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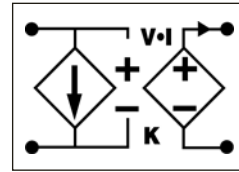
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# BCM™ Bus Converter

- 384 V to 12 V V•I Chip™ Converter
- 300 Watt (450 Watt for 1 ms)
- High density – up to 1017 W/in<sup>3</sup>
- Small footprint – 260 W/in<sup>2</sup>
- Low weight – 0.5 oz (15 g)
- ZVS / ZCS isolated sine amplitude converter
- Typical efficiency 95%
- 125°C operation (T<sub>J</sub>)
- <1 μs transient response
- >3.5 million hours MTBF
- No output filtering required



**V<sub>in</sub> = 360 - 400 V**  
**V<sub>out</sub> = 11.3 - 12.5 V**  
**I<sub>out</sub> = 25 A**  
**K = 1/32**  
**R<sub>out</sub> = 20.0 mΩ max**



## Product Description

The V•I Chip bus converter is a high efficiency (>95%), narrow input range Sine Amplitude Converter™ (SAC™) operating from a 360 to 400 Vdc primary bus to deliver an isolated low voltage secondary (ELV).

The off-line BCM provides an isolated 11.3 -12.5 V distribution bus and is ideal for use in silver boxes and PFC front ends. Due to the fast response time and low noise of the BCM, the need for limited life aluminum electrolytic or tantalum capacitors at the input of POL converters is reduced—or eliminated—resulting in savings of board area, materials and total system cost.

The BCM achieves a power density of 1017 W/in<sup>3</sup> in a V•I Chip package compatible with standard pick-and-place and surface mount assembly processes. The V•I Chip package provides flexible thermal management through its low junction-to-case and junction-to-board thermal resistance. Owing to its high conversion efficiency and safe operating temperature range, the BCM does not require a discrete heat sink in typical applications. Low junction-to-case and junction-to-lead thermal impedances assure low junction temperatures and long life in the harshest environments.

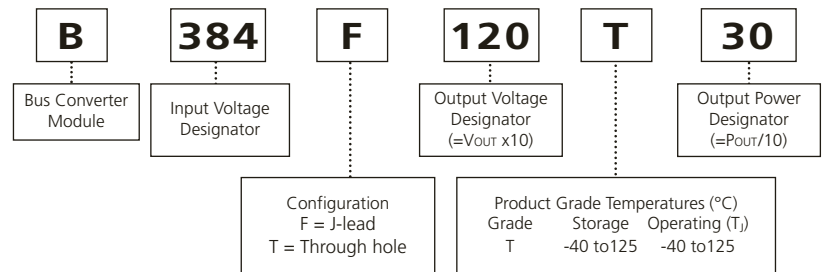
## Absolute Maximum Ratings

Parameter	Values	Unit	Notes
+In to -In	-1.0 to 440	Vdc	
	500	Vdc	For 100 ms
PC to -In	-0.3 to 7.0	Vdc	
+Out to -Out	-0.5 to 16.0	Vdc	
Isolation voltage	4242	Vdc	Input to Output
Output current	27.7	A	Continuous
Peak output current	37.5	A	For 1 ms
Output power	300	W	Continuous
Peak output power	450	W	For 1 ms
Case temperature	225	°C	MSL 5
	245	°C	MSL 6
Operating junction temperature <sup>(1)</sup>	-40 to 125	°C	T-Grade
Storage temperature	-40 to 125	°C	T-Grade

### Note:

(1) The referenced junction is defined as the semiconductor having the highest temperature. This temperature is monitored by a shutdown comparator.

## Part Numbering



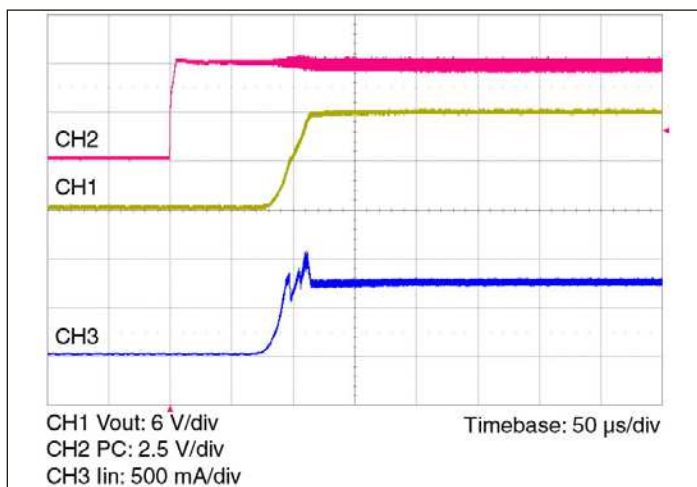


## Specifications

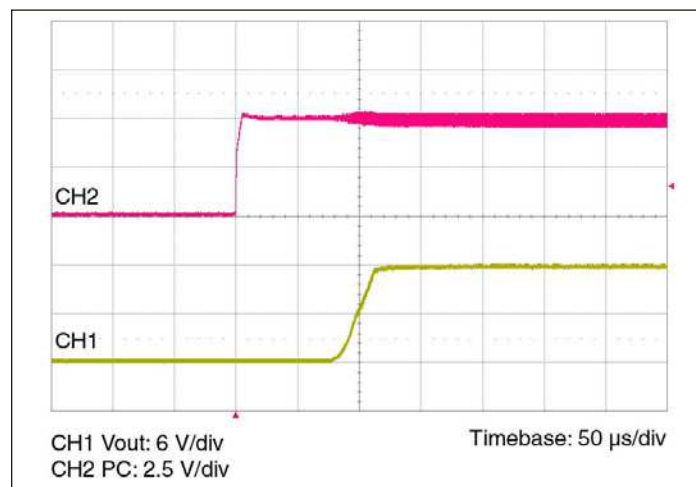
**Input** (Conditions are at 384 Vin, full load, and 25°C ambient unless otherwise specified)

Parameter	Min	Typ	Max	Unit	Note
Input voltage range	360	384	400	Vdc	
Input dV/dt			1	V/μs	
Input undervoltage turn-on			320	Vdc	
Input undervoltage turn-off	280			Vdc	
Input overvoltage turn-on	400			Vdc	
Input overvoltage turn-off			440	Vdc	
Input quiescent current		1.1		mA	PC low
Inrush current overshoot		0.28		A	Using test circuit in Figure 20; See Figure 1
Input current			0.9	Adc	
Input reflected ripple current		456		mA p-p	Using test circuit in Figure 20; See Figure 4
No load power dissipation		5.8	8.3	W	
Internal input capacitance		0.2		μF	
Internal input inductance		5		nH	
Recommended external input capacitance		2.2		μF	200 nH maximum source inductance; See Figure 20

