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Features

- Number of Keys:
 - Comms Mode: 1 to 12 keys (1 to 9 if wheel or slider option enabled)
 - Standalone Mode: 1 to 5 keys
- Technology:
 - Patented spread-spectrum QTouchADC charge-transfer
- Number of Output Lines:
 - Comms Mode: Up to 10 channels can be configured as outputs (but they will replace the keys)
 - Standalone Mode: 1 to 5 channels can be configured as outputs
- Key Outline Sizes:
 - 5 mm x 5 mm or larger (panel thickness dependent)
- Key Spacings:
 - 6 mm or wider, center to center (panel thickness, human factors dependent)
- Key Design:
 - Single solid or ring shaped electrodes; widely different sizes and shapes possible
- Proximity Electrode Design:
 - Single solid electrodes; Key Design, Loop, PCB Trace - different sizes and shapes possible
- Wheel Size:
 - Typically 30 mm – 50 mm diameter
- Wheel Electrode Design:
 - Spatially interpolated wheel up to 80 mm diameter
 - Typical width of segments 12 mm
- Slider Electrode Design:
 - Spatially interpolated, resistorless design
 - Typical length 50 mm – 100 mm, typical width 12 mm
 - Can be an arc or other irregular shape
- Substrates:
 - FR-4, low cost CEM-1 or FR-2 PCB materials; polyamide FPCB; PET films, glass
- Adjacent Metal:
 - Compatible with grounded metal immediately next to keys
- Layers Required:
 - One; electrodes and components can be on same side
- Electrode Materials:
 - Etched copper, silver, carbon, indium tin oxide (ITO), PEDOT
- Panel Materials:
 - Plastic, glass, composites, painted surfaces (nonconductive paints)
- Key Panel Thickness:
 - Up to 15 mm glass (key size dependent)
 - Up to 10 mm plastic (key size dependent)
- Wheel/slider Panel Thickness:
 - Up to 4 mm glass
 - Up to 3 mm plastic
- Key Sensitivity:
 - Comms Mode – individually settable via simple commands over I²C interface
 - Standalone mode – settings are fixed
- Interface:
 - I²C slave mode (400 kHz)
 - CHANGE status indication pin
- Signal Processing:
 - Self-calibration, Auto drift compensation, Noise filtering, Adjacent Key Suppression[®] (AKS[®])
- Power:
 - 1.8 V to 5.5 V
- Packages:
 - 20-pin SOIC/TSSOP RoHS compliant IC
 - 20-pad VQFN RoHS compliant IC



QTouch 12-channel Touch Sensor IC

AT42QT2120

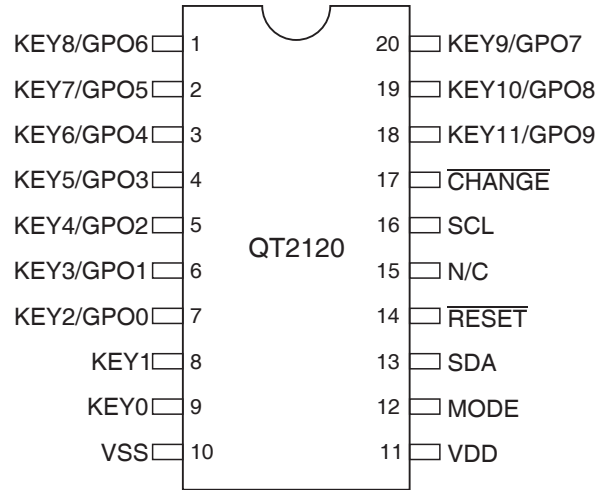
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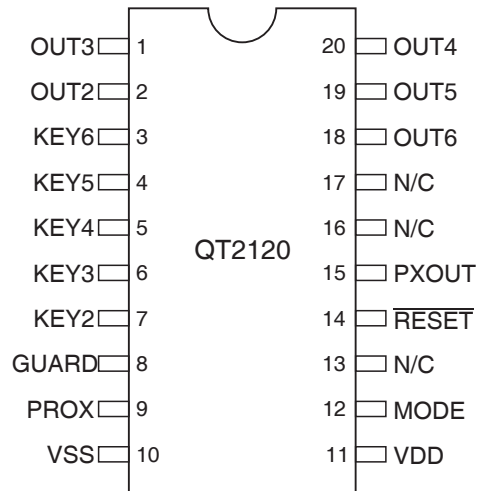
1. Pinouts and Schematics

1.1 Pinouts

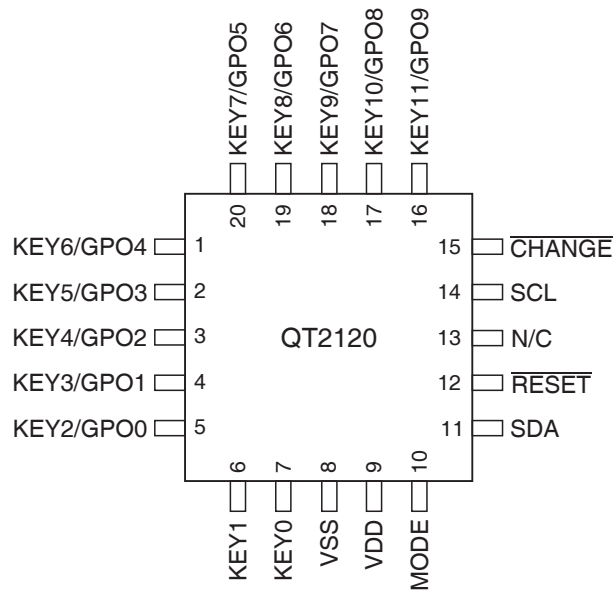
1.1.1 20-pin SOIC/TSSOP – Comms Mode



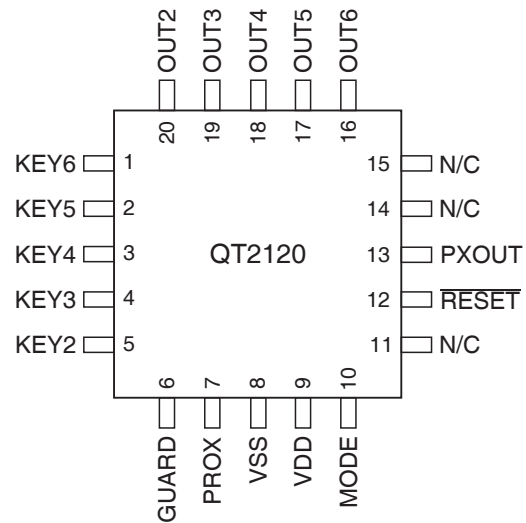
1.1.2 20-pin SOIC/TSSOP – Standalone Mode



1.1.3 20-pin VQFN – Comms Mode



1.1.4 20-pin VQFN – Standalone Mode



1.2 Pin Descriptions

1.2.1 20-pin SOIC/TSSOP

Table 1-1. Pin Listings (20-pin SOIC/TSSOP)

