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SAW Components

Data Sheet B3558





SAW ComponentsB3558Low-loss Filter449,10 MHzData Sheet

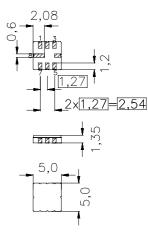
Ceramic package QCC8C

Features

- RF low-loss filter for remote control receivers
- Package for Surface Mounted Technology (SMT)
- Balanced and unbalanced operation possible

Terminals

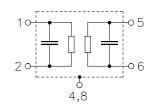
Ni, gold plated



typ. dimensions in mm, approx. weight 0,1 g

Pin configuration

1	Input Ground
2	Input
5	Output
6	Output Ground
4,8	Case - Ground
3,7	to be grounded



Туре	Ordering code	Marking and package according to	Packing according to
B3558	B39451-B3558-U310	C61157-A7-A56	F61074-V8070-Z000

Electrostactic Sensitive Device (ESD)

Maximum ratings

Operable temperature range	T _A	-30/+85	°C	
Storage temperature range	T _{stg}	-45/+90	°C	
DC voltage	V _{DC}	0	V	
Source power	P_S	10	dBm	source impedance 50 Ω





SAW Components					B3558
Low-loss Filter				449,	10 MHz
Data Sheet					
Characteristics					
Terminating source impedance: $Z_{\rm S}$		and matc	ning network ning network		
		min.	typ.	max.	
Center frequency (center frequency between 3 dB points)	f _C	_	449,14	—	MHz
Minimum insertion attenuation 449,00 449,28 MHz	$lpha_{min}$	_	2,2	3,2	dB
Pass band (relative to α_{min})					
448,965 449,315 MH		_	0,8	3,0	dB
448,930 449,350 MH	Z		1,5	6,0	dB
Relative attenuation (relative to α_{min})	α_{rel}				
10,00 428,00 MHz		45	50	—	dB
428,00 439,00 MHz		40	45		dB
439,00 448,10 MHz	:	15	25	—	dB
450,10 459,00 MHz	:	15	25	—	dB
459,00 550,00 MHz			45	—	dB
550,00 1000,00 MHz		45	50	—	dB
Impedance for pass band matching ²⁾					
Input: $Z_{IN} = R_{IN} C_{IN}$		_	210 2,70	_	$\Omega \parallel pF$
Output: $Z_{OUT} = R_{OUT} C_{OUT}$		_	210 2,70		$\Omega \parallel pF$
Temperature coefficient of frequency 1)	TC _f		-0,03		ppm/K ²
Frequency inversion point	T_0	20	_	40	°C

¹⁾Temperature dependence of f_C : $f_C(T_A) = f_C(T_0) (1 + TC_f(T_A - T_0)^2)$

²⁾ Impedance for passband matching bases on an ideal, perfect matching of the SAW filter to source- and to load impedance (here 50 Ohm). After the SAW filter is removed and input impedance into the input matching / output matching network is calculated.

The conjugate complex value of these characteristic impedances are the input and output impedances for flat passband. For more details, we refer to EPCOS application note #4.



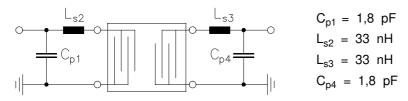
SAW Components				B3558
Low-loss Filter			449,	10 MHz
Data Sheet				
Characteristics				
Reference temperature: $T_A = -30$				
		ning network		
Terminating load impedance: $Z_{\rm L} = 50$ G	2 and match	ning network	K	
	min.	typ.	max.	
Center frequency f _C	_	449,10	_	MHz
(center frequency between 3 dB points)				
Minimum insertion attenuation α_{min}				
449,00 449,28 MHz	_	2,2	3,4	dB
Pass band (relative to α_{\min})				
448,965 449,235 MHz	—	0,5	3,0	dB
448,930 449,270 MHz	-	1,5	6,0	dB
Relative attenuation (relative to α_{min}) α_{rel}				
10,00 428,00 MHz	45	50	_	dB
428,00 439,00 MHz	40	45	—	dB
439,00 448,02 MHz	15	25		dB
450,10 459,00 MHz	15	25		dB
459,00 550,00 MHz	40	45		dB
550,00 1000,00 MHz	45	50	—	dB
Impedance for pass band matching ²⁾				
Input: $Z_{IN} = R_{IN} C_{IN}$	-	210 2,70		$\Omega \parallel pF$
Output: $Z_{OUT} = R_{OUT} C_{OUT}$	_	210 2,70	_	$\Omega \parallel pF$

