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FEATURES

JESD204B (Subclass 1) coded serial digital outputs

Support for lane rates up to 16 Gbps per lane

Noise density

–152 dBFS/Hz at 2.56 GSPS at full-scale voltage = 1.7 V p-p

–154 dBFS/Hz at 2.56 GSPS at full-scale voltage = 2.0 V p-p

–154.2 dBFS/Hz at 2.0 GSPS at full-scale voltage = 1.7 V p-p

–155.3 dBFS/Hz at 2.0 GSPS at full-scale voltage = 2.0 V p-p

1.55 W total power per channel at 2.56 GSPS (default settings)

SFDR at 2.56 GSPS encode

73 dBFS at 1.8 GHz A_{IN} at –2.0 dBFS

59 dBFS at 5.53 GHz A_{IN} at –2.0 dBFS

full-scale voltage = 1.1 V p-p

SNR at 2.56 GSPS encode

59.7 dBFS at 1.8 GHz A_{IN} at –2.0 dBFS

53.0 dBFS at 5.53 GHz A_{IN} at –2.0 dBFS

full-scale voltage = 1.1 V p-p

SFDR at 2.0 GSPS encode

78 dBFS at 900 MHz A_{IN} at –2.0 dBFS

62 dBFS at 5.53 GHz A_{IN} at –2.0 dBFS

full-scale voltage = 1.1 V p-p

SNR at 2.0 GSPS encode

62.7 dBFS at 900 MHz A_{IN} at –2.0 dBFS

53.1 dBFS at 5.5 GHz A_{IN} at –2.0 dBFS

full-scale voltage = 1.1 V p-p

0.975 V, 1.9 V, and 2.5 V dc supply operation

9 GHz analog input full power bandwidth (–3 dB)

Amplitude detect bits for efficient AGC implementation

Programmable FIR filters for analog channel loss equalization

2 integrated, wideband digital processors per channel

48-bit NCO

Programmable decimation rates

Phase coherent NCO switching

Up to 4 channels available

Serial port control

Supports 100 MHz SPI writes and 50 MHz SPI reads

Integer clock with divide by 2 and divide by 4 options

Flexible JESD204B lane configurations

On-chip dither

APPLICATIONS

Diversity multiband and multimode digital receivers

3G/4G, TD-SCDMA, W-CDMA, and GSM, LTE, LTE-A

Electronic test and measurement systems

Phased array radar and electronic warfare

DOCSIS 3.0 CMTS upstream receive paths

HFC digital reverse path receivers

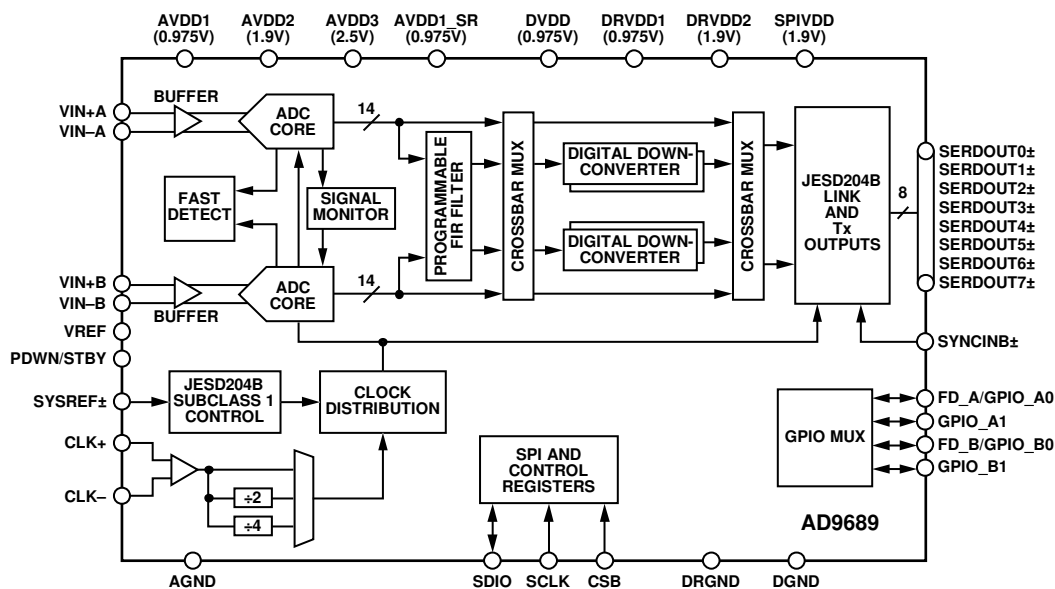
FUNCTIONAL BLOCK DIAGRAM


Figure 1.

Rev. A

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REVISION HISTORY

10/2017—Rev. 0 to Rev. A

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Added 2.0 GSPS Section and Figure 6 to Figure 11; Renumbered Sequentially	16
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9/2017—Revision 0: Initial Version

GENERAL DESCRIPTION

The AD9689 is a dual, 14-bit, 2.0 GSPS/2.6 GSPS analog-to-digital converter (ADC). The device has an on-chip buffer and a sample-and-hold circuit designed for low power, small size, and ease of use. This product is designed to support communications applications capable of direct sampling wide bandwidth analog signals of up to 5 GHz. The -3 dB bandwidth of the ADC input is 9 GHz. The AD9689 is optimized for wide input bandwidth, high sampling rate, excellent linearity, and low power in a small package.

The dual ADC cores feature a multistage, differential pipelined architecture with integrated output error correction logic. Each ADC features wide bandwidth inputs supporting a variety of user-selectable input ranges. An integrated voltage reference eases design considerations. The analog input and clock signals are differential inputs. The ADC data outputs are internally connected to four digital downconverters (DDCs) through a crossbar mux. Each DDC consists of multiple cascaded signal processing stages: a 48-bit frequency translator (numerically controlled oscillator (NCO)), and decimation rates. The NCO has the option to select preset bands over the general-purpose input/output (GPIO) pins, which enables the selection of up to three bands. Operation of the AD9689 between the DDC modes is selectable via SPI-programmable profiles.

In addition to the DDC blocks, the AD9689 has several functions that simplify the automatic gain control (AGC) function in a communications receiver. The programmable threshold detector allows monitoring of the incoming signal power using the fast detect control bits in Register 0x0245 of the ADC. If the input signal level exceeds the programmable threshold, the fast detect indicator goes high. Because this threshold indicator has low latency, the user can quickly turn down the system gain to avoid an overrange condition at the ADC input. In addition to the fast detect outputs, the AD9689 also offers signal monitoring capability. The signal monitoring block provides additional information about the signal being digitized by the ADC.

The user can configure the Subclass 1 JESD204B-based high speed serialized output in a variety of one-lane, two-lane, four-lane, and eight-lane configurations, depending on the DDC configuration and the acceptable lane rate of the receiving logic device. Multidevice synchronization is supported through the $\text{SYSREF}\pm$ and $\text{SYNCINB}\pm$ input pins.

The AD9689 has flexible power-down options that allow significant power savings when desired. All of these features can be programmed using a 3-wire serial port interface (SPI).

The AD9689 is available in a Pb-free, 196-ball BGA, specified over the -40°C to $+85^{\circ}\text{C}$ ambient temperature range. This product is protected by a U.S. patent.

Note that throughout this data sheet, multifunction pins, such as $\text{FD_A}/\text{GPIO_A0}$, are referred to either by the entire pin name or by a single function of the pin, for example, FD_A , when only that function is relevant.

PRODUCT HIGHLIGHTS

1. Wide, input -3 dB bandwidth of 9 GHz supports direct radio frequency (RF) sampling of signals up to about 5 GHz.
2. Four integrated, wideband decimation filters and NCO blocks supporting multiband receivers.
3. Fast NCO switching enabled through the GPIO pins.
4. SPI controls various product features and functions to meet specific system requirements.
5. Programmable fast overrange detection and signal monitoring.
6. On-chip temperature diode for system thermal management.
7. 12 mm \times 12 mm, 196-ball BGA.
8. Pin, package, feature, and memory map compatible with the [AD9208](#) 14-bit, 3.0 GSPS, JESD204B dual ADC.

SPECIFICATIONS

DC SPECIFICATIONS

AVDD1 = 0.975 V, AVDD1_SR = 0.975 V, AVDD2 = 1.9 V, AVDD3 = 2.5 V, DVDD = 0.975 V, DRVDD1 = 0.975 V, DRVDD2 = 1.9 V, SPIVDD = 1.9 V, sampling rate = 2.0 GHz/2.56 GHz, clock divider = 2, 1.7 V p-p full-scale differential input, input amplitude (A_{IN}) = -2.0 dBFS, L = 8, M = 2, F = 1, $-10^{\circ}\text{C} \leq T_J \leq +120^{\circ}\text{C}$,¹ unless otherwise noted. Typical specifications represent performance at $T_J = 70^{\circ}\text{C}$ ($T_A = 25^{\circ}\text{C}$).

Table 1.

Parameter	2.0 GSPS			2.6 GSPS			Unit
	Min	Typ	Max	Min	Typ	Max	
RESOLUTION	14			14			Bits
ACCURACY							
No Missing Codes		Guaranteed			Guaranteed		
Offset Error				0			%FSR
Offset Matching		0		0			%FSR
Gain Error	-2.9	±1	+1.8	-4.9	±1	+5.6	%FSR
Gain Matching		±0.2			±0.2		%FSR
Differential Nonlinearity (DNL)	-0.62	±0.4	+0.79	-0.65	±0.4	+0.75	LSB
Integral Nonlinearity (INL)	-9.9	±2	+8.1	-16	±6	+13	LSB
TEMPERATURE DRIFT							
Offset Error		±7.7			±3.7		ppm/°C
Gain Error		15			58		ppm/°C
INTERNAL VOLTAGE REFERENCE		0.5			0.5		V
INPUT REFERRED NOISE		3.8			4.6		LSB rms
ANALOG INPUTS							
Differential Input Voltage Range	1.1	1.7	2.0	1.1	1.7	2.0	V p-p
Common-Mode Voltage (V_{CM})		1.4			1.4		V
Differential Input Capacitance		0.35			0.35		pF
-3 dB Bandwidth		9			9		GHz
POWER SUPPLY							
AVDD1	0.95	0.975	1.0	0.95	0.975	1.0	V
AVDD2	1.85	1.9	1.95	1.85	1.9	1.95	V
AVDD3	2.44	2.5	2.56	2.44	2.5	2.56	V
AVDD1_SR	0.95	0.975	1.0	0.95	0.975	1.0	V
DVDD	0.95	0.975	1.0	0.95	0.975	1.0	V
DRVDD1	0.95	0.975	1.0	0.95	0.975	1.0	V
DRVDD2	1.85	1.9	1.95	1.85	1.9	1.95	V
SPIVDD	1.85	1.9	1.95	1.85	1.9	1.95	V
I_{AVDD1}		455	605		590	693	mA
I_{AVDD2}		585	670		810	882	mA
I_{AVDD3}		65	72		65	73	mA
I_{AVDD1_SR}		25	41		25	43	mA
I_{DVDD}		340	800		405	833	mA
I_{DRVDD1}^2		320	432		390	500	mA
I_{DRVDD2}		25	30		25	30	mA
I_{SPIVDD}		1	5		1	5	mA

POWER CONSUMPTION			
Total Power Dissipation (Including Output Drivers) ³	2.45	3.1	W
Power-Down Dissipation	265	300	mW
Standby ⁴	1.3	1.5	W

¹ The junction temperature (T_J) range of -10°C to $+120^\circ\text{C}$ translates to an ambient temperature (T_A) range of -40°C to $+85^\circ\text{C}$.

² All lanes running. Power dissipation on DRVDDx changes with lane rate and number of lanes used.

³ Default mode. No DDCs used.

⁴ Can be controlled by the SPI.

AC SPECIFICATIONS

AVDD1 = 0.975 V, AVDD1_SR = 0.975 V, AVDD2 = 1.9 V, AVDD3 = 2.5 V, DVDD = 0.975 V, DRVDD1 = 0.975 V, DRVDD2 = 1.9 V, SPIVDD = 1.9 V, sampling rate = 2.0 GHz/2.56 GHz, clock divider = 2, 1.7 V p-p full-scale differential input, input amplitude (A_{IN}) = -2.0 dBFS, default SPI settings, $-10^\circ\text{C} \leq T_J \leq +120^\circ\text{C}$,¹ unless otherwise noted. Typical specifications represent performance at $T_J = 70^\circ\text{C}$ ($T_A = 25^\circ\text{C}$).

Table 2.

