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### FEATURES

- RF frequency: 700 MHz to 3000 MHz, continuous
- LO input frequency: 200 MHz to 2700 MHz, high-side or low-side injection
- IF range: 40 MHz to 500 MHz
- Power conversion gain of 9.0 dB
- Phase noise performance of  $-144$  dBc/Hz at 800 kHz offset supporting stringent GSM standards in both 800 MHz to 900 MHz and 1800 MHz to 1900 MHz bands
- Single-sideband (SSB) noise figure of 11.3 dB
- Input IP3 of 30 dBm
- Input P1dB of 10.6 dBm
- Typical LO input drive of 0 dBm
- Single-ended, 50  $\Omega$  RF port
- Single-ended or balanced LO input port
- Serial port interface (SPI) control on all functions
- Exposed pad, 7 mm  $\times$  7 mm, 48-lead LFCSP

### APPLICATIONS

- Multiband/multistandard cellular base station diversity receivers
- Wideband radio link diversity downconverters
- Multimode cellular extenders and picocells

### GENERAL DESCRIPTION

The **ADRF6614** is a dual radio frequency (RF) mixer and intermediate frequency (IF) amplifier with an integrated phase-locked loop (PLL) and voltage controlled oscillators (VCOs). The **ADRF6614** uses revolutionary broadband square wave limiting local oscillator (LO) amplifiers to achieve a wideband RF bandwidth of 700 MHz to 3000 MHz. Unlike narrow-band sine wave LO amplifier solutions, the LO can be applied above or below the RF input over a wide bandwidth. Energy storage elements are not utilized in the LO amplifier, thus dc current consumption also decreases with decreasing LO frequency.

The **ADRF6614** utilizes highly linear, doubly balanced passive mixer cores with integrated RF and LO balancing circuits to allow single-ended operation. Integrated RF baluns allow optimal performance over the 700 MHz to 3000 MHz RF input frequency. The balanced passive mixer arrangement provides outstanding LO to RF and LO to IF leakages, excellent RF to IF isolation, and excellent intermodulation performance over the full RF bandwidth.

The balanced mixer cores provide extremely high input linearity, allowing the device to be used in demanding wideband applications where in-band blocking signals may otherwise result in the degradation of dynamic range. Noise performance under blocking is comparable to narrow-band passive mixer designs. High linearity

### FUNCTIONAL BLOCK DIAGRAM

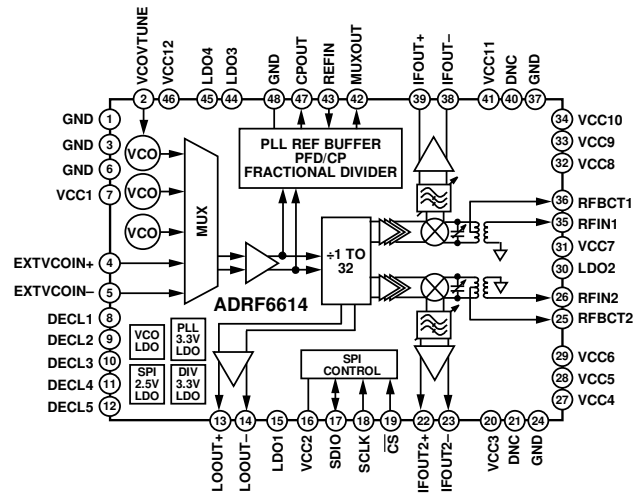


Figure 1.

IF buffer amplifiers follow the passive mixer cores, yielding typical power conversion gains of 9.0 dB, and can be matched to a wide range of output impedances.

The PLL architecture supports both integer-N and fractional-N operation and can generate the entire LO frequency range of 200 MHz to 2700 MHz using an external reference input frequency anywhere in the range of 12 MHz to 320 MHz. An external loop filter provides flexibility in trading off phase noise vs. acquisition time. To reduce fractional spurs in fractional-N mode, a  $\Sigma$ - $\Delta$  modulator controls the post VCO-programmable divider. The device integrates six VCO cores, four of which provide complete frequency coverage between 200 MHz and 2700 MHz, and meet the GSM phase noise requirements in the 800 MHz and 900 MHz bands. Two additional GSM only cores enable the **ADRF6614** to meet the GSM phase noise requirements in the digital cellular system 1800 MHz (DCS1800) and personal communications service 1900 MHz (PCS1900) bands.

All features of the **ADRF6614** are controlled via a 3-wire SPI, resulting in optimum performance and minimum external components.

The **ADRF6614** is fabricated using a BiCMOS, high performance IC process. The device is available in a 7 mm  $\times$  7 mm, 48-lead LFCSP package and operates over a  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  temperature range. An evaluation board is available.

Rev. 0

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## COMPARABLE PARTS

View a parametric search of comparable parts.

## EVALUATION KITS

- ADRF6614 Evaluation Board

## DOCUMENTATION

### Data Sheet

- ADRF6614: 700 MHz to 3000 MHz, Dual Passive Receive Mixer with Integrated PLL and VCO Data Sheet

### User Guides

- UG-968: Evaluating the ADRF6612/ADRF6614, 700 MHz to 3000 MHz Rx Dual Mixer with Integrated Fractional-N PLL and VCO

## DESIGN RESOURCES

- ADRF6614 Material Declaration
- PCN-PDN Information
- Quality And Reliability
- Symbols and Footprints

## DISCUSSIONS

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## TABLE OF CONTENTS

Features .....	1	Spurious Performance .....	32
Applications.....	1	Theory of Operation .....	34
Functional Block Diagram .....	1	RF Subsystem.....	34
General Description .....	1	External LO Generation .....	34
Revision History .....	2	Internal LO Generation .....	34
Specifications.....	3	Applications Information .....	38
RF Specifications .....	3	Basic Connections by Pin Description .....	39
Synthesizer/PLL Specifications.....	4	Mixer Optimization .....	40
VCO Specifications, Open-Loop.....	7	RF Input Balun Insertion Loss Optimization.....	40
Logic Input and Power Specifications .....	8	IIP3 Optimization .....	40
Digital Logic Specifications.....	9	VGS Programming.....	41
Absolute Maximum Ratings.....	10	Low-Pass Filter Programming.....	41
Thermal Resistance .....	10	GSM Mode of Operation.....	43
ESD Caution.....	10	Register Summary .....	44
Pin Configuration and Function Descriptions.....	11	Register Details .....	45
Typical Performance Characteristics .....	13	Evaluation Board .....	55
Mixer, High Performance Mode.....	13	Outline Dimensions .....	61
Mixer, High Efficiency Mode.....	22	Ordering Guide .....	61
Synthesizer.....	23		

## REVISION HISTORY

3/16—Revision 0: Initial Version

## SPECIFICATIONS

### RF SPECIFICATIONS

$T_A = 25^\circ\text{C}$ ,  $f_{RF} = 1900\text{ MHz}$ ,  $f_{LO} = 1697\text{ MHz}$ ,  $Z_O = 50\ \Omega$ , frequency of the reference ( $f_{REF}$ ) = 122.88 MHz,  $f_{REF}$  power = 4 dBm,  $f_{PFD} = 1.536\text{ MHz}$ , low-side LO injection, optimum RF balun (RFB) and low-pass filter (LPF) settings, unless otherwise noted.

**Table 1. High Performance Mode**

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
<b>RF INTERFACE</b>					
Return Loss	Tunable to >20 dB broadband via serial port		17.9		dB
Input Impedance			50		$\Omega$
RF Frequency Range ( $f_{RF}$ )		700		3000	MHz
<b>IF OUTPUT INTERFACE</b>					
Output Impedance	Differential impedance, $f = 200\text{ MHz}$		300  1.5		$\Omega$   pF
IF Frequency Range ( $f_{IF}$ )		40		500	MHz
DC Bias Voltage <sup>1</sup>	Externally generated		IFOUTx $\pm$		V
<b>EXTERNAL LO INPUT</b>					
External LO Power Input	External VCO input supports divide by 1, 2, 4, 8, 16, and 32 Low-side or high-side LO, internally or externally generated	-5	0	+5	dBm
Return Loss			-11		dB
Input Impedance			50		$\Omega$
External VCO Input Frequency		250		5700	MHz
LO Frequency Range		250		2850	MHz
<b>DYNAMIC PERFORMANCE</b>					
Power Conversion Gain	4:1 IF port transformer and printed circuit board (PCB) loss removed		9.0		dB
Voltage Conversion Gain	$Z_{SOURCE} = 50\ \Omega$ , differential $Z_{LOAD} = 200\ \Omega$		15.0		dB
SSB Noise Figure			11.3		dB
IF Output Phase Noise Under Blocking	10 dBm blocker present 10 MHz above desired the RF input, $f_{RF} = 1900\text{ MHz}$ , $f_{BLOCK} = 1910\text{ MHz}$ , $f_{LO} = 1697\text{ MHz}$ , $f_{IF} = 203\text{ MHz}$ , $f_{BLOCKER} = 213\text{ MHz}$		-153		dBc/Hz
Input Third-Order Intercept (IIP3)	$f_{RF1} = 1900\text{ MHz}$ , $f_{RF2} = 1901\text{ MHz}$ , $f_{LO} = 1697\text{ MHz}$ , each RF tone at -10 dBm		30		dBm
Input Second-Order Intercept (IIP2)	$f_{RF1} = 1900\text{ MHz}$ , $f_{RF2} = 1950\text{ MHz}$ , $f_{LO} = 1697\text{ MHz}$ , each RF tone at -10 dBm		60		dBm
Input 1 dB Compression Point (P1dB)			10.6		dBm
LO to IF Output Leakage	Unfiltered IF output		-35		dBm
LO to RF Input Leakage			-45		dBm
RF to IF Output Isolation			-22		dB
IF/2 Spurious	-10 dBm input power		-72		dBc
IF/3 Spurious	-10 dBm input power		-69		dBc
<b>POWER INTERFACE</b>					
VCC1, VCC2, VCC7, VCC12		3.55	3.7	3.85	V
Supply Voltage			260		
Quiescent Current					
VCC3, VCC4, VCC5, VCC6, VCC8, VCC9, VCC10, VCC11, IFOUT1+, IFOUT1-, IFOUT2+, IFOUT2-		3.55	5	5.25	V
Supply Voltage			214		
Quiescent Current					
<b>LO OUTPUT (LOOUT+, LOOUT-)</b>					
Frequency Range ( $f_{LO}$ )	Adjustable via SPI in four steps, in 50 $\Omega$ balanced load	200		2700	MHz
Output Level		-5		+7	dBm
Output Impedance		Balanced		50	$\Omega$

<sup>1</sup> Supply voltage must be applied from the external circuit through choke inductors.

$T_A = 25^\circ\text{C}$ ,  $f_{RF} = 1900\text{ MHz}$ ,  $f_{LO} = 1697\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $f_{REF} = 122.88\text{ MHz}$ ,  $f_{REF}\text{ power} = 4\text{ dBm}$ ,  $f_{PFD} = 1.536\text{ MHz}$ , low-side LO injection, optimum RFB and LPF settings, unless otherwise noted.

Table 2. High Efficiency Mode

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
<b>DYNAMIC PERFORMANCE</b>					
Power Conversion Gain	4:1 IF port transformer and PCB loss removed		8.7		dB
Voltage Conversion Gain	$Z_{SOURCE} = 50\ \Omega$ , differential $Z_{LOAD} = 200\ \Omega$		14.7		dB
SSB Noise Figure			10.7		dB
IIP3	$f_{RF1} = 1900\text{ MHz}$ , $f_{RF2} = 1901\text{ MHz}$ , $f_{LO} = 1697\text{ MHz}$ , each RF tone at $-10\text{ dBm}$		20.5		dBm
IIP2	$f_{RF1} = 1900\text{ MHz}$ , $f_{RF2} = 1950\text{ MHz}$ , $f_{LO} = 1697\text{ MHz}$ , each RF tone at $-10\text{ dBm}$		53		dBm
Input P1dB			8.2		dBm
LO to IF Output Leakage	Unfiltered IF output		-45.0		dBm
LO to RF Input Leakage			-52.0		dBm
RF to IF Output Isolation			-22.8		dB
IF/2 Spurious	$-10\text{ dBm}$ input power		-58		dBc
IF/3 Spurious	$-10\text{ dBm}$ input power		-58		dBc
<b>POWER INTERFACE</b>					
VCC1, VCC2, VCC7, VCC12					
Supply Voltage		3.55	3.7	3.85	V
Quiescent Current			260		mA
VCC3, VCC4, VCC5, VCC6, VCC8, VCC9, VCC10, VCC11, IFOUT1+, IFOUT1-, IFOUT2+, IFOUT2-					
Supply Voltage		3.55	3.7	5.25	V
Quiescent Current			210		mA

## SYNTHESIZER/PLL SPECIFICATIONS

High performance mode,  $T_A = 25^\circ\text{C}$ , measured on LO output,  $f_{LO} = 1700\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $f_{REF} = 122.88\text{ MHz}$ ,  $f_{PFD} = 1.536\text{ MHz}$ ,  $f_{REF}\text{ power}$  ( $P_{REFIN}$ ) =  $4\text{ dBm}$ ,  $CSCALE = 8\text{ mA}$ ,  $bleed = 0\ \mu\text{A}$ ,  $ABLDLY = 0.9\text{ ns}$ , integer mode loop filter, unless otherwise noted.

