
Industrial and medical ultrasound
Radar receivers
GENERAL DESCRIPTION

The AD8264 is a quad, linear-in-dB, general-purpose variable gain amplifier (VGA) with a preamplifier (preamp), and a flexible differential output buffer. DC coupling, combined with wide bandwidth, makes this amplifier a very good pulse processor. Each channel includes a single-ended input preamp/VGA section to preserve the wide bandwidth and fast slew rate for low distortion pulse applications. A 6 dB differential output buffer with common-mode and offset adjustments enable direct coupling to most modern high speed analog-to-digital converters (ADCs), using the converter reference output for perfect dc matching levels.

The -3 dB bandwidth of the preamp/VGA is dc to 235 MHz, and the bandwidth of the differential driver is 80 MHz. The floating gain control interface provides a precise linear-in-dB scale of 20 dB/V and is easy to interface to a variety of external circuits. The gain of each channel is adjusted independently, and all channels are referenced to a single pin, GNLO. Combined with a multioutput, digital-to-analog converter (DAC), each section of the AD8264 can be used for active calibration or as a trim amplifier.

Operation from a bipolar power supply enables amplification of negative voltage pulses generated by current-sinking pulses into a grounded load, such as is typical of photodiodes or photo-multiplier tubes (PMT). Delay-free processing of wideband video signals is also possible.

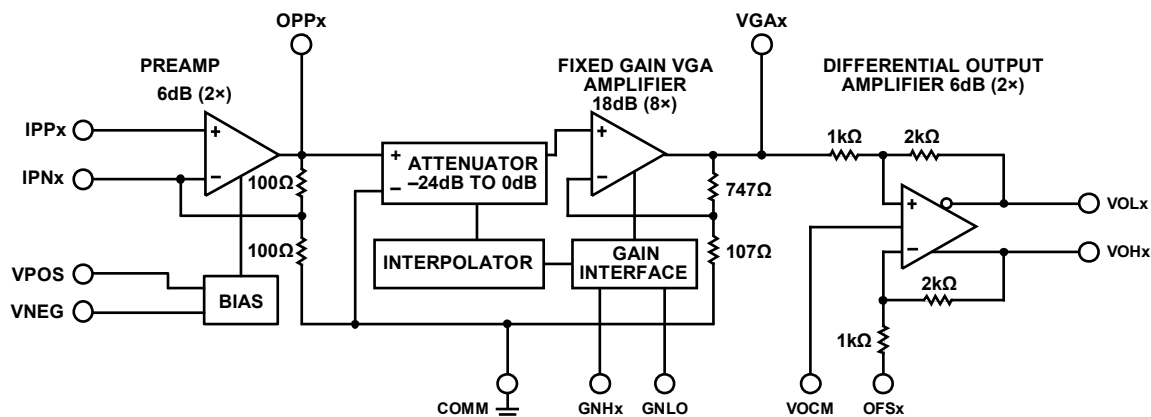
**FUNCTIONAL BLOCK DIAGRAM
ONE CHANNEL SHOWN**


Figure 1.

07796-001

Rev. B
Document Feedback

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AD8264* PRODUCT PAGE QUICK LINKS

Last Content Update: 02/23/2017

COMPARABLE PARTS

View a parametric search of comparable parts.

EVALUATION KITS

- AD8264 Evaluation Board

DOCUMENTATION

Data Sheet

- AD8264: Quad, 235 MHz, DC Coupled VGA and Differential Output Amplifier Data Sheet

DESIGN RESOURCES

- AD8264 Material Declaration
- PCN-PDN Information
- Quality And Reliability
- Symbols and Footprints

DISCUSSIONS

View all AD8264 EngineerZone Discussions.

SAMPLE AND BUY

Visit the product page to see pricing options.

TECHNICAL SUPPORT

Submit a technical question or find your regional support number.

DOCUMENT FEEDBACK

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REVISION HISTORY

1/16—Rev. A to Rev. B

Changes to Features Section, General Description Section, and Figure 1	1
Changes to Figure 2.....	7
Changes to VGA Section	28
Updated Outline Dimensions	37
Changes to Ordering Guide	37

1/11—Rev. 0 to Rev. A

Changes to Figure 1.....	1
Changes to Connecting and Using the AD8264-EVALZ Section and Figure 117	34
Changes to Figure 118	35

5/09—Revision 0: Initial Version

SPECIFICATIONS

$V_S = \pm 2.5$ V, $T_A = 25^\circ\text{C}$, $f = 10$ MHz, $C_L = 5$ pF, $R_L = 500$ Ω per output (VGAX, VOHX, VOLX), $V_{\text{GAIN}} = (V_{\text{GNHX}} - V_{\text{GNLO}}) = 0$ V, $V_{\text{VOCM}} = \text{GND}$, $V_{\text{OFSX}} = \text{GND}$, gain range = 6 dB to 30 dB, unless otherwise specified.

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit	
GENERAL PERFORMANCE						
-3 dB Small Signal Bandwidth (VGAX)	$V_{\text{OUT}} = 10$ mV p-p		235		MHz	
-3 dB Large Signal Bandwidth (VGAX)	$V_{\text{OUT}} = 1$ V p-p		150		MHz	
-3 dB Small Signal Bandwidth (Differential Output) ¹	$V_{\text{OUT}} = 100$ mV p-p		80		MHz	
-3 dB Large Signal Bandwidth (Differential Output) ¹	$V_{\text{OUT}} = 2$ V p-p		80		MHz	
Slew Rate	VGAX, $V_{\text{OUT}} = 2$ V p-p		380		V/ μs	
	VGAX, $V_{\text{OUT}} = 1$ V p-p		290		V/ μs	
	Differential output, $V_{\text{OUT}} = 2$ V p-p		470		V/ μs	
	Differential output, $V_{\text{OUT}} = 1$ V p-p		220		V/ μs	
	Pins IPPx		-8	-5	-3	μA
	Input Bias Current	Pins IPPx at dc; $\Delta V_{\text{IN}}/\Delta I_{\text{BIAS}}$		4.2		M Ω
Input Resistance	Pins IPPx		2		pF	
Input Capacitance	Pins IPPx at 10 MHz		7.9		k Ω	
Input Impedance			2.3		nV/ $\sqrt{\text{Hz}}$	
Input Voltage Noise			2		pA/ $\sqrt{\text{Hz}}$	
Input Current Noise			9		dB	
Noise Figure (Differential Output)	$V_{\text{GAIN}} = 0.7$ V, $R_S = 50$ Ω , unterminated		72		nV/ $\sqrt{\text{Hz}}$	
Output-Referred Noise (Differential Output)	$V_{\text{GAIN}} = 0.7$ V (Gain = 30 dB)		45		nV/ $\sqrt{\text{Hz}}$	
	$V_{\text{GAIN}} = -0.7$ V (Gain = 6 dB)				nV/ $\sqrt{\text{Hz}}$	
Output Impedance	VGAX, dc to 10 MHz		3.5		Ω	
	Differential output, dc to 10 MHz		<1		Ω	
Output Signal Range	Preamp		$ V_S - 1.3$		V	
	VGAX, $R_L \geq 500$ Ω		$ V_S - 1.3$		V	
	Differential amplifier, $R_L \geq 500$ Ω per side		$ V_S - 0.5$		V	
Output Offset Voltage	Preamp offset	-6	<1	+6	mV	
	VGAX offset, $V_{\text{GAIN}} = 0.7$ V	-18	<5	+18	mV	
	Differential output offset, $V_{\text{GAIN}} = 0.7$ V	-38	<10	+38	mV	
DYNAMIC PERFORMANCE						
Harmonic Distortion						
VGAX = 1 V p-p, differential output = 2 V p-p (measured at VGAX)						
	$f = 1$ MHz		-73		dBc	
HD2			-68		dBc	
HD3			-71		dBc	
HD2	$f = 10$ MHz		-61		dBc	
HD3			-60		dBc	
HD2	$f = 35$ MHz		-53		dBc	
HD3					dBc	
VGAX = 1 V p-p, differential output = 2 V p-p (measured at differential output)						
	$f = 1$ MHz		-78		dBc	
HD2			-66		dBc	
HD3			-71		dBc	
HD2	$f = 10$ MHz		-43		dBc	
HD3			-56		dBc	
HD2	$f = 35$ MHz		-20		dBc	
HD3			7		dBm ²	
Input 1 dB Compression Point	$V_{\text{GAIN}} = -0.7$ V, $f = 10$ MHz		-9.6		dBm	
	$V_{\text{GAIN}} = +0.7$ V, $f = 10$ MHz				dBm	

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
Two-Tone Intermodulation Distortion (IMD3)	VGAX = 1 V p-p, f ₁ = 10 MHz, f ₂ = 11 MHz		-68		dBc
	VGAX = 1 V p-p, f ₁ = 35 MHz, f ₂ = 36 MHz		-51		dBc
Output Third-Order Intercept	V _{OUT} = 2 V p-p, f ₁ = 10 MHz, f ₂ = 11 MHz		-49		dBc
	V _{OUT} = 2 V p-p, f ₁ = 35 MHz, f ₂ = 36 MHz		-34		dBc
	VGAX = 1 V p-p, f = 10 MHz		32		dBm
			19		dBV _{RMS}
	VGAX = 1 V p-p, f = 35 MHz		23		dBm
			10		dBV _{RMS}
Overload Recovery	V _{OUT} = 2 V p-p, f = 10 MHz		30		dBm
			17		dBV _{RMS}
	V _{OUT} = 2 V p-p, f = 35 MHz		21		dBm
Group Delay Variation	V _{GAIN} = 0.7 V, V _{IN} stepped from 0.1 V p-p to 1 V p-p		8		dBV _{RMS}
	1 MHz < f < 100 MHz, full gain range		±1		ns
ACCURACY					
Absolute Gain Error ³	-0.7 V < V _{GAIN} < -0.6 V	0	0.2 to 2	3	dB
	-0.6 V < V _{GAIN} < -0.5 V	-1.25	±0.35	+1.25	dB
	-0.5 V < V _{GAIN} < +0.5 V	-1	±0.25	+1	dB
	0.5 V < V _{GAIN} < 0.6 V	-1.25	±0.35	+1.25	dB
	0.6 V < V _{GAIN} < 0.7 V	-3	-0.2 to -2	0	dB
Gain Law Conformance ⁴	-0.5 V < V _{GAIN} < +0.5 V, ±2.5 V ≤ V _S ≤ ±5 V		±0.2		dB
	-0.5 V < V _{GAIN} < +0.5 V, -40°C ≤ T _A ≤ +105°C		±0.3		dB
Channel-to-Channel Matching	Single IC, -0.5 V < V _{GAIN} < +0.5 V, -40°C ≤ T _A ≤ +105°C	-0.5	±0.1 to ±0.25	+0.5	dB
	Multiple ICs, -0.5 V < V _{GAIN} < +0.5 V, -40°C ≤ T _A ≤ +105°C		±0.25		dB
GAIN CONTROL INTERFACE					
Gain Scaling Factor	-0.5 V < V _{GAIN} < +0.5 V	19.5	20.0	20.5	dB/V
	-40°C ≤ T _A ≤ +105°C		20 ± 0.5		dB/V
Gain Range			24		dB
Gain Intercept to VGAX	-40°C ≤ T _A ≤ +105°C	11.5	11.9	12.2	dB
			11.9 ± 0.4		dB
Gain Intercept to Differential Output	-40°C ≤ T _A ≤ +105°C	17.5	17.9	18.2	dB
			17.9 ± 0.4		dB
GNHx Input Voltage Range	GNLO = 0 V, no gain foldover	-V _S		+V _S	V
Input Resistance	ΔV _{IN} /ΔI _{BIAS} , -0.7 V < V _{GAIN} < +0.7 V		70		MΩ
GNHx Input Bias Current	-0.7 V < V _{GAIN} < 0.7 V	-0.9	-0.4	0	μA
	-0.7 V < V _{GAIN} < 0.7 V, -40°C ≤ T _A ≤ +105°C		-0.4 ± 0.2		μA
GNLO Input Bias Current	-0.7 V < V _{GAIN} < 0.7 V		-1.2		μA
	-0.7 V < V _{GAIN} < 0.7 V, -40°C ≤ T _A ≤ +105°C		-1.2 ± 0.4		μA
Response Time	24 dB gain change		200		ns
OUTPUT BUFFER					
VOCM Input Bias Current	-40°C ≤ T _A ≤ +105°C	0.3	1.5	2.5	nA
			1.5 ± 0.3		nA
VOCM Input Voltage Range	OFSx = 0 V, VGAX = 0 V	-1.4		+1.4	V
Gain (VGAX to Differential Output)	-40°C ≤ T _A ≤ +105°C	5.75	6	6.25	dB
			6 ± 0.5		dB

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
POWER SUPPLY					
Supply Voltage		±2.5		±5	V
Power Consumption					
Quiescent Current	$V_S = \pm 2.5\text{ V}$	65	79	88	mA
	$V_S = \pm 2.5\text{ V}, -40^\circ\text{C} \leq T_A \leq +105^\circ\text{C}$		79 ± 25		mA
	$V_S = \pm 3.3\text{ V}$	70	85	95	mA
	$V_S = \pm 3.3\text{ V}, -40^\circ\text{C} \leq T_A \leq +105^\circ\text{C}$		85 ± 30		mA
	$V_S = \pm 5\text{ V}$	81	99	110	mA
	$V_S = \pm 5\text{ V}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}^5$		99 ± 30		mA
Power Dissipation	$V_S = \pm 2.5\text{ V}$		395		mW
	$V_S = \pm 3.3\text{ V}$		560		mW
	$V_S = \pm 5\text{ V}$		990		mW
PSRR	From VPOS to differential output, $V_{\text{GAIN}} = 0.7\text{ V}$		-15		dB
	From VNEG to differential output, $V_{\text{GAIN}} = 0.7\text{ V}$		-15		dB

¹ Differential Output = (VOHx – VOLx).

² All dBm values are calculated with 50 Ω reference, unless otherwise noted.

³ Conformance to theoretical gain expression (see Equation 1 in the Theory of Operation section).

⁴ Conformance to best-fit dB linear curve.

⁵ For supplies greater than ±3.3 V, the operating temperature range is limited to $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$.