Pin	Name (Comms)	Name (Standalone)	Type	Description	If Unused...
1	KEY8/ GPO6	OUT3	I/O	Comms mode: Key 8 / General-purpose output Standalone mode: push-pull output for key 3	Leave open
2	KEY7/ GPO5	OUT2	I/O	Comms mode: Key 7 / General-purpose output Standalone mode: push-pull output for key 2	Leave open
3	KEY6/ GPO4	KEY6	I/O	Comms mode: Key 6 / General-purpose output Standalone mode: Key 6	Leave open
4	KEY5/ GPO3	KEY5	I/O	Comms mode: Key 5 / General-purpose output Standalone mode: Key 5	Leave open
5	KEY4/ GPO2	KEY4	I/O	Comms mode: Key 4 / General-purpose output Standalone mode: Key 4	Leave open
6	KEY3/ GPO1	KEY3	I/O	Comms mode: Key 3 / General-purpose output Standalone mode: Key 3	Leave open
7	KEY2/ GPO0	KEY2	I/O	Comms mode: Key/slider/wheel 2 / General-purpose output Standalone mode: Key 2	Leave open
8	KEY1	GUARD	I/O	Comms mode: Key/slider/wheel 1 Standalone mode: Guard channel	Leave open
9	KEY0	PROX	I/O	Comms mode: Key/slider/wheel 0 Standalone mode: Proximity channel	Leave open
10	VSS	VSS	P	Ground	–
11	VDD	VDD	P	Power	–
12	MODE	MODE	I	Mode selection pin Comms mode: connect to Vss Standalone mode: connect to Vdd	–
13	SDA	N/C	OD	Comms mode: Serial Interface Data Standalone mode: Unused	Pull up to Vdd
14	$\overline{\text{RESET}}$	$\overline{\text{RESET}}$	I	Active low reset; has internal pull-up 60 k Ω resistor	Tie to Vdd
15	N/C	PXOUT	O	Comms mode: no connection Standalone mode: open drain output for proximity channel	Leave open
16	SCL	N/C	OD	Comms mode: Serial Interface Clock Standalone mode: Unused	Pull up to Vdd
17	$\overline{\text{CHANGE}}$	N/C	OD	Comms mode: Active low state change interrupt (external pull-up resistor needed) Standalone mode: Unused	Pull up to Vdd
18	KEY11/ GPO9	OUT6	I/O	Comms mode: Key 11 / General-purpose output Standalone mode: push-pull output for key 6	Leave open
19	KEY10/ GPO8	OUT5	I/O	Comms mode: Key 10 / General-purpose output Standalone mode: push-pull output for key 5	Leave open
20	KEY9/ GPO7	OUT4	I/O	Comms mode: Key 9 / General-purpose output Standalone mode: push-pull output for key 4	Leave open

I Input only I/O Input and output OD Open drain output P Ground or power

1.2.2 20-pin VQFN

Table 1-2. Pin Listings (20-pin VQFN)

Pin	Name (Comms)	Name (Standalone)	Type	Description	If Unused...
1	KEY6/ GPO4	KEY6	I/O	Comms mode: Key 6 / General-purpose output Standalone mode: Key 6	Leave open
2	KEY5/ GPO3	KEY5	I/O	Comms mode: Key 5 / General-purpose output Standalone mode: Key 5	Leave open
3	KEY4/ GPO2	KEY4	I/O	Comms mode: Key 4 / General-purpose output Standalone mode: Key 4	Leave open
4	KEY3/ GPO1	KEY3	I/O	Comms mode: Key 3 / General-purpose output Standalone mode: Key 3	Leave open
5	KEY2/ GPO0	KEY2	I/O	Comms mode: Key/slider/wheel 2 / General-purpose output Standalone mode: Key 2	Leave open
6	KEY1	GUARD	I/O	Comms mode: Key/slider/wheel 1 Standalone mode: Guard channel	Leave open
7	KEY0	PROX	I/O	Comms mode: Key/slider/wheel 0 Standalone mode: Proximity channel	Leave open
8	VSS	VSS	P	Ground	–
9	VDD	VDD	P	Power	–
10	MODE	MODE	I	Mode selection pin Comms mode: connect to Vss Standalone mode: connect to Vdd	–
11	SDA	N/C	OD	Comms mode: Serial Interface Data Standalone mode: Unused	Pull up to Vdd
12	<u>RESET</u>	<u>RESET</u>	I	Active low reset; has internal pull-up 60 kΩ resistor	Tie to Vdd
13	N/C	PXOUT	OD	Comms mode: no connection Standalone mode: open drain output for proximity channel	Leave open
14	SCL	N/C	OD	Comms mode: Serial Interface Clock Standalone mode: Unused	Pull up to Vdd
15	<u>CHANGE</u>	N/C	OD	Comms mode: Active low state change interrupt (external pull-up resistor needed) Standalone mode: Unused	Pull up to Vdd
16	KEY11/ GPO9	OUT6	I/O	Comms mode: Key 11 / General-purpose output Standalone mode: push-pull output for key 6	Leave open
17	KEY10/ GPO8	OUT5	I/O	Comms mode: Key 10 / General-purpose output Standalone mode: push-pull output for key 5	Leave open
18	KEY9/ GPO7	OUT4	I/O	Comms mode: Key 9 / General-purpose output Standalone mode: push-pull output for key 4	Leave open
19	KEY8/ GPO6	OUT3	I/O	Comms mode: Key 8 / General-purpose output Standalone mode: push-pull output for key 3	Leave open
20	KEY7/ GPO5	OUT2	I/O	Comms mode: Key 7 / General-purpose output Standalone mode: push-pull output for key 2	Leave open

I Input only I/O Input and output OD Open drain output P Ground or power O Output only

