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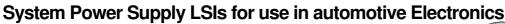
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Multifunction System Power Supply IC (with Watchdog Timer) BD4911FM



•Description

The BD4911FM multiple-output system power supply features microcontroller output and is capable of operating at super-low dark current levels. With a built-in reset, with a microcontroller delay, BATTERY/ACCESSORY voltage detection, and mute function, WDT (Watch Dog Timer)function, the BD4911FM is ideal for car audio and satellite navigation systems.

Features

1) Main regulator Vo1 can be switched between 5.0 V and 3.3 V output using the SEL input voltage, and can be used with an external boost transistor.

Sub-regulator Vo2 generates 3.3 V output.

Both regulators feature an output current of 150 mA.

- Built-in WDT reset; monitor time can be set with the CTW capacitor. WDT operation can be switched on and off using INH input.
- 3) Circuit current during standby operation (WDT off): 100 μ A (TYP)
- 4) Built-in reduced power detection circuits for Vo1 and Vo2 output
- 5) Built-in elevated output detection circuit for Vo1
- 6) The IC monitors the BuM voltage and outputs it to the BuDET1 and BuDET2 pins.
- 7) The IC monitors the ACCM voltage and outputs it to the ACCM pin when BuDET1 and BuDET2 are low.
- 8) Built-in mute circuit. Pulse width can be set with the CE capacitor.
- 9) Built-in power-on reset function for which RESET output can be set with the CTP capacitor.
- 10) Vo1 generates low-saturation voltage type PMOS output.
- 11) Built-in overcurrent protection circuit
- 12) Built-in overvoltage protection circuit
- 13) Built-in thermal shutdown circuit
- 14) The 28-pin HSOP-M28 package is ideal for space-saving designs.

Applications

Car audio and satellite navigation systems

•Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol	Limits	Unit
Supply Voltage 1	VIN1	-0.3 to 36	V
Supply Voltage 2	VIN2	-0.3 to 36	V
Power Dissipation	Pd	2200 (*1)	mW
Operating Temperature Range	Topr	-40 to +85	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Peak Supply Voltage 1	VIN1 Peak	50 (*2)	V
Peak Supply Voltage 2	VIN2 Peak	50 (*2)	V
Surge Applied Current 1	IACC (S+)	+3 (*2)	mA
Surge Applied Current 2	IACC (S-)	-12 (*3)	mA
Maximum Junction Temperature	Tjmax	150	°C

*1: When mounted on a PCB (70 mm \times 70 mm \times 1.6 mm glass epoxy).

*2: tr \ge 1 ms, Bias voltage/current is less than 200 ms or shorter

*3: tr \geq 1 ms, Bias voltage/current 60 ms or shorter

•Recommended operating ranges (Ta = 25°C)

Parameter	Symbol		Limits	Unit	Comment	
Farameter	Symbol	Min.	Typ. Max.		Unit	Comment
Recommended Power	VIN1	Vo1+1.2	13.2	16.0	V	When using built-in transistor.
	VINIA	Vo1 + 0.5 +	40.0	10.0	V	When using external boost
Supply Voltage Range 1	VIN1	eternal TrVBE	13.2	16.0	v	transistor.
Recommended Power	VIN2	4.5	13.2	16.0	V	
Supply Voltage Range 2	VIINZ	4.5	13.2	10.0	v	
Recommended Power	Vo1	1.2	—	5.2	V	DESET DET4
Supply Voltage Range 3	VOT					RESET, DET1
Recommended Power	Vo1	2.5		5.0	V	BuDET, ACCDET, MUTE,
Supply Voltage Range 4	VOT	2.5		5.2	v	WDT
Recommended Power	Vo2	1.2		3.4	V	DET2
Supply Voltage Range 5	V02	1.2	—	3.4	v	DETZ

*Electrical characteristics are not guaranteed (especially when operating on reduced voltage).

•External element setting time (Ta = 25°C)

Parameter	Symbol	Equation (TYP)	Unit	Recommendation	Unit	Condition
Runaway Operation Detection Time 1	TWD1	$700\times CTW~(\mu F)$	mS	0.1 to 2.2	μF	SEL > 1.5 V
Runaway Operation Detection Time 2	TWD2	$462\times CTW~(\mu F)$	mS	0.1 to 2.2	μF	SEL < 1.0 V
Runaway Operation Reset Time 1	TWR1	$340 \times CTW \ (\mu F)$	mS	0.1 to 2.2	μF	SEL > 1.5 V
Runaway Operation Reset Time 2	TWR2	$260 \times CTW (\mu F)$	mS	0.1 to 2.2	μF	SEL < 1.0 V
Power-on Reset Time	TPR	$1000 \times CTP (\mu F)$	mS	0.047 to 0.22	μF	
CD Delay Time	Td	$100 \times CD \; (\mu F)$	mS	0.047 to 0.22	μF	
MUTE Pulse Width	Tm	$1 \times CE (\mu F)$	S	0.1 to 2.2	μF	
BuM Detection Voltage (rising)	VTHBu	1.252 × (R1+R2)/R2	V	R1: 0.5 to 2	MΩ	
BuM Detection Voltage (falling)	VTLBu	1.184 × (R1+R2)/R2	V	R2: 0.1 to 0.5	1015.2	
ACCM Detection Voltage (rising)	VTHACC	1.252 × (R3 + R4)/R4	V	R3: 50 to 200	kΩ	
ACCM Detection Voltage (falling)	ATLACC	1.184 × (R3+R4)/R4	V	R4: 10 to 50		

•Electrical characteristics (Unless otherwise specified, Ta = 25°C, VIN1 = VIN2 = 13.2 V)

