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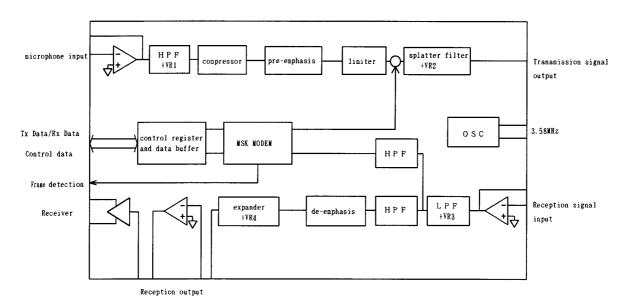
☐ Package: 24 pin VSOP

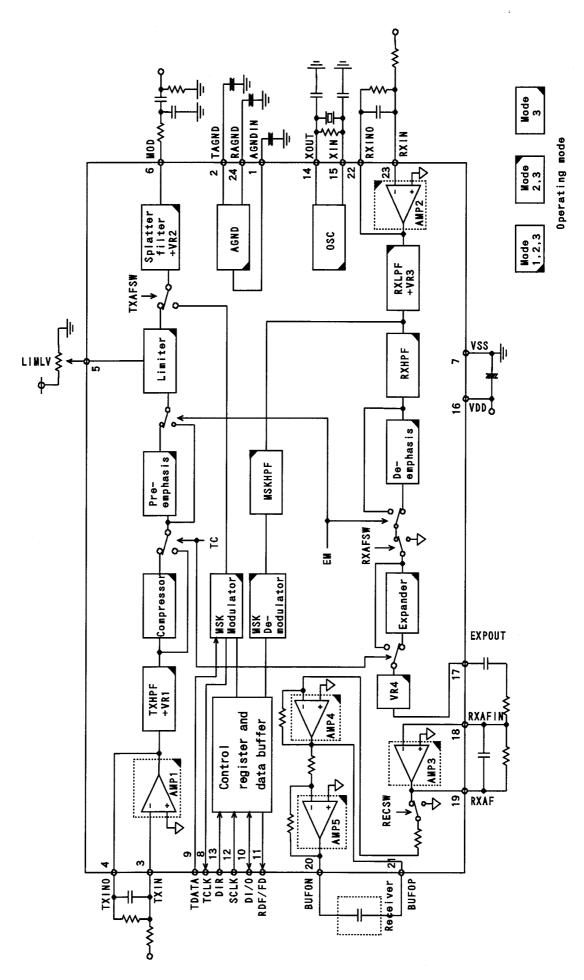
AK2361

Base-band LSI for Cordless Telephones

Features
Built-in voice filter for cordless telephone, MSK MODEM (2400bps) and COMPANDOR Low / wide operation voltage range (1.9 V to 5.5V)
Built-in COMPANDOR output transient response circuit and time constant circuit No external component is needed for COMPANDOR
Built-in buffer amplifier for ceramic receiver driving.
Built-in electronic volume for microphone sensitivity and modulator/demodulator
sensitivity
Receiving level switchable in 8 steps (-12 to +9dB)
Built-in muting function for voice transmitting and receiving
External adjustment for the limiter level
Built-in amplifier for transmission and reception gain adjustment
Low power CMOS and power-down function
Built-in 3.58MHz oscillator circuit
Built-in frame detection function for the MSK demodulator
Control register and MSK MODEM data buffer controlled by serial interface
Few external component is necessary resulting cost reduction and small set size.
AK2358/58A pin compatible

Block Diagram





Description

The AK2361, a base-band LSI for cordless telephone, has built-in voice filters, a 2400bps MSK MODEM for data communication, a frame detection circuit and a COMPANDOR for noise reduction.

The CMOS process provides low power operation. Application of 24 pin VSOP package with the feature of significant reduction of external component provides minimum mounting area. The time constant circuit for the COMPANDOR output transient response is built into the LSI.

Using a 2400bps MSK MODEM for data communication has realized high data reliability and high speed communication at the same time.

This LSI is suitable for cordless system telephones etc. which requires complicated protocol control.

An oscillation circuit with a 3.58MHz crystal oscillator is built in, and no other frequency source is required for the MSK MODEM. The oscillator also can be used for the other DTMF generator etc.

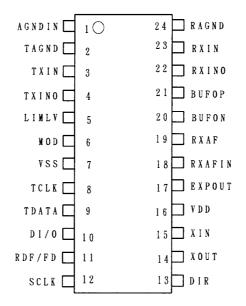
Built-in electronic volumes provided for transmission and reception part realize automatic adjustment of the microphone sensitivity and the modulator/demodulator sensitivity by external EEPROM and microprocessor.

The transmission part is composed of high-pass filter, compressor, pre-emphasis circuit, limiter, MSK modulator, splatter filter, electronic volume control, etc.

The reception part is composed of band pass filter, de-emphasis circuit, buffer amplifier, expander, MSK demodulator, frame detection circuit, electronic volume control, etc.