Parameter ²	2.0 GSPS			2.6 GSPS			Unit
	Min	Typ	Max	Min	Typ	Max	
NOISE DENSITY ³							
Full Scale = 1.7 V p-p		-154.2			-152		dBFS/Hz
Full Scale = 2.0 V p-p		-155.3			-154		dBFS/Hz
CODE ERROR RATE (CER)							
AVDD1 = 0.975 V		7×10^{-15}			9×10^{-9}		Errors
AVDD1 = 1.0 V		3×10^{-15}			4.5×10^{-10}		Errors
SIGNAL-TO-NOISE RATIO (SNR)							
$f_{IN} = 155$ MHz		63.7			61.3		dBFS
$f_{IN} = 155$ MHz (Full Scale = 2.0 V p-p)		65.0			62.5		dBFS
$f_{IN} = 750$ MHz		63.1			61.0		dBFS
$f_{IN} = 900$ MHz	60.2	62.7			60.9		dBFS
$f_{IN} = 1800$ MHz		60.9		56.0	59.7		dBFS
$f_{IN} = 2100$ MHz		59.9			59.3		dBFS
$f_{IN} = 3300$ MHz		58.3			58.0		dBFS
$f_{IN} = 4350$ MHz (Full Scale = 1.1 V p-p)		54.4			54.0		dBFS
$f_{IN} = 5530$ MHz (Full Scale = 1.1 V p-p)		53.1			53.0		dBFS
SIGNAL-TO-NOISE-AND-DISTORTION RATIO (SINAD)							
$f_{IN} = 155$ MHz		63.5			61.2		dBFS
$f_{IN} = 155$ MHz (Full Scale = 2.0 V p-p)		64.7			62.4		dBFS
$f_{IN} = 750$ MHz		62.8			60.7		dBFS
$f_{IN} = 900$ MHz	59.6	62.5			60.5		dBFS
$f_{IN} = 1800$ MHz		60.8		52.4	59.4		dBFS
$f_{IN} = 2100$ MHz		59.7			59.1		dBFS
$f_{IN} = 3300$ MHz		55.3			56.6		dBFS
$f_{IN} = 4350$ MHz (Full Scale = 1.1 V p-p)		53.2			51.0		dBFS
$f_{IN} = 5530$ MHz (Full Scale = 1.1 V p-p)		52.3			49.5		dBFS
EFFECTIVE NUMBER OF BITS (ENOB)							
$f_{IN} = 155$ MHz		10.3			9.9		Bits
$f_{IN} = 155$ MHz (Full Scale = 2.0 V p-p)		10.5			10.1		Bits
$f_{IN} = 750$ MHz		10.1			9.8		Bits
$f_{IN} = 900$ MHz	9.6	10.1			9.8		Bits
$f_{IN} = 1800$ MHz		9.8		8.4	9.6		Bits
$f_{IN} = 2100$ MHz		9.6			9.5		Bits
$f_{IN} = 3300$ MHz		8.9			9.1		Bits
$f_{IN} = 4350$ MHz (Full Scale = 1.1 V p-p)		8.6			8.2		Bits
$f_{IN} = 5530$ MHz (Full Scale = 1.1 V p-p)		8.4			7.9		Bits

Parameter ²	2.0 GSPS			2.6 GSPS			Unit
	Min	Typ	Max	Min	Typ	Max	
SPURIOUS FREE DYNAMIC RANGE (SFDR), SECOND OR THIRD HARMONIC ^{4,5}							
$f_{IN} = 155$ MHz		77			78		dBFS
$f_{IN} = 155$ MHz (Full Scale = 2.0 V p-p)		77			78		dBFS
$f_{IN} = 750$ MHz		77			73		dBFS
$f_{IN} = 900$ MHz	66	78			74		dBFS
$f_{IN} = 1800$ MHz		76		58	73		dBFS
$f_{IN} = 2100$ MHz		76			73		dBFS
$f_{IN} = 3300$ MHz		60			64		dBFS
$f_{IN} = 4350$ MHz (Full Scale = 1.1 V p-p)		61			60		dBFS
$f_{IN} = 5530$ MHz (Full Scale = 1.1 V p-p)		62			59		dBFS
WORST OTHER, EXCLUDING SECOND OR THIRD HARMONIC							
$f_{IN} = 155$ MHz		-99			-96		dBFS
$f_{IN} = 155$ MHz (Full Scale = 2.0 V p-p)		-95			-98		dBFS
$f_{IN} = 750$ MHz		-100			-97		dBFS
$f_{IN} = 900$ MHz		-94	-80		-96		dBFS
$f_{IN} = 1800$ MHz		-91			-88	-74	dBFS
$f_{IN} = 2100$ MHz		-86			-94		dBFS
$f_{IN} = 3300$ MHz		-85			-85		dBFS
$f_{IN} = 4350$ MHz (Full Scale = 1.1 V p-p)		-83			-84		dBFS
$f_{IN} = 5530$ MHz (Full Scale = 1.1 V p-p)		-82			-82		dBFS
TWO-TONE INTERMODULATION DISTORTION (IMD), A_{IN1} AND $A_{IN2} = -8.0$ dBFS							
$f_{IN1} = 1841$ MHz, $f_{IN2} = 1846$ MHz		-72			-72		dBFS
$f_{IN1} = 2137$ MHz, $f_{IN2} = 2142$ MHz		-74			-76		dBFS
CROSSTALK ⁶		>90			>90		dB
ANALOG INPUT BANDWIDTH, FULL POWER ⁷		5			5		GHz

¹ The junction temperature (T_J) range of -10°C to $+120^\circ\text{C}$ translates to an ambient temperature (T_A) range of -40°C to $+85^\circ\text{C}$.

² See AN-835 for definitions and for details on how these tests were completed.

³ Noise density is measured at a low analog input frequency (30 MHz).

⁴ The input configuration component values are found in Table 9. Refer to Table 10 for the recommended buffer settings.

⁵ Figure 79 shows the differential transformer coupled configuration. Figure 80 is the input network configuration for frequencies > 5 GHz.

⁶ Crosstalk is measured at 950 MHz with a -2.0 dBFS analog input on one channel, and no input on the adjacent channel.

⁷ Full power bandwidth is the bandwidth of operation in which proper ADC performance can be achieved.

DIGITAL SPECIFICATIONS

AVDD1 = 0.975 V, AVDD1_SR = 0.975 V, AVDD2 = 1.9 V, AVDD3 = 2.5 V, DVDD = 0.975 V, DRVDD1 = 0.975 V, DRVDD2 = 1.9 V, SPIVDD = 1.9 V, $-10^{\circ}\text{C} \leq T_j \leq +120^{\circ}\text{C}$,¹ unless otherwise noted. Typical specifications represent performance at $T_j = 70^{\circ}\text{C}$ ($T_A = 25^{\circ}\text{C}$).

Table 3.

Parameter	Min	Typ	Max	Unit
CLOCK INPUTS (CLK+, CLK–)				
Logic Compliance		LVDS/LVPECL		
Differential Input Voltage	300	800	1800	mV p-p
Input Common-Mode Voltage		0.675		V
Input Resistance (Differential)		106		Ω
Input Capacitance		0.9		pF
Differential Input Return Loss at 2.6 GHz ²		9.4		dB
SYSTEM REFERENCE (SYSREF) INPUTS (SYSREF+, SYSREF–)				
Logic Compliance		LVDS/LVPECL		
Differential Input Voltage	400	800	1800	mV p-p
Input Common-Mode Voltage		0.675	2.0	V
Input Resistance (Differential)		18		k Ω
Input Capacitance (Differential)		1		pF
LOGIC INPUTS (SDIO, SCLK, CSB, PDWN/STBY, FD_A/GPIO_A0, FD_B/GPIO_B0, GPIO_A1, GPIO_B1)				
Logic Compliance		CMOS		
Logic 1 Voltage	$0.65 \times \text{SPIVDD}$			V
Logic 0 Voltage	0		$0.35 \times \text{SPIVDD}$	V
Input Resistance		30		k Ω
LOGIC OUTPUTS (SDIO, FD_A, FD_B)				
Logic Compliance		CMOS		
Logic 1 Voltage ($I_{OH} = 4 \text{ mA}$)	$\text{SPIVDD} - 0.45\text{V}$			V
Logic 0 Voltage ($I_{OL} = 4 \text{ mA}$)	0		0.45	V
SYNCHRONIZATION INPUT (SYNCINB+/SYNCINB–)				
Logic Compliance		LVDS/LVPECL		
Differential Input Voltage	400	800	1800	mV p-p
Input Common-Mode Voltage		0.675	2.0	V
Input Resistance (Differential)		18		k Ω
Input Capacitance		1		pF
SYNCINB+ INPUT				
Logic Compliance		CMOS		
Logic 1 Voltage	$0.9 \times \text{DRVDD1}$		$2 \times \text{DRVDD1}$	V
Logic 0 Voltage			$0.1 \times \text{DRVDD1}$	V
Input Resistance		2.6		k Ω
DIGITAL OUTPUTS (SERDOUTx_{\pm}, $x = 0$ TO 7)				
Logic Compliance		SST		
Differential Output Voltage	360	560	770	mV p-p
Differential Termination Impedance	80	100	120	Ω

¹ The junction temperature (T_j) range of -10°C to $+120^{\circ}\text{C}$ translates to an ambient temperature (T_A) range of -40°C to $+85^{\circ}\text{C}$.

² Reference impedance = 100 Ω .

SWITCHING SPECIFICATIONS

AVDD1 = 0.975 V, AVDD1_SR = 0.975 V, AVDD2 = 1.9 V, AVDD3 = 2.5 V, DVDD = 0.975 V, DRVDD1 = 0.975 V, DRVDD2 = 1.9 V, SPIVDD = 1.9 V, default SPI settings, $-10^{\circ}\text{C} \leq T_j \leq +120^{\circ}\text{C}$,¹ unless otherwise noted. Typical specifications represent performance at $T_j = 70^{\circ}\text{C}$ ($T_A = 25^{\circ}\text{C}$).

Table 4.

Parameter	2.0 GSPS			2.6 GSPS			Unit
	Min	Typ	Max	Min	Typ	Max	
CLOCK							
Clock Rate at CLK+/CLK– Pins			6			6	GHz
Sample Rate ²	1200	2000	2100	1900	2600	2700	MSPS
Clock Pulse Width High	238.096			185.185			ps
Clock Pulse Width Low	238.096			185.185			ps
OUTPUT PARAMETERS							
Unit Interval (UI) ³	62.5	66.67	592.6	62.5	66.67	592.6	ps
Rise Time (t_R) (20% to 80% into 100 Ω Load)		26			26		ps
Fall Time (t_F) (20% to 80% into 100 Ω Load)		26			26		ps
Phase-Locked Loop (PLL) Lock Time		5			5		ms
Data Rate per Channel (Nonreturn to Zero) ⁴	1.6875	13	16	1.6875	13	16	Gbps
LATENCY⁵							
Pipeline Latency ⁶		75			75		Clock cycles
Fast Detect Latency		26			26		Clock cycles
NCO Channel Selection to Output			8			8	Clock cycles
WAKE-UP TIME							
Standby		400			400		μs
Power-Down		15			15		ms
APERTURE							
Delay (t_A)		250			250		ps
Uncertainty (Jitter, t_j)		55			55		fs rms
Out of Range Recovery Time		1			1		Clock cycles

¹ The junction temperature (T_j) range of -10°C to $+120^{\circ}\text{C}$ translates to an ambient temperature (T_A) range of -40°C to $+85^{\circ}\text{C}$.

² The maximum sample rate is the clock rate after the divider.

³ Baud rate = $1/\text{UI}$. A subset of this range can be supported.

⁴ Default L = 8. This number can be changed based on the sample rate and decimation ratio.

⁵ No DDCs used. L = 8, M = 2, and F = 1.

⁶ Refer to the Latency section for more details.

TIMING SPECIFICATIONS

Table 5.

