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Datasheet

AS1528/AS1529 Micro-Power, 10-Bit, 150ksps, ADC, Single Ended or Differential

1 General Description

The AS1528/AS1529 are low-power, 10-bit analog-todigital converters (ADCs) designed to operate with a single +2.7V to +5.25V supply. Excellent dynamic performance, low power consumption, and simplicity make these devices perfect for portable battery-powered dataacquisition applications.

The devices are available as the standard products listed in Table 1.

Table 1. Standard Products

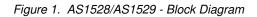
Model	Input Type	Input Voltage
AS1528	1-Channel, Pseudo / True-Differential	0 to VREF / -VREF/2 to VREF/2
AS1529	2-Channel, Single- Ended	0 to VREF

The devices feature a successive-approximation register (SAR), automatic shutdown, fast wakeup (1.4 μ s), and low-power consumption at the maximum sampling rate of 150ksps.

Automatic shutdown $(0.2\mu A)$ between conversions results in reduced power consumption (at slower throughput rates).

Data access are made via an external clock through the SPI-/QSPI-/MICROWIRE-compatible 3-wire high-speed serial interface.

The AS1529/AS1528 are available in a 8-pin TDFN (3x3mm) package.

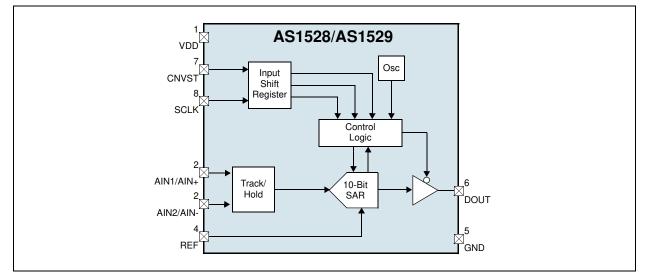


2 Key Features

- Single-Supply Operation: +2.7V to +5.25V
- Automatic Shutdown Between Conversions
- Low Power Consumption
 - 350µA @ 150ksps
 - 245µA @ 100ksps
 - 24µA @ 10ksps
 - 2.5µA @ 1ksps
 - 200nA in Automatic Shutdown Mode
- True-Differential Track/Hold, 150kHz Sampling Rate Software-Configurable Unipolar/Bipolar Conversion (AS1528)
- Input Common Mode Range from GND to VDD
- 3-Wire SPI-/QSPI-/MICROWIRE-Compatible Serial Interface
- Internal Conversion Clock
- 8-pin TDFN (3x3mm) Package

3 Applications

The devices are ideal for remote sensors, data-acquisition, data logging devices, lab instruments, or for any other space-limited A/D devices with low power consumption and single-supply requirements.



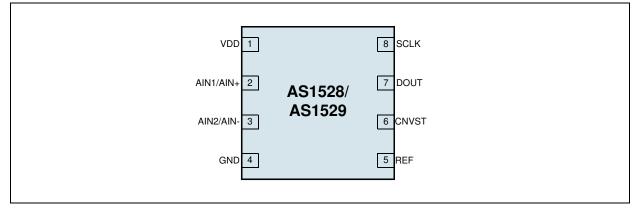
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4 Pinout

Pin Assignment

Figure 2. Pin Assignments (Top View)



Pin Description

Table 2. Pin Description

Pin Number	Pin Name	Description
1	VDD	Positive Supply Voltage. +2.7V to +5.25V.Note:Bypass with a 0.1μF capacitor to GND.
2	AIN1/AIN+	Analog Input Channel 1 (AS1529) or Positive Analog Input (AS1528)
3	AIN2/AIN-	Analog Input Channel 2 (AS1529) or Negative Analog Input (AS1528)
4	GND	Ground
5	REF	External Reference Voltage Input. Sets the analog voltage range. Note: Bypass with a 4.7μF capacitor to GND.
6	CNVST	 Conversion Start. A rising edge powers up the device and puts the track/ hold circuitry in track mode. At the falling edge of this pin, the device enters hold mode and begins a conversion. Note: This pin also selects the input channel (AS1529) or input polarity (AS1528).
7	DOUT	 Serial Data Output. This pin transitions the falling edge of SCLK and goes low at the start of a conversion and delivers the MSB at the completion of a conversion. Note: This pin goes high impedance once data has been fully clocked out.
8	SCLK	Serial Clock Input. Clocks out data at DOUT with the MSB first.

5 Absolute Maximum Ratings

Stresses beyond those listed in Table 3 may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in Electrical Characteristics on page 5 is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 3.	Absolute	Maximum	Ratings
----------	----------	---------	---------

Parameter	Min	Max	Units	Comments
VDD to GND	-0.3	+6	V	
CNVST, SCLK, DOUT, REF, AIN1/ AIN+, AIN2/AIN- to GND	-0.3	VDD + 0.3	V	
Current into Any Pin		50	mA	
Continuous Power Dissipation	1491		mW	TAMB = $+70^{\circ}$ C; derate 19.5mW/ $^{\circ}$ C above $+70^{\circ}$ C
Operating Temperature Range	-40	+85	°C	
Storage Temperature Range	-60	+150	°C	
Package Body Temperature		+260	°C	The reflow peak soldering temperature (body temperature) specified is in accordance with IPC/JEDEC J-STD-020D "Moisture/Reflow Sensitivity Classification for Non-Hermetic Solid State Surface Mount Devices". The lead finish for Pb-free leaded packages is matte tin (100% Sn).

