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# BGX7100

## Transmitter IQ modulator

Rev. 5 — 3 September 2012

Product data sheet

### 1. General description

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The BGX7100 device combines high performance, high linearity I and Q modulation paths for use in radio frequency up-conversion. It supports RF frequency outputs in the range from 400 MHz to 4000 MHz. The BGX7100 IQ modulator is performance independent of the IQ common mode voltage. The modulator provides a typical output power at 1 dB gain compression ( $P_{L(1dB)}$ ) value of 12 dBm and a typical 27 dBm output third-order intercept point ( $IP3_o$ ). Unadjusted sideband suppression and carrier feedthrough are 50 dBc and -45 dBm respectively. A hardware control pin provides a fast power-down/power-up mode functionality which allows significant power saving.

### 2. Features and benefits

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- 400 MHz to 4000 MHz frequency operating range
- Stable performance across 0.25 V to 3.3 V common-mode voltage input
- Independent low-current power-down hardware control pin
- 12 dBm output -1 dB compression point
- 27 dBm output third-order intercept point (typical)
- Integrated active biasing
- Single 5 V supply
- 180  $\Omega$  differential IQ input impedance
- Matched 50  $\Omega$  single-ended RF output impedance
- ESD protection at all pins

### 3. Applications

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- Mobile network infrastructure
- Microwave and broadband
- RF and IF applications
- Industrial applications

### 4. Device family

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The BGX7100 operates in the RF frequency range of 400 MHz to 4000 MHz with modulation bandwidths up to 400 MHz.

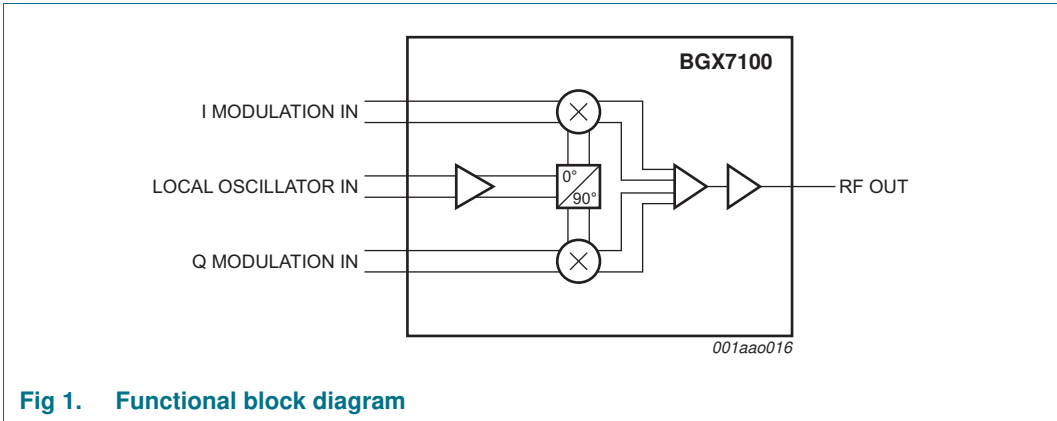


## 5. Ordering information

Table 1. Ordering information

Type number	Package		Version
	Name	Description	
BGX7100HN	HVQFN24	plastic thermal enhanced very thin quad flat package; no leads; 24 terminals; body 4 × 4 × 0.85 mm	SOT616-3

## 6. Functional diagram



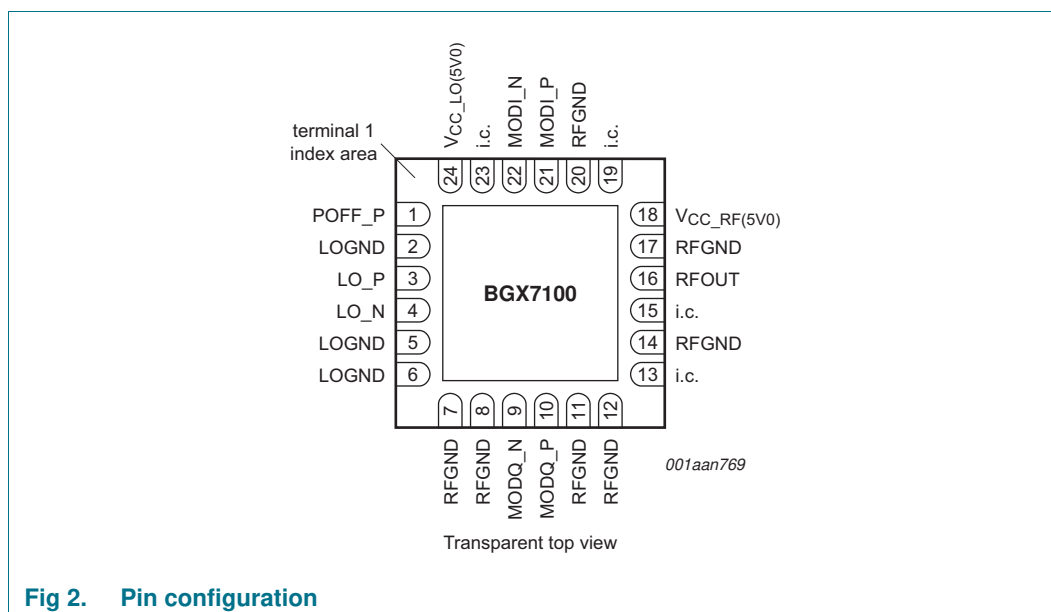
Differential I and Q baseband inputs are each fed to an associated upconverter mixer. The Local Oscillator (LO) carrier input is buffered and split into 0 degree and 90 degree signals. The in-phase signal is passed to the I mixer and the 90 degree phase-changed signal is passed to the Q mixer. The outputs of the mixers are summed to produce the resulting RF output signal.

## 7. Pinning information

### 7.1 Pinning

The BGX7100 device pinout is designed to allow easy interfacing when mounted on a Printed-Circuit Board (PCB). When viewing the device from above, the two differential IQ baseband input paths are at the top and bottom. The common LO input is at the left and the RF output at the right. Multiple power and ground pins allow for independent supply domains, improving isolation between blocks. A small package footprint is chosen to reduce bond-wire induced series inductance in the RF ports.

The input and output pin matching is described in [Section 12 “Application information”](#).



## 7.2 Pin description

**Table 2. Pin description**

Symbol	Pin	Type <sup>[1]</sup>	Description
POFF_P	1	I	active HIGH logic input to power-down modulator
LOGND	2	G	LO ground
LO_P	3	I	LO positive input <sup>[2]</sup>
LO_N	4	I	LO negative input <sup>[2]</sup>
LOGND	5	G	LO ground
LOGND	6	G	LO ground
RFGND	7	G	RF ground
RFGND	8	G	RF ground
MODQ_N	9	I	modulator quadrature negative input
MODQ_P	10	I	modulator quadrature positive input
RFGND	11	G	RF ground
RFGND	12	G	RF ground
i.c.	13	-	internally connected; to be tied to ground
RFGND	14	G	RF ground
i.c.	15	-	internally connected; to be tied to ground
RFOUT	16	O	modulator single-ended RF output <sup>[2]</sup>
RFGND	17	G	RF ground
V <sub>CC_RF</sub> (5V0)	18	P	RF analog power supply 5 V
i.c.	19	-	internally connected; to be tied to ground
RFGND	20	G	RF ground
MODI_P	21	I	modulator in-phase positive input
MODI_N	22	I	modulator in-phase negative input



Table 2. Pin description ...continued

Symbol	Pin	Type <sup>[1]</sup>	Description
i.c.	23	-	internally connected; to be tied to ground
V <sub>CC_LO(5V0)</sub>	24	P	LO analog power supply 5 V
Exposed die pad	-	G	exposed die pad; must be connected to RF ground

[1] G = ground; I = input; O = output; P = power.