Parameter	Description	Min	Typ	Max	Unit
CLK+ to SYSREF+ TIMING REQUIREMENTS					
t_{SU_SR}	Device clock to SYSREF+ setup time		-65		ps
t_{H_SR}	Device clock to SYSREF+ hold time		95		ps
SPI TIMING REQUIREMENTS					
t_{DS}	Setup time between the data and the rising edge of SCLK	2			ns
t_{DH}	Hold time between the data and the rising edge of SCLK	2			ns
t_{CLK} for SPI Reads	Period of the SCLK	20			ns
t_{CLK} for SPI Writes	Period of the SCLK	10			ns
t_S	Setup time between CSB and SCLK	2			ns
t_H	Hold time between CSB and SCLK	2			ns
t_{HIGH} for SPI Reads	Minimum period that SCLK must be in a logic high state	8			ns
t_{HIGH} for SPI Writes	Minimum period that SCLK must be in a logic high state	4			ns
t_{LOW} for SPI Reads	Minimum period that SCLK must be in a logic low state	8			ns
t_{LOW} for SPI Writes	Minimum period that SCLK must be in a logic low state	4			ns
t_{ACCESS}	Maximum time delay between the falling edge of SCLK and output data valid for a read operation		5	8	ns
t_{DIS_SDIO}	Time required for the SDIO pin to switch from an output to an input, relative to the SCLK rising edge (not shown in Figure 4)	2			ns

Timing Diagrams

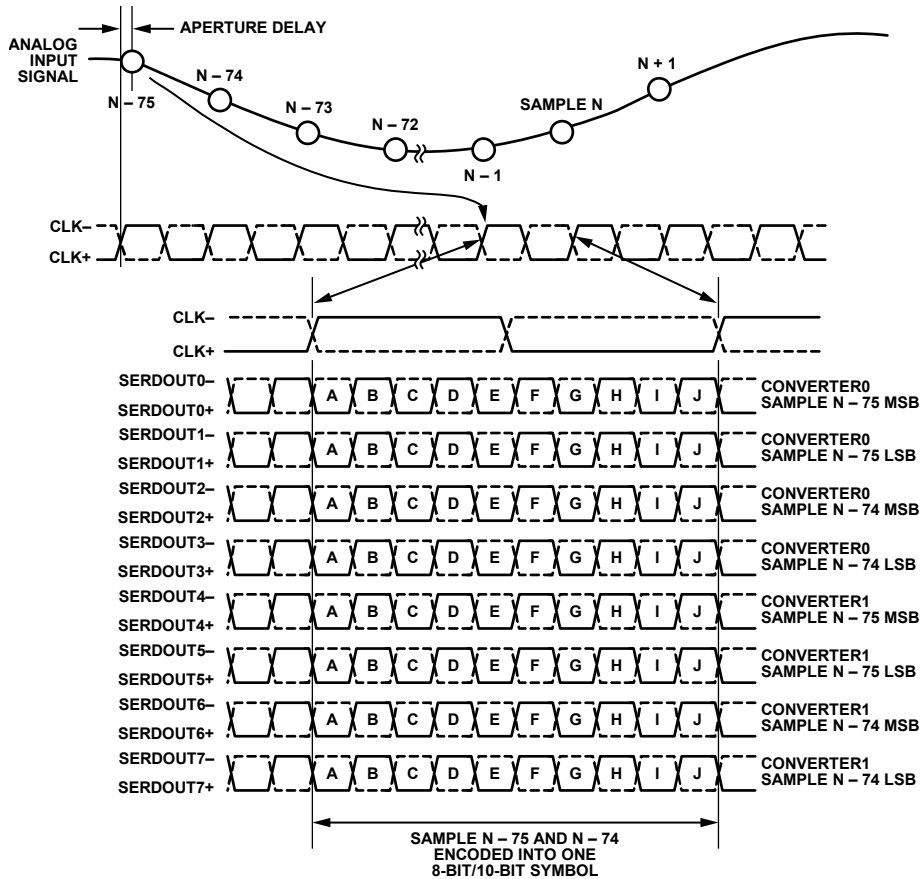


Figure 2. Data Output Timing Diagram

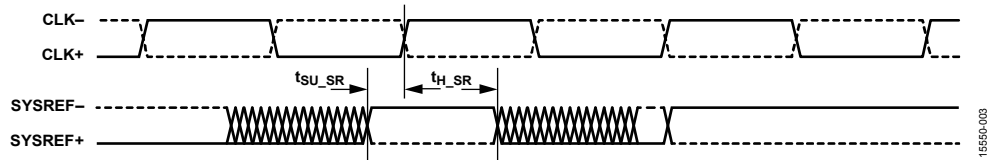


Figure 3. CLK+ to SYSREF+ Setup and Hold Timing Diagram

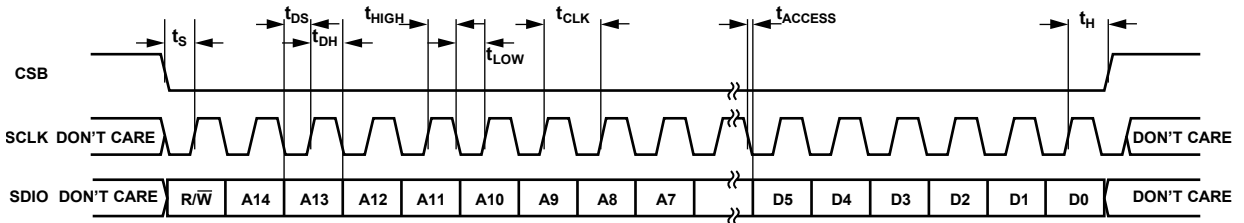


Figure 4. SPI Interface Timing Diagram

ABSOLUTE MAXIMUM RATINGS

Table 6.

Parameter	Rating
Electrical	
AVDD1 to AGND	1.05 V
AVDD1_SR to AGND	1.05 V
AVDD2 to AGND	2.0 V
AVDD3 to AGND	2.70 V
DVDD to DGND	1.05 V
DRVDD1 to DRGND	1.05 V
DRVDD2 to DRGND	2.0 V
SPIVDD to DGND	2.0 V
AGND to DRGND	-0.3 V to +0.3 V
AGND to DGND	-0.3 V to +0.3 V
DGND to DRGND	-0.3 V to +0.3 V
VIN±x to AGND	AGND - 0.3 V to AVDD3 + 0.3 V
CLK± to AGND	AGND - 0.3 V to AVDD1 + 0.3 V
SCLK, SDIO, CSB to DGND	DGND - 0.3 V to SPIVDD + 0.3 V
PDWN/STBY to DGND	DGND - 0.3 V to SPIVDD + 0.3 V
SYSREF± to AGND	2.5 V
SYNCINB± to DRGND	2.5 V
Junction Temperature Range (T _J)	-40°C to +125°C
Storage Temperature Range, Ambient (T _A)	-65°C to +150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Close attention to PCB thermal design is required. θ_{JA} is the natural convection junction-to-ambient thermal resistance measured in a one cubic foot sealed enclosure. θ_{JC} is the junction to case thermal resistance.

Table 7. Thermal Resistance

Package Type	θ_{JA}	θ_{JC_TOP}	Ψ_{JB}	Unit
BP-196-4 ¹	16.26	1.4	5.44	°C/W

¹ Test Condition 1: Thermal impedance simulated values are based on JEDEC 252P thermal test board with 190 thermal vias. See JEDEC JESD51.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

	1	2	3	4	5	6	7	8	9	10	11	12	13	14
A	AVDD2	AVDD2	AVDD1	AVDD1 ¹	AVDD1 ¹	AGND ¹	CLK+	CLK-	AGND ¹	AVDD1 ¹	AVDD1 ¹	AVDD1	AVDD2	AVDD2
B	AVDD2	AVDD2	AVDD1	AVDD1 ¹	AGND	AGND ¹	AGND ¹	AGND ¹	AGND ¹	AGND	AVDD1 ¹	AVDD1	AVDD2	AVDD2
C	AVDD2	AVDD2	AVDD1	AGND	AGND	AGND ¹	AGND ¹	AGND ¹	AGND ¹	AGND	AGND	AVDD1	AVDD2	AVDD2
D	AVDD3	AGND	AGND	AGND	AGND	AGND	AGND ¹	AGND ¹	AGND	AGND	AGND	AGND	AGND	AVDD3
E	VIN-B	AGND	AGND	AGND	AGND	AGND ²	AVDD1_SR	AGND ²	AGND	AGND	AGND	AGND	AGND	VIN-A
F	VIN+B	AGND	AGND	AGND	AGND	AGND	SYSREF+	SYSREF-	AGND	AGND	AGND	AGND	AGND	VIN+A
G	AVDD3	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AVDD3
H	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	VREF	AGND	AGND	AGND	AGND
J	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND	AGND
K	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³	AGND ³
L	DGND	GPIO_B1	SPIVDD	FD_B/ GPIO_B0	CSB	SCLK	SDIO	PDWN/ STBY	FD_A/ GPIO_A0	SPIVDD	GPIO_A1	DGND	DGND	DGND
M	DGND	DGND	DRGND	DRGND	DRVDD1	DRVDD1	DRVDD1	DRVDD1	DRGND	DRGND	DRVDD1	DRGND	DRVDD2	DVDD
N	DVDD	DVDD	DRGND	SERDOUT7+	SERDOUT6+	SERDOUT5+	SERDOUT4+	SERDOUT3+	SERDOUT2+	SERDOUT1+	SERDOUT0+	DRGND	SYNCINB+	DVDD
P	DVDD	DVDD	DRGND	SERDOUT7-	SERDOUT6-	SERDOUT5-	SERDOUT4-	SERDOUT3-	SERDOUT2-	SERDOUT1-	SERDOUT0-	DRGND	SYNCINB-	DVDD

¹DENOTES CLOCK DOMAIN.
²DENOTES SYSREF± DOMAIN.
³DENOTES ISOLATION DOMAIN.

Figure 5. Pin Configuration (Top View)

15560-005

Table 8. Pin Function Descriptions¹

Pin No.	Mnemonic	Type	Description
Power Supplies			
A3, A12, B3, B12, C3, C12	AVDD1	Power	Analog Power Supply (0.975 V Nominal).
A4, A5, A10, A11, B4, B11	AVDD1 ²	Power	Analog Power Supply for the Clock Domain (0.975 V Nominal).
A1, A2, A13, A14, B1, B2, B13, B14, C1, C2, C13, C14	AVDD2	Power	Analog Power Supply (1.9 V Nominal).
D1, D14, G1, G14	AVDD3	Power	Analog Power Supply (2.5 V Nominal).
E7	AVDD1_SR	Power	Analog Power Supply for SYSREF± (0.975 V Nominal).
L3, L10	SPIVDD	Power	Digital Power Supply for SPI (1.9 V Nominal).
M14, N1, N2, N14, P1, P2, P14	DVDD	Power	Digital Power Supply (0.975 V Nominal).
M5 to M8, M11	DRVDD1	Power	Digital Driver Power Supply (0.975 V Nominal).
M13	DRVDD2	Power	Digital Driver Power Supply (1.9 V Nominal).
B5, B10, C4, C5, C10, C11, D2 to D6, D9 to D13, E2 to E5, E9 to E13, F2 to F6, F9 to F13, G2 to G13, H1 to H9, H11 to H14, J1 to J14	AGND	Ground	Analog Ground. These pins connect to the analog ground plane.
A6, A9, B6 to B9, C6 to C9, D7, D8	AGND ²	Ground	Ground Reference for the Clock Domain.
E6, E8	AGND ³	Ground	Ground Reference for SYSREF±.
K1 to K14	AGND ⁴	Ground	Isolation Ground.
L1, L12 to L14, M1, M2	DGND	Ground	Digital Control Ground Supply. These pins connect to the digital ground plane.
M3, M4, M9, M10, M12, N3, N12, P3, P12	DRGND	Ground	Digital Driver Ground Supply. These pins connect to the digital driver ground plane.
Analog			
E1, F1	VIN–B, VIN+B	Input	ADC B Analog Input Complement/True.
E14, F14	VIN–A, VIN+A	Input	ADC A Analog Input Complement/True.
A7, A8	CLK+, CLK–	Input	Clock Input True/Complement.
H10	VREF	Input/output/ do not connect (DNC)	0.50 V Reference Voltage Input/Do Not Connect. This pin is configurable through the SPI as a no connect or an input. Do not connect this pin if using the internal reference. This pin requires a 0.50 V reference voltage input if using an external voltage reference source.
CMOS Inputs/Outputs			
L2	GPIO_B1	Input/output	GPIO B1.
L4	FD_B/GPIO_B0	Input/output	Fast Detect Outputs for Channel B/GPIO B0.
L9	FD_A/GPIO_A0	Input/output	Fast Detect Outputs for Channel A/GPIO A0.
L11	GPIO_A1	Input/output	GPIO A1.
Digital Inputs			
F7, F8	SYSREF+, SYSREF–	Input	Active High JESD204B LVDS System Reference Input True/Complement.
N13	SYNCINB+	Input	Active Low JESD204B LVDS/CMOS Sync Input True.
P13	SYNCINB–	Input	Active Low JESD204B LVDS Sync Input Complement.
Data Outputs			
N4, P4	SERDOUT7+, SERDOUT7–	Output	Lane 7 Output Data True/Complement.
N5, P5	SERDOUT6+, SERDOUT6–	Output	Lane 6 Output Data True/Complement.
N6, P6	SERDOUT5+, SERDOUT5–	Output	Lane 5 Output Data True/Complement.
N7, P7	SERDOUT4+, SERDOUT4–	Output	Lane 4 Output Data True/Complement.
N8, P8	SERDOUT3+, SERDOUT3–	Output	Lane 3 Output Data True/Complement.
N9, P9	SERDOUT2+, SERDOUT2–	Output	Lane 2 Output Data True/Complement.
N10, P10	SERDOUT1+, SERDOUT1–	Output	Lane 1 Output Data True/Complement.
N11, P11	SERDOUT0+, SERDOUT0–	Output	Lane 0 Output Data True/Complement.