Parameter	Symbol	N.41	Limits		Unit	Condition
	,	Min.	Тур.	Max.		
[Overall]	11/11/14	65	05	105	^	
VIN1 Supply Current	IVIN1	65	95	125	μA	
VIN2 Supply Current	IVIN2	5	10	20	μA	
Total Supply Current	IVINA	65	100	135	μA	ACC = 0 V, WDINH = H
Over veltage Detection Veltage	IVINB	65	130	195	μΑ	ACC = 13.2 V
Overvoltage Detection Voltage	VOVP	28	31	34	V	All regulator output off
Overvoltage Detection Hysteresis	VOVPHY	0.5	1	1.5	V	All regulator output reset
Width						
Main Regulator (REG1)]						$\lambda (1) (4 - 0.0) (4 - 40) (4 $
VO1 Output Voltage 1	VO1-1	4.8	5.0	5.2	V	VIN1 = 6.2 V to 16 V,
VO1 Line Regulation 1	∆VO1I-1		1	30	m\/	lo1 = 0 mA to 150 mA, SEL > 1.5 V
			1		mV	VIN1 = 6.2 V to 16 V, SEL > 1.5 V
VO1 Load Regulation 1	∆VO1L-1		7	50	mV	lo1 = 0 mA to 150 mA, SEL > 1.5 V
VO1 Output Voltage 2	VO1-2	3.168	3.3	3.432	V	VIN1 = 4.5 to 16 V, Io1 = 0150 mA,
VO1 Line Degulation 2			1	20	m\/	SEL < 1.0 V
VO1 Line Regulation 2	∆VO1I-2		1	30	mV	VIN1 = 4.5 V to 16 V, SEL < 1.0 V
VO1 Load Regulation 2	∆VO1L-2		7	30	mV	lo1 = 0 mA to 150 mA, SEL < 1.0 V
Minimum VO1 Output Short Protection Start Current	VO1-L	2.5	400	-	V	VIN1 = 3.0 V, Io1 = 0 mA
	lo1max	150	400	600	mA	$f_{\rm m} = 420 \text{Jm} = 40 \text{d} \text{D} \text{/} \text{Jm} = 450 \text{m} \text{A}$
Power Supply Ripple Rejection Ratio	RRV01	45	55		dB	fin = 120 Hz, -10 dBV, lo1 = 150 mA
VO1 Sink Current	IVO1in	35	90	145	μA	Vo1 = 5 V, VIN1 = ACC = OPEN,
						WDINH = SEL = 5 V
External Boost Transistor Current-li			0.4	1.0	•	
OCP Input Current		0	0.1	1.0	μΑ	VOCP = VIN1 = 16 V
OCP Detection Voltage	VOCP1	360	400	440	mV	Voltage differential with VIN1
OCP Detection Voltage (During	VOCP2	20	32	50	mV	Vo1 = 0 V, voltage differential with
Output Ground Fault) Elevated Output Detection Circuit (VIN1
	/*	E 20	E 40	F 69	V	
Elevated Output Detection Voltage 1 Elevated Output Detection Voltage 2		5.30 3.5	5.49 3.62	5.68 3.75	V V	SEL > 1.5 V SEL < 1.0 V
· · ·	VCVER2				V	VO1 > 5.68 V, lo = 100 μA
Elevated Output Detection Output	VCOMP	—	0.1	0.4	v	
Output Off Delay Time	TmVoff		—	50	μS	Vo1: 3.1→4.8 V (tr = 0.01 V/μS) VIN1 = 4.8 V, Ro = 1 kΩ
						$VINT = 4.6 V, RO = T K\Omega$
Sub-regulator (REG2)]						V(N) = V(N) = 4.5 V(to 16) (102 - 100)
VO2 Output Voltage	VO2	3.168	3.3	3.432	V	VIN1 = VIN2 = 4.5 V to 16V, lo2 =
VO2 Line Degulation			1	20	m\/	0 mA to 150 mA,
VO2 Line Regulation			1 7	30	mV	VIN1 = VIN2 = 4.5 V to 16V
VO2 Load Regulation	∆VO2L		1	30	mV	lo2 = 0 mA to 150 mA
Minimum VO2 Output	VO2-L	2.5	400	-		VIN1 = VIN2 = 3.0 V, lo2 = 0 mA
Short Protection Start Current	lo2max	150	400	600	mA dD	$f_{\rm D} = 120 \text{Hz} = 10 \text{d} \text{D} (102 = 150 \text{m})$
Power Supply Ripple Rejection Ratio	RRVO2	45	55	—	dB	fin = 120 Hz, -10 dBV, lo2 = 150 mA
Regulator Voltage Selection Circuit		1 20	1.05	1 20	V	
SEL Threshold SEL Input Current	VTHSEL	1.20	1.25	1.30		
	ISEL	1	2	4	μA	VSEL = 5 V
VO1 Reduced-voltage Detection Ci			A 4 -	4.00	14	λ (a) folling $\Omega \Sigma \to 4 \Sigma \lambda$
VO1 Detection Voltage 1	VTLP1-1	4.00	4.15	4.30	V	Vo1 falling, SEL > 1.5 V
Reset Voltage 1	VTHP1-1	4.10	4.35	4.60	V	Vo1 rising, SEL > 1.5 V
Hysteresis Width 1	VHSP1-1	0.1	0.2	0.3	V	SEL > 1.5 V
VO1 Detection Voltage 2	VTLP1-2	2.85	2.95	3.05	V	Vo1 falling, SEL < 1.0 V
Reset Voltage 2	VTHP1-2	2.92	3.09	3.26	V	Vo1 rising, SEL < 1.0 V
Hysteresis Width 2	VHSP1-2	0.07	0.14	0.21	V	SEL < 1.0 V
DET1 Output On Resistance	RDET1		270	600	Ω	IDET1 = 1 mA
DET1 Output Saturation Voltage	VDET1L	—	0.1	0.4	V	IDET1 = 2 μA, Vo1 = 1.2 V

• This IC is not designed to be radiation-resistant.