[2] AC coupling required as shown in [Figure 4 "Typical wideband application diagram"](#).

## 8. Functional description

### 8.1 General

Each IQ baseband input has a 180  $\Omega$  differential input impedance allowing straightforward matching, from the DAC output through the baseband filter. The device allows operation with IQ input common-mode voltages between 0.25 V and 3.3 V allowing direct connection to a broad family of DACs. The LO and RF ports provide broadband 50  $\Omega$  termination to RF source and loads.

The chip can be placed in inactive mode (see [Section 8.2 "Shutdown control"](#)).

### 8.2 Shutdown control

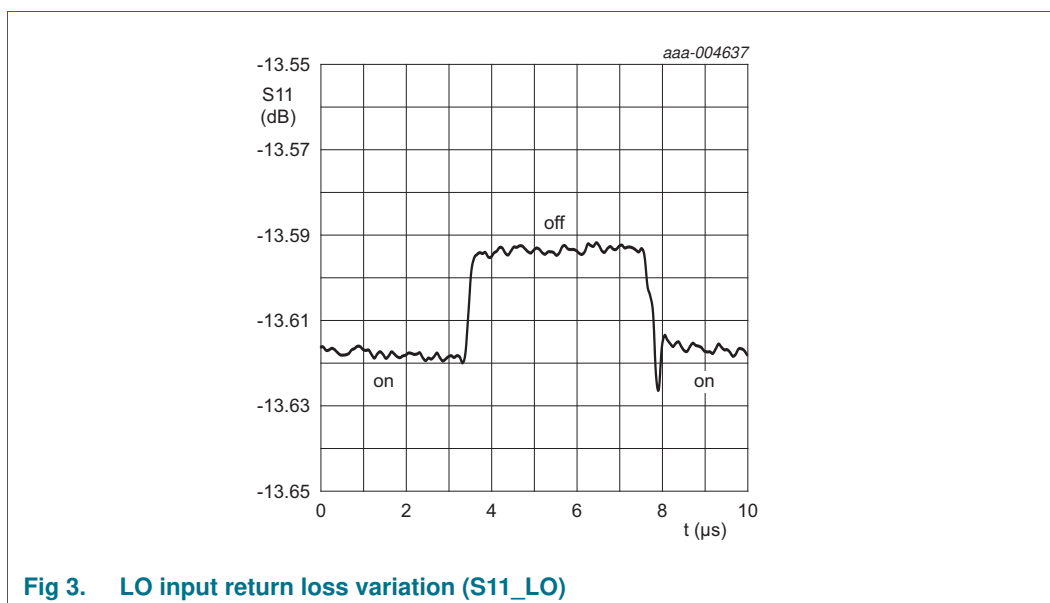
Table 3. Shutdown control

Mode	Mode description	Functional description	POFF_P
Idle	modulator fully off; minimal supply current	shutdown enabled	> 1.5 V
Active	modulator active mode	shutdown disabled	< 0.5 V

The modulator can be placed into inactive mode by the voltage level at power-up disable pin (pin 1, POFF\_P). The time required to pass between active and low-current states is less than 1  $\mu$ s.

The shutdown feature of IQ modulator during switching does not induce any unlock of the LO synthesizer in base station application thanks to the low impedance variation of the LO input.

The graph (see [Figure 3](#)) describes the impact on LO impedance variation during the switching time.



## 9. Limiting values

**Table 4. Limiting values**

*In accordance with the Absolute Maximum Rating System (IEC 60134).*

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{CC}$	supply voltage		-	5.5	V
$P_{i(lo)}$	local oscillator input power		-	16	dBm
$P_{o(RF)}$	RF output power		-	20	dBm
$T_{mb}$	mounting base temperature		-40	+85	°C
$T_j$	junction temperature		-	+150	°C
$T_{stg}$	storage temperature		-65	+150	°C
$V_{ESD}$	electrostatic discharge voltage	EIA/JESD22-A114 (HBM)	-2500	+2500	V
		EIA/JESD22-C101 (FCDM)	-650	+650	V

**Table 4. Limiting values ...continued**

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
<b>Pin POFF_P</b>					
V <sub>i</sub>	input voltage	active HIGH logic input to power-down modulator	-	3.5	V
<b>Pins MODI_N, MODI_P, MODQ_N and MODQ_P</b>					
V <sub>i</sub>	input voltage		0	5	V
V <sub>ID</sub>	differential input voltage	DC	-2	+2	V

## 10. Thermal characteristics

**Table 5. Thermal characteristics**

Symbol	Parameter	Conditions	Typ	Unit
R <sub>th(j-mb)</sub>	thermal resistance from junction to mounting base		10	K/W

## 11. Characteristics

**Table 6. Characteristics**

Modulation source resistance per pin = 90 Ω; POFF\_P connected to GND (shutdown disabled); V<sub>CC</sub> = 5 V; T<sub>mb</sub> range = -40 °C to +85 °C; P<sub>i(lo)</sub> = 0 dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>CC</sub>	supply voltage		4.75	5	5.25	V
I <sub>CC(tot)</sub>	total supply current	modulator in active mode				
		f <sub>lo</sub> = 900 MHz	-	165	-	mA
		f <sub>lo</sub> = 2 GHz	-	173	-	mA
		f <sub>lo</sub> = 2.5 GHz	-	178	-	mA
		f <sub>lo</sub> = 3.5 GHz	-	184	-	mA
		modulator in inactive mode; T <sub>mb</sub> = 25 °C	-	6	-	mA
f <sub>lo</sub>	local oscillator frequency	[1]	400	-	4000	MHz
P <sub>i(lo)</sub>	local oscillator input power	[1]	-9	0	+6	dBm
<b>Pins MODI_x and MODQ_x[2]</b>						
V <sub>i(cm)</sub>	common-mode input voltage		0.25	-	3.3	V
S <sub>22_RF</sub>	RF output return loss		-	10	-	dB
S <sub>11_LO</sub>	LO input return loss		-	12	-	dB
<b>MODI and MODQ[3]</b>						
BW <sub>mod</sub>	modulation bandwidth	gain fall off < 1 dB; R <sub>S</sub> = 90 Ω	-	400	-	MHz
R <sub>i(dif)</sub>	differential input resistance		-	180	-	Ω
C <sub>i(dif)</sub>	differential input capacitance		-	1.8	-	pF

[1] Operation outside this range is possible but parameters are not guaranteed.

[2] x = N or P.

[3] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

**Table 7. Characteristics at 750 MHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	29	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	71	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-159	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158.5	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	55	-	dBc
CF	carrier feedthrough	unadjusted	-	-55	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

**Table 8. Characteristics at 910 MHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	29	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	72	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-159	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158.5	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	49	-	dBc
CF	carrier feedthrough	unadjusted	-	-55	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.