6 Electrical Characteristics

 $V_{DD} = +2.7 \text{ to } +5.25 \text{V}, \text{ VREF} = +2.5 \text{V}, 4.7 \mu \text{F}$ Capacitor at REF; fscL κ = 8MHz (50% Duty Cycle); AIN- = GND (AS1528) TAMB = TMIN to TMAX (unless otherwise specified). Typical Values at TAMB = +25°C. Unipolar Mode (AS1528). Table 4. Electrical Characteristics

GND (AS SINAD THD	Resolution Relative Accuracy Differential Non-Lineraity Offset Error Gain Error ¹ Gain Temp Coefficient Offset Temp Coefficient Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match		10	61.5 -79.5 84 20	±0.275 ±0.275 ±2 ±2	Bits LSB LSB LSB ppm/ºC ppm/ºC LSB LSB , AIN- = dB dB dB dB	
DNL Dynamic GND (AS SINAD THD SFDR SFDR Conversi tconv tAcq	Relative Accuracy Differential Non-Lineraity Offset Error Gain Error ¹ Gain Temp Coefficient Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fin (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	e) = 10kHz, VIN = 2.5VP-P, 150ksps, fscLk		±0.25 ±0.3 ±0.025 ±0.025 (50% du 61.5 -79.5 84 20	±0.275 ±2 ±2	LSB LSB LSB ppm/ºC Dpm/ºC LSB LSB , AIN- = dB dB dB	
DNL Dynamic GND (AS SINAD THD SFDR SFDR Conversi tconv tAcq	Differential Non-Lineraity Offset Error Gain Error ¹ Gain Temp Coefficient Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fiN (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	e) = 10kHz, VIN = 2.5VP-P, 150ksps, fscLk	K = 8MHz	±0.25 ±0.3 ±0.025 ±0.025 (50% du 61.5 -79.5 84 20	±0.275 ±2 ±2	LSB LSB ppm/ºC ppm/ºC LSB LSB , AIN- = dB dB	
Dynamic GND (AS SINAD THD SFDR Conversi tconv tacq	Offset Error Gain Error ¹ Gain Temp Coefficient Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fin (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	e) = 10kHz, VIN = 2.5VP-P, 150ksps, fscLk	< = 8MHz	±0.25 ±0.3 ±0.025 ±0.025 (50% du 61.5 -79.5 84 20	±2 ±2	LSB ppm/ºC ppm/ºC LSB , AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	Gain Error ¹ Gain Temp Coefficient Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fiN (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	K = 8MHz	±0.25 ±0.3 ±0.025 ±0.025 (50% du 61.5 -79.5 84 20	±2	LSB ppm/ºC ppm/ºC LSB LSB , AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	Gain Temp Coefficient Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fiN (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	< = 8MHz	±0.3 ±0.3 ±0.025 ±0.025 (50% du 61.5 -79.5 84 20		ppm/ºC ppm/ºC LSB LSB , AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	Offset Temp Coefficient Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fin (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	< = 8MHz	±0.3 ±0.025 ±0.025 (50% du 61.5 -79.5 84 20	ty cycle)	ppm/ºC LSB LSB , AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	Channel-to-Channel Offset Match Channel-to-Channel Gain Match Specifications – (fiN (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	< = 8MHz	±0.025 ±0.025 (50% du 61.5 -79.5 84 20	ity cycle)	LSB LSB , AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	Match Channel-to-Channel Gain Match Specifications – (fin (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	< = 8MHz	±0.025 (50% du 61.5 -79.5 84 20	ty cycle)	LSB , AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	Match Specifications – (fiN (sinewave 1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	< = 8MHz	(50% du 61.5 -79.5 84 20	ity cycle)	, AIN- = dB dB dB	
GÍND (AS SINAD THD SFDR Conversi tconv tACQ	1528) Signal-to-Noise Plus Distortion Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point	< = 8MHz	61.5 -79.5 84 20	ity cycle)	dB dB dB	
THD SFDR Conversi tconv tACQ	Total Harmonic Distortion (to the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point		-79.5 84 20		dB dB	
SFDR Conversi tconv tACQ	the 5th Harmonic) Spurious-Free Dynamic Range Full Power Bandwidth	-3dB Point		84 20		dB	
Conversi tconv tACQ	Full Power Bandwidth	-3dB Point		20		-	
tconv tacq						MHz	
tconv tacq	Full Linear Bandwidth	-0.1dB Point					
tconv tacq				400		kHz	
tACQ	ion Rate		P				
	Conversion Time	Exclusive of tACQ		3.3	3.7	μs	
fsclk	Track/Hold Acquisition Time				1.4	μs	
fsclk	Aperture Delay			30		ns	
	Max Serial Clock Frequency				8	MHz	
	Serial Clock Duty Cycle		30		70	%	
Analog Ir	nput						
	VIN Range ²	Unipolar	0		VREF	V	
	VIN Range	Bipolar	-VREF/2		VREF/2	v	
	Input Leakage Current	No Channel Selected or Conversion Halted		±0.01	±1	μA	
	Input Capacitance	Track Mode		20		pF	
	input Capacitance	Hold Mode		5		pF	
External	Reference Input						
Vref	VIN Range		1.0		Vdd + 50mV	V	
		VREF = +2.5V @ 150ksps		11	25		
IREF		VREF = +4.096V @ 150ksps		19		μA	
	Input Current	VREF = +4.096V @ 150KSpS					

