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## LED level meter driver, 12-point $\times 2$ channel, VU scale, bar display BA6820F / BA6822S / BA6822F

The BA6820F, BA6822S and BA6822F are two-channel, 12-point LED drivers for VU-scale bar-level meters.
The ICs are available in 22-pin SOP (BA6820F and BA6822F) and 22-pin shrink-DIP (BA6822S) packages, and use a dynamic-drive system that can drive two 12-point displays. The ICs have independent AC and DC inputs, and the AC input mode has a peak hold circuit (with cancel function). The ICs also feature a power-on mute, so steps to prevent erroneous LED lighting at power on are not required.
The input resistance of the $A C$ inputs is $3 k \Omega$ (Typ.) for the BA6820F and $0.3 \mathrm{k} \Omega$ (Typ.) for the BA6822S/F.

## -Applications

Level meters for VCRs, stereo cassette players, audio amplifiers and karaoke equipment.

## -Features

1) Uses dynamic-drive system to display two 12-point channels. 22-pin SOP and 22-pin shrink DIP packages.
2) Independent $A C$ and $D C$ inputs provided (one input displays one channel). Switching function allows two-mode display.
3) Upper 8 points have peak hold function in AC mode (two seconds). A peak-hold cancel function is also provided.
4) A mute function is provided, and muting can be applied externally.
5) Power on mute function.

Block diagram


Absolute maximum ratings
(BA6820F/BA6822F) $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| Parameter | Symbol | Limits | Unit |
| :--- | :---: | :---: | :---: |
| Power supply voltage | Vcc | 7.0 | V |
| Power dissipation | Pd | $450^{*}$ | mW |
| Operating temperature | Topr | $-20 \sim+70$ | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg | $-55 \sim+125$ | ${ }^{\circ} \mathrm{C}$ |
| Maximum output driver current | Iom | 40 | mA |
| Maximum display switch driver current | IDM | 20 | mA |

* Reduced by 4.5 mW for each increase in Ta of $1^{\circ} \mathrm{C}$ over $25^{\circ} \mathrm{C}$.
(BA6822S) $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| Parameter | Symbol | Limits | Unit |
| :--- | :---: | :---: | :---: |
| Power supply voltage | $\mathrm{V}_{\mathrm{CC}}$ | 7.0 | V |
| Power dissipation | Pd | $1000^{*}$ | mW |
| Operating temperature | Topr | $-20 \sim+70$ | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg | $-55 \sim+125$ | ${ }^{\circ} \mathrm{C}$ |
| Maximum output driver current | Iom | 60 | mA |
| Maximum display switch driver current | IDM | 30 | mA |

* Reduced by 10 mW for each increase in Ta of $1^{\circ} \mathrm{C}$ over $25^{\circ} \mathrm{C}$.
-Recommended operating conditions ( $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Min. | Typ. | Max. | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :--- |
| Oscillator frequency | fosc | - | 4 | - | kHz | $\mathrm{C}=0.01 \mu \mathrm{~F}$ <br> $\mathrm{R}=41 \mathrm{k} \Omega$ |

－Electrical characteristics（unless otherwise noted， $\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{cc}}=5.0 \mathrm{~V}$ and $\mathrm{f}=1 \mathrm{kHz}$ ）