1.3 Schematics

Figure 1-1. 20-pin SOIC/TSSOP – Comms Mode

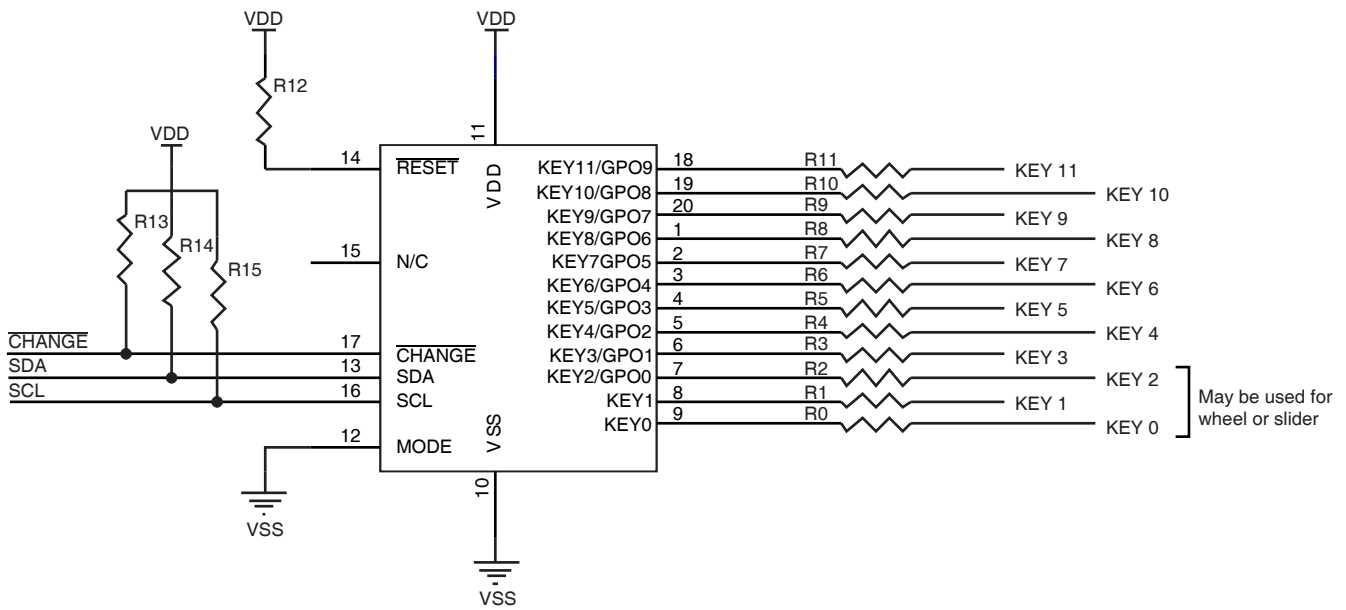


Figure 1-2. 20-pin SOIC/TSSOP – Standalone Mode

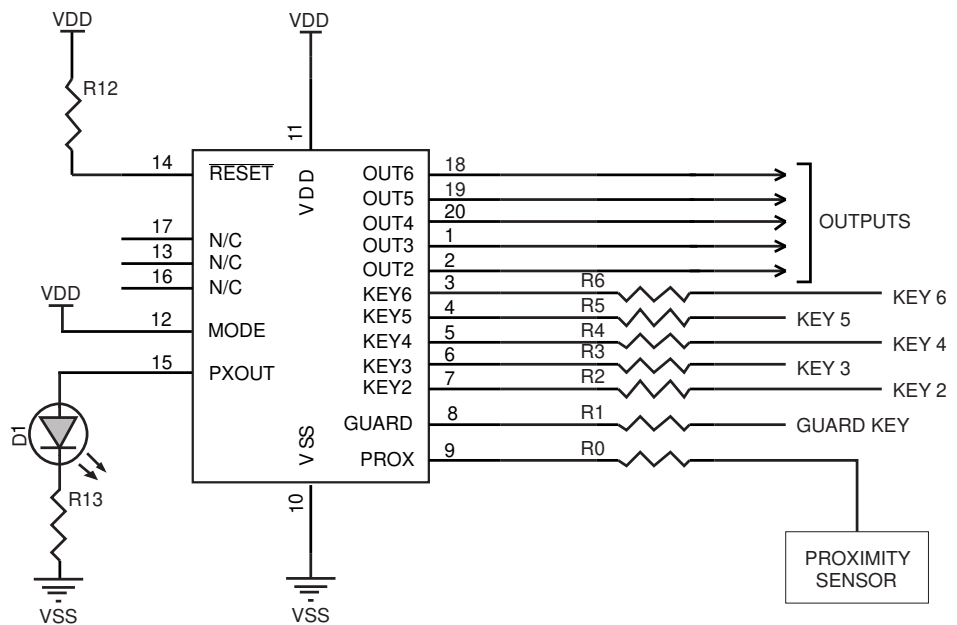


Figure 1-3. 20-pin VQFN – Comms Mode

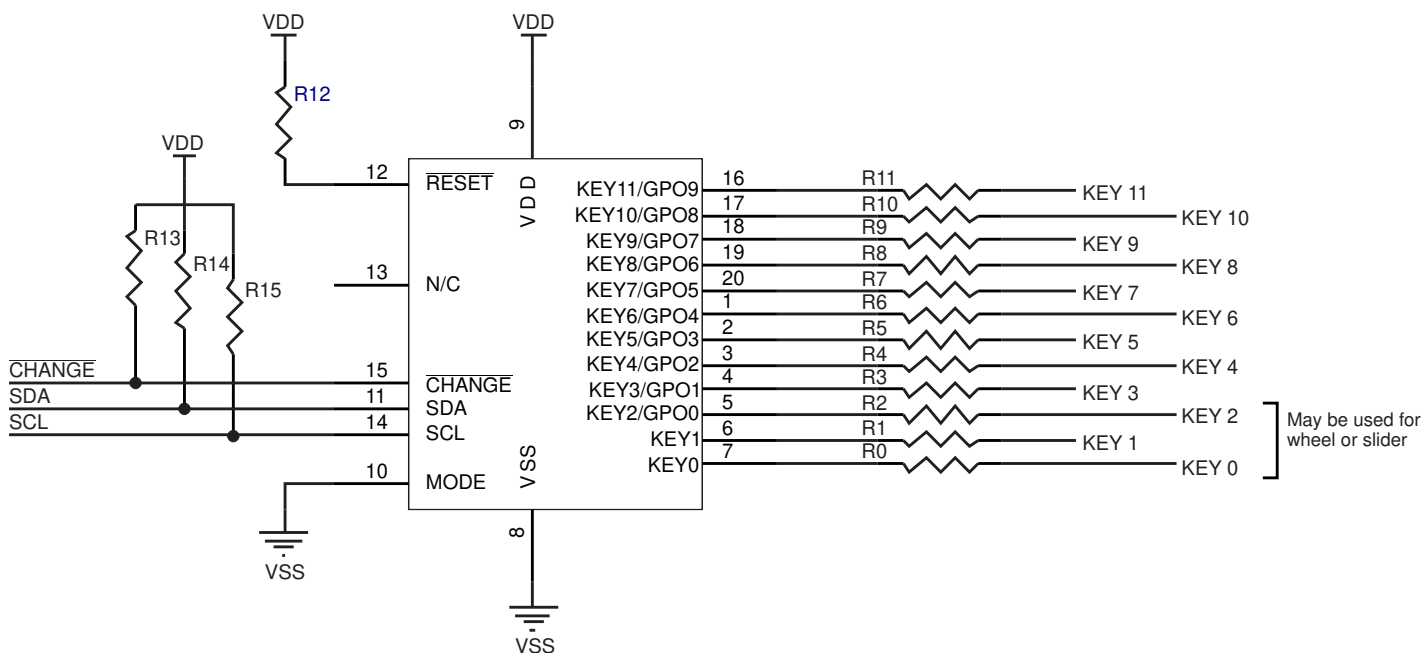
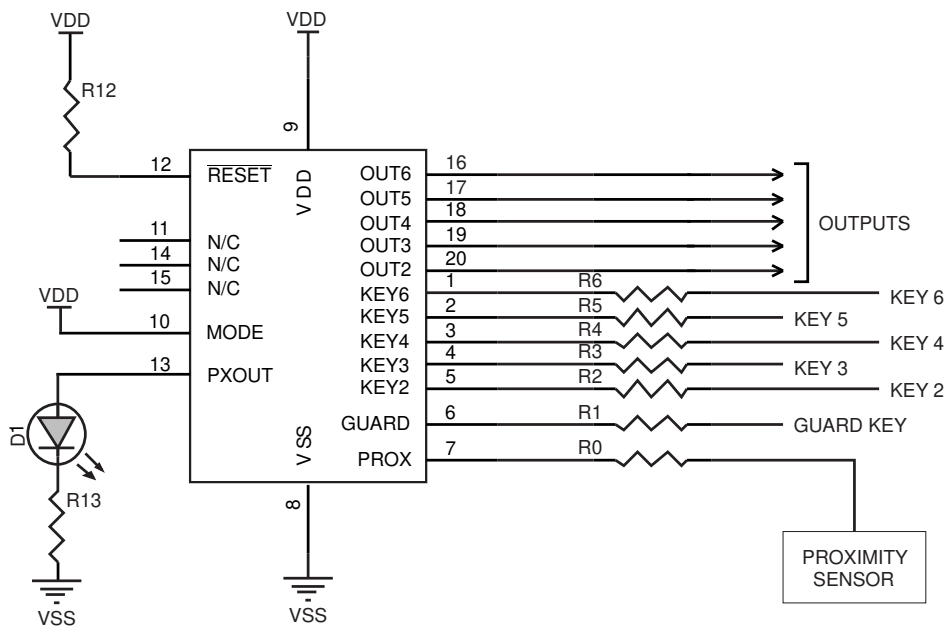


Figure 1-4. 20-pin VQFN – Standalone Mode



2. Overview

2.1 Introduction

The AT42QT2120 (QT2120) is a QTouchADC sensor driver. The device can sense from one to 12 keys, dependent on mode. Three of the keys can be used as sense channels for a slider or wheel, leaving a maximum of nine standard touch keys. The device also supports the use of proximity sensors and a guard channel.

