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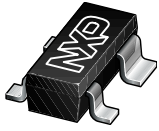
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BFU530XR

NPN wideband silicon RF transistor

Rev. 1 — 5 March 2014

Product data sheet

1. Product profile

1.1 General description

NPN silicon RF transistor for high speed, low noise applications in a plastic, 4-pin dual-emitter SOT143R package.

The BFU530XR is part of the BFU5 family of transistors, suitable for small signal to medium power applications up to 2 GHz.

1.2 Features and benefits

- Low noise, high breakdown RF transistor
- AEC-Q101 qualified
- Minimum noise figure (NF_{min}) = 0.65 dB at 900 MHz
- Maximum stable gain 21 dB at 900 MHz
- 11 GHz f_T silicon technology

1.3 Applications

- Applications requiring high supply voltages and high breakdown voltages
- Broadband amplifiers up to 2 GHz
- Low noise amplifiers for ISM applications
- ISM band oscillators

1.4 Quick reference data

Table 1. Quick reference data

$T_{amb} = 25\text{ }^{\circ}\text{C}$ unless otherwise specified

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|-----------|---------------------------|---|-----|------|-----|------|
| V_{CB} | collector-base voltage | open emitter | - | - | 24 | V |
| V_{CE} | collector-emitter voltage | open base | - | - | 12 | V |
| | | shorted base | - | - | 24 | V |
| V_{EB} | emitter-base voltage | open collector | - | - | 2 | V |
| I_C | collector current | | - | 10 | 40 | mA |
| P_{tot} | total power dissipation | $T_{sp} \leq 87\text{ }^{\circ}\text{C}$ [1] | - | - | 450 | mW |
| h_{FE} | DC current gain | $I_C = 10\text{ mA}$; $V_{CE} = 8\text{ V}$ | 60 | 95 | 200 | |
| C_c | collector capacitance | $V_{CB} = 8\text{ V}$; $f = 1\text{ MHz}$ | - | 0.36 | - | pF |
| f_T | transition frequency | $I_C = 15\text{ mA}$; $V_{CE} = 8\text{ V}$; $f = 900\text{ MHz}$ | - | 11 | - | GHz |



Table 1. Quick reference data ...continued

$T_{amb} = 25\text{ }^{\circ}\text{C}$ unless otherwise specified

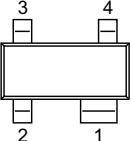
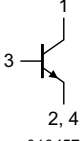
| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|--------------|---------------------------------------|--|-----|------|-----|------|
| $G_{p(max)}$ | maximum power gain | $I_C = 10\text{ mA}$; $V_{CE} = 8\text{ V}$; $f = 900\text{ MHz}$ | - | 21 | - | dB |
| NF_{min} | minimum noise figure | $I_C = 1\text{ mA}$; $V_{CE} = 8\text{ V}$; $f = 900\text{ MHz}$; $\Gamma_S = \Gamma_{opt}$ | - | 0.65 | - | dB |
| $P_{L(1dB)}$ | output power at 1 dB gain compression | $I_C = 15\text{ mA}$; $V_{CE} = 8\text{ V}$; $Z_S = Z_L = 50\text{ }\Omega$; $f = 900\text{ MHz}$ | - | 10.5 | - | dBm |

[1] T_{sp} is the temperature at the solder point of the collector lead.

[2] If $K > 1$ then $G_{p(max)}$ is the maximum power gain. If $K < 1$ then $G_{p(max)} = MSG$.

2. Pinning information

Table 2. Discrete pinning

| Pin | Description | Simplified outline | Graphic symbol |
|-----|-------------|--|--|
| 1 | collector |  |  aaa-010457 |
| 2 | emitter | | |
| 3 | base | | |
| 4 | emitter | | |

3. Ordering information

Table 3. Ordering information

| Type number | Package | | Version |
|-------------|---------|---|---------|
| | Name | Description | |
| BFU530XR | - | plastic surface-mounted package; reverse pinning; 4 leads | SOT143R |
| OM7964 | - | Customer evaluation kit for BFU520XR, BFU530XR and BFU550XR [1] | - |

[1] The customer evaluation kit contains the following:

- Unpopulated RF amplifier Printed-Circuit Board (PCB)
- Unpopulated RF amplifier Printed-Circuit Board (PCB) with emitter degeneration
- Four SMA connectors for fitting unpopulated Printed-Circuit Board (PCB)
- BFU520XR, BFU530XR and BFU550XR samples
- USB stick with data sheets, application notes, models, S-parameter and noise files

4. Marking

Table 4. Marking

| Type number | Marking | Description |
|-------------|---------|--------------------------|
| BFU530XR | *TK | * = t : made in Malaysia |
| | | * = w : made in China |

5. Design support

Table 5. Available design support

Download from the BFU530XR product information page on <http://www.nxp.com>.

| Support item | Available | Remarks |
|---|-----------|---|
| Device models for Agilent EEsof EDA ADS | yes | Based on Mextram device model. |
| SPICE model | yes | Based on Gummel-Poon device model. |
| S-parameters | yes | |
| Noise parameters | yes | |
| Customer evaluation kit | yes | See Section 3 and Section 10 . |
| Solder pattern | yes | |
| Application notes | yes | See Section 10.1 and Section 10.2 . |

6. Limiting values

Table 6. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

| Symbol | Parameter | Conditions | Min | Max | Unit |
|-----------|---------------------------------|---|-----|------|------|
| V_{CB} | collector-base voltage | open emitter | - | 30 | V |
| V_{CE} | collector-emitter voltage | open base | - | 16 | V |
| | | shorted base | - | 30 | V |
| V_{EB} | emitter-base voltage | open collector | - | 3 | V |
| I_C | collector current | | - | 65 | mA |
| T_{stg} | storage temperature | | -65 | +150 | °C |
| V_{ESD} | electrostatic discharge voltage | Human Body Model (HBM) According to JEDEC standard 22-A114E | - | ±150 | V |
| | | Charged Device Model (CDM) According to JEDEC standard 22-C101B | - | ±2 | kV |

7. Recommended operating conditions

Table 7. Characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|-----------|---------------------------|-----------------------------|-----|-----|------|------|
| V_{CB} | collector-base voltage | open emitter | - | - | 24 | V |
| V_{CE} | collector-emitter voltage | open base | - | - | 12 | V |
| | | shorted base | - | - | 24 | V |
| V_{EB} | emitter-base voltage | open collector | - | - | 2 | V |
| I_C | collector current | | - | - | 40 | mA |
| P_i | input power | $Z_S = 50 \Omega$ | - | - | 10 | dBm |
| T_j | junction temperature | | -40 | - | +150 | °C |
| P_{tot} | total power dissipation | $T_{sp} \leq 87 \text{ °C}$ | [1] | - | 450 | mW |

[1] T_{sp} is the temperature at the solder point of the collector lead.

8. Thermal characteristics

Table 8. Thermal characteristics

| Symbol | Parameter | Conditions | Typ | Unit |
|----------------|--|------------|---------|------|
| $R_{th(j-sp)}$ | thermal resistance from junction to solder point | | [1] 140 | K/W |

[1] T_{sp} is the temperature at the solder point of the collector lead.

T_{sp} has the following relation to the ambient temperature T_{amb} :

$$T_{sp} = T_{amb} + P \times R_{th(sp-a)}$$

With P being the power dissipation and $R_{th(sp-a)}$ being the thermal resistance between the solder point and ambient. $R_{th(sp-a)}$ is determined by the heat transfer properties in the application.