Pin Arrangement

24 pin VSOP



Circuit Configuration

Block	Function
AMP1	The operational amplifier for voice signal transmission gain
	adjustment and for the filter to eliminate aliasing noise by the
	SCF(switched capacitor filter) in the following stage. Use an
	external resistor and capacitor to set the gain less than 30dB and
	the cut-off frequency to about 10kHz.
TXHPF	The SCF circuit to eliminate the low frequency component less
	than 300Hz from the transmission voice signal.
Compressor	The circuit to compress the amplitude of the transmission voice
	signal.
Pre-emphasis	The circuit to emphasis the high-frequency component of the
	transmission voice signal to improve the S/N of the modulation
	signal.
Limiter	The amplitude-limiting circuit to suppress the frequency deviation
	of the modulation signal. The limitation level can be adjusted by
	applying a DC voltage to the LIMLV pin. If the LIMLV pin is open,
	the default limitation level is applied.
Splatter filter	The SCF circuit to eliminate the high frequency component higher
	than 3.0kHz from the limiter output signal or the MSK modulator
	signal.
MSK modulator	The circuit to generate a 2400bps MSK signal according to the
	received digital signal logic from the TDATA pin.
AMP2	The operational amplifier to adjust the reception demodulation
	signal gain and for the filter to eliminate the aliasing noise of
	the SCF in the following stage. Set the gain to less than 30dB
	and the cut-off frequency to about 10kHz by external resister and
	capacitor.
RXLPF	The SCF circuit to eliminate the high frequency component higher
	than 3.0kHz from the limiter output signal or the MSK modulator
	signal.
RXHPF	The SCF circuit to eliminate the low frequency component lower
	than 300Hz from the reception voice signal.
De-emphasis	The circuit to de-emphasis the emphasized signal by pre-emphasis
	circuit.
Expander	The circuit to expand the signal amplitude compressed by the
	compressor.

Block	Function
AMP3	The operational amplifier used on the smoothing filter of the
	reception SCF output. Set the gain to OdB and the cut-off
	frequency to about 20kHz by external resister and capacitor.
MSKHPF	The SCF circuit to eliminate the low frequency component lower
	than 100Hz from the reception MSK signal.
MSK demodulator	The circuit to reproduce the 2400bps receiving data and the clock
,	from the received MSK signal in the RXIN pin.
AMP4	The inverting and the non-inverting buffer amplifier to drive the
AMP5	ceramic receiver.
AGND	The circuit to generate the reference voltage for the internal
	analog signal.
Oscillation	The circuit to oscillate the 3.58 MHz reference clock using an
circuit	external crystal oscillator and resistor.
VR1	The volume to control the input amplitude of the transmission
	voice signal. The adjustment range is -8dB to +7dB by 1dB step.
VR2	The volume to control the MOD output amplitude. The adjustment
	range is -4dB to +3.5dB by 0.5dB step.
VR3	The volume to control the input amplitude of the reception
	demodulation signal. The adjustment range is -4dB to +3.5dB by
	0.5dB step.
VR4	The volume to control the receiving voice amplitude. The
	adjustment range is -12dB to +9dB by 3dB step.
Control register	The control register controls the status of internal switches and
and data buffer	internal volumes of the LSI by serial data consists of 2 address
	bits and 8 data bits. At the start up a power-on-reset circuit
	works and the default values are set to the control register.
	(see control register map.)
	The data buffer stores 8 bits of the MSK receiving data to smooth
	the signal interface with CPU.

Pin/Function

Pin No.	Pin name	I/O	Function
1	AGNDIN	I	Analog ground input pin.
			Connect the capacitor to stabilize the analog ground.
2	TAGND	0	Analog ground pin for the transmission system.
	INUND		
			Connect the capacitor to stabilize the analog
			ground.
3	TXIN	I	Transmission voice input pin.
			This is the inverting input pin for AMP1. It
			composes a microphone amplifier with a external
			resister and a capacitor.
4	TXINO	0	AMP1 output pin.
5	LIMLV	I	Limitation level adjustment pin.
			The limitation level can be adjusted by applying a
			DC voltage to this pin. The default limitation level
			is adopted if no voltage is applied.