The QT2120 includes all signal processing functions necessary to provide stable sensing under a wide variety of changing conditions, and the outputs are fully debounced. Only a few external parts are required for operation and no external Cs capacitors are required.

The QT2120 modulates its bursts in a spread-spectrum fashion in order to heavily suppress the effects of external noise, and to suppress RF emissions. The QT2120 uses a QTouchADC method of acquisition. This provides greater noise immunity and eliminates the need for external sampling capacitors, allowing touch sensing using a single pin.

The QT2120 can operate in two ways; comms and standalone.

2.2 Modes

2.2.1 Comms Mode

The QT2120 can operate in comms mode where a host can communicate with the device via an I²C bus. This allows the user to configure settings such as Threshold, Adjacent Key Suppression (AKS), Detect Integrator, Low Power (LP) Mode, Guard Channel and Max Time On for keys.

2.2.2 Standalone Mode

The QT2120 can operate in a standalone mode where an I²C-compatible interface is not required. To enter standalone mode, connect the Mode pin to Vdd before powering up the QT2120.

In standalone mode, the start-up values are hard coded in firmware and cannot be changed. The default start-up values are used. This means that key detection is reported via its respective input/output.

The Guard channel feature is automatically implemented on key 1 in standalone mode. This means that this channel has a higher sensitivity and is used to protect against false triggering, perhaps by a hand covering all keys.

A proximity sensor is also available on channel (key) 0 in standalone mode.

2.3 Keys

Dependent on mode, the QT2120 can have a minimum of one key and a maximum of 12 keys. These can be constructed in different shapes and sizes. See “Features” on page 1 for the recommended dimensions.

The possible combinations of keys are:

Comms mode:

- 1 to 12 keys
or
- 1 to 9 keys plus 1 slider/wheel
- Key channels 2 to 11 can be reassigned as general outputs, if required

Note: Any number of keys can be configured as proximity channels.

Standalone mode:

- 1 to 5 keys plus corresponding discrete outputs
- 1 Guard Channel
- 1 Proximity sensor

Unused keys should be disabled by setting bit 0 of their control bytes to 1 (see [Section 5.17 on page 26](#)).

The Key Status register (see [Section 5.5 on page 21](#)) can be read to determine the touch status of the corresponding key. It is recommended using the open-drain $\overline{\text{CHANGE}}$ line to detect when a change of status has occurred.

2.4 Output Lines

In comms mode some pins, normally used for touch keys, can be used as output pins. If the Key Control bit 0 (EN) is set to 1, the pin can be used as an output. The state of the pin is then controlled by Key Control bit 1 (GPO).

In Standalone mode pins OUT2 to OUT6 are driven by KEY2 to KEY6 respectively. The OUT pins drive high during touch and can be used to drive, for example, LEDs.

2.5 Acquisition/Low Power Mode (LP)

There are 255 different acquisition times possible. These are controlled via the LP Mode byte (see [Section 5.9 on page 22](#)) which can be written to via I²C-compatible communication.

LP mode controls the intervals between acquisition measurements. Longer intervals consume lower power but have an increased response time. During calibration, touch and during the detect integrator (DI) period, the LP mode is temporarily set to LP mode 1 for a faster response.

The QT2120 operation is based on a fixed cycle time of approximately 16 ms. The LP mode setting indicates how many of these periods exist per measurement cycle. For example, If LP mode = 1, there is an acquisition every cycle (16 ms). If LP mode = 3, there is an acquisition every 3 cycles (48 ms). If a high PULSE setting is selected then the acquisition time may exceed 16 ms.

An LP setting of 0 will send the device into Power-down mode. To wake the device from this mode a nonzero LP setting should be written to the LP address at location 8.

2.6 Adjacent Key Suppression (AKS) Technology

The device includes the Atmel patented Adjacent Key Suppression (AKS) technology, to allow the use of tightly spaced keys on a keypad with no loss of selectability by the user.

There can be up to three AKS groups, implemented so that only one key in the group may be reported as being touched at any one time. Once a key in a particular AKS group is in detect no other key in that group can go into detect. Only when the key in detect goes out of detection can another key go into detect state.

Keys which are members of the AKS groups can be set in the Key Control register (see [Section 5.17 on page 26](#)). Keys outside the group may be in detect simultaneously.

Note: To use a key as a guard channel, its AKS group should be set to be the same as that of the keys it is to protect.

2.7 $\overline{\text{CHANGE}}$ Line (Comms Mode Only)

The $\overline{\text{CHANGE}}$ line is active low and signals when there is a change of state in the Detection Status and/or Key Status bytes. It is cleared (allowed to float high) when the host reads the status bytes.

If the status bytes change back to their original state before the host has read the status bytes (for example, a touch followed by a release), the $\overline{\text{CHANGE}}$ line will be held low. In this case, a read to any memory location will clear the $\overline{\text{CHANGE}}$ line.

The $\overline{\text{CHANGE}}$ line is open-drain and should be connected via a 47 k Ω resistor to Vdd. It is necessary for minimum power operation as it ensures that the QT2120 can sleep for as long as possible. Communications wake up the QT2120 from sleep causing a higher power consumption if the part is randomly polled.

Note that the $\overline{\text{CHANGE}}$ line is pulled low 85 ms after power-up or reset. The $\overline{\text{CHANGE}}$ line is pulled low approximately for another 16 ms before any bursting on the touch pins will occur. If any of the pins are required to be outputs then the relevant Key Control settings should be written within this 16 ms time to prevent bursting on pins required as outputs. Also note that the $\overline{\text{CHANGE}}$ line is cleared during a read of the Detection Status bytes when all bytes differing from the previous read have been read.

2.8 Types of Reset

2.8.1 External Reset

An external reset logic line can be used if desired, fed into the $\overline{\text{RESET}}$ pin. However, under most conditions it is acceptable to tie $\overline{\text{RESET}}$ to Vdd. The minimum reset pulse width is 2 μs .

2.8.2 Soft Reset

The host can cause a device reset by writing a nonzero value to the Reset byte. This soft reset triggers the internal watchdog timer on a 125 ms interval. After 125 ms the device executes a full reset.

The device NACKs any attempts to communicate with it for approximately 200 ms after the soft reset command. Communication can begin as soon the $\overline{\text{CHANGE}}$ line is first asserted.

Note: The device can process a Soft Reset command while in Power Down (LPM = 0) mode, causing a chip reset.

2.9 Calibration

Writing a nonzero value to the calibration byte can force a recalibration at any time. This can be useful to clear out a stuck key condition after a prolonged period of uninterrupted detection. A calibration command executes 15 burst cycles at LPM 1 and sets the CALIBRATE bit of the Detection Status register during the calibration sequence.

Note: A calibration command should be sent whenever Key Control bit 0 (EN) is changed. This changes the use of the key from a standard touch key to an output pin and vice-versa.