Symbol	Parameter	Condition	Min	Тур	Max	Unit	
Digital In	puts/Outputs (CNVST, SCLK, I	DOUT)			•		
VIL	Input Low Voltage				0.3Vdd	V	
Vін	Input High Voltage		0.7Vdd			V	
ILEAK	Input Leakage Current			±0.01	±1.0	μA	
CIN	Input Capacitance			15		pF	
Vol	Output Low Voltage	ISINK = 2mA			0.4		
VOL	Oulput Low Voltage	ISINK = 4mA			0.8	V	
Vон	Output High Voltage	ISOURCE = 1.5mA	0.7Vdd			V	
	Tri-State Leakage Current	CNVST = GND		±0.05	±5	μA	
	Tri-State Output Capacitance	CNVST = GND		15		pF	
Power R	equirements		•				
Vdd	Positive Supply Voltage		2.7		5.25	V	
		VDD = +3V, fSAMPLE = 150ksps		350	400		
		VDD = +3V, fSAMPLE = 100ksps		245			
		VDD = +3V, fsample = 10ksps		24			
		VDD = +3V, fsample = 1ksps		2.5			
IDD	Positive Supply Current	VDD = +5V, fsample = 150ksps		485	550	μA	
		VDD = +5V, fSAMPLE = 100ksps		330			
		VDD = +5V, fsample = 10ksps		33			
		VDD = +5V, fSAMPLE = 1ksps		3.7			
		Automatic Shutdown Mode		0.2	1		
PSR	Dower Supply Dejection	VDD = +5V ±5%, Full Scale Input		±0.3		m\/	
ron	Power Supply Rejection	VDD = +2.7V to 3.6V, Full Scale Input		±0.4		mV	

Table 4. Electrical Characteristics (Continued)

1. Offset nulled.

2. The absolute input voltage range for the analog inputs is from GND to VDD.

Timing Characteristics

VDD = +2.7 to +5.25V, VREF = +2.5V, $4.7\mu F$ Capacitor at REF; fSCLK = 8MHz (50% Duty Cycle); AIN- = GND (AS1528) TAMB = TMIN to TMAX (unless otherwise specified). Typical Values at TAMB = $+25^{\circ}C$.

Parameter	Symbol	Conditions	Min	Тур	Max	Units
SCLK Pulse Width High	tсн		38			ns
SCLK Pulse Width Low	tc∟		38			ns
SCLK Falling-to-DOUT Transition	tdot	CLOAD = 30pF (see Figure 3, Figure 4, Figure 18 on page 11, Figure 19 on page 11)		28	60	ns
SCLK Rising-to-DOUT ¹ Disable	tDOD	CLOAD = 30pF (see Figure 3, Figure 4, Figure 18 on page 11, Figure 19 on page 11)	100	200	500	ns
CNVST Falling-to-MSB Vlid	tCONV	CLOAD = 30pF (see Figure 3, Figure 4, Figure 18 on page 11, Figure 19 on page 11)		3.3	3.7	μs
CNVST Pulse Width	tCSW		30			ns

Table 5. Timing Characteristics

1. Guaranteed by Design and Characterisation.

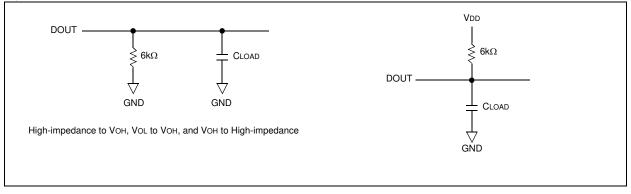
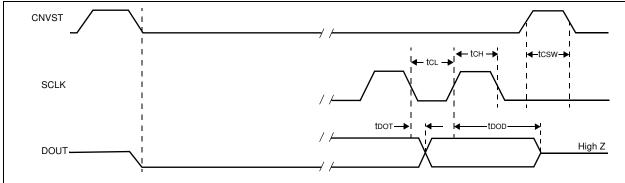


Figure 4. Detailed Serial Interface Timing Diagram



7 Typical Operating Characteristics

VDD = 5V; VREF = 2.5V, fSCLK = 8MHz(50% duty), $CREF = 4.7\mu$ F, $TAMB = +25^{\circ}C$ (unless otherwise specified). Figure 5. Integral Nonlinearity vs. Digital Output Code Figure 6. Differential Nonlinearity vs. Digital Output Code