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Voltage	
Supply Voltage (VPOS, VNEG)	±6 V
Input Voltage (INPx)	VPOS, VNEG
Gain Voltage (GNHx, GNLO)	VPOS, VNEG
Power Dissipation	2.5 W
Temperature	
Operating Temperature Range	−40°C to +105°C
Storage Temperature Range	−65°C to +150°C
Lead Temperature (Soldering, 60 sec)	300°C
Package Glass Transition Temperature (T _G)	150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

θ_{JA} is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages. The θ_{JA} values in Table 3 assume a 4-layer JEDEC standard board with zero airflow.

Table 3. Thermal Resistance

Package Type	θ_{JA}	θ_{JC}	Unit
40-Lead LFCSP ¹	31.0	2.3	°C/W

¹ 4-Layer JEDEC board (2S2P).

MAXIMUM POWER DISSIPATION

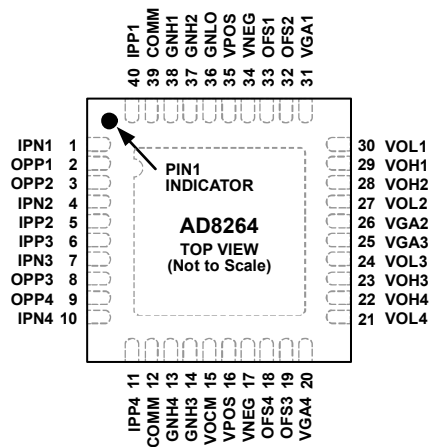
The maximum safe power dissipation for the AD8264 is limited by the associated rise in junction temperature (T_J) on the die. At approximately 150°C, which is the glass transition temperature, the properties of the plastic change. Even temporarily exceeding this temperature limit may change the stresses that the package exerts on the die, permanently shifting the parametric performance of the amplifiers. Exceeding a temperature of 150°C for an extended period can cause changes in silicon devices, potentially resulting in a loss of functionality.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES
 1. EXPOSED PAD (PIN 0) NEEDS AN ELECTRICAL CONNECTION TO GROUND. FOR PROPER RF GROUNDING AND INCREASED RELIABILITY, THE PAD MUST BECONNECTED TO THE GROUND PLANE.

Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
0 (EP), 12, 39	COMM	Ground. Exposed pad (EP, Pin 0) needs an electrical connection to ground. For proper RF grounding and increased reliability, the pad must be connected to the ground plane.
1, 4, 7, 10	IPN1, IPN2, IPN3, IPN4	Negative Preamp Inputs for Channel 1 Through Channel 4. Normally, no external connection is needed.
2, 3, 8, 9	OPP1, OPP2, OPP3, OPP4	Preamp Output for Channel 1 Through Channel 4. This pin is internally connected to the attenuator (VGA) input, and normally, no external connection is needed.
5, 6, 11, 40	IPP1, IPP2, IPP3, IPP4	Positive Preamp Input for Channel 1 Through Channel 4. High impedance.
13, 14, 37, 38	GNH1, GNH2, GNH3, GNH4	Positive Gain Control Voltage Input for Channel 1 Through Channel 4. This pin is referenced to GNLO (Pin 36).
15	VOCM	This pin sets the differential output amplifier (VOHx and VOLx) common-mode voltage.
16, 35	VPOS	Positive Supply (Internally Tied Together).
17, 34	VNEG	Negative Supply (Internally Tied Together).
18, 19, 32, 33	OFS1, OFS2, OFS3, OFS4	Voltage sets the differential output offset for Channel 1 through Channel 4. This is the noninverting input to the differential amplifier, and it has the same bandwidth as the inverting input (VGAx).
20, 25, 26, 31	VGA4, VGA3, VGA2, VGA1	VGA Output for Channel 1 Through Channel 4.
21, 24, 27, 30	VOL1, VOL2, VOL3, VOL4	Negative Differential Amplifier Output for Channel 1 Through Channel 4.
22, 23, 28, 29	VOH1, VOH2, VOH3, VOH4	Positive Differential Amplifier Output for Channel 1 Through Channel 4.
36	GNLO	Negative Gain Control Input (Reference for GNHx Pins).

TYPICAL PERFORMANCE CHARACTERISTICS

$V_S = \pm 2.5\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 10\text{ MHz}$, $C_L = 5\text{ pF}$, $R_L = 500\ \Omega$ per output (V_{GAx} , V_{OHx} , V_{OLx}), $V_{GAIN} = (V_{GNHx} - V_{GNLO}) = 0\text{ V}$, $V_{VOCM} = \text{GND}$, $V_{OFsx} = \text{GND}$, gain range = 6 dB to 30 dB, unless otherwise specified.

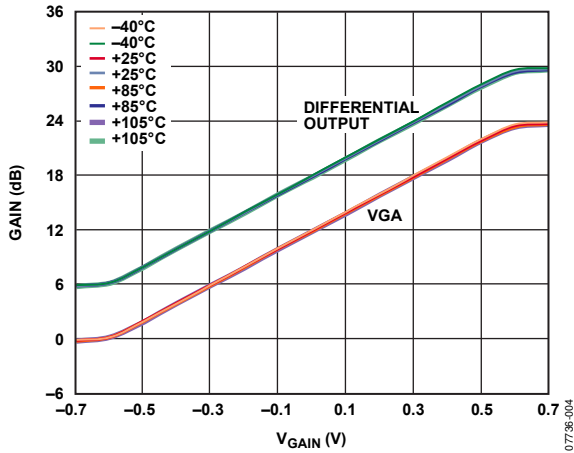


Figure 3. Gain vs. V_{GAIN} vs. Temperature

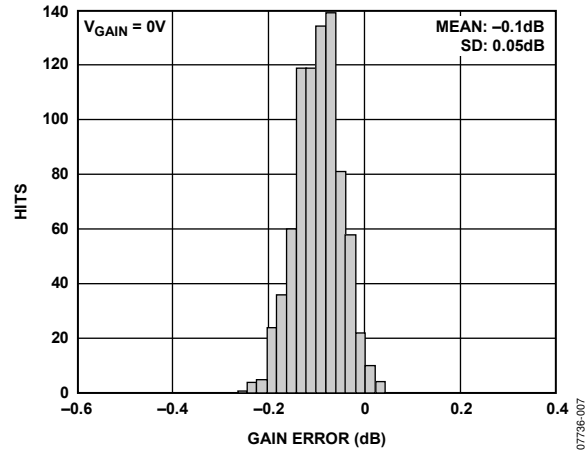


Figure 6. VGA Absolute Gain Error Histogram

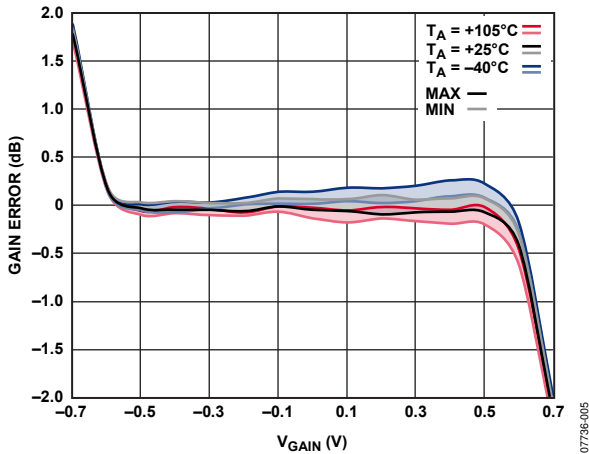


Figure 4. Gain Error vs. V_{GAIN} vs. Temperature

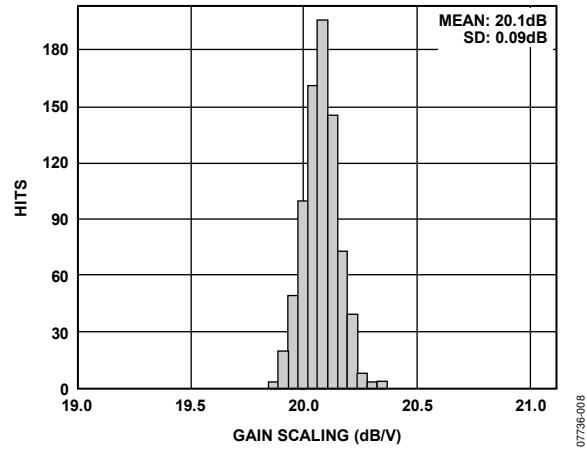


Figure 7. Gain Scale Factor Histogram ($-0.4\text{ V} < V_{GAIN} < +0.4\text{ V}$)

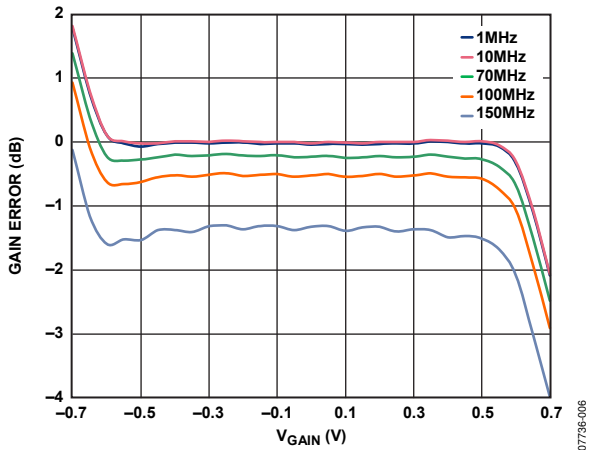


Figure 5. Gain Error vs. V_{GAIN} at Various Frequencies to V_{GAx}

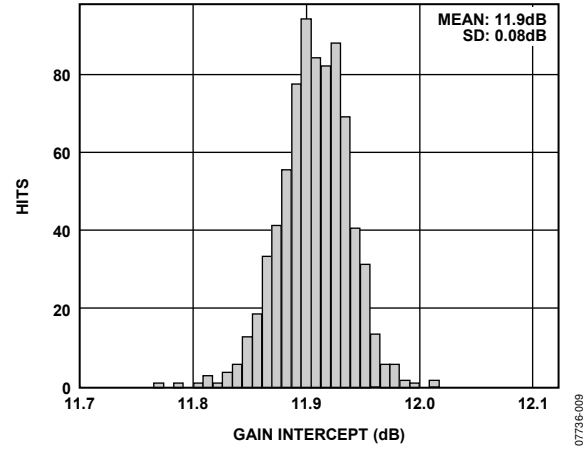


Figure 8. VGA Gain Intercept Histogram

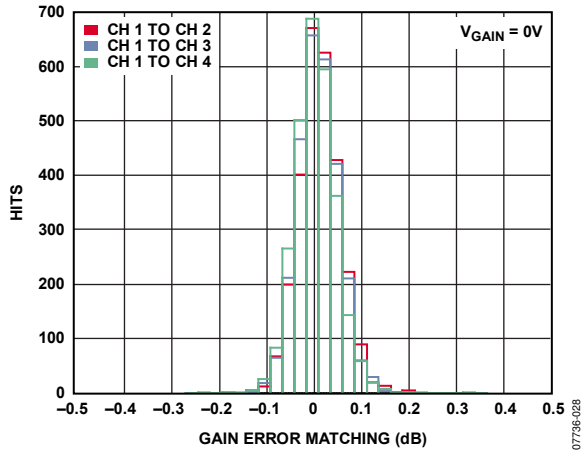


Figure 9. Channel-to-Channel Gain Match Histogram

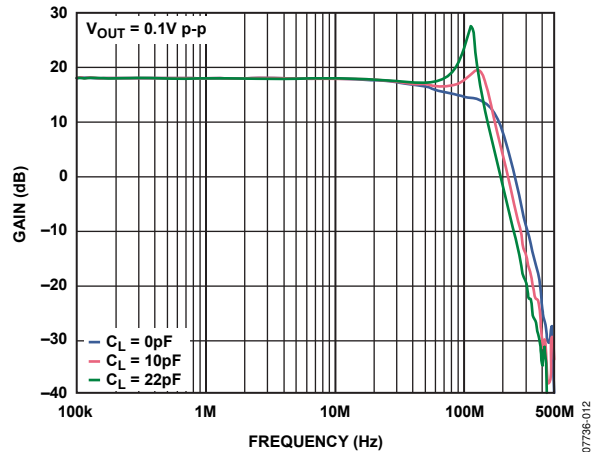


Figure 12. Frequency Response to Differential Output for Various Capacitive Loads

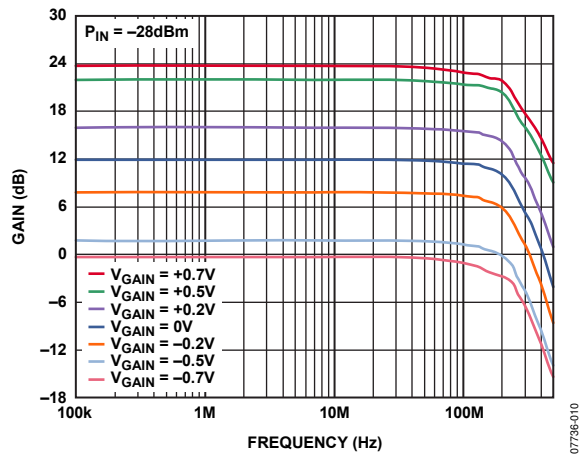


Figure 10. Frequency Response vs. Gain to V_GAx for Various Values of V_GAIN

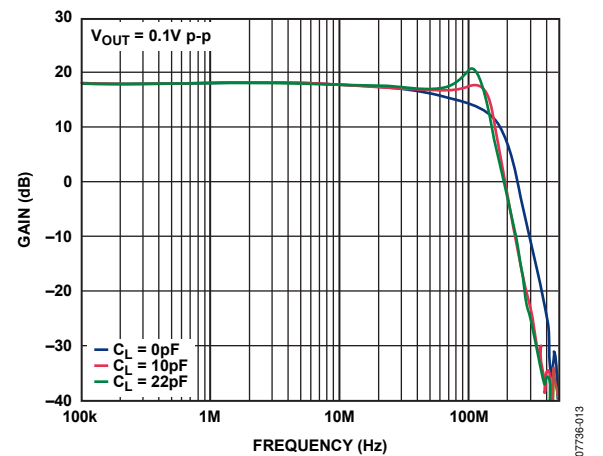


Figure 13. Frequency Response to Differential Output for Various Capacitive Loads with Series R = 10 Ω