Pin No.	Mnemonic	Type	Description
Digital Controls			
L5	CSB	Input	SPI Chip Select (Active Low).
L6	SCLK	Input	SPI Serial Clock.
L7	SDIO	Input/output	SPI Serial Data Input/Output.
L8	PDWN/STBY	Input	Power-Down Input (Active High). The operation of this pin depends on the SPI mode and can be configured as power-down or standby.

¹ See the Theory of Operation section and the Applications Information section for more information on isolating the planes for optimal performance.

² Denotes clock domain.

³ Denotes SYSREF± domain.

⁴ Denotes isolation domain.

TYPICAL PERFORMANCE CHARACTERISTICS

2.0 GSPS

AVDD1 = 0.975 V, AVDD1_SR = 0.975 V, AVDD2 = 1.9 V, AVDD3 = 2.5 V, DVDD = 0.975 V, DRVDD1 = 0.975 V, DRVDD2 = 1.9 V, SPIVDD = 1.9 V, sampling rate = 2.0 GHz, clock divider = 2, 1.7 V p-p full-scale differential input, input amplitude (A_{IN}) = -2.0 dBFS, $T_J = 70^\circ\text{C}$ ($T_A = 25^\circ\text{C}$), 128k fast Fourier transform (FFT) sample, unless otherwise noted. See Table 10 for the recommended settings.

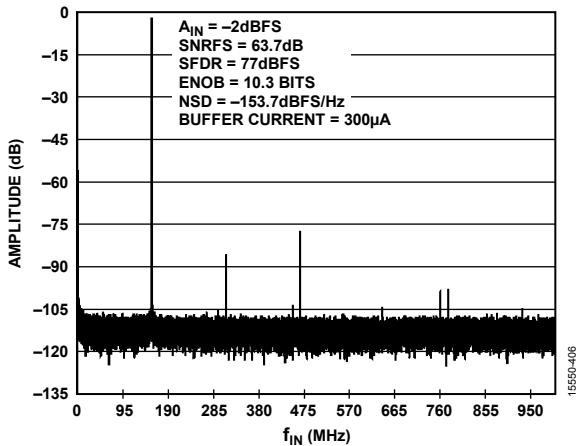


Figure 6. Single-Tone FFT at $f_{IN} = 155$ MHz

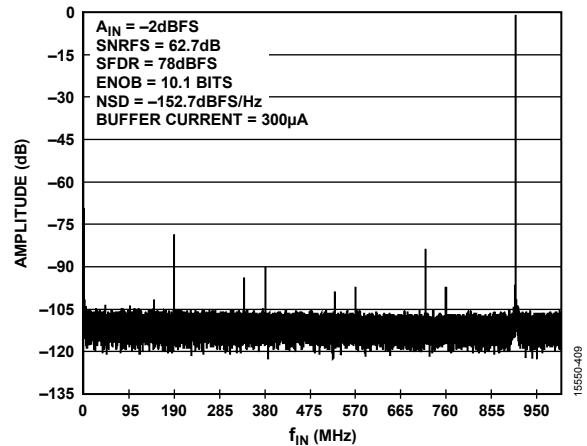


Figure 9. Single-Tone FFT at $f_{IN} = 905$ MHz

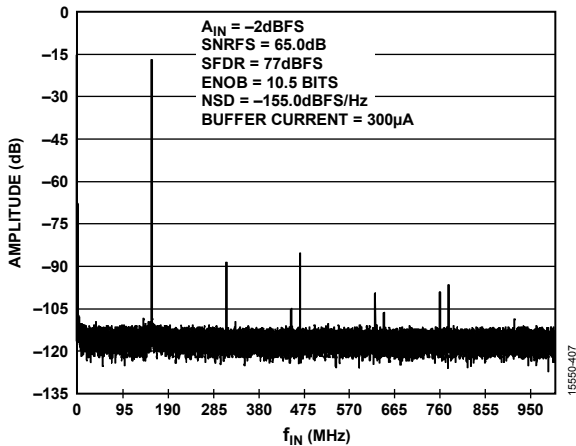


Figure 7. Single-Tone FFT at $f_{IN} = 155$ MHz, Full-Scale Voltage = 2.04 V p-p

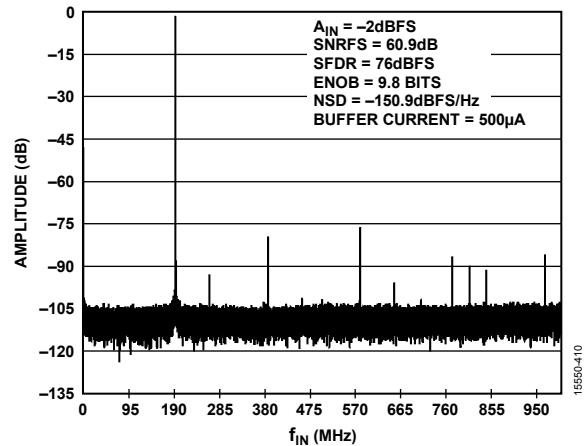


Figure 10. Single-Tone FFT at $f_{IN} = 1807$ MHz

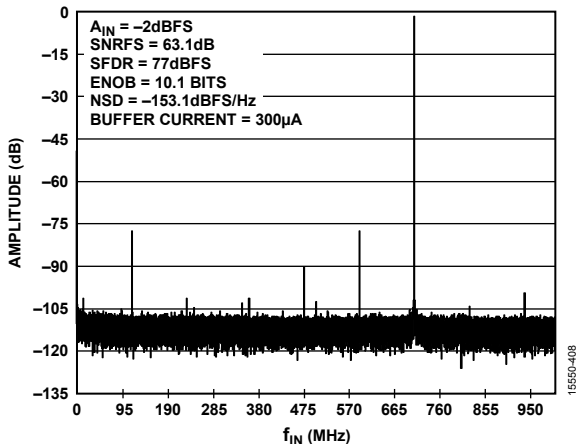


Figure 8. Single-Tone FFT at $f_{IN} = 750$ MHz

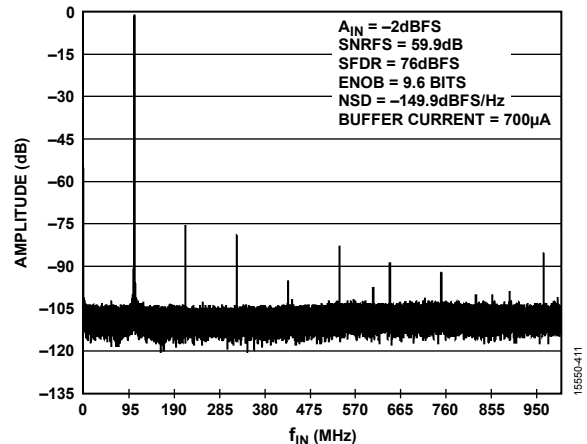


Figure 11. Single-Tone FFT at $f_{IN} = 2100$ MHz

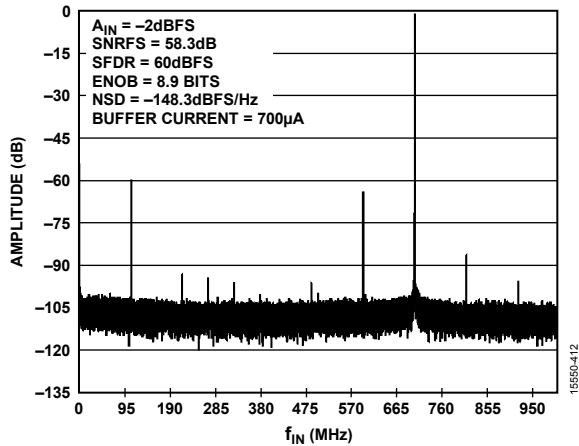


Figure 12. Single-Tone FFT at $f_{IN} = 3300 \text{ MHz}$

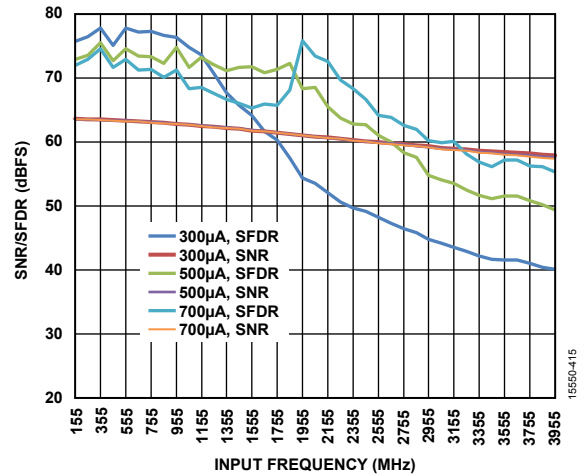


Figure 15. SNR/SFDR vs. Input Frequency (f_{IN}) for Various Buffer Currents

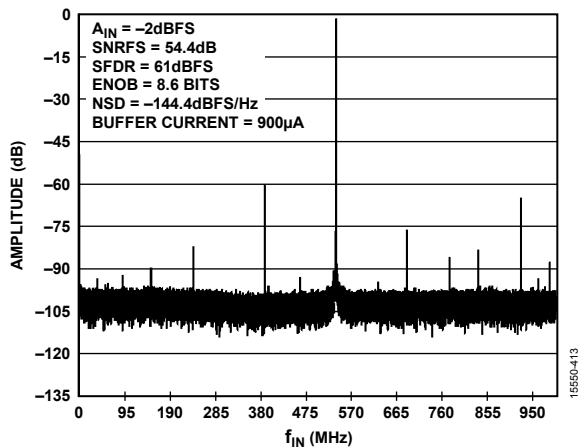


Figure 13. Single-Tone FFT at $f_{IN} = 4350 \text{ MHz}$; Full-Scale Voltage = 1.1 V_{p-p}

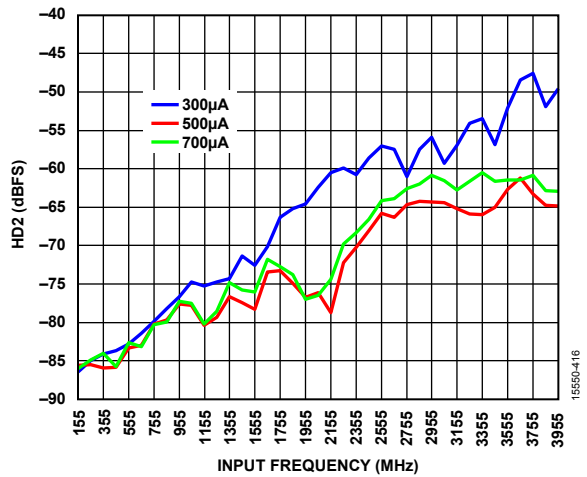


Figure 16. HD2 vs. Input Frequency (f_{IN}) for Various Buffer Currents

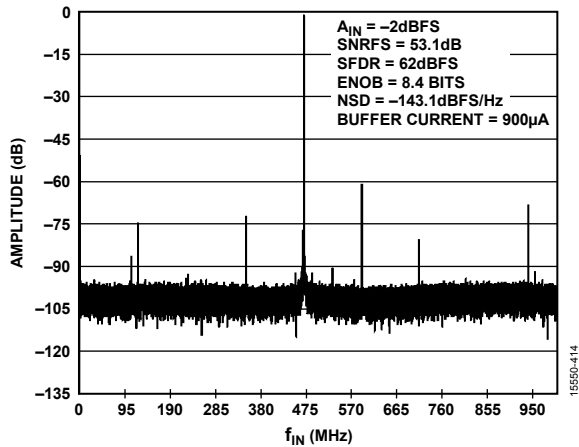


Figure 14. Single-Tone FFT at $f_{IN} = 5400 \text{ MHz}$; Full-Scale Voltage = 1.1 V_{p-p}