Table 3. Integer Mode

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
<b>SYNTHESIZER SPECIFICATIONS</b>					
Frequency Range ( $f_{LO}$ )	Synthesizer specifications referenced to $1 \times LO$ Internally generated LO	200		2700	MHz
Figure of Merit (FOM) <sup>1</sup>	$P_{REFIN} = 6.5\text{ dBm}$		-223		dBc/Hz/Hz
Phase and Frequency Detector (PFD) Frequency ( $f_{PFD}$ )		0.8		70	MHz
Reference Spurs	$f_{PFD} = 1.536\text{ MHz}$ $1 \times f_{PFD}$ $4 \times f_{PFD}$ $>4 \times f_{PFD}$		-105		dBc
			-105		dBc
			-90		dBc
<b>CHARGE PUMP</b>					
Pump Current	Programmable to $250\ \mu\text{A}$ , $500\ \mu\text{A}$ , ..., $8\text{ mA}$		8	8.75	mA
Output Compliance Range		0.7		2.5	V
<b>REFERENCE CHARACTERISTICS</b>					
REFIN Input Frequency	REFIN, MUXOUT pins	12		320	MHz
REFIN Input Capacitance			4		pF
Reference Divider Value	Programmable to 0.5, 1, 2, 3, ..., 2047	0.5		2047	
MUXOUT Output Level	$V_{OL}$ (lock detect output selected)			0.25	V
	$V_{OH}$ (lock detect output selected)	2.7			V
MUXOUT Duty Cycle	Reference output selected		50		%

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
VCO_0					
Phase Noise, Locked	$f_{LO} = 2.55$ GHz				
	1 kHz offset		-87		dBc/Hz
	50 kHz offset		-94.9		dBc/Hz
	100 kHz offset		-103.3		dBc/Hz
	1 MHz offset		-132.9		dBc/Hz
	10 MHz offset		-154.1		dBc/Hz
	40 MHz offset		-155.2		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.87		°rms
VCO_1					
Phase Noise, Locked	$f_{LO} = 2.22$ GHz				
	1 kHz offset		-90		dBc/Hz
	50 kHz offset		-98.4		dBc/Hz
	100 kHz offset		-106.5		dBc/Hz
	1 MHz offset		-136.1		dBc/Hz
	10 MHz offset		-154.8		dBc/Hz
	40 MHz offset		-155.5		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.63		°rms
VCO_2					
Phase Noise, Locked	$f_{LO} = 1.9$ GHz				
	1 kHz offset		-90		dBc/Hz
	50 kHz offset		-98.1		dBc/Hz
	100 kHz offset		-109.8		dBc/Hz
	1 MHz offset		-137.1		dBc/Hz
	10 MHz offset		-155.7		dBc/Hz
	40 MHz offset		-156.2		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.61		°rms
VCO_3					
Phase Noise, Locked	$f_{LO} = 1.6$ GHz				
	1 kHz offset		-89		dBc/Hz
	50 kHz offset		-97.2		dBc/Hz
	100 kHz offset		-107		dBc/Hz
	1 MHz offset		-136.2		dBc/Hz
	10 MHz offset		-155.7		dBc/Hz
	40 MHz offset		-157.3		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.64		°rms
VCO_4					
Phase Noise, Locked	$f_{LO} = 1.57$ GHz				
	1 kHz offset		-90		dBc/Hz
	50 kHz offset		-109		dBc/Hz
	100 kHz offset		-119		dBc/Hz
	800 kHz offset		-144		dBc/Hz
	1 MHz offset		-145		dBc/Hz
	10 MHz offset		-156		dBc/Hz
	40 MHz offset		-156		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.26		°rms
VCO_5					
Phase Noise, Locked	$f_{LO} = 1.68$ GHz				
	1 kHz offset		-93		dBc/Hz
	50 kHz offset		-107		dBc/Hz
	100 kHz offset		-118		dBc/Hz
	800 kHz offset		-144		dBc/Hz
	1 MHz offset		-145		dBc/Hz

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
Integrated Phase Noise	10 MHz offset		-157		dBc/Hz
	40 MHz offset		-157.5		dBc/Hz
	1 kHz to 40 MHz integration bandwidth		0.27		°rms

<sup>1</sup> The FOM is computed as phase noise (dBc/Hz) - 10log<sub>10</sub>(f<sub>FPD</sub>) - 20log<sub>10</sub>(f<sub>LO</sub>/f<sub>FPD</sub>). The FOM was measured across the full LO range, with f<sub>REF</sub> = 122.88 MHz and f<sub>REF</sub> power = 6.5 dBm with a 1.536 MHz f<sub>FPD</sub>. The FOM was computed at 50 kHz offset.

High performance mode, T<sub>A</sub> = 25°C, measured on LO output, f<sub>LO</sub> = 1700 MHz, Z<sub>O</sub> = 50 Ω, f<sub>REF</sub> = 122.88 MHz, f<sub>FPD</sub> = 30.72 MHz, f<sub>REF</sub> power = 4 dBm, CSCALE = 250 μA, bleed = 93.75 μA, ABLDLY = 0 ns, fractional mode loop filter, unless otherwise noted.

**Table 4. Fractional Mode**

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
SYNTHESIZER SPECIFICATIONS	Synthesizer specifications referenced to 1 × LO				
FOM <sup>1</sup>	P <sub>REFIN</sub> = 6.5 dBm		219		dBc/Hz/Hz
REFERENCE CHARACTERISTICS	REFIN, MUXOUT pins				
VCO_0					
Phase Noise, Locked	f <sub>LO</sub> = 2.55 GHz				
	1 kHz offset		-92.5		dBc/Hz
	50 kHz offset		-97.4		dBc/Hz
	100 kHz offset		-109.7		dBc/Hz
	1 MHz offset		-137.6		dBc/Hz
	10 MHz offset		-153.6		dBc/Hz
	40 MHz offset		-155.5		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.36		°rms
VCO_1					
Phase Noise, Locked	f <sub>LO</sub> = 2.22 GHz				
	1 kHz offset		-93.6		dBc/Hz
	50 kHz offset		-101.8		dBc/Hz
	100 kHz offset		-112.5		dBc/Hz
	1 MHz offset		-140.5		dBc/Hz
	10 MHz offset		-154.3		dBc/Hz
	40 MHz offset		-155.3		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.32		°rms
VCO_2					
Phase Noise, Locked	f <sub>LO</sub> = 1.9 GHz				
	1 kHz offset		-94.2		dBc/Hz
	50 kHz offset		-101.7		dBc/Hz
	100 kHz offset		-112.4		dBc/Hz
	1 MHz offset		-141.3		dBc/Hz
	10 MHz offset		-155.8		dBc/Hz
	40 MHz offset		-156.8		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.32		°rms
VCO_3					
Phase Noise, Locked	f <sub>LO</sub> = 1.6 GHz				
	1 kHz offset		-93.1		dBc/Hz
	50 kHz offset		-99.8		dBc/Hz
	100 kHz offset		-110.9		dBc/Hz
	1 MHz offset		-140.2		dBc/Hz
	10 MHz offset		-155.7		dBc/Hz
	40 MHz offset		-157.2		dBc/Hz
Integrated Phase Noise	1 kHz to 40 MHz integration bandwidth		0.33		°rms

<sup>1</sup> The FOM is computed as phase noise (dBc/Hz) - 10log<sub>10</sub>(f<sub>FPD</sub>) - 20log<sub>10</sub>(f<sub>LO</sub>/f<sub>FPD</sub>). The FOM was measured across the full LO range, with f<sub>REF</sub> = 122.88 MHz and f<sub>REF</sub> power = 6.5 dBm with a 30.72 MHz f<sub>FPD</sub>. The FOM was computed at 45 kHz offset.



**VCO SPECIFICATIONS, OPEN-LOOP**High performance mode,  $T_A = 25^\circ\text{C}$ , measured on LO output, unless otherwise noted.

Table 5.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
VCO_0 PHASE NOISE	$f_{LO} = 2.55\text{ GHz}$				
	1 kHz offset		-50		dBc/Hz
	50 kHz offset		-104.4		dBc/Hz
	100 kHz offset		-112.6		dBc/Hz
	1 MHz offset		-137.7		dBc/Hz
	10 MHz offset		-154		dBc/Hz
VCO_1 PHASE NOISE	$f_{LO} = 2.15\text{ GHz}$				
	1 kHz offset		-54		dBc/Hz
	50 kHz offset		-106.1		dBc/Hz
	100 kHz offset		-115		dBc/Hz
	1 MHz offset		-138.9		dBc/Hz
	10 MHz offset		-155.8		dBc/Hz
VCO_2 PHASE NOISE	$f_{LO} = 1.9\text{ GHz}$				
	1 kHz offset		-53.6		dBc/Hz
	50 kHz offset		-106.6		dBc/Hz
	100 kHz offset		-114.6		dBc/Hz
	1 MHz offset		-140.8		dBc/Hz
	10 MHz offset		-155.4		dBc/Hz
VCO_3 PHASE NOISE	$f_{LO} = 1.6\text{ GHz}$				
	1 kHz offset		-48.5		dBc/Hz
	50 kHz offset		-106		dBc/Hz
	100 kHz offset		-115.3		dBc/Hz
	800 kHz offset		-139.2		dBc/Hz
	1 MHz offset		-140.2		dBc/Hz
VCO_4 PHASE NOISE	$f_{VCO} = 3.14\text{ GHz}$				
	1 kHz offset		-53.8		dBc/Hz
	50 kHz offset		-110.3		dBc/Hz
	100 kHz offset		-118		dBc/Hz
	800 kHz offset		-139.5		dBc/Hz
	1 MHz offset		-140.6		dBc/Hz
VCO_5 PHASE NOISE	$f_{VCO} = 3.36\text{ GHz}$				
	1 kHz offset		-54		dBc/Hz
	50 kHz offset		-108.3		dBc/Hz
	100 kHz offset		-116.3		dBc/Hz
	800 kHz offset		-138.5		dBc/Hz
	1 MHz offset		-140		dBc/Hz
VCO_5 PHASE NOISE	$f_{VCO} = 3.36\text{ GHz}$				
	10 MHz offset		-156.3		dBc/Hz
	40 MHz offset		-157.8		dBc/Hz

**LOGIC INPUT/OUTPUT AND POWER SPECIFICATIONS**

$T_A = 25^\circ\text{C}$ ,  $f_{RF} = 1900\text{ MHz}$ ,  $f_{LO} = 1697\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $f_{REF} = 122.88\text{ MHz}$ ,  $f_{REF}\text{ power} = 4\text{ dBm}$ ,  $f_{PPD} = 1.536\text{ MHz}$ , low-side LO injection, optimum RFB and LPF settings, unless otherwise noted.

**Table 6.**

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
<b>LOGIC INPUT/OUTPUTS</b>					
SCLK, SDIO, $\overline{\text{CS}}$					
Input Voltage					
High, $V_{IH}$		1.4		3.3	V
Low, $V_{IL}$		0		0.7	V
Output Voltage					
High, $V_{OH}$	$I_{OH} = -100\ \mu\text{A}$	2.3			V
Low, $V_{OL}$	$I_{OL} = 100\ \mu\text{A}$			0.2	V
Input Current, $I_{INH}/I_{INL}$			0.1		$\mu\text{A}$
<b>POWER SUPPLIES</b>					
High Performance Mode					
Voltage Range					
VCC1, VCC2, VCC7, VCC12					
		3.55	3.7	5.25	V
VCC3, VCC4, VCC5, VCC6, VCC8, VCC9, VCC10, VCC11, IFOUT1+, IFOUT1-, IFOUT2+, IFOUT2-					
		4.75	5	5.25	V
Power Dissipation					
Internal LO mode (internal PLL)					
External LO output enabled					
			2.7		W
External LO output disabled					
			2.5		W
High Efficiency Mode					
Voltage Range					
VCC1, VCC2, VCC3, VCC4, VCC5, VCC6, VCC7, VCC8, VCC9, VCC10, VCC11, VCC12, IFOUT1+, IFOUT1-, IFOUT2+, IFOUT2-					
		3.55	3.7	3.85	V
Power Dissipation					
Internal LO mode (internal PLL)					
External LO output enabled					
			2.0		W
External LO output disabled					
			1.8		W

DIGITAL LOGIC SPECIFICATIONS

Table 7.