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$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$	ile charging).		
Hysteresis WidthVHSP20.070.140.21VDET2 Output On ResistanceRDET2-270600 Ω IDET2 = 1 mADET2 Output Saturation VoltageVDET2L-0.10.4VIDET2 = 2 μ A, Vo1 = 1.2[Power-on Reset Timer (CTP, RESET)]CTP Charge Resistance 1RCTP10.60.91.2M Ω When RESET is low (whiCTP Charge Resistance2RCTP25.58.511.5 $k\Omega$ When RESET is high (aft is complete).CTP Rising ThresholdVTHP13.003.333.66VSEL > 1.5 VVTHP21.982.22.42VSEL < 1.0 V	ile charging).		
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	ile charging).		
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	ile charging).		
$[Power-on Reset Timer (CTP, RESET)]$ $CTP Charge Resistance 1 RCTP1 0.6 0.9 1.2 M\Omega When RESET is low (whi CTP Charge Resistance2 RCTP2 5.5 8.5 11.5 k\Omega When RESET is high (aft is complete). VTHP1 3.00 3.33 3.66 V SEL > 1.5 V VTHP2 1.98 2.2 2.42 V SEL < 1.0 V$	ile charging).		
$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$			
CTP Charge Resistance2 RCTP2 5.5 8.5 11.5 kΩ When RESET is high (aft is complete). CTP Rising Threshold VTHP1 3.00 3.33 3.66 V SEL > 1.5 V VTHP2 1.98 2.2 2.42 V SEL < 1.0 V			
CTP Charge Resistance2 RCTP2 5.5 8.5 11.5 KΩ is complete). CTP Rising Threshold VTHP1 3.00 3.33 3.66 V SEL > 1.5 V VTHP2 1.98 2.2 2.42 V SEL < 1.0 V	ter charging		
CTP Rising Threshold VTHP2 1.98 2.2 2.42 V SEL < 1.0 V			
VIHP2 1.98 2.2 2.42 V SEL < 1.0 V			
OTD Folling Threshold VTLP1 1.50 1.67 1.83 V SEL > 1.5 V			
CTP Falling Threshold VTLP2 0.9 1.07 1.03 V OEL > 1.0 V			
Power-on Reset Time TPR 60 100 140 mS CTP = $0.1 \mu\text{F}$			
Reset On Delay Time TDR 10 50 100 μ S CTP = 0.1 μ F			
	IRST = 1 mA		
RESET Low Output Voltage VRSTL — — 0.4 V IRST = 2 μ A, Vo1 = 1.2 V	/		
[Watchdog Timer (WDT, WDINH, CTW)]			
Runaway Operation DetectionTWD1420700980mS $CT = 1 \ \mu F$, VO1 = 5 VTime from	n final WDT out to RESET		
Runaway Operation DetectionTWD2277.2462.0646.8mS $CT = 1 \mu F$, VO1 = 3.3 Vinversion			
Runaway Operation Reset Time 1 TWR1 204 340 476 mS $CT = 1 \mu F$, VO1 = 5 V			
Runaway Operation Reset Time 2 TWR2 156 260 364 mS CT = 1 μ F, VO1 = 3.3 V			
CTW Charge Current (Source) IHCTW 3.3 4.7 6.1 µA			
CTW Discharge Current (Sink) ILCTW 3.4 4.8 6.2 µA			
CTW/ Papid Discharge Resistance RCTWon1 — 16 50 Ω When WDT signal input i	s active.		
CTW Rapid Discharge Resistance $RCTWon2 - 0.5 2.5 k\Omega$ ACC: L, WDINH: H			
VTHW1 3.00 3.33 3.66 V SEL > 1.5 V CTW Rising Threshold VTHW1 4.00 9.00 9.10 V//// SEL > 1.5 V			
VTHW2 1.98 2.20 2.42 V SEL < 1.0 V			
CTW/ Falling Threshold VTLW1 1.50 1.67 1.83 V SEL > 1.5 V			
CTW Falling Threshold VTLW1 1.00 1.01 1.00 0			
VTHWDT1 3.00 3.33 3.66 V SEL > 1.5 V			
WDT Rrising Threshold VTHWDT1 0.00 0			