2.10 Guard Channel

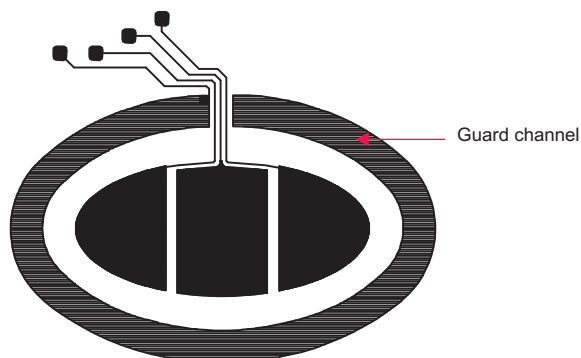
A guard channel to help prevent false detection is available in both modes. This is fixed on key 1 for standalone mode and programmable for comms mode by setting Key Control bit 4 (GUARD) (see [Section 5.17 on page 26](#)).

Guard channel keys should be more sensitive than the other keys and physically bigger. Because the guard channel key is physically bigger it becomes more susceptible to noise so it should have a higher Oversampling (see [Section 5.18 on page 27](#)) than the other keys. In standalone mode it is assigned to key 1 and cannot be changed.

In comms mode any key can be selected to be a guard key by setting Key Control bit 4 (GUARD).

The guard channel is connected to a sensor pad which detects the presence of touch. Because of its larger size and sensitivity it goes into touch before the keys it surrounds (if, for example, a hand covers all the keys).

Figure 2-1. Guard Channel Example



2.11 Signal Processing

2.11.1 Detect Threshold

The device detects a touch when the signal has crossed a threshold level and remained there for a specified number of counts (see [Section 5.11 on page 24](#)). This can be altered on a key-by-key basis using the key Detect Threshold I²C-compatible commands.

This detect threshold is based on the reference value of the particular key. The delta of the key is obtained by subtracting the reference value from the signal value (the signal value rises when touch is present).

In standalone mode the detect threshold is set to a fixed value of 10 counts of change with respect to the internal reference level for the guard channel and 10 counts for the other six keys (including proximity channel). The reference level has the ability to adjust itself slowly in accordance with the drift compensation mechanism.

The drift mechanism will drift toward touch at a rate of $160 \text{ ms} \times 20 = 3.2$ seconds (Towards Touch Drift (TTD) register) and away from touch at a rate of $160 \text{ ms} \times 5 = 0.8$ seconds (Away from Touch Drift (ATD) register). These values are fixed in standalone mode but can be configured in comms mode see [Section 5.10 on page 22](#).

2.11.2 Detect Integrator

The device features a fast detection integrator counter (DI filter), which acts to filter out noise at the small expense of a slower response time. The DI filter requires a programmable number of consecutive samples confirmed in detection before the key is declared to be touched. The minimum number for the DI filter is 1. A setting of 0 for the DI also defaults to 1. The DI has a maximum usable value of 32. Values above this will prevent a key from entering touch.

The DI is also implemented when a touch is removed.

2.11.3 Cx Limitations

The recommended range for key capacitance C_x is $1 \text{ pF} - 30 \text{ pF}$. Larger values of C_x will give reduced sensitivity.

2.11.4 Touch Recalibration Delay

If an object or material obstructs the sense pad the signal may rise enough to create a detection, preventing further operation. To prevent this, the sensor includes a timer which monitors detections. If a detection exceeds the timer setting the sensor performs a key recalibration. This is known as the Touch Recalibration Delay (TRD) and is set to approximately 30 s in standalone mode.

In comms mode this feature can be changed by setting a value in the range $1 - 255$ ($160 \text{ ms} - 40,800 \text{ ms}$) in steps of 160 ms. A setting of 0 disables the TRD.

TRD is a global setting and applies to all keys.

2.11.5 Away from Touch Recalibration

If a keys signal jumps in the negative direction (with respect to its reference) by more than the Away from Touch Recalibration setting (25% of detect threshold), then a recalibration of that key takes place.

Note: The minimum Away from Touch Recalibration is hard limited to 4 counts.

2.11.6 Drift Hold Time

Drift Hold Time (DHT) is used to restrict drift on all keys while one or more keys are activated. DHT restricts the drifting on all keys until approximately four seconds after all touches have been removed.

This feature is particularly useful in cases of high-density keypads where touching a key or hovering a finger over the keypad would cause untouched keys to drift, and therefore create a sensitivity shift, and ultimately inhibit touch detection. In Comms mode this value is settable, see [Section 5.13 on page 24](#).

The QT2120 will remain in fast mode ($LP = 1$) for the duration of the DHT counter. The total DHT time is $160 \text{ ms} \times \text{DHT value}$. The default setting for DHT is 25, so $160 \text{ ms} \times 25 = \sim 4 \text{ seconds}$. The QT2120 will not drift or re-enter slow LP mode during this time.

2.11.7 Hysteresis

Hysteresis is fixed at 12.5% of the Detect Threshold. When a key enters a detect state once the DI count has been reached, the Detect threshold (DTHR) value is changed by a small amount (12.5% of DTHR) in the direction away from touch. This is done to effect hysteresis and so makes it less likely a key will dither in and out of detect. DTHR is restored once the key drops out of detect.

Note: The minimum value for hysteresis is 2 counts.

3. Wiring and Parts

3.1 Rs Resistors

Series resistors R_s ($R_{s0} - R_{s11}$ for comms mode and $R_{s0} - R_{s6}$ for standalone mode) are in line with the electrode connections and should be used to limit electrostatic discharge (ESD) currents and to suppress radio frequency interference (RFI). Series resistors are recommended for noise reduction. They should be approximately $4.7 \text{ k}\Omega$ to $20 \text{ k}\Omega$ each. For maximum noise rejection the value may be up to $100 \text{ k}\Omega$. Care should be taken in this case that the sensor keys are fully charged. The Charge Time may need to be increased (see [Section 5.15 on page 26](#)). Each count increase will extend the charge pulse by approximately $1 \mu\text{s}$.

3.2 LED Traces and Other Switching Signals

Digital switching signals near the sense lines induce transients into the acquired signals, deteriorating the signal-to-noise (SNR) performance of the device. Such signals should be routed away from the sensing traces and electrodes, or the design should be such that these lines are not switched during the course of signal acquisition (bursts).

LED terminals which are multiplexed or switched into a floating state, and which are within, or physically very near, a key (even if on another nearby PCB) should be bypassed to either V_{ss} or V_{dd} with at least a 10 nF capacitor. This is to suppress capacitive coupling effects which can induce false signal shifts. The bypass capacitor does not need to be next to the LED, in fact it can be quite distant. The bypass capacitor is noncritical and can be of any type.

LED terminals which are constantly connected to V_{ss} or V_{dd} do not need further bypassing.

3.3 PCB Cleanliness

Modern no-clean flux is generally compatible with capacitive sensing circuits.



CAUTION: If a PCB is reworked to correct soldering faults relating to the device, or to any associated traces or components, be sure that you fully understand the nature of the flux used during the rework process. Leakage currents from hygroscopic ionic residues can stop capacitive sensors from functioning. If you have any doubts, a thorough cleaning after rework may be the only safe option.

If a PCB is reworked in any way, clean it thoroughly to remove all traces of the flux residue around the capacitive sensor components. Dry it thoroughly before any further testing is conducted.

3.4 Power Supply

See [Section 6.2 on page 30](#) for the power supply range. If the power supply fluctuates slowly with temperature, the device tracks and compensates for these changes automatically with only minor changes in sensitivity. If the supply voltage drifts or shifts quickly, the drift compensation mechanism is not able to keep up, causing sensitivity anomalies or false detections.

The power should be clean and come from a separate regulator if possible. However, this device is designed to minimize the effects of unstable power, and except in extreme conditions should not require a separate Low Dropout (LDO) regulator.



CAUTION: A regulator IC shared with other logic can result in erratic operation and is **not** advised.