| Parameter |  | Symbol | Min． | Typ． | Max． | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating power supply voltage |  | Vcc | 4.5 | 5.0 | 5.5 | V | － |
| Quiescent current |  | la | － | 10 | 15 | mA | Not input，when output off |
| ＜1／2 divider amplifier 1） |  |  |  |  |  |  |  |
| Quiescent input voltage |  | Voo | － | 30 | 100 | mV | $\mathrm{l}_{\mathrm{N}}=0, \mathrm{R}_{\mathrm{L}}=47 \mathrm{k} \Omega$ |
| Input resistance | BA6820F | Rin | 2 | 3 | 4 | $k \Omega$ | － |
|  | BA6822S／F |  | 0.2 | 0.3 | 0.4 |  |  |
| Output voltage | BA6820F | Vout | 600 | 1000 | 1400 | mV | $\begin{aligned} & \mathrm{liv}=-100 \mu \mathrm{~A} \\ & \mathrm{R}_{\mathrm{L}}=47 \mathrm{k} \Omega \end{aligned}$ |
|  | BA6822S／F |  | 500 | 890 | 1400 |  |  |
| Maximum input current |  | lim | 1.2 | 2 | － | mA | － |
| Crosstalk |  | CT | － | 40 | 120 | mV | $\mathrm{V}_{\mathrm{IN}}=1.8 \mathrm{Vrms}$ |
| Differential output voltage |  | $\triangle$ Vout | －250 | 0 | 250 | mV | $\mathrm{lin}=-100 \mu \mathrm{~A}, \mathrm{R}_{\mathrm{L}}=47 \mathrm{k} \Omega$ |
| Output voltage linearity | BA6820F | $\Delta \mathrm{V} / \Delta \mathrm{l}$ | 400 | 650 | 900 | mV | $\mathrm{lin}=-10 \sim-100 \mu \mathrm{~A}$ |
|  | BA6822S／F |  | 350 | 650 | 900 |  |  |
| 〈DC input） |  |  |  |  |  |  |  |
| DC input resistance |  | Rinde | 30 | 50 | 70 | $k \Omega$ | － |
| 〈Oscillator〉 |  |  |  |  |  |  |  |
| Oscillator frequency |  | fosc | 3.5 | 4.0 | 4.5 | kHz | $\mathrm{C}=0.01 \mu \mathrm{~F}, \mathrm{R}=41 \mathrm{k} \Omega$ |
| ＜Display output） |  |  |  |  |  |  |  |
| Peak hold time＊1 |  | Thold | － | 2 | － | $s$ | $\mathrm{fosc}=4 \mathrm{kHz}$ |
| Display switch output duty cycle |  | duty | － | 7／32 | － | － | － |
| Display switch output low level voltage |  | Vol | － | 0.1 | 0.3 | V | $\mathrm{I}_{\mathrm{L}}=10 \mathrm{~mA}, \quad$ pins 12 to 15 |
| Display switch output leakage current |  | Idieak | － | － | 10 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{CE}}=5 \mathrm{~V}$ ，pins12 to15 |
| LED driver output low level voltage |  | Vol | － | 0.3 | 0.7 | V | $\mathrm{Io}_{\mathrm{o}}=30 \mathrm{~mA}$ ，pins16 to21 |
| LED driver output leakage current |  | IDleak | － | － | 10 | $\mu \mathrm{A}$ | $\mathrm{VCE}=5 \mathrm{~V}$ ，pins16 to 21 |
| Mute time at power on |  | Tmute | － | 1 | － | s | fosc $=4 \mathrm{kHz}$ |
| AC／DC switching threshold level |  | VTH6 | 2.2 | 2.5 | 2.8 | V | ＂H＂；AC，＂L＂；DC6pin |
| Forced mute input threshold |  | $V_{\text {TH7 }}$ | 2.2 | 2.5 | 2.8 | V | ＂ $\mathrm{H}^{\prime}$ ：mute（pin 7） |
| Peak hold cancel input threshold |  | $V_{\text {TH8 }}$ | 2.2 | 2.5 | 2.8 | V | ＂ $\mathrm{H}^{\prime \prime}$ ：cancel（pin 8） |
| 〈Comparator〉 |  |  |  |  |  |  |  |
| AC comparator level 12 |  | VC12AC | 8.5 | 10 | 12 | dB | Output：pin 16＊3 |
| AC comparator level 11 |  | VC11ac | 5.5 | 7 | 8.5 | dB | Output：pin 16＊3 |
| AC comparator level 10 |  | VCloac | 3.0 | 4 | 5.5 | dB | Output：pin 17＊3 |
| AC comparator level 9 |  | Vcaac | 1.0 | 2 | 3.0 | dB | Output：pin 17＊3 |
| AC comparator level 8 |  | Vcbac | － | 0 | － | dB | Output：pin 18＊3 |
| AC comparator level 7 |  | Vctac | $-3.0$ | －2 | $-1.0$ | dB | Output：pin 18＊3 |
| AC comparator level 6 |  | $V_{\text {cbac }}$ | $-5.5$ | －4 | －3．0 | dB | Output：pin 19＊3 |
| AC comparator level 5 |  | Vc5ac | －8．5 | －7 | $-5.5$ | dB | Output：pin 19＊3 |
| AC comparator level 4 |  | Vc4ac | －15 | －10 | －8．5 | dB | Output：pin 20＊3 |
| AC comparator level 3 |  | Vc3ac | －25 | －20 | －15 | dB | Output：pin 20＊3 |
| AC comparator level 2 |  | VC2AC | －33 | $-30$ | －25 | dB | Output：pin 21＊3 |
| AC comparator level 1 |  | VC1AC | －55 | $-38$ | －33 | dB | Output：pin 21＊3 |