**Table 9. Characteristics at 1.840 GHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	27	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	69	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-158.5	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	47	-	dBc
CF	carrier feedthrough	unadjusted	-	-50	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

**Table 10. Characteristics at 1.960 GHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	27	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	72.5	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-158.5	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	49	-	dBc
CF	carrier feedthrough	unadjusted	-	-48	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

**Table 11. Characteristics at 2.140 GHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	27	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	74	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-158.5	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	51	-	dBc
CF	carrier feedthrough	unadjusted	-	-45	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

**Table 12. Characteristics at 2.650 GHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	26	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	62	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-158	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	60	-	dBc
CF	carrier feedthrough	unadjusted	-	-45	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

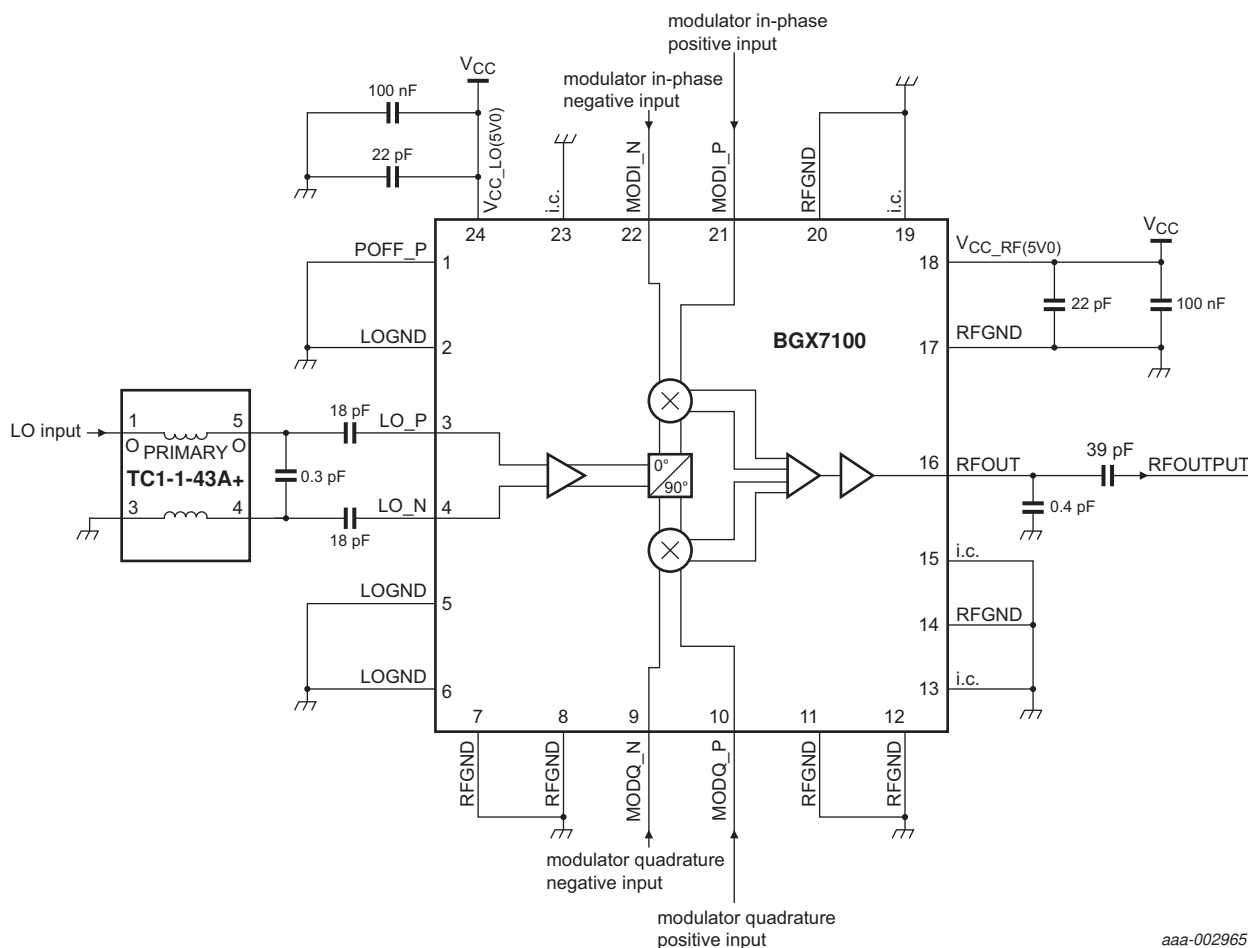
**Table 13. Characteristics at 3.650 GHz**

Modulation source resistance per pin = 90  $\Omega$ ; POFF\_P connected to GND (shutdown disabled);  $V_{CC} = 5$  V;  
 $T_{mb}$  range =  $-40$  °C to  $+85$  °C;  $P_{i(lo)} = 0$  dBm; IQ frequency = 5 MHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_o$	output power	1 V (p-p) differential on MODI and MODQ[1]	-	-0.2	-	dBm
$P_{L(1dB)}$	output power at 1 dB gain compression		-	11.5	-	dBm
$IP3_o$	output third-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	25	-	dBm
$IP2_o$	output second-order intercept point	IQ frequency 1 = 4.5 MHz; IQ frequency 2 = 5.5 MHz; output power per tone = -10 dBm	-	60	-	dBm
$N_{flr(o)}$	output noise floor	no modulation present	-	-158	-	dBm/Hz
		modulation at MODI and MODQ[1]; $P_{o(RF)} = -10$ dBm	-	-158	-	dBm/Hz
SBS	sideband suppression	unadjusted	-	53	-	dBc
CF	carrier feedthrough	unadjusted	-	-43	-	dBm

[1] MODI = MODI\_P – MODI\_N and MODQ = MODQ\_P – MODQ\_N.

## 12. Application information



aaa-002965

Fig 4. Typical wideband application diagram

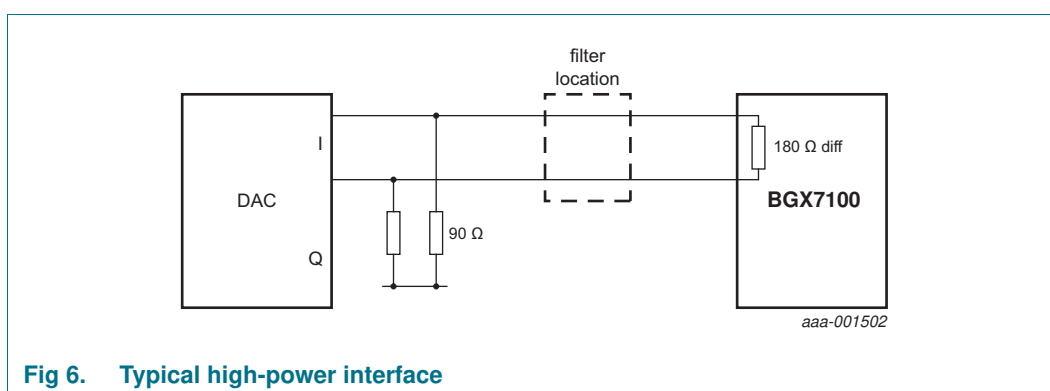
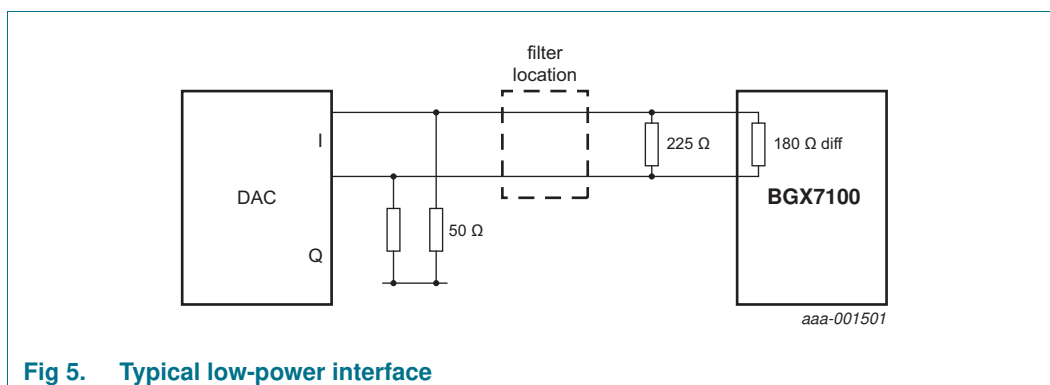
Figure 4 shows a typical wideband (from 0.4 GHz to 4 GHz) application circuit. Refer to the application note for narrowband optimum component values.