The heat transfer properties are set by the application board materials, the board layout and the environment e.g. housing.

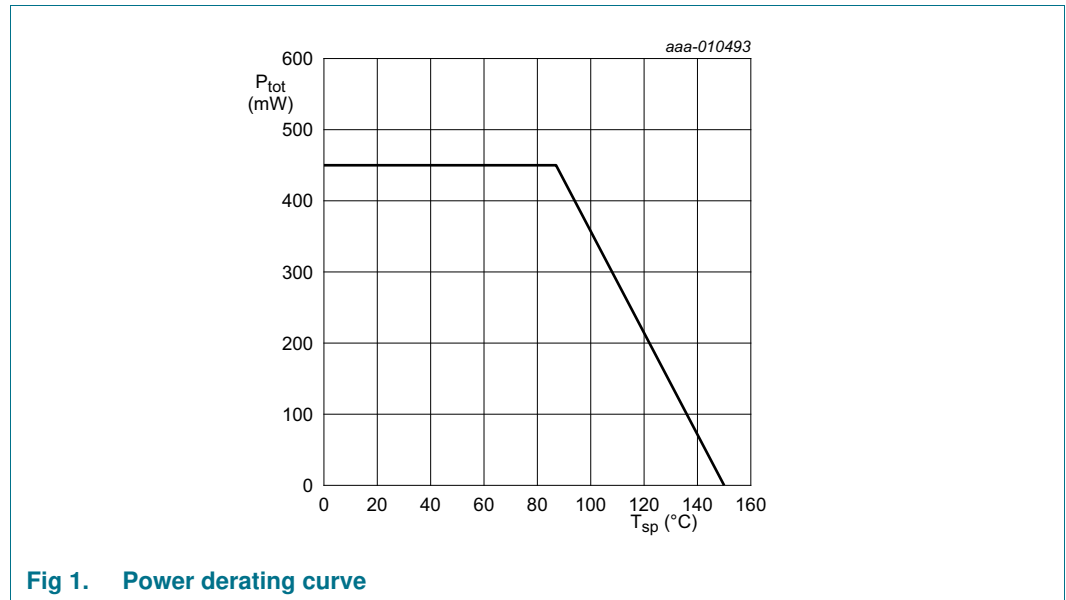


Fig 1. Power derating curve

9. Characteristics

Table 9. Characteristics

$T_{amb} = 25\text{ °C}$ unless otherwise specified

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|---------------|-------------------------------------|---|-----|------|-----|------|
| $V_{(BR)CBO}$ | collector-base breakdown voltage | $I_C = 100\text{ nA}; I_E = 0\text{ mA}$ | 24 | - | - | V |
| $V_{(BR)CEO}$ | collector-emitter breakdown voltage | $I_C = 150\text{ nA}; I_B = 0\text{ mA}$ | 12 | - | - | V |
| I_C | collector current | | - | 10 | 40 | mA |
| I_{CBO} | collector-base cut-off current | $I_E = 0\text{ mA}; V_{CB} = 8\text{ V}$ | - | <1 | - | nA |
| h_{FE} | DC current gain | $I_C = 10\text{ mA}; V_{CE} = 8\text{ V}$ | 60 | 95 | 200 | |
| C_{EBS} | emitter-base capacitance | $V_{CE} = 8\text{ V}; f = 1\text{ MHz}$ | - | 0.71 | - | pF |
| C_{CES} | collector-emitter capacitance | $V_{EB} = 0.5\text{ V}; f = 1\text{ MHz}$ | - | 0.44 | - | pF |
| C_{CBS} | collector-base capacitance | $V_{CB} = 8\text{ V}; f = 1\text{ MHz}$ | - | 0.36 | - | pF |
| f_T | transition frequency | $I_C = 15\text{ mA}; V_{CE} = 8\text{ V}; f = 900\text{ MHz}$ | - | 11 | - | GHz |

Table 9. Characteristics ...continued
 $T_{amb} = 25\text{ }^{\circ}\text{C}$ unless otherwise specified

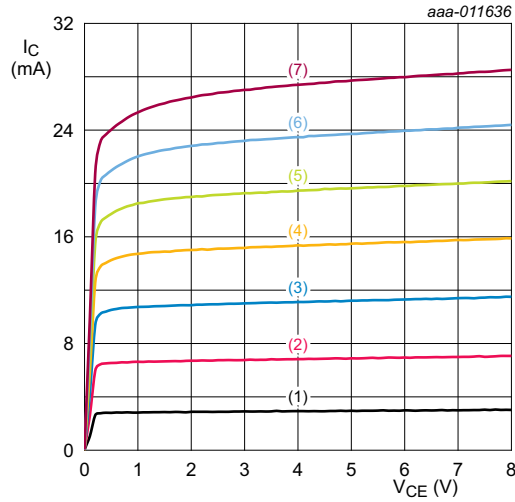
| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|--------------|----------------------|---|-----|------|-----|------|
| $G_{p(max)}$ | maximum power gain | $f = 433\text{ MHz}; V_{CE} = 8\text{ V}$ [1] | | | | |
| | | $I_C = 1\text{ mA}$ | - | 15.5 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 24.5 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 26 | - | dB |
| | | $f = 900\text{ MHz}; V_{CE} = 8\text{ V}$ [1] | | | | |
| | | $I_C = 1\text{ mA}$ | - | 12.5 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 21 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 21.5 | - | dB |
| | | $f = 1800\text{ MHz}; V_{CE} = 8\text{ V}$ [1] | | | | |
| | | $I_C = 1\text{ mA}$ | - | 10.5 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 17 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 16.5 | - | dB |
| $ S_{21} ^2$ | insertion power gain | $f = 433\text{ MHz}; V_{CE} = 8\text{ V}$ | | | | |
| | | $I_C = 1\text{ mA}$ | - | 10.5 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 23 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 23.5 | - | dB |
| | | $f = 900\text{ MHz}; V_{CE} = 8\text{ V}$ | | | | |
| | | $I_C = 1\text{ mA}$ | - | 8.5 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 18 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 18 | - | dB |
| | | $f = 1800\text{ MHz}; V_{CE} = 8\text{ V}$ | | | | |
| | | $I_C = 1\text{ mA}$ | - | 5.5 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 12 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 12.5 | - | dB |
| NF_{min} | minimum noise figure | $f = 433\text{ MHz}; V_{CE} = 8\text{ V}; \Gamma_S = \Gamma_{opt}$ | | | | |
| | | $I_C = 1\text{ mA}$ | - | 0.55 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 0.85 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 0.95 | - | dB |
| | | $f = 900\text{ MHz}; V_{CE} = 8\text{ V}; \Gamma_S = \Gamma_{opt}$ | | | | |
| | | $I_C = 1\text{ mA}$ | - | 0.65 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 0.9 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 1.0 | - | dB |
| | | $f = 1800\text{ MHz}; V_{CE} = 8\text{ V}; \Gamma_S = \Gamma_{opt}$ | | | | |
| | | $I_C = 1\text{ mA}$ | - | 0.85 | - | dB |
| | | $I_C = 10\text{ mA}$ | - | 1.0 | - | dB |
| | | $I_C = 15\text{ mA}$ | - | 1.1 | - | dB |