| Parameter |  | Symbol | Min. | Typ. | Max. | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $A C$ sensitivity | BA6820F | Vinac | 170 | 308 | 450 | $\mathrm{m} \mathrm{V}_{\mathrm{ms}}$ | AC comparator level 8 on level |
|  | BA6822S/F |  | 220 | 400 | 580 |  |  |
| DC comparator level 12 |  | VC12de | 2.78 | 3.10 | 3.44 | V | Output: pin 16*4 |
| DC comparator level 11 |  | $V_{\text {C110 }}$ | 2.35 | 2.64 | 2.90 | V | Output: pin 16*4 |
| DC comparator level 10 |  | $V_{\text {clobc }}$ | 1.98 | 2.23 | 2.48 | V | Output: pin 17*4 |
| DC comparator level 9 |  | Vcgoc | 1.75 | 2.00 | 2.25 | V | Output: pin 17*4 |
| DC comparator level 8 |  | V 6800 | 1.50 | 1.76 | 2.00 | V | Output: pin 18*4 |
| DC comparator level 7 |  | Vcroc | 1.35 | 1.58 | 1.85 | V | Output: pin 18*4 |
| DC comparator level 6 |  | V 6800 | 1.20 | 1.42 | 1.65 | $V$ | Output: pin 19*4 |
| DC comparator level 5 |  | VCSDC | 1.00 | 1.18 | 1.35 | $V$ | Output: pin 19*4 |
| DC comparator level 4 |  | VCade | 0.80 | 1.00 | 1.15 | V | Output: pin 20*4 |
| DC comparator level 3 |  | $V_{C 300}$ | 0.35 | 0.51 | 0.70 | V | Output: pin 20*4 |
| DC comparator level 2 |  | Vczod | 0.15 | 0.25 | 0.35 | V | Output: pin $21 * 4$ |
| DC comparator level 1 |  | Vadc | 0.05 | 0.11 | 0.20 | V | Output: pin 21*4 |

* 1 Peak hold for comparator levels for 5 to 12. There is no peak hold for DC input.
* 2 Display switch output duty cycle.


Display switch output duty cycle $=\mathrm{T} 1 / \mathrm{T}_{2}$

* 3 The reference value for the AC comparator is the level of comparator 8 .
* 4 Only ch2 is effective in DC input mode.
- Measurement circuit

$R 1$ and $R 3$ are $0 \Omega$ for the BA6820F, and $2.7 \mathrm{k} \Omega$ for the BA6822S.

Fig. 1

## -Application example



Fig. 2
-Attached components

- $\mathrm{C}_{1}$ and $\mathrm{C}_{2}: 1 \mu \mathrm{~F}$ input coupling capacitors (electrolytic) for the square-root compression amplifier. The rectified voltage output from pins 2 and 4 is attenuated by 1.5 dB at the frequency determined by the following formula (refer to the data for the relationship between the pin 2 and 4 DC voltage frequency characteristics and the LED lighting level).

$$
\mathrm{fC}=\frac{1}{2 \pi \cdot \mathrm{C}_{1}\left(\mathrm{RiN}_{\mathrm{I}}+\mathrm{VR}_{1}\right)} \quad\left(\mathrm{RiN}_{\mathrm{I}}=\text { input resistance }\right)
$$

When the LEDs light due to low-frequency mechanical noise or induced hum, reduce the values of $\mathrm{C}_{1}$ and $\mathrm{C}_{2}$ to cut the cut the low-frequency frequency characteristic.