Symbol	Description	Min	Typ	Max	Unit
$t_{CLK}$	Serial clock period	38			ns
$t_{DS}$	Setup time between data and rising edge of SCLK	8			ns
$t_{DH}$	Hold time between data and rising edge of SCLK	8			ns
$t_S$	Setup time between falling edge of $\overline{CS}$ and SCLK	10			ns
$t_H$	Hold time between rising edge of $\overline{CS}$ and SCLK	10			ns
$t_{HIGH}$	Minimum period for SCLK to be in a logic high state	10			ns
$t_{LOW}$	Minimum period for SCLK to be in a logic low state	10			ns
$t_{ACCESS}$	Maximum delay between falling edge of SCLK and output data Valid for a read operation			231	ns
$t_z$	Maximum delay between $\overline{CS}$ deactivation and SDIO bus return to high impedance			5	ns

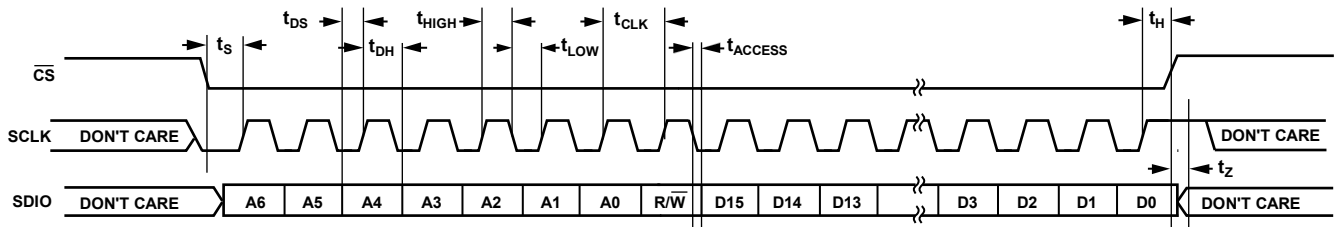


Figure 2. Setup and Hold Timing Measurements

14115-002

## ABSOLUTE MAXIMUM RATINGS

Table 8.

Parameter	Rating
Supply Voltage (VCC1, VCC2, VCC3, VCC4, VCC5, VCC6, VCC7, VCC8, VCC9, VCC10, VCC11, VCC12, IFOUT1+, IFOUT1-, IFOUT2+, IFOUT2-)	-0.5 V to +5.5 V
Digital Input/Output (SCLK, SDIO, $\overline{CS}$ )	-0.3 V to +3.6 V
RFINx	20 dBm
EXTVCOIN+, EXTVCOIN-	13 dBm
Maximum Junction Temperature	150°C
Operating Temperature Range	-40°C to +85°C
Storage Temperature Range	-65°C to +150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## THERMAL RESISTANCE

$\theta_{JC}$  is specified for the worst case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 9. Thermal Resistance

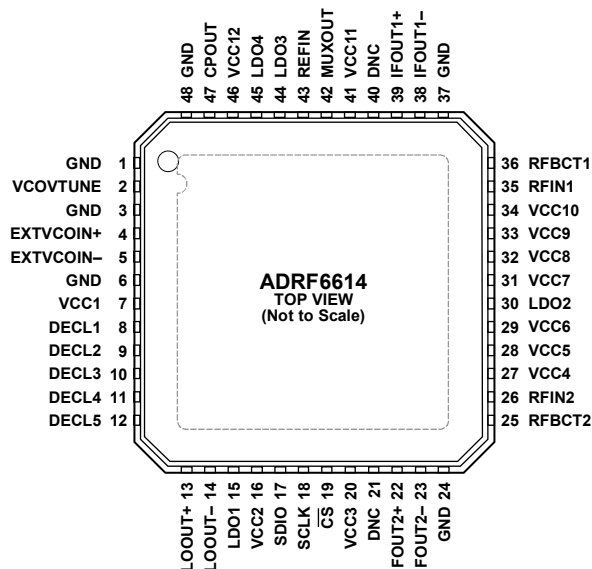
Package Type	$\theta_{JC}$	Unit
48-Lead LFCSP	1.62	°C/W

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



## NOTES

1. DNC = DO NOT CONNECT.
2. THE EXPOSED PAD MUST BE CONNECTED TO A GROUND PLANE WITH LOW THERMAL IMPEDANCE.

14115-003

Figure 3. Pin Configuration

Table 10. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	GND	Common Ground Connection for External Loop Filter.
2	VCOVTUNE	Control Voltage for Internal VCO.
3, 6	GND	Common Ground for External VCO.
4, 5	EXTVCOIN+, EXTVCOIN-	Inputs from External VCO to Internal Divider.
7	VCC1	3.7 V VCO Supply.
8, 9	DECL1, DECL2	LDO Output Decouplers for VCO.
10, 11	DECL3, DECL4	External Decouplers for VCO Buffer.
12	DECL5	External Decoupler for VCO Circuitry.
13, 14	LOOUT+, LOOUT-	Differential Outputs of Internally Generated LO.
15	LDO1	External Decoupling for Internal 2.5 V SPI Port LDO.
16	VCC2	3.7 V Supply for Programmable SPI Port.
17	SDIO	Serial Data Input/Output for Programmable SPI Port.
18	SCLK	Clock for Programmable SPI Port.
19	$\overline{CS}$	SPI Chip Select, Active Low.
20, 41	VCC3, VCC11	5 V Biases for Channel 1 and Channel 2 IF.
21, 40	DNC	Do Not Connect. Do not connect these pins externally.
22, 23	IFOUT2+, IFOUT2-	Channel 2 Differential IF Outputs.
24, 37	GND	Ground Connections for Channel 1 and Channel 2 IF Stage.
25	RFBCT2	Balun Center Tap Connection for Channel 2 RF Input.
26	RFIN2	Channel 2 RF Input.
27, 28, 29	VCC4, VCC5, VCC6	5 V Supplies for Mixer LO Amplifiers.
30	LDO2	External Decoupling for Internal 3.3 V PLL/Divider LDO.
31	VCC7	3.7 V Supply for Mixer LO Divider Chain.
32, 33, 34	VCC8, VCC9, VCC10	5 V Supplies for Mixer LO Amplifiers.
35	RFIN1	Channel 1 RF Input.
36	RFBCT1	Balun Center Tap Connection for Channel 1 RF Input.
38, 39	IFOUT1-, IFOUT1+	Channel 1 Differential IF Outputs.
42	MUXOUT	Internal Multiplexer Output.

Pin No.	Mnemonic	Description
43	REFIN	Reference Input for Internal PLL (Single-Ended, CMOS).
44	LDO3	External Decoupling for Internal 2.5 V PLL LDO.
45	LDO4	External Decoupling for Internal 3.3 V PLL LDO.
46	VCC12	3.7 V Supply for Internal PLL.
47	CPOUT	Charge Pump Output.
48	GND	Common Ground for External Charge Pump.
	EPAD	Exposed Pad. The exposed pad must be connected to a ground plane with low thermal impedance.

# TYPICAL PERFORMANCE CHARACTERISTICS

## MIXER, HIGH PERFORMANCE MODE

$T_A = 25^\circ\text{C}$ ,  $f_{RF} = 1900\text{ MHz}$ ,  $f_{LO} = 1697\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $f_{REF} = 122.88\text{ MHz}$ ,  $f_{REF}\text{ power} = 4\text{ dBm}$ , low-side LO injection, optimum RFB and LPF settings, unless otherwise noted. For integer mode:  $f_{FPD} = 1.536\text{ MHz}$ ,  $CSCALE = 8\text{ mA}$ ,  $bleed = 0\ \mu\text{A}$ ,  $ABLDLY = 0.9\text{ ns}$ . For fractional mode:  $f_{FPD} = 30.72\text{ MHz}$ ,  $CSCALE = 250\ \mu\text{A}$ ,  $bleed = 93.75\ \mu\text{A}$ ,  $ABLDLY = 0.0\text{ ns}$ .

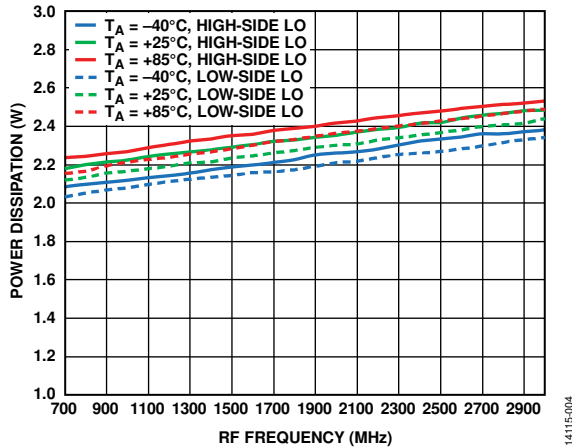


Figure 4. Power Dissipation vs. RF Frequency over Three Temperatures

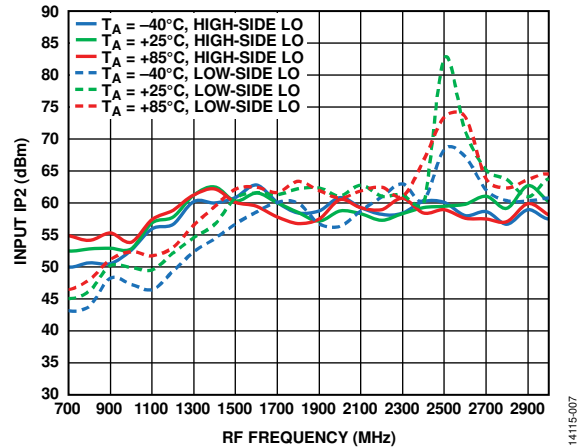


Figure 7. Input IP2 vs. RF Frequency over Three Temperatures

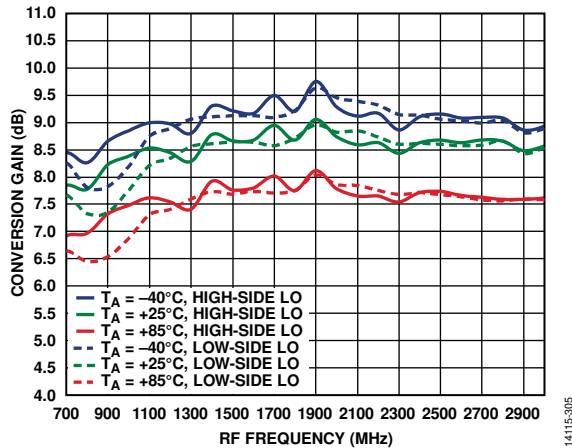


Figure 5. Power Conversion Gain vs. RF Frequency over Three Temperatures, IF Balun and Board Loss Removed