²⁾ Impedance for passband matching bases on an ideal, perfect matching of the SAW filter to source- and to load impedance (here 50 Ohm). After the SAW filter is removed and input impedance into the input matching / output matching network is calculated.

The conjugate complex value of these characteristic impedances are the input and output impedances for flat passband. For more details, we refer to EPCOS application note #4.



Matching network to 50 Ω (element values depend on pcb layout and equivalent circuit)



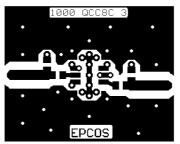
Minimising the crosstalk

For a good ultimate rejection a low crosstalk is necessary. Low crosstalk can be realised with a good RF layout. The major crosstalk mechanism is caused by the "ground-loop" problem.

Grounding loops are created if input-and output transducer GND are connected on the top-side of the PCB and fed to the system grounding plane by a common via hole. To avoid the common ground path, the ground pin of the input- and output transducer are fed to the system ground plane (bottom PCB plane) by their own via hole. The transducers' grounding pins should be isolated from the upper grounding plane.

A common GND inductivity of 0.5nH degrades the ultimate rejection (crosstalk) by 20dB.

The optimised PCB layout, including matching network for transformation to 50 Ohm, is shown here. In this PCB layout the grounding loops are minimised to realise good ultimate rejection.



Optimised PCB layout for SAW filters in QCC8C package, pinning 2,5 (top side, scale 1:1)

The bottom side is a copper plane (system ground area). The input and output grounding pins are isolated and connected to the common ground by separated via holes.

For good contact of the upper grounding area with the lower side it is necessary to place enough via holes.

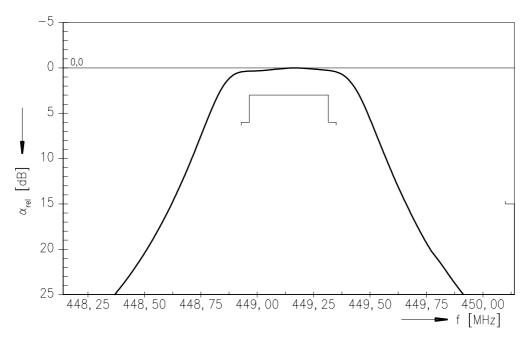


Low-loss Filter

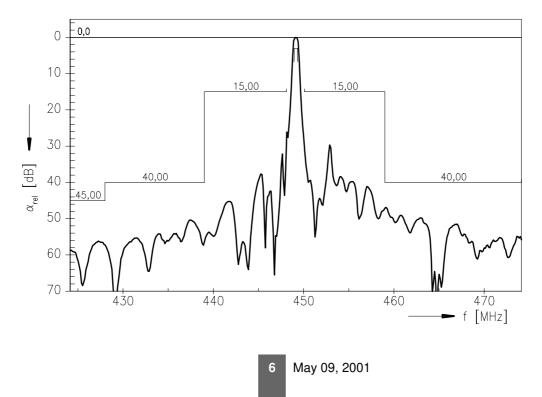
B3558 449,10 MHz

Data Sheet

Normalized frequency response



Normalized frequency response (wideband)





SAW Components	B3558
Low-loss Filter	449,10 MHz

Data Sheet

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