- VR1 and $\mathrm{VR}_{2}$ : $5 \mathrm{k} \Omega$ to $100 \mathrm{k} \Omega$ potentiometers for sensitivity and L/R balance adjustment (refer to item 2 of Circuit operation).
- $\mathrm{C}_{3}$ and $\mathrm{C}_{4}: 22 \mu \mathrm{~F}$ electrolytic capacitors.
- $\mathrm{R}_{1}$ and $\mathrm{R}_{2}: 47 \mathrm{k} \Omega$ carbon-film resistors.

These resistors set the attack and release times (refer to item 6 of the Circuit operation).

- C5: $0.01 \mu \mathrm{~F}$ mylar capacitor.

R3: $39 \mathrm{k} \Omega$ carbon-film resistor.
This resistor sets the oscillator frequency (refer to the Operation notes).

- $\mathrm{C}_{6}: 100 \mu \mathrm{~F}$ electrolytic capacitor. This capacitor is used as a ripple filter. The positive lead should be connected close to Vcc (pin 22), and the negative lead close to the junction of pins 10 and 11.

Circuit operation
(1) Function

There are 12 LEDs each for the left and right channels, and these are divided into 4 groups of six. A dynamicdrive technique is used to drive the LEDs in order, and provide 12 display points for each channel. A 12-point VU -scale bar display is produced over the display range -38 dB to +10 dB . The top eight points have a peak hold function which may be cancelled if required (there is no peak hold function for DC mode).
Independent AC (two channels) and DC (one channel) inputs are provided, and a control input pin is used to switch between them.
In DC mode, only channel 2 operates.
A power-on mute circuit is provided, so no measures are required to prevent erroneous lighting of the LEDs when power is applied.

- Q1 to Q4: 2SA854S Transistors for display switching. If the current for one LED is Iled, the values of Ic and $\mathrm{P}_{\mathrm{c}}$ for each of $Q_{1}$ to $Q_{4}$ when all LEDs are lit are given by:

$$
\begin{aligned}
& \mathrm{I}_{\mathrm{c}}=6 \times \mathrm{I} \text { Led }(\text { for duty cycle of } 7 / 32) \\
& \mathrm{P}_{\mathrm{c}}=6 \times \mathrm{I}_{\text {Led }} \times \mathrm{V}_{\text {sat }} \times \frac{7}{32}
\end{aligned}
$$

Choose transistors that have sufficient margin for Ic and Pc.

- $\mathrm{R}_{4}$ to $\mathrm{R}_{7}$ : $2.2 \mathrm{k} \Omega$ carbon-film resistors.

These resistors limit the base current for transistors $Q_{1}$ to $Q_{2}$. The base current is the collector current divided by $h_{\text {FE, }}$ but as the transistors are operated in the saturation region, hfe will drop considerably. Choose the resistors so that sufficient base current flows.

- LED ${ }_{1}$ to LED $_{12}$ : SLB-26
- $\mathrm{R}_{8}$ to $\mathrm{R}_{13}: 330 \Omega$ carbon-film resistors. These resistors set the LED current. Four LEDs are connected to each resistor, but as the LEDs light sequentially, only the current for one LED flows in a resistor.

$$
R 8 \sim 13=\frac{V_{\text {cc }}-V_{\text {LED }}-V_{\text {sat }}\left(Q_{1}-Q_{4}\right)-V_{\text {DL }}}{I_{\text {LED }}}
$$

VDL = Display switch output "L" voltage.