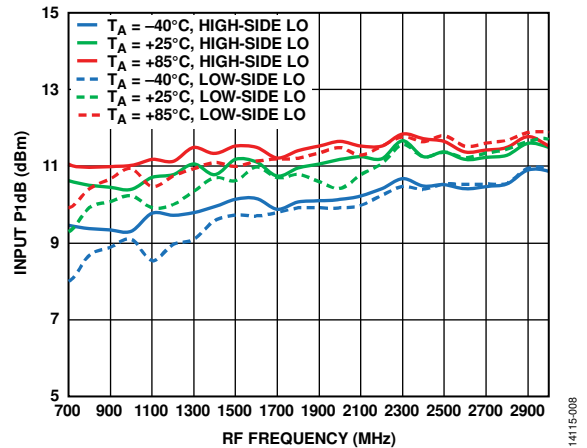


Figure 8. Input P1dB vs. RF Frequency over Three Temperatures

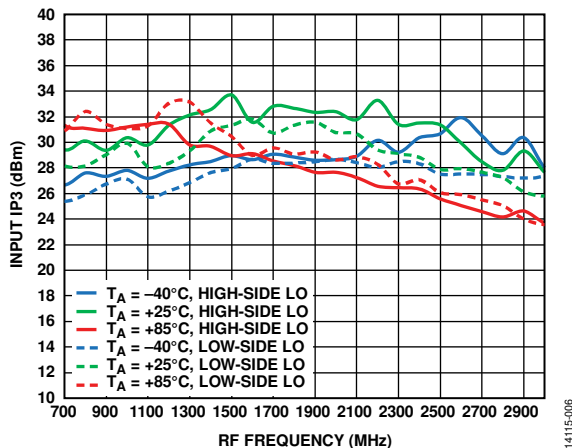


Figure 6. Input IP3 vs. RF Frequency over Three Temperatures

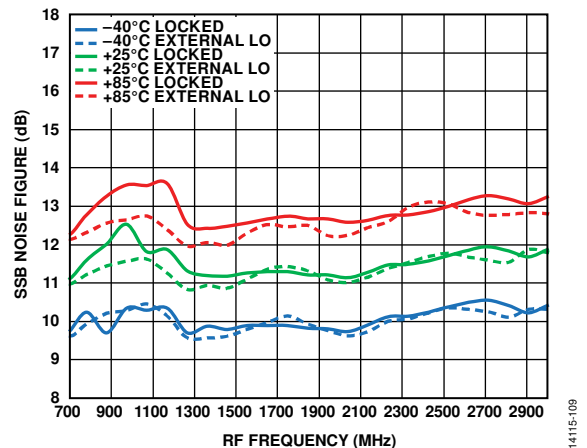


Figure 9. SSB Noise Figure vs. RF Frequency over Three Temperatures

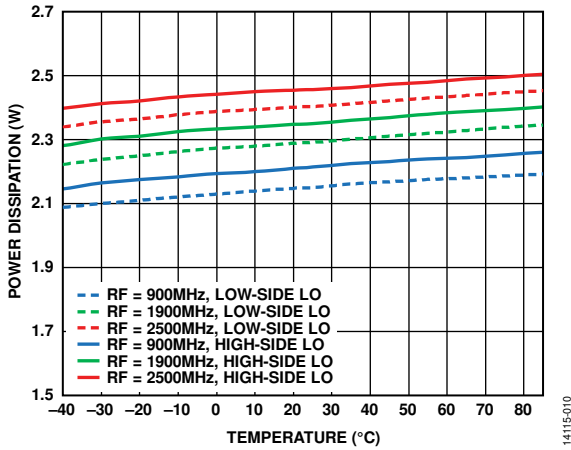


Figure 10. Power Dissipation vs. Temperature for Three RF Frequencies

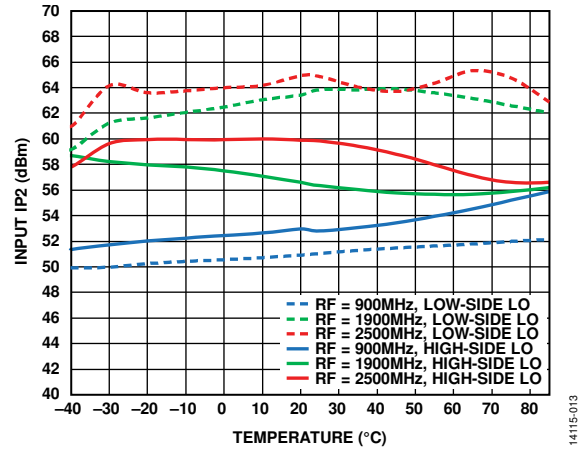


Figure 13. Input IP2 vs. Temperature for Three RF Frequencies

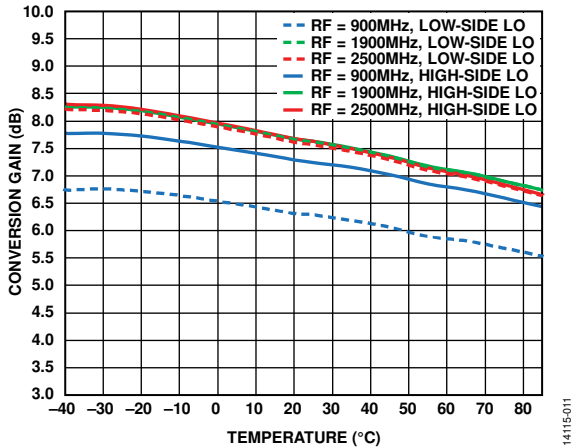


Figure 11. Power Conversion Gain vs. Temperature for Three RF Frequencies

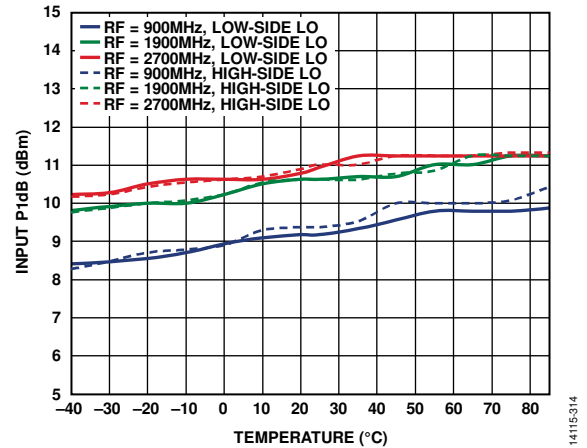


Figure 14. Input P1dB vs. Temperature for Three RF Frequencies

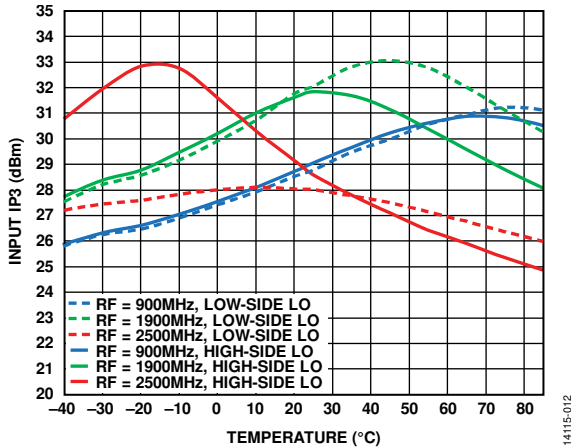


Figure 12. Input IP3 vs. Temperature for Three RF Frequencies

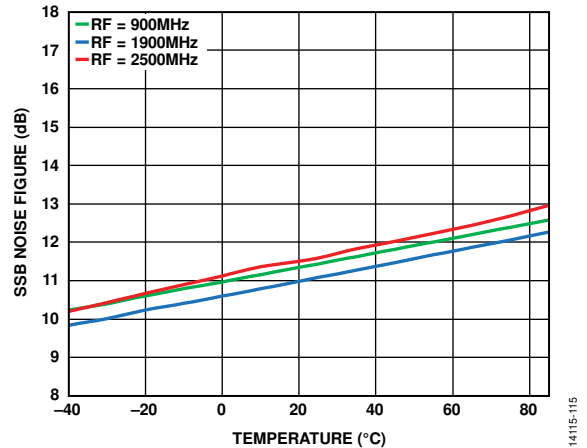


Figure 15. SSB Noise Figure vs. Temperature for Three RF Frequencies



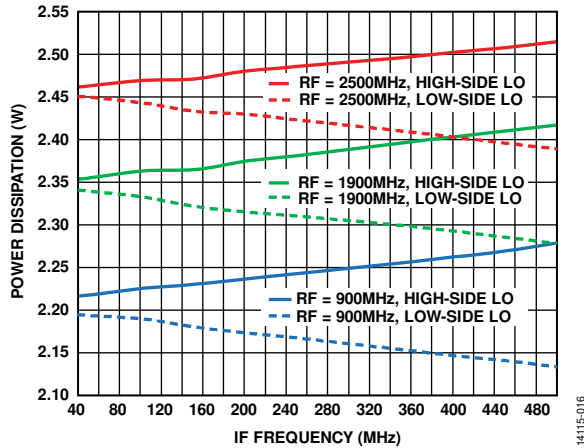


Figure 16. Power Dissipation vs. IF Frequency for Three RF Frequencies

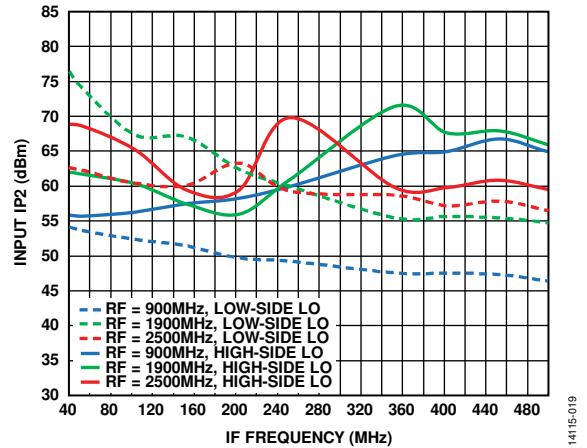


Figure 19. Input IP2 vs. IF Frequency for Three RF Frequencies

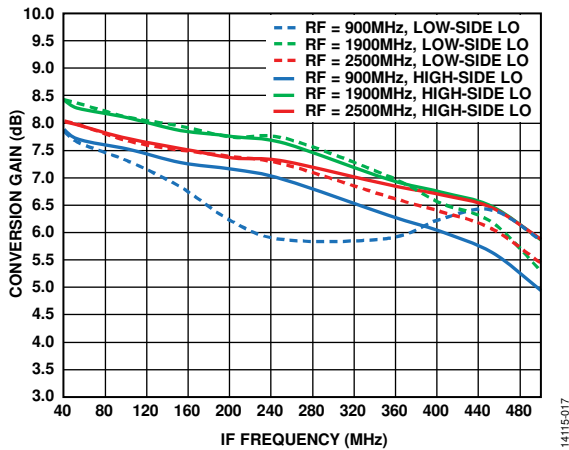


Figure 17. Power Conversion Gain vs. IF Frequency for Three RF Frequencies

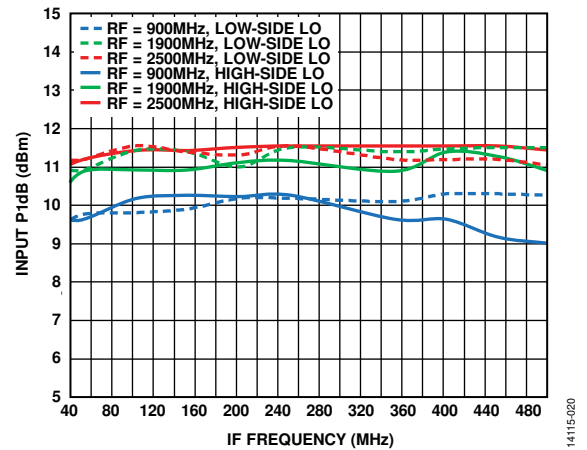


Figure 20. Input P1dB vs. IF Frequency for Three RF Frequencies

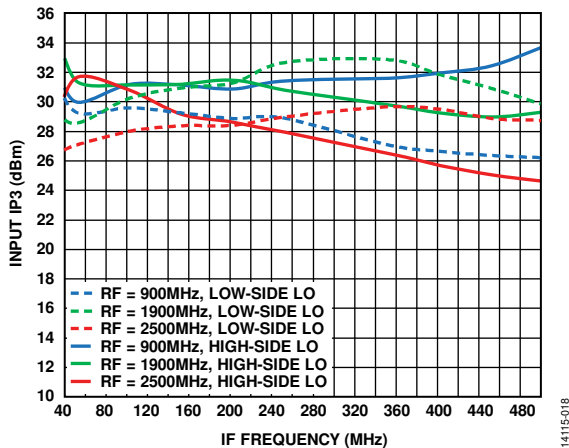


Figure 18. Input IP3 vs. IF Frequency for Three RF Frequencies