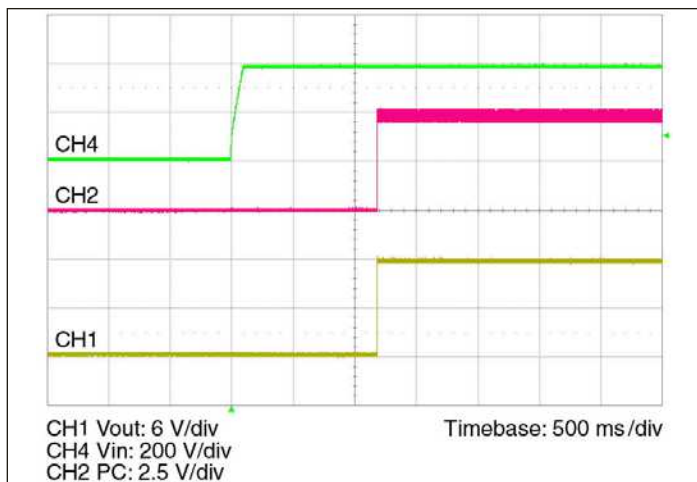
## Input Waveforms



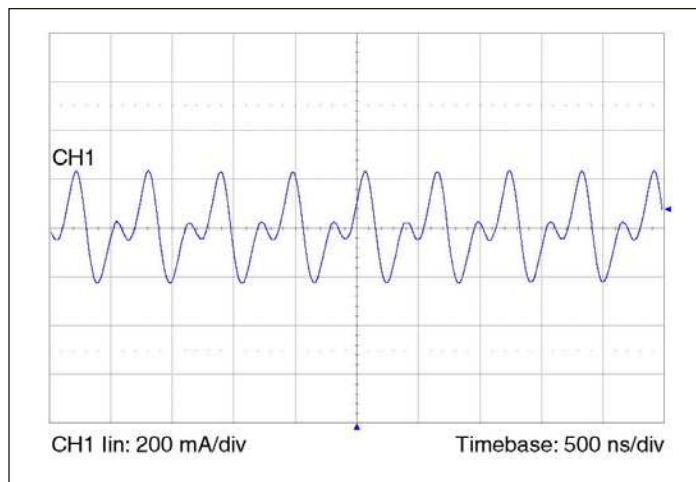
**Figure 1** — Inrush transient current at full load and 384 Vin with PC enabled



**Figure 2** — Output voltage turn-on waveform with PC enabled at full load and 384 Vin



**Figure 3** — Output voltage turn-on waveform with input turn-on at full load and 384 Vin



**Figure 4** — Input reflected ripple current at full load and 384 Vin

## Specifications (continued)

**Output** (Conditions are at 384 V<sub>in</sub>, full load, and 25°C ambient unless otherwise specified)

Parameter	Min	Typ	Max	Unit	Note
Output voltage	11.3		12.5	V <sub>dc</sub>	No load
	10.8		12.0	V <sub>dc</sub>	Full load
Output power	0		300	W	364 - 400 V <sub>IN</sub>
Rated DC current	0		27.7	A <sub>dc</sub>	P <sub>OUT</sub> ≤ 300 W
Peak repetitive power			450	W	Max pulse width 1ms, max duty cycle 10%, baseline power 50%
Current share accuracy		5	10	%	See Parallel Operation on Page 11
Efficiency					
Half load	94.1	95.2		%	See Figure 5
Full load	94.2	95.3		%	See Figure 5
Internal output inductance		1.1		nH	
Internal output capacitance		31		μF	Effective value
Load capacitance			1,000	μF	
Output overvoltage setpoint	12.5			V <sub>dc</sub>	
Output ripple voltage					
No external bypass		197	400	mV p-p	See Figures 7 and 9
10 μF bypass capacitor		23		mV p-p	See Figure 8
Short circuit protection set point	28.2			A <sub>dc</sub>	Module will shut down
Average short circuit current		0.23		A	
Effective switching frequency	3.3	3.4	3.5	MHz	Fixed, 1.7 MHz per phase
Line regulation					
K	0.0309	1/32	0.0316		V <sub>OUT</sub> = K • V <sub>IN</sub> at no load
Load regulation					
R <sub>OUT</sub>		15.0	20.0	mΩ	
Transient response					
Voltage overshoot		74		mV	100% load step; See Figures 10 and 11
Response time		200		ns	See Figures 10 and 11
Recovery time		1		μs	See Figures 10 and 11
Output overshoot					
Input turn-on		0		mV	No output filter; See Figure 3
PC enable		0		mV	No output filter; See Figure 2
Output turn-on delay					
From application of power		1180		ms	No output filter; See Figure 3
From release of PC pin		240		ms	No output filter

## Output Waveforms

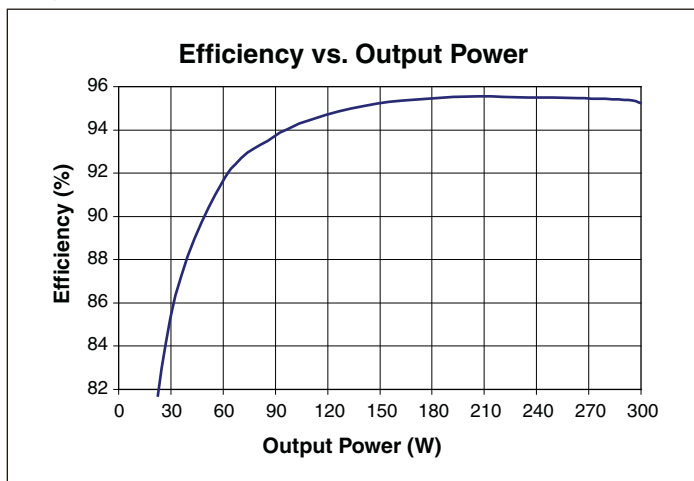


Figure 5 — Efficiency vs. output power at 384 V<sub>in</sub>

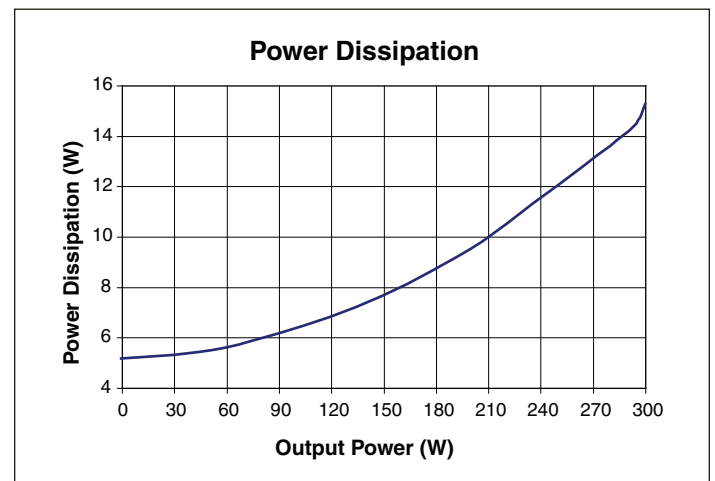


Figure 6 — Power dissipation as a function of output power

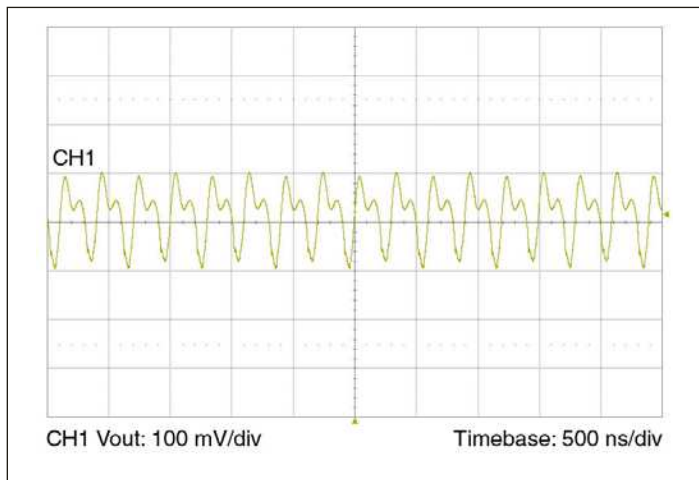


Figure 7 — Output voltage ripple at full load and 384 Vin without any external bypass capacitor.