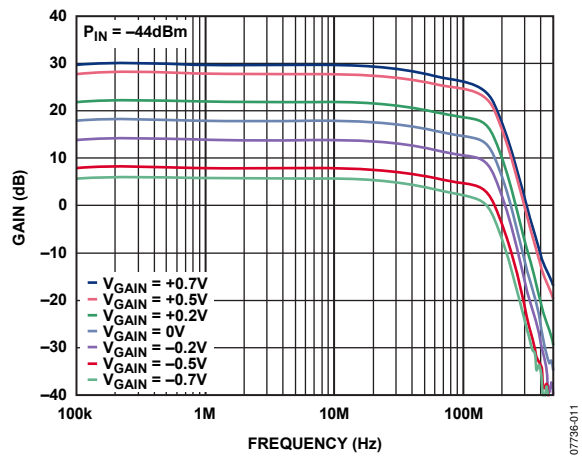


Figure 11. Frequency Response vs. Gain to Differential Output for Various Values of V_GAIN

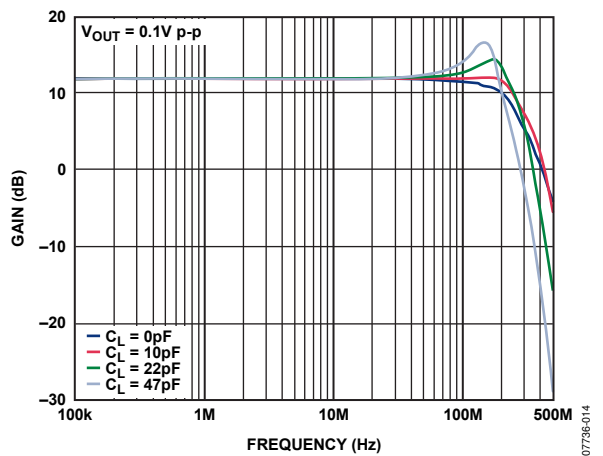


Figure 14. Small Signal Frequency Response to V_GAx for Various Capacitive Loads

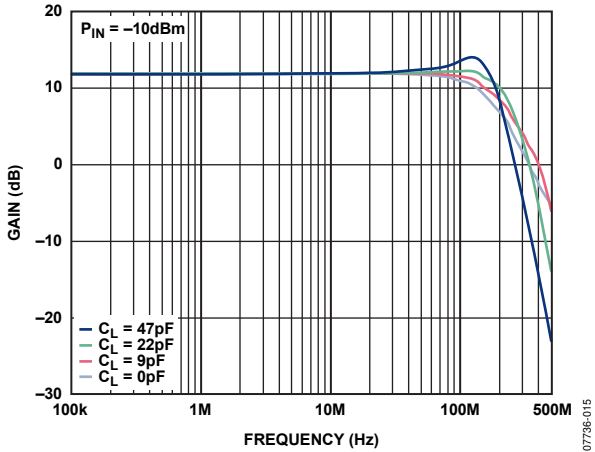


Figure 15. Large Signal Frequency Response to VGAX for Various Capacitive Loads

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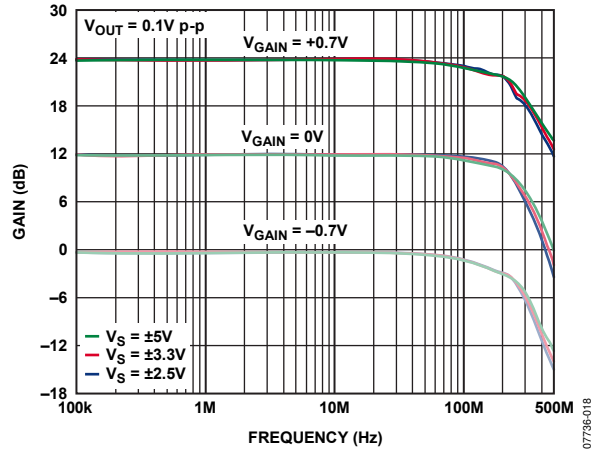


Figure 18. Small Signal Frequency Response vs. Gain to VGAX for Various Supply Voltages

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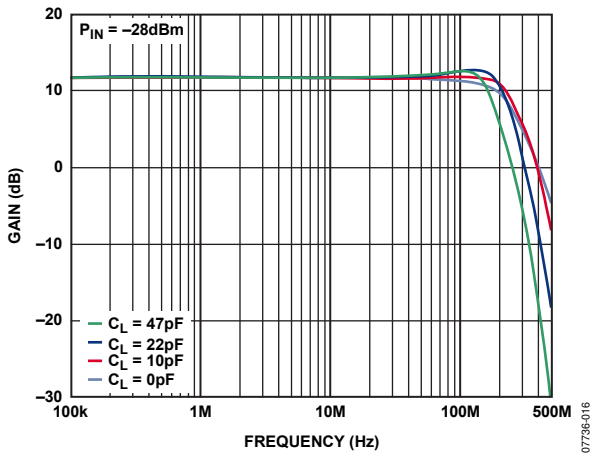


Figure 16. Small Signal Frequency Response to VGAX for Various Capacitive Loads with Series R = 10 Ω

07736-016

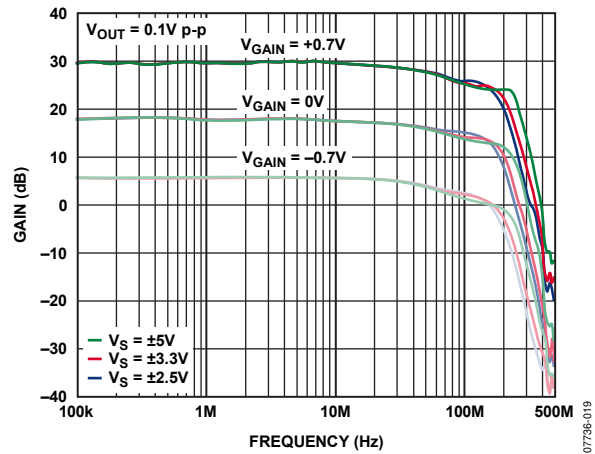


Figure 19. Small Signal Frequency Response vs. Gain to Differential Output for Various Supply Voltages

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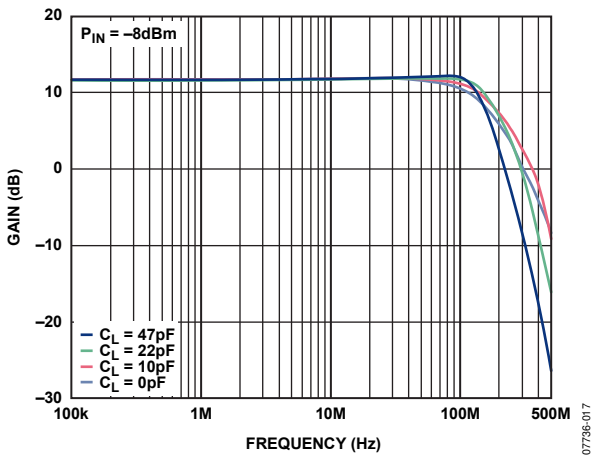


Figure 17. Large Signal Frequency Response to VGAX for Various Capacitive Loads with Series R = 10 Ω

07736-017

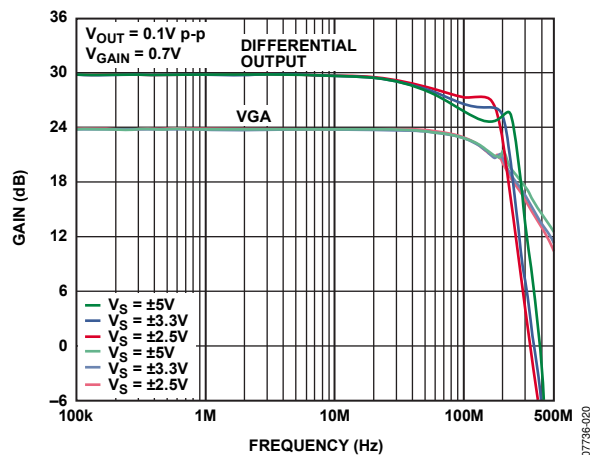


Figure 20. Large Signal Frequency Response to VGAX and Differential Output for Various Supply Voltages

07736-020

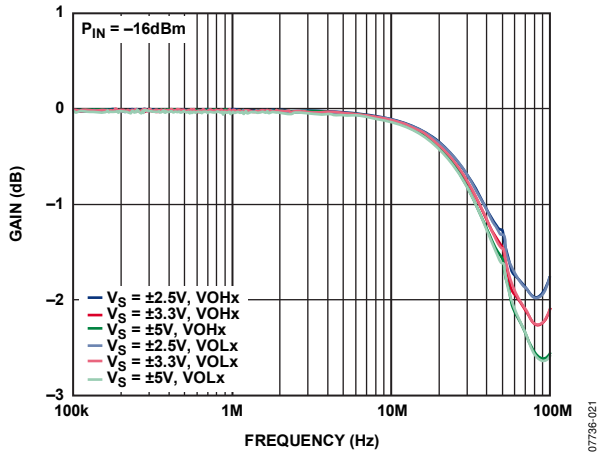


Figure 21. Frequency Response from VOCM to VOHx and VOLx for Various Supplies

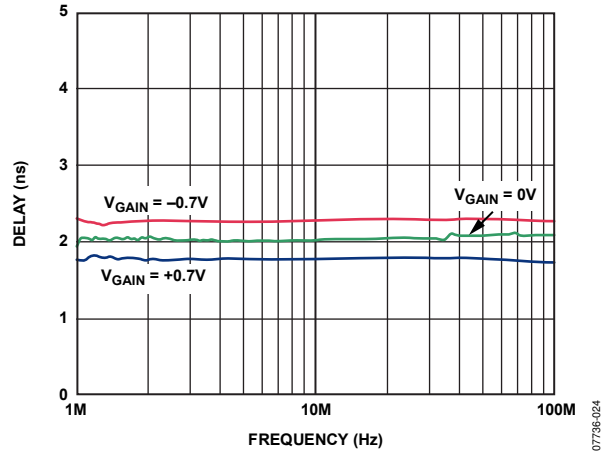


Figure 24. Group Delay vs. Frequency to VGAIN

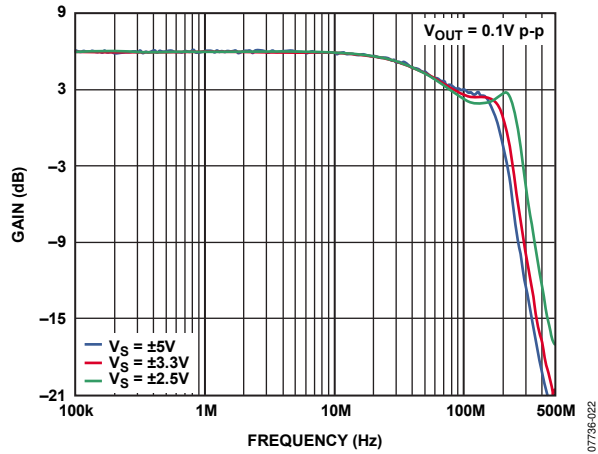


Figure 22. Frequency Response from OFSx to Differential Output for Various Supply Voltages