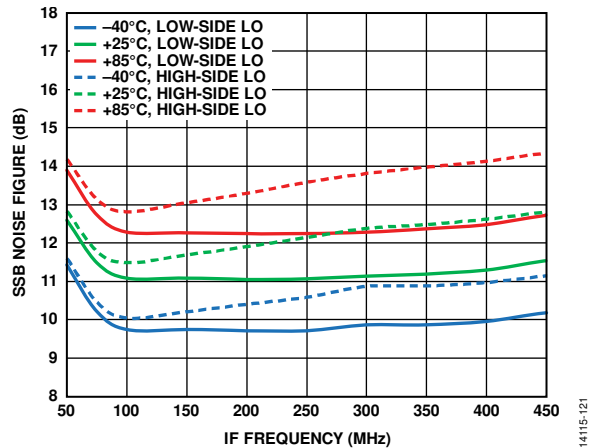


Figure 21. SSB Noise Figure vs. IF Frequency for Three Temperatures

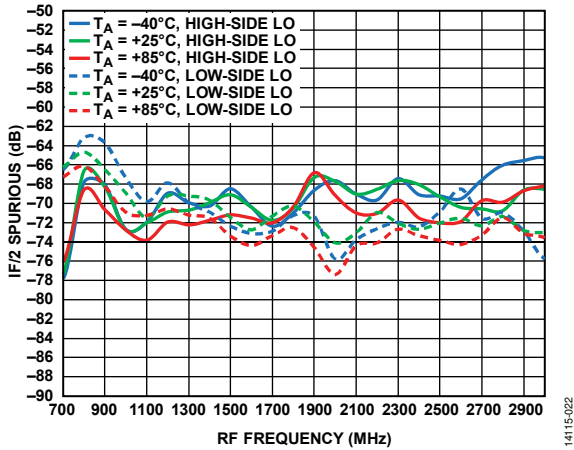


Figure 22. IF/2 Spurious vs. RF Frequency over Three Temperatures

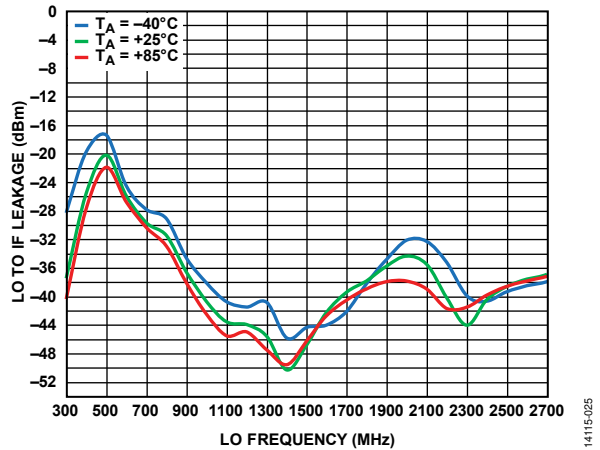


Figure 25. LO to IF Leakage vs. LO Frequency over Three Temperatures

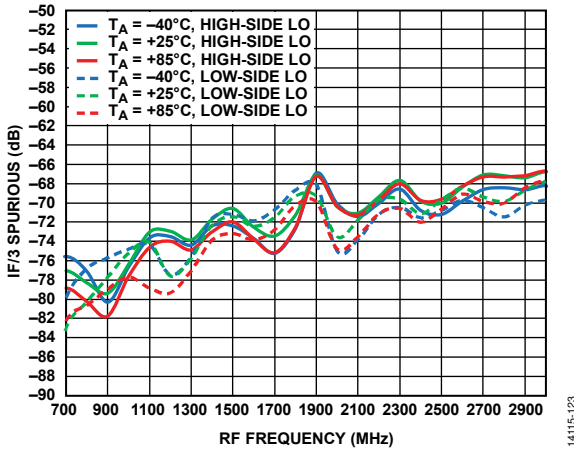


Figure 23. IF/3 Spurious vs. RF Frequency over Three Temperatures

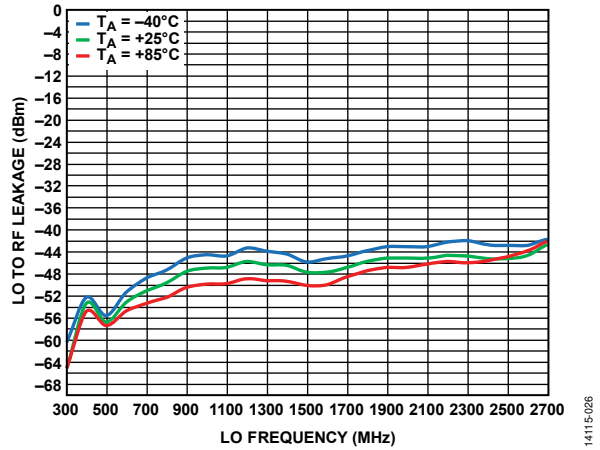


Figure 26. LO to RF Leakage vs. LO Frequency over Three Temperatures

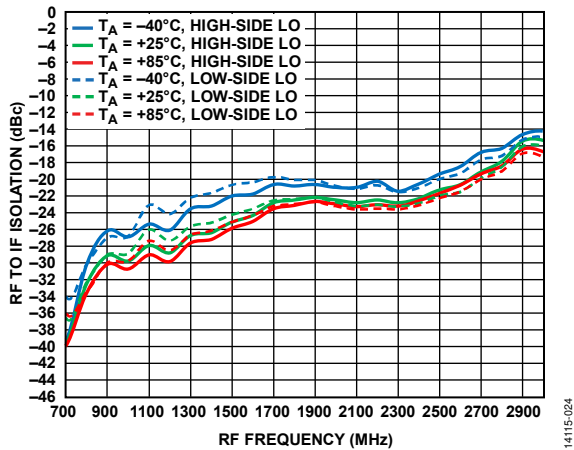


Figure 24. RF to IF Isolation vs. RF Frequency over Three Temperatures

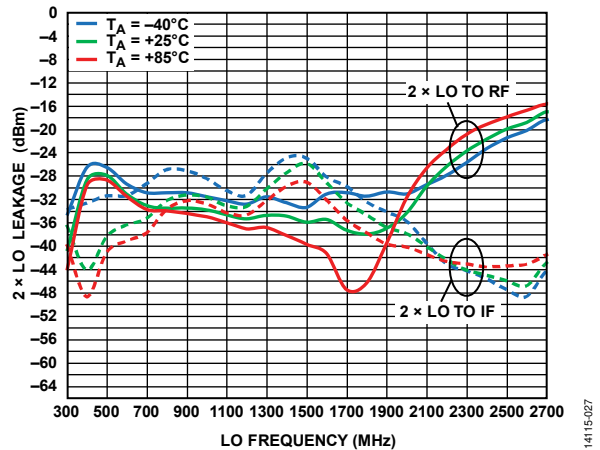


Figure 27. 2 x LO Leakage vs. LO Frequency over Three Temperatures (2 x LO to RF and 2 x LO to IF)

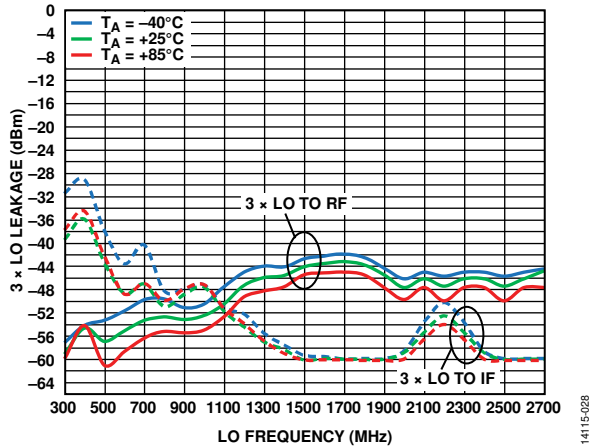


Figure 28. 3 x LO Leakage vs. LO Frequency over Three Temperatures (3 x LO to RF and 3 x LO to IF)

14115-028

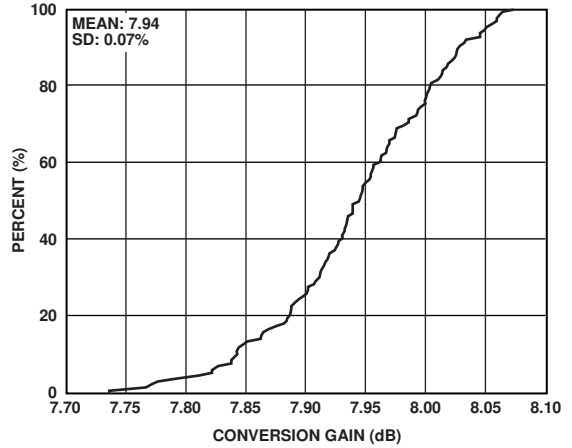


Figure 31. Conversion Gain Distribution

14115-131

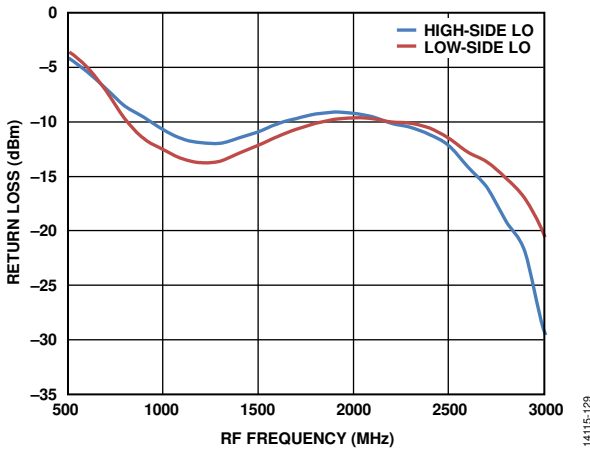


Figure 29. RF Port Return Loss, Fixed IF LO Return Loss

14115-129

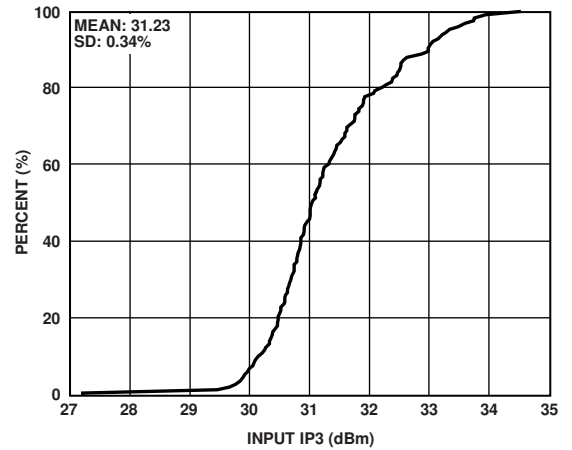


Figure 32. Input IP3 Distribution

14115-132

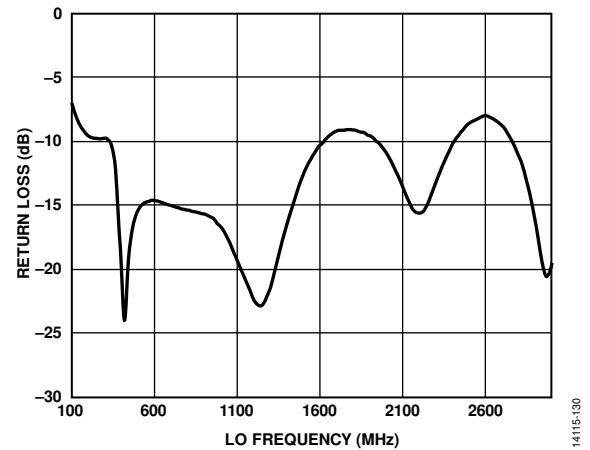


Figure 30. LO Return Loss

14115-130

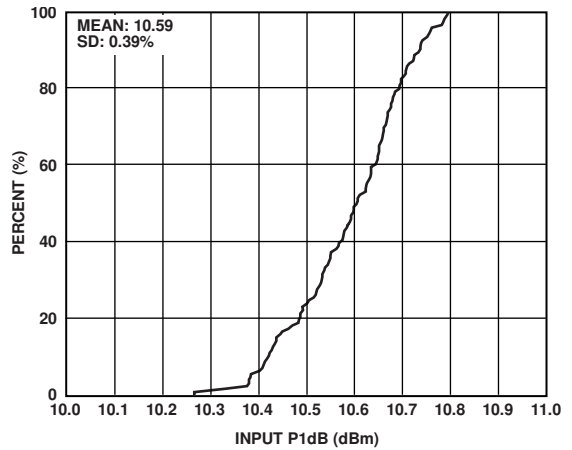


Figure 33. Input P1dB Distribution

14115-133

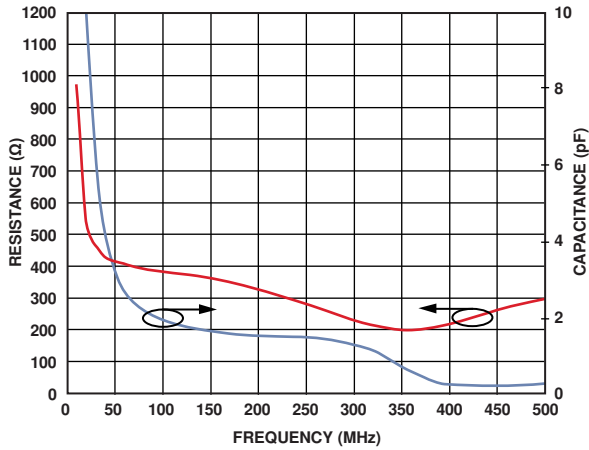


Figure 34. IF Output Impedance (R Parallel C Equivalent)