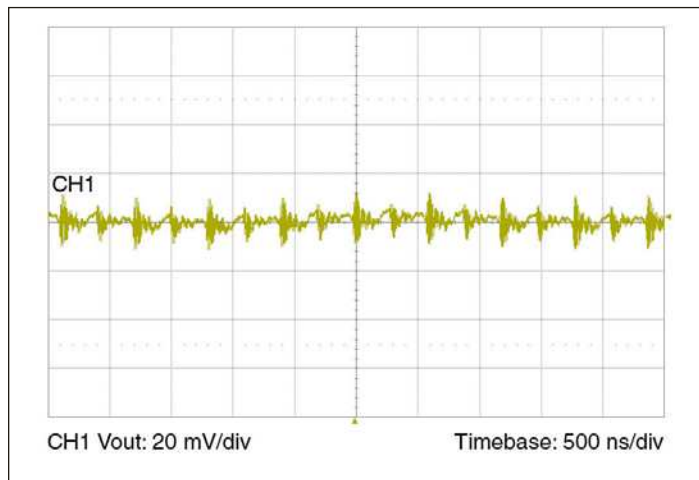


Figure 8 — Output voltage ripple at full load and 384 Vin with 10  $\mu$ F ceramic external bypass capacitor and 20 nH of distribution inductance.

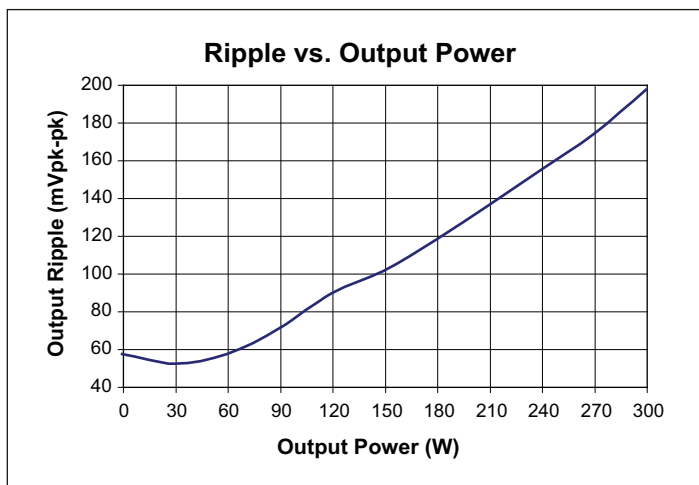


Figure 9 — Output voltage ripple vs. output power at 384 Vin without any external bypass capacitor.

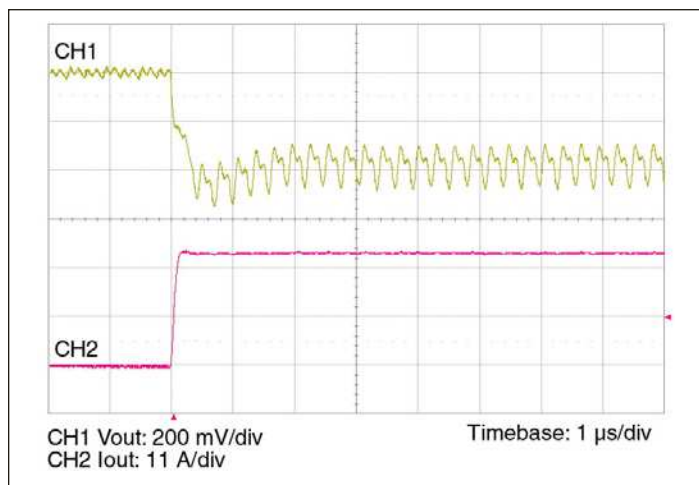


Figure 10 — 0-25 A load step with 2.2  $\mu$ F input capacitor and no output capacitor.

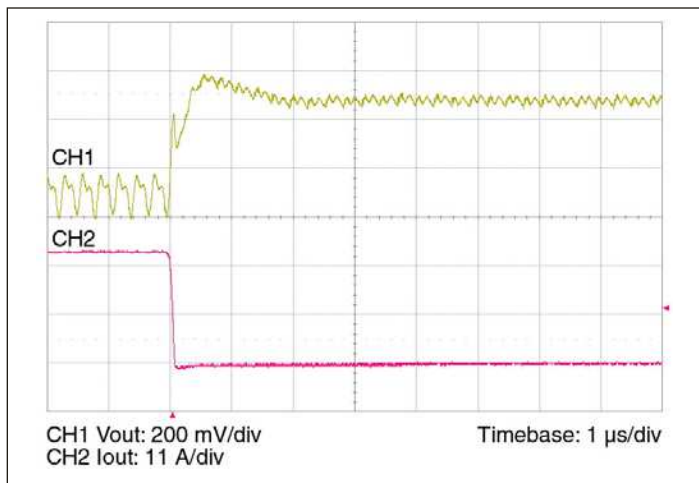


Figure 11 — 25-0 A load step with 2.2  $\mu$ F input capacitor and no output capacitor.

**General**

Parameter	Min	Typ	Max	Unit	Note
MTBF					
MIL-HDBK-217F		3.5		Mhrs	25°C, GB
Isolation specifications					
Voltage	4242			Vdc	Input to Output
Capacitance		500		pF	Input to Output
Resistance	10			MΩ	Input to Output
Agency approvals					
		cTÜVus			UL/CSA 60950-1, EN 60950-1
		CE Mark			Low Voltage Directive
		RoHS			
Mechanical					
Weight		0.53/15		oz/g	See Mechanical Drawings, Figures 15 – 18
Dimensions					
Length		1.28/32,5		in/mm	
Width		0.87/22		in/mm	
Height		0.265/6,73		in/mm	
Thermal					
Over temperature shutdown	125	130	135	°C	Junction temperature
Thermal capacity		9.3		Ws/°C	
Junction-to-case thermal impedance ( $R_{\theta JC}$ )		1.1		°C/W	
Junction-to-board thermal impedance ( $R_{\theta JB}$ )		2.1		°C/W	