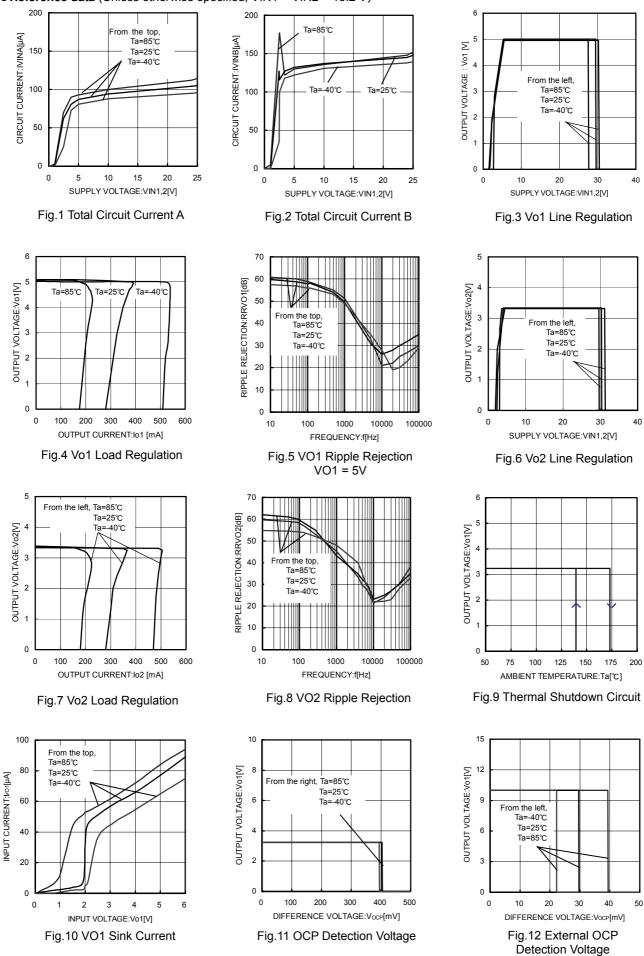
● This IC is not designed to be radiation-resistant.

•Electrical characteristics (Unless otherwise specified, Ta = 25°C, VIN1 = VIN2 = 13.2 V)

•Electrical characteristics (Unles		specifieu,		C, VINT		15.2 V)
Parameter	Symbol		Limits		Unit	Condition
	-	Min.	Typ.	Max.		
WDT Falling Threshold	VTLWDT1	1.50	1.67	1.83	V	SEL > 1.5 V
	VTLWDT2	0.9	1.0	1.1	V	SEL < 1.0 V
WDT Input Current	IWDT	0.5	1.0	2.0	μΑ	VWDT = 5 V, SEL > 1.5 V
WDT Ddge Pulse Width	TPULSE	100	190	300	μS	
WDINH Input Voltage At High Level	VIH	0.8X Vo1			V	
WDINH Input Voltage Low Level	VIL			0.3X Vo1	V	
WDT Input Current	IWDINH	1	2	4	μA	VWDINH = 5 V, SEL < 1.0 V
[Bu Voltage Detection Circuit (BuM,	BuDET)]					
BuM Detection Voltage (Rising) BuDET: H→L, AccDET: H→L	VTHB	1.214	1.252	1.290	V	IC without heat sink
BuM Detection Voltage (Falling) BuDET: L→H, AccDET: L→H	VTLB	1.148	1.184	1.220	V	IC without heat sink
BuDET1 High Output Voltage	VBDTH1	Vo1 -0.4	Vo1 -0.2	_	V	lout = -5 mA
BuDET1 Low Output Voltage	VBDTL1	_	0.15	0.40	V	lout = 5 mA
BuDET2 High Output Voltage	VBDTH2	Vo2 -0.4	Vo2 -0.2		V	lout = -5 mA
BuDET2 Low Output Voltage	VBDTL2	0.4 —	0.15	0.40	V	lout = 5 mA
· •	IBM1	0	4	110	nA	BuM = 1 V
BuM Input Current	IBM1	0	20	110	nA	BuM = 2 V
[MUTE One-shot Pulse Generation	1		20	110	10.0	
	Tm1	0.6	1.0	1.4	S	CE = 1 μF, SEL > 1.5 V
MUTE Pulse Width	Tm2	0.6	1.0	1.4	S	$CE = 1 \ \mu F, SEL < 1.0 \ V$
	Td1	0	5	10	μS	$CE = 1 \ \mu F, SEL > 1.5 \ V$
MUTE Pulse On Dalay Time	Td2	0	8	16	μS	$CE = 1 \ \mu F, SEL < 1.0 \ V$
						When MUTE is on (resistance while
CE Charge Resistance 1	RTM1	0.7	1.0	1.3	MΩ	charging).
CE Charge Resistance 2	RTM2	8.0	11.5	15.0	kΩ	When MUTE is off (resistance when stabilized after charging).
CE Rapid Discharge Resistance	RCEon		4	20	Ω	Must satisfy Td.
CE Output Saturation Voltage	VCEL		0.1	0.3	V	CE output on, ICE = $0 \mu A$
CE Output Saturation Voltage	VCLL VTHCE1	3.00	3.33	3.66	V	SEL > 1.5 V
CE-CMP Threshold (rising)	VTHCE1	1.98	2.2	2.42	V	SEL < 1.0 V
	VTLCE1	1.50	1.67	1.83	V	SEL > 1.5 V
CE-CMP Threshold (falling)	VTLCE2	0.9	1.07	1.00	v	SEL < 1.0 V
MUTE Output Saturation Voltage	VMUTEL		0.2	0.4	V	IMUTE = 5 mA
[ACC Voltage Detection & Delay Cir		CD ACC		0.1	•	
ACCM Detection Voltage (Rising)	VTHA	1.214	1.252	1.290	V	IC without heat sink, BuM = H
ACCM Detection Voltage (Falling)	VTLA	1.148	1.184	1.220	V	IC without heat sink, BuM = H
ACCM Positive Clamp Voltage	VHACC	8	11	14	V	ACCM = +5 mA
ACCM Negative Clamp Voltage	VLACC	-0.30	-0.15	0	V	IACCM = -12 mA
ACCM Input Current 1	IACC1	-5	-1	0	μA	ACCM = 0 V
ACCM Input Current 2	IACC2	0	10	110	nA	ACCM = 2 V
· ·	TdLH	6	10	14	mS	$CD = 0.1 \mu\text{F}$
CD Delay Time	TdHL	6	10	14	mS	CD = 0.1 μF
CD Charge Resistance	ICDH	60	90	120	kΩ	
CD Discharge Resistance	ICDL	60	90	120	kΩ	
	VTHCD1	3.00	3.33	3.66	V	SEL > 1.5 V
CD-CMP Threshold (Rising)	VTHCD2	1.98	2.2	2.42	V	SEL < 1.0 V
CD-CMP Threshold (Falling)	VTLCD1	1.50	1.67	1.84	V	SEL > 1.5 V
	VTLCD2	0.9	1.0	1.1	V	SEL < 1.0 V
 ACCDET Output Saturation Voltage This IC is not designed to be radia 	VADTL	—	0.2	0.4	V	IADT = 5 mA

● This IC is not designed to be radiation-resistant.