Table 9. Characteristics ...continued
 $T_{amb} = 25\text{ }^{\circ}\text{C}$ unless otherwise specified

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|---------------------|---------------------------------------|--|-----|------|-----|------|
| G _{ass} | associated gain | f = 433 MHz; V _{CE} = 8 V; $\Gamma_S = \Gamma_{opt}$ | | | | |
| | | I _C = 1 mA | - | 23.5 | - | dB |
| | | I _C = 10 mA | - | 25 | - | dB |
| | | I _C = 15 mA | - | 25 | - | dB |
| | | f = 900 MHz; V _{CE} = 8 V; $\Gamma_S = \Gamma_{opt}$ | | | | |
| | | I _C = 1 mA | - | 16 | - | dB |
| | | I _C = 10 mA | - | 19 | - | dB |
| | | I _C = 15 mA | - | 19.5 | - | dB |
| | | f = 1800 MHz; V _{CE} = 8 V; $\Gamma_S = \Gamma_{opt}$ | | | | |
| | | I _C = 1 mA | - | 10 | - | dB |
| | | I _C = 10 mA | - | 13.5 | - | dB |
| | | I _C = 15 mA | - | 14 | - | dB |
| P _{L(1dB)} | output power at 1 dB gain compression | f = 433 MHz; V _{CE} = 8 V; Z _S = Z _L = 50 Ω | | | | |
| | | I _C = 10 mA | - | 6.5 | - | dBm |
| | | I _C = 15 mA | - | 9.5 | - | dBm |
| | | f = 900 MHz; V _{CE} = 8 V; Z _S = Z _L = 50 Ω | | | | |
| | | I _C = 10 mA | - | 7.5 | - | dBm |
| | | I _C = 15 mA | - | 10.5 | - | dBm |
| | | f = 1800 MHz; V _{CE} = 8 V; Z _S = Z _L = 50 Ω | | | | |
| | | I _C = 10 mA | - | 8 | - | dBm |
| | | I _C = 15 mA | - | 10 | - | dBm |
| IP _{3o} | output third-order intercept point | f ₁ = 433 MHz; f ₂ = 434 MHz; V _{CE} = 8 V; Z _S = Z _L = 50 Ω | | | | |
| | | I _C = 10 mA | - | 16 | - | dBm |
| | | I _C = 15 mA | - | 19 | - | dBm |
| | | f ₁ = 900 MHz; f ₂ = 901 MHz; V _{CE} = 8 V; Z _S = Z _L = 50 Ω | | | | |
| | | I _C = 10 mA | - | 17 | - | dBm |
| | | I _C = 15 mA | - | 20 | - | dBm |
| | | f ₁ = 1800 MHz; f ₂ = 1801 MHz; V _{CE} = 8 V; Z _S = Z _L = 50 Ω | | | | |
| | | I _C = 10 mA | - | 18 | - | dBm |
| | | I _C = 15 mA | - | 20 | - | dBm |

[1] If $K > 1$ then G_{p(max)} is the maximum power gain. If $K < 1$ then G_{p(max)} = MSG.

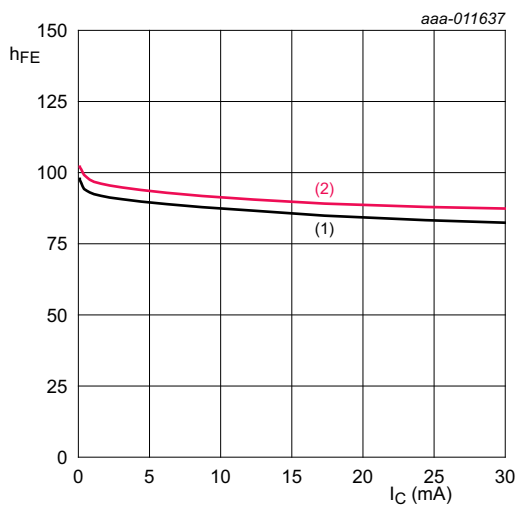
9.1 Graphs



$T_{amb} = 25\text{ }^\circ\text{C}$.

- (1) $I_B = 25\text{ }\mu\text{A}$
- (2) $I_B = 75\text{ }\mu\text{A}$
- (3) $I_B = 125\text{ }\mu\text{A}$
- (4) $I_B = 175\text{ }\mu\text{A}$
- (5) $I_B = 225\text{ }\mu\text{A}$
- (6) $I_B = 275\text{ }\mu\text{A}$
- (7) $I_B = 325\text{ }\mu\text{A}$

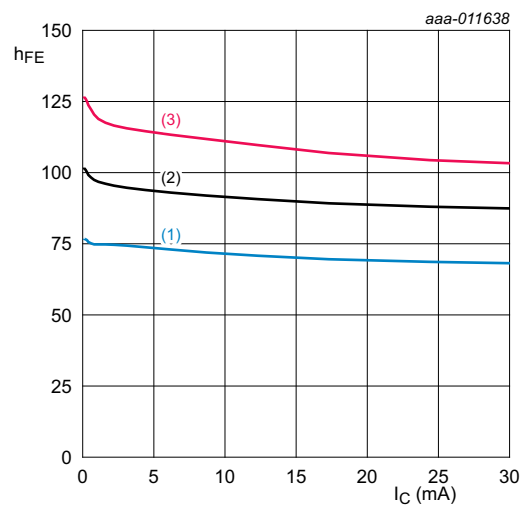
Fig 2. Collector current as a function of collector-emitter voltage; typical values



$T_{amb} = 25\text{ }^\circ\text{C}$.

- (1) $V_{CE} = 3.0\text{ V}$
- (2) $V_{CE} = 8.0\text{ V}$

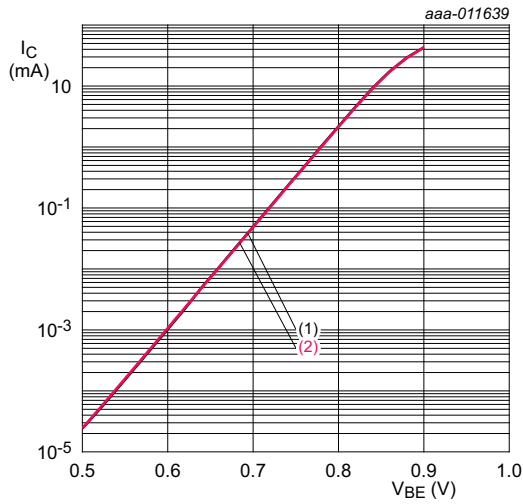
Fig 3. DC current gain as function of collector current; typical values



$V_{CE} = 8\text{ V}$.