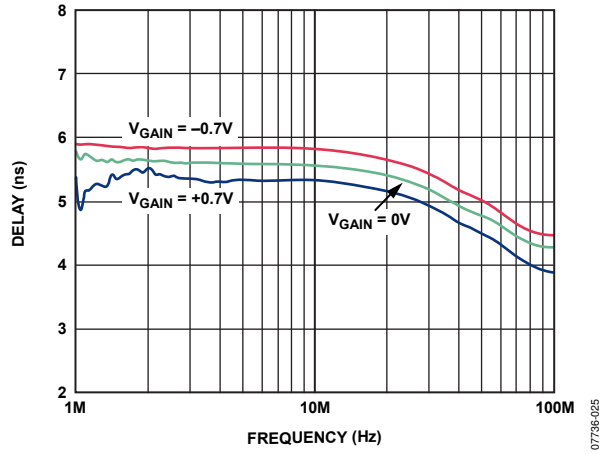


Figure 25. Group Delay vs. Frequency to Differential Output

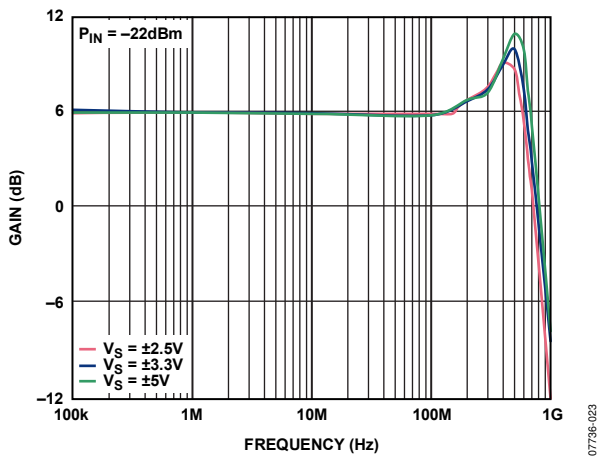


Figure 23. Preamp Frequency Response to OPPx

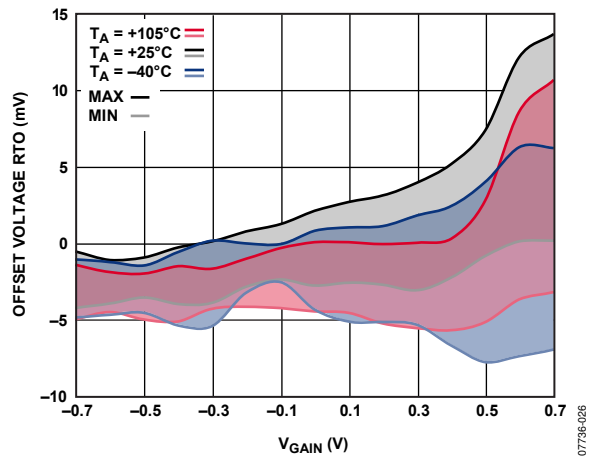


Figure 26. Differential Output Offset Voltage vs. V_GAIN vs. Temperature

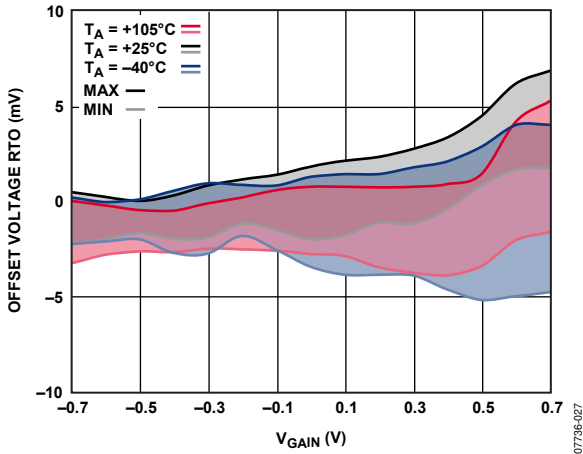


Figure 27. VGAX Output Offset Voltage vs. VGAIN vs. Temperature

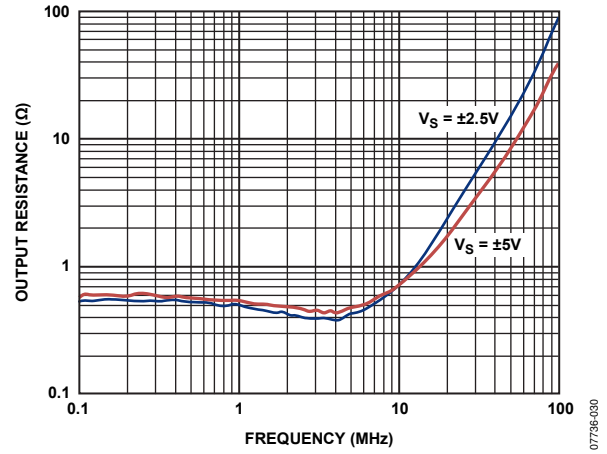


Figure 30. Output Resistance (VOHX, VOLX) vs. Frequency

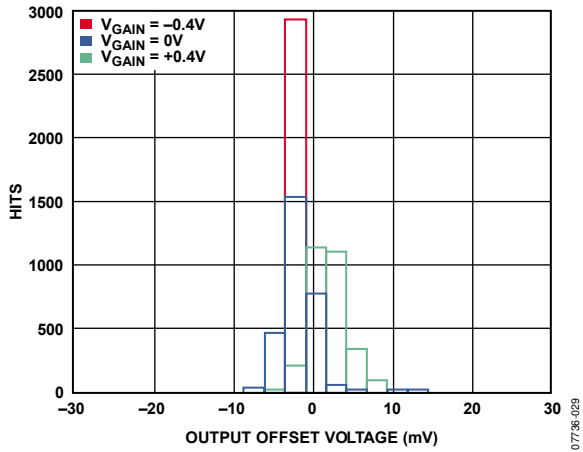


Figure 28. Output Offset Histogram to VGAX

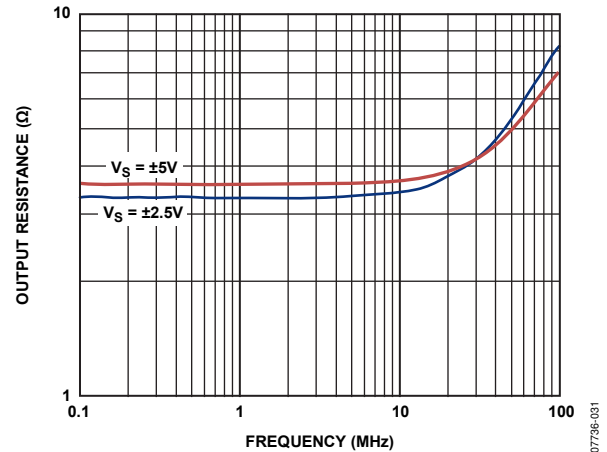


Figure 31. Output Resistance (VGAX) vs. Frequency

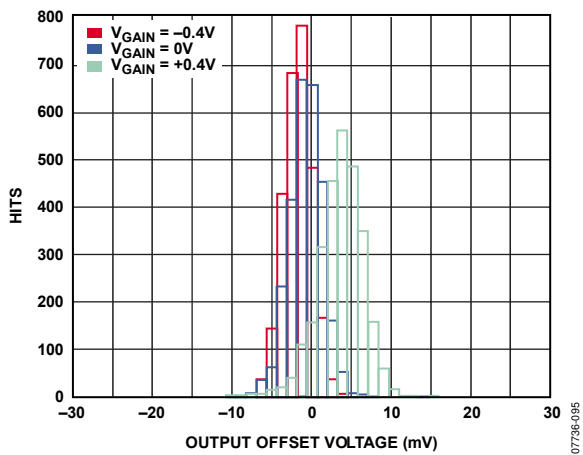


Figure 29. Output Offset Histogram to Differential Output

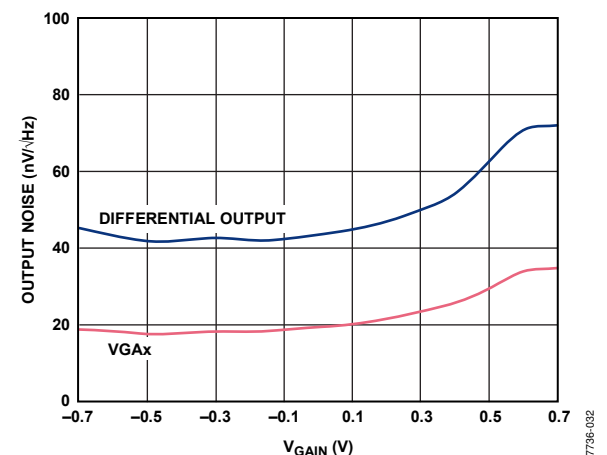


Figure 32. Output Referred Noise to VGAX and Differential Output vs. VGAIN

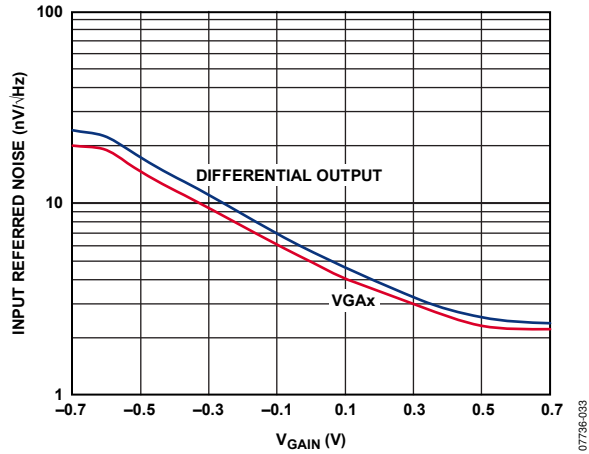


Figure 33. Input Referred Noise from VGAx and Differential Output vs. V_{GAIN}

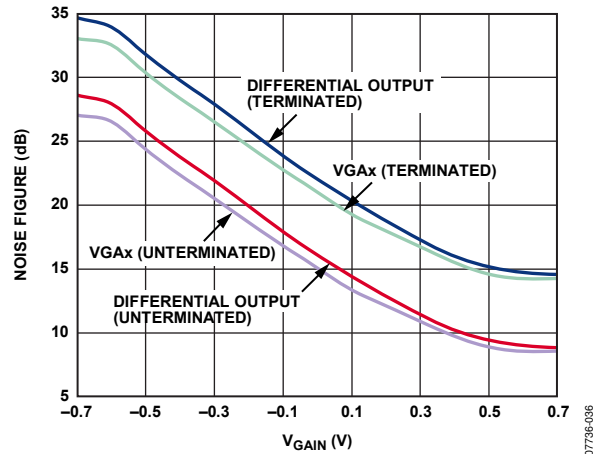


Figure 36. Noise Figure vs. V_{GAIN}

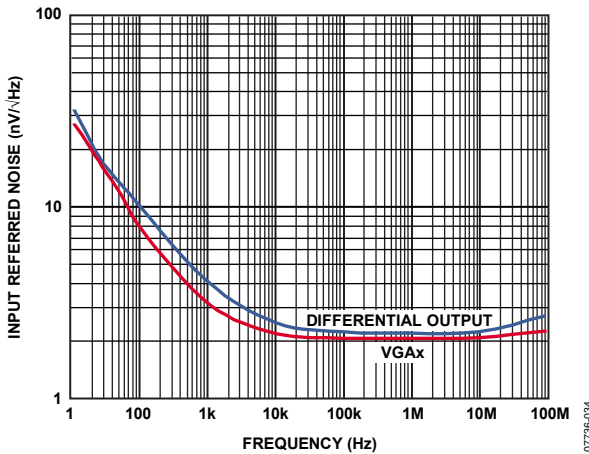


Figure 34. Input Referred Noise vs. Frequency at Maximum Gain

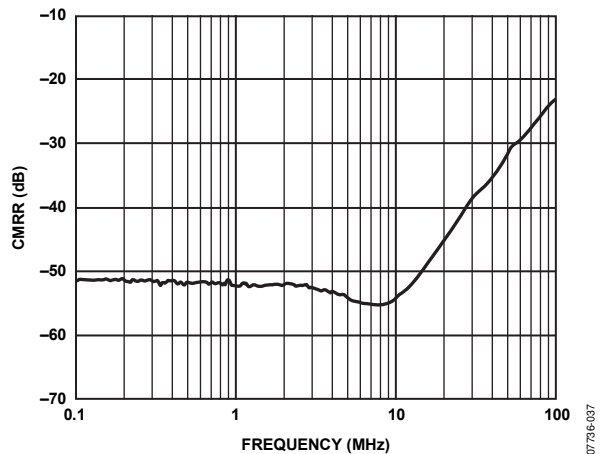


Figure 37. V_OCM Common-Mode Rejection Ratio vs. Frequency

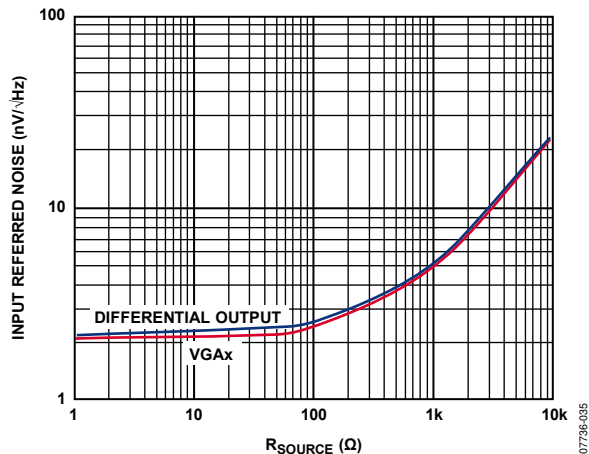


Figure 35. Input Referred Noise vs. R_{SOURCE}

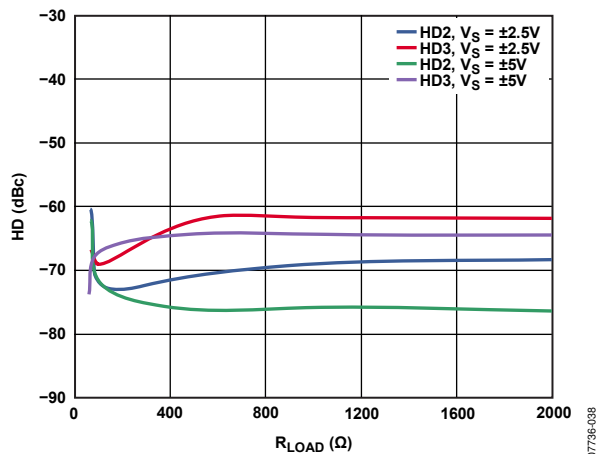


Figure 38. Harmonic Distortion to VGAx vs. R_{LOAD} and Various Supplies

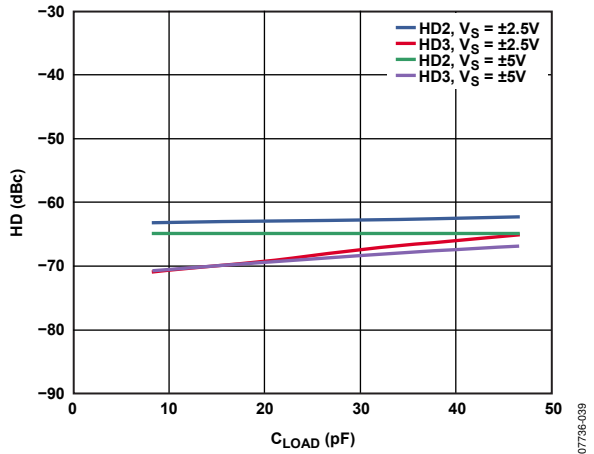


Figure 39. Harmonic Distortion to VGAX vs. C_{LOAD}

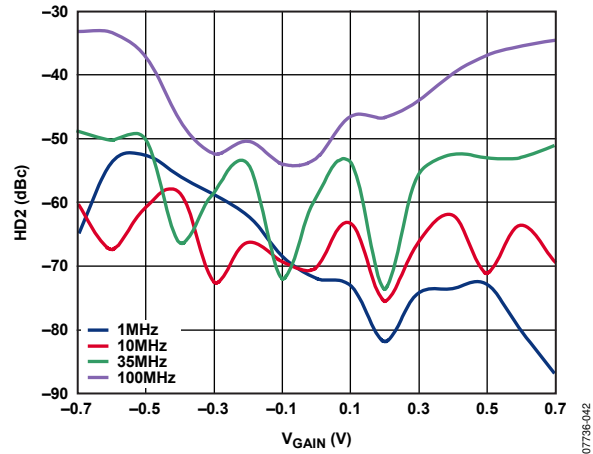


Figure 42. HD_2 vs. V_{GAIN} vs. Frequency to VGAX

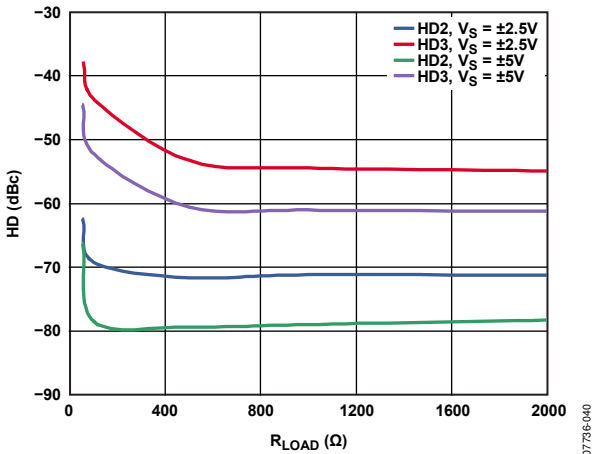


Figure 40. Harmonic Distortion to Differential Output vs. R_{LOAD} and Various Supplies