### 12.1 External DAC interfacing

Nominal DAC single-ended output currents are between 0 mA and 20 mA.

If the DAC outputs are only designed for 1 V peak-to-peak differential (250 mV peak-single) then the single-ended impedance at the DAC needs to be limited to 25  $\Omega$ . This can be split as 50  $\Omega$  load resistors at the DAC outputs and a 225  $\Omega$  differential resistor in parallel to the modulator inputs (see Figure 5). In this way, the differential filter can be properly terminated by 100  $\Omega$  at both ends.

If the DAC outputs can withstand a higher swing without performance degradation, then 90  $\Omega$  load resistors can be placed at the DAC outputs. No external resistors are needed in this case, only the differential filter needs to be designed to have 180  $\Omega$  at both ends (see Figure 6).

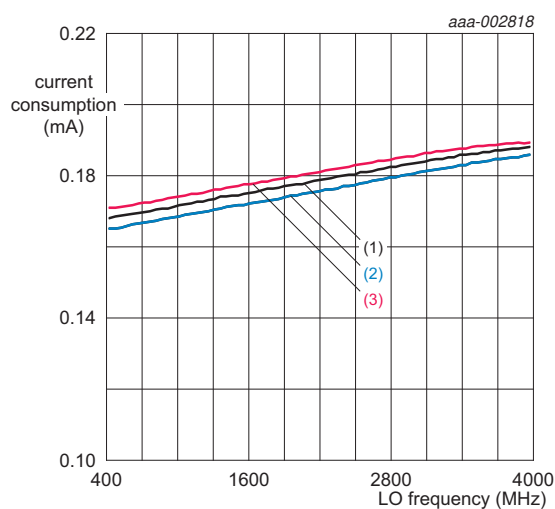


## 12.2 RF

Good RF port matching typically requires some reactive components to tune-out residual inductance or capacitance. As the LO inputs and RF output are internally DC biased, both pins need a series AC-coupling capacitor.

## 13. Test information

Parameters for the following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; IQ frequency = 5 MHz; IQ amplitude = 0.5 V (p-p) differential sine wave;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.

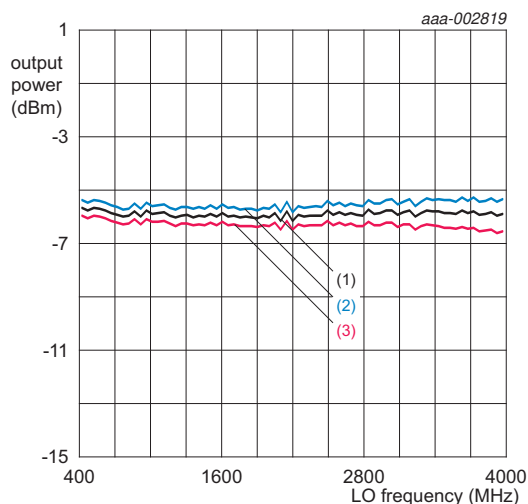


- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

**Fig 7. Current consumption versus  $f_{lo}$  and  $T_{mb}$**

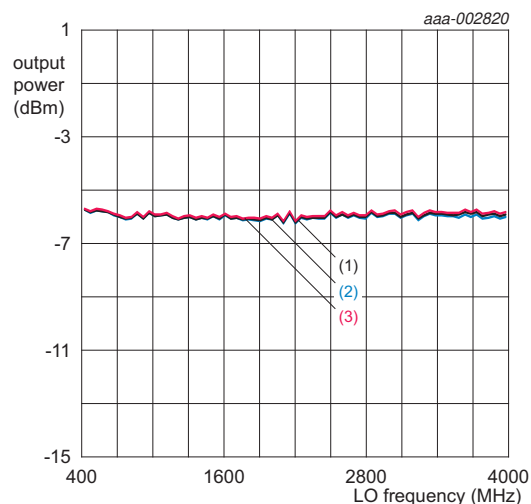


Parameters for the five following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; IQ frequency = 5 MHz; IQ amplitude = 0.5 V (p-p) differential sine wave;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



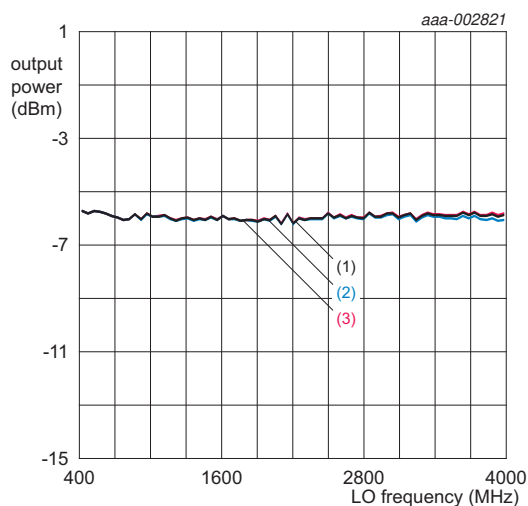
- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

Fig 8.  $P_o$  versus  $f_{lo}$  and  $T_{mb}$



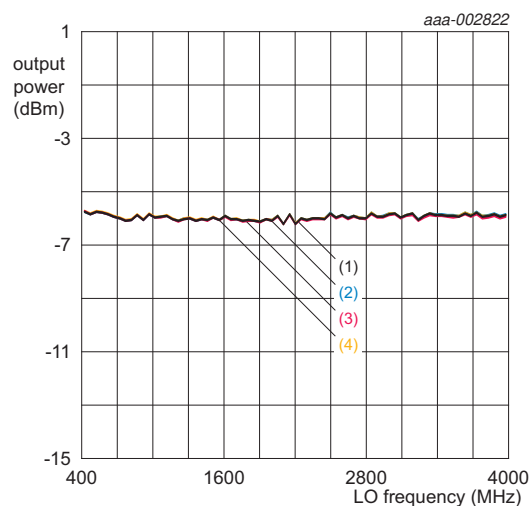
- (1)  $V_{CC} = 5\text{ V}$ .
- (2)  $V_{CC} = 4.75\text{ V}$ .
- (3)  $V_{CC} = 5.25\text{ V}$ .