Pin No.	Pin name	I / O	Function
6	MOD	0	Output pin of the modulated transmission signal.
LIMITE			A load impedance larger than $10k\Omega$ can be driven.
. 7	VSS		Negative power supply pin.
8	TCLK	0	Clock output pin for the MSK data transmission.
			A 2.4kHz clock is put out by setting the internal
			register TDE to "0". If the register is set to "1",
			it goes "H" level.
9	TDATA	I	MSK transmission data input pin.
			Data are latched synchronizing with the TCLK rising
			edge.
1 0	DI/O	I / O	Serial data input and output pin.
1 1	RFD/FD	0	MSK signal reception flag output and Frame detection
			signal output pin.
			This pin puts out two types of information,
			depending on the status of the internal register
			FSL. If FSL is "1", it is MSK signal reception mode,
			so the pin reaches low after 8 bits of the MSK
			reception signal have been written to the data
			register. If FSL is "0", it is the frame detection
			signal output mode, so the low pulse is put out
1.0	OCLV	T	after a frame pattern is detected.
1 2	SCLK	I	Clock input pin for serial data I/O.
$\frac{13}{14}$	DIR XOUT	I	Serial data I/O control pin. Crystal oscillator connection pin.
1 4	A001	1	The reference clock IC is generated by connecting a
			3.58MHz crystal oscillator parallel to a $1M\Omega$
			resistor between this pin and XIN pin. In case of
			external clock operation, connect XOUT pin to VSS
			and apply the clock to XIN.
1 5	XIN	0	Crystal oscillator connection pin.
1 6	VDD	_	Positive power supply pin.
1 7	EXP O UT	0	Expander output pin.
1 8	RXAFIN	I	Reception voice input pin.
			This is the inverting input of AMP3. It composes a
			smoothing filter by external resistor and capacitor.
1 9	RXAF	0	Reception voice output pin.
			This is the output pin of AMP3. A load impedance
			more than $10k\Omega$ can be driven.
2 0	BUF O N	О	Receiver amplifier output pins.
2 1	BUFOP	0	Connect the ceramic receiver between these two pins.
2 2	RXINO	0	AMP2 output pin.
2 3	RXIN	I	Demodulated receiving signal input pin.
			This is the inverting input of AMP2. It composes a
			prefilter with external resistor and capacitor.
2 4	RAGND	0	Analog ground pin for the reception system.
			Connect the capacitor to stabilize analog ground.

Absolute Maximum Ratings

VSS=0V; Note 1)

Parameter	Symbol	min	max	Unit
Power supply voltage: (VDD)	VA+	-0.3	6.5	V
Input current (except the power supply pins)	IIN	-	± 10	mA
Analog input voltage	VINA	-0.3	(VA+)+0.3	V
Digital input voltage	VIND	-0.3	(VA+)+0.3	V
Storage temperature	Tstg	-55	130	$^{\circ}$ C

Note 1): All voltages with respect to the VSS pin.

Warning: Operation at or beyond these limits may result in permanent damage to the device.

Normal operation is not guaranteed at these extremes.

Recommended Operating Conditions

VSS=0V; Note 1)

Parameter		Symbol	min	typ	max	Unit
Operation temperature		Ta	-10		70	°C
Power supply voltage:	(VDD)	VD+	1. 9	2. 0	5. 5	V
Analog reference volt	age	AGND		1/2VD+		V
Current consumption	Mode 0	Idd0		0.1	0.8	mA
	Mode 1	Idd1		0.9	1.9	
	Mode 2	Idd2		1.5	2.9	
	Mode 3	I dd3		5.0	10	

Note 1): All voltages with respect to the VSS pin.

Analog Characteristics

f=1kHz, TC="1", EM="1", VR1 to VR4=0dB:unless otherwise specified,

0dBm=0.775Vrms

0dBx=-5dBm at AVDD=2V Note 8)

1) TX system

Para	meter	min	typ	max	Unit
Standard input level	@TX INO		-10		dBx
Absolute gain	TXINO→MOD Note 1)	2. 0	3. 5	5.0	dB
Limiter level	MOD 1kHz Note 1)				
	No external R	-4.5	-3.5	-2.5	יימף.
	Adjustment range by			-2.5	dBx
	external R				
Compressor linearity	TXINO→MOD Note 1) 2)				
	TXINO=-44dBx	-20	-17.0	-14	dB
	TXIN0=-50dBx	-24	-20.0	-16	
Noise without input	TXINO→MOD Note 1) 3)			-36.5	dBm
Compressor distortion	TXINO→MOD			-35	dB
	TXIN0=-10dBx			- 50	uБ
Transmission MSK	@MOD Note 1)	-4.5	-3.5	-2.5	dBx
signal level	1.2kHz signal output	-4. J	ე. ე	- 4. 0	UDX
Transmission MSK	@MOD Note 1)			-32	dB
signal distortion	1.2kHz signal output			J.L	uр

2) RX system

Para	ameter	min	typ	max	Unit
Standard input level	@RX INO		-10		dBx
Absolute gain	RXINO→BUFON, BUFOP Note 1)	-1.5	0	+1.5	dB
Expander linearity	RXINO→BUFON, BUFOP Note 1) 4)				
	RX I NO=-25 dBx RX I NO=-30 dBx	-33.0 -45.0	-30. 0 -40. 0	-27. 0 -35. 0	dB
Noise with no input	RXINO→BUFON, BUFOP Note 1) 3)			-70	dBm
Expander distortion	RXINO→RXAF RXINO=-5dBx			-35	dB
Reception MSK signal level	@RXINO 1.2kHz signal output	-14	-7	-1	dBx