14115-034

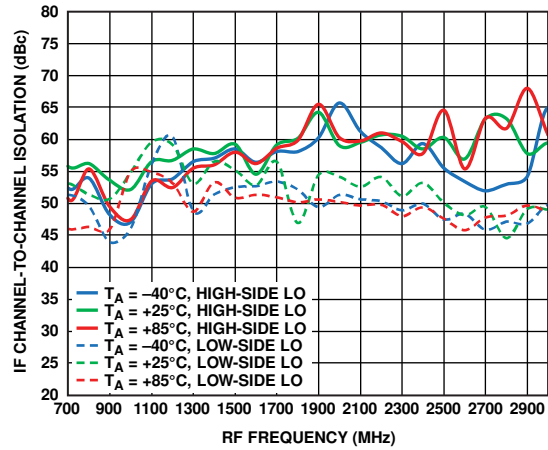


Figure 37. IF Channel to Channel Isolation vs. RF Frequency over Three Temperatures

14115-037

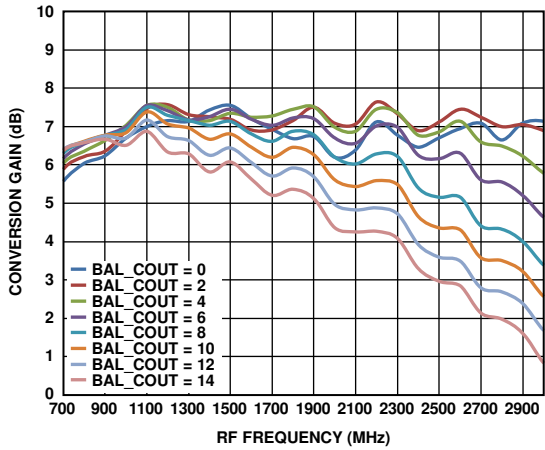


Figure 35. Conversion Gain vs. RF Frequency for All RFB Settings, VGS Bit and LPF Use Optimum Settings

14115-035

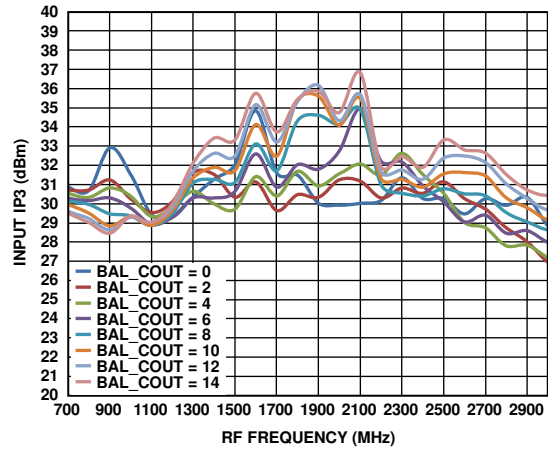


Figure 38. Input IP3 vs. RF Frequency for All RFB Settings, VGS Bit and LPF Use Optimum Settings

14115-038

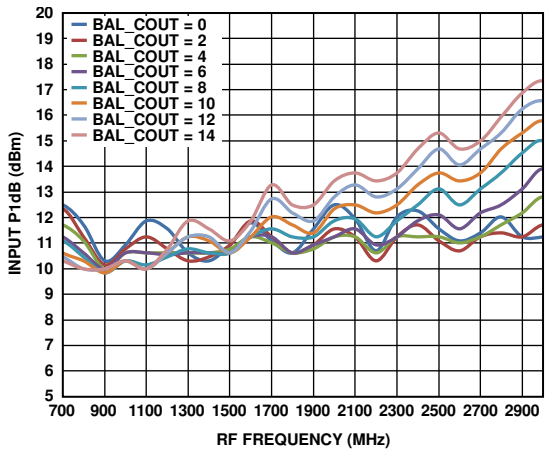


Figure 36. Input P1dB vs. RF Frequency for All RFB Settings, VGS Bit and LPF Use Optimum Settings

14115-036

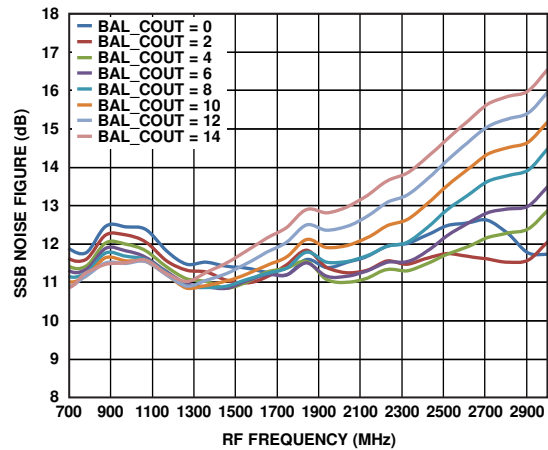


Figure 39. SSB Noise Figure vs. RF Frequency for All RFB Settings, VGS Bit and LPF Use Optimum Settings

14115-039

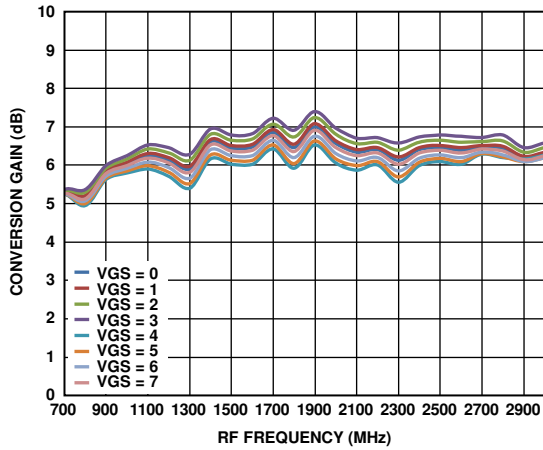


Figure 40. Conversion Gain vs. RF Frequency for All VGS Bit Settings, RFB and LPF Use Optimum Settings

14115-040

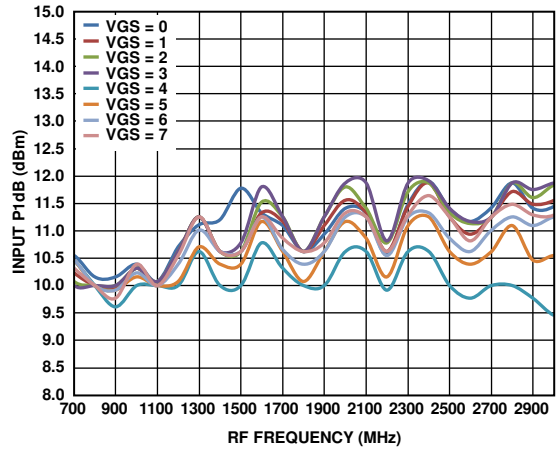


Figure 43. Input P1dB vs. RF Frequency for All VGS Bit Settings, RFB and LPF Use Optimum Settings

14115-043

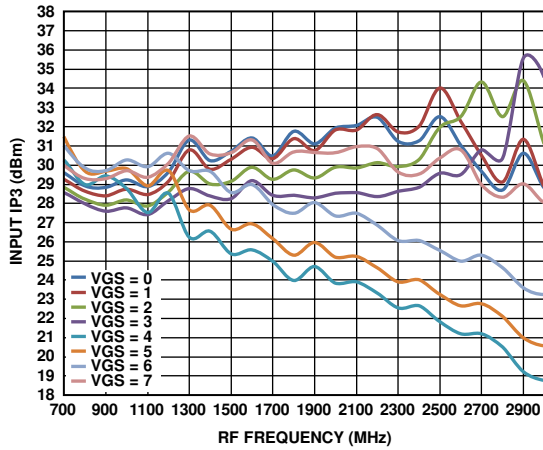


Figure 41. Input IP3 vs. RF Frequency for All VGS Bit Settings, RFB and LPF Use Optimum Settings

14115-141

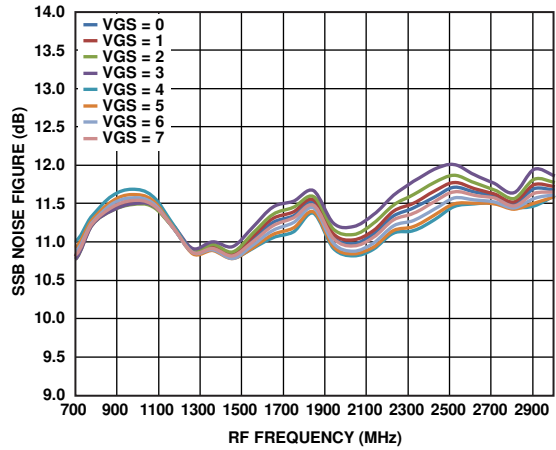


Figure 44. SSB Noise Figure vs. RF Frequency for All VGS Bit Settings, RFB and LPF Use Optimum Settings

14115-144

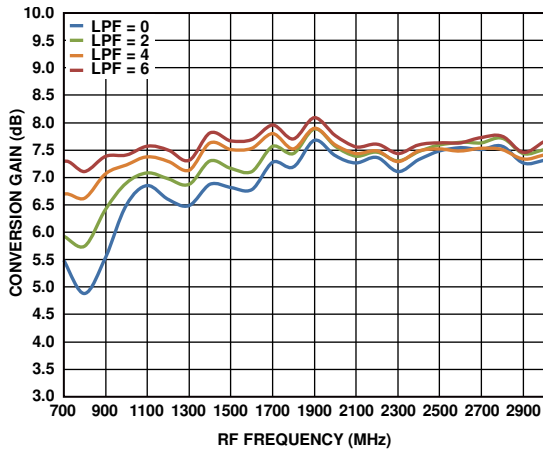


Figure 42. Conversion Gain vs. RF Frequency for All LPF Settings, RFB and VGS Bit Use Optimum Settings

14115-146

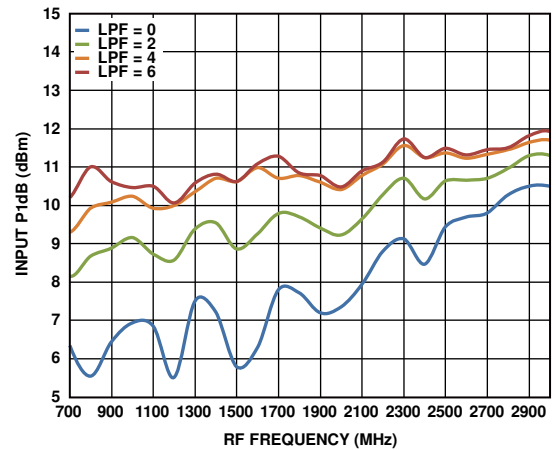


Figure 45. Input P1dB vs. RF Frequency for All LPF Settings, RFB and VGS Bit Use Optimum Settings

14115-148

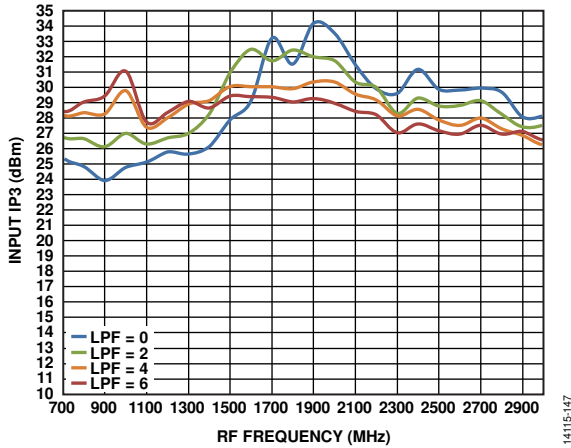


Figure 46. Input IP3 vs. RF Frequency for All LPF Settings, RFB and VGS Bit Use Optimum Settings

14115-147

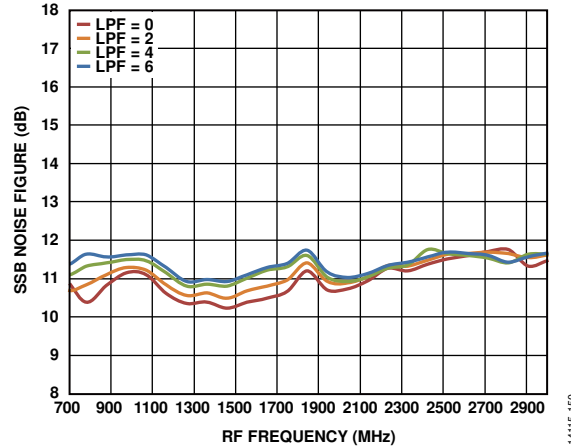


Figure 49. SSB Noise Figure vs. RF Frequency for All LPF Settings, RFB and VGS Bit Use Optimum Settings