•Reference data (Unless otherwise specified, VIN1 = VIN2 = 13.2 V)



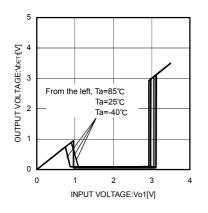


Fig.13 Vo1 Output Detection Voltage

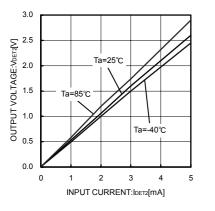


Fig.16 DET2 Output Voltage

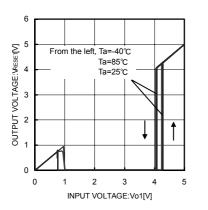
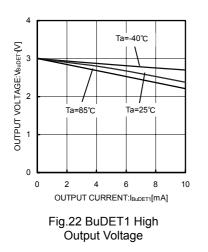


Fig.19 RESET Threshold Voltage



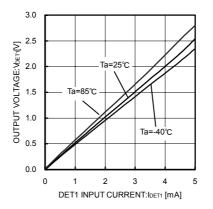


Fig.14 DET1 Output Voltage

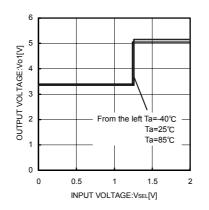


Fig.17 SEL Threshold Voltage

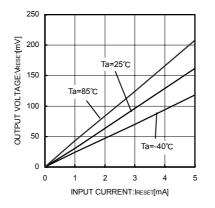
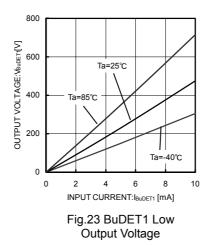


Fig.20 RESET Output Voltage



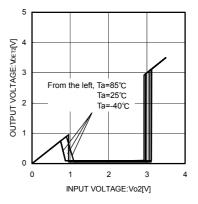


Fig.15 Vo2 Output Detection Voltage

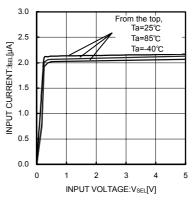


Fig.18 SEL Input Current

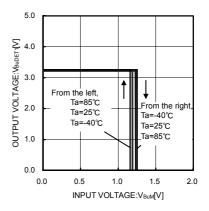


Fig.21 BuDET1 Detection Voltage

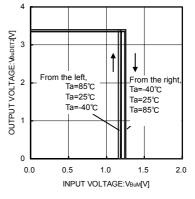


Fig.24 BuDET2 Detection Voltage

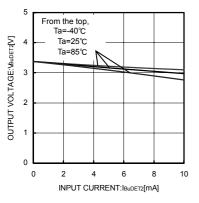


Fig.25 BuDET2 High Output Voltage

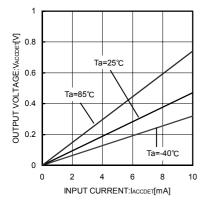
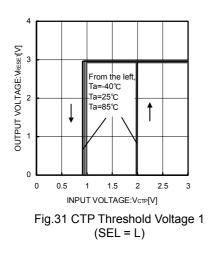
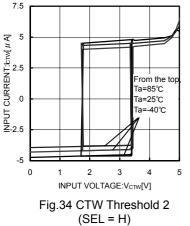


Fig.28 ACCDET Output Voltage





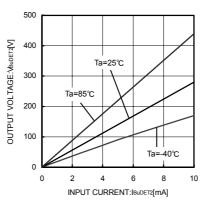


Fig.26 BuDET2 Low Output Voltage

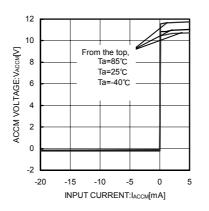
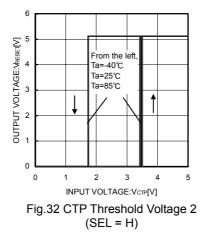
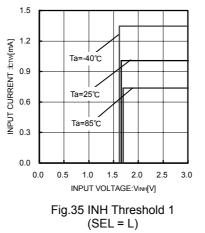


Fig. 29 ACCM Positive/Negative Clamp Voltage





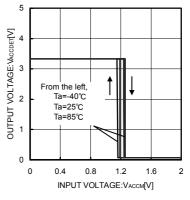


Fig.27 ACCDET Detection Voltage

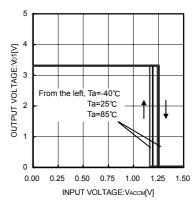
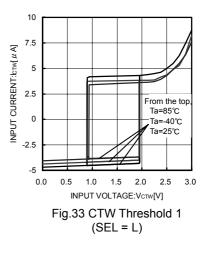
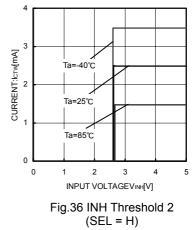