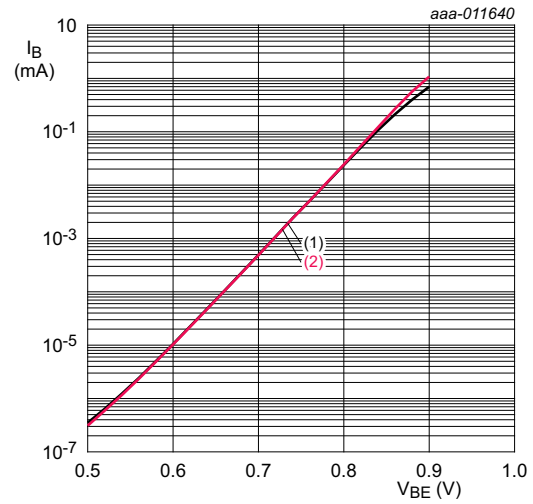
- (1) $T_{amb} = -40\text{ }^\circ\text{C}$
- (2) $T_{amb} = +25\text{ }^\circ\text{C}$
- (3) $T_{amb} = +125\text{ }^\circ\text{C}$

Fig 4. DC current gain as function of collector current; typical values



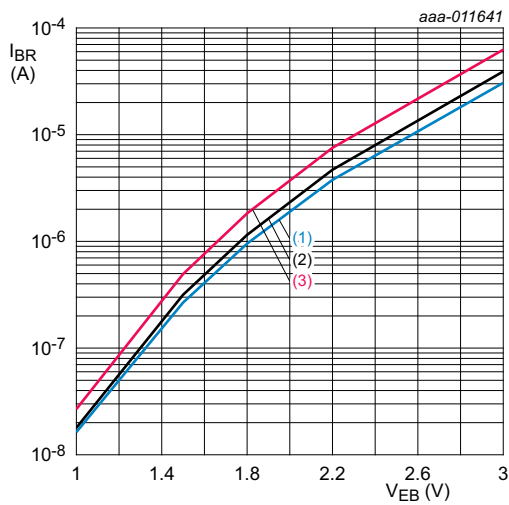
$T_{amb} = 25$ °C.
 (1) $V_{CE} = 3.0$ V
 (2) $V_{CE} = 8.0$ V

Fig 5. Collector current as a function of base-emitter voltage; typical values



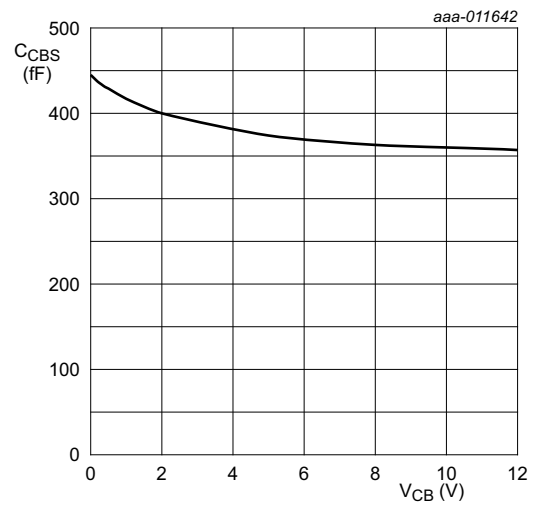
$T_{amb} = 25$ °C.
 (1) $V_{CE} = 3.0$ V
 (2) $V_{CE} = 8.0$ V

Fig 6. Base current as a function of base-emitter voltage; typical values



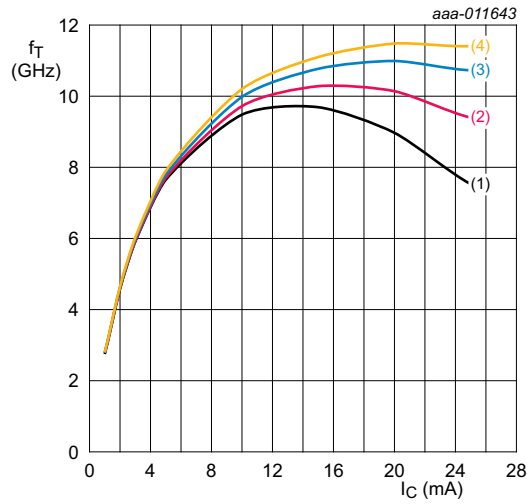
$V_{CE} = 3$ V.
 (1) $T_{amb} = -40$ °C
 (2) $T_{amb} = +25$ °C
 (3) $T_{amb} = +125$ °C

Fig 7. Reverse base current as a function of emitter-base voltage; typical values



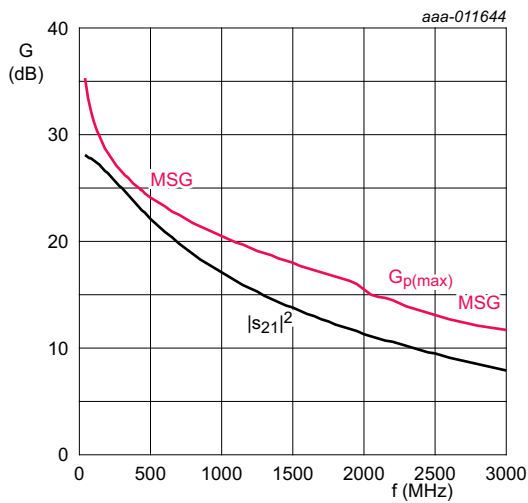
$I_C = 0$ mA; $f = 1$ MHz; $T_{amb} = 25$ °C.

Fig 8. Collector-base capacitance as a function of collector-base voltage; typical values



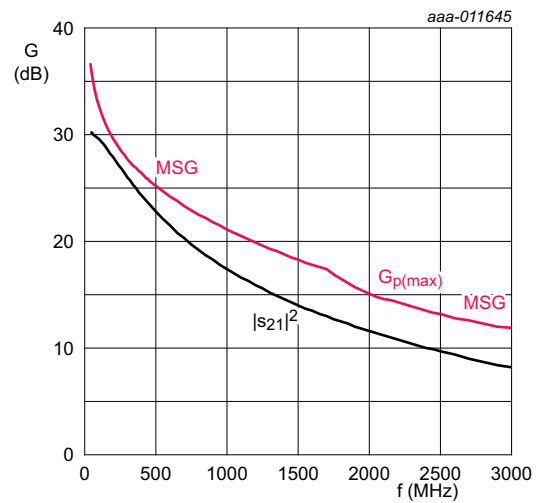
- $T_{amb} = 25\text{ }^{\circ}\text{C}.$
- (1) $V_{CE} = 3.3\text{ V}$
 - (2) $V_{CE} = 5.0\text{ V}$
 - (3) $V_{CE} = 8.0\text{ V}$
 - (4) $V_{CE} = 12.0\text{ V}$

Fig 9. Transition frequency as a function of collector current; typical values



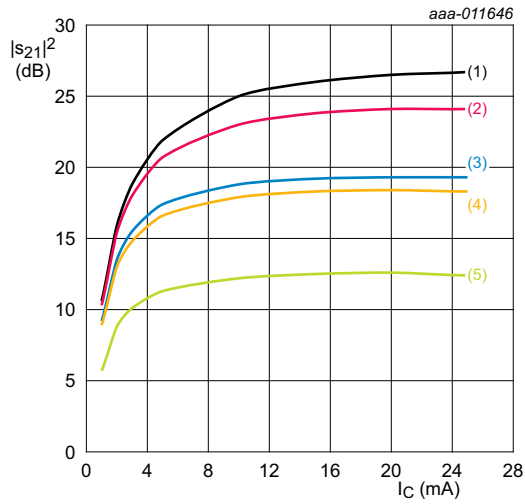
$I_C = 10\text{ mA}; V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C}.$