14115-150

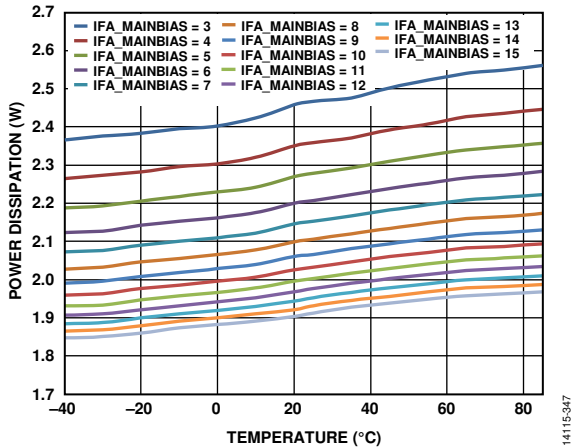


Figure 47. Power Dissipation vs. Temperature for Various IFA\_MAINBIAS Settings

14115-347

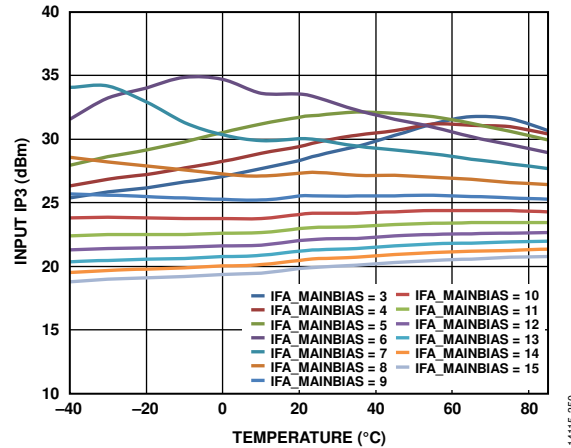


Figure 50. Input IP3 vs. Temperature for Various IFA\_MAINBIAS Settings

14115-350

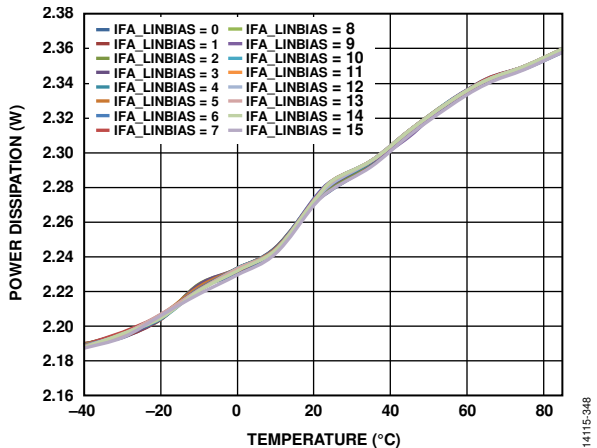


Figure 48. Power Dissipation vs. Temperature for Various IFA\_LINBIAS Settings

14115-348

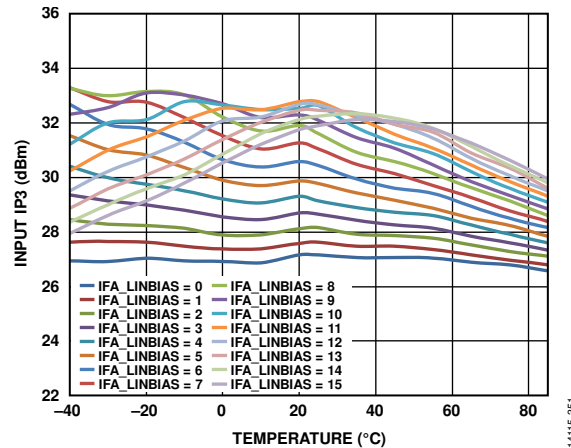


Figure 51. Input IP3 vs. Temperature for Various IFA\_LINBIAS Settings

14115-351

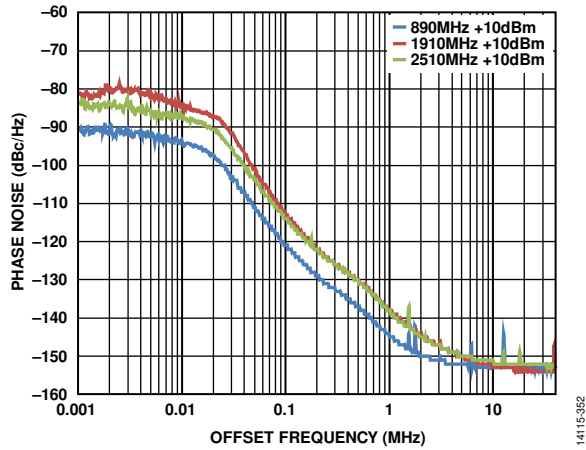


Figure 52. Phase Noise at IF Output vs. Offset Frequency with 10 dBm Blocker in Integer Mode

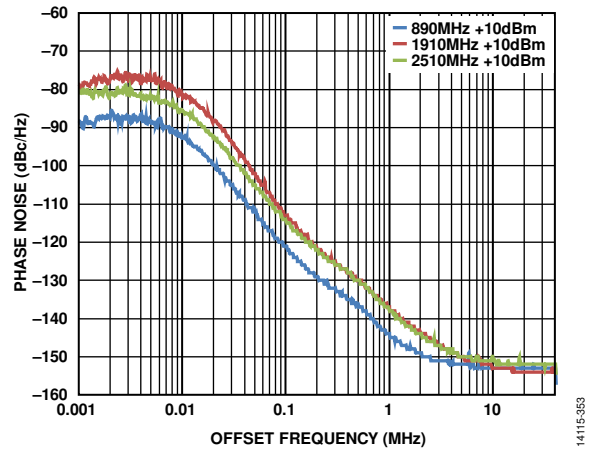


Figure 53. Phase Noise at IF Output vs. Offset Frequency with 10 dBm Blocker in Fractional Mode

MIXER, HIGH EFFICIENCY MODE

$T_A = 25^\circ\text{C}$ ,  $f_{RF} = 1900\text{ MHz}$ ,  $f_{LO} = 1697\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $f_{REF} = 122.88\text{ MHz}$ ,  $f_{REF}\text{ power} = 4\text{ dBm}$ ,  $f_{PPD} = 1.536\text{ MHz}$ , low-side LO injection, optimum RFB and LPF settings, unless otherwise noted.

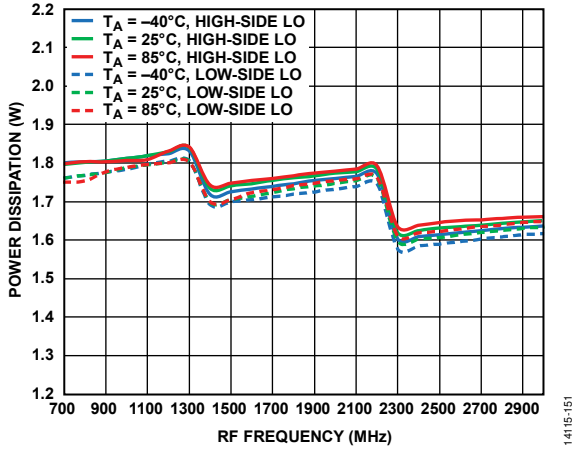


Figure 54. Power Dissipation vs. RF Frequency over Three Temperatures

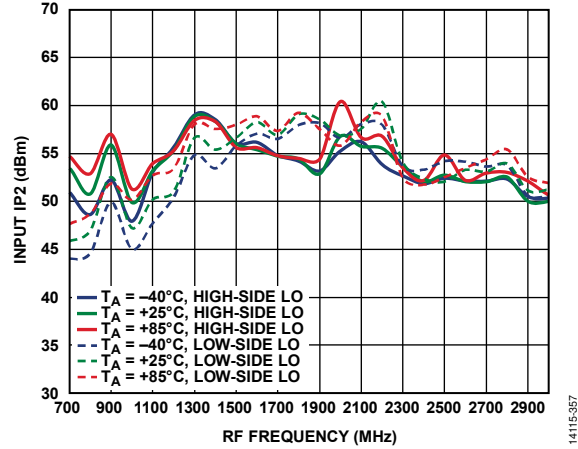


Figure 57. Input IP2 vs. RF Frequency over Three Temperatures

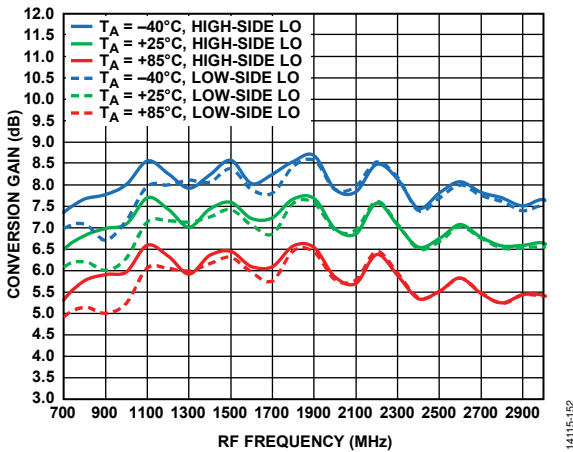


Figure 55. Conversion Gain vs. RF Frequency over Three Temperatures

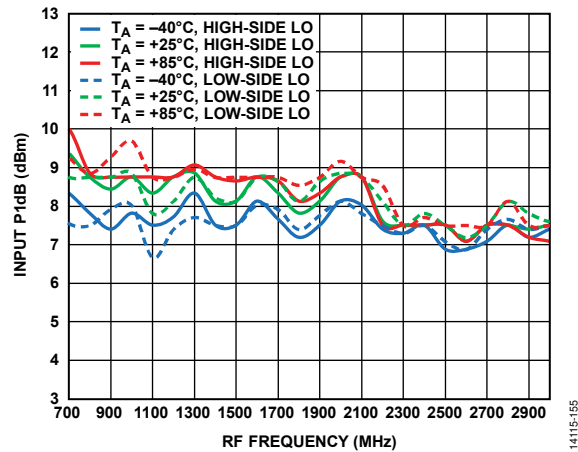


Figure 58. Input P1dB vs. RF Frequency over Three Temperatures

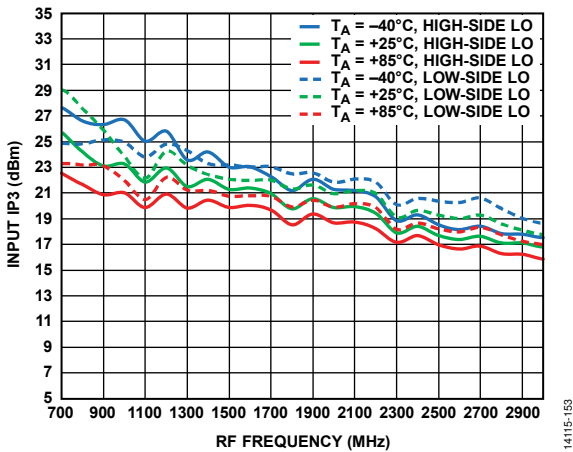


Figure 56. Input IP3 vs. RF Frequency over Three Temperatures

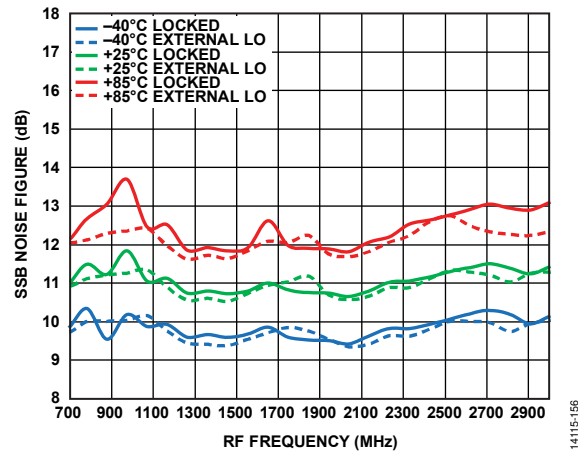


Figure 59. SSB Noise Figure vs. RF Frequency over Three Temperatures



**SYNTHESIZER**

$V_S$  = high performance mode,  $T_A = 25^\circ\text{C}$ , measured on LO output,  $f_{LO} = 1700\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $f_{REF} = 122.88\text{ MHz}$ ,  $f_{PPFD} = 1.536\text{ MHz}$ ,  $f_{REF}$  power = 4 dBm, integer mode loop filter, unless otherwise noted.

