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Configuration	Silicon Monolithic Integrated Circuit
Product Name	TFT Liquid Crystal Power Management IC for Car Navigation System
Model Name	BD6171KV
Outline Drawing	Figure - 1 (Molded Plastic)
Block Diagram	Figure - 2
Functions	<ol style="list-style-type: none"> 1. 2 ch PWM controller with variable stepdown converter. 2. Variable output voltage series regulator. 3. Variable output voltage positive / negative charge pump inverter. 4. Oscillator is enabled by an external pulse synchronization. 5. Overcurrent protection (OCP) circuitry. 6. Undervoltage Lockout protection (UVLO). 7. Thermal Shutdown Protection (TSD). 8. Each channel can be individually turned ON/OFF. 9. VQUPF48 package

○ Absolute Maximum Ratings (Ta=25°C)

Specifications	Symbol	Rating	Unit
Voltage between VCC, DVCC-GND	Vcc, DVcc	36	V
Voltage between PVCC1,2-GND	PVcc1,2	36	V
Voltage between VCC-PVCC1,2	Vcc-PVcc1,2	0.3	V
Voltage between VREF-GND	VREF	7	V
Voltage between VREGA-GND	VREGA	7	V
Voltage between VCC-VRGB	VRGB	7	V
Voltage between VCC-OUT1,2	VOUT1,2	7	V
Voltage between PVCC3-GND	PVcc3	36	V
Voltage between PVCC4,5-GND	PVcc4,5	12	V
Voltage between OUT4,5-GND	VOUT4,5	PVcc4,5	V
Voltage between CL1,2-GND	VCL1,2	PVcc1,2	V
Voltage between VS1,2-GND	VVS1,2	PVcc1,2	V
Voltage between VODET1,2,3,4,5-GND	VODET1,2,3,4,5	VREGA	V
Voltage between SYNC-GND	VSYNC	7	V
Power Dissipation	Pd	900(*1)	mW
Operating Temperature Range	Topr	-40~+85	°C
Ambient Storage Temperature Range	Tstg	-55~+125	°C

(*1) When mounted on a 70 × 70 × 1.6 mm³ glass epoxy substrate, for every temperature increase of 1°C, power dissipation decreases by 9 mW (at temperatures of 25°C or greater).

Application example

The application circuit is recommended for use. Make sure to confirm the adequacy of the characteristics.

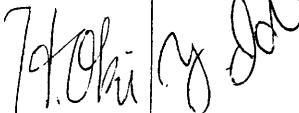
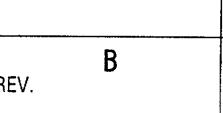
When using the circuit with changes to the external circuit constants, make sure to leave an adequate margin for external components including static and transitional characteristics as well as dispersion of the IC.

Note that ROHM cannot provide adequate confirmation of patents.

The product described in this specification is designed to be used with ordinary electronic equipment or devices (such as audio-visual equipment, office-automation equipment, communications devices, electrical appliances, and electronic toys).

Should you intend to use this product with equipment or devices which require an extremely high level of reliability and the malfunction of which would directly endanger human life (such as medical instruments, transportation equipment, aerospace machinery, nuclear-reactor controllers, fuel controllers and other safety devices), please be sure to consult with our sales representative in advance.

ROHM assumes no responsibility for use of any circuits described herein, conveys no license under any patent or other right, and makes no representations that the circuits are free from patent infringement.

DESIGN 	CHECK 	APPROVAL 	DATE : 2/13/04	SPECIFICATION No. : TSZ02201-BD6171KV-1-2	
				REV. B	ROHM CO., LTD.

○ Recommended Operating Conditions

Specifications	Symbol	Standard Value			Unit	Conditions
		Min.	Standard	Max.		
Supply Voltage	Vcc,DVcc	6.0	—	24.0	V	
PWM Controller Input Voltage	PVcc1,2	6.0	—	24.0	V	
Series REG Input Voltage	PVcc3	5.0	—	24.0	V	
Charge Pump Input Voltage	PVcc4,5	3.0	—	12.0	V	
Oscillator Frequency	Fosc	50	—	500	kHz	
External Synchronization Frequency	FSYNC	60	—	600	kHz	
Oscillator Timing Resistor Value	RT	30	—	680	kΩ	
Overcurrent Protection Cancellation Timing Capacitor Value	Cocp1,2	—	—	1000	pF	

○ Electrical Properties (Ta=25°C, Vcc,DVcc,PVcc1,2,3=13.2V, PVcc4,5=8V, Fosc=500kHz, EN1~5=5V unless otherwise specified.)

Specification	Symbol	Standard Value			Unit	Conditions	Measurement Circuit
		Min.	Standard	Max.			
【Entire Device】							
Standby Current	Iccst	—	—	10	μA	EN1~5=0V	1
Operating Current	Icc	—	5	10	mA	FB1,FB2=0V	1
【Reference Voltages】							
Output Voltage	VREF	2.97	3.00	3.03	V	IREF=-0.1mA	1
Line Regulation	Δ VLREF	—	—	10	mV	Vcc=7~18V, IREF=-0.1mA	1
Load Regulation	Δ	—	—	10	mV	IREF=-0.1mA~1mA	1
【Internal Regulator】							
VREGA Output Voltage	VREGA	4.5	5.0	5.5	V		1
VRGB Output Voltage	VRGB	VCC -5.5	VCC -5.0	VCC -4.5	V	When switching between PVcc1,2-OUT1,2 with a 1000 pF connection	1

VREGB Voltage	Dropout Voltage	Δ VREGB	—	1.8	2.2	V	Voltage between VREGB-GND When switching between Vcc=5.6V,PVcc1,2-OUT1, 2with a 1000 pF connection	1
【Enable Unit】 EN1~5								
Threshold Voltage	VEN	1.2	1.8	2.4	V			1
Inflow Current	IEN	—	17	35	μA	EN1~5=5V		1

◎ The design of this device does not incorporate measures used to harden against radiation.

○ Electrical Properties (Ta=25°C, Vcc,DVcc,PVcc1,2,3=13.2V, PVcc4,5=8V, Fosc=500kHz, EN1~5=5V unless otherwise specified.)

Specification	Symbol	Standard Value			Unit	Conditions	Measure- ment Circuit
		Min.	Standard	Max.			

Series REG Unit

【Output】 (Measured at a 5V output setting, unless otherwise specified.)							
Feedback Voltage	VFB3	0.97	1.00	1.03	V	Io3=-80mA	2
Load Regulation	ΔVLO3	—	80	160	mV	Io3=0~ -80mA	2
Line Regulation	ΔVLI3	—	20	40	mV	Vcc=10~16V, Io3=-80mA	2
Output Current Ability	Io3	100	—	—	mA	VO3=4.75V	2
Dropout Voltage	ΔVo3	—	0.4	0.7	V	PVCC3=4.75V, Io3=-80mA	2
Ripple Rejection	RR3	45	60	—	dB	f=100Hz, Io3=-80mA	2
Output Voltage	Vo3	—	5.0	—	V	Io3=-80mA	2
Output Short Current	Io3S	—	25	—	mA	Vo3=0V	2
FB3 Outflow Current	IFB3	—	—	1	μA	VFB3=0.97V	2
【Output Voltage Detection】							
Output Detection Voltage	VDET3	0.55	0.6	0.65	V	FB3 Voltage	2
Hysteresis	VHYS3	0.1	0.2	0.3	V		2
Output High Voltage	VDETH3	VREGA-0.4	—	—	V	IDET3=-1 μA	2
Output Low Voltage	VDETL3	—	—	0.4	V	IDET3=25 μA	2

PWM Controller Unit

【Oscillator】							
Oscillator	Fosc	270	300	330	kHz	RT=91kΩ	3
Oscillator Frequency Ratio	ΔFosc	—	—	2	%	Vcc=7~18V	3
【Frequency Synchronization】							
Synchronization Frequency	FSYNC	—	375	—	kHz	FSYNC=375kHz	3
SYNC Threshold Voltage	VSYNC	1.2	1.4	1.6	V		3
SYNC Outflow Current	ISYNC	-1	—	—	μA	VSYNC=1.4V	3

【Error Amp】

Feedback Voltage	VINV1,2	0.98	1.00	1.02	V		3
Input Bias Current	IINV1,2	-1	—	—	μA	VINV1,2=0.97V	3
Maximum Output Voltage	VFBH1,2	2.8	—	—	V	VINV1,2=0.5V	3
Minimum Output Voltage	VFBL1,2	—	—	0.1	V	VINV1,2=1.5V	3
Output Sink Current	IFBSINK 1,2	-6.0	-2.0	-0.5	mA	VFB1,2=1.5V	3
Output Source Current	IFBSOUR CE1,2	60	100	160	μA	VFB1,2=1.5V	3
Soft Start Charging Current	ISS1,2	-4.0	-2.5	-1.0	μA	VSS1,2=1.0V	3
Soft Start Threshold Voltage	VSS1,2	—	1.0	—	V	VINV1,2=1.0V, SS Voltage	3
Soft Start Standby Voltage	VSSSTD 1,2	—	10	100	mV	SS Voltage	3

© The design of this device does not incorporate measures used to harden against radiation.