3) Overall characteristics

	Parameter		min	typ	max	Unit
Absolute gai	n TXINO→BUFON, BUFOP	Note 5)	-0.5	0	+2.5	dB
	TXINO = -10dBx					
Distortion	TXINO→BUFON, BUFOP	Note 3) 5)		-50	-43	dB
	TX I N 0 = - 1 0 d B x					
Crosstalk	@BUFON, BUFOP	Note 1) 6)				
Transmission	n → Reception				-60	dBx
	TXINO=0dBx TC=	0"				
Crosstalk	@MOD	Note 1) 6)	·			
Reception →	· Transmission				-56.5	dBx
	RXINO=0dBx $TC=$	0"				

4) Filter characteristics

Parameter		min	typ	max	Unit
Transmission overall characteristics (S	ee Fig. 1)				
TXINO → MOD	100Hz		•	-40	
TC="0" EM="1"	300Hz	-12	-10.5	-9	
Relative value with OdB gain	2.5kHz	6.5	8	9.5	dB
at 1kHz	3kHz	6.5	8	9. 5	
	5kHz			-7	
Reception overall characteristics (See	Fig. 2)				
RXINO → EXPOUT	100Hz			-4	
TC="0" EM="1"	250Hz		12	13.5	
Relative value with OdB gain	300Hz	9	10.5		dB
at lkHz	3kHz	-10.5	-9	-7.5	
	5kHz			-15	

- Note 1) With the external circuit shown in the application circuit example.
- Note 2) Relative value with OdB as the MOD output level at the time of input of standard input level (-10dBx) to TXINO.
- Note 3) With the C-message filter.
- Note 4) Relative value with OdB as the BUFON, BUFOP output level at the time of input of standard input level (-10dBx) to RXINO.
- Note 5) With the external circuit shown in the application circuit example. Further, the AMP2 gain should be -3.5dB, and MOD and RXIN should be in loop connection.
- Note 6) TC="0"
- Note 7) The dBx is standardized unit valid for various power supply voltages from 1.9 to 5.5V. If the voltage is 2V, 0dBx should be -5dBm. With the other voltage as X [V], $0dBx = -5 + 20 \log (X/2) [dBm]$.

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☐ Filter characteristics

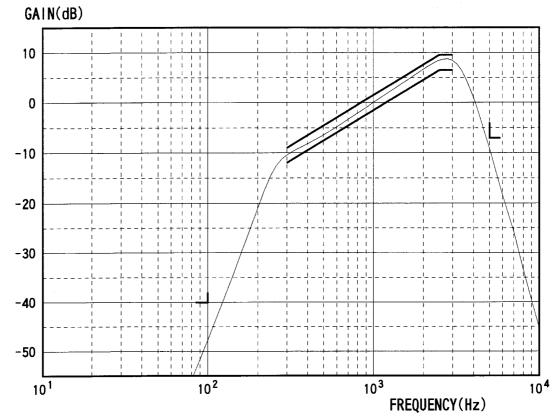


Fig. 1 Transmission overall characteristics

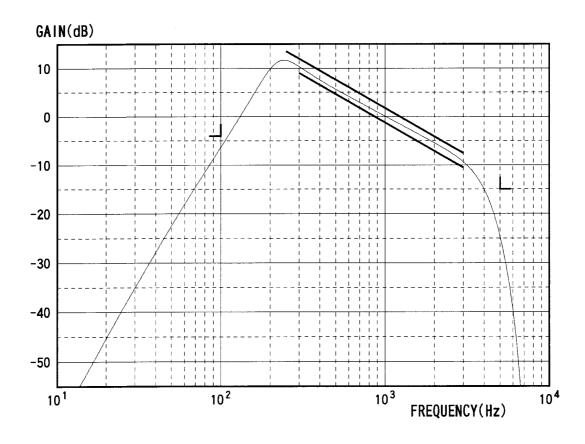
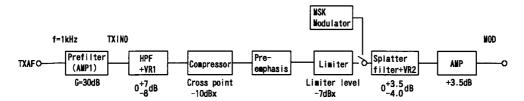
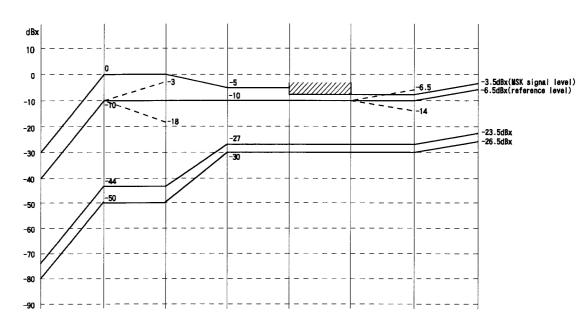


Fig. 2 Reception overall characteristics

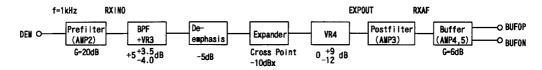
Level Diagram

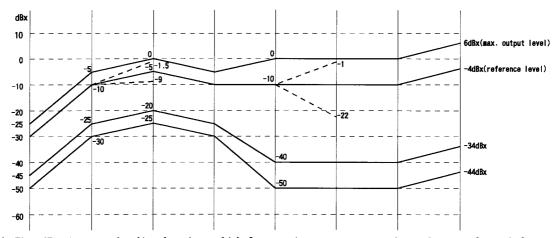
1) TX system





2) RX system





Note) The dBx is standardized unit valid for various power supply voltages from 1.9 to 5.5V. If the voltage is 2V, 0dBx should be -5dBm. With the other voltage as X [V], $0dBx = -5 + 20 \log (X/2) [dBm]$.