**Auxiliary Pins** (Conditions are at 48 Vin, full load, and 25°C ambient unless otherwise specified)

Parameter	Min	Typ	Max	Unit	Note
Primary control (PC)					
DC voltage	4.8	5.0	5.2	Vdc	
Module disable voltage	2.4	2.5		Vdc	
Module enable voltage		2.5	2.6	Vdc	
Current limit	2.4	2.5	2.9	mA	Source only
Enable delay time		240		ms	
Disable delay time		40		μs	See Figure 12, time from PC low to output low

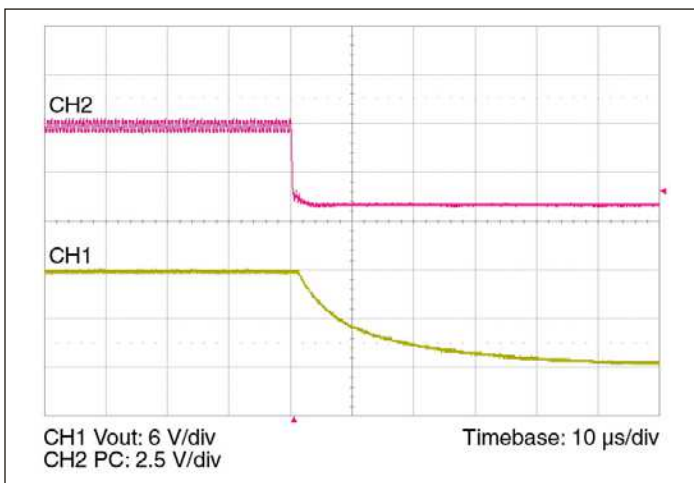


Figure 12 —  $V_{OUT}$  at full load vs. PC disable

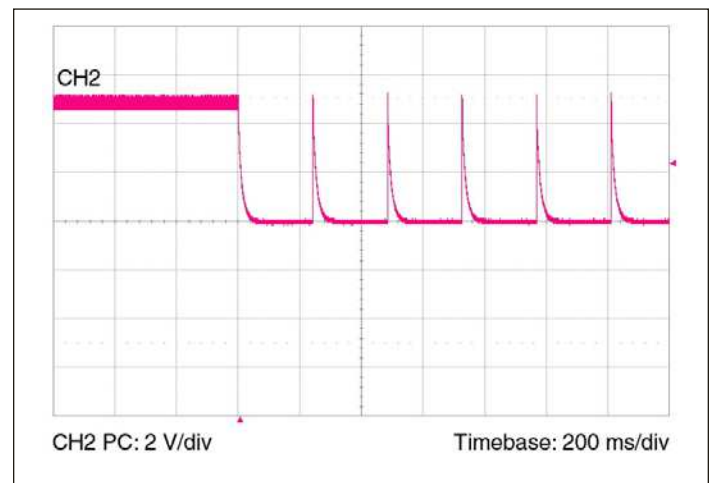


Figure 13 — PC signal during fault

### +In / -In – DC Voltage Input Ports

The V•I Chip input voltage range should not be exceeded. An internal under / over voltage lockout-function prevents operation outside of the normal operating input range. The BCM turns on within an input voltage window bounded by the “Input under-voltage turn-on” and “Input over-voltage turn-off” levels, as specified. The V•I Chip may be protected against accidental application of a reverse input voltage by the addition of a rectifier in series with the positive input, or a reverse rectifier in shunt with the positive input located on the load side of the input fuse.

The connection of the V•I Chip to its power source should be implemented with minimal distribution inductance. If the interconnect inductance exceeds 100 nH, the input should be bypassed with a RC damper to retain low source impedance and stable operation. With an interconnect inductance of 200 nH, the RC damper may be 2.2  $\mu$ F in series with 0.3 $\Omega$ . A single electrolytic or equivalent low-Q capacitor may be used in place of the series RC bypass.

### PC – Primary Control

The Primary Control port is a multifunction node that provides the following functions:

**Enable / Disable** – If the PC port is left floating, the BCM output is enabled. Once this port is pulled lower than 2.4 Vdc with respect to -In, the output is disabled. This action can be realized by employing a relay, opto-coupler, or open collector transistor. Refer to Figures 1-3, 12 and 13 for the typical enable / disable characteristics. This port should not be toggled at a rate higher than 1 Hz. The PC port should also not be driven by or pulled up to an external voltage source.

**Primary Auxiliary Supply** – The PC port can source up to 2.4 mA at 5.0 Vdc. The PC port should never be used to sink current.

**Alarm** – The BCM contains circuitry that monitors output overload, input over voltage or under voltage, and internal junction temperatures. In response to an abnormal condition in any of the monitored parameters, the PC port will toggle. Refer to Figure 13 for PC alarm characteristics.

### TM and RSV – Reserved for factory use.

### +Out / -Out – DC Voltage Output Ports

Two sets of contacts are provided for the +Out port. They must be connected in parallel with low interconnect resistance. Similarly, two sets of contacts are provided for the -Out port. They must be connected in parallel with low interconnect resistance. Within the specified operating range, the average output voltage is defined by the Level 1 DC behavioral model of Figure 21. The current source capability of the BCM is rated in the specifications section of this document.

The low output impedance of the BCM reduces or eliminates the need for limited life aluminum electrolytic or tantalum capacitors at the input of POL converters.

Total load capacitance at the output of the BCM should not exceed the specified maximum. Owing to the wide bandwidth and low output impedance of the BCM, low frequency bypass capacitance and significant energy storage may be more densely and efficiently provided by adding capacitance at the input of the BCM.