Fig 9.  $P_o$  versus  $f_{lo}$  and  $V_{CC}$



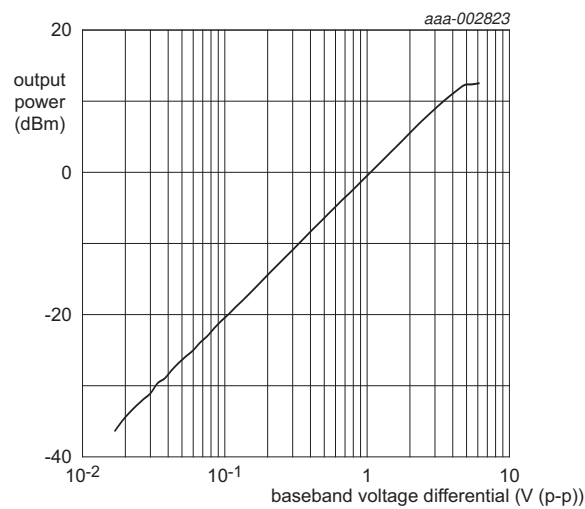
- (1)  $P_{i(lo)} = 0\text{ dBm}$ .
- (2)  $P_{i(lo)} = -3\text{ dBm}$ .
- (3)  $P_{i(lo)} = +3\text{ dBm}$ .

Fig 10.  $P_o$  versus  $f_{lo}$  and  $P_{i(lo)}$



- (1)  $V_{i(cm)} = 0.5\text{ V}$ .
- (2)  $V_{i(cm)} = 0.25\text{ V}$ .
- (3)  $V_{i(cm)} = 1.5\text{ V}$ .
- (4)  $V_{i(cm)} = 2.5\text{ V}$ .

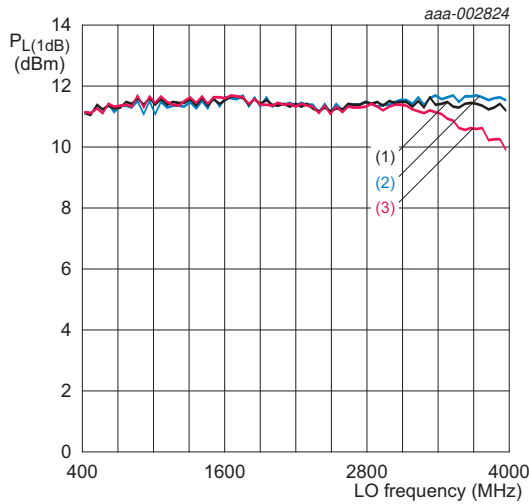
Fig 11.  $P_o$  versus  $f_{lo}$  and  $V_{i(cm)}$



(1)  $f_{lo} = 2140$  MHz.

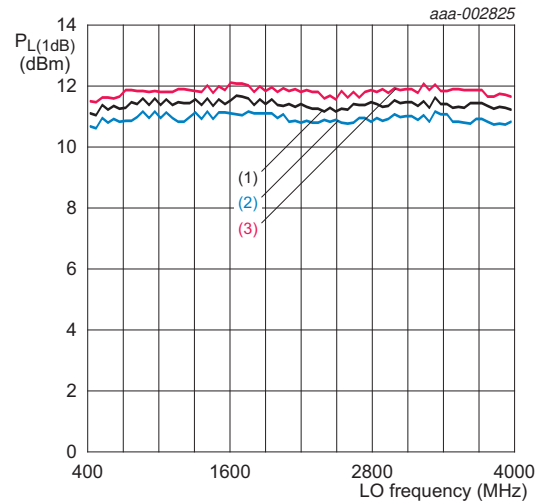
Fig 12.  $P_o$  versus baseband voltage at 2140 MHz

Parameters for the four following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; IQ frequency = 5 MHz; IQ amplitude = 0.5 V (p-p) differential sine wave;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



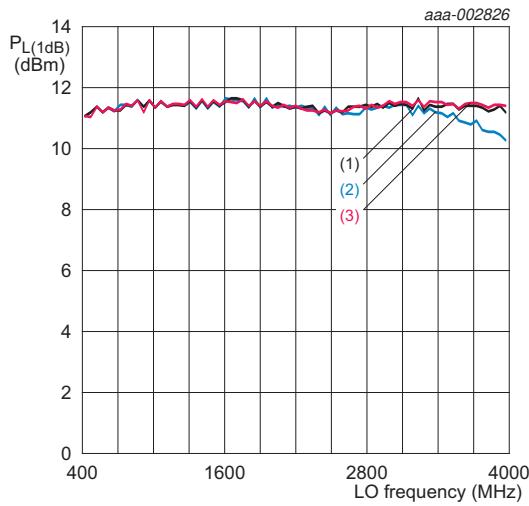
- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

Fig 13.  $P_{L(1dB)}$  versus  $f_{LO}$  and  $T_{mb}$



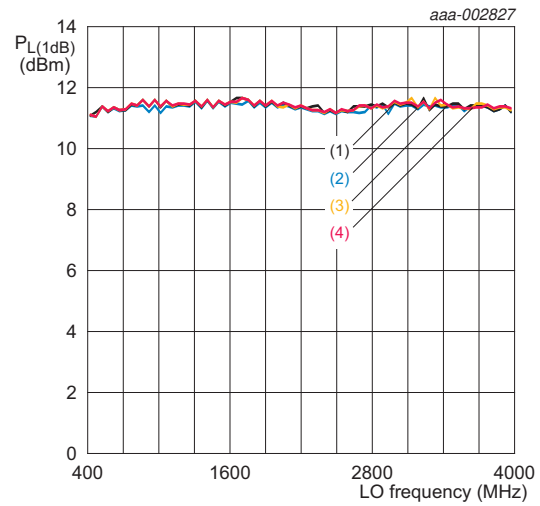
- (1)  $V_{CC} = 5\text{ V}$ .
- (2)  $V_{CC} = 4.75\text{ V}$ .
- (3)  $V_{CC} = 5.25\text{ V}$ .

Fig 14.  $P_{L(1dB)}$  versus  $f_{LO}$  and  $V_{CC}$



- (1)  $P_{i(lo)} = 0\text{ dBm}$ .
- (2)  $P_{i(lo)} = -3\text{ dBm}$ .
- (3)  $P_{i(lo)} = +3\text{ dBm}$ .

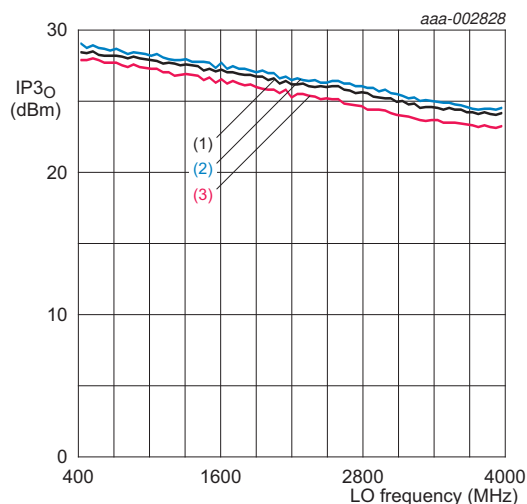
Fig 15.  $P_{L(1dB)}$  versus  $f_{LO}$  and  $P_{i(lo)}$



- (1)  $V_{i(cm)} = 0.5\text{ V}$ .
- (2)  $V_{i(cm)} = 0.25\text{ V}$ .
- (3)  $V_{i(cm)} = 1.5\text{ V}$ .
- (4)  $V_{i(cm)} = 2.5\text{ V}$ .