Fig 10. Gain as a function of frequency; typical values



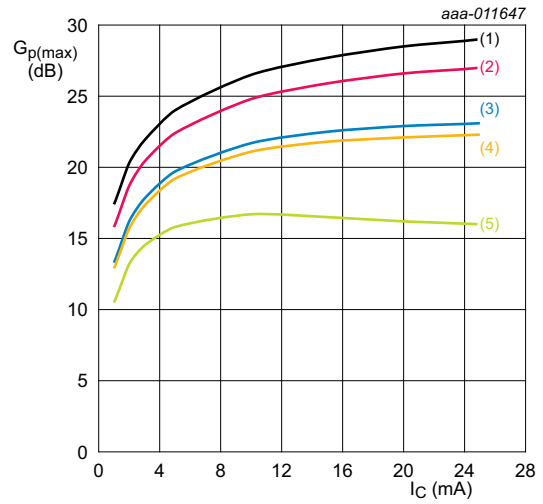
$I_C = 15\text{ mA}; V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C}.$

Fig 11. Gain as a function of frequency; typical values



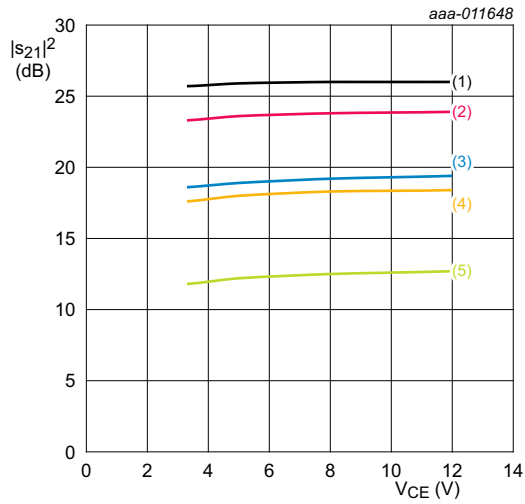
$V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^\circ\text{C}.$
 (1) $f = 300\text{ MHz}$
 (2) $f = 433\text{ MHz}$
 (3) $f = 800\text{ MHz}$
 (4) $f = 900\text{ MHz}$
 (5) $f = 1800\text{ MHz}$

Fig 12. Insertion power gain as a function of collector current; typical values



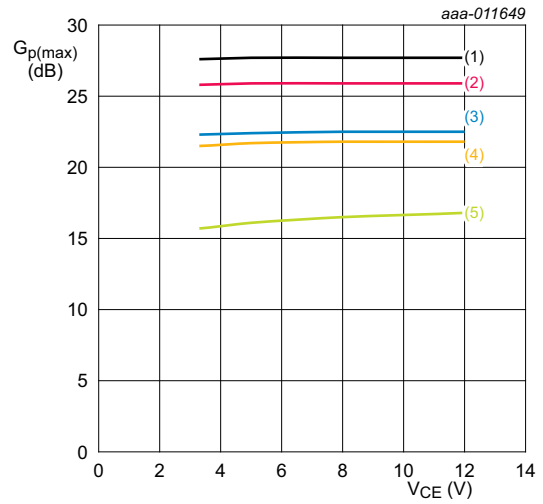
$V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^\circ\text{C}.$
 If $K > 1$ then $G_{p(max)}$ = maximum power gain. If $K < 1$ then $G_{p(max)}$ = MSG.
 (1) $f = 300\text{ MHz}$
 (2) $f = 433\text{ MHz}$
 (3) $f = 800\text{ MHz}$
 (4) $f = 900\text{ MHz}$
 (5) $f = 1800\text{ MHz}$

Fig 13. Maximum power gain as a function of collector current; typical values



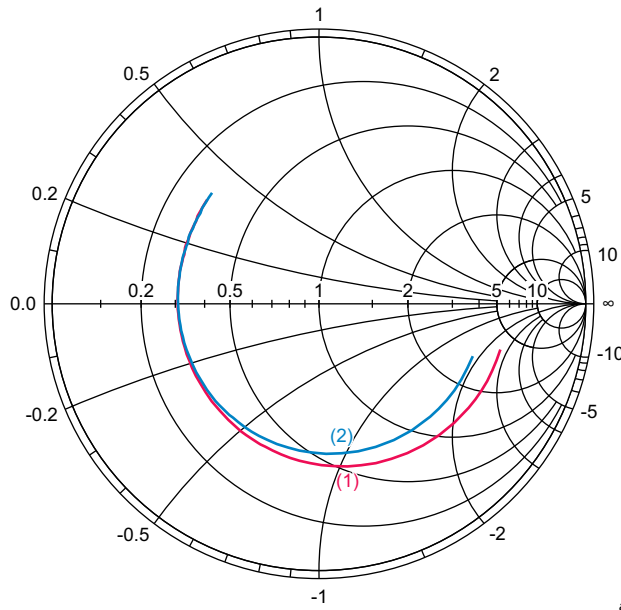
$I_C = 15 \text{ mA}; T_{\text{amb}} = 25 \text{ }^\circ\text{C}.$
 (1) $f = 300 \text{ MHz}$
 (2) $f = 433 \text{ MHz}$
 (3) $f = 800 \text{ MHz}$
 (4) $f = 900 \text{ MHz}$
 (5) $f = 1800 \text{ MHz}$

Fig 14. Insertion power gain as a function of collector-emitter voltage; typical values



$I_C = 15 \text{ mA}; T_{\text{amb}} = 25 \text{ }^\circ\text{C}.$
 If $K > 1$ then $G_{p(\text{max})}$ = maximum power gain. If $K < 1$ then $G_{p(\text{max})}$ = MSG.
 (1) $f = 300 \text{ MHz}$
 (2) $f = 433 \text{ MHz}$
 (3) $f = 800 \text{ MHz}$
 (4) $f = 900 \text{ MHz}$
 (5) $f = 1800 \text{ MHz}$

Fig 15. Maximum power gain as a function of collector-emitter voltage; typical values