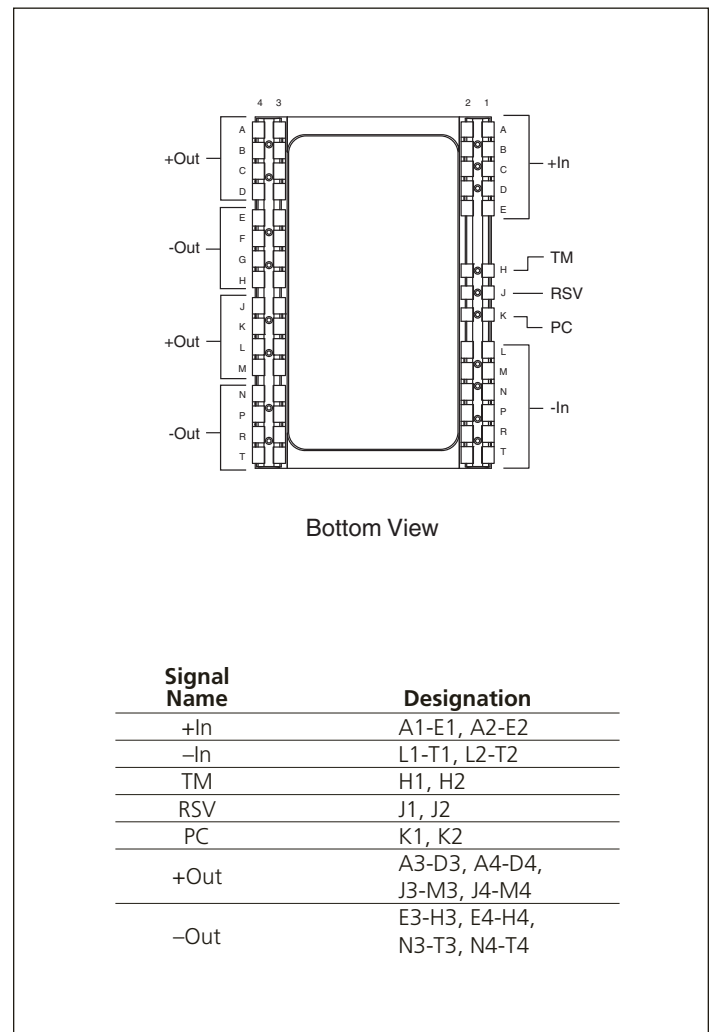


Figure 14 — BCM pin configuration

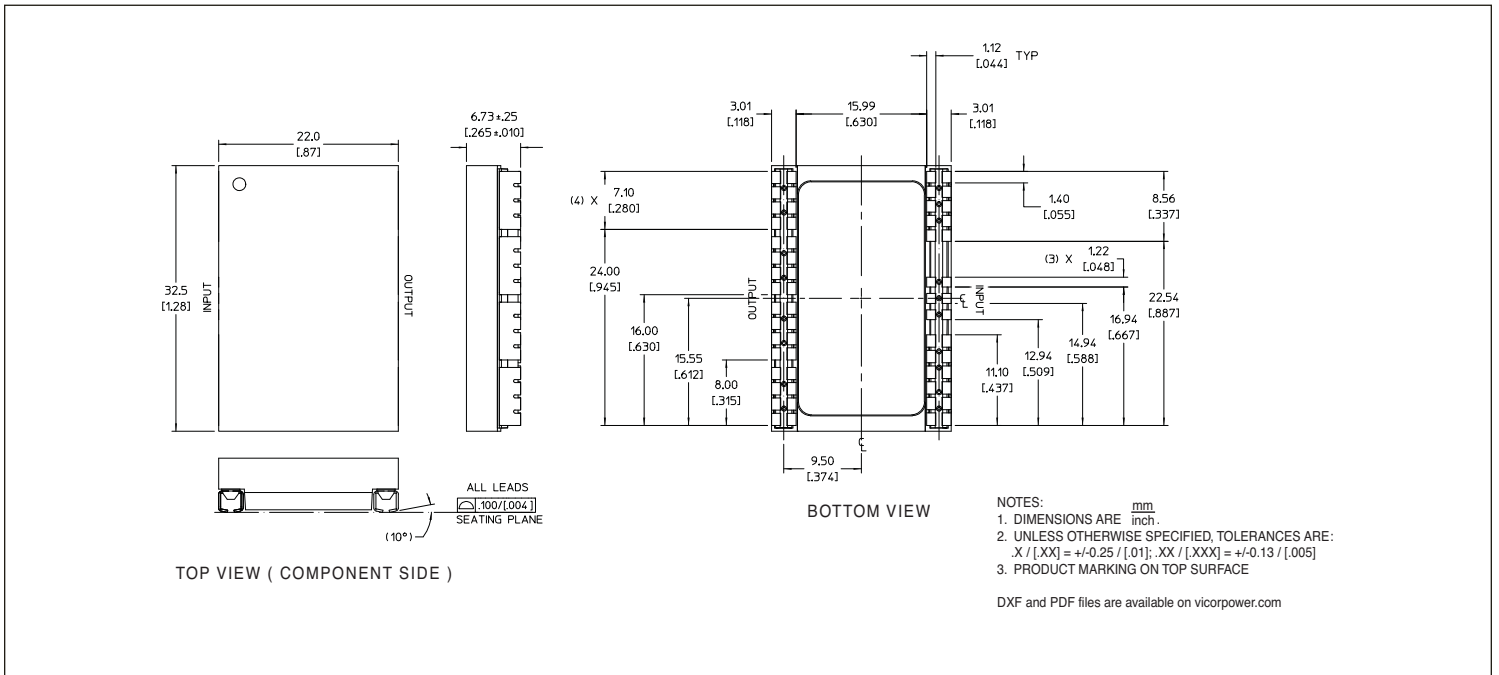


Figure 15 — BCM J-Lead mechanical outline; Onboard mounting

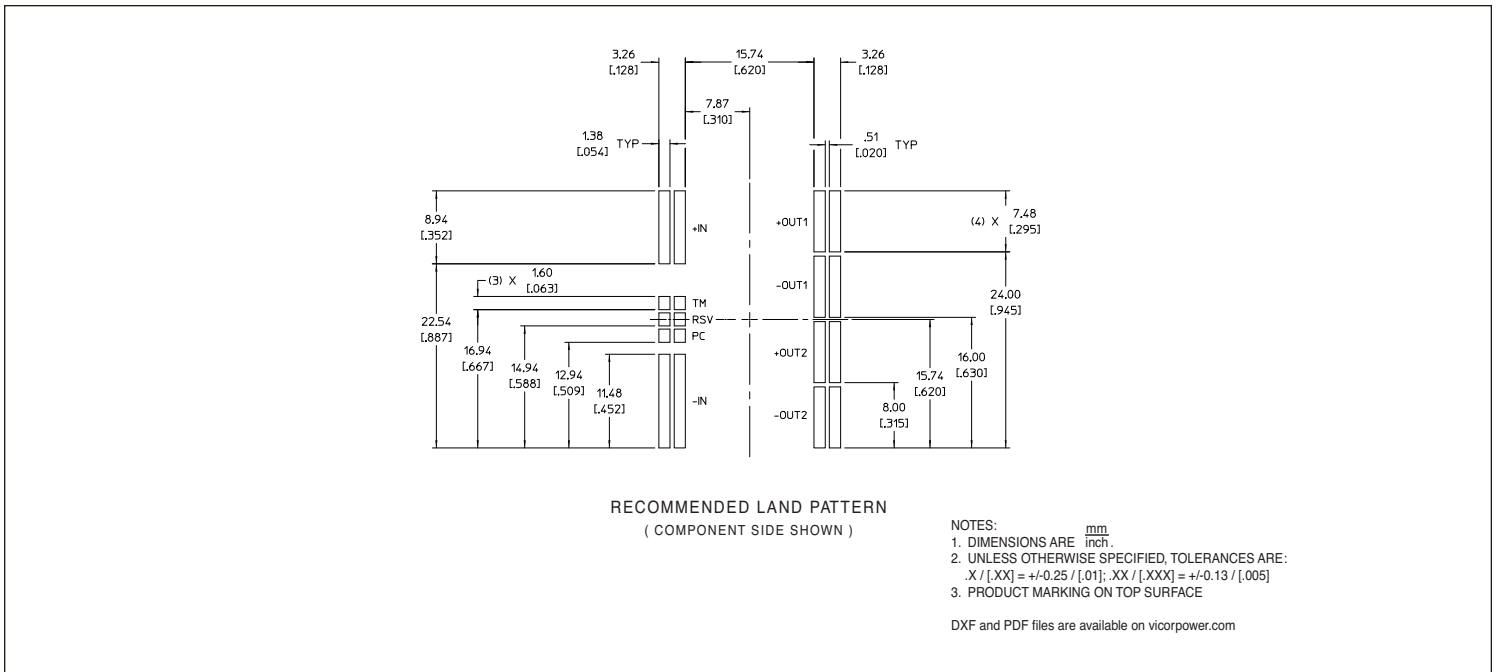