Fig.30 ACC Detection Voltage





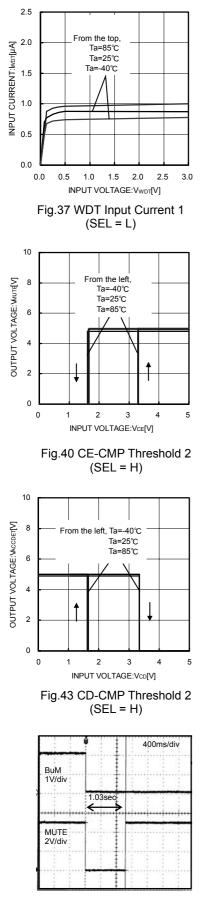
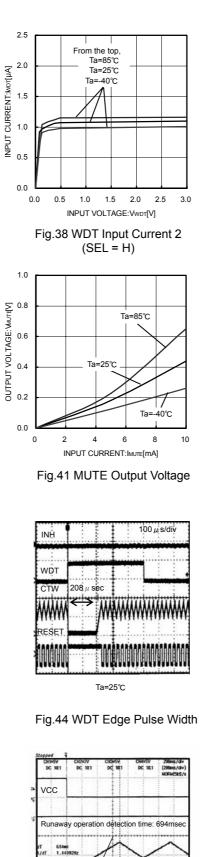
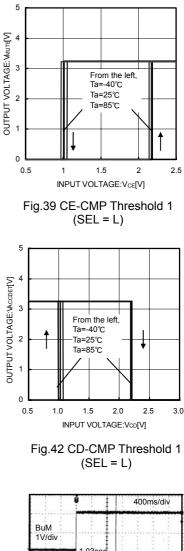




Fig.46 MUTE Timer Time (High→Low)

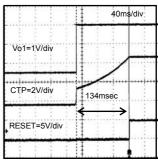




BuM 1V/div			 	
····	1.03sec ←→	. 	 •••[••	•••••••
MUTE 2V/div			 	
				-

Ta=25℃

Fig.45 MUTE Timer Time (Low→High)



Ta=25℃

Fig.48 Power-on Reset Time

RESET Time

Run

←

Ta=25℃

Fig.47 Runaway Operation

way operation

reset time: 340 mse

СТ

RESET

Block diagram, IC package & timing chart

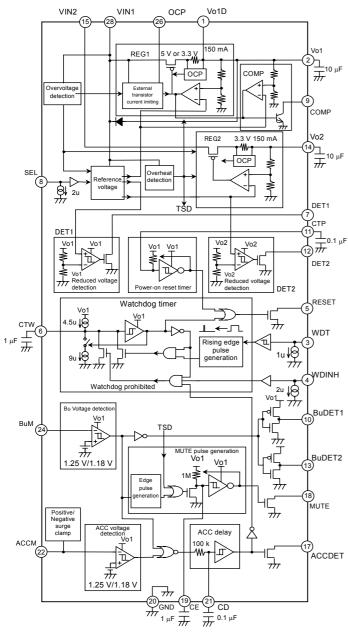
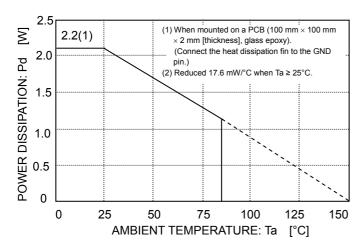
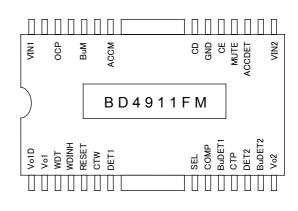


Fig.49 Block Diagram

No	э.	Pin name	Function
1		Vo1D	Main regulator built-in boost transistor output pin
2	2 Vo1		Main regulator output voltage monitor/Vo1 circuit
			power supply pin
3	3	WDT	Watchdog pulse signal input pin
			Watchdog prohibit signal input pin (prohibits WDT
4	ł	WDINH	when ACC is low or WDINH is high).
			Reset signal output pin
5	5	RESET	Reset signal output pin
6		CTW	Runaway operation detection time setting
0	,	CTW	capacitor connection pin
7	,	DET1	Vo1 reduced-voltage detection signal output pin
. 8		SEL	Main regulator output voltage and reduced-voltage
. 0)	SEL	detection voltage selection input pin
9)	COMP	Elevated output detection output
10	<u> </u>		+B drop detection signal output pin (Vo1 at high
1(0	BuDET1	output)
		OTD	Power-on reset time setting capacitor connection
1'	I	CTP	pin
12	2	DET2	Vo2 reduced-voltage detection signal output pin
	<u>_</u>		+B drop detection signal output pin (Vo2 at high
1:	3	BuDET2	output)
14	4	Vo2	Sub-regulator output pin
1	5	VIN2	Sub-regulator input pin
16	6	N.C.	NC pin
17	7	ACCDET	ACC voltage detection signal output pin
18	8	MUTE	MUTE output pin
19	9	CE	MUTE pulse width setting capacitor connection pin
20	0	GND	GND pin
	4	20	ACCDET delay time setting capacitor connection
2'	1	CD	pin
22	2	ACCM	Acc voltage monitor input pin
23	3	N.C.	NC pin
24	4	BuM	Battery voltage monitor input pin
2	5	N.C.	NC pin
~			Main regulator external boost transistor current
26	o	OCP	limit setting resistance connection pin
2	7	N.C.	NC pin
28	8	VIN1	Main regulator connection pin
FI	N	FIN	Heat dissipation fin (sub) (Connect to GND pin.)