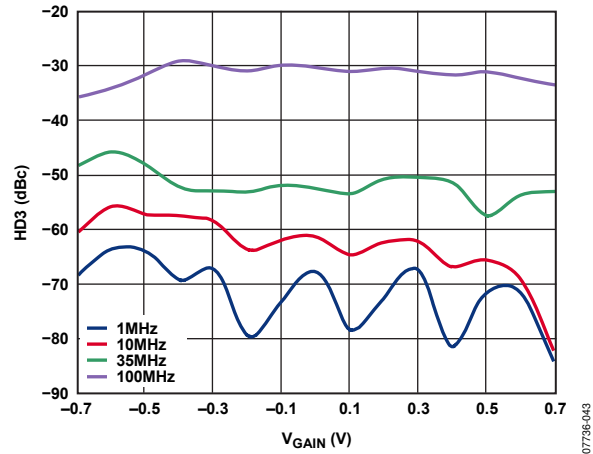


Figure 43. HD_3 vs. V_{GAIN} vs. Frequency to VGAX

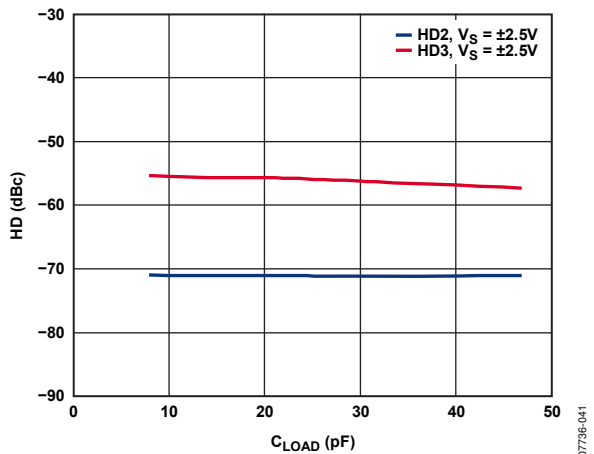


Figure 41. Harmonic Distortion to Differential Output vs. C_{LOAD}

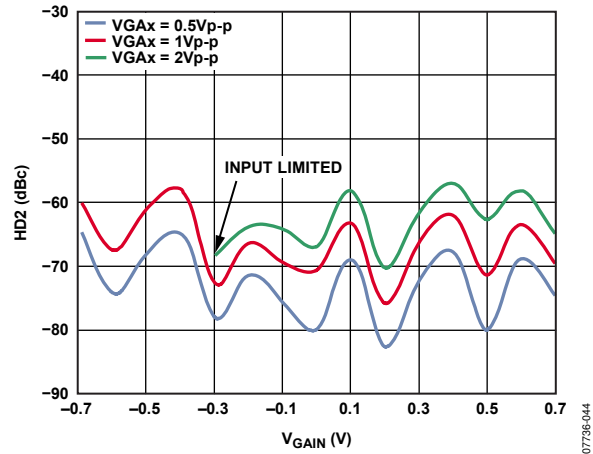


Figure 44. HD_2 vs. Amplitude to VGAX

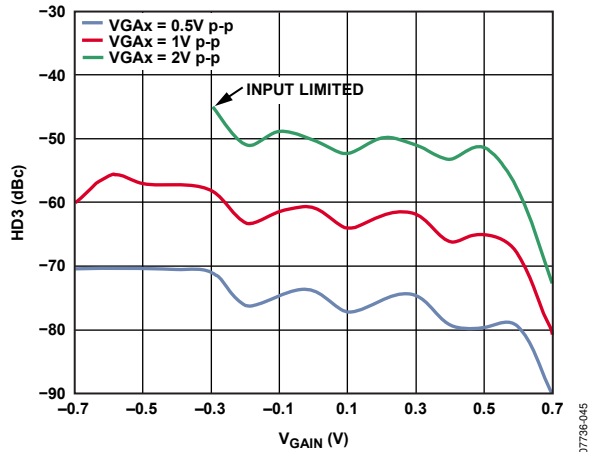


Figure 45. HD3 vs. Amplitude to VGAX

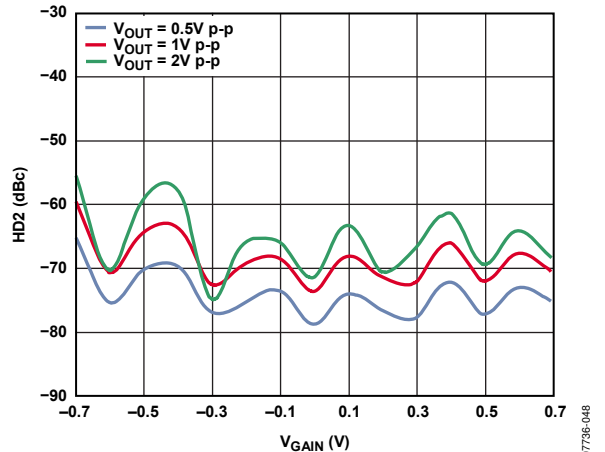


Figure 48. HD2 vs. Amplitude to Differential Output

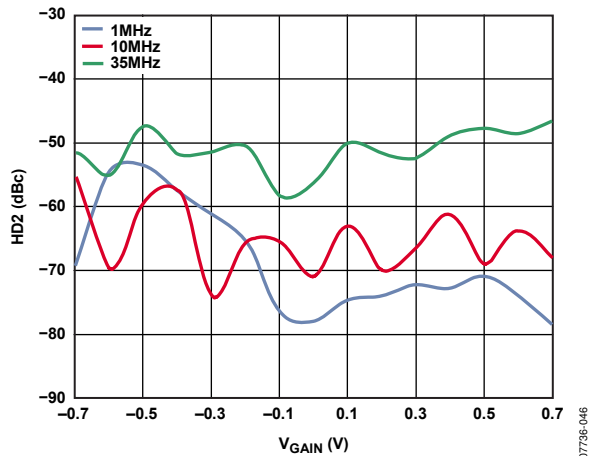


Figure 46. HD2 vs. V_{GAIN} vs. Frequency to Differential Output

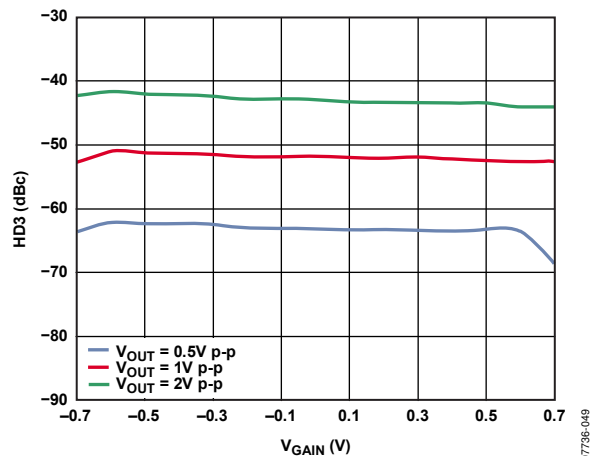


Figure 49. HD3 vs. Amplitude to Differential Output

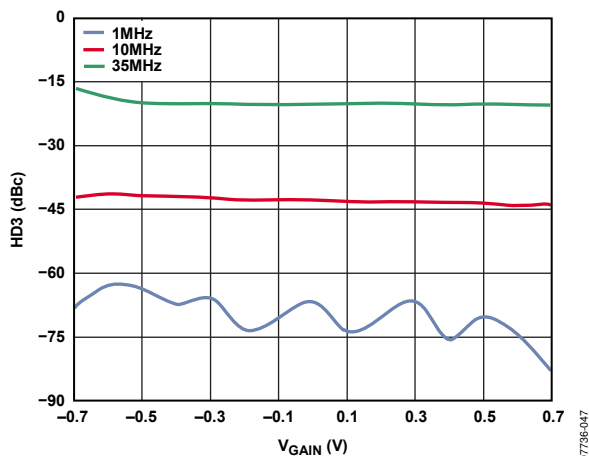


Figure 47. HD3 vs. V_{GAIN} vs. Frequency to Differential Output

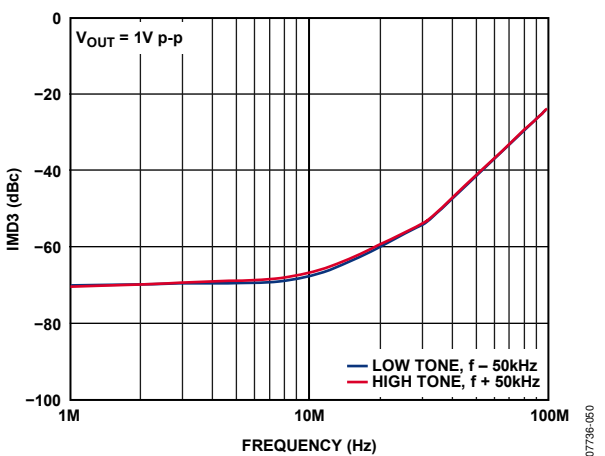


Figure 50. IMD3 vs. Frequency to VGAX

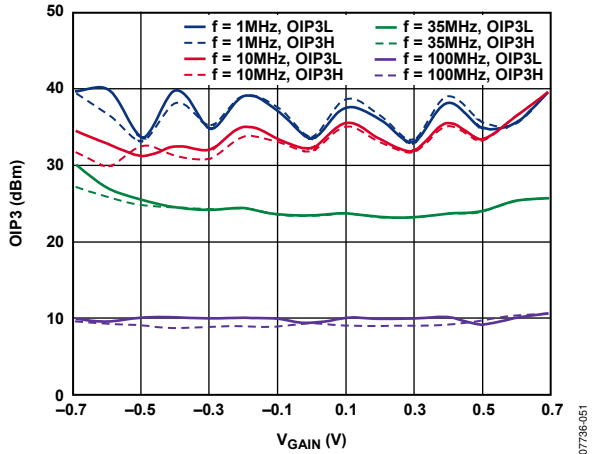


Figure 51. OIP3 vs. V_{GAIN} vs. Frequency to VGAx

07736-051

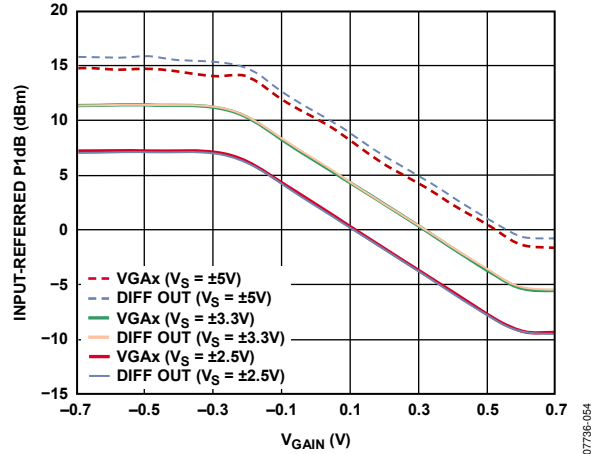


Figure 54. Input P1dB vs. V_{GAIN}

07736-054

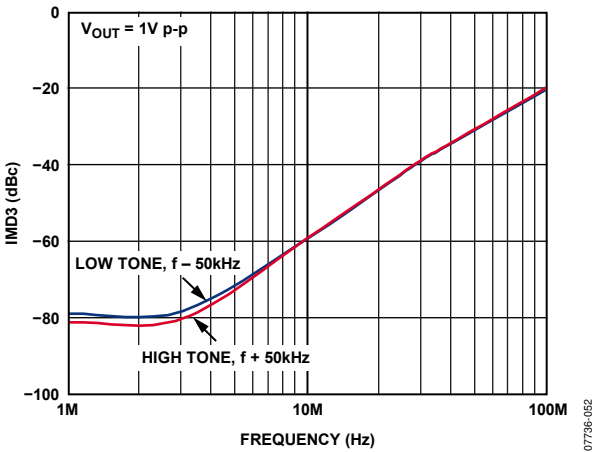


Figure 52. IMD3 vs. Frequency to Differential Output

07736-052

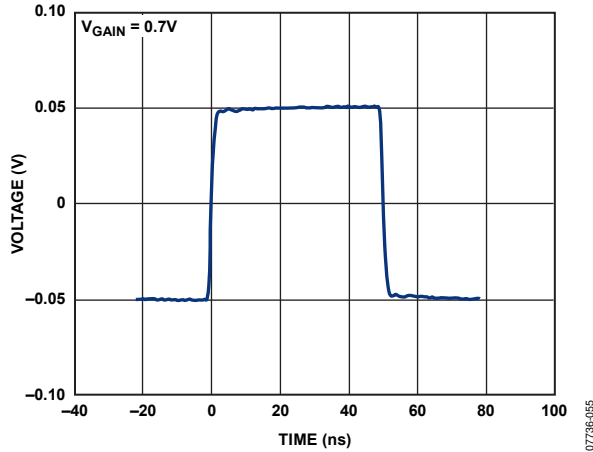


Figure 55. Small Signal Pulse Response to VGAx

07736-055

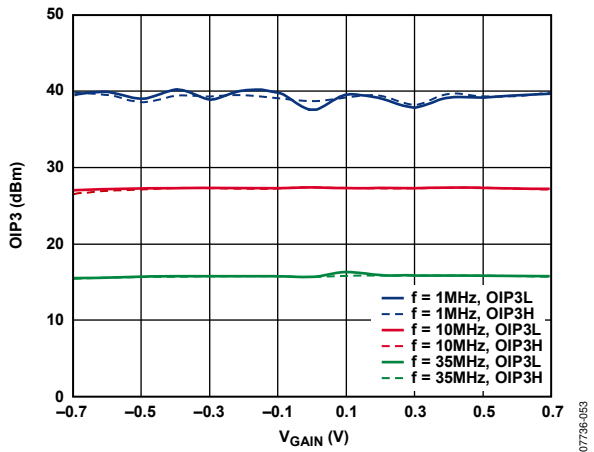


Figure 53. OIP3 vs. Frequency to Differential Output

07736-053

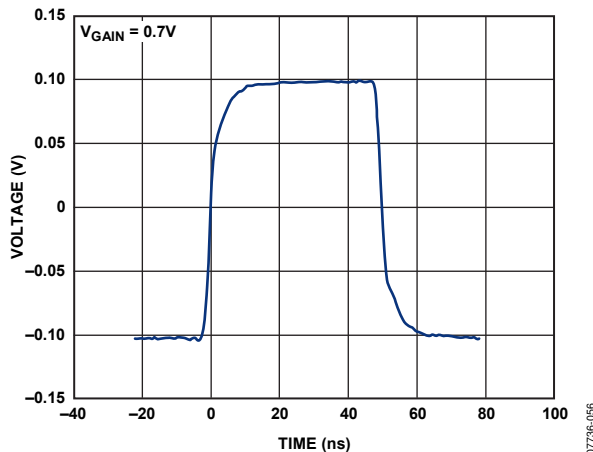


Figure 56. Small Signal Pulse Response to Differential Output

07736-056

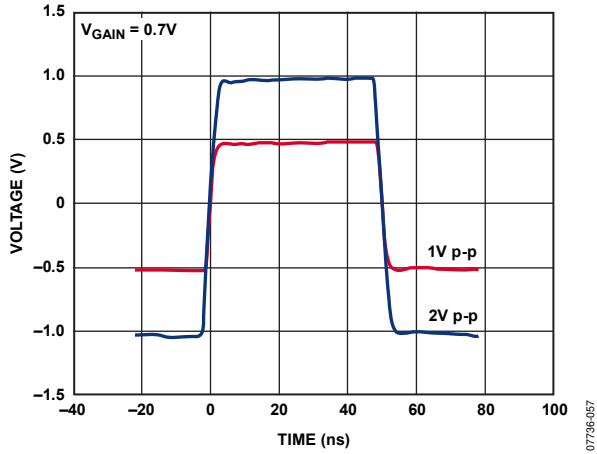


Figure 57. Large Signal Pulse Response to VGAX

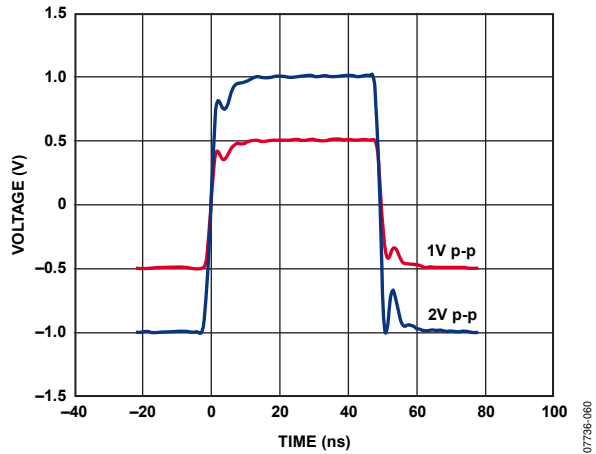