- Electrical Properties (Ta=25°C, Vcc,DVcc,PVcc1,2,3=13.2V, PVcc4,5=8V, Fosc=500kHz, EN1~5=5V unless otherwise specified.)

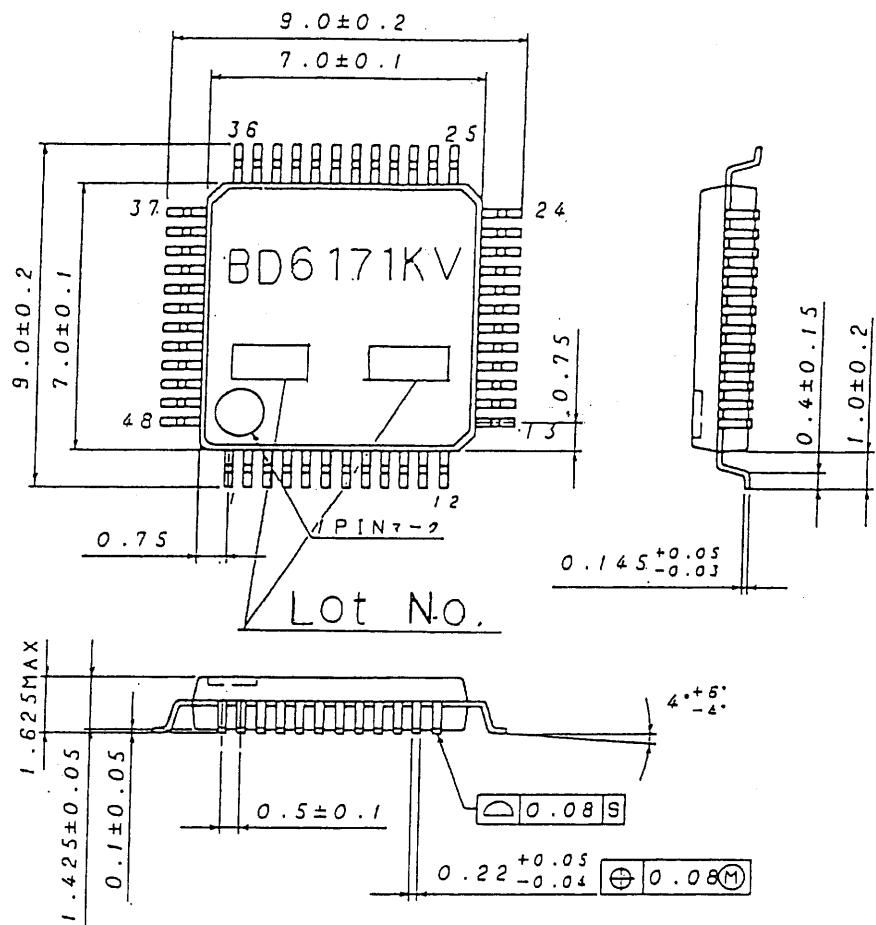
Specification	Symbol	Standard Value			Unit	Conditions	Measurement Circuit
		Min.	Standard	Max.			
【FET Driver】							
Output Source Current	IoSOURCE1,2	—	-100	—	mA	PVcc1,2-OUT1,2=0.4V	3
Output Sink Current	IoSINK1,2	—	100	—	mA	OUT1,2-VREGB=1.3V	3
PchON Resistance	RONP1,2	—	3.0	6.0	Ω	IOUT1,2=-100mA	3
NchON Resistance	RONN1,2	—	13.0	20.0	Ω	IOUT1,2=100mA	3
Rise Time	Tr1,2	—	20	—	nsec	When switching between PVcc1,2-OUT1,2 with a 1000 pF connection ※Design Assurance	3
Fall Time	Tf1,2	—	20	—	nsec	When switching between PVcc1,2-OUT1,2 with a 1000 pF connection ※Design Assurance	3
【Overcurrent Protection】							
VS Threshold Voltage	VVS1,2	VCC -0.22	VCC -0.20	VCC -0.18	V	RCL1,2=20KΩ	4
VS Inflow Current	IVS1,2	—	—	1	μA	VS1,2=PVCC1,2	4
CL Inflow Current	ICL1,2	9	10	11	μA	VCL1,2=13.2V	4
【Overcurrent Protection Cancellation】							
Charging Current	IOCP1,2	-4.0	-2.5	-1.0	μA	VOCP1,2=1.0V	4
Threshold Voltage	VOCP1,2	1.9	2.0	2.1	V		4
Standby Voltage	VOCPST B1,2	—	0.6	1.0	V		4
【Input Undervoltage Malfunction Prevention】							
Threshold Voltage	VUVLO	5.6	5.7	5.8	V	Vcc sweep down	4
Hysteresis Voltage	VHYSUVL O	0.05	0.1	0.15	V		4
【Output Voltage Detection】							
Output Detection Voltage	VDET1,2	0.55	0.6	0.65	V	VINV1,2 Voltage	3
Hysteresis Voltage	VHYS1,2	0.1	0.2	0.3	V		3
Output High Voltage	VDETH1,2	VREGA -0.4	—	—	V	IDET1,2=-1 μA	3
Output Low Voltage	VDETL1,2	—	—	0.4	V	IDET1,2=25 μA	3

○ The design of this device does not incorporate measures used to harden against radiation.

○ Electrical Properties (Ta=25°C, Vcc,DVcc,PVcc1,2,3=13.2V, PVcc4,5=8V, Fosc=500kHz, EN1 ~ 5=5V unless otherwise specified.)

Specification	Symbol	Standard Value			Unit	Conditions	Measurement Circuit				
		Min.	Standard	Max.							
Charge Pump Unit											
【Output】											
Output Current	Source IoSOURCE4,5	—	—	-10	mA	OUT4,5=PVCC4,5-150mV	5				
Output Sink Current	IoSINK4,5	10	—	—	mA	OUT4,5=150mV	5				
【Error Amp】											
Feedback Voltage (Positive Voltage)	VFB4	0.97	1.00	1.03	V		5				
Input Bias Current (Positive Voltage)	IFB4	-1	—	—	μA	FB4=0.97V	5				
Feedback Voltage (Negative Voltage)	VFB5	-0.03	0.0	0.03	V		5				
Input Bias Current (Negative Voltage)	IFB5	-1	—	—	μA	FB5=-0.03V	5				
【Output Voltage Detection】											
Output Detection Voltage (Positive Voltage)	VDET4	0.55	0.6	0.65	V	FB4	5				
Hysteresis Voltage (Positive Voltage)	VHYS4	0.1	0.2	0.3	V		5				
Output Detection Voltage (Negative Voltage)	VDET5	0.55	0.6	0.65	V	FB5	5				
Hysteresis Voltage (Negative Voltage)	VHYS5	0.1	0.2	0.3	V		5				
Output High Voltage	VDETH4,5	VREGA- 0.4	—	—	V	IDET4,5=-1 μA	5				
Output Low Voltage	VDETL4,5	—	—	0.4	V	IDET4,5=25 μA	5				

○ The design of this device does not incorporate measures used to harden against radiation.