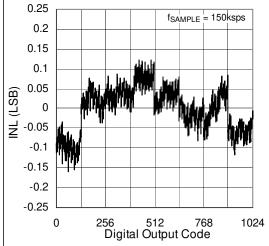
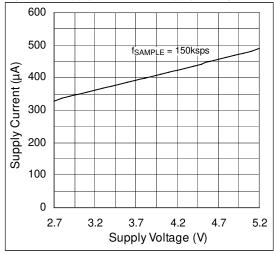
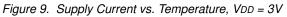
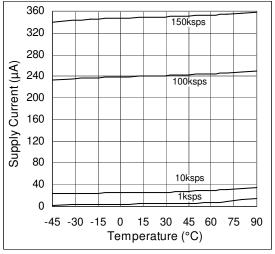


Figure 7. Supply Current vs. Supply Voltage







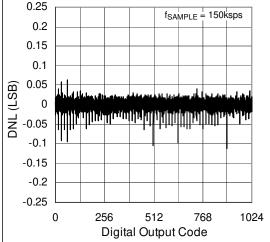


Figure 8. Supply Current vs. Temperature

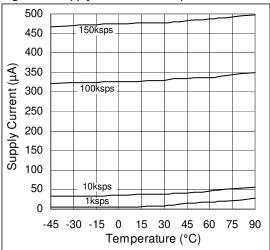
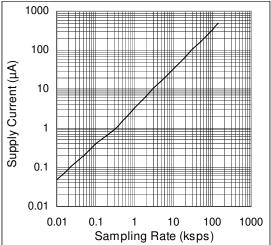
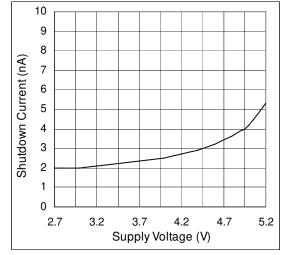


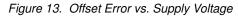
Figure 10. Supply Current vs. Sampling Rate

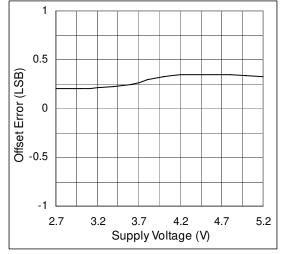


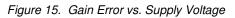
6.25

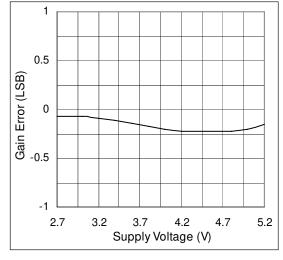
Figure 11. Shutdown Current vs. Supply Voltage

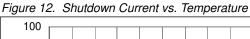












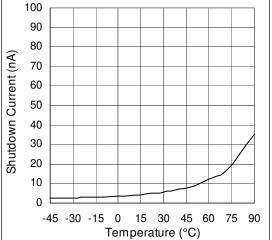


Figure 14. Offset Error vs. Temperature

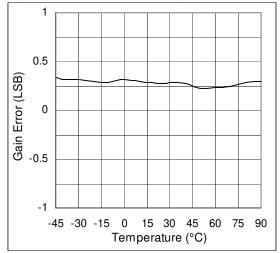
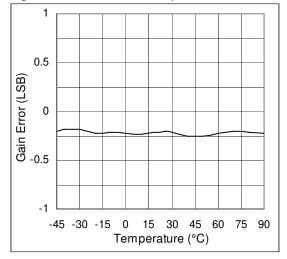


Figure 16. Gain Error vs. Temperature



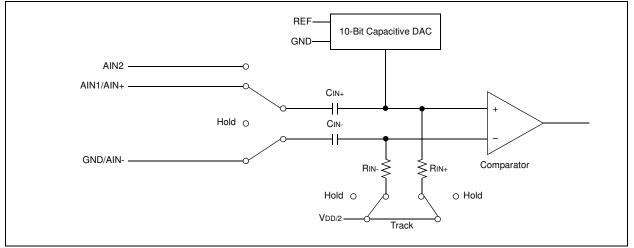
8 Detailed Description

The AS1528/AS1529 employ a successive approximation conversion (SAR) technique and integrated track/hold circuitry to convert analog signals into 10-bit digital output. The serial interface provides easy interfacing to microprocessors. Figure 17 shows the simplified internal structure for the AS1529 (2-channels, single ended) and the AS1528 (1-channel, true differential).

True Differential Analog Input Track/Hold

The equivalent circuit of Figure 17 shows the device input architecture which is composed of track/hold circuitry, input multiplexer, comparator, and switched-capacitor DAC. The track/hold circuitry enters its tracking mode on the rising edge of CNVST. The positive input capacitor is connected to AIN1 or AIN2 (AS1529) or AIN+ (AS1528). The negative input capacitor is connected to GND (AS1529) or AIN- (AS1528).





The track/hold circuitry enters its hold mode on the falling edge of CNVST and the difference between the sampled positive and negative input voltages is converted. The time required for the track/hold to acquire an input signal is determined by how quickly its input capacitance is charged. If the input signal's source impedance is high, the acquisition time lengthens, and CNVST must be held high for a longer period of time. The acquisition time (tACQ) is the maximum time needed for the signal to be acquired, plus the power-up time. tACQ is calculated by:

$$tACQ = 9 x (Rs + RiN) x 20pF + tPWR$$
(EQ 1)

Where:

Rs is the source impedance of the input signal; RIN = $1.5k\Omega$;

- tPWR of $1\mu s$ is the power-up time of the device.
- Note: tACQ is never less than 1.4μs and any source impedance below 300. does not significantly affect the AS1528/ AS1529 AC performance. A high-impedance source can be accommodated either by lengthening tACQ or by placing a 1μF capacitor between the positive and negative analog inputs.