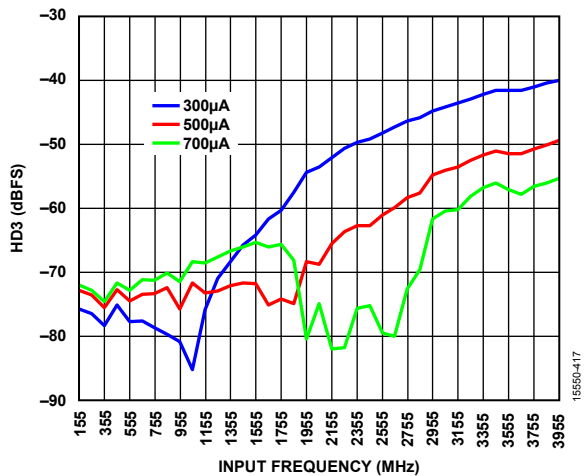


Figure 17. HD3 vs. Input Frequency (f_{IN}) for Various Buffer Currents

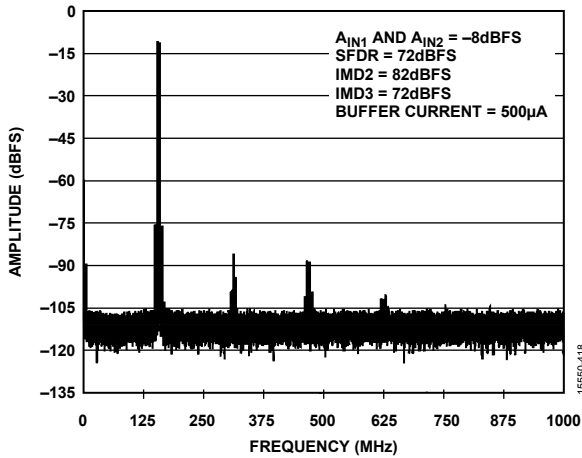


Figure 18. Two-Tone FFT; $f_{IN1} = 1841$ MHz, $f_{IN2} = 1846$ MHz; A_{IN1} and $A_{IN2} = -8$ dBFS

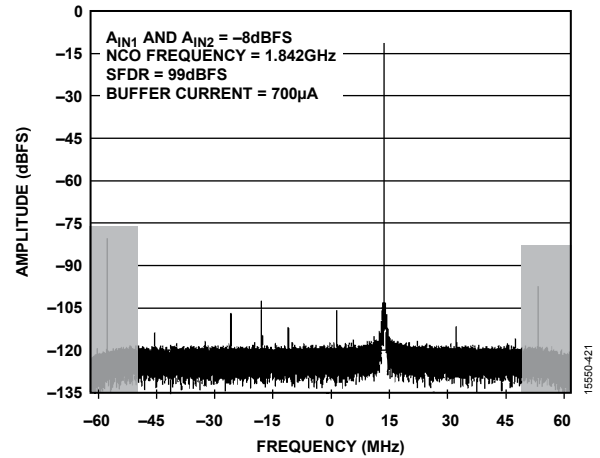


Figure 21. Two-Tone FFT; $f_{IN1} = 947.5$ MHz, $f_{IN2} = 1855.5$ MHz $f_{CLK} = 1.96608$ GHz; Decimation Ratio = 16, NCO Frequency = 1842.5 MHz

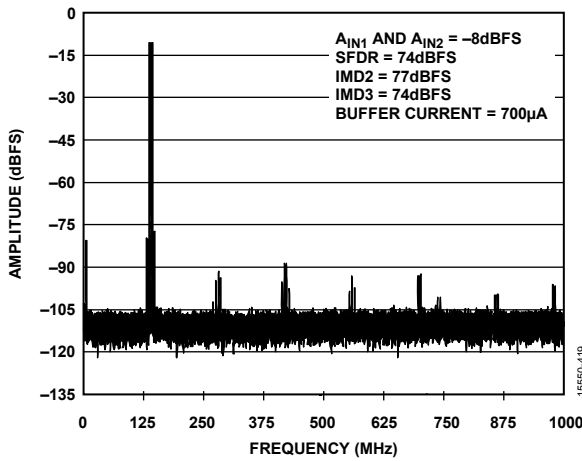


Figure 19. Two-Tone FFT; $f_{IN1} = 2137$ MHz, $f_{IN2} = 2142$ MHz; A_{IN1} and $A_{IN2} = -8$ dBFS

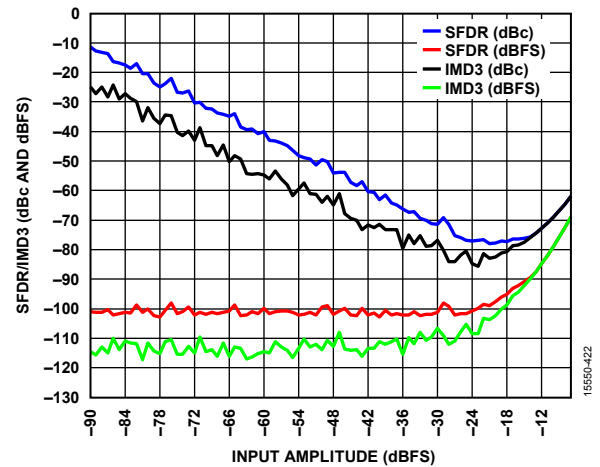


Figure 22. Two-Tone SFDR/IMD3 vs. Input Amplitude (A_{IN}) with $f_{IN1} = 1841.5$ MHz, $f_{IN2} = 1846.5$ MHz

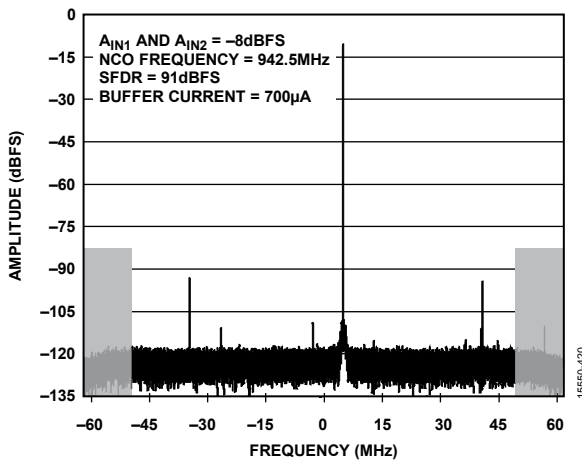


Figure 20. Two-Tone FFT; $f_{IN1} = 947.5$ MHz, $f_{IN2} = 1855.5$ MHz $f_{CLK} = 1.96608$ GHz; Decimation Ratio = 16, NCO Frequency = 942.5 MHz

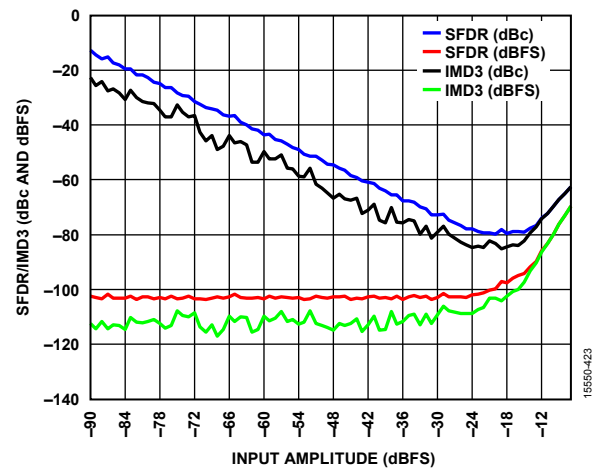


Figure 23. Two-Tone SFDR/IMD3 vs. Input Amplitude (A_{IN}) with $f_{IN1} = 2137.5$ MHz, $f_{IN2} = 2142.5$ MHz

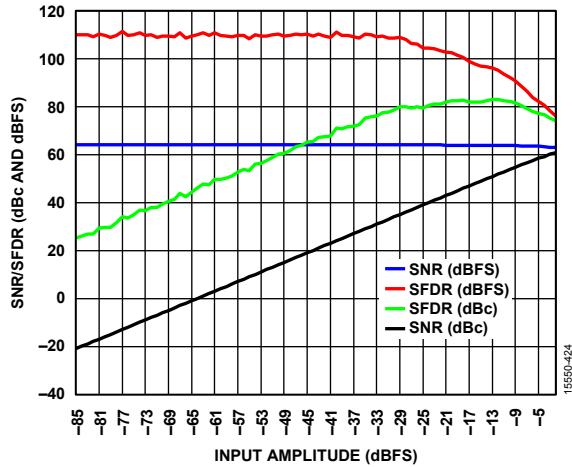


Figure 24. SNR/SFDR vs. Input Amplitude (A_{IN}), $f_{IN} = 900$ MHz

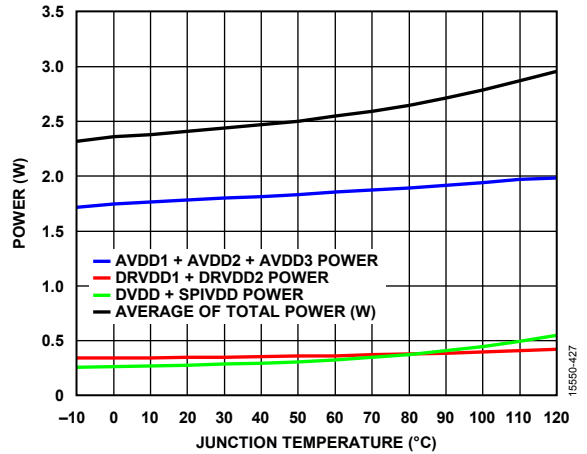


Figure 27. Power vs. Junction Temperature (T_J), $f_{IN} = 900$ MHz

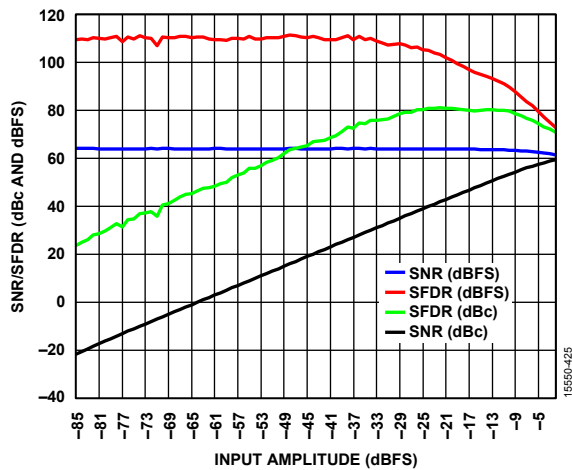


Figure 25. SNR/SFDR vs. Input Amplitude (A_{IN}), $f_{IN} = 1800$ MHz

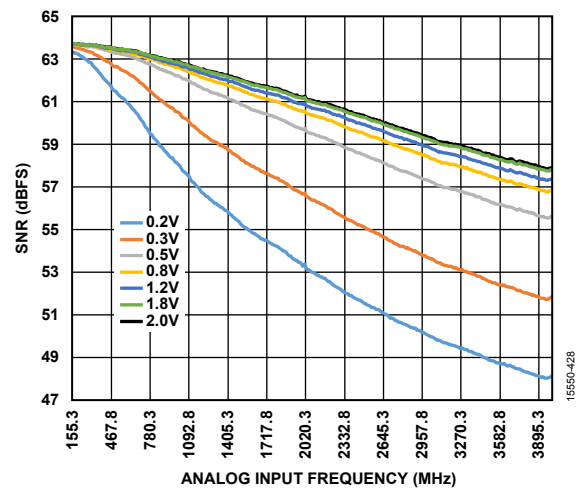


Figure 28. SNR vs. Analog Input Frequency (f_{IN}) for Various Clock Amplitude in Differential Peak-to-Peak Voltages