(UNIT: mm)

Figure - 1 Outline Drawing

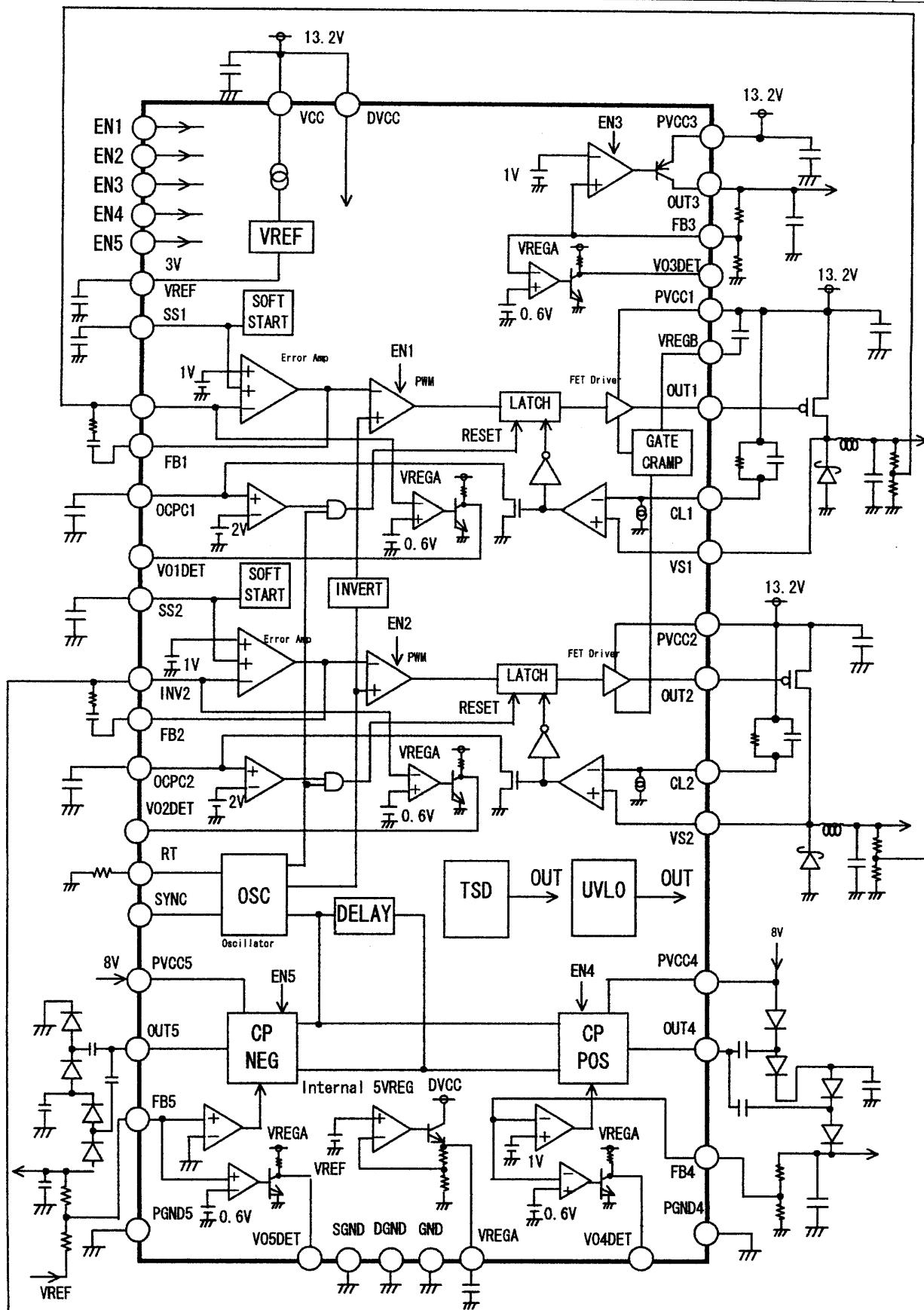
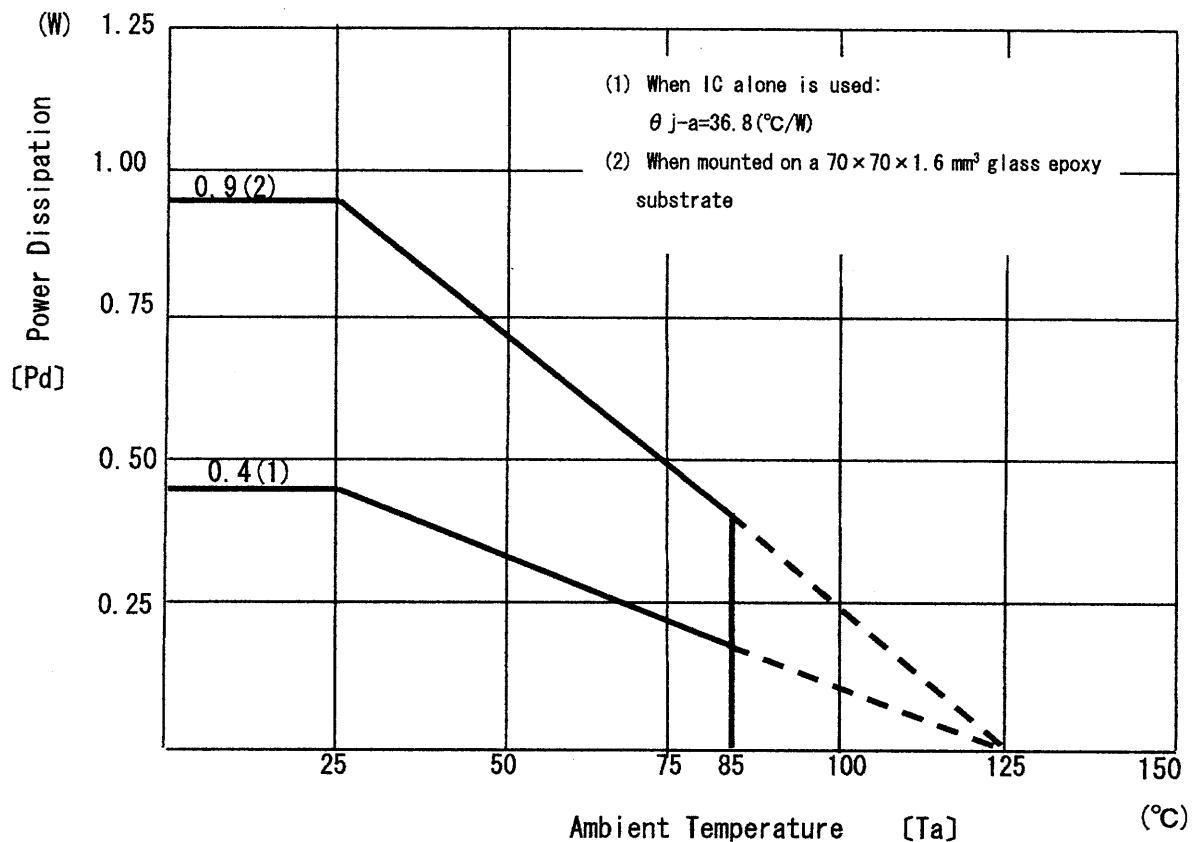