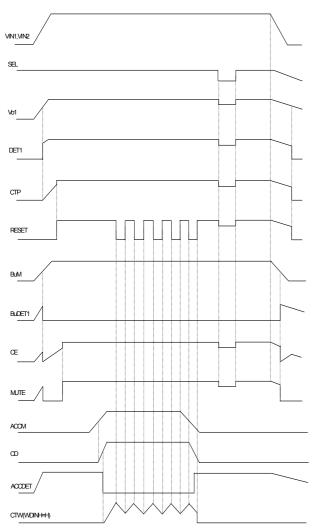






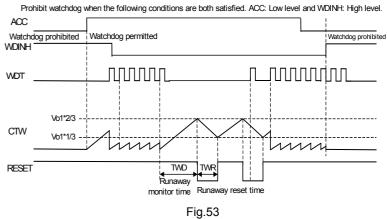
UNIT: (mm)

Fig.51 Pin Assign Diagram



OI/O Timing Chart

OWatchdog timer and runaway operation reset timing chart 1



OWatchdog timer and runaway operation reset timing chart 2

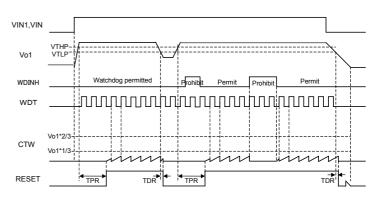
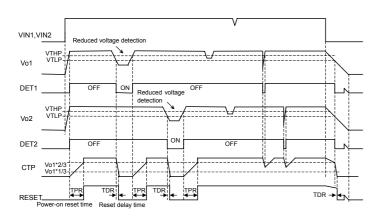


Fig.54

Fig.52



OPower-on reset and reduced-voltage reset timing chart

Fig.55

OBuDET1, BuDET2, ACCDET, and MUTE timing chart

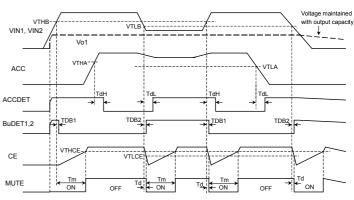
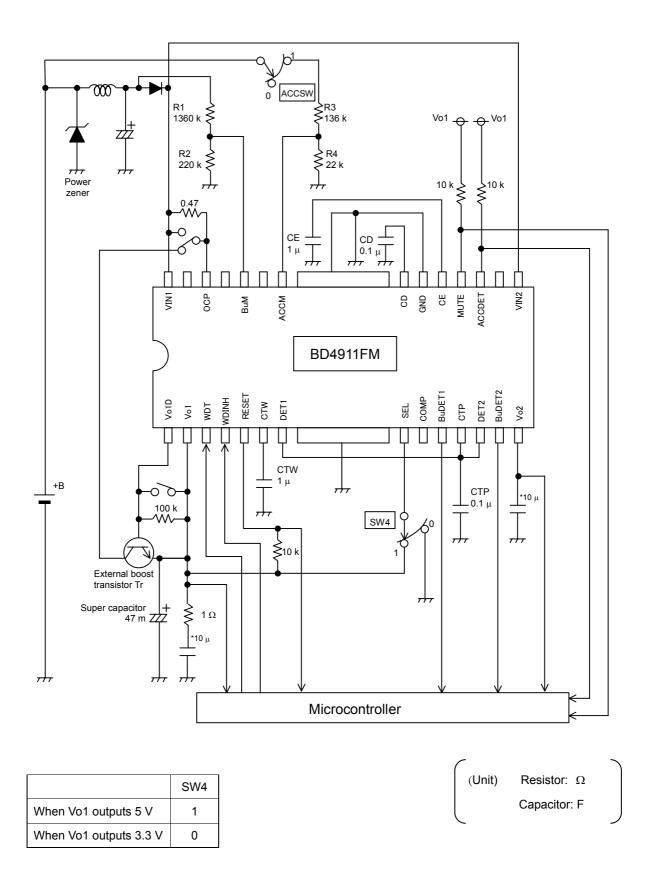


Fig.56



*Select the output capacitance carefully after referring to all operating precautions.



Operation Notes

1. Absolute maximum ratings

An excess in the absolute maximum ratings, such as supply voltage, temperature range of operating conditions, etc., can break down the devices, thus making impossible to identify breaking mode, such as a short circuit or an open circuit. If any over rated values will expect to exceed the absolute maximum ratings, consider adding circuit protection devices, such as fuses.

2. GND voltage

The potential of GND pin must be minimum potential in all operating conditions.

3. Thermal design

Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) in actual operating conditions.

4. Inter-pin shorts and mounting errors

Use caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error or if pins are shorted together.

- 5. Actions in strong electromagnetic field Use caution when using the IC in the presence of a strong electromagnetic field as doing so may cause the IC to malfunction.
- 6. Testing on application boards

When testing the IC on an application board, connecting a capacitor to a pin with low impedance subjects the IC to stress. Always discharge capacitors after each process or step. Always turn the IC's power supply off before connecting it to or removing it from a jig or fixture during the inspection process. Ground the IC during assembly steps as an antistatic measure. Use similar precaution when transporting or storing the IC.

7. Regarding input pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of these P layers with the N layers of other elements, creating a parasitic diode or transistor. For example, the relation between each potential is as follows:

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes can occur inevitable in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Accordingly, methods by which parasitic diodes operate, such as applying a voltage that is lower than the GND (P substrate) voltage to an input pin, should not be used.