A single ceramic 0.1 μF bypass capacitor, with short traces, should be placed very close to the power pins of the IC. Failure to do so can result in device oscillation, high current consumption and erratic operation.

It is assumed that a larger bypass capacitor (like 1 μF) is somewhere else in the power circuit; for example, near the regulator.

4. I²C-compatible Communications (Comms Mode Only)

4.1 I²C-compatible Protocol

4.1.1 Protocol

The I²C-compatible protocol is based around access to an address table (see [Table 5-1 on page 18](#)) and supports multibyte reads and writes. The maximum clock rate is 400 kHz.

See [Section A on page 41](#) for an overview of I²C bus operation.

4.1.2 Signals

The I²C-compatible interface requires two signals to operate:

- **SDA** – Serial Data
- **SCL** – Serial Clock

A third line, $\overline{\text{CHANGE}}$, is used to signal when the device has seen a change in the status byte:

- **$\overline{\text{CHANGE}}$** : Open-drain, active low when the device status has changed since the last I²C read. After reading the four status bytes ⁽¹⁾ (or all the status bytes which have changed since the previous read), this pin floats (high) again if it is pulled up with an external resistor. If the status bytes change back to their original state before the host has read the status bytes (for example, a touch followed by a release), the $\overline{\text{CHANGE}}$ line is held low. In this case, a read to any memory location clears the $\overline{\text{CHANGE}}$ line.

4.2 I²C-compatible Address

There is one preset I²C-compatible address of 0x1C (28). This is not changeable.

4.3 Data Read/Write

4.3.1 Address Pointer

The internal address pointer is initialized to address 0.

4.3.2 Writing Data to the Device

The sequence of events required to write data to the device is shown next.

Host to Device Device Tx to Host



Table 4-1. Description of Write Data Bits

Key	Description
S	START condition
SLA+W	Slave address plus write bit
A	Acknowledge bit
MemAddress	Target memory address within device
Data	Data to be written
P	Stop condition

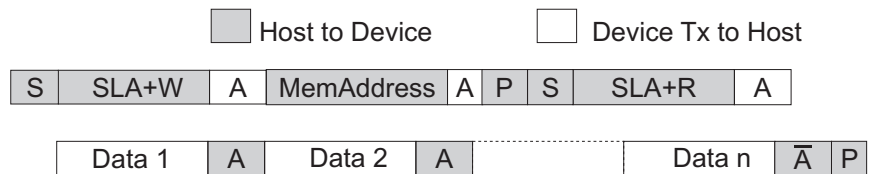
1. Detection Status Byte, Key Status Byte[0], Key Status Byte[1], Slider Position

1. The host initiates the transfer by sending the START condition
2. The host follows this by sending the slave address of the device together with the WRITE bit.
3. The device sends an ACK.
4. The host then sends the memory address within the device it wishes to write to.
5. The device sends an ACK.
6. The host transmits one or more data bytes; each is acknowledged by the device (unless trying to write to an invalid address). Valid write address are 5 – 51.
7. If the host sends more than one data byte, they are written to consecutive memory addresses.
8. The device automatically increments the target memory address after writing each data byte.
9. After writing the last data byte, the host should send the STOP condition.

Note: the host should not try to write to addresses outside the range 0x06 to 0x33 (6 – 51) because this is the limit of the device’s internal memory address.

4.3.3 Reading Data From the Device

The sequence of events required to read data from the device is shown next.



1. The host initiates the transfer by sending the START condition
2. The host follows this by sending the slave address of the device together with the WRITE bit.
3. The device sends an ACK.
4. The host then sends the memory address within the device it wishes to read from.
5. The device sends an ACK if the address to be read from is less than 0x63 otherwise it sends a NACK).
6. The host must then send a STOP and a START condition followed by the slave address again but this time accompanied by the READ bit.

Note: Alternatively, instead of step 6, a repeated START can be sent so the host does not need to relinquish control of the bus.

7. The device returns an ACK, followed by a data byte.
8. The host must return either an ACK or NACK.
 - a. If the host returns an ACK, the device subsequently transmits the data byte from the next address. Each time a data byte is transmitted, the device automatically increments the internal address. The device continues to return data bytes until the host responds with a NACK.
 - b. If the host returns a NACK, it should then terminate the transfer by issuing the STOP condition.
9. The device resets the internal address to the location indicated by the memory address sent to it previously. Therefore, there is no need to send the memory address again when reading from the same location.

Note: Reading the 16-bit reference and signal values is not an automatic operation; reading the first byte of a 16-bit value does not lock the other byte. As a result glitches in the reported value may be seen as values increase from 255 to 256, or decrease from 256 to 255. This device also supports the use of a repeated START condition as an alternative to the Stop condition.

4.4 SDA, SCL

The I²C-compatible bus transmits data and clock with SDA and SCL respectively. They are open-drain; that is I²C-compatible master and slave devices can only drive these lines low or leave them open. The termination resistors pull the line up to Vdd if no I²C-compatible device is pulling it down.

The termination resistors commonly range from 1 k Ω to 10 k Ω and should be chosen so that the rise times on SDA and SCL meet the I²C-compatible specifications (300 ns maximum for 400 kHz operation).

4.5 Standalone Mode

If I²C-compatible communications are not required, then standalone mode can be enabled by connecting the MODE pin to Vdd. See [Section 2.4 on page 9](#) for more information.

In Standalone mode (Mode pin connected to Vdd at start-up) the chip is configured to specific settings:

- Key0 is configured as a proximity channel. If this key goes into detect then PXOUT is asserted high.
- Key1 is configured as a guard channel and should have a PCB layout which reflects this.
- Keys 2 – 6 are standard QTouchADC keys and have pins OUT 2 – 6 configured to reflect their respective touch status.
- Keys 1 – 6 are configured to have the same AKS group setting.

5. Setups

5.1 Introduction

The device calibrates and processes signals using a number of algorithms specifically designed to provide for high survivability in the face of adverse environmental challenges. User-defined Setups are employed to alter these algorithms to suit each application. These Setups are loaded into the device over the I²C-compatible serial interfaces. In standalone mode these settings are fixed to predetermined values.

Table 5-1. Internal Register Address Allocation

Address	Use	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	R/W
0	Chip ID	Chip Id = 0x3E (62)								R
1	Firmware Version	Major version				Minor version				R
2	Detection Status	CALIBRATE	OVERFLOW	–	–	–	–	SDET	TDET	R
3	Key Status	KEY7	KEY6	KEY5	KEY4	KEY3	KEY2	KEY1	KEY0	R
4		Reserved				KEY11	KEY10	KEY9	KEY8	R
5	Slider Position	Slider position								R
6	Calibrate	Calibrate Command								R/W
7	Reset	Reset Command								R/W
8	LP	Low Power (LP) Mode								R/W
9	TTD	0	Towards Touch Drift compensation							R/W
10	ATD	0	Away from Touch Drift compensation							R/W
11	DI	Detection integrator								R/W
12	TRD	Touch Recal Delay								R/W
13	DHT	Drift Hold Time								R/W
14	Slider Options	EN	WHEEL	Reserved						R/W
15	Charge Time	Reserved				Charge Time				–
16	Key 0 Detect Threshold	Detect Threshold level for key 0								R/W
17	Key 1 Detect Threshold	Detect Threshold level for key 1								R/W
18	Key 2 Detect Threshold	Detect Threshold level for key 2								R/W
19	Key 3 Detect Threshold	Detect Threshold level for key 3								R/W
20	Key 4 Detect Threshold	Detect Threshold level for key 4								R/W
21	Key 5 Detect Threshold	Detect Threshold level for key 5								R/W
22	Key 6 Detect Threshold	Detect Threshold level for key 6								R/W
23	Key 7 Detect Threshold	Detect Threshold level for key 7								R/W
24	Key 8 Detect Threshold	Detect Threshold level for key 8								R/W
25	Key 9 Detect Threshold	Detect Threshold level for key 9								R/W
26	Key 10 Detect Threshold	Detect Threshold level for key 10								R/W
27	Key 11 Detect Threshold	Detect Threshold level for key 11								R/W
28	Key 0 Control	Reserved			GUARD	AKS		GPO	EN	R/W
29	Key 1 Control	Reserved			GUARD	AKS		GPO	EN	R/W
30	Key 2 Control	Reserved			GUARD	AKS		GPO	EN	R/W