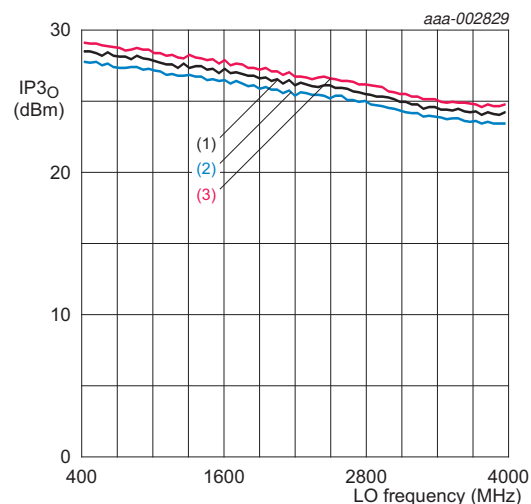
Fig 16.  $P_{L(1dB)}$  versus  $f_{LO}$  and  $V_{i(cm)}$

Parameters for the four following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; two tones; tone 1: IQ frequency = 4.5 MHz and tone 2: IQ frequency = 5.5 MHz;  $P_o$  per tone =  $-10\text{ dBm}$ ;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



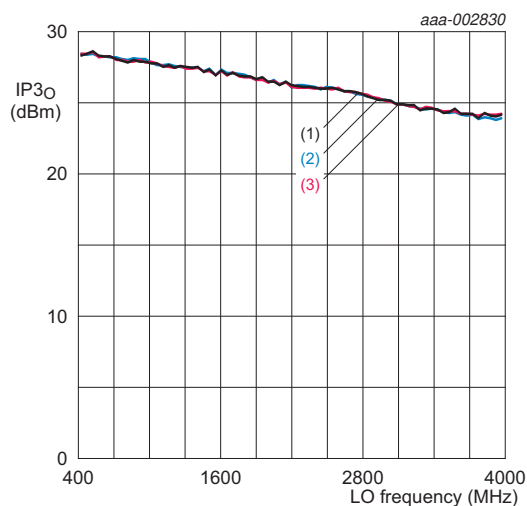
- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

Fig 17.  $IP3_o$  versus  $f_{lo}$  and  $T_{mb}$



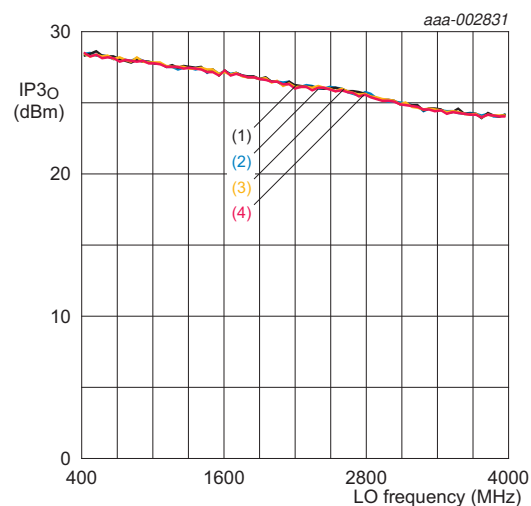
- (1)  $V_{CC} = 5\text{ V}$ .
- (2)  $V_{CC} = 4.75\text{ V}$ .
- (3)  $V_{CC} = 5.25\text{ V}$ .

Fig 18.  $IP3_o$  versus  $f_{lo}$  and  $V_{CC}$



- (1)  $P_{i(lo)} = 0\text{ dBm}$ .
- (2)  $P_{i(lo)} = -3\text{ dBm}$ .
- (3)  $P_{i(lo)} = +3\text{ dBm}$ .

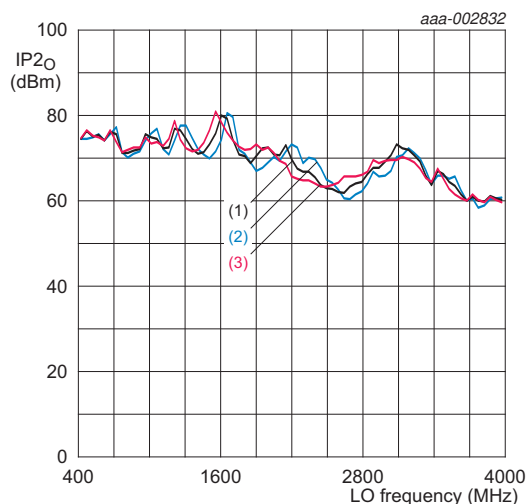
Fig 19.  $IP3_o$  versus  $f_{lo}$  and  $P_{i(lo)}$



- (1)  $V_{i(cm)} = 0.5\text{ V}$ .
- (2)  $V_{i(cm)} = 0.25\text{ V}$ .
- (3)  $V_{i(cm)} = 1.5\text{ V}$ .
- (4)  $V_{i(cm)} = 2.5\text{ V}$ .

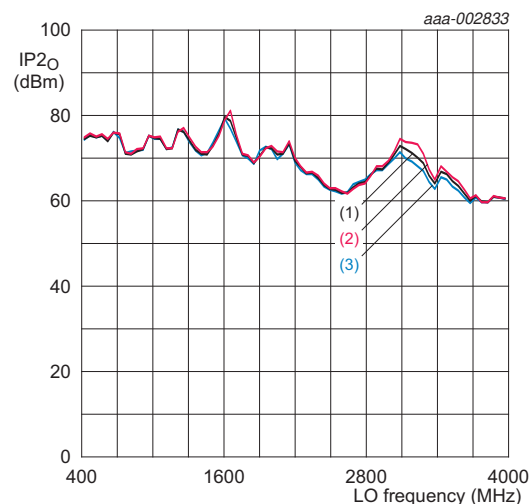
Fig 20.  $IP3_o$  versus  $f_{lo}$  and  $V_{i(cm)}$

Parameters for the four following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; two tones; tone 1: IQ frequency = 4.5 MHz and tone 2: IQ frequency = 5.5 MHz;  $P_o$  per tone =  $-10\text{ dBm}$ ;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



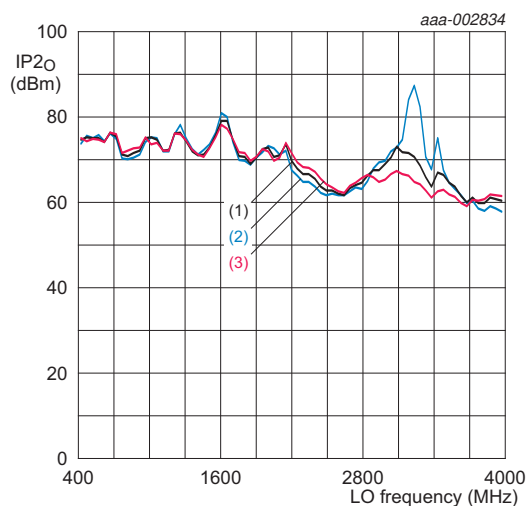
- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

Fig 21.  $IP2_o$  versus  $f_{lo}$  and  $T_{mb}$



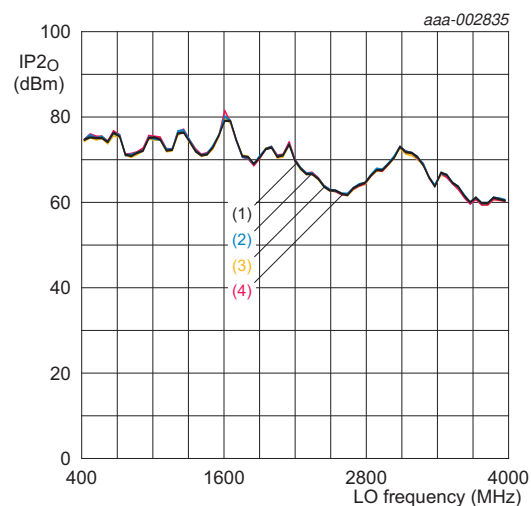
- (1)  $V_{CC} = 5\text{ V}$ .
- (2)  $V_{CC} = 4.75\text{ V}$ .
- (3)  $V_{CC} = 5.25\text{ V}$ .