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Digital Characteristics

1. DC Characteristics

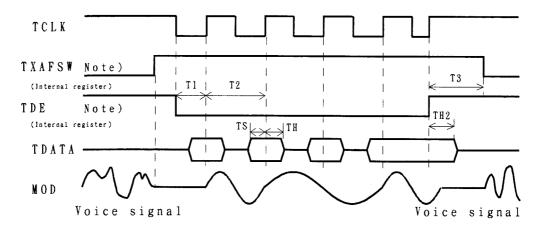
Parameter	Pin	Symbol	min	typ	max	Unit
High-level input voltage 1	(1)	V _{IH1}	70%VD+			V
Low-level input voltage 1	(1)	V 1 L 1			30%VD+	V
High-level input voltage 2	(2)	V 1 H 2	80%VD+			V
Low-level input voltage 2	(2)	V 1 L 2			20%VD+	V
High-level input current V _{IH} =VD+	(1)(2)	IIH			10	μΑ
Low-level input current V _{1L} =0V	(1)(2)	IIL	-10			μΑ
High-level output voltage IoH=0.1mA	(3)	V _{он}	90%VD+			V
Low-level output voltage Iol=0.6mA	(3)	Vol			0.3	V

- (1) TDATA, DI/0
- (2) SCLK, DIR
- (3) TCLK, RDF, DI/O

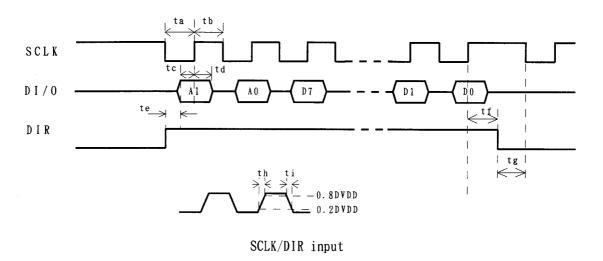
2. AC Characteristics

Parameter	Symbol	min	typ	max	Unit
Master clock frequency	fclk	•**	3.579545	:	MHz
MSK Modulator timing					
TDE Falling to TCLK Rising	T1		208. 3		μs
TCLK period	T2		416.7		μs
TDE Rising to TXAFSW Falling	Т3	2			ms
TDATA Set up time	TS	1			μs
TDATA Hold time	TH	1			μs
TDATA Hold time2	TH2	2			μs
MSK Demodulator timing					
RCLK Period & FD pules width	T	402.2	416. 7		μs
Serial data input timing					
Clock pulse width 1	ta	500			ns
Clock pulse width 2	tb	500			ns
SDATA Set Up time	tc	100			ns
SDATA Hold time	td	100			ns
DIR Set up time	te	100			ns
DIR Hold time	tf	100			ns
DIR falling to SCLK falling time	tg	100			ns
SCLK/DIR input rising time	th			1	μs
SCLK/DIR input falling time	ti			1	μs
RDF falling to SCLK falling time	tj	100			ns
SCLK rising to RDF falling time	tk	600			ns

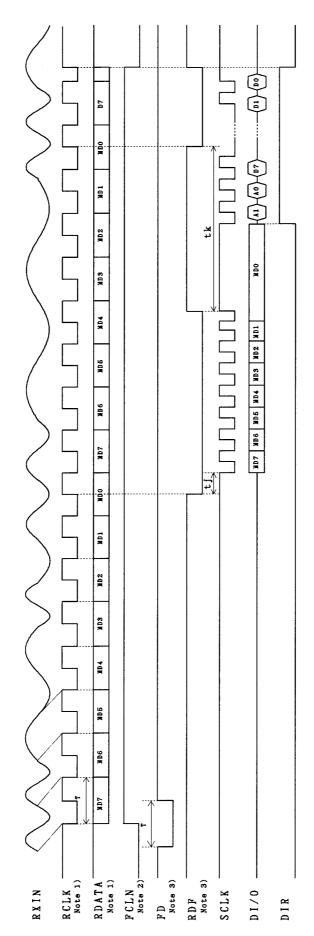
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MSK modulator



Note) The timing to rewrite the internal registers TXAFSW and TDE is synchronized with the falling edge of DIR.



Note 1) Internal node Note 2) Internal register Note 3) If the internal register FSL is "O", the status of "FD" is put out to the RFD/FD pin. If the FCL is "1", the status of "RFD" is put out.