The attack and release times for the display can be set using external components.
(2) Input block (pins 1, 3 and 5)

Pins 1 and 3 are the $A C$ inputs, and pin 5 is the DC input. Pin 6 is used to switch between AC and DC input. When the pin 6 input is " H ", AC input is selected (pins 1 and 3 ), and when it is " $L$ ", DC input (pin 5 ) is selected. Therefore, if pins 1 and 3 are used for audio input, and pin 5 as the input for signal meter output (DC) from a tuner, the switch input (pin 6) can be used to make LED display do two jobs.
Pins 1 and 3 have low input impedance, so we recommend that you connect potentiometers (VR1 and VR2: $5 \mathrm{k} \Omega$ to $100 \mathrm{k} \Omega$ ) in series with the inputs to adjust the balance between ch1 and ch2 and set the levels.
(3) Peak hold

The peak hold circuit momentarily holds the maximum signal level in AC input mode. The peak hold function applies to comparator levels 5 to 12 ( 8 points). When the oscillator frequency is 4 kHz , the peak hold time is two seconds (Typ.), and can be changed by changing the oscillator frequency. The peak hold function can be turned off by setting pin 8 high.
The peak hold function does not operate in DC input mode.
(4) Output block

The two sets of 12 LEDs for the BA6820F and BA6822S/F are divided into 4 groups of six. A dynamicdrive technique is used to drive the LEDs in order, and display the input level.

Group 1
Channel 1 LEDs 1, 3, 5, 7, 9 and 11
(odd-numbered LEDs)
Group 2
Channel 1 LEDs 2, 4, 6, 8, 10 and 12
(even-numbered LEDs)
Group 3
Channel 2 LEDs 1, 3, 5, 7, 9 and 11
(odd-numbered LEDs)
Group 4
Channel 2 LEDs 2, 4, 6, 8, 10 and 12
(even-numbered LEDs)
Therefore, six LED drive outputs, and four display switch terminals are provided.
The output timing chart for when all LEDs are lit is shown in Fig. 3.
In DC input mode, only channel 2 operates.
$D_{1}$ to $D_{4}$ in the timing chart operate in the same way as for AC mode (i.e. $\mathrm{D}_{1}$ and $\mathrm{D}_{2}$ operate), but $\mathrm{O}_{1}$ to $\mathrm{O}_{6}$ are high for the channel 1 period.

1) Display switch outputs (pins 12 to 15)

To successively switch between the four LED groups, these outputs are active low (see Fig. 4).
The outputs are pulled up by $36 \mathrm{k} \Omega$ resistors (between the base and emitter of $Q_{1}$ to $Q_{4}$ ), so leak current bypass resistors are not required.


Timing chart for all LEDs on ( $\mathrm{f}=4 \mathrm{kHz}$ )
Fig. 3


Fig. 4
2) LED drive outputs (pins 16 to 21)

These are open-collector outputs that drive the LEDs (see Fig. 5).
One complete cycle of the timing chart starting from $\mathrm{O}_{1}$ (pin 21) is as follows: if the channel 1 input is higher than comparator level 1 during the first-point period for channel 1 , a " $L$ " level is output. Next, if the channel 1 input is higher than comparator level 2 during the second-point period for channel 1 , a " $L$ " level is output, if the channel 2 input is higher than comparator level 1 during the firstpoint period for channel 2 , a " $L$ " level is output, and then if the channel 2 input is higher than comparator level 2 during the second-point period for channel 2 , a "L" level is output.This cycle is then repeated for outputs $\mathrm{O}_{2}$ to $\mathrm{O}_{6}$.
(5) Mode switch pins (pins 6, 7 and 8)

- AC/DC mode switch pin (pin 6)

When "H" AC input (pins 1 and 3 ) is selected, and when "L" DC input (pin 5) is selected.
"H": 1, 3pin "L": 5pin

- Mute (pin 7)

When "H", mute is applied, and no display is output. Normally "L".

- Peak hold (pin 8)

When " H ", peak hold for AC input is cancelled.
The circuit for these three mode switches is given in Fig. 6.

Pins 6 to 8 are pulled down by a $50 \mathrm{k} \Omega$ resistor, so there is no need for the inputs to be connected to GND for a " L " input.
(6) Attack and release times (pins 2 and 4)

The output response to AC input signals is set by the RC circuits connected to pins 2 and 4 ( $\mathrm{R}_{1}$ and $\mathrm{C}_{3}$ for channel 1 , and $\mathrm{R}_{2}$ and $\mathrm{C}_{4}$ for channel 2).

| $\mathbf{R}$ | C | Attack time | Release time |
| :---: | :---: | :---: | :---: |
| $47 \mathrm{k} \Omega$ | $22 \mu \mathrm{~F}$ | 14 ms | 2.7 s |
| $10 \mathrm{k} \Omega$ | $10 \mu \mathrm{~F}$ | 5 ms | 0.3 s |

Attack time: the time for pins 2 and 4 to go from 0 V to 2 V (when all 12 LEDs are lit) when the input signal goes from no signal to a 1.8 V input signal with $\mathrm{f}=1 \mathrm{kHz}$ (the input level at which the 12th LED lights is +6 dB ).