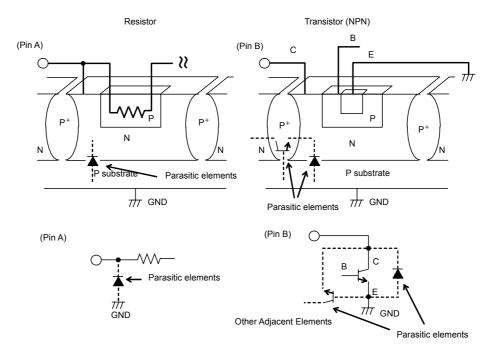


Fig.58 Example of a Simple Monolithic IC Architecture

8. Ground Wiring Pattern

When using both small signal and large current GND patterns, it is recommended to isolate the two ground patterns, placing a single ground point at the ground potential of application so that the pattern wiring resistance and voltage variations caused by large currents do not cause variations in the small signal ground voltage. Be careful not to change the GND wiring pattern of any external components, either.

9. Recommended operating ranges

Proper circuit functionality is guaranteed within the operating temperature range for power supply voltages that fall within the recommended ranges.

Although standard electrical characteristics values are not guaranteed, characteristics values will not vary suddenly within these ranges.

10. Output capacitors

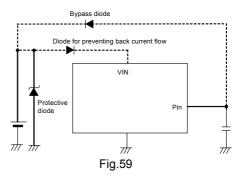
Capacitors for stopping oscillation must be placed between each output pin and the GND pin.

It is recommended to use a 10 μ F ceramic capacitor (B characteristics). When using an external boost transistor, the ceramic capacitor described above should be connected in series with a 1 Ω resistor. Abrupt input voltage and load fluctuations can affect output voltages. Output capacitor capacitance values should be determined after sufficient testing of the actual application.

11. Applications or inspection processes with modes where the potentials of the VIN pin and other pins may be reversed from their normal states may cause damage to the IC's internal circuitry or elements. For example, such damage might occur when VIN is shorted with the GND pin while an external capacitor is charged. Use capacitors that fall within the range listed for each pin in Table 1. It is recommended to insert a diode to prevent back current flow in series with VIN, or bypass diodes between VIN and each pin. If the VIN pin carries a lower voltage than the GND pin, insert a protective diode between the VIN and GND pins.

Output pin	Output capacitor				
Vo1	10 μF to 2200 μF				
Vo2	10 μF to 2200 μF				
Table 1					





12. Overcurrent protection circuits

The IC incorporates a built-in overcurrent protection circuit for each output pin. Each circuit is specifically designed for the current capacity of the corresponding pin and acts to prevent damage to the IC when an overcurrent flows. The protection circuits use drooping type current limiting (when using the built-in transistor) or dropping fold-back type current limiting (when using the external boost transistor). They are designed to limit current flow by not latching up in the event of a large and instantaneous current flow originating from a large capacitor or other component. Their design allows for sufficient safety margins. These protection circuits are effective in preventing damage due to sudden and unexpected accidents. However, the IC should not be used in applications characterized by the continuous operation or transitioning of the protection circuits (for example, applications where the IC is continuously connected to a load that significantly exceeds the output current capacity).

Use caution regarding thermal design, as the output current capacity varies negatively with the temperature characteristics.

13. Overvoltage protection circuit

Overvoltage protection is designed to turn off all output voltages when the voltage differential between the VIN and GND pins exceeds approximately 31 V (at room temperature). Use caution when determining the power supply voltage range to use.

14. Thermal shutdown circuit (TSD)

This IC incorporates a built-in thermal shutdown circuit for the protection from thermal destruction. The IC should be used within the specified power dissipation range. However, in the event that the IC continues to be operated in excess of its power dissipation limits, the attendant rise in the chip's temperature Tj will trigger the thermal shutdown circuit to turn off all output power elements. The circuit will automatically reset once the chip's temperature Tj drops. Operation of the thermal shutdown circuit presumes that the IC's absolute maximum ratings have been exceeded. Application designs should never make use of the thermal shutdown circuit.

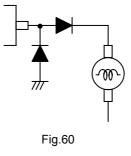
15. Ground precautions

Pattern routes connecting the ground points, indicated in application circuits example, to the GND pin should be sufficiently short and should be positioned to avoid electrical interference.

16. Bypass capacitor between the VIN and GND pins

It is recommended to insert a 0.47 μF to 10 mF bypass capacitor between the VIN and GND pins. Capacitance values vary with application. Capacitors should be tested in actual implementations, and designs should allow for sufficient margins. Failure to use the optimum capacitance value may lead to output oscillation and other issues.

17. Applications with modes where the potentials of the input pins (VIN1, VIN2) and GND pins and other output pins may be reversed from their normal states may cause damage to the IC's internal circuitry. In particular, it is recommended to create a bypass route with diodes or other components when loads including large inductance components are connected where BEMF may be generated during startup or when output is turned off.



18. Always verify the characteristics of example application circuits prior to their use. When changing other external circuit b constants, allow for sufficient margins after considering the variability of both the ROHM IC and external components, including both static and transient characteristics.

Selecting a model name when ordering



Part number BD4911

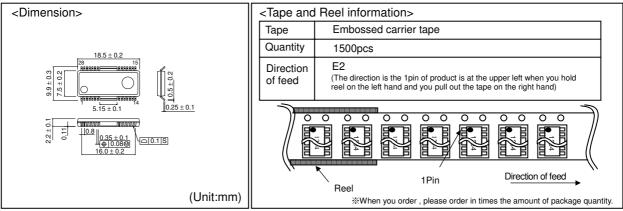


Package type FM: HSOPM28



Taping type E2 = Reel-wound embossed taping (HSOPM28)

HSOP-M28



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