MSK demodulator

Control Register Map

■ Register composition

	Addr	ess		Data						
	A 1	A 0	D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0
Control register 1	0	0	FSL	BS2	BS1	FCLN	-	TDE	TXAF -SW	RXAF -SW
Volume register	0	1	1	1	1	1	RECSW		V R 4	
Volume register	1	0		V]	R 2	****		V R	1	
Control register 2 + volume register	1	1	TC	TC EM FRPT - VR3						
Reception data register			MSK MODEM reception data							

The reception data register is a read only register, and the others are write only registers.

The reception data register has no address information proceeding to the Data. Set the bits D4 to D7 of volume register address "01" to all "1". If they are set to except all "1", it is changed to test mode. Do not use this test mode. If the bit D3 of address "00" and D4 of address "11" is written either "0" or "1", setting of AK2361 is not changed.

Register map

1) Control register 1

Addr	ess	Data							
A 1	A 0	D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0
0	0	FSL	BS2	BS1	FCLN	-	TDE	TXAFSW	RXAFSW
(Defa	ult)	1	1	0	0	_	1	0	0

a) Transmission signal control

TDE	TXAFSW	Transmission output
1	0	Voice signal
1	1	Mute
0	1	MSK signal

b) Reception signal control

RXAFSW	RECSW	RXAF	BUFOP/BUFON
1		Mute	Mute
0	1	ON	Mute
0	0	ON	ON

c) Frame detection circuit 0N/0FF

FCLN	
1	The frame detection function is not used (OFF).
0	The frame detection function is used (ON).

Note) FCLN automatically changes from 0 to 1 when a synchronized frame is detected.

d) Power-down mode

BS2	BS1	Mode name	Voice system + transmission MSK	Reception MSK	0scillator
1	1	mode0	OFF	OFF	OFF
0	1	mode1	OFF	OFF	O N
1	0	mode2	OFF	O N	O N
0	0	mode3	O N	O N	O N

e) RDF/FD selection

FSL	
1	The MSK signal reception flag (RDF) is put out from the RDF/FD pin.
0	The frame detection signal (FD) is put out from the RDF/FD pin.

2) Control register 2

Addr	ess	Data							
A 1	A 0	D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0
1	1	TC	EM	FRPT	-		VR	3	
(Defa	ult)	1	1	0	_	1	0	0	0

Data name	Function							
FRPT	Frame detector detection pattern	"1": 1100010011010110 (base unit)						
		"0": 1001001100110110 (portable unit)						
ΕM	Emphasis circuit	"1": Passage (ON)						
		"0": Bypass (OFF)						
ТС	COMPANDOR circuit	"1": Passage (ON)						
		"0": Bypass (OFF)						

3) Volume register

Addr	ess	Data							
A 1	A 0	D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0
0	1	1	1	1	1	RECSW	VR42	VR41	VR40
1	0	VR23	VR22	VR21	VR20	VR13	VR12	VR11	VR10
1	1	TC	EM	FRPT		VR33	VR32	VR31	VR30

a) VR1 volume control

VR13	VR12	VR11	VR10	Volume gain (dB)
0	0	0	0	-8.0
0	0	0	1	-7.0
0	0	1	0	- 6 . 0
0	0	1	1	-5.0
0	1	0	0	-4.0
0	1	0	1	- 3. 0
0	1	1	0	-2.0
0	1	1	1	-1.0
1	0	0	0	0
1	0	0	1	+1.0
1	0	1	0	+ 2. 0
1	0	1	1	+ 3. 0
1	l 1	0	0	+ 4. 0
1	1	0	1	+ 5. 0
1	1	1	0	+6.0
1	1	1	1	+7.0

b) VR2, VR3 volume control

VR23	VR22	VR21	VR20	Volume main (dD)
VR33	VR32	VR31	VR30	Volume gain (dB)
0	0	0	0	-4.0
0	0	0	1	- 3. 5
0	0	1	0	- 3. 0
0	0	1	1	- 2. 5
0	1	0	0	- 2. 0
0	1	0	1	-1.5
0	1	1	0	-1.0
0	1	1	1	- 0. 5
1	0	0	0	0
1	0	0	1	+ 0. 5
1	0	1	0	+ 1. 0
1	0	1	1	+ 1. 5
1	1	0	0	+ 2. 0
1	1	0	1	+ 2. 5
1	1	1	0	+ 3. 0
1	1	1	1	+ 3. 5

c) VR4 volume control

VR42	VR41	VR40	Volume gain (dB)
0	0	0	-12
0	0	1	- 9
0	1	0	<u> </u>
0	1	1	- 3
1	0	0	0
1	0	1	+ 3
1	1	0	+ 6
1	1	1	+ 9

Note) By reset, the gain of all volumes are set to 0dB and RECSW bit is changed to "0".