Figure - 2 Block Diagram

○ Heat Dissipation Graph



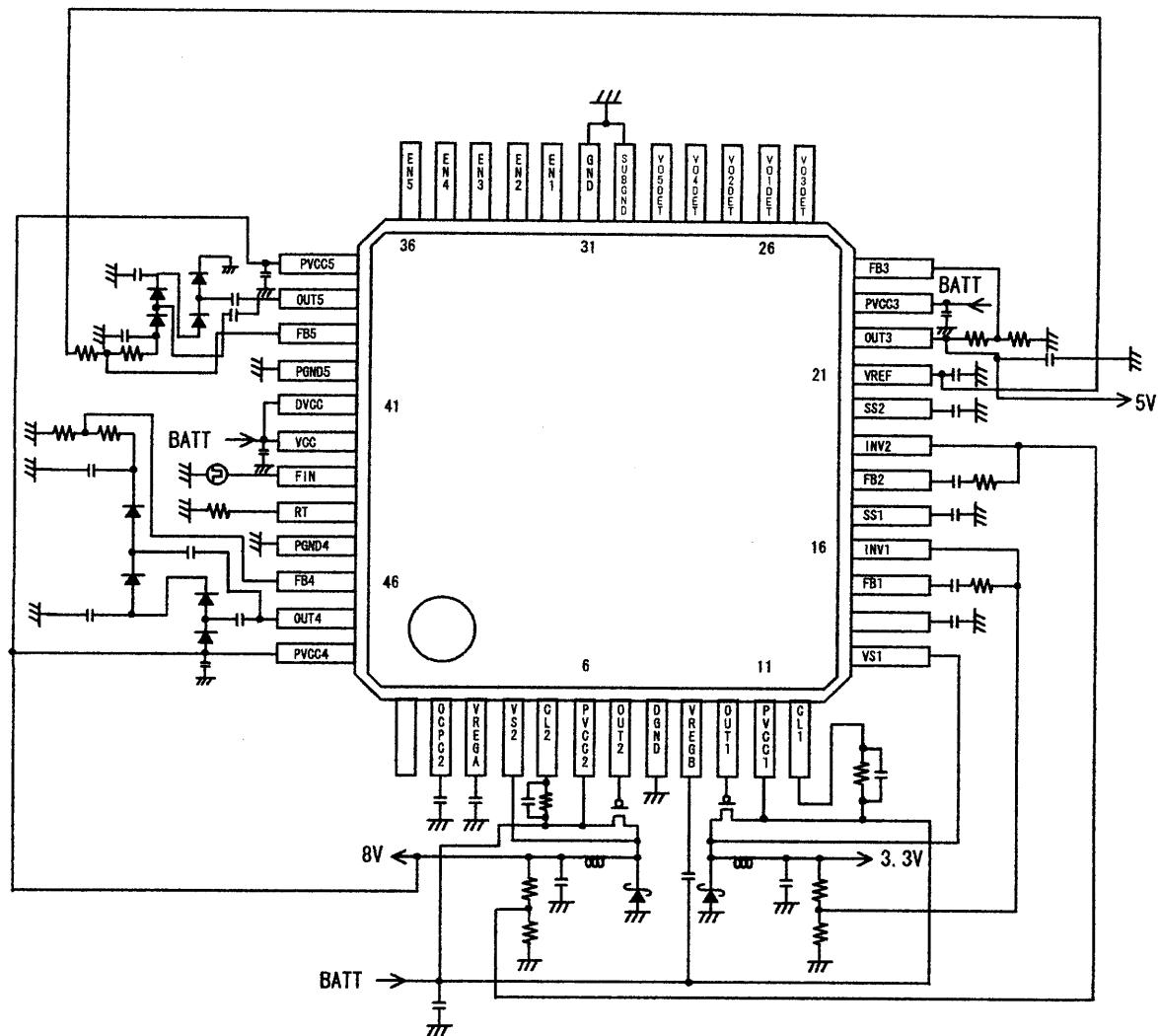


Figure - 3 Pin Plot Diagram (Example of Recommended Circuit)