Figure 16 — BCM PCB land layout information



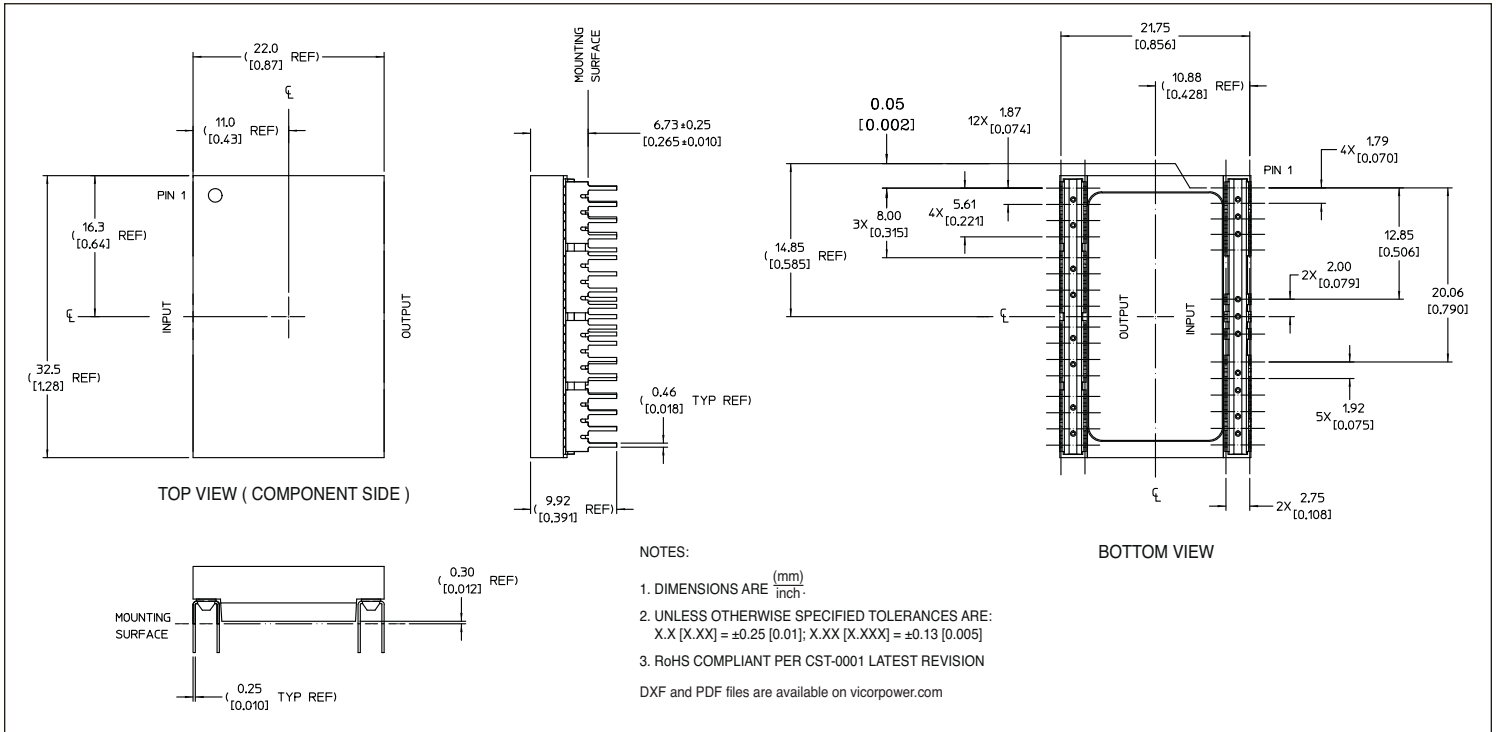


Figure 17 — BCM through-hole mechanical outline

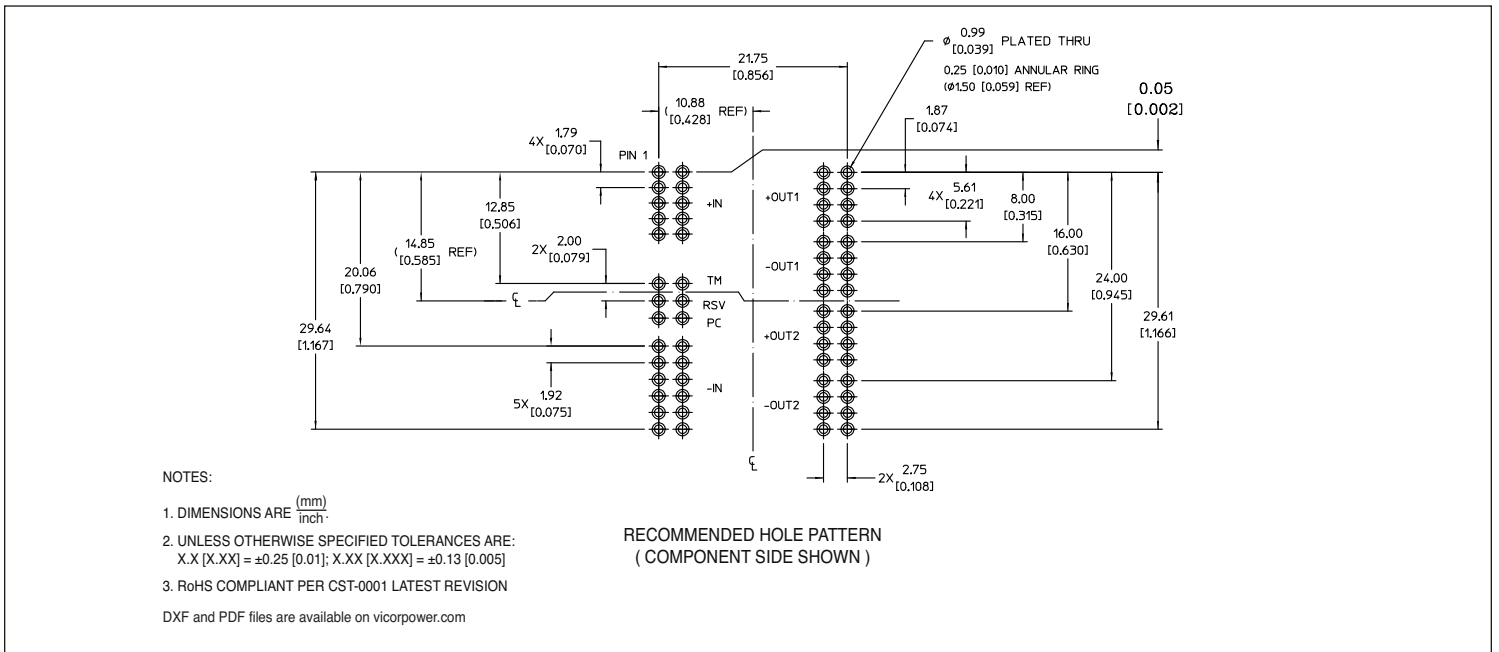
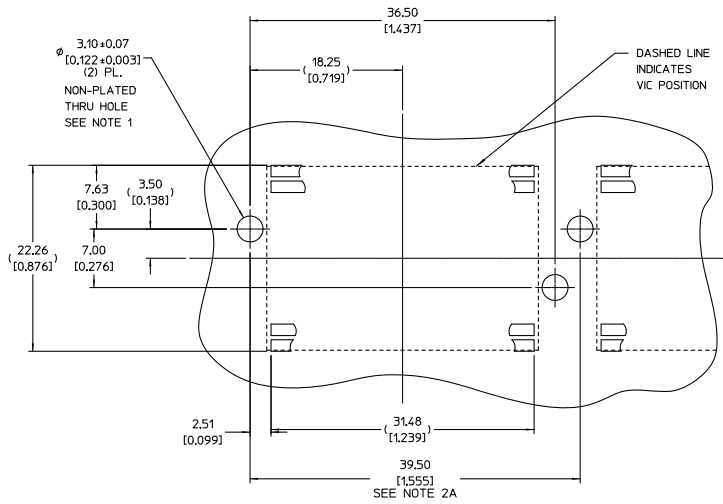