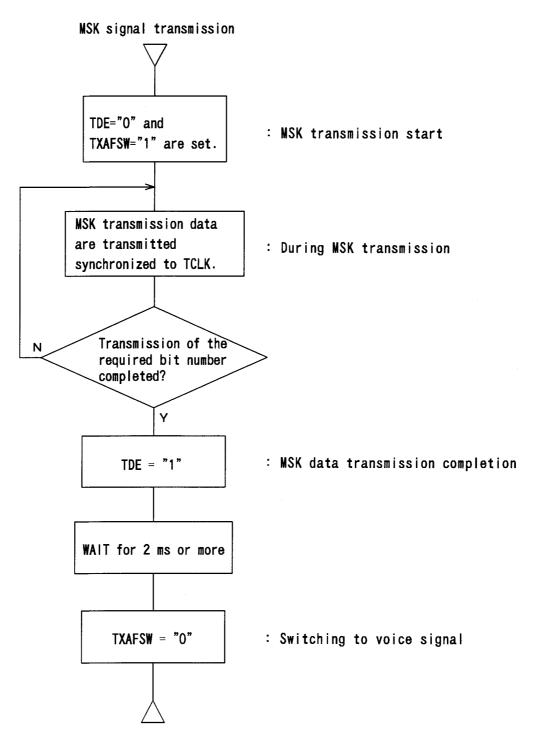
4) MSK MODEM reception data

Data							
D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0
RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0

Data name		Function	
RD0	MSK reception data	"1":1.2kHz	
} }		"0":2.4kHz	
R D 7	RD7 is the first received data.		

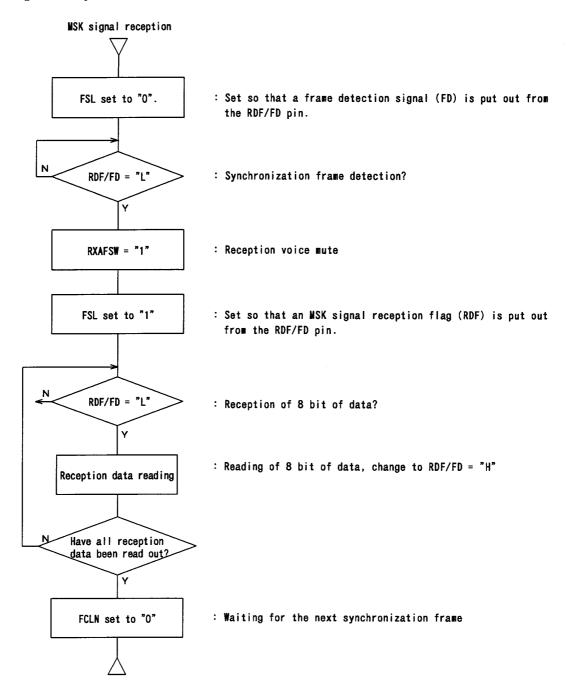
MSK Modem

MSK signal transmission flow



- (1) Set the serial register "TDE" to "0" and "TXAFSW" to "1", so that MSK transmission state is provided.
- (2) A 2400Hz clock is put out from TCLK. Synchronizing with the rising edge of TCLK, AK2361 reads the MSK transmission data from TDATA pin and put out them to MOD pin.
- (3) After the transmission of the necessary number of signal bit, "TDE" of the serial register is set to "1".
- (4) Afterwards, before switching to a voice signal transmission mode, wait at least 2ms after "TDE" has set to "1" to complete the MSK signal final bit transmission. Then set TXAFSW register to "0".

MSK Signal Reception



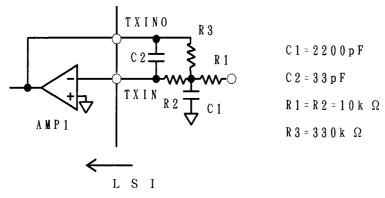
- (1) If the internal register "FCLN" is "O", the internal nodes RDATA, RCLK are fixed to
- (2) After a synchronization frame is detected, FD goes to "L" during the period "T", then FCLN is set to "1".
- (3) RDATA and RCLK put out the data following to the synchronized frame pattern, and these are stored in the internal buffer.
- (4) After 8 bit of reception data have been entered to the internal buffer, RDF goes "L".
 (5) After the CPU detect that RDF is "L", it puts out 8 clock bits to SCLK, then read 8 bit of reception data from the DI/O pin.
- (6) With input of 8 clock bits to SCLK, RDF goes "H".(7) Afterwards, by repeating the steps (4) and (5) the necessary data bits are read.
- (8) After the necessary data have been read, DIR goes "H", "FCLN" is set to "O" via the serial interface, the internal nodes RDATA and RCLK are set to "1", then the system waits for the next synchronization frame.

Application Circuit Example

■Application Circuit

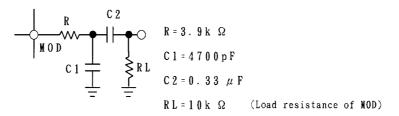
OAMP1

Use as a transmitting microphone amplifier. The gain should be less than 30dB. To eliminate high frequency noise component over than 100kHz from input signal, 1st order or 2nd order anti-aliasing filter is necessary. The following drawing is one example of the 2nd order anti-aliasing filter, which has 30dB gain and 10kHz cut-off frequency.