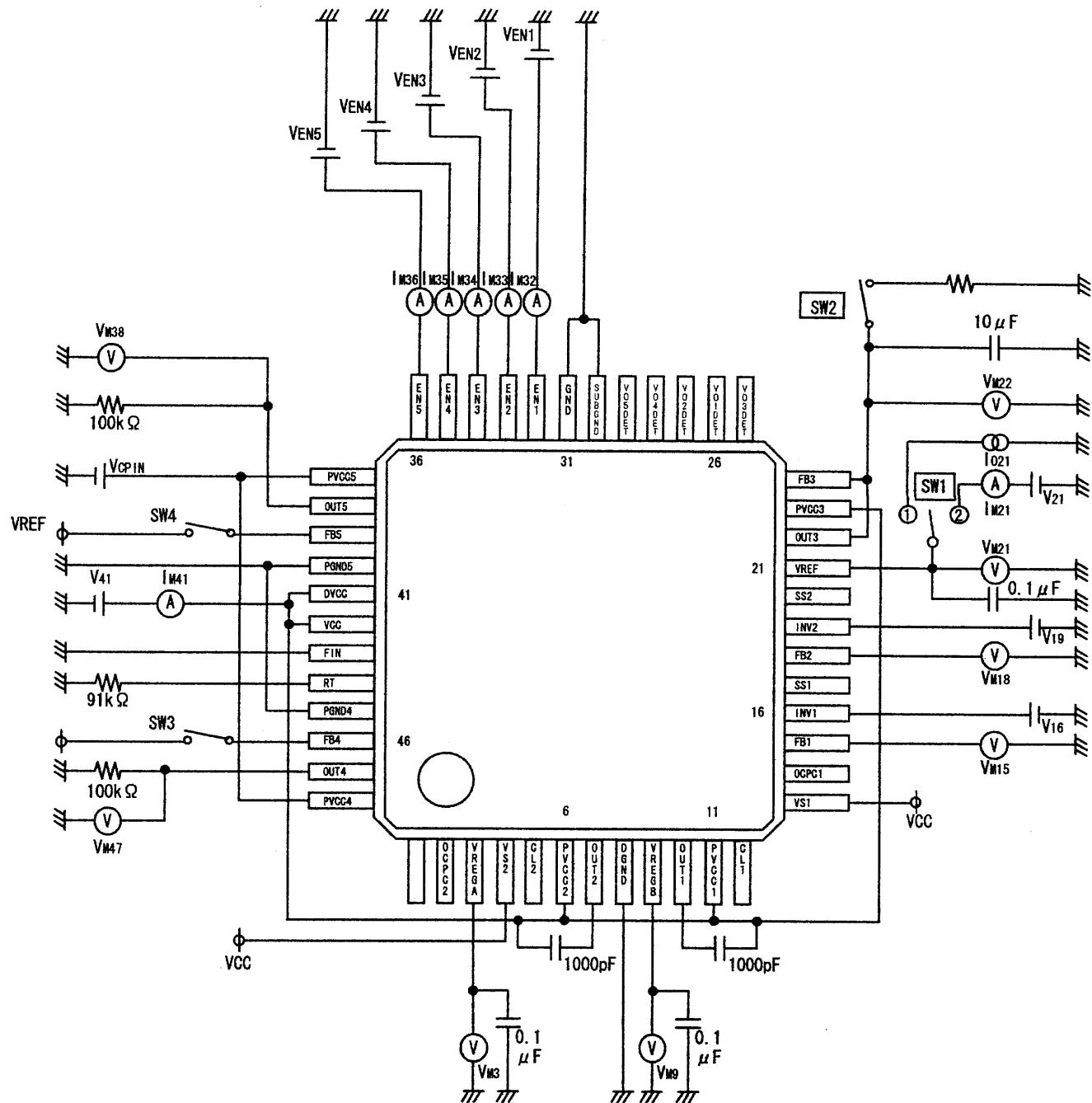


Figure - 4 Measurement Circuit Diagram - 1

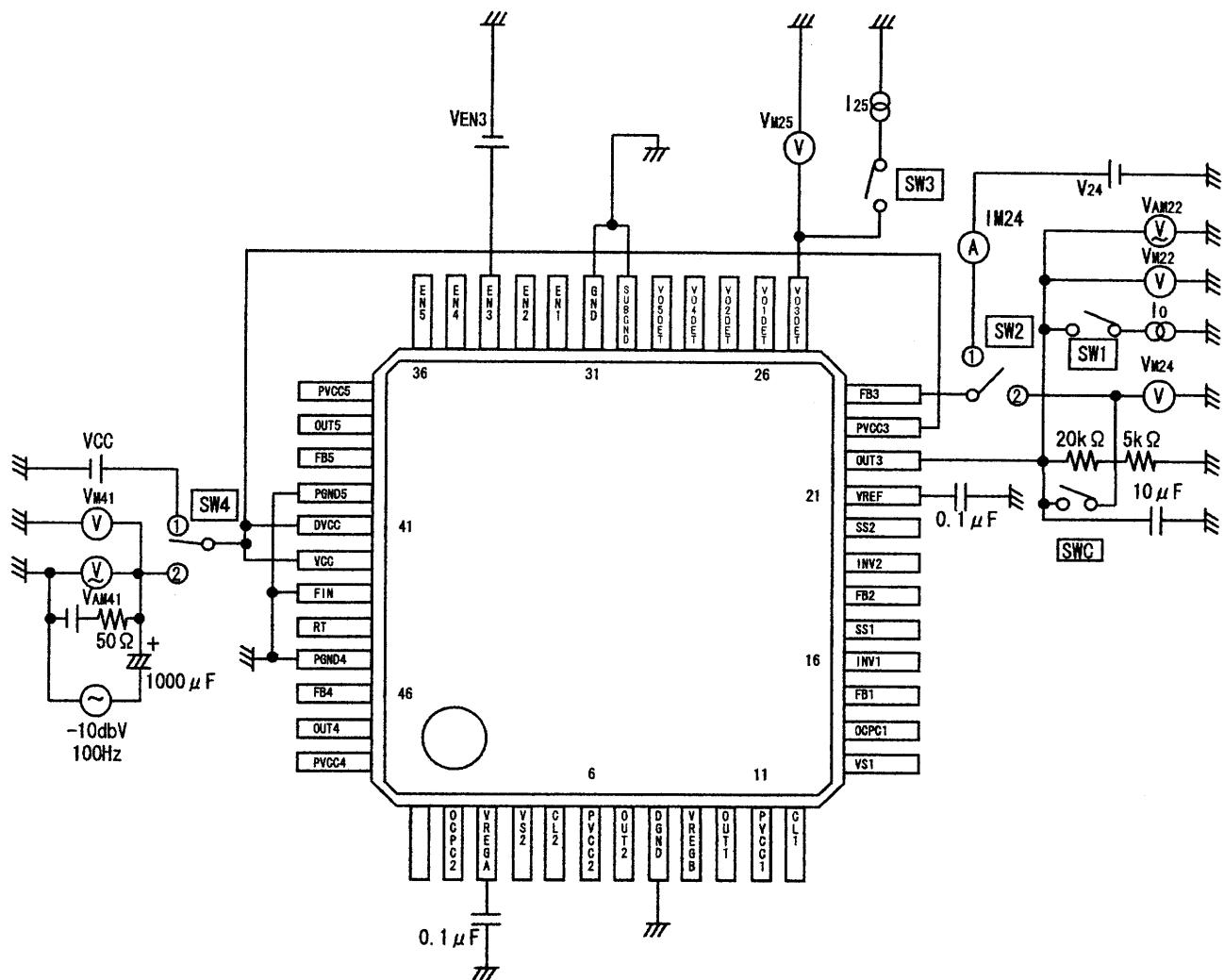


Figure - 5 Measurement Circuit Diagram - 2

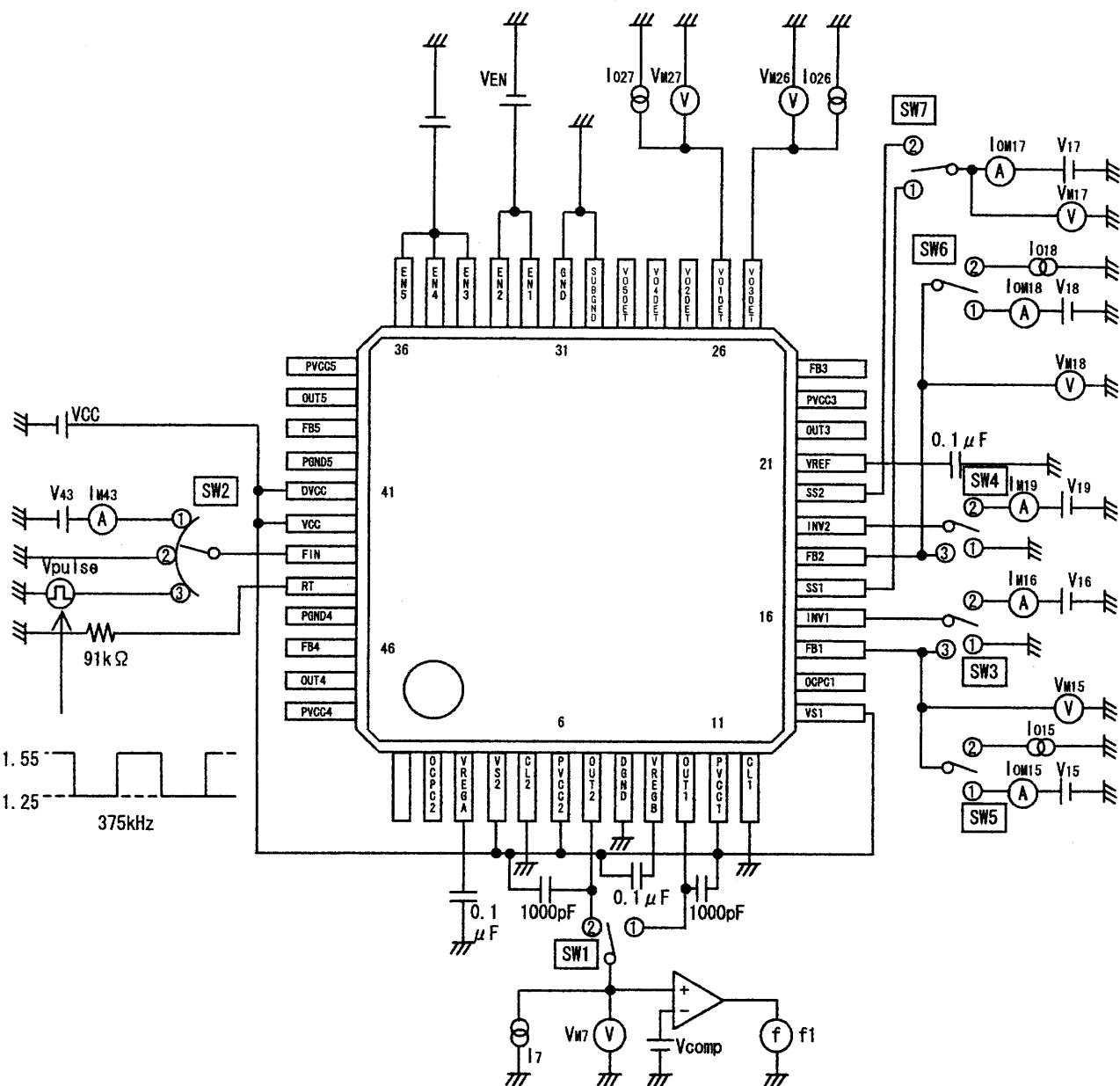


Figure - 6 Measurement Circuit Diagram - 3

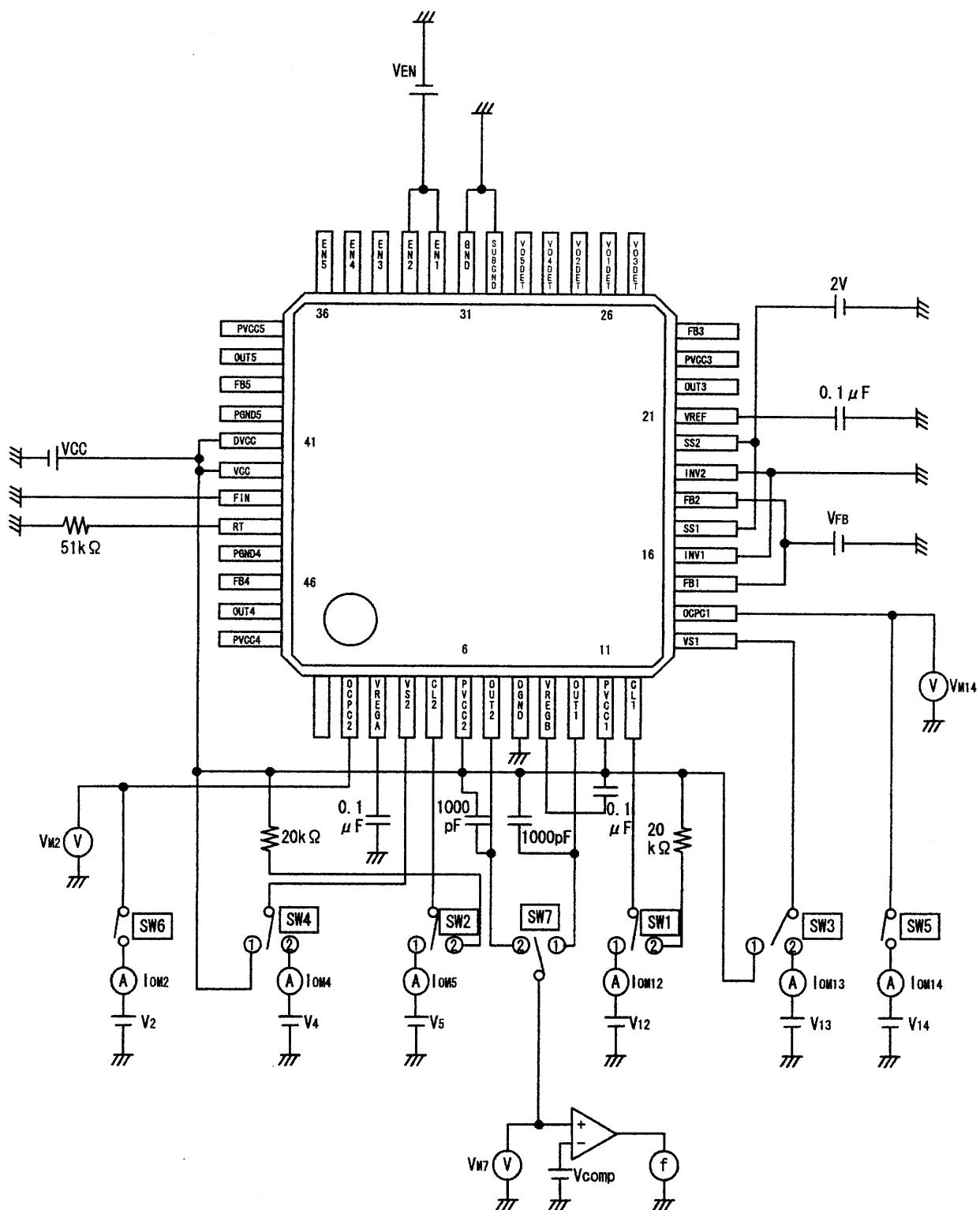


Figure - 7 Measurement Circuit Diagram - 4

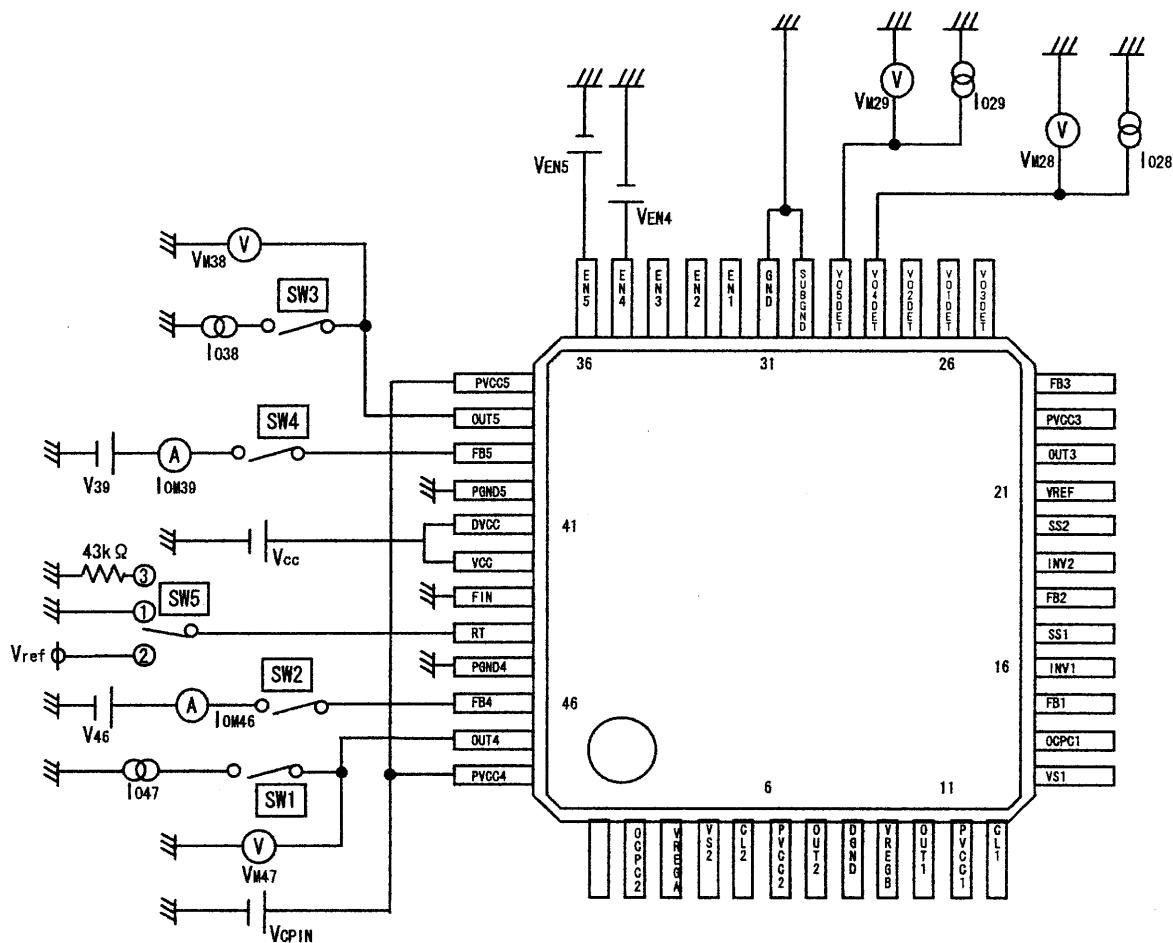
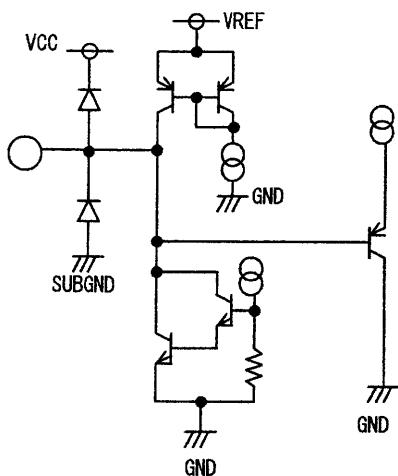


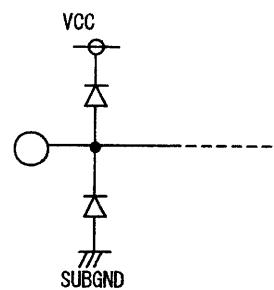
Figure - 8 Measurement Circuit Diagram – 5

○ Terminal Internal Equivalence Circuits