Fig 22.  $IP2_o$  versus  $f_{lo}$  and  $V_{CC}$



- (1)  $P_{i(lo)} = 0\text{ dBm}$ .
- (2)  $P_{i(lo)} = -3\text{ dBm}$ .
- (3)  $P_{i(lo)} = +3\text{ dBm}$ .

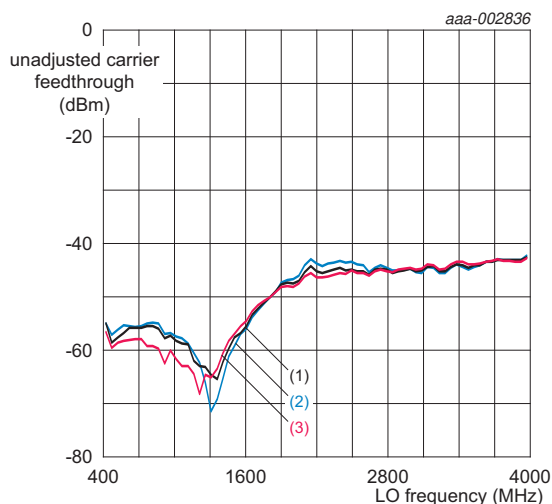
Fig 23.  $IP2_o$  versus  $f_{lo}$  and  $P_{i(lo)}$



- (1)  $V_{i(cm)} = 0.5\text{ V}$ .
- (2)  $V_{i(cm)} = 0.25\text{ V}$ .
- (3)  $V_{i(cm)} = 1.5\text{ V}$ .
- (4)  $V_{i(cm)} = 2.5\text{ V}$ .

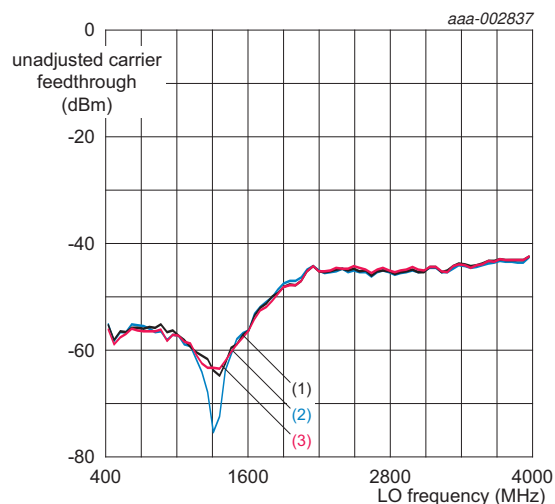
Fig 24.  $IP2_o$  versus  $f_{lo}$  and  $V_{i(cm)}$

Parameters for the five following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; IQ frequency = 5 MHz; IQ amplitude = 0.5 V (p-p) differential sine wave;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



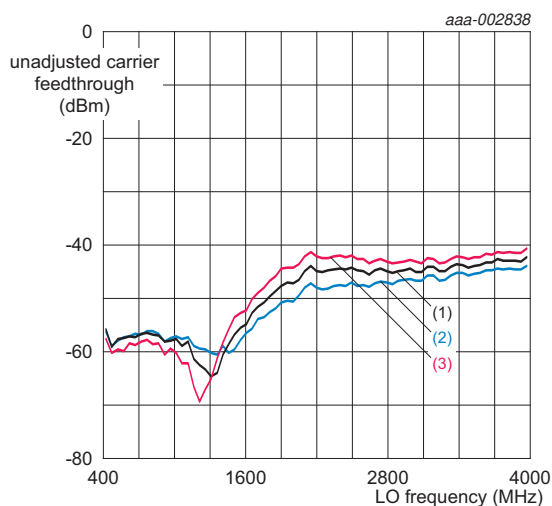
- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

Fig 25. Unadjusted CF versus  $f_{LO}$  and  $T_{mb}$



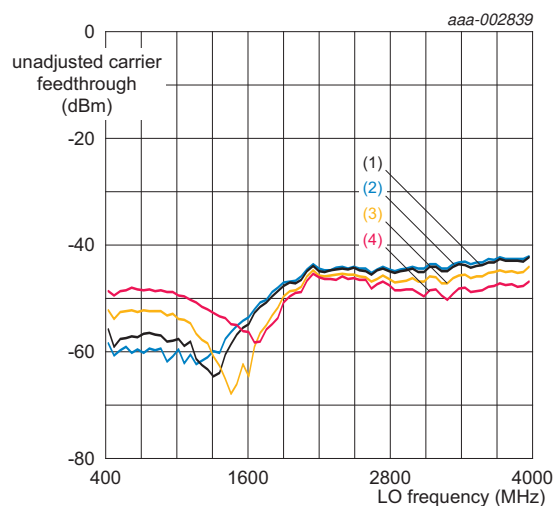
- (1)  $V_{CC} = 5\text{ V}$ .
- (2)  $V_{CC} = 4.75\text{ V}$ .
- (3)  $V_{CC} = 5.25\text{ V}$ .

Fig 26. Unadjusted CF versus  $f_{LO}$  and  $V_{CC}$



- (1)  $P_{i(lo)} = 0\text{ dBm}$ .
- (2)  $P_{i(lo)} = -3\text{ dBm}$ .
- (3)  $P_{i(lo)} = +3\text{ dBm}$ .

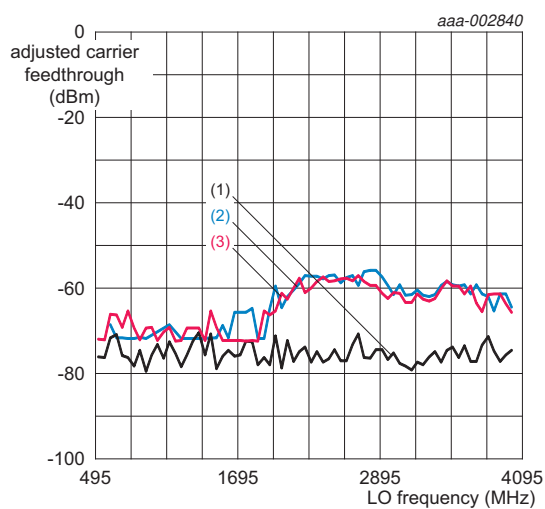
Fig 27. Unadjusted CF versus  $f_{LO}$  and  $P_{i(lo)}$



- (1)  $V_{i(cm)} = 0.5\text{ V}$ .
- (2)  $V_{i(cm)} = 0.25\text{ V}$ .
- (3)  $V_{i(cm)} = 1.5\text{ V}$ .
- (4)  $V_{i(cm)} = 2.5\text{ V}$ .