Selecting AIN1 or AIN2 (AS1529)

Select one of the AS1529 two positive input channels using the CNVST pin (see page 3). If AIN1 is selected (see Figure 18), drive CNVST high to power up the AS1529 and place the track/hold circuitry in track mode with AIN1 connected to the positive input capacitor. Hold CNVST high for tACQ to fully acquire the signal. Drive CNVST low to place the track/hold circuitry in hold mode. The AS1529 then performs a conversion and shutdown automatically. The MSB is available at DOUT after 3.7µs. Data can then be clocked out using SCLK. Clock out all 12 bits of data before driving CNVST high for the next conversion. If all 12 bits of data are not clocked out before CNVST is driven high, AIN2 is selected for the next conversion.

, <u> </u>	Sampling Instant
SCLK	
DOUT ^{High} Z I	High Z MSB B8 B7 B6 B5 B4 B3 B2 B1 LSB ZERO ZERO

Eiguro 18 Single Conversion AINIT VG GND (AS1520) Unipolar Made AINIT VG AINIT (AS1520)

If AIN2 is selected (see Figure 19), drive CNVST high for at least 30ns. Next, drive CNVST low for at least 30ns, and then high again. This powers up the AS1529 and places the track/hold circuitry in track mode with AIN2 connected to the positive input capacitor. Next hold CNVST high for tACQ to fully acquire the signal. Drive CNVST low to place the track/hold circuitry in hold mode. The AS1529 then performs a conversion and shuts down automatically. The MSB is available at DOUT after 3.7µs. Data can then be clocked out using SCLK.

Note: If all 12 bits of data are not clocked out before CNVST is driven high, AIN2 is selected for the next conversion.

Selecting Unipolar or Bipolar Conversions (AS1528)

True-differential conversion (with the AS1528 unipolar and bipolar modes) is selected using pin CNVST (see page 3). AIN+ and AIN- are sampled at the falling edge of CNVST. In unipolar mode, AIN+ can exceed AIN- by up to VREF. The output format is straight binary. In bipolar mode, either input can exceed the other by up to VREF/2. The output format is two's complement. In both modes, the input common mode range can go from GND to VDD.

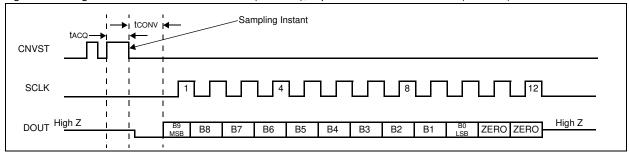


Figure 19. Single Conversion – AIN2 vs. GND (AS1529), Bipolar Mode AIN+ vs. AIN- (AS1528)

Note: In unipolar and bipolar modes, AIN+ and AIN- must not exceed VDD by more than 50mV or be lower than GND by more than 50mV.

If unipolar mode is selected (see Figure 18), drive CNVST high to power up the AS1528 and place the track/hold circuitry in track mode with AIN+ and AIN- connected to the input capacitors. Hold CNVST high for tACQ to fully acquire the signal. Drive CNVST low to place the track/hold circuitry in hold mode. The AS1528 then performs a conversion and shutdown automatically. The MSB is available at DOUT after 3.7us. Data can then be clocked out using SCLK.

Clock out all 12 bits of data before driving CNVST high for the next conversion. If all 12 bits of data are not clocked out before CNVST is driven high, bipolar mode is selected for the next conversion.

If bipolar mode is selected (see Figure 19), drive CNVST high for at least 30ns. Next, drive CNVST low for at least 30ns and then high again. This places the track/hold circuitry in track mode with AIN+ and AIN- connected to the input capacitors.

Next hold CNVST high for tACQ to fully acquire the signal. Drive CNVST low to place the track/hold circuitry in hold mode. The AS1528 then performs a conversion and shuts down automatically. The MSB is available at DOUT after 3.7µs. Data can then be clocked out using SCLK.

Note: If all 12 bits of data are not clocked out before CNVST is driven high, bipolar mode is selected for the next conversion.

Input Bandwidth

The AS1528/AS1529 input tracking circuitry has a 20MHz small signal bandwidth, so it is possible to digitize highspeed transient events and measure periodic signals with bandwidths exceeding the AS1528/AS1529 sampling rate by using undersampling techniques.

Note: To avoid high-frequency signals being aliased into the frequency band of interest, anti-alias filtering is recommended.

Analog Input Protection

Internal protection diodes that clamp the analog input to VDD and GND allow the analog input pins to swing from GND - 0.3V to VDD + 0.3V without damage. Both inputs must not exceed VDD by more than 50mV or be lower than GND by more than 50mV for accurate conversions.

Note: If an off-channel analog input voltage exceeds the supply voltages, the input current should be limited to 2mA.