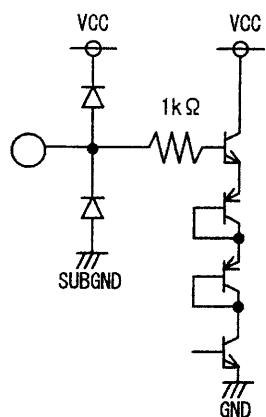
- OUTC2(2pin), OUTC1(14pin)



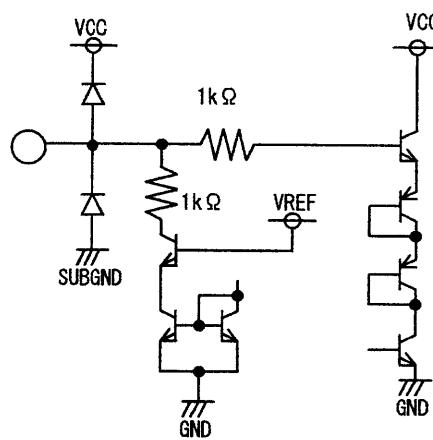
- VREGA(3pin)



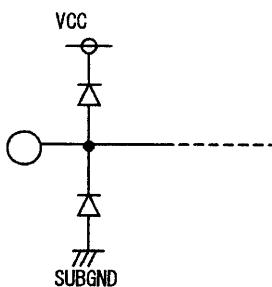
- VS2(4pin), VS1(13pin)



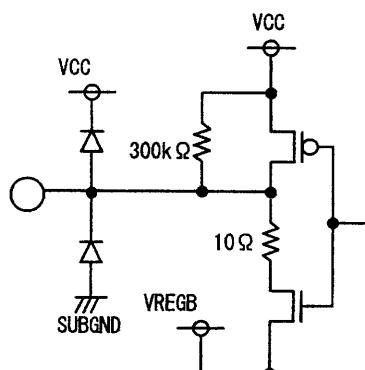
- CL2(5pin), CL1(12pin)



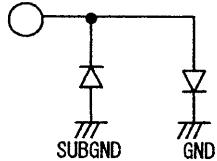
- PVCC2(6pin), PVCC1(11pin)



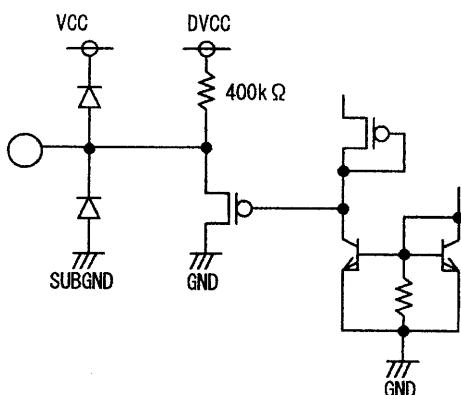
- OUTH2(7pin), OUTH1(10pin)



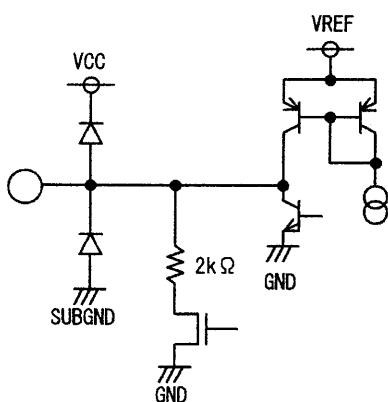
• DGND(8pin)



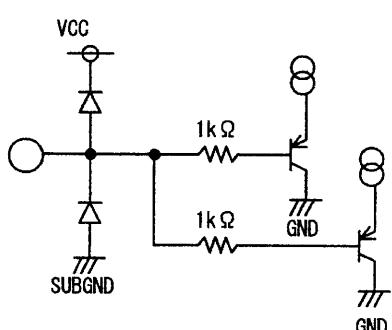
• VREGB(9pin)



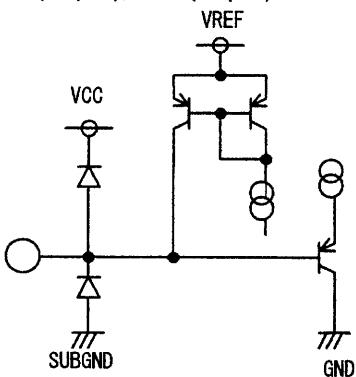
• FBI(15pin), FB2(18pin)