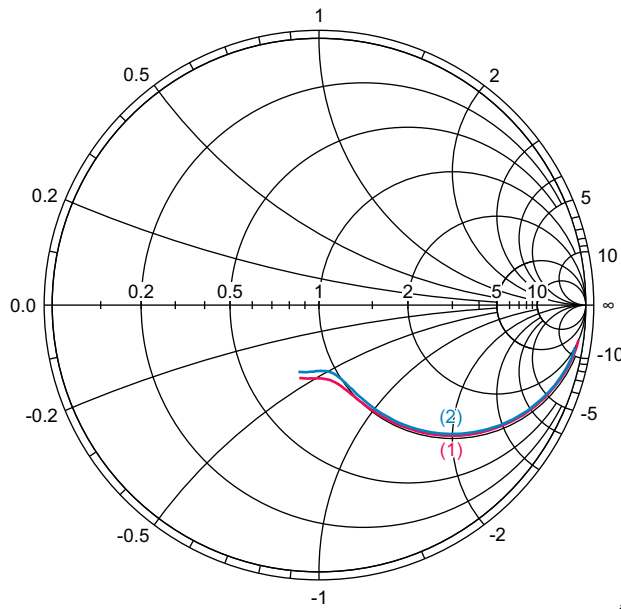


aaa-011650

$V_{CE} = 8\text{ V}; 40\text{ MHz} \leq f \leq 3\text{ GHz}.$

- (1) $I_C = 10\text{ mA}$
- (2) $I_C = 15\text{ mA}$

Fig 16. Input reflection coefficient (s_{11}); typical values

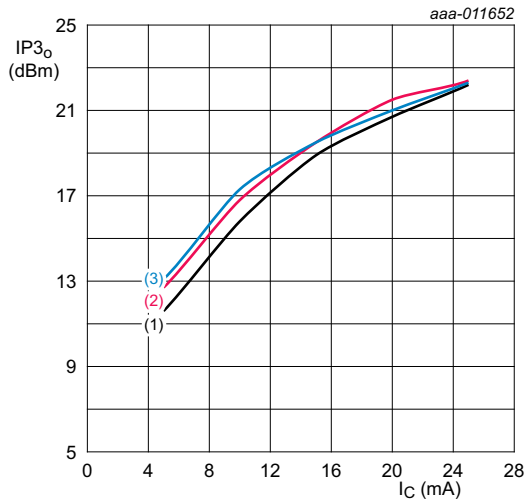


aaa-011651

$V_{CE} = 8\text{ V}; 40\text{ MHz} \leq f \leq 3\text{ GHz}.$

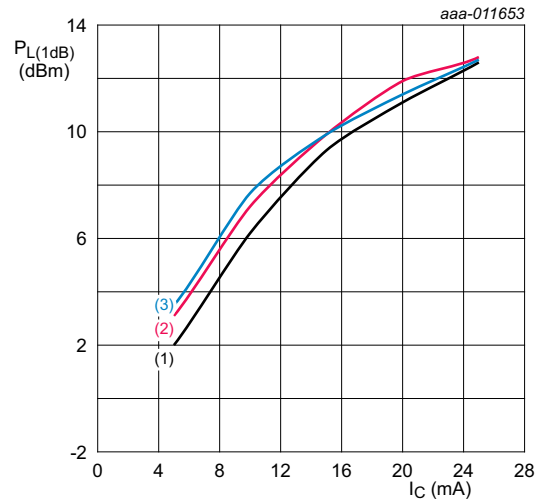
- (1) $I_C = 10\text{ mA}$
- (2) $I_C = 15\text{ mA}$

Fig 17. Output reflection coefficient (s_{22}); typical values



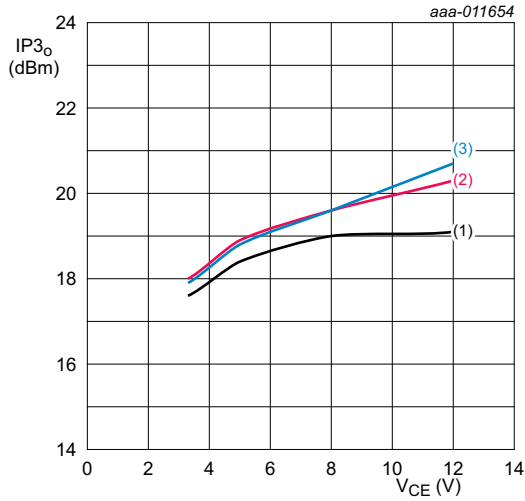
$V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C}.$
 (1) $f_1 = 433\text{ MHz}; f_2 = 434\text{ MHz}$
 (2) $f_1 = 900\text{ MHz}; f_2 = 901\text{ MHz}$
 (3) $f_1 = 1800\text{ MHz}; f_2 = 1801\text{ MHz}$

Fig 18. Output third-order intercept point as a function of collector current; typical values



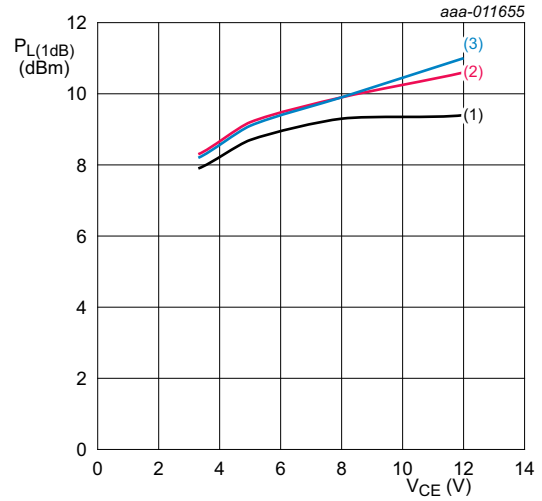
$V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C}.$
 (1) $f = 433\text{ MHz}$
 (2) $f = 900\text{ MHz}$
 (3) $f = 1800\text{ MHz}$

Fig 19. Output power at 1 dB gain compression as a function of collector current; typical values



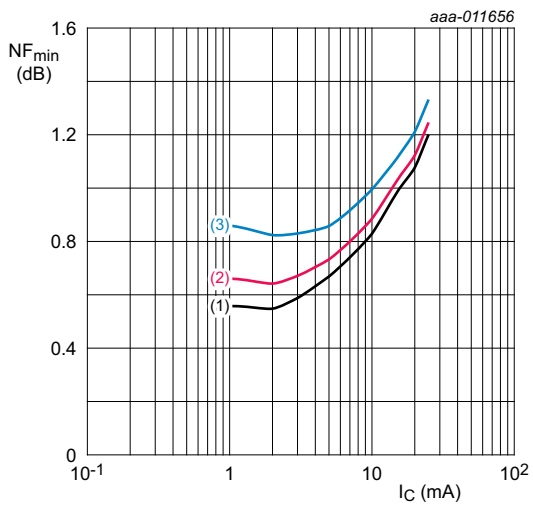
$I_C = 15\text{ mA}; T_{amb} = 25\text{ }^{\circ}\text{C}.$
 (1) $f_1 = 433\text{ MHz}; f_2 = 434\text{ MHz}$
 (2) $f_1 = 900\text{ MHz}; f_2 = 901\text{ MHz}$
 (3) $f_1 = 1800\text{ MHz}; f_2 = 1801\text{ MHz}$

Fig 20. Output third-order intercept point as a function of collector-emitter voltage; typical values



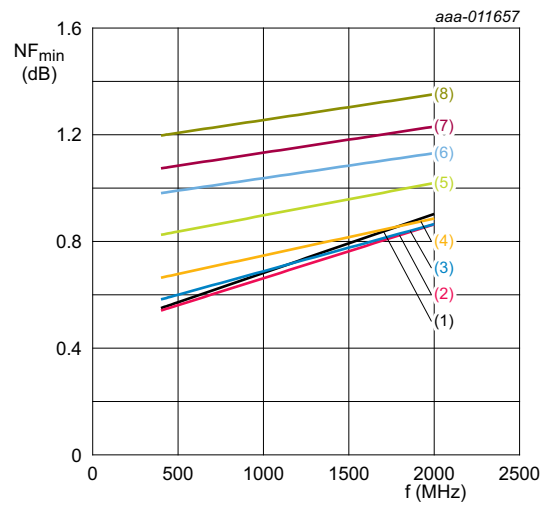
$I_C = 15\text{ mA}; T_{amb} = 25\text{ }^{\circ}\text{C}.$
 (1) $f = 433\text{ MHz}$
 (2) $f = 900\text{ MHz}$
 (3) $f = 1800\text{ MHz}$

Fig 21. Output power at 1 dB gain compression as a function of collector-emitter voltage; typical values