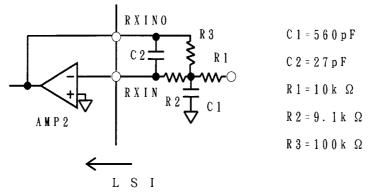
Smoothing filter for MOD output signal

Realize low-pass filter to eliminate 112kHz clock signal component from MOD pin output signal. The following is one example of the 1st order low-pass filter which has 8.7kHz cut-off frequency. $10k\,\Omega$ of the modulator load resistor(RL) provide 3.3dB signal attenuation.



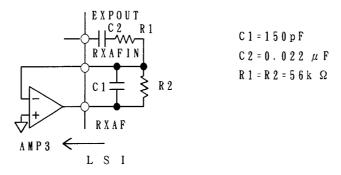
OAMP 2

The amplifier for the receiving gain adjustment and anti-aliasing filtering to eliminate high frequency noise component over 100kHz. The gain should be less than 30dB. The following is an example of the 2nd order low pass filter, which has 20 dB gain and 40kHz cut-off frequency.

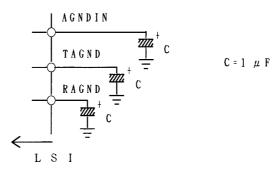


OAMP3

The smoothing filter to eliminate 448kHz clock component from EXPOUT signal is provided by this amplifier. Also it works to adjust the receiving gain. Adding the other pass signal may be possible. The following is one example of the 1st order low-pass filter, which has 0dB gain , 19kHz cut-off frequency.

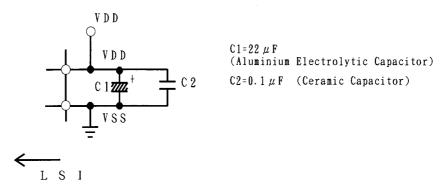


To stabilize the AGND potential, connect capacitors larger than 0.3 μ F between TAGND pin, RAGND pin and AVSS pin. Also between AGNDIN pin and AVSS pin some capacitor is necessary to reduce the ripple of the power.



○VDD stabilizing capacitor

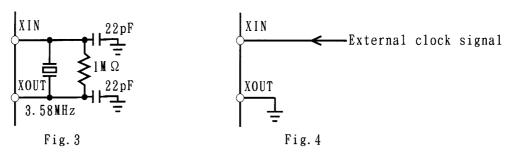
To reduce the noise on YDD, connect capacitors between YDD and VSS.

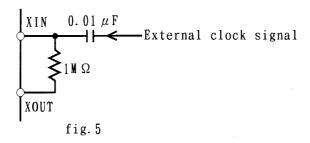


OCrystal oscillator

- Crystal resonator, resistor and capacitors should be connected as shown Fig. 3 for on-chip oscillator operation.

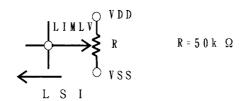
- For external clock operation, if the high(H) level of the input clock signal amplitude equals to or is greater than 1.5V, and the low(L) level equals to or is smaller than 0.5V, then connection should be made as shown in Fig. 4. If the input clock signal amplitude (peak-to-peak) equals to or is smaller than 1V, and equals to or is greater than 200mV, then AC coupling should be as illustrated in Fig. 5.





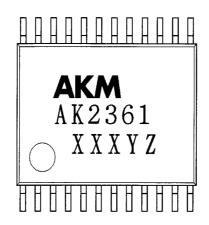
OLimit level adjusting resistor

The limiting level can be controlled externally by applying DC voltage to LIMIV pin. Applied DC voltage should be larger than TAGND, then the limiting level is shown as $TAGND \pm Va(V)$, while Va is the voltage between LIMIV and TAGND. Keeping LIMIV pin open provides default limit level. See following example.



Package

Marking



[Contents of XXXYZ]

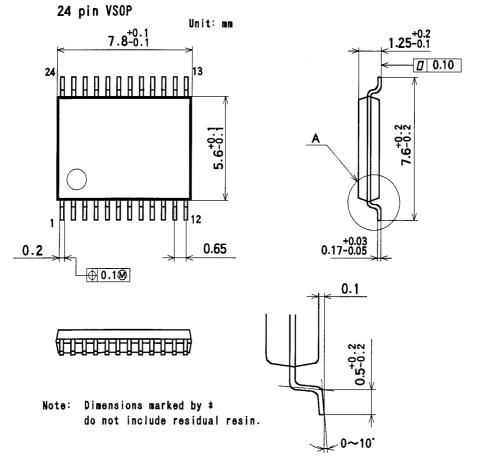
XXX: Date of manufacture

Last digit of the year, week number of the year as 2 digits

Y: Production lot number

Z: Assembled place

■ Shape and dimensions of the package



Detail of part A

[Material] Resin: Low-stress type epoxy resin

Lead frame: Cu