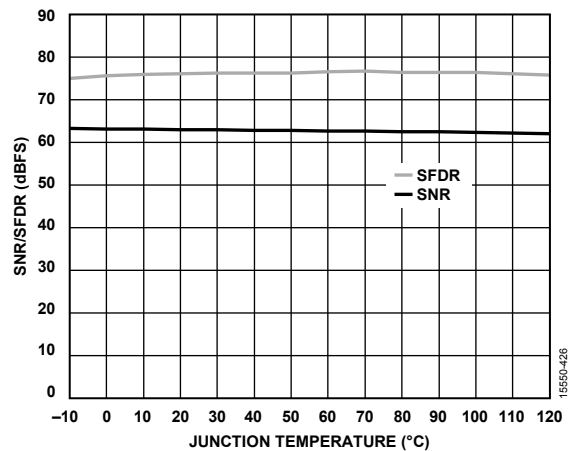


Figure 26. SNR/SFDR vs. Junction Temperature (T_J), $f_{IN} = 900$ MHz

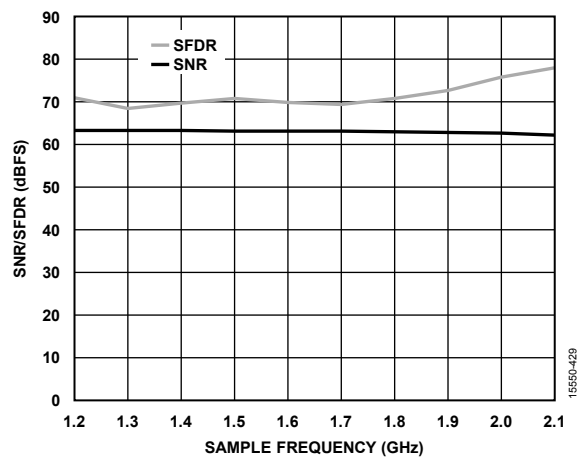


Figure 29. SNR/SFDR vs. Sample Frequency (f_s), $f_{IN} = 900$ MHz

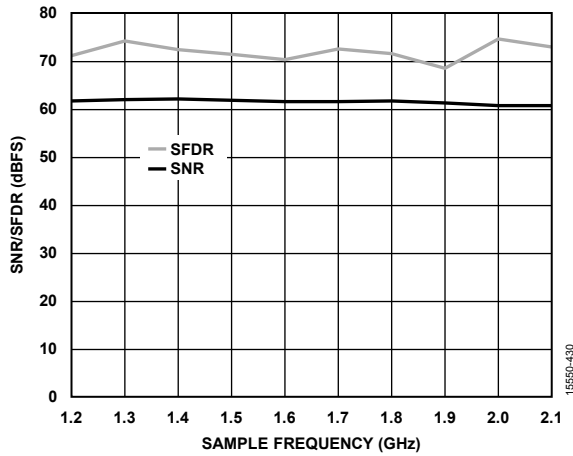


Figure 30. SNR/SFDR vs. Sample Frequency (f_s), $f_{IN} = 1.8$ GHz

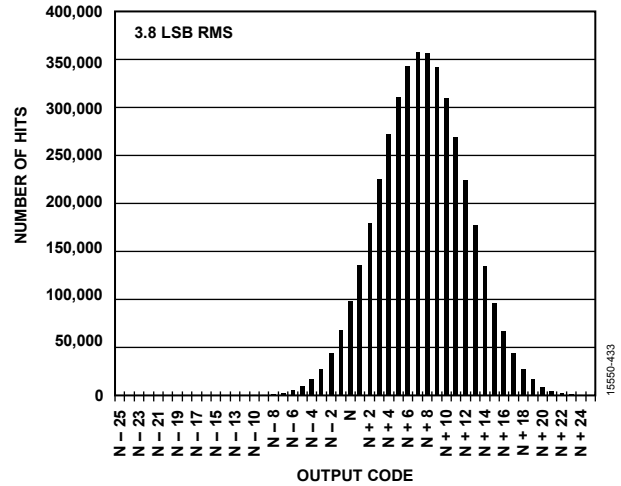


Figure 33. Input Referred Noise Histogram

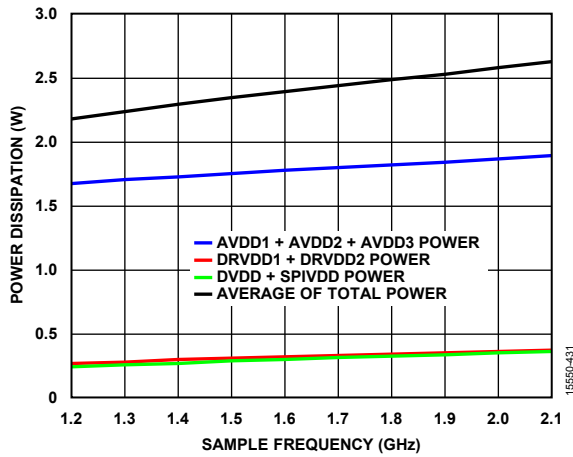


Figure 31. Power Dissipation vs. Sample Frequency (f_s), $f_{IN} = 1.8$ GHz

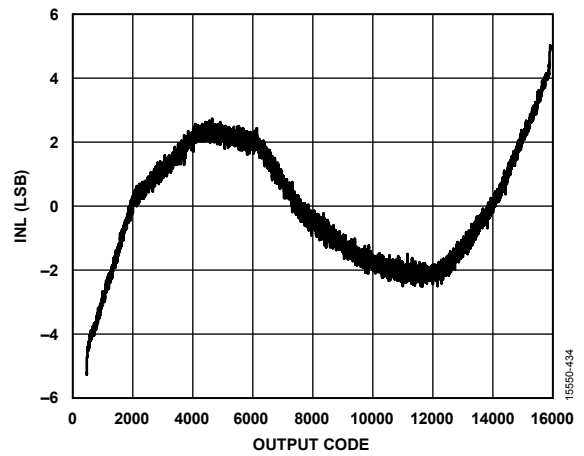


Figure 34. INL, $f_{IN} = 155$ MHz

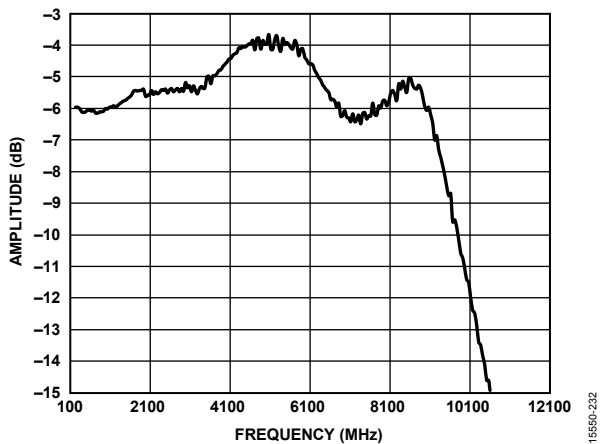


Figure 32. Input Bandwidth (See Figure 80 for the Input Configuration)

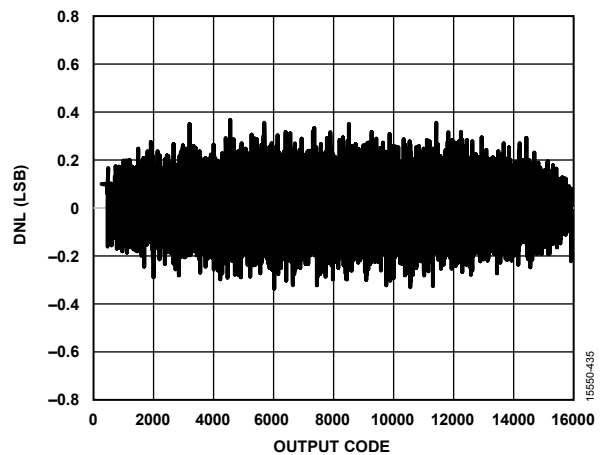


Figure 35. DNL, $f_{IN} = 155$ MHz

2.6 GSPS

AVDD1 = 0.975 V, AVDD1_SR = 0.975 V, AVDD2 = 1.9 V, AVDD3 = 2.5 V, DVDD = 0.975 V, DRVDD1 = 0.975 V, DRVDD2 = 1.9 V, SPIVDD = 1.9 V, sampling rate = 2.56 GHz, clock divider = 2, 1.7 V p-p full-scale differential input, input amplitude (A_{IN}) = -2.0 dBFS, $T_J = 70^\circ\text{C}$ ($T_A = 25^\circ\text{C}$), 128 k FFT sample, unless otherwise noted. See Table 10 for the recommended settings.

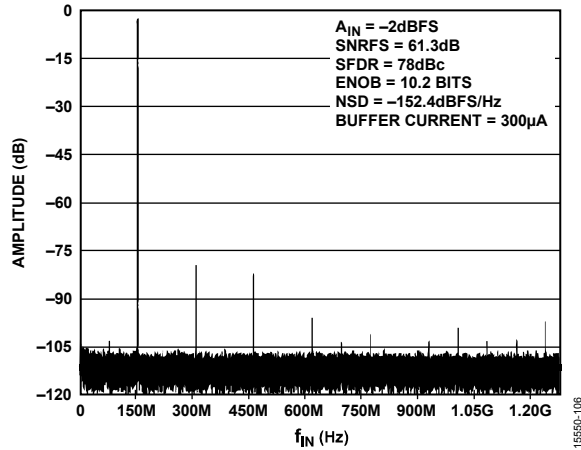


Figure 36. Single-Tone FFT at $f_{IN} = 155$ MHz

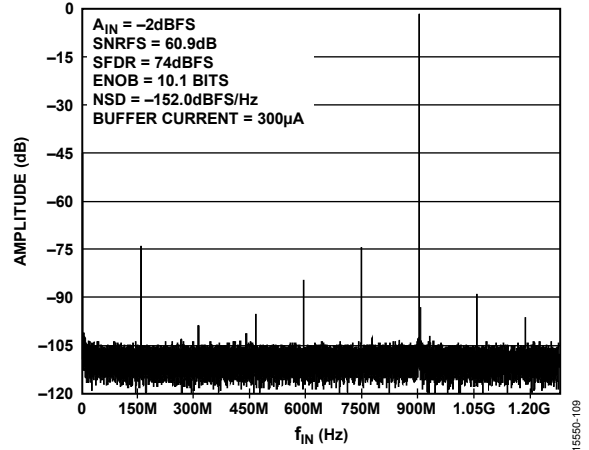


Figure 39. Single-Tone FFT at $f_{IN} = 905$ MHz

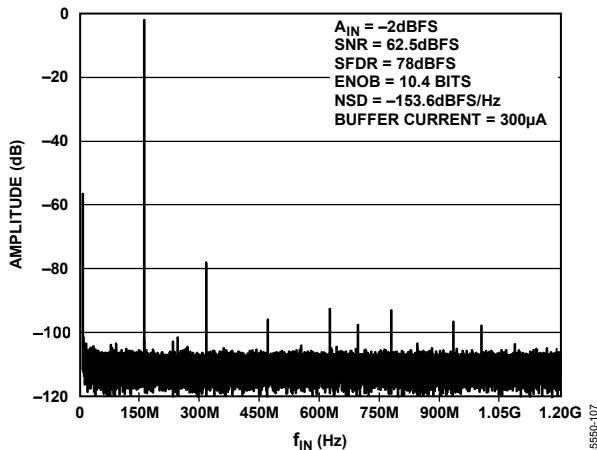


Figure 37. Single-Tone FFT at $f_{IN} = 155$ MHz, Full-Scale Voltage = 2.04 V p-p