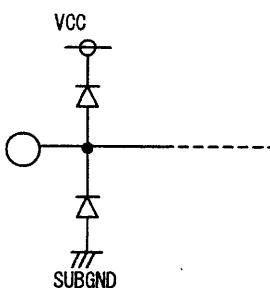
• INV1(16pin), INV2(19pin)



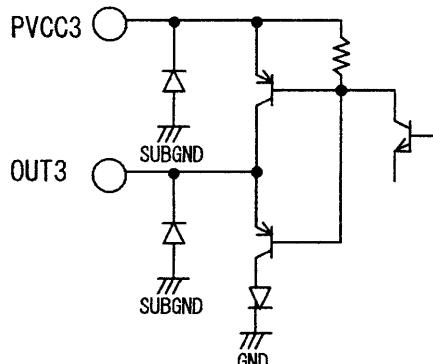
• SS1(17pin), SS2(20pin)



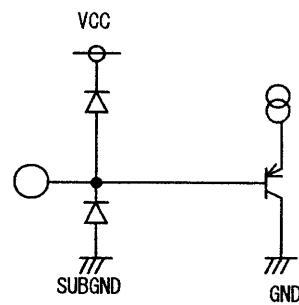
• VREF(21pin)



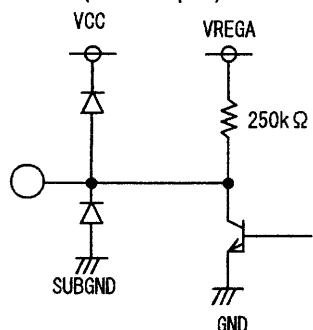
- OUT3(22pin), PVCC3(23pin)



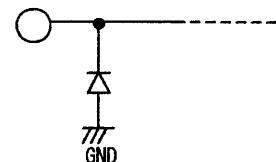
- FB3(24pin)



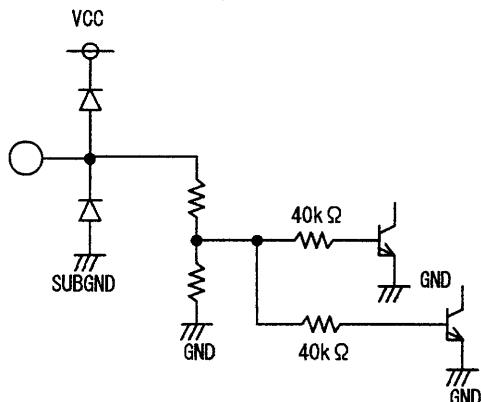
- VoDET(25~29pin)



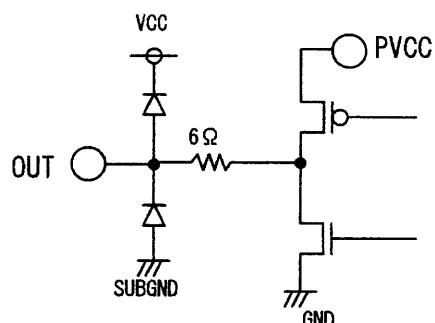
- SUBGND(30pin)



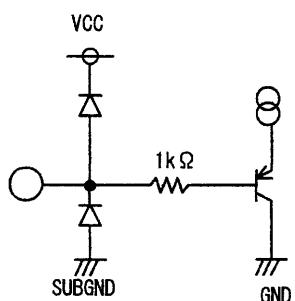
- EN1~5(31~36pin)



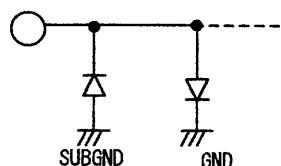
- PVCC5(37pin), PVCC4(48pin), OUT5(38pin), OUT4(47pin)



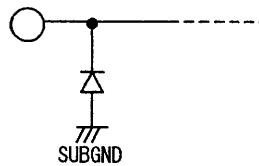
- FB5(39pin), FB4(46pin)



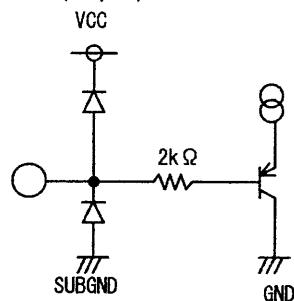
- PGND5(40pin), PGND(45pin)



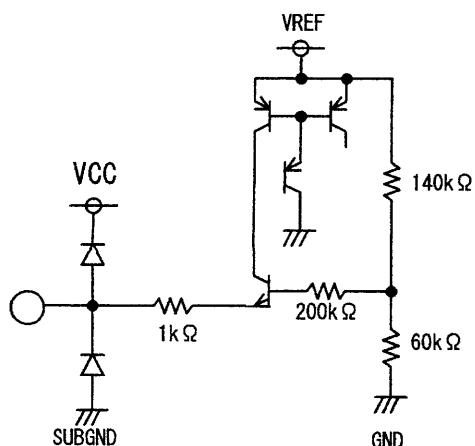
• VCC(42pin)



• FIN(43pin)



• RT(44pin)



○ Explanation of Each Function (Typical values are used for all numeric values below.)

1. PWM Controller Functions

(1) Reference Voltage Block

This circuit generates a constant, temperature-compensated voltage from the power input supplied to the VCC terminal. The voltage is 3V.

(2) Internal Regulator Block (VREGA)

This circuit generates a 5V voltage from the power input supplied to the VCC terminal. A low ESR capacitor must be connected between VREGA—GND.

(3) Internal Regulator Block (VREGB)

This circuit generates a voltage (Vcc-5V) from the voltage supplied to the VCC terminal. This voltage serves as the voltage supply for the main FET driver unit. A low ESR capacitor must be connected between VREGB—VCC.

(4) Oscillator Block

This circuit generates a triangular waveform via the connection of a resistance, which is used to set the frequency, to the RT terminal. It allows for external synchronization of the oscillator via the input of a pulsed waveform to the SYNC terminal, at a frequency higher than the specified oscillator frequency, thus enabling the fine adjustment of the oscillator frequency (within 25%).

(5) Error Amp Block

This circuit detects the output voltage at the INV terminal, amplifies the error in relation to the output voltage setting and then outputs the resulting error voltage from the FB terminal.

The comparison voltage used is 1V.

An arbitrary degree of phase compensation can be specified externally by connecting a resistance and a capacitor between the INV terminal and the FB terminal.

(6) PWM Comparator Unit

This unit converts the voltage output from the error amp into a PWM waveform and then outputs it to the FET driver.

(7) FET Driver Unit

Outputs are driven by Pch FETs, with all outputs being push/pull.

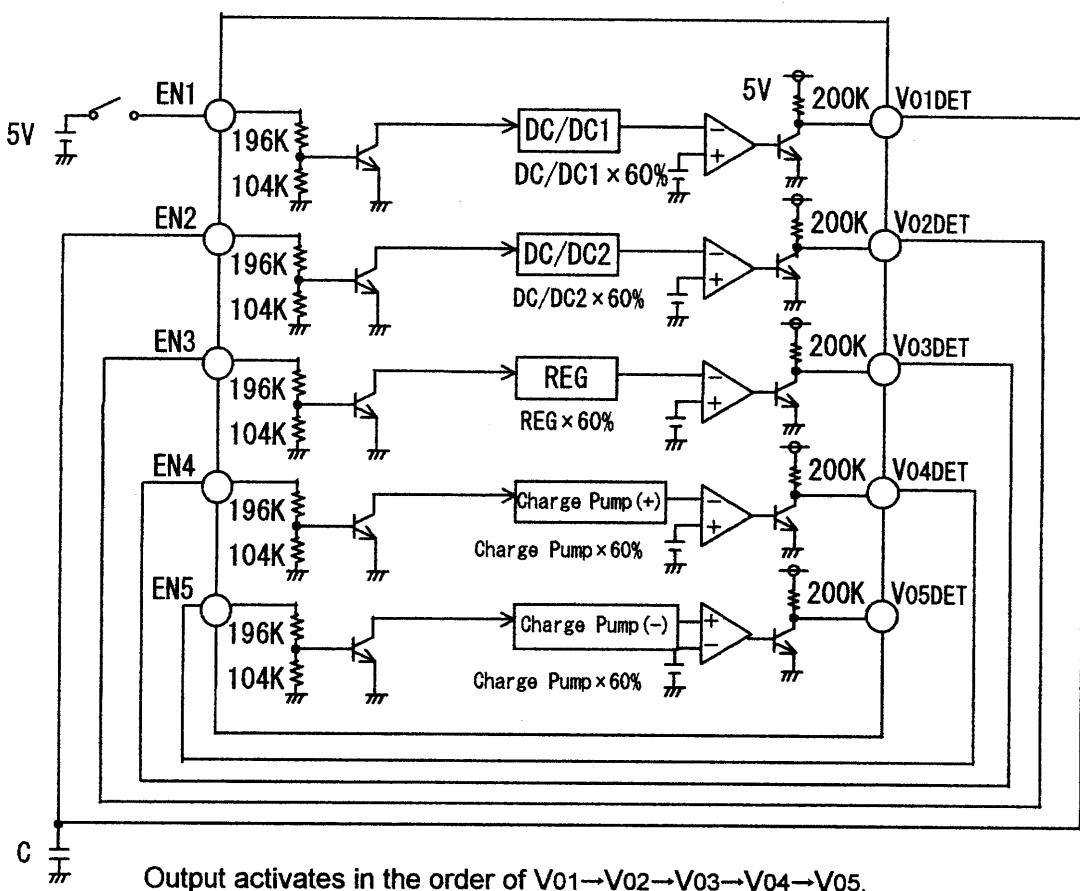
Each output voltage is clamped by the VREGB output voltage.