Internal Clock

The AS1528/AS1529 operate from an internal clock, which is accurate within 5% of the 4MHz clock rate. This results in a worst-case conversion time of 3.7μ s. The internal clock releases the system microprocessor from running the SAR conversion clock and allows the conversion results to be read back at the processor's convenience, at any clock rate from 0 to 8MHz.

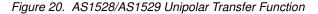
Output Data Format

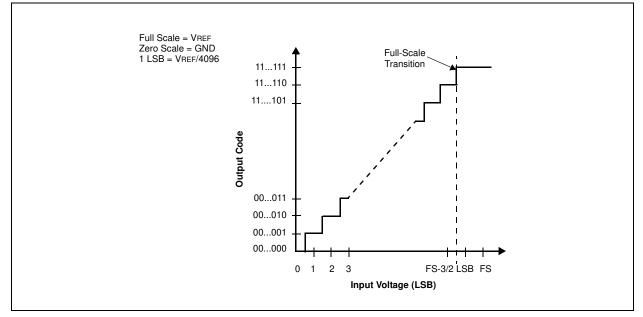
Figure 18 on page 11 and Figure 19 on page 11 illustrate the conversion timing for the AS1528/AS1529. The 10-bit conversion result is output in MSB-first format. Data on DOUT transitions on the falling edge of SCLK. All 10 bits must be clocked out before CNVST transitions again.

For the AS1528, data is straight binary for unipolar mode and two's complement for bipolar mode. For the AS1529, data is always straight binary.

Transfer Function

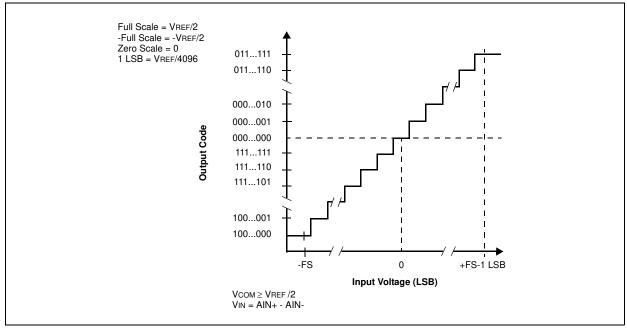
Figure 20 on page 12 shows the unipolar transfer function for the AS1528/AS1529. Figure 21 on page 13 shows the bipolar transfer function for the AS1528. Code transitions occur halfway between successive-integer LSB values.





Datasheet - Detailed Description





9 Application Information

Automatic Shutdown Mode

With CNVST low, the AS1528/AS1529 default to automatic shutdown (< 0.2μ A) mode after power-up and between conversions. After detecting a rising edge of CNVST, the AS1528/AS1529 powers up, sets DOUT low, and enters track mode.

After detecting a falling edge of CNVST, the device enters hold mode and begins the conversion. A maximum of 3.7µs later, the device completes conversion, enters shutdown, and MSB is available at DOUT.

External Reference

An external reference is required for the AS1528/AS1529. Use a 4.7μ F bypass capacitor for best performance. The reference input structure allows a voltage range of +1V to VDD + 50mV.

Performing a Conversion

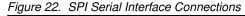
- 1. Use a general-purpose I/O line on the CPU to hold CNVST low between conversions.
- Drive CNVST high to acquire AIN1(AS1529) or unipolar mode (AS1528). To acquire AIN2 (AS1529) or bipolar mode (AS1528), drive CNVST low and high again.
- 3. Hold CNVST high for $1.4\mu s$.
- Drive CNVST low and wait approximately 3.7μs for conversion to complete. After 3.7μs, the MSB is available at DOUT.
- 5. Activate SCLK for a minimum of 12 rising clock edges. DOUT transitions on SCLK's falling edge and is available in MSB-first format. Observe the SCLK to DOUT valid timing characteristic. Clock data into the μ P on SCLK's rising edge.

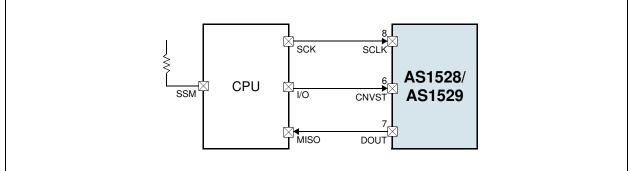
Standard Interface Connections

The AS1528/AS1529 serial interface is fully compatible with SPI, QSPI, and MICROWIRE. If a serial interface is available, establish the processor's serial interface as a master so that the CPU generates the serial clock for the AS1528/ AS1529 and select a clock frequency up to 8MHz.

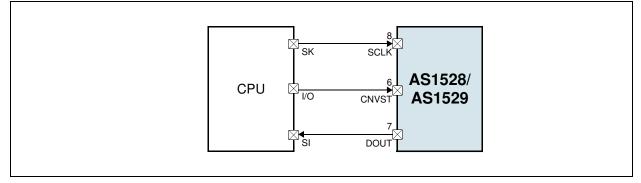
SPI and Microwire Interface

When using an SPI (Figure 22) or Microwire interface (Figure 23), set CPOL = CPHA = 0. Two 8-bit readings are necessary to obtain the entire 10-bit result from the AS1528/AS1529. DOUT data transitions on the serial clock's falling edge and is clocked into the processor on SCLK's rising edge. The first 8-bit data stream contains the first 8-bits of DOUT starting with the MSB. The second 8-bit data stream contains the remaining four result bits. DOUT then goes high impedance.