Figure 60. OFSx Large Signal Pulse Response

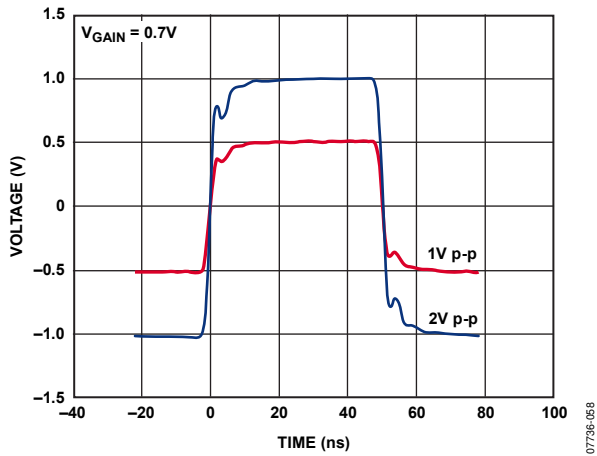


Figure 58. Large Signal Pulse Response to Differential Output

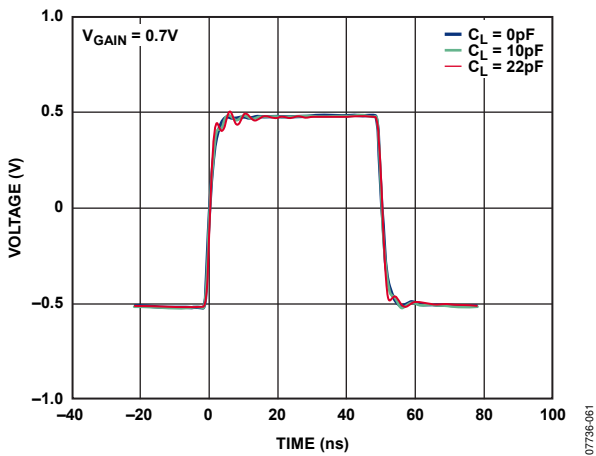


Figure 61. Large Signal Pulse Response to VGAX for Various Capacitive Loads

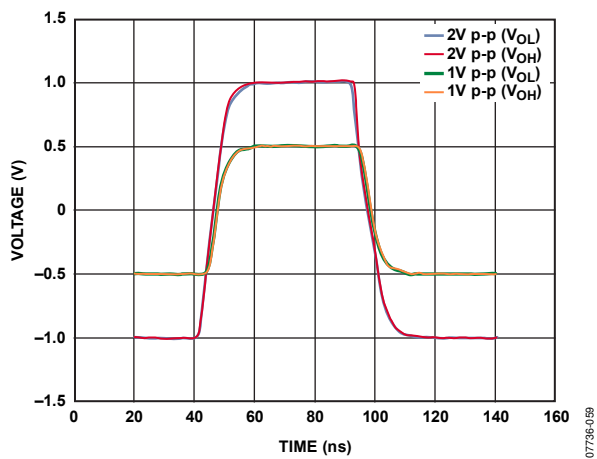


Figure 59. VOCM Large Signal Pulse Response

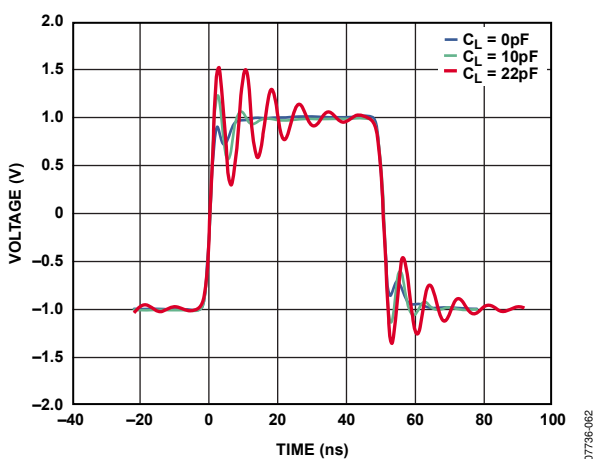


Figure 62. Large Signal Pulse Response to Differential Output for Various Capacitive Loads

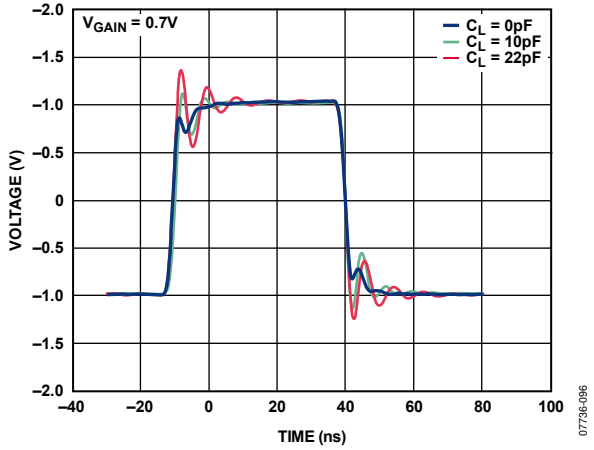


Figure 63. Large Signal Pulse Response to Differential Output for Various Capacitive Loads with Series $R = 10 \Omega$

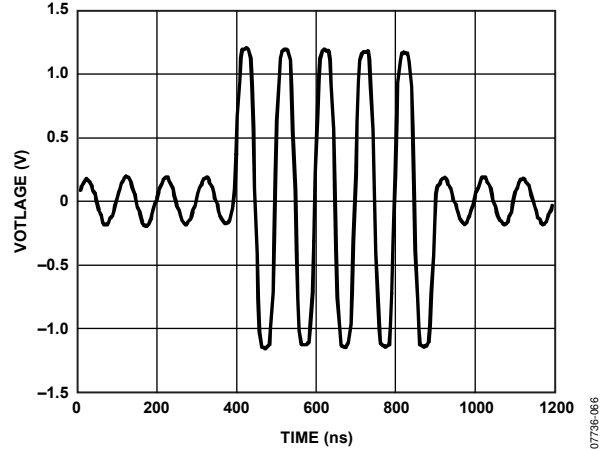


Figure 66. Preamp Overdrive Recovery

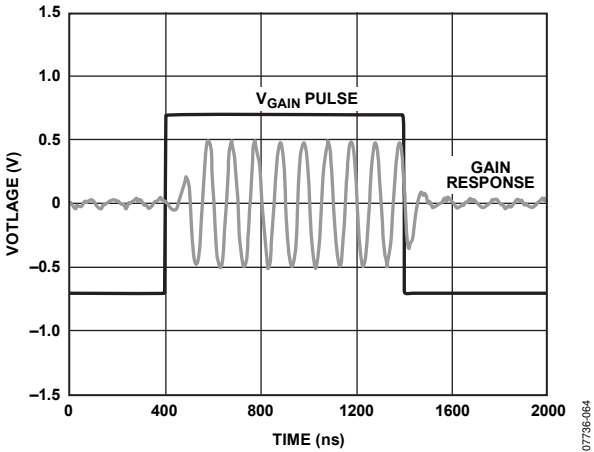


Figure 64. VGAX Response to Change in V_{GAIN}

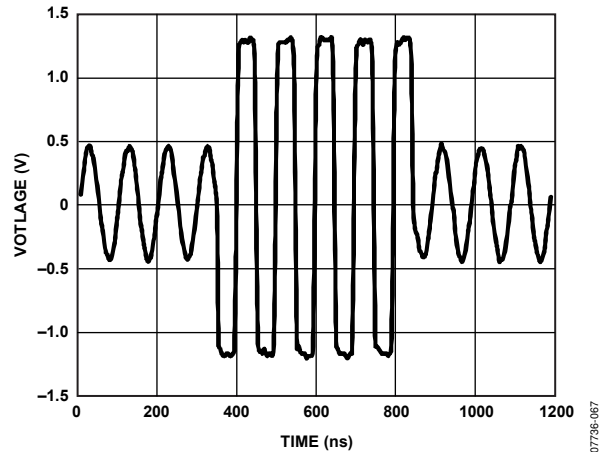


Figure 67. VGA Overdrive Recovery

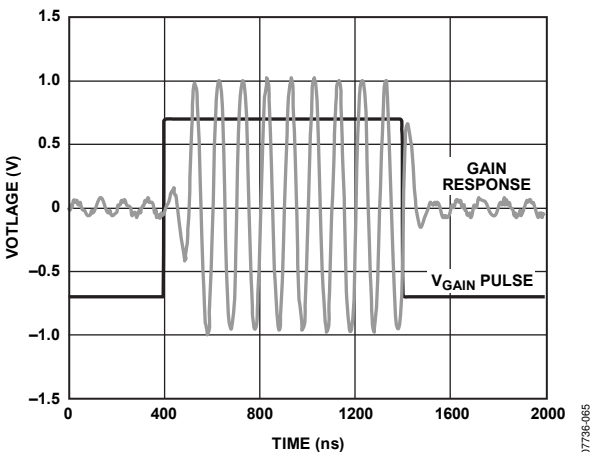


Figure 65. Differential Output Response to Change in V_{GAIN}

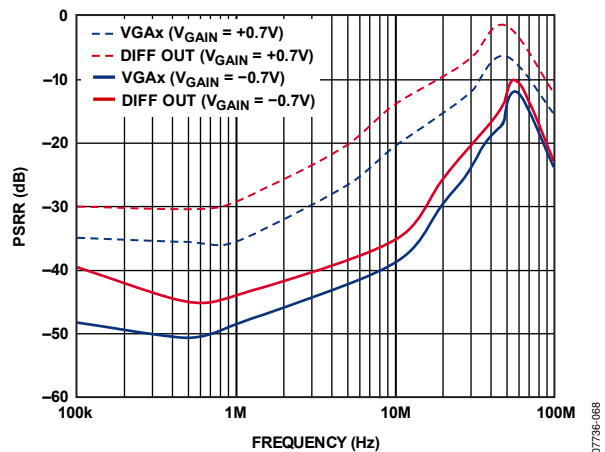


Figure 68. Power Supply Rejection vs. Frequency (VPOS)

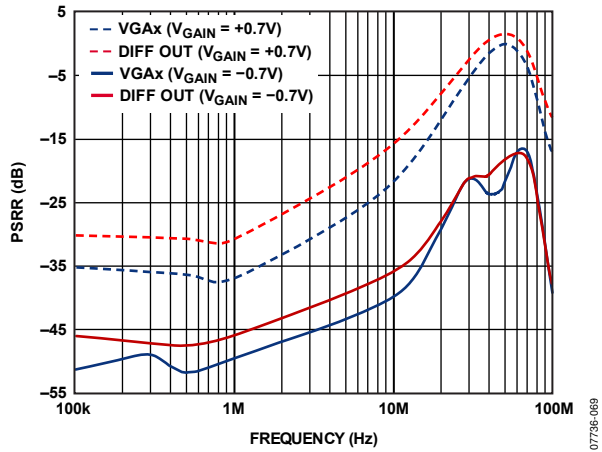


Figure 69. Power Supply Rejection vs. Frequency (VNEG)

07738-069

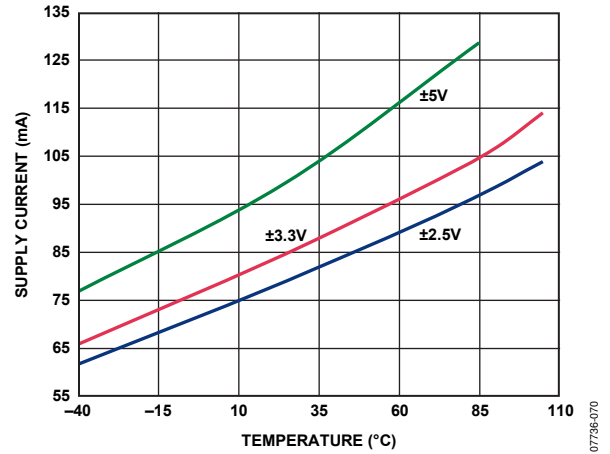


Figure 70. Quiescent Supply Current vs. Temperature

07738-070

TEST CIRCUITS

$V_S = \pm 2.5\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 10\text{ MHz}$, $C_L = 5\text{ pF}$, $R_L = 500\ \Omega$ per output (V_{GAx} , V_{OHx} , V_{OLx}), $V_{GAIN} = (V_{GNHx} - V_{GNLO}) = 0\text{ V}$, $V_{VOCM} = \text{GND}$, $V_{OFSx} = \text{GND}$, gain range = 6 dB to 30 dB, unless otherwise specified.