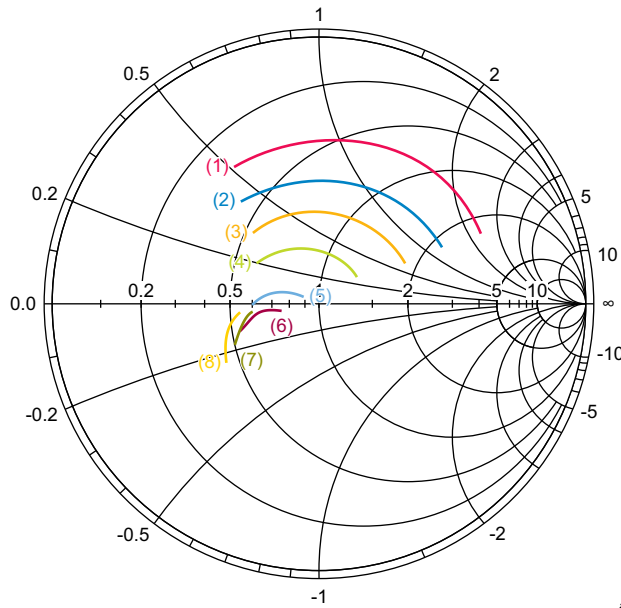
$V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^\circ\text{C}; \Gamma_S = \Gamma_{opt}$
 (1) $f = 433\text{ MHz}$
 (2) $f = 900\text{ MHz}$
 (3) $f = 1800\text{ MHz}$

Fig 22. Minimum noise figure as a function of collector current; typical values



$V_{CE} = 8\text{ V}; T_{amb} = 25\text{ }^\circ\text{C}; \Gamma_S = \Gamma_{opt}$
 (1) $I_C = 1\text{ mA}$
 (2) $I_C = 2\text{ mA}$
 (3) $I_C = 3\text{ mA}$
 (4) $I_C = 5\text{ mA}$
 (5) $I_C = 10\text{ mA}$
 (6) $I_C = 15\text{ mA}$
 (7) $I_C = 20\text{ mA}$
 (8) $I_C = 25\text{ mA}$

Fig 23. Minimum noise figure as a function of frequency; typical values



$V_{CE} = 8\text{ V}; 400\text{ MHz} \leq f \leq 2\text{ GHz}.$

- (1) $I_C = 1\text{ mA}$
- (2) $I_C = 2\text{ mA}$
- (3) $I_C = 3\text{ mA}$
- (4) $I_C = 5\text{ mA}$
- (5) $I_C = 10\text{ mA}$
- (6) $I_C = 15\text{ mA}$
- (7) $I_C = 20\text{ mA}$
- (8) $I_C = 25\text{ mA}$

Fig 24. Optimum reflection coefficient (Γ_{opt}); typical values

10. Application information

More information about the following application example can be found in the application notes. See [Section 5 “Design support”](#).

The following application example can be implemented using the evaluation kit. See [Section 3 “Ordering information”](#) for the order type number.

The following application example can be simulated using the simulation package. See [Section 5 “Design support”](#).

10.1 Application example: 433 ISM band LNA

433 ISM band LNA, optimized for low noise.

More detailed information of the application example can be found in the application note: *AN11441*

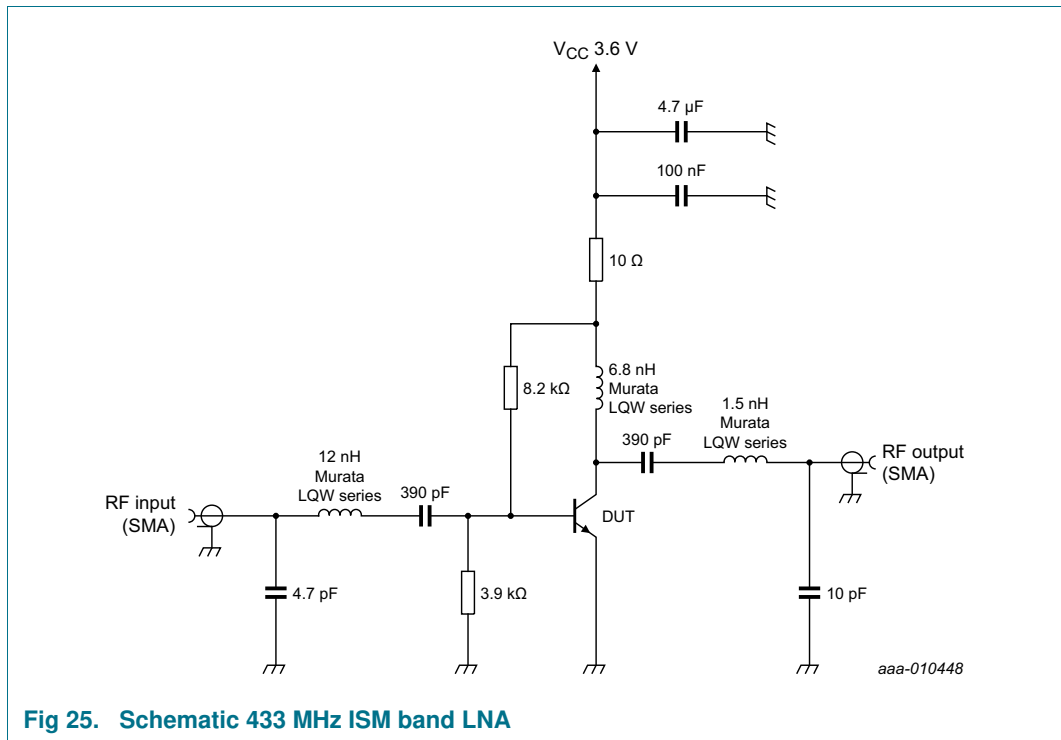


Fig 25. Schematic 433 MHz ISM band LNA

Remark: fine tuning of components maybe required depending on PCB parasitics.

Table 10. Application performance data at 433 MHz

$I_{CC} = 10 \text{ mA}; V_{CC} = 3.6 \text{ V}$

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|------------------|------------------------------------|---|-----|-----|-----|------|
| $ S_{21} ^2$ | insertion power gain | | - | 18 | - | dB |
| NF | noise figure | | - | 1.1 | - | dB |
| IP3 _o | output third-order intercept point | $f_1 = 433 \text{ MHz}; f_2 = 433.1 \text{ MHz}; P_1 = -30 \text{ dBm per carrier}$ | - | 9 | - | dBm |

10.2 Application example: 866 ISM band LNA

866 ISM band LNA, optimized for low noise.

More detailed information of the application example can be found in the application note: AN11442