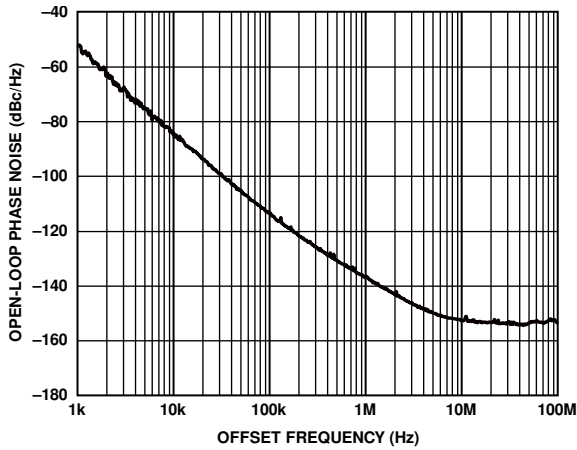


Figure 60. VCO\_0 Open-Loop Phase Noise vs. Offset Frequency,  $f_{VCO_0} = 2.55\text{ GHz}$ , Divide by Two Selected, VCOVTUNE = 1.5 V

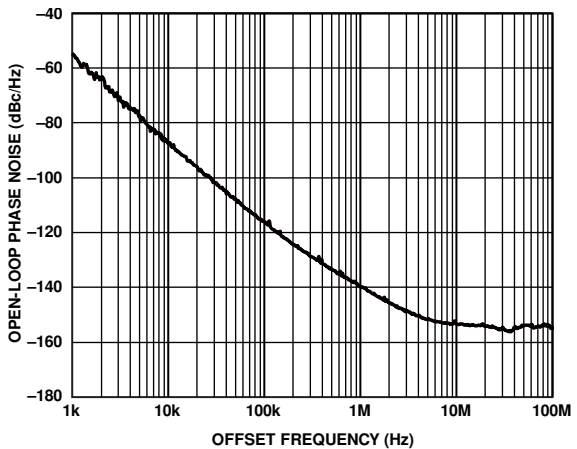


Figure 61. VCO\_1 Open-Loop Phase Noise vs. Offset Frequency,  $f_{VCO_1} = 2.2\text{ GHz}$ , Divide by Two Selected, VCOVTUNE = 1.5 V

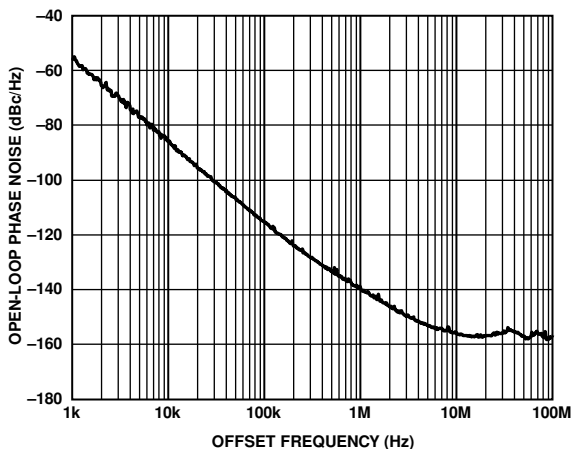


Figure 62. VCO\_2 Open-Loop Phase Noise vs. Offset Frequency,  $f_{VCO_2} = 1.9\text{ GHz}$ , Divide by Two Selected, VCOVTUNE = 1.5 V

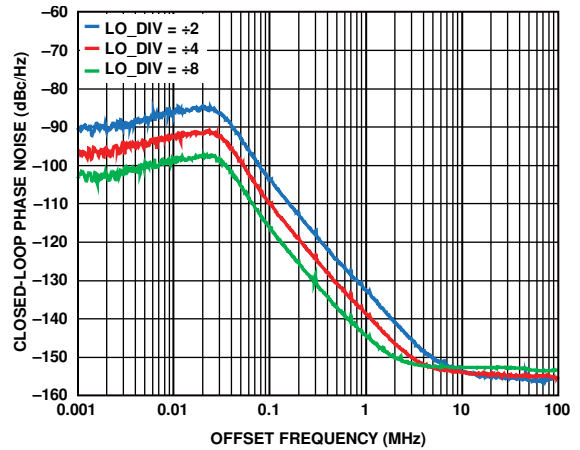


Figure 63. VCO\_0 Closed-Loop Phase Noise vs. Offset Frequency for Various LO\_DIV Dividers,  $f_{VCO_0} = 5.1\text{ GHz}$

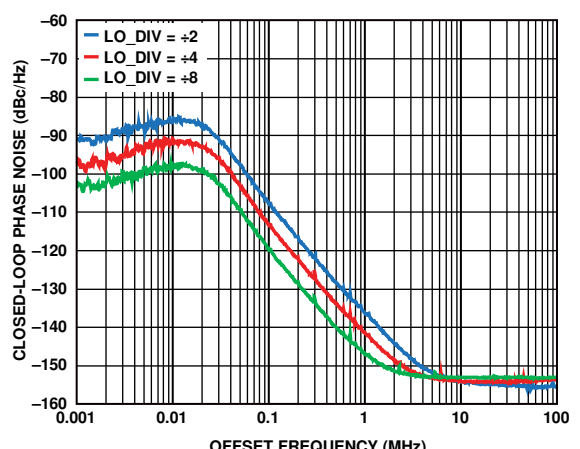


Figure 64. VCO\_1 Closed-Loop Phase Noise vs. Offset Frequency for Various LO\_DIV Dividers,  $f_{VCO_1} = 4.5\text{ GHz}$

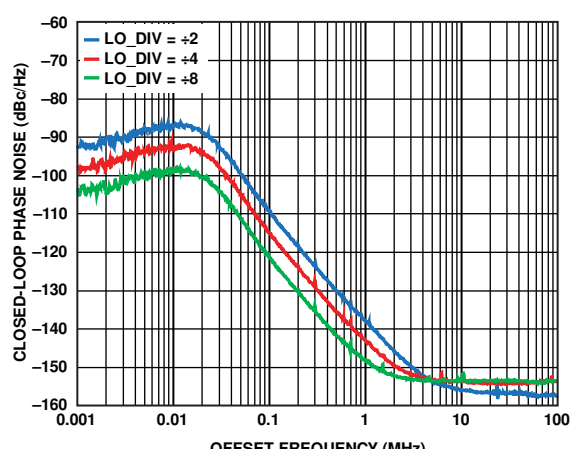


Figure 65. VCO\_2 Closed-Loop Phase Noise vs. Offset Frequency for Various LO\_DIV Dividers,  $f_{VCO_2} = 3.8\text{ GHz}$

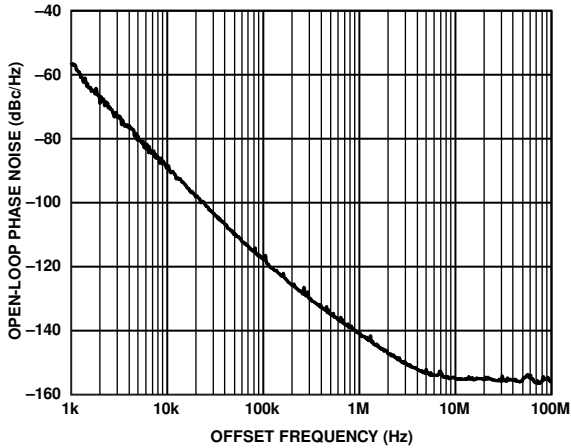


Figure 66. VCO\_3 Open-Loop Phase Noise vs. Offset Frequency,  $f_{VCO\_3} = 1.6$  GHz, Divide by Two Selected,  $VCOVTUNE = 1.5$  V

14115-163

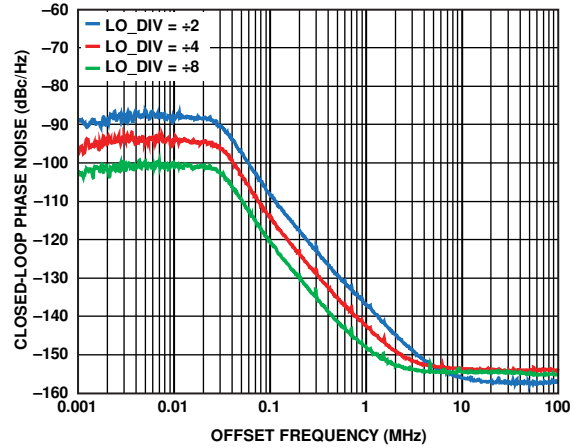


Figure 69. VCO\_3 Closed-Loop Phase Noise for Various LO\_DIV Dividers vs. Offset Frequency,  $f_{VCO\_3} = 3.2$  GHz

14115-066

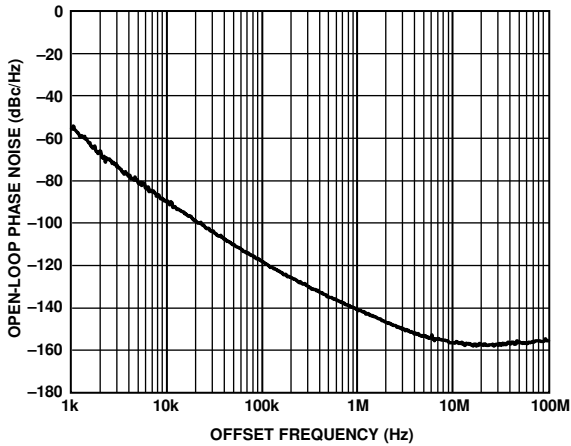


Figure 67. VCO\_4 Open-Loop Phase Noise vs. Offset Frequency,  $f_{VCO\_4} = 3.087$  GHz, Divide by One Selected,  $VCOVTUNE = 1.5$  V

14115-567

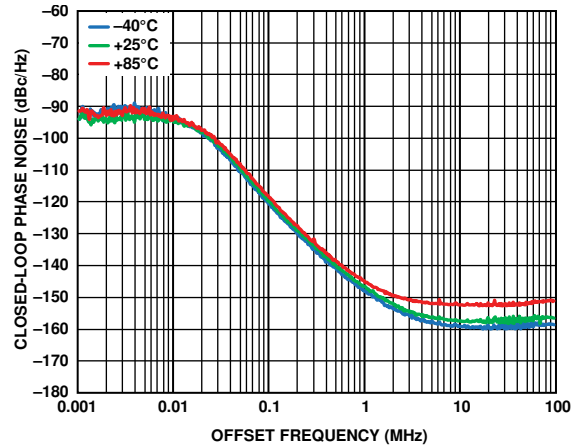


Figure 70. VCO\_4 Closed-Loop Phase Noise for Various Temperatures vs. Offset Frequency,  $f_{VCO\_4} = 1.536$  GHz, Divide by Two Selected

14115-570

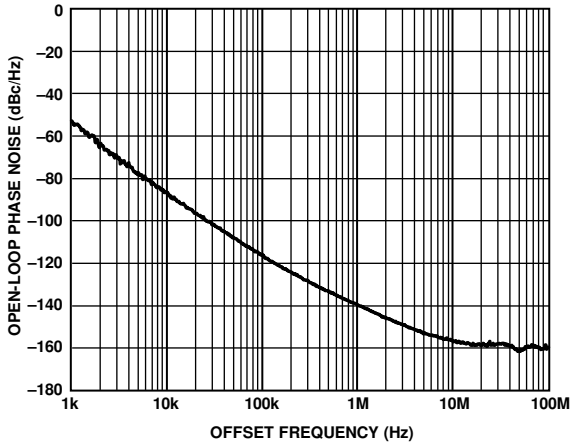


Figure 68. VCO\_5 Open-Loop Phase Noise vs. Offset Frequency,  $f_{VCO\_5} = 3.375$  GHz, Divide by One Selected,  $VCOVTUNE = 1.5$  V

14115-568

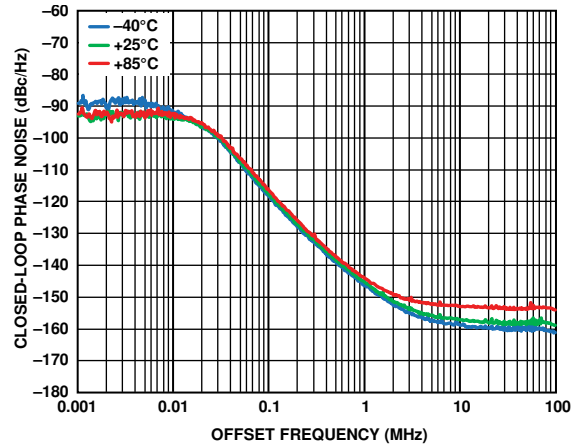


Figure 71. VCO\_5 Closed-Loop Phase Noise for Various Temperatures vs. Offset Frequency,  $f_{VCO\_5} = 1.688$  GHz, Divide by Two Selected

14115-571