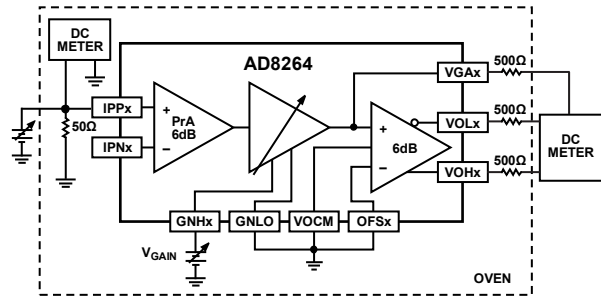


Figure 71. Gain vs. V_{GAIN} vs. Temperature (See Figure 3 and Figure 4)

07736-119

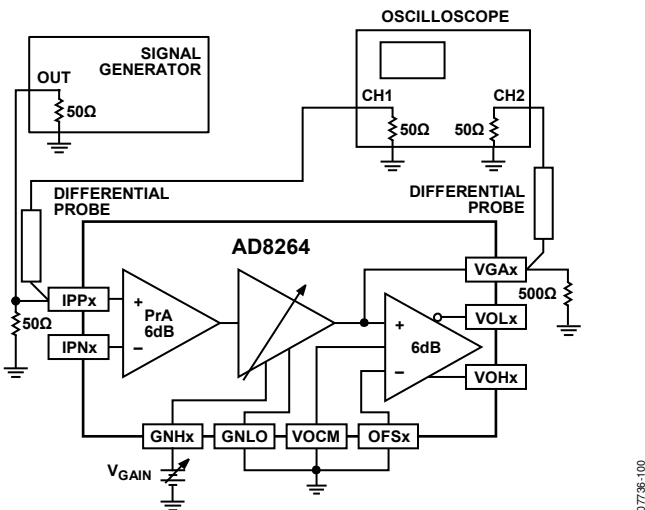


Figure 72. Gain Error vs. V_{GAIN} at Various Frequencies to V_{GAx} (See Figure 5)

07736-100

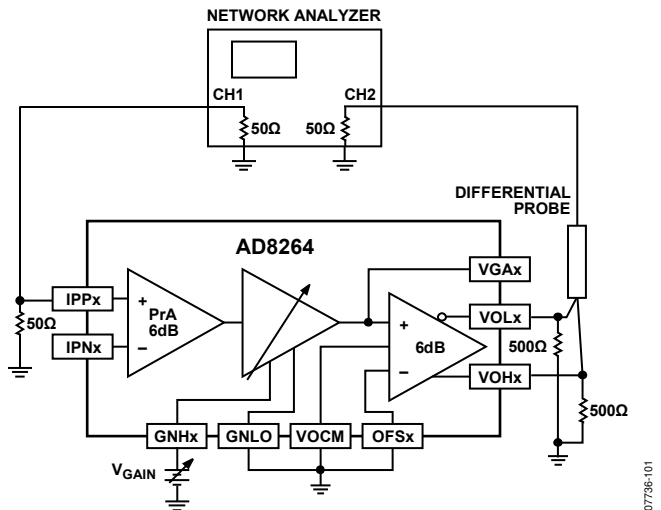


Figure 74. Frequency Response vs. Gain to Differential Output for Various Values of V_{GAIN} (See Figure 11)

07736-101

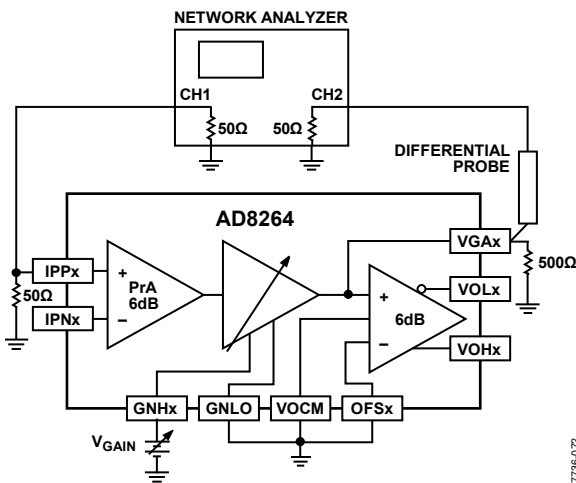


Figure 73. Frequency Response vs. Gain to V_{GAx} for Various Values of V_{GAIN} , $V_{GAIN} = V_{GNHx} - V_{GNLO}$ (See Figure 10)

07736-072

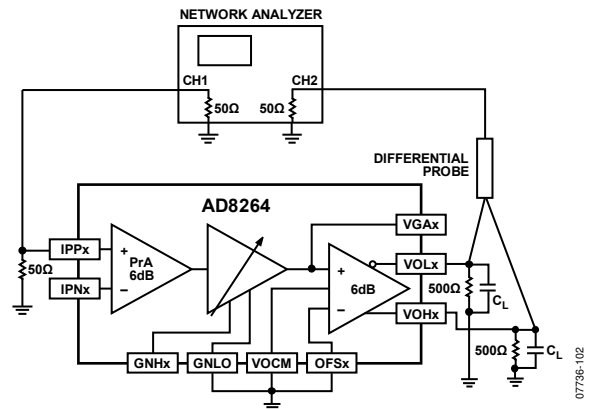


Figure 75. Frequency Response to Differential Output for Various Capacitive Loads (See Figure 12)

07736-102

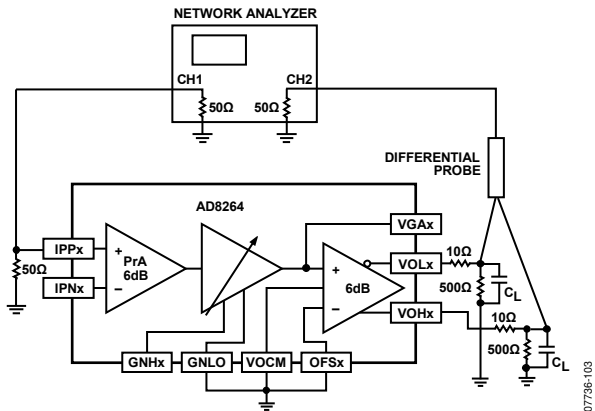


Figure 76. Frequency Response to Differential Output for Various Capacitive Loads with Series R = 10 Ω (See Figure 13)

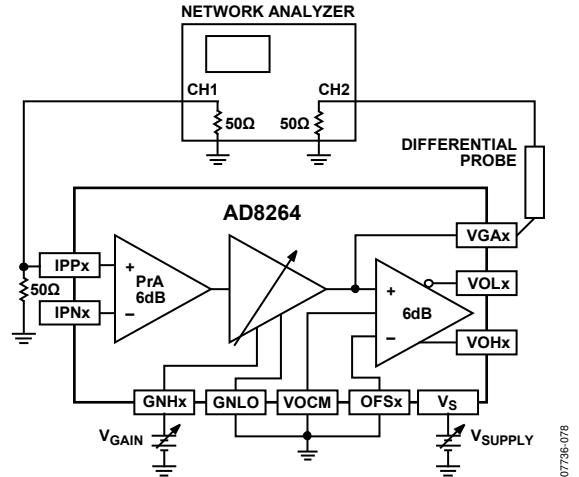


Figure 79. Frequency Response vs. Gain to VGAx for Various Supply Voltages (See Figure 18)

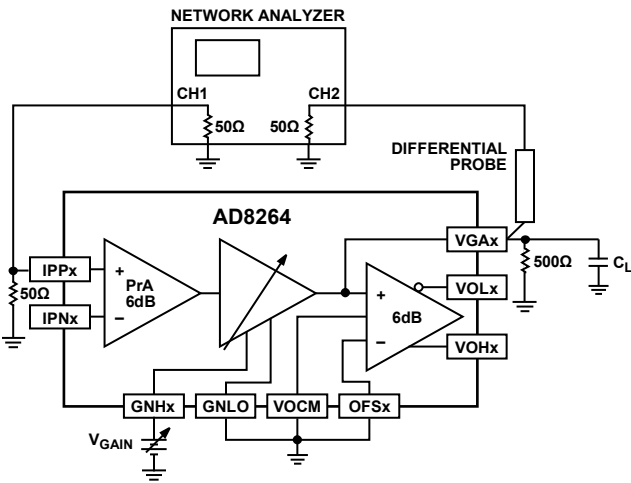


Figure 77. Frequency Response to VGAx for Various Capacitive Loads (See Figure 14)

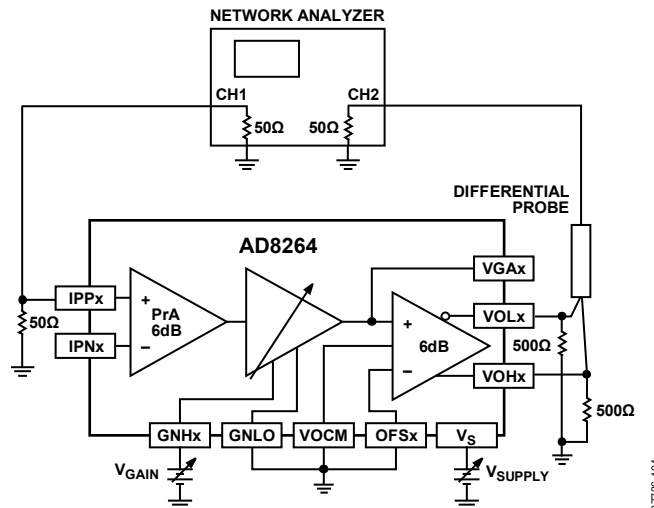


Figure 80. Frequency Response vs. Gain to Differential Output for Various Supply Voltages (See Figure 19)

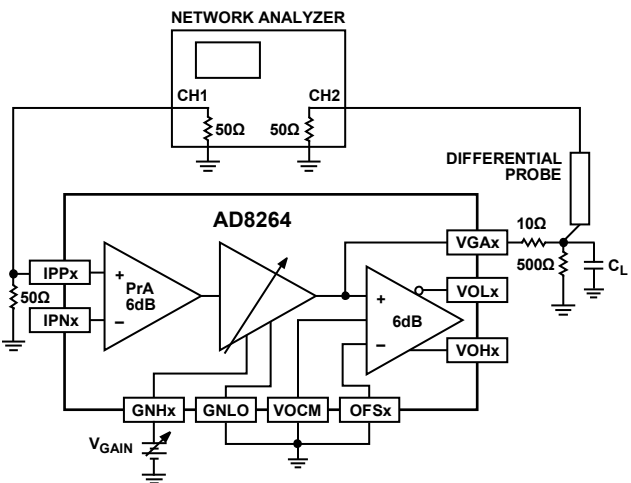


Figure 78. Frequency Response to VGAx for Various Capacitive Loads with Series R = 10 Ω (See Figure 16)

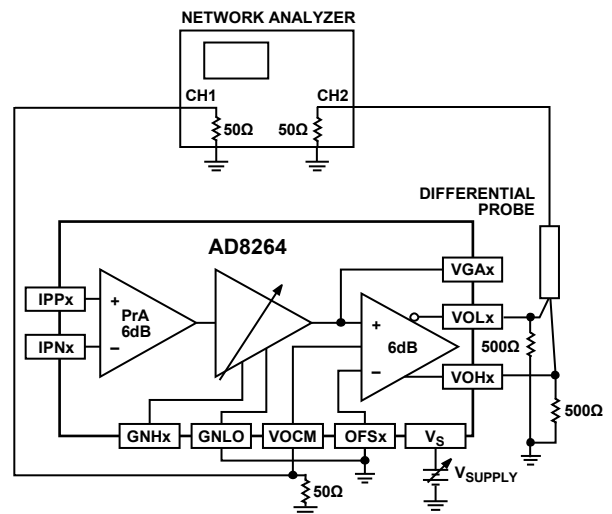


Figure 81. VOCM Frequency Response to Differential Output (See Figure 21)

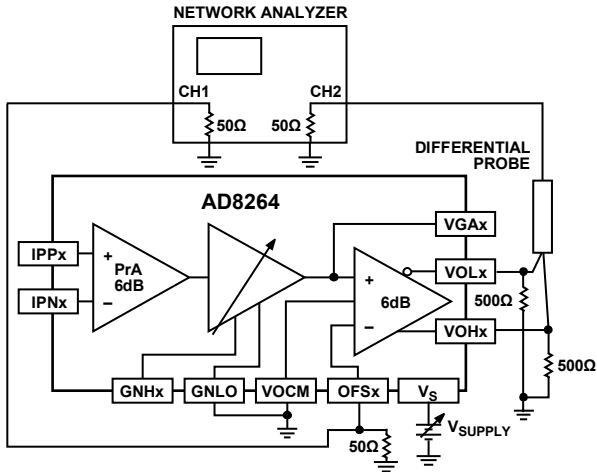


Figure 82. OFSx Frequency Response to Differential Output (See Figure 22)

07796-106

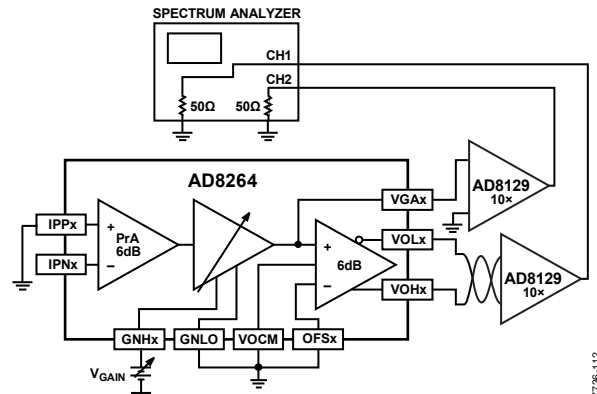


Figure 85. Output Referred Noise vs. V_{GAIN} (See Figure 32)