Table 5-1. Internal Register Address Allocation (Continued)

Address	Use	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	R/W
31	Key 3 Control	Reserved			GUARD	AKS		GPO	EN	R/W
32	Key 4 Control	Reserved			GUARD	AKS		GPO	EN	R/W
33	Key 5 Control	Reserved			GUARD	AKS		GPO	EN	R/W
34	Key 6 Control	Reserved			GUARD	AKS		GPO	EN	R/W
35	Key 7 Control	Reserved			GUARD	AKS		GPO	EN	R/W
36	Key 8 Control	Reserved			GUARD	AKS		GPO	EN	R/W
37	Key 9 Control	Reserved			GUARD	AKS		GPO	EN	R/W
38	Key 10 Control	Reserved			GUARD	AKS		GPO	EN	R/W
39	Key 11 Control	Reserved			GUARD	AKS		GPO	EN	R/W
40	Key 0 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
41	Key 1 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
42	Key 2 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
43	Key 3 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
44	Key 4 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
45	Key 5 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
46	Key 6 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
47	Key 7 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
48	Key 8 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
49	Key 9 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
50	Key 10 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
51	Key 11 Pulse Scale	PULSE3	PULSE2	PULSE1	PULSE0	SCALE3	SCALE2	SCALE1	SCALE0	R/W
52–53	Key Signal 0	Key signal 0 (MSByte) – Key signal 0 (LSByte)								R
54–55	Key Signal 1	Key signal 1 (MSByte) – Key signal 1 (LSByte)								R
56–57	Key Signal 2	Key signal 2 (MSByte) – Key signal 2 (LSByte)								R
58–59	Key Signal 3	Key signal 3 (MSByte) – Key signal 3 (LSByte)								R
60–61	Key Signal 4	Key signal 4 (MSByte) – Key signal 4 (LSByte)								R
62–63	Key Signal 5	Key signal 5 (MSByte) – Key signal 5 (LSByte)								R
64–65	Key Signal 6	Key signal 6 (MSByte) – Key signal 6 (LSByte)								R
66–67	Key Signal 7	Key signal 7 (MSByte) – Key signal 7 (LSByte)								R
68–69	Key Signal 8	Key signal 8 (MSByte) – Key signal 8 (LSByte)								R
70–71	Key Signal 9	Key signal 9 (MSByte) – Key signal 9 (LSByte)								R
72–73	Key Signal 10	Key signal 10 (MSByte) – Key signal 10 (LSByte)								R
74–75	Key Signal 11	Key signal 11 (MSByte) – Key signal 11 (LSByte)								R
76–77	Reference Data 0	Reference data 0 (MSByte) – Reference data 0 (LSByte)								R
78–79	Reference Data 1	Reference data 1 (MSByte) – Reference data 1 (LSByte)								R
80–81	Reference Data 2	Reference data 2 (MSByte) – Reference data 2 (LSByte)								R
82–83	Reference Data 3	Reference data 3 (MSByte) – Reference data 3 (LSByte)								R
84–85	Reference Data 4	Reference data 4 (MSByte) – Reference data 4 (LSByte)								R

Table 5-1. Internal Register Address Allocation (Continued)

Address	Use	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	R/W
86–87	Reference Data 5	Reference data 5 (MSByte) – Reference data 5 (LSByte)								R
88–89	Reference Data 6	Reference data 6 (MSByte) – Reference data 6 (LSByte)								R
90–91	Reference Data 7	Reference data 7 (MSByte) – Reference data 7 (LSByte)								R
92–93	Reference Data 8	Reference data 8 (MSByte) – Reference data 8 (LSByte)								R
94–95	Reference Data 9	Reference data 9 (MSByte) – Reference data 9 (LSByte)								R
96–97	Reference Data 10	Reference data 10 (MSByte) – Reference data 10 (LSByte)								R
98–99	Reference Data 11	Reference data 11 (MSByte) – Reference data 11 (LSByte)								R

5.2 Address 0: Chip ID

Table 5-2. Chip ID

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	CHIP ID							

CHIP ID: Holds the chip ID; always 0x3E.

5.3 Address 1: Firmware Version

Table 5-3. Firmware Version

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
1	MAJOR VERSION				MINOR VERSION			

MAJOR VERSION: Holds the major firmware version (for example revision 1.5).

MINOR VERSION: Holds the minor firmware version (for example revision 1.5).

5.4 Address 2: Detection Status

Table 5-4. Detection Status

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
2	CALIBRATE	OVERFLOW	–	–	–	–	SDET	TDET

CALIBRATE: This bit is set during a calibration sequence.

OVERFLOW: This bit is set if the time to acquire all key signals exceeds 16 ms.

SDET: This bit is set if any of the slider/wheel channels are in detect.

TDET: This bit is set if any of the keys are in detect.

Note: If the slider or wheel is enabled then the SDET bit will be set when it is in detect. Also the relevant Key Status bit (0 – 2) and TDET will be set. These bits can be ignored if the SDET bit is set as the slider/wheel takes priority.

A change in these bytes will cause the $\overline{\text{CHANGE}}$ line to trigger.

5.5 Addresses 3 – 4: Key Status

Table 5-5. Key Status

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
3	KEY7	KEY6	KEY5	KEY4	KEY3	KEY2	KEY1	KEY0
4	Reserved				KEY11	KEY10	KEY9	KEY8

KEY0 – KEY11: These bits indicate which keys are in detection, if any. Touched keys report as 1, untouched or disabled keys report as 0. A change in these bytes will cause the CHANGE line to trigger.

5.6 Address 5: Slider Position

Table 5-6. Slider Position

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
5	SLIDER POSITION							

SLIDER POSITION: Reports the slider/wheel position. This value is only valid when the SDET bit in the Detection Status byte is set. A change in this value will cause the CHANGE line to assert low.

5.7 Address 6: Calibrate

Table 5-7. Calibrate

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
6	CALIBRATE COMMAND							

CALIBRATE COMMAND: Writing any nonzero value into this address triggers the device to start a calibration cycle. The CALIBRATE flag in the detection status register is set when the calibration begins and clears when the calibration has finished.

5.8 Address 7: Reset

Table 5-8. Reset

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
7	RESET COMMAND							

RESET COMMAND: Writing any nonzero value to this address triggers the device to reset.

5.9 Address 8: Low Power (LP) Mode

Table 5-9. LP Mode

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
8	LP MODE							

LP MODE: This 8-bit value determines the number of 16 ms intervals between key measurements. Longer intervals between measurements yield a lower power consumption but at the expense of a slower response to touch.