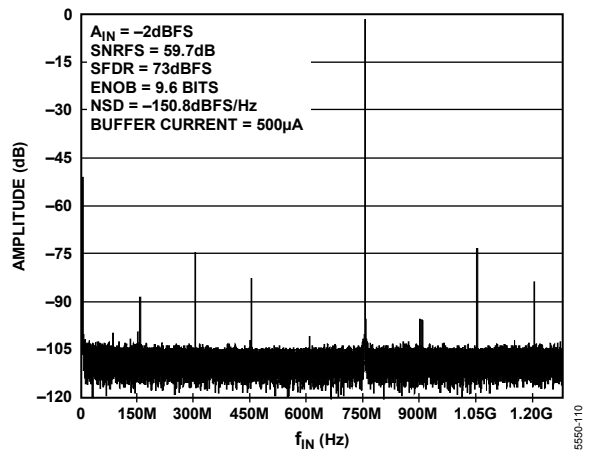


Figure 40. Single-Tone FFT at $f_{IN} = 1807$ MHz

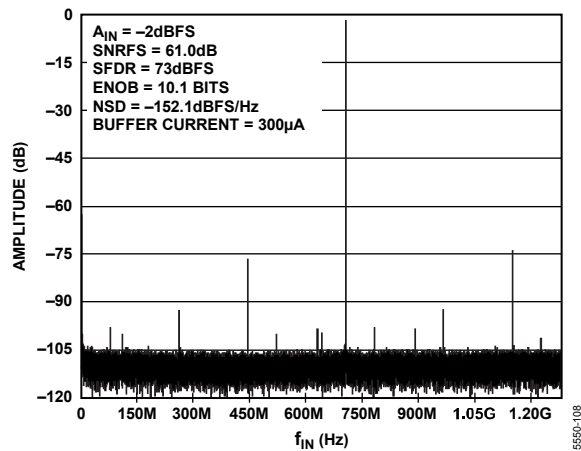


Figure 38. Single-Tone FFT at $f_{IN} = 750$ MHz

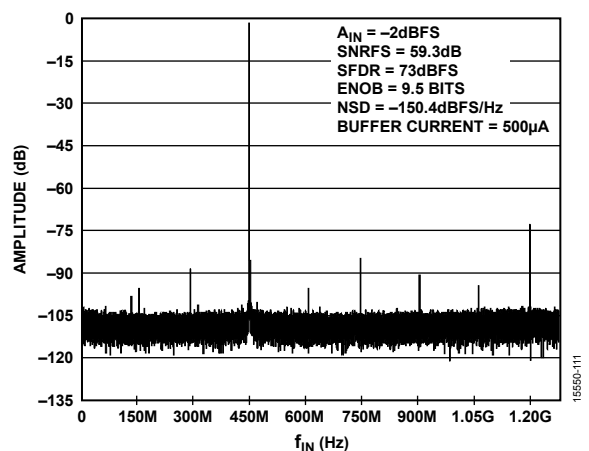


Figure 41. Single-Tone FFT at $f_{IN} = 2100$ MHz

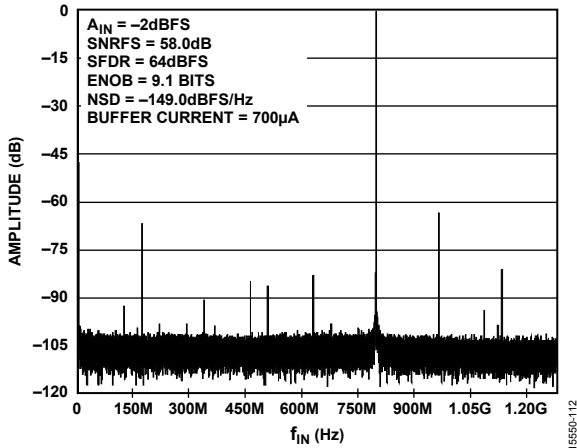


Figure 42. Single-Tone FFT at $f_{IN} = 3300$ MHz

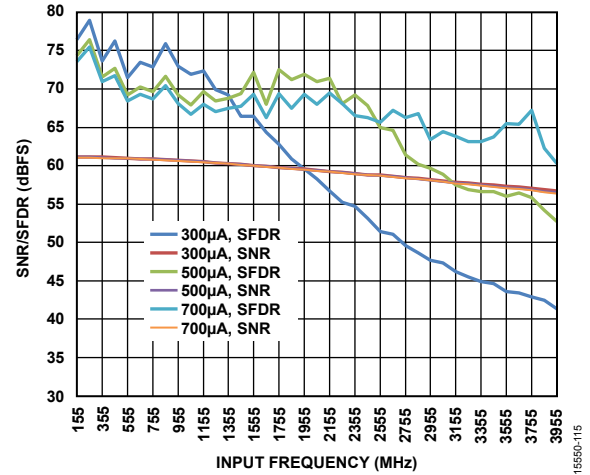


Figure 45. SNR/SFDR vs. Input Frequency (f_{IN}) for Various Buffer Currents

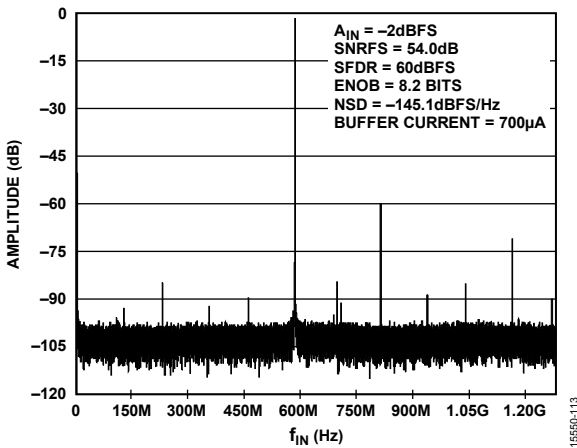


Figure 43. Single-Tone FFT at $f_{IN} = 4350$ MHz; Full-Scale Voltage = 1.1 V p-p

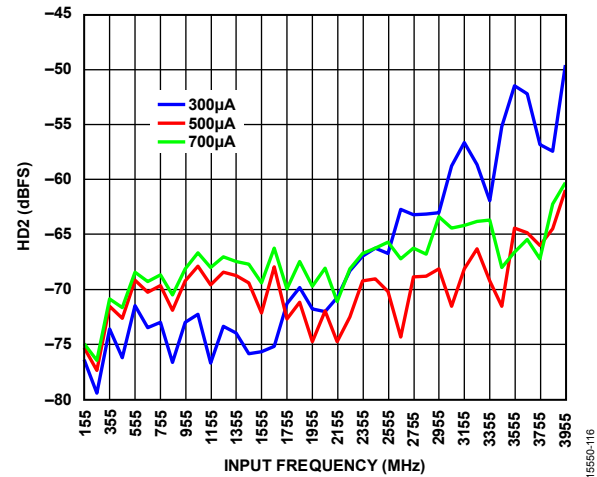


Figure 46. Second Harmonics (HD_2) vs. Input Frequency (f_{IN}) for Various Buffer Currents

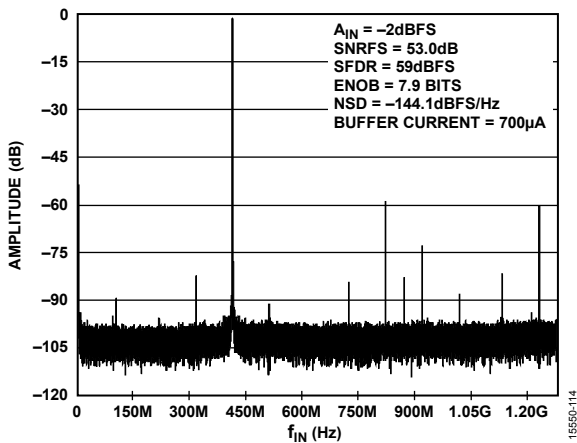


Figure 44. Single-Tone FFT at $f_{IN} = 5400$ MHz; Full-Scale Voltage = 1.1 V p-p

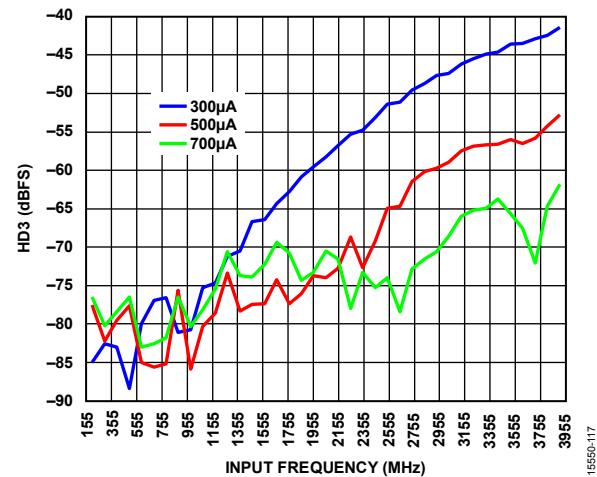


Figure 47. Third Harmonics (HD_3) vs. Input Frequency (f_{IN}) for Various Buffer Currents

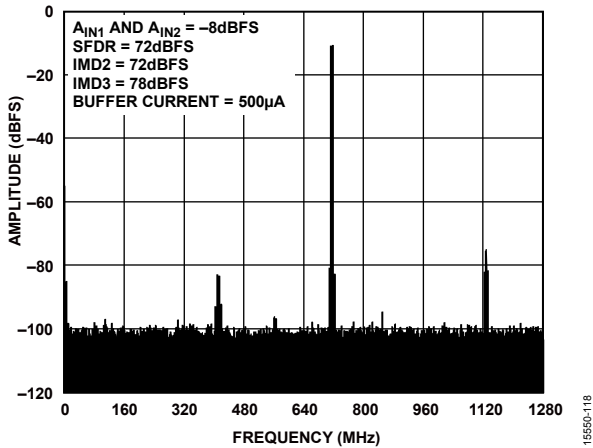


Figure 48. Two-Tone FFT; $f_{IN1} = 1841$ MHz, $f_{IN2} = 1846$ MHz; A_{IN1} and $A_{IN2} = -8$ dBFS

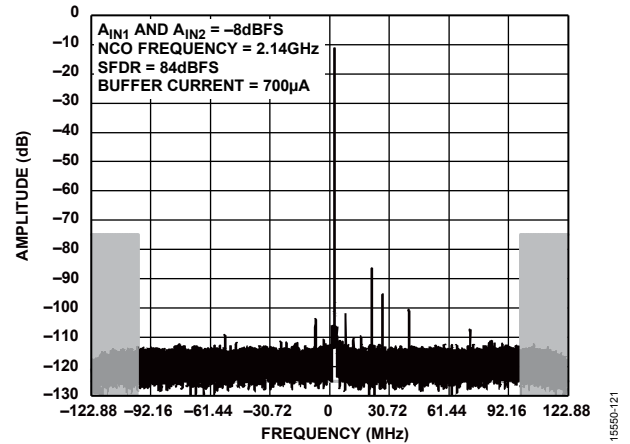


Figure 51. Two-Tone FFT; $f_{IN1} = 1846.5$ MHz, $f_{IN2} = 2142.5$ MHz; $f_{CLK} = 2.4576$ GHz; Decimation Ratio = 10, NCO Frequency = 2140 MHz

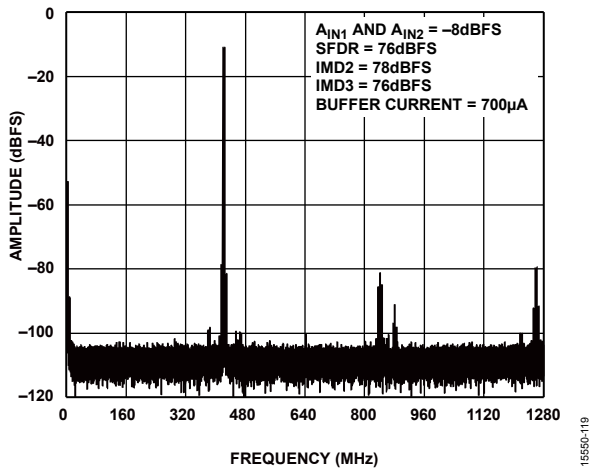


Figure 49. Two-Tone FFT; $f_{IN1} = 2137$ MHz, $f_{IN2} = 2142$ MHz; A_{IN1} and $A_{IN2} = -8$ dBFS

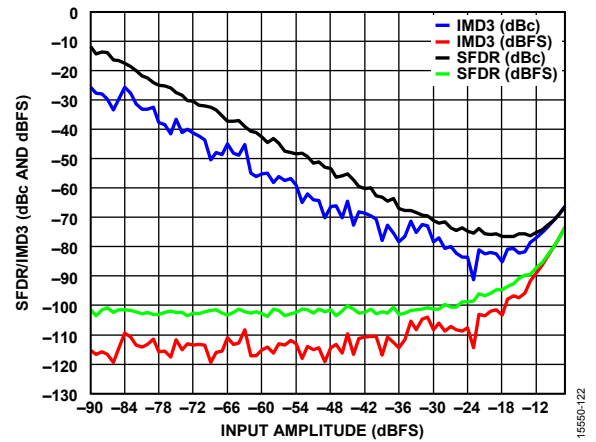


Figure 52. Two-Tone SFDR/IMD3 vs. Input Amplitude (A_{IN}) with $f_{IN1} = 1841.5$ MHz, $f_{IN2} = 1846.5$ MHz

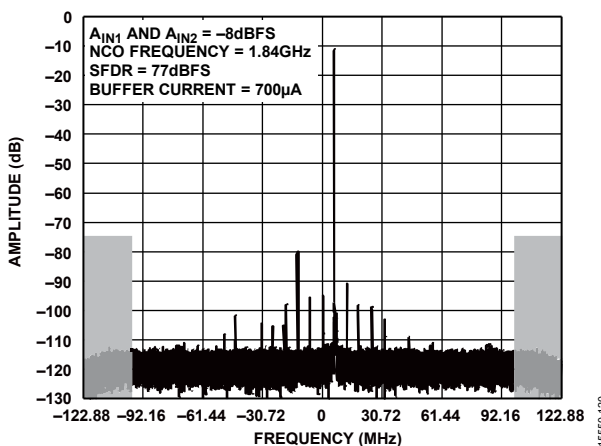


Figure 50. Two-Tone FFT; $f_{IN1} = 1846.5$ MHz, $f_{IN2} = 2142.5$ MHz; $f_{CLK} = 2.4576$ GHz; Decimation Ratio = 10, NCO Frequency = 1840 MHz