Figure 18 — BCM through-hole PCB layout information

## RECOMMENDED LAND PATTERN (NO GROUNDING CLIPS)

TOP SIDE SHOWN



NOTES: 1. MAINTAIN 3.50 [0.138] DIA. KEEP-OUT ZONE FREE OF COPPER, ALL PCB LAYERS.

2. (A) MINIMUM RECOMMENDED PITCH IS 39.50 [1.555], THIS PROVIDES 7.00 [0.275] COMPONENT EDGE-TO-EDGE SPACING, AND 0.50 [0.020] CLEARANCE BETWEEN VICOR HEAT SINKS.

(B) MINIMUM RECOMMENDED PITCH IS 41.00 [1.614], THIS PROVIDES 8.50 [0.334] COMPONENT EDGE-TO-EDGE SPACING, AND 2.00 [0.079] CLEARANCE BETWEEN VICOR HEAT SINKS.

3. V-I CHIP LAND PATTERN SHOWN FOR REFERENCE ONLY; ACTUAL LAND PATTERN MAY DIFFER. DIMENSIONS FROM EDGES OF LAND PATTERN TO PUSH-PIN HOLES WILL BE THE SAME FOR ALL FULL SIZE V-I CHIP PRODUCTS.

4. RoHS COMPLIANT PER CST-0001 LATEST REVISION.

5. UNLESS OTHERWISE SPECIFIED: DIMENSIONS ARE MM [INCH]. TOLERANCES ARE:  
X.X [X.XX] = ±0.3 [0.01]  
X.XX [X.XXX] = ±0.13 [0.005]

6. PLATED THROUGH HOLES FOR GROUNDING CLIPS (33855) SHOWN FOR REFERENCE. HEATSINK ORIENTATION AND DEVICE PITCH WILL DICTATE FINAL GROUNDING SOLUTION.

## RECOMMENDED LAND PATTERN (With GROUNDING CLIPS)

TOP SIDE SHOWN

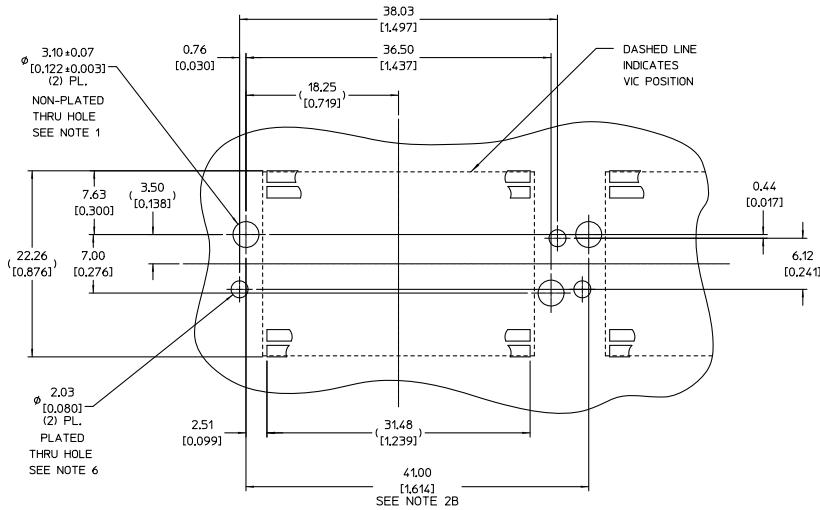
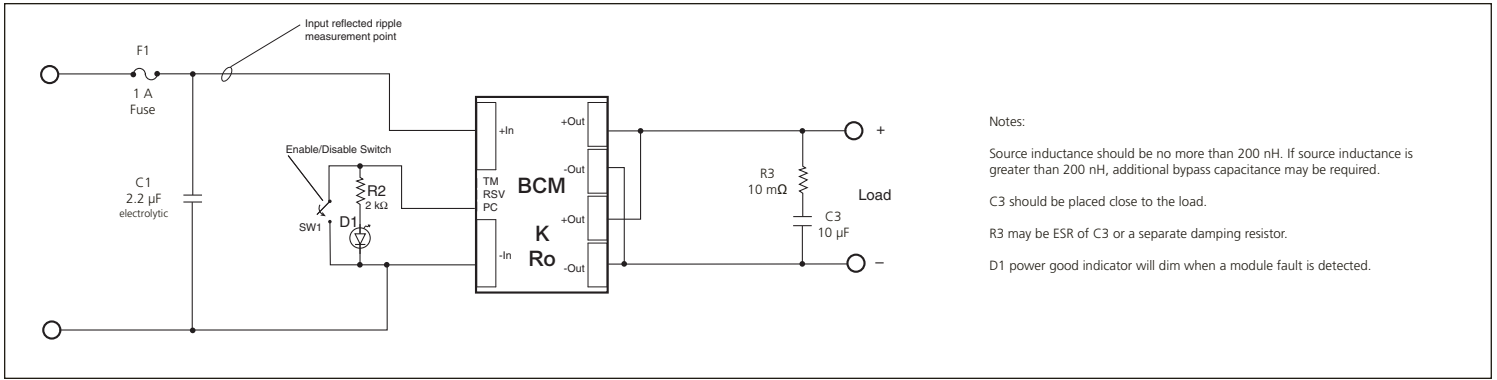
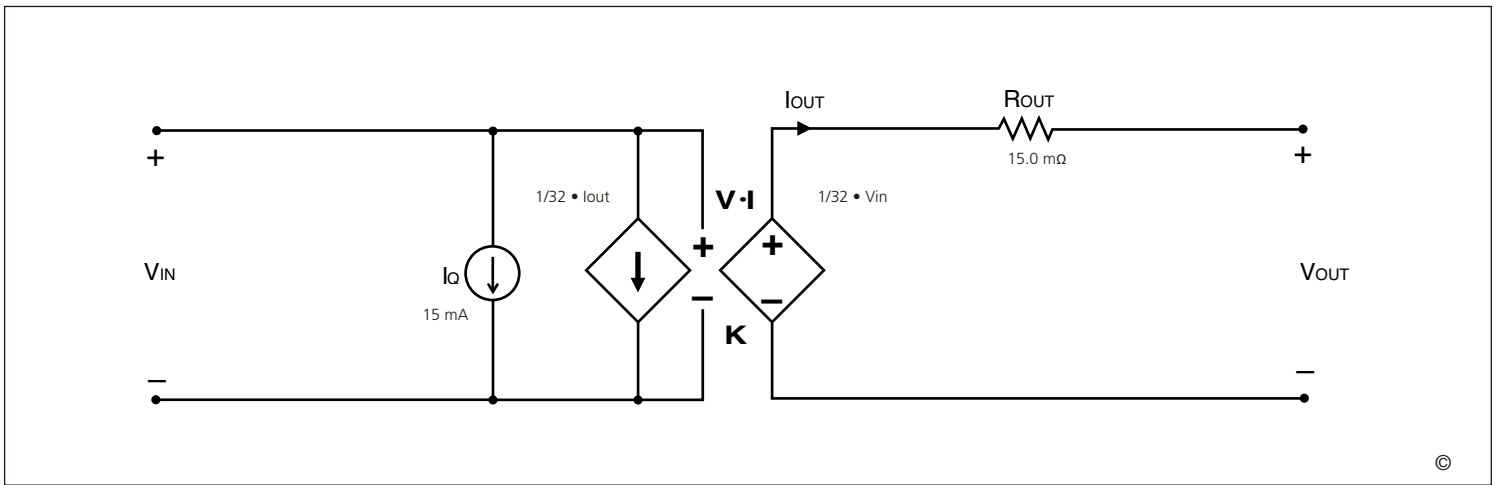