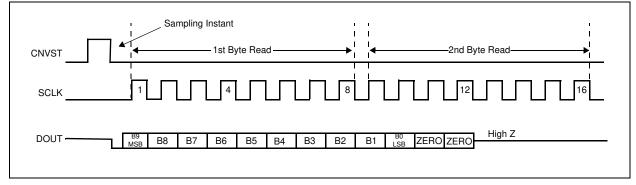










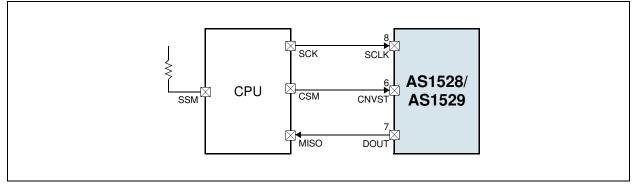


QSPI Interface

Using the high-speed QSPI interface (Figure 25) with CPOL = 0 and CPHA = 0, the AS1528/AS1529 support a maximum fSCLK of 8MHz. One 12- to 16-bit reads are necessary to obtain the entire 10-bit result from the AS1528/AS1529.

DOUT data transitions on the serial clock's falling edge and is clocked into the processor on SCLK's rising edge. The first 10 bits are the data. DOUT then goes high impedance (see Figure 23).

Figure 25. QSPI Serial Interface Connections



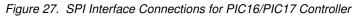
rigure 26.	QSPT Seriar Interface Triffing (CPOL = CPHA = 0)
	Sampling Instant
CNVST	
SCLK	
DOUT -	MSB B8 B7 B6 B5 B4 B3 B2 B1 LSB ZERO ZERO

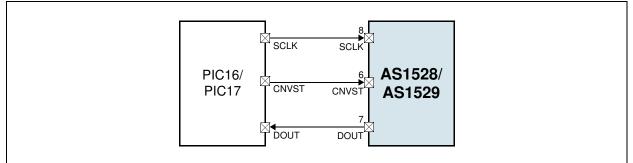
aura OC OCDI Carial Interface Timing (CDOI CDUIA

PIC16 and SSP Module and PIC17 Interface

The AS1528/AS1529 are compatible with a PIC16/PIC17 controllers, using the synchronous serial port (SSP) module

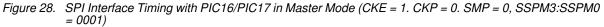
To establish SPI communication, connect the PIC16/PIC17 controllers as shown in Figure 27 and configure the PIC16/ PIC17 as system master. This is done by initializing its synchronous serial port control register (SSPCON) and synchronous serial port status register (SSPSTAT) to the bit patterns shown in Table 6 on page 17 and Table 7 on page 17.





In SPI mode, the PIC16/PIC17 processor allow 8 bits of data to be synchronously transmitted and received simultaneously. Two consecutive 8-bit readings (see Figure 28) are necessary to obtain the entire 10-bit result from the AS1528/ AS1529. DOUT data transitions on the serial clock's falling edge and is clocked into the processor on SCLK's rising edge.

The first 8-bit data stream contains the first 8 data bits starting with the MSB. The second data stream contains the remaining bits, D3 through D0.



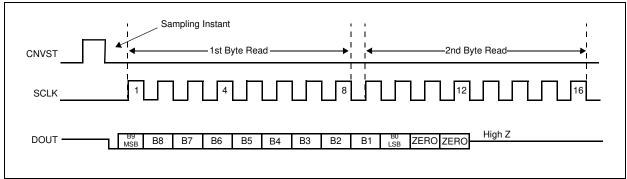




Table 6. SSPCON Register Settings

Control Bit AS1528/AS1529 Setting			Synchronous Serial Port Control Register (SSPCON)
WCOL	Bit 7	Х	Write Collision Detection Bit
SSPOV	Bit 6	Х	Receive Overflow Detect Bit
SSPEN	Bit 5	1	 Synchronous Serial Port Enable 0: Disables serial port and configures these pins as I/O port pins. 1: Enables serial port and configures SCK, SDO, and SCI pins as serial port pins.
CKP	Bit 4	0	Clock Polarity Select Bit. CKP = 0 for SPI master mode selection.
SSPM3:1	Bit 3:1	0	Synchronous Serial Port Mode Select Bit. Sets SPI master mode and
SSPM0	Bit 0	1	selects FCLK = fOSC / 16.

Table 7. SSPSTAT Register Settings

Control Bit		AS1528/AS1529 Setting	Synchronous Serial Status Register (SSPSTAT)	
SMP	Bit 7	0	SPI Data Input Sample Phase. Input data is sampled at the middle of the data output time.	
CKE	Bit 6	1	SPI Clock Edge Select Bit. Data is transmitted on the rising edge of the serial clock.	
D/A	Bit 5	Х	Data Address Bit	
Р	Bit 4	Х	Stop Bit	
S	Bit 3	Х	Start Bit	
R/W	Bit 2	Х	Read/Write Bit Information	
UA	Bit 1	Х	Update Address	
BF	Bit 0	Х	Buffer Full Status Bit	

Layout and Grounding Considerations

The AS1528/AS1529 require proper layout and design procedures for optimum performance.