2. Channel Control Function

Each output can be controlled individually, via terminals EN1 - EN5.

As well, by connecting the DET terminal of each output to terminals EN1 - EN5, the order of activation can be controlled arbitrarily for each output.

(Example)



Output activates in the order of V01 → V02 → V03 → V04 → V05.

Furthermore, when the activation of each output must be delayed, a specific delay time can be set by connecting a capacitor C to an EN terminal, as shown in the diagram above.

3. Protection Functions

(1) Overcurrent Protection Circuit (OCP)

In the DC/DC unit, if the external Pch MOS drain voltage drops to a lower value than the specified external voltage when the output FET is ON, this circuit will cause the output to latch OFF. Subsequently, this circuit begins to charge the capacitor connected to the OCP terminal. Once the capacitor reaches 2.0V, the latch OFF operation will cancel and the circuit will reactivate. At that time, if the unit is still in the overcurrent condition, the output will turn OFF. If the unit has returned to normal load conditions, then normal operation will resume.

(2) Input Undervoltage Malfunction Prevention Circuit (UVLO)

In order to prevent the IC from malfunctioning during power-up or during a power interruption, whenever the supply voltage decreases to approximately 5.7V or less, this circuit turns all of the outputs off. There is approximately 0.1V of hysteresis width between the detection voltage and the UVLO cancellation voltage, thus preventing malfunctions due to input voltage fluctuations at the threshold online.

(3) Overheating Protection Circuit (TSD)

This circuit detects any excess chip heat and turns all of the outputs off, in order to prevent IC damage.

A hysteresis width is present between the overheat detection temperature and the cancellation temperature, thus preventing malfunctions caused by temperature fluctuations at the threshold online.

4. Oscillator External Pulse Synchronization Function

Permits external pulse synchronization of the oscillator through the connection of a resistance to the RT terminal and the input of a synchronizing signal to the SYNC terminal.

The input must be a pulse waveform having a higher frequency than that determined by RT. However, the frequency differential must be specified to within 25%.

In addition, the duty cycle of the pulse signal must be set within the range of 10% - 90%.

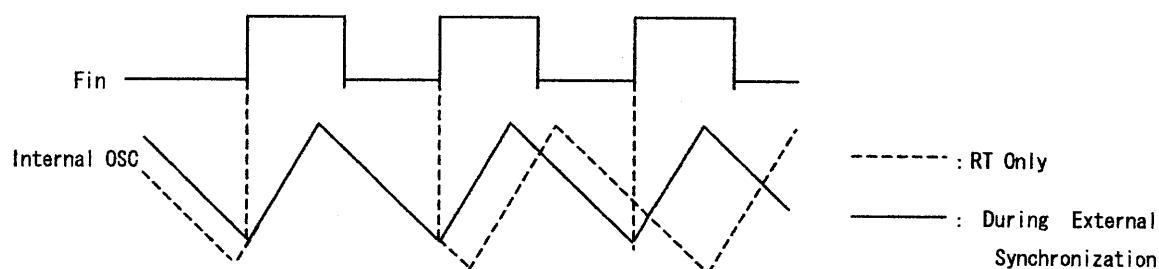


Figure - 9 CT Waveform During External Pulse Synchronization

If the external synchronization function is not utilized, the SYNC terminal must be shorted to GND.

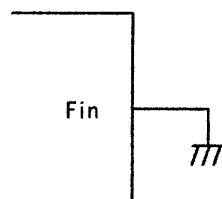
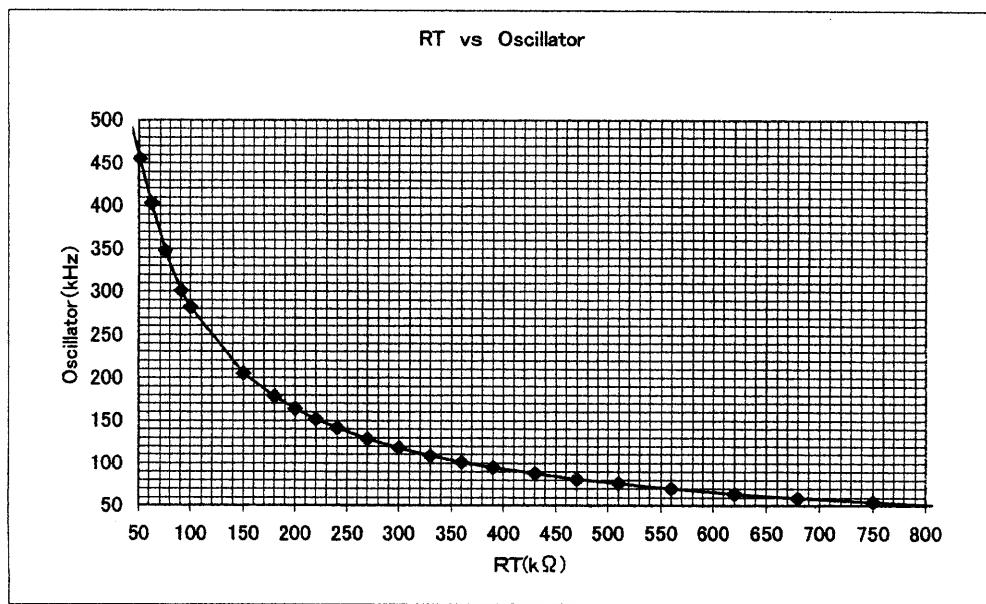


Figure - 10 If External Synchronization Function is Not Utilized



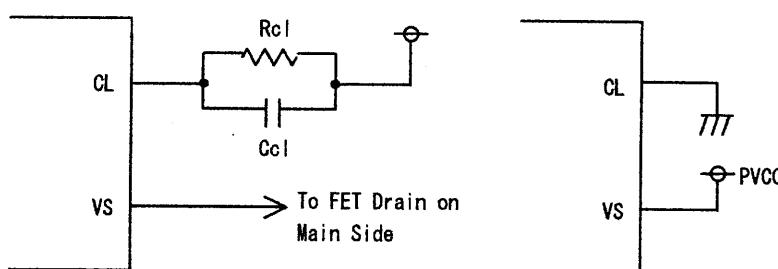
5. Setting the Overcurrent Detection Level

The level of overcurrent detection for the overcurrent protection circuit (I_{ocp}) is determined by the resistance connected to the ON-resistor (R_{ON}) of the external main FET and the resistance connected between CL-VCC (R_{cl}).

$$I_{ocp} = \frac{R_{cl}}{R_{ON}} \times 10^{-5} \text{ [A]} \quad (\text{typ.})$$

In addition, a capacitor (C_{cl}) must be connected in parallel to R_{cl}, in order to prevent incorrect detection due to noise.

If the overcurrent protection circuit is not utilized, the VS terminal must be short-circuited to PVCC and the CL terminal must be short-circuited to GND. (Refer to Figure 11)



If Overcurrent Protection Circuit is Used

If Overcurrent Protection Circuit is Not Used

Figure - 11 CL and VS Terminal Connections

6. Setting the Timer to Reset Overcurrent Protection

The time required for the unit to reset from the overcurrent protection state (T_{ocprst}) is determined by the size of the capacitor (C_{ocp}) connected to the OCPC terminals.

$$T_{ocprst} = 8 \times 10^5 \times C_{ocp} \quad [\text{sec}] \quad (\text{typ.})$$

7. Setting the soft start timer

The soft start time is determined by the size of the capacitor (C_{ss}) connected between SS1, SS2 to GND.

$$T_{ss} = \frac{V_{ss} \times C_{ss}}{I_{ss}}$$

The soft start is from VIN standup to Output voltage 100%.