Release time: the time for pins 2 and 4 to go from 2 V to 120 mV (the level at which the first LED lights) when the input signal goes from a 1.8 V input signal to no signal (the input level at which the 12th LED lights is +6 dB ).
The recommended values are $10 \mathrm{k} \Omega$ to $47 \mathrm{k} \Omega$ for $\mathrm{R}_{1}$ and $\mathrm{R}_{2}$ and $10 \mu \mathrm{~F}$ to $22 \mu \mathrm{~F}$ for $\mathrm{C}_{3}$ and $\mathrm{C}_{4}$. In DC mode, the attack and release times are zero seconds regardless of the RC circuits (see Fig. 7).
(7) Oscillator frequency (pin 9)
$\mathrm{R}_{3}$ and $\mathrm{C}_{5}$ connected to pin 9 determine the oscillator frequency.
The relationship between the oscillator frequency and the display switching period is as follows:


The LED drive output, display switch output period, peak hold time, and mute time when power is applied are all determined by the oscillation frequency, and are varied when the frequency is changed.


Fig. 5


Fig. 6


Fig. 7
(8) GND terminal (pins 10 and 11)

Pin 11 is the GND for the pin 12 to pin 21 driver circuits, and pin 10 is the GND for the other signal circuits (see Fig. 8).
Pins 10 and 11 are not connected internally, so both must be connected to GND.
The large LED current that flows from pin 11 has a large high-frequency component, so care must be taken with regard to the PCB layout to avoid faulty operation of the IC, and incorrect lighting of the LEDs due to noise.

## Operation notes

(1) Use the potentiometers VR1 and VR2 connected to pins 1 and 3 to set the 0dB levels and adjust the channel 1 and channel 2 balance.
(2) The recommended oscillator frequency is 4 kHz (pin 9). The external component values for $\mathrm{C}_{5}$ and $\mathrm{R}_{3}$ are $0.01 \mu \mathrm{~F}$ and $41 \mathrm{k} \Omega$ (or $39 \mathrm{k} \Omega$ ). If you wish to use other values, they should be in the ranges $0.001 \mu \mathrm{~F}$ to $0.1 \mu \mathrm{~F}$ and $10 \mathrm{k} \Omega$ to $100 \mathrm{k} \Omega$. (see Fig. 9).
(3) The dynamic ranges for the BA6820F and BA6822S/F comparators are large, and the comparators will operate even with low-level signals.
For this reason, noise may cause the LEDs to light when they should not, and to prevent this, GND line noise should be kept sufficiently low.
(4) Take care to ensure the none of the pins have a voltage applied to them that is higher than Vcc or lower than GND (e.g. supply the LED current from a source other than $V_{c c}$ ).
(5) In general, the level meter output block is switched on and off in the range GND to Vcc, and a large voltage with a large high-frequency component is generated.
Therefore, if the input and output lines are close, the out-


Fig. 8
put may be fed back to the input and cause oscillation, incorrect LED lighting, or faulty operation of the IC. Design the PCB artwork so that there is no coupling between the input and output lines or with other units, in the same way as for linear amplifier PCB design. If they must be physically close, insert a capacitor between the input and GND to reduce the AC input impedance, and prevent noise from entering the input.


Fig. 9

- Electrical characteristic curves


Fig. 10 Frequency characteristics


Fig. 13 Output saturation voltage vs. sink current (display switch output)


Fig. 11 Pin 2 and pin 4 DC output voltage vs. pin 1 and pin 3 $A C$ input voltage


Fig. 12 Output saturation voltage vs. sink current (LED output)