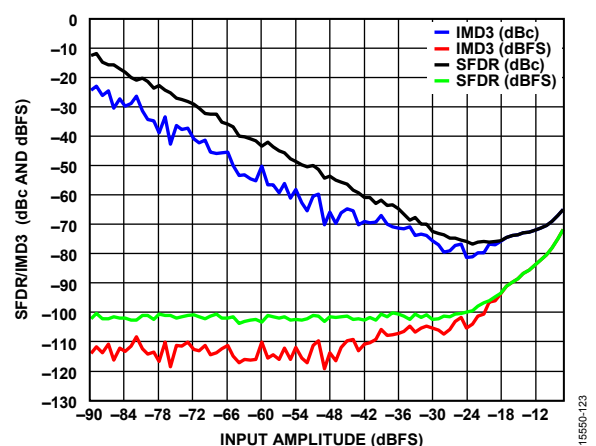


Figure 53. Two-Tone SFDR/IMD3 vs. Input Amplitude (A_{IN}) with $f_{IN1} = 2137.5$ MHz, $f_{IN2} = 2142.5$ MHz

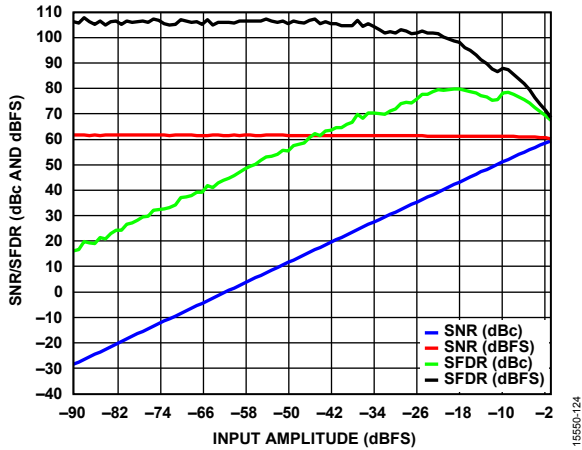


Figure 54. SNR/SFDR vs. Input Amplitude (A_{IN}), $f_{IN} = 900$ MHz

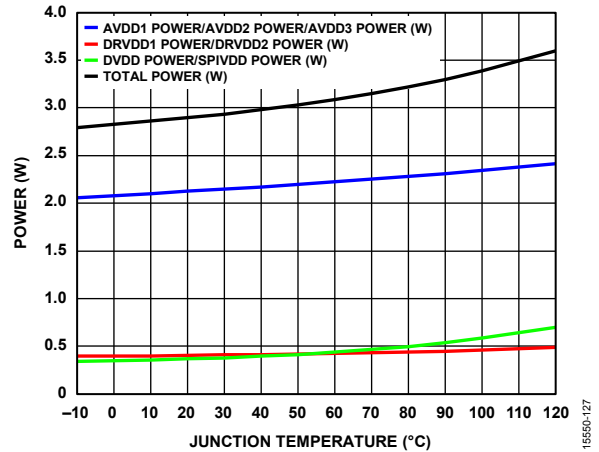


Figure 57. Power vs. Junction Temperature (T_J), $f_{IN} = 900$ MHz

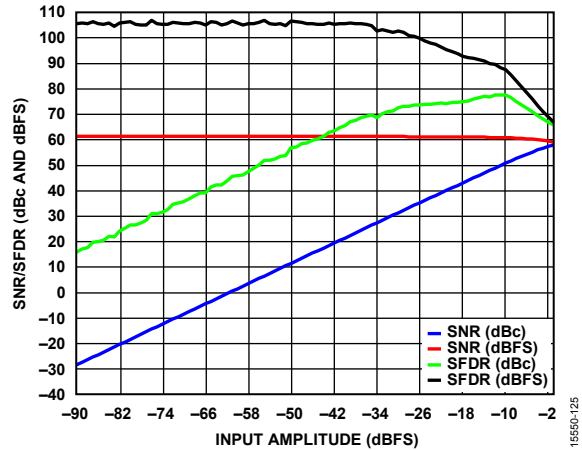


Figure 55. SNR/SFDR vs. Input Amplitude (A_{IN}), $f_{IN} = 1800$ MHz

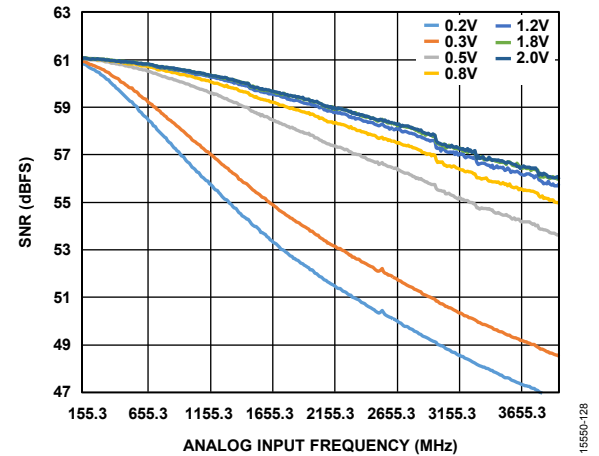


Figure 58. SNR vs. Analog Input Frequency (f_{IN}) for Various Clock Amplitude in Differential Peak-to-Peak Voltages

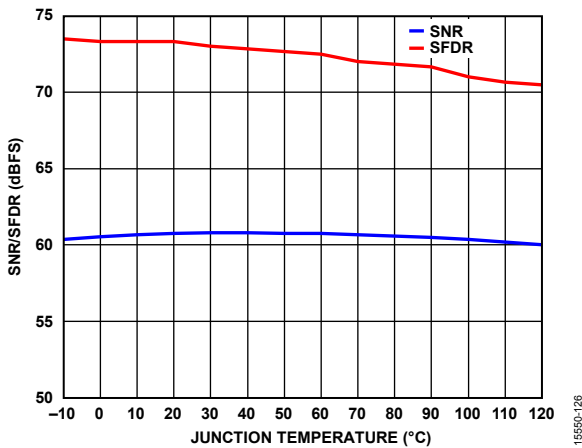


Figure 56. SNR/SFDR vs. Junction Temperature (T_J), $f_{IN} = 900$ MHz

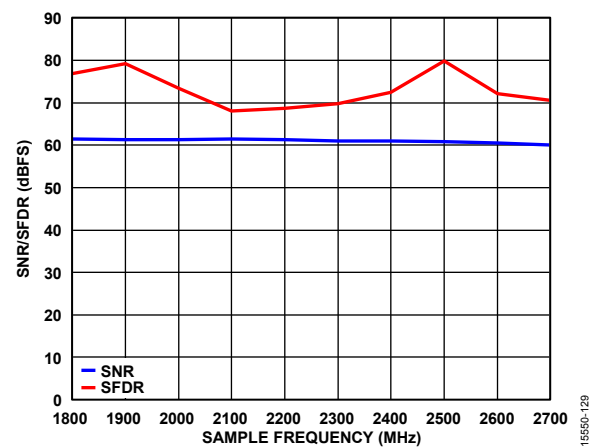


Figure 59. SNR/SFDR vs. Sample Frequency (f_s), $f_{IN} = 900$ MHz

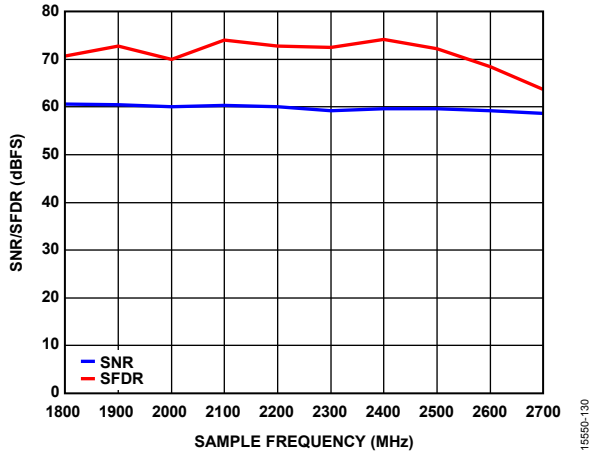


Figure 60. SNR/SFDR vs. Sample Frequency (f_s), $f_{IN} = 1.8$ GHz

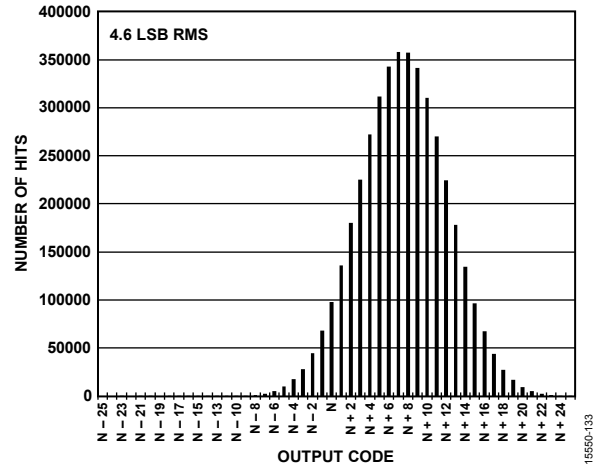


Figure 63. Input Referred Noise Histogram

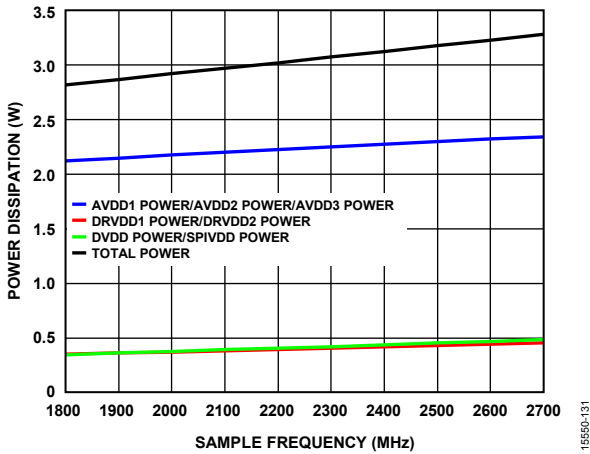


Figure 61. Power Dissipation vs. Sample Frequency (f_s), $f_{IN} = 1.8$ GHz

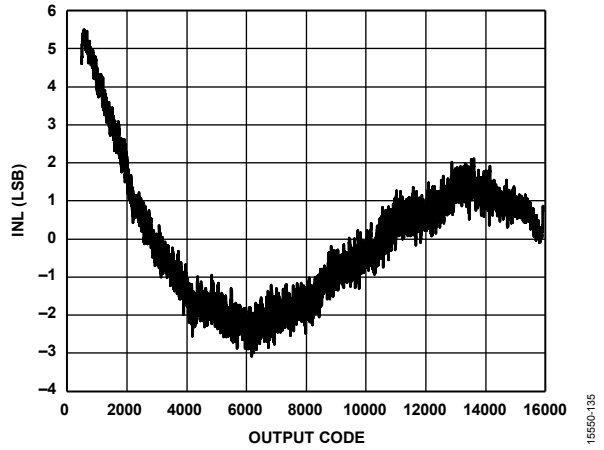


Figure 64. INL, $f_{IN} = 155$ MHz

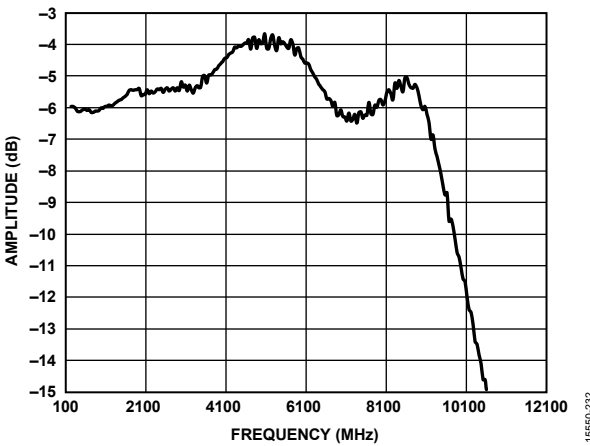


Figure 62. Input Bandwidth (See Figure 80 for the Input Configuration)

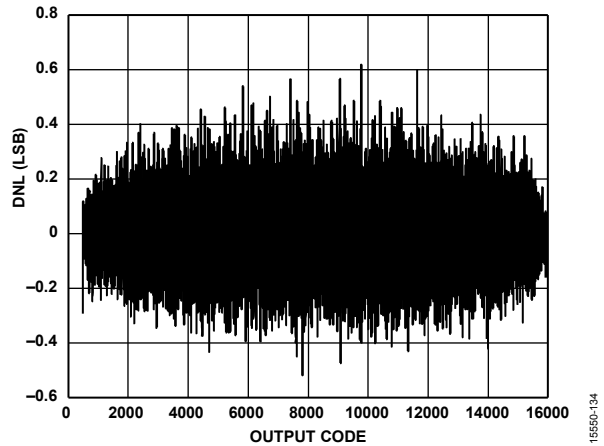


Figure 65. DNL, $f_{IN} = 155$ MHz