Figure 19 — Hole location for push pin heat sink relative to V-I Chip



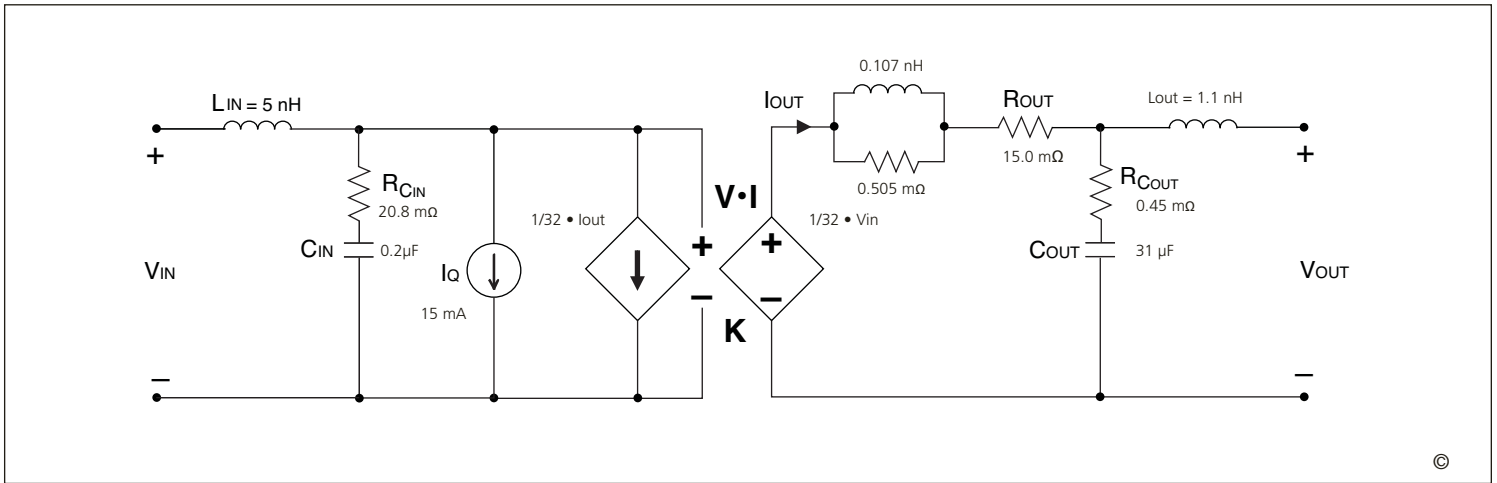
**Figure 20** — BCM test circuit

**V•I Chip Bus Converter Level 1 DC Behavioral Model for 384 V to 12 V, 300 W**



**Figure 21** — This model characterizes the DC operation of the V•I Chip bus converter, including the converter transfer function and its losses. The model enables estimates or simulations of output voltage as a function of input voltage and output load, as well as total converter power dissipation or heat generation.

**V•I Chip Bus Converter Level 2 Transient Behavioral Model for 384 V to 12 V, 300 W**



**Figure 22** — This model characterizes the AC operation of the V•I Chip bus converter including response to output load or input voltage transients or steady state modulations. The model enables estimates or simulations of input and output voltages under transient conditions, including response to a stepped load with or without external filtering elements.

### Parallel Operation

The BCM will inherently current share when operated in an array. Arrays may be used for higher power or redundancy in an application.

Current sharing accuracy is maximized when the source and load impedance presented to each BCM within an array are equal. The recommended method to achieve matched impedances is to dedicate common copper planes within the PCB to deliver and return the current to the array, rather than rely upon traces of varying lengths. In typical applications the current being delivered to the load is larger than that sourced from the input, allowing traces to be utilized on the input side if necessary. The use of dedicated power planes is, however, preferable.

The BCM power train and control architecture allow bi-directional power transfer, including reverse power processing from the BCM output to its input. Reverse power transfer is enabled if the BCM input is within its operating range and the BCM is otherwise enabled. The BCM's ability to process power in reverse improves the BCM transient response to an output load dump.

### Input Impedance Recommendations

To take full advantage of the BCM capabilities, the impedance presented to its input terminals must be low from DC to approximately 5 MHz. The source should exhibit low inductance (less than 100 nH) and should have a critically damped response. If the interconnect inductance exceeds 100 nH, the BCM input pins should be bypassed with an RC damper (e.g., 2.2  $\mu$ F in series with 0.3 ohm) to retain low source impedance and stable operations. Given the wide bandwidth of the BCM, the source response is generally the limiting factor in the overall system response.

Anomalies in the response of the source will appear at the output of the BCM multiplied by its K factor. The DC resistance of the source should be kept as low as possible to minimize voltage deviations. This is especially important if the BCM is operated near low or high line as the over/under voltage detection circuitry could be activated.

### Input Fuse Recommendations

V•I Chips are not internally fused in order to provide flexibility in configuring power systems. However, input line fusing of V•I Chips must always be incorporated within the power system. A fast acting fuse should be placed in series with the +In port.

### Application Notes

For BCM and V•I Chip application notes on soldering, thermal management, board layout, and system design click on the link below:

[http://www.vicorpower.com/technical\\_library/application\\_information/chips/](http://www.vicorpower.com/technical_library/application_information/chips/)



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