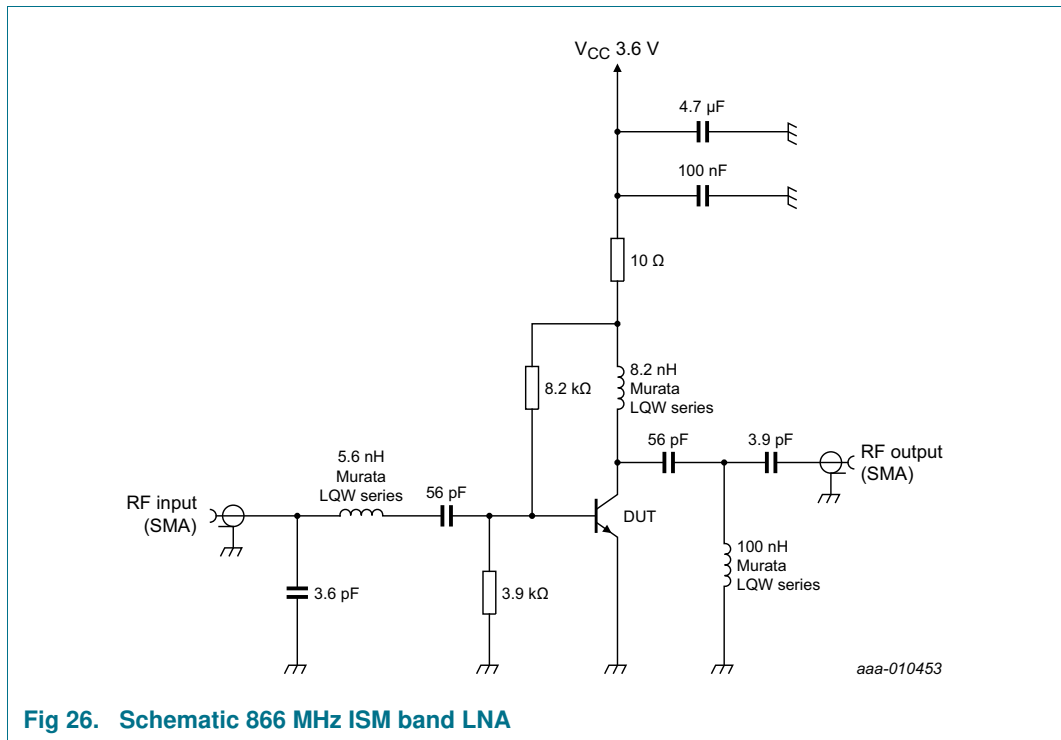


Fig 26. Schematic 866 MHz ISM band LNA

Remark: fine tuning of components may be required depending on PCB parasitics.

Table 11. Application performance data at 866 MHz

$I_{CC} = 10 \text{ mA}$; $V_{CC} = 3.6 \text{ V}$

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|------------------|------------------------------------|---|-----|-----|-----|------|
| $ S_{21} ^2$ | insertion power gain | | - | 16 | - | dB |
| NF | noise figure | | - | 1.1 | - | dB |
| IP3 _o | output third-order intercept point | $f_1 = 866.1 \text{ MHz}$; $f_2 = 866.2 \text{ MHz}$; $P_i = -30 \text{ dBm}$ per carrier | - | 17 | - | dBm |

11. Package outline

Plastic surface-mounted package; reverse pinning; 4 leads

SOT143R

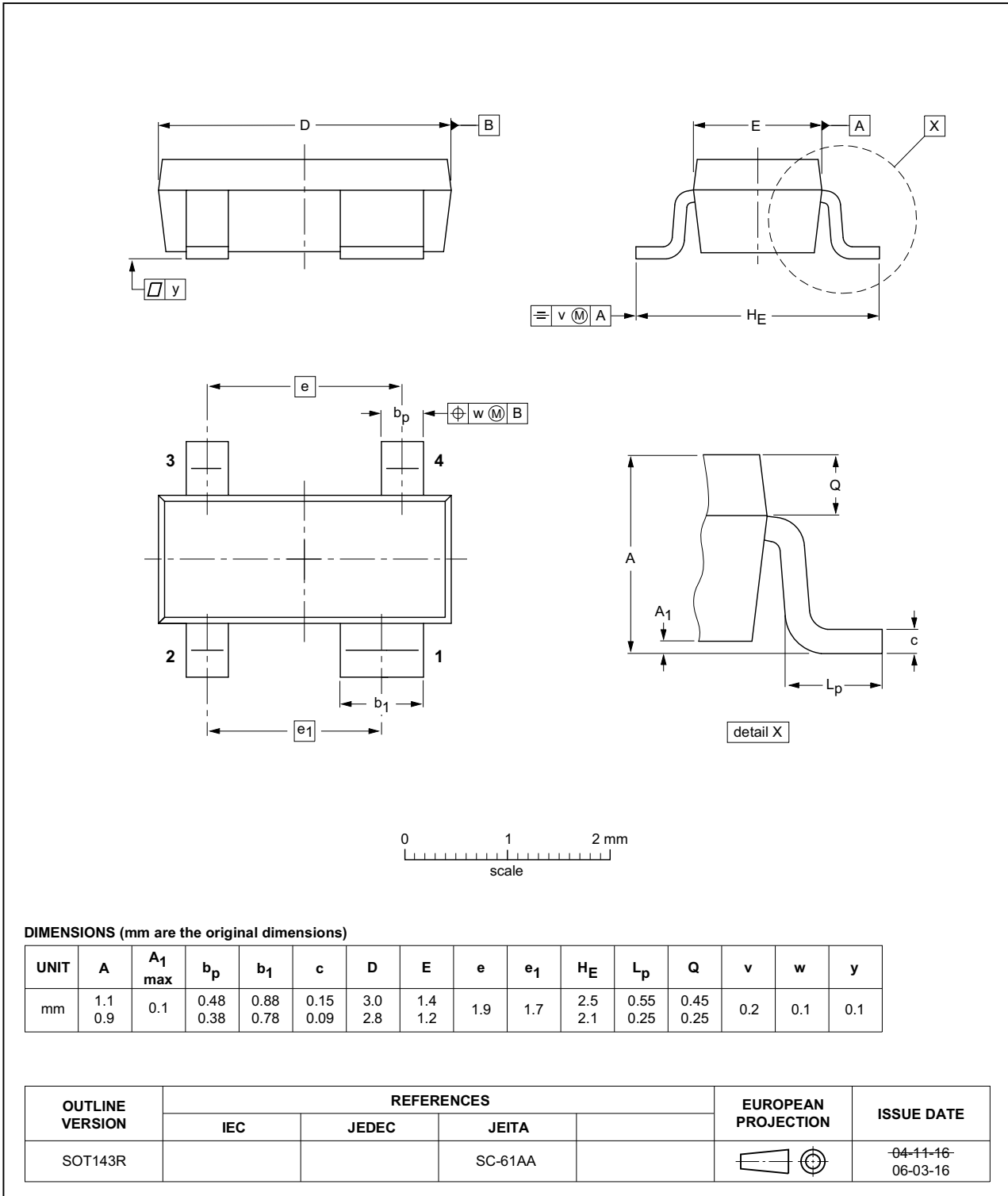


Fig 27. Package outline SOT143R

12. Handling information

CAUTION



This device is sensitive to ElectroStatic Discharge (ESD). Observe precautions for handling electrostatic sensitive devices.

Such precautions are described in the *ANSI/ESD S20.20*, *IEC/ST 61340-5*, *JESD625-A* or equivalent standards.

13. Abbreviations

Table 12. Abbreviations

| Acronym | Description |
|---------|------------------------------------|
| AEC | Automotive Electronics Council |
| ISM | Industrial, Scientific and Medical |
| LNA | Low-Noise Amplifier |
| MSG | Maximum Stable Gain |
| NPN | Negative-Positive-Negative |
| SMA | SubMiniature version A |

14. Revision history

Table 13. Revision history

| Document ID | Release date | Data sheet status | Change notice | Supersedes |
|--------------|--------------|--------------------|---------------|------------|
| BFU530XR v.1 | 20140305 | Product data sheet | - | - |

15. Legal information

15.1 Data sheet status

| Document status ^{[1][2]} | Product status ^[3] | Definition |
|-----------------------------------|-------------------------------|---|
| Objective [short] data sheet | Development | This document contains data from the objective specification for product development. |
| Preliminary [short] data sheet | Qualification | This document contains data from the preliminary specification. |
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[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

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