- Use printed circuit boards; wirewrap boards should not be used.
- Separate analog and digital traces from each other. Analog and digital traces should not run parallel to each other (especially clock traces).
- Digital traces should not run beneath the AS1528/AS1529.
- Use a single-point analog ground at GND, separate from the digital ground (see Figure 29). Connect all other analog grounds and DGND to this star ground point for further noise reduction. No other digital system ground should be connected to this single-point analog ground. The ground return to the power supply for this ground should be low impedance and as short as possible for noise-free operation.
- High-frequency noise in the VDD power supply may affect the AS1528/AS1529 high-speed comparator. Bypass this supply to the single-point analog ground with 0.1µF and 4.7µF bypass capacitors (see Figure 29). The bypass capacitors should be placed as close to the device as possible for optimum power supply noise-rejection. If the power supply is very noisy, a 10Ω resistor can be connected as a low-pass filter to attenuate supply noise
- Power components such as the inductor, converter IC, filter capacitors, and output diode should be placed as close together as possible, and their traces should be kept short, direct, and wide.
- Keep the voltage feedback network very close to the device, within 5mm (0.2") of the pin.
- Keep noisy traces, such as those from the pin LX, away from the voltage feedback network and guarded from them using grounded copper traces.

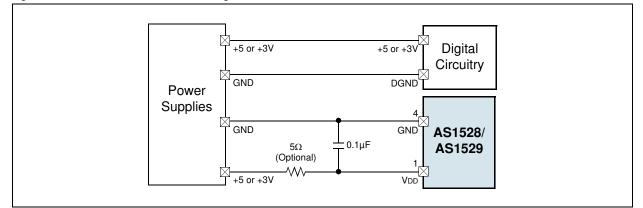


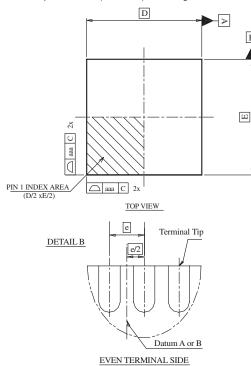
Figure 29. Recommended Ground Design

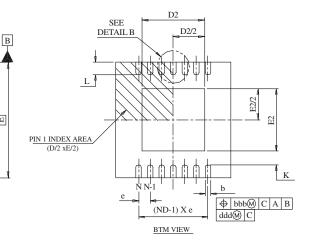


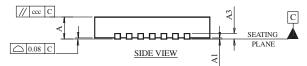
10 Package Drawings and Markings

The devices are available in a 8-pin TDFN (3x3mm) package.

Figure 30. 8-pin TDFN (3x3mm) Packagee







Symbol	Min	Тур	Max	Notes
А	0.70	0.75	0.80	1, 2
A1	0.00	0.02	0.05	1, 2
A3		0.20 REF		1, 2
L1			0.15	1, 2
L2			0.13	1, 2
aaa		0.15		1, 2
bbb		0.10		1, 2
CCC		0.10		1, 2
ddd		0.05		1, 2
eee		0.08		1, 2
ggg		0.10		1, 2

Symbol	Min	Тур	Max	Notes
D BSC		3.00		1, 2
E BSC		3.00		1, 2
D2	1.60		2.50	1, 2
E2	1.35		1.75	1, 2
L	0.30	0.40	0.50	1, 2
θ	0º		14º	1, 2
K	0.20			1, 2
b	0.25	0.30	0.35	1, 2, 5
е		0.65		
Ν		8		1, 2
ND		4		1, 2, 5

Notes:

- 1. Figure 30 is shown for illustration only.
- 2. All dimensions are in millimeters; angles in degrees.
- 3. Dimensioning and tolerancing conform to ASME Y14.5 M-1994.
- 4. N is the total number of terminals.

5. The terminal #1 identifier and terminal numbering convention shall conform to *JEDEC 95-1, SPP-012*. Details of terminal #1 identifier are optional, but must be located within the zone indicated. The terminal #1 identifier may be either a mold or marked feature.

6. Dimension b applies to metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.

- 7. ND refers to the maximum number of terminals on side D.
- 8. Unilateral coplanarity zone applies to the exposed heat sink slug as well as the terminals

11 Ordering Information

The devices are available as the standard products shown in Table 8.

Table 8. Ordering Information

Ordering Code	Description	Delivery Form	Package
AS1528-BTDT	150ksps, 10-Bit, 1-Channel True-Differential ADC	Tape & Reel	8-pin TDFN (3x3mm)
AS1528-BTDR	150ksps, 10-Bit, 1-Channel True-Differential ADC	Tray	8-pin TDFN (3x3mm)
AS1529-BTDT	150ksps, 10-Bit, 2-Channel Single-Ended ADC	Tape & Reel	8-pin TDFN (3x3mm)
AS1529-BTDR	150ksps, 10-Bit, 2-Channel Single-Ended ADC	Tray	8-pin TDFN (3x3mm)

Note: All products are RoHS compliant and Pb-free.

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