Setting	Time
0	Power Down
1	16 ms
2	32 ms
3	48 ms
4	64 ms
...254	4.064 s
255	4.08 s

Default: 1 (16 ms between key acquisitions)

To wake the device from Power-down mode a nonzero LP setting should be written to this address. The QT2120 can also be reset during power-down mode by writing a nonzero value to the reset register (address 7).

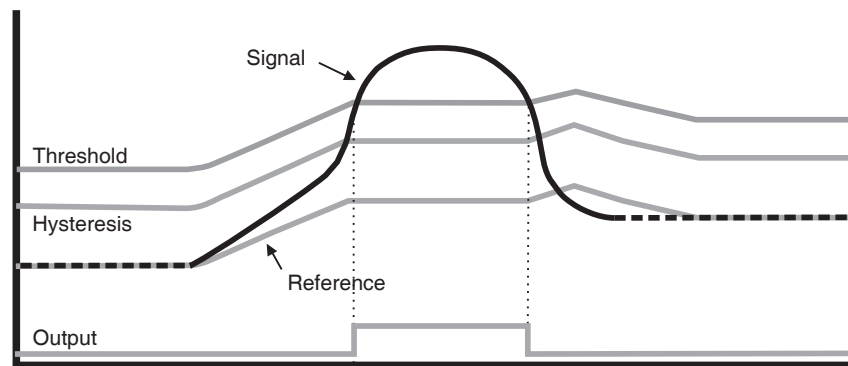
5.10 Address 9 – 10: Toward Touch and Away from Touch Drift (TTD, ATD)

Table 5-10. Toward Touch and Away from Touch Drift

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
9	0	TOWARD TOUCH DRIFT						
10	0	AWAY FROM TOUCH DRIFT						

TOWARD TOUCH DRIFT and **AWAY FROM TOUCH DRIFT:** Signals can drift because of changes in Cx and Cs over time and temperature. It is crucial that such drift be compensated for, else false detections and sensitivity shifts can occur.

Drift compensation (see [Figure 5-1](#)) is performed by making the reference level track the raw signal at a slow rate, but only while there is no detection in effect. The rate of adjustment must be performed slowly, otherwise legitimate detections could be ignored. The parameters can be configured in increments of 0.16s.

Figure 5-1. Thresholds and Away From Touch Drift

The device drift compensates using a slew-rate limited change to the reference level; the threshold and hysteresis values are slaved to this reference.

When a finger is sensed, the signal increases due to capacitance being added to C_x . An isolated, untouched foreign object (a coin, or a water film) will cause the signal to drop very slightly due to an enhancement of coupling.

Once a finger is sensed, the drift compensation mechanism ceases since the signal is legitimately detecting an object. Drift compensation only works when the signal in question has not crossed the negative threshold level.

The drift compensation mechanism can be asymmetric; the drift-compensation can be made to occur in one direction faster than it does in the other simply by changing the TTD and ATD Setup parameters. This is a global configuration.

Specifically, drift compensation should be set to compensate faster for decreasing signals than for increasing signals. Increasing signals should not be compensated quickly, since an approaching finger could be compensated for partially or entirely before even touching the touchpad (Toward Touch Drift (TTD)).

However, an obstruction over the sense pad, for which the sensor has already made full allowance, could suddenly be removed leaving the sensor with an artificially suppressed reference level and thus become insensitive to touch. In this latter case, the sensor should compensate for the object's removal by lowering the reference level relatively quickly (Away from Touch Drift (ATD)).

Drift compensation and the detection time-outs work together to provide for robust, adaptive sensing. The time-outs provide abrupt changes in reference calibration depending on the duration of the signal 'event'.

If ATD or TTD is set to 0 then the drift compensation in the respective direction is disabled.

Note: it is recommended that the drift compensation rate be more than four times the LP mode period. This is to prevent undersampling, which decreases the algorithm's efficiency.

Default TTD: 20 (3.2 s / reference level)

Default ATD: 5 (0.8 s / reference level)

5.11 Address 11: Detection Integrator (DI)

Table 5-11. Detection Integrator

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
11	DI							

DI: Allows the DI level to be set for each key. This 8-bit value controls the number of consecutive measurements that must be confirmed as having passed the key threshold before that key is registered as being in detect. The minimum value for the DI filter is 1. A settings of 0 for the DI also defaults to 1.

Default: 4 (maximum = 32)

5.12 Address 12: Touch Recal Delay (TRD)

Table 5-12. Touch Recal Delay

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
12	TRD							

If an object unintentionally contacts a key resulting in a detection for a prolonged interval it is usually desirable to recalibrate the key in order to restore its function, perhaps after a time delay of some seconds.

The Touch Recal Delay timer monitors such detections; if a detection event exceeds the timer's setting, the key will be automatically recalibrated. After a recalibration has taken place, the affected key will once again function normally even if it is still being contacted by the foreign object. This feature is set globally.

TRD can be disabled by setting it to zero (infinite timeout) in which case the key will never autorecalibrate during a continuous detection (but the host could still command it).

TRD is set globally, which can range in value from 1 – 255. TRD above 0 is expressed in 0.16 s increments.

Default: 255 (Comms) 255 (Standalone)

5.13 Address 13: Drift Hold Time (DHT)

Table 5-13. Drift Hold Time

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
13	DHT							

This is used to restrict drift on all keys while one or more keys are activated. DHT defines the length of time the drift is halted after a key detection. When DHT = 0, drifting is never suspended, even during a valid touch of another key.

This feature is particularly useful in cases of high-density keypads where touching a key or hovering a finger over the keypad would cause untouched keys to drift, and therefore create a sensitivity shift, and ultimately inhibit any touch detection. It is expressed in 0.16 s increments.

DHT default value: 25

DHT range: 0 – 255

5.14 Address 14: Slider Options

Table 5-14. Slider Options

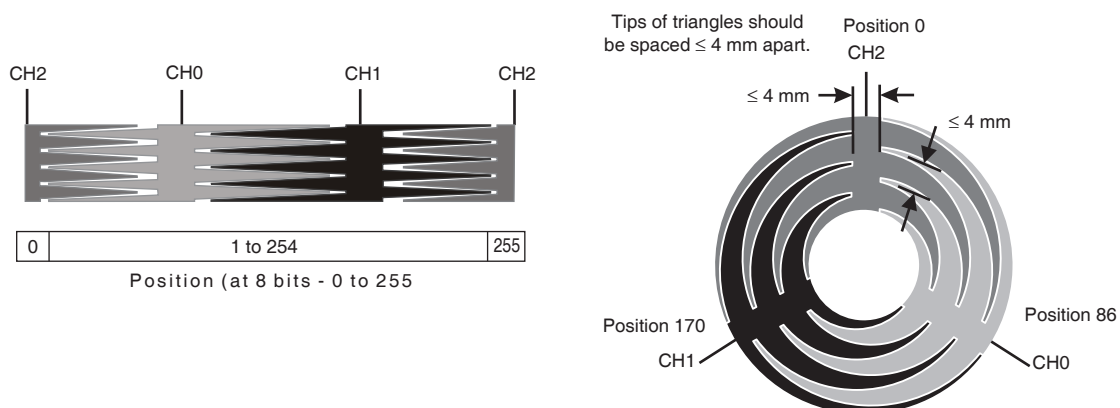
Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
14	EN	WHEEL	Reserved					

EN: Setting this bit enables a Slider or Wheel to be configured. Only the first three channels (0, 1 and 2) can be used.

WHEEL: Setting this bit allows a wheel to be configured. If not set, and EN is enabled, it defaults to a slider.

The range of both is from 0 – 255.

Figure 5-2. Slider/Wheel Settings



EN default value: 0

WHEEL default value: 0