07796-112

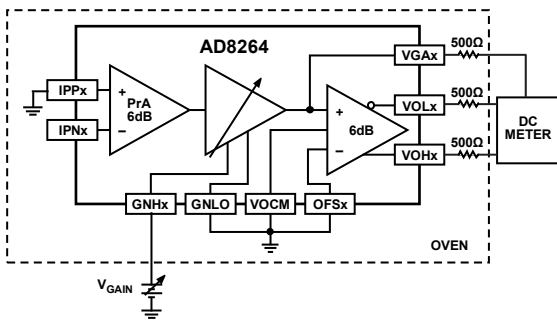


Figure 83. Output Offset Voltage vs. V_{GAIN} vs. Temperature (See Figure 26 and Figure 27)

07796-110

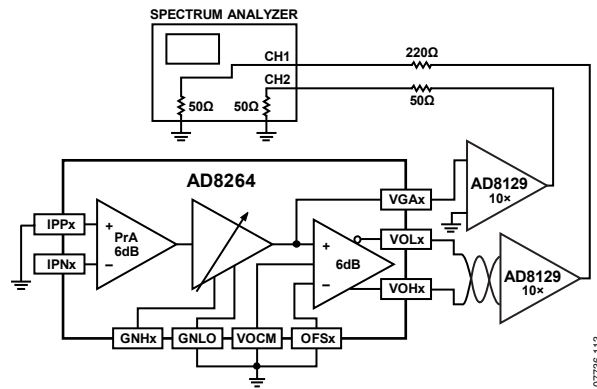


Figure 86. Input Referred Noise vs. Frequency (See Figure 34)

07796-113

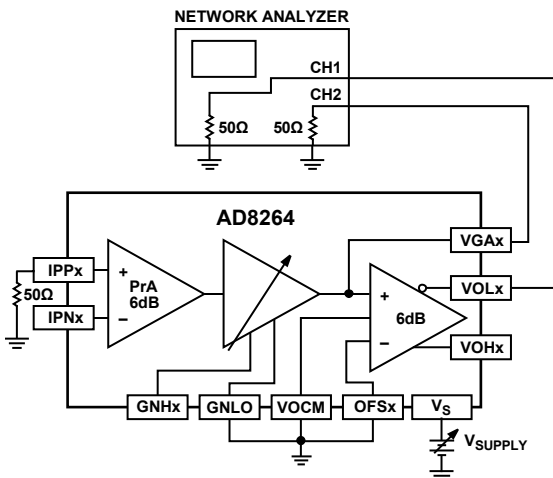


Figure 84. Output Resistance vs. Frequency (See Figure 30 and Figure 31)

07796-111

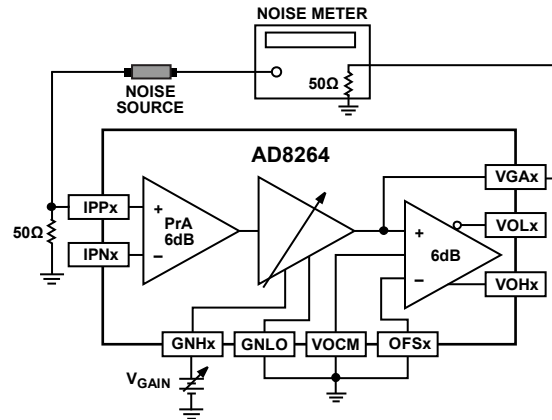


Figure 87. Noise Figure vs. V_{GAIN} (See Figure 36)

07796-115

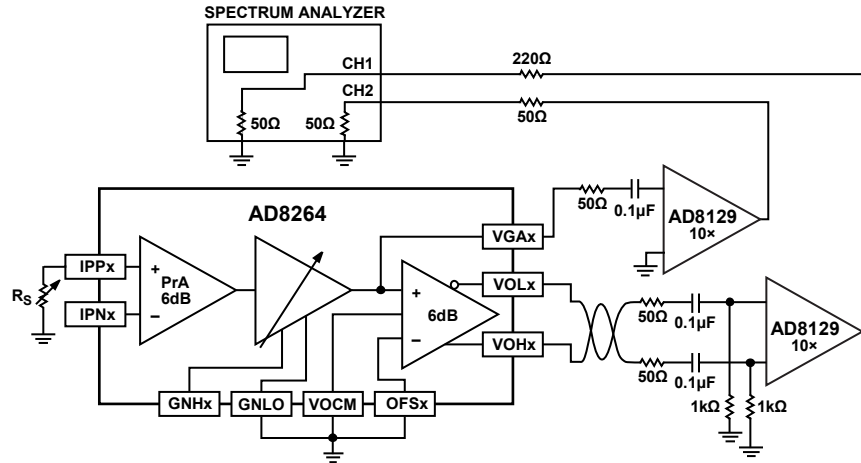


Figure 88. Input Referred Noise vs. R_{SOURCE} (See Figure 35)

07736-114

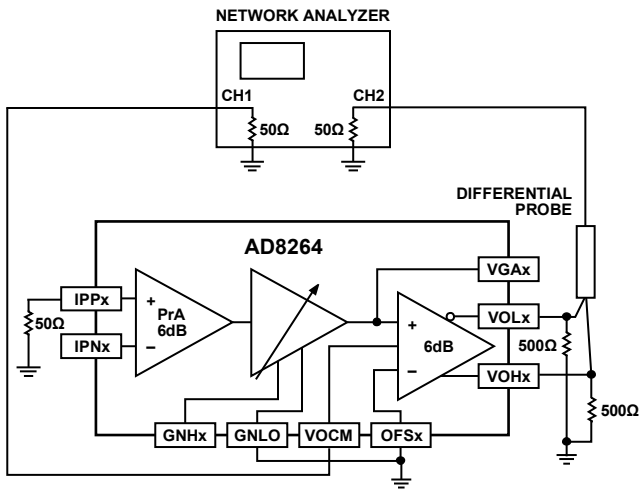


Figure 89. VOCM Common-Mode Rejection vs. Frequency (See Figure 37)

07736-116

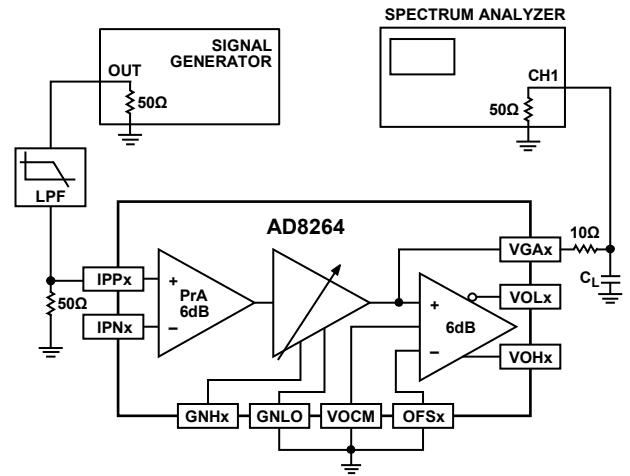


Figure 91. Harmonic Distortion to VGAX vs. C_{LOAD} (Figure 39)

07736-118

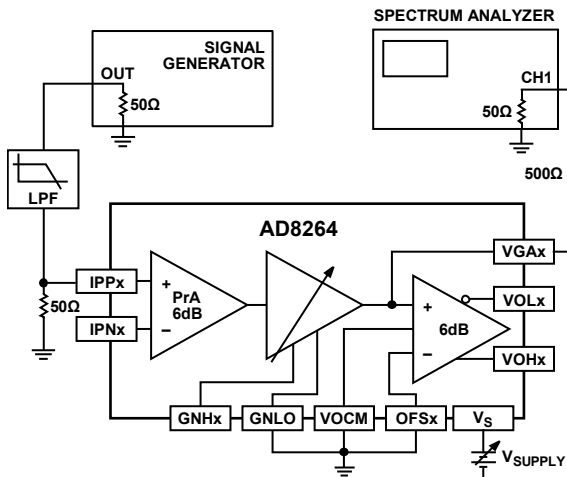


Figure 90. Test Circuit Harmonic Distortion to VGAX vs. R_{LOAD} and Various Supplies (See Figure 38)

07736-117

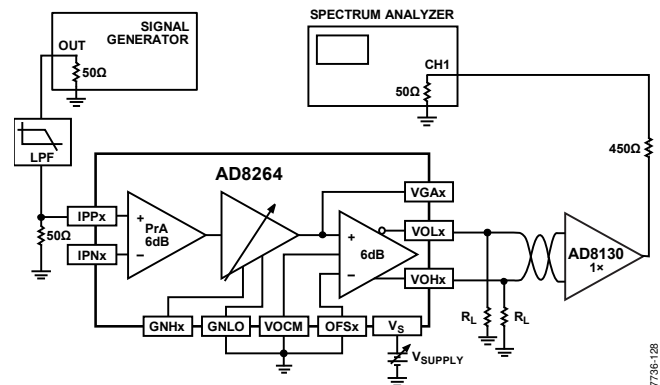


Figure 92. Harmonic Distortion to Differential Output vs. R_{LOAD} and Various Supplies (See Figure 40)

07736-128

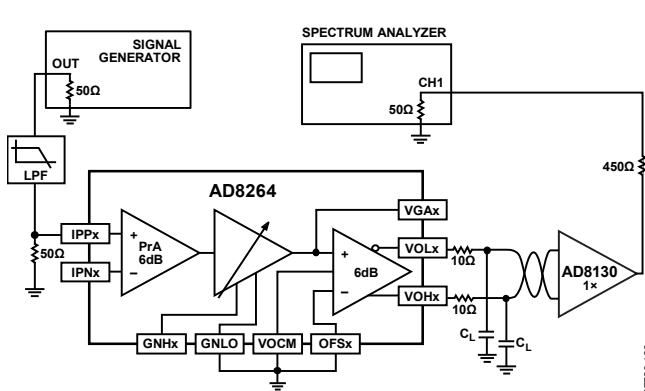


Figure 93. Harmonic Distortion to Differential Output vs. C_{LOAD} (See Figure 41)

07736-129

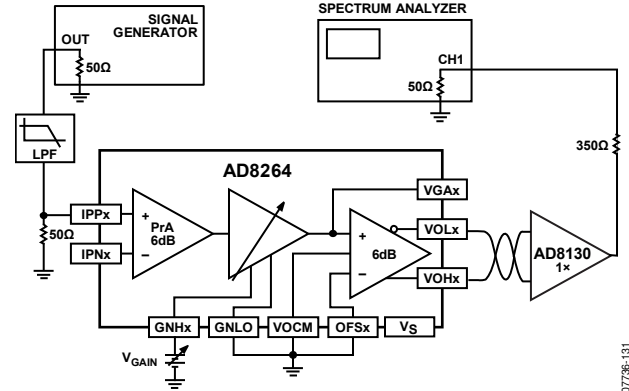


Figure 95. HD2 and HD3 to Differential Output (See Figure 46 through Figure 49)

07736-131

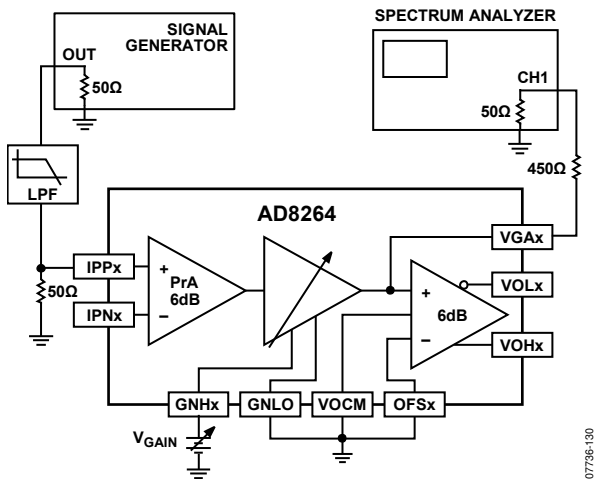


Figure 94. HD2 and HD3 to VGAx (See Figure 42 Through Figure 45)

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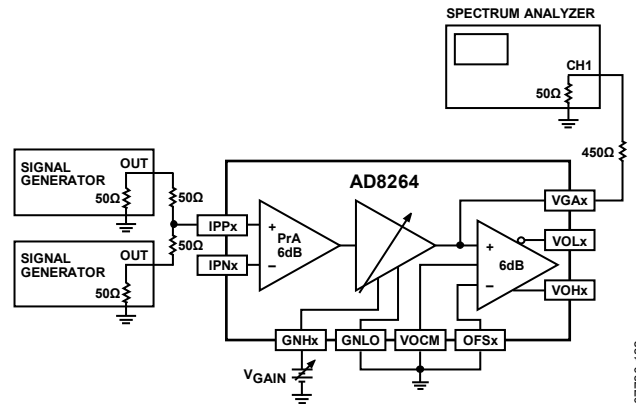


Figure 96. IMD3 and OIP3 to VGAx (See Figure 50 and Figure 51)

07736-132

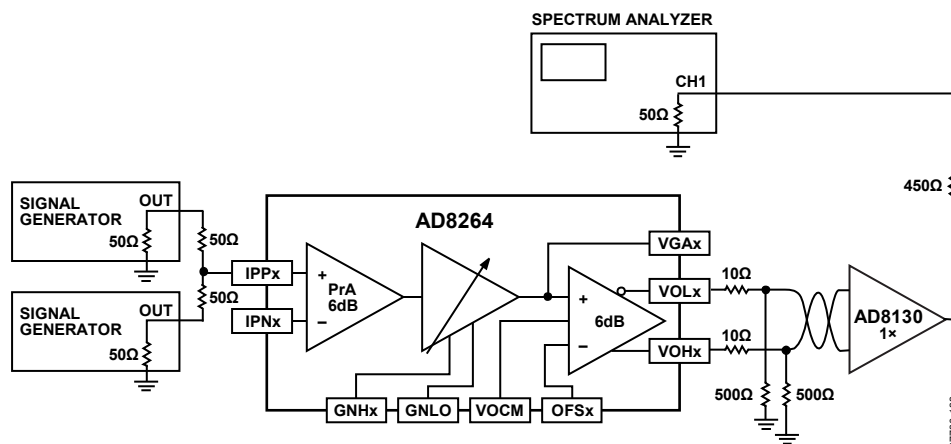


Figure 97. IMD3 and OIP3 to Differential Output (See Figure 52 and Figure 53)

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