Fig 28. Unadjusted CF versus  $f_{LO}$  and  $V_{i(cm)}$





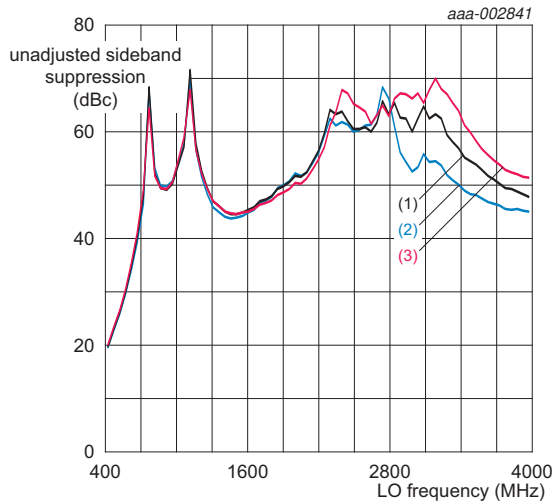
(1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .

(2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .

(3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

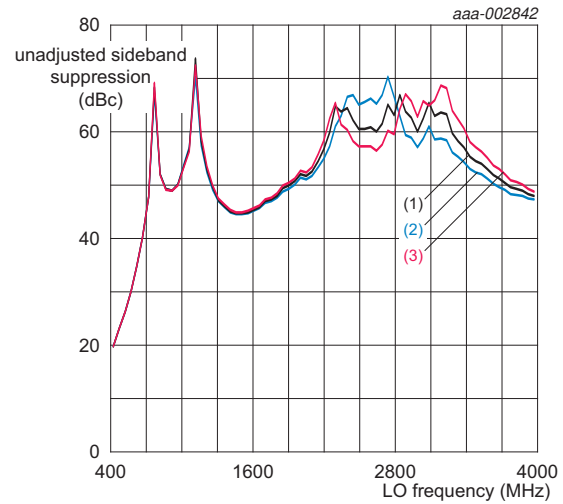
**Fig 29. Adjusted CF versus  $f_{lo}$  and  $T_{mb}$  after nulling at  $25\text{ }^{\circ}\text{C}$**

Parameters for the five following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $P_{i(lo)} = 0\text{ dBm}$ ; IQ frequency = 5 MHz; IQ amplitude = 0.5 V (p-p) differential sine wave;  $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



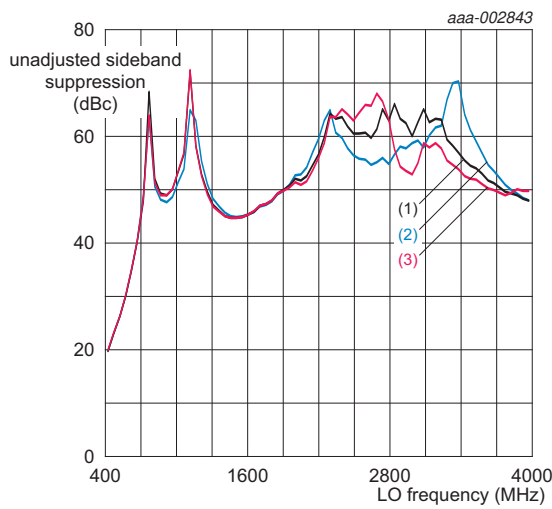
- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

Fig 30. Unadjusted SBS versus  $f_{lo}$  and  $T_{mb}$



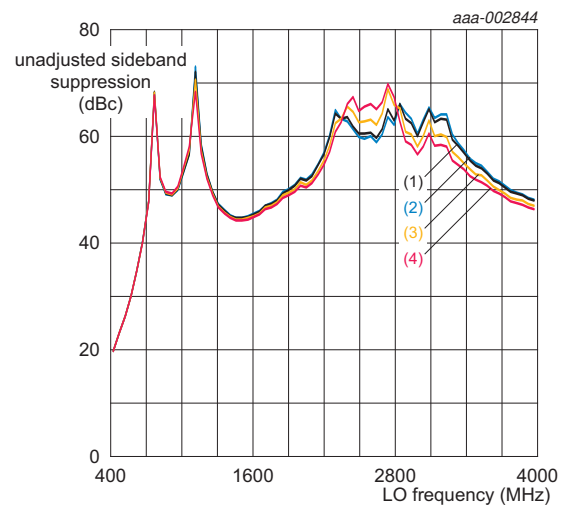
- (1)  $V_{CC} = 5\text{ V}$ .
- (2)  $V_{CC} = 4.75\text{ V}$ .
- (3)  $V_{CC} = 5.25\text{ V}$ .

Fig 31. Unadjusted SBS versus  $f_{lo}$  and  $V_{CC}$



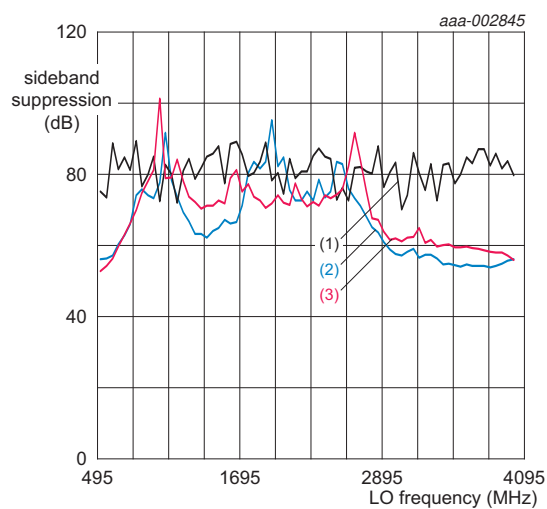
- (1)  $P_{i(lo)} = 0\text{ dBm}$ .
- (2)  $P_{i(lo)} = -3\text{ dBm}$ .
- (3)  $P_{i(lo)} = +3\text{ dBm}$ .

Fig 32. Unadjusted SBS versus  $f_{lo}$  and  $P_{i(lo)}$



- (1)  $V_{i(cm)} = 0.5\text{ V}$ .
- (2)  $V_{i(cm)} = 0.25\text{ V}$ .
- (3)  $V_{i(cm)} = 1.5\text{ V}$ .
- (4)  $V_{i(cm)} = 2.5\text{ V}$ .

Fig 33. Unadjusted SBS versus  $f_{lo}$  and  $V_{i(cm)}$



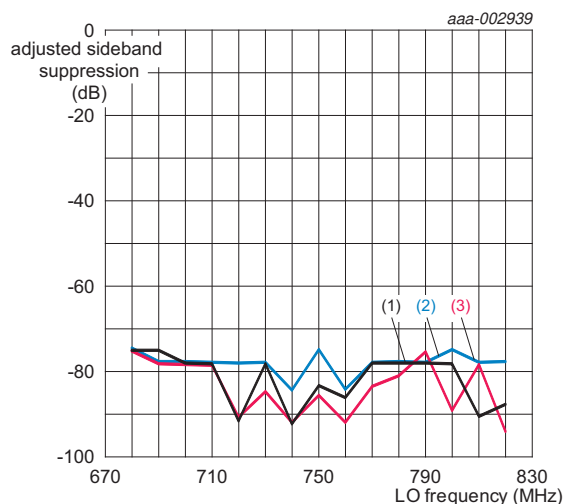
(1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .

(2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .

(3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

**Fig 34. Adjusted SBS versus  $f_{lo}$  and  $T_{mb}$  after nulling at  $25\text{ }^{\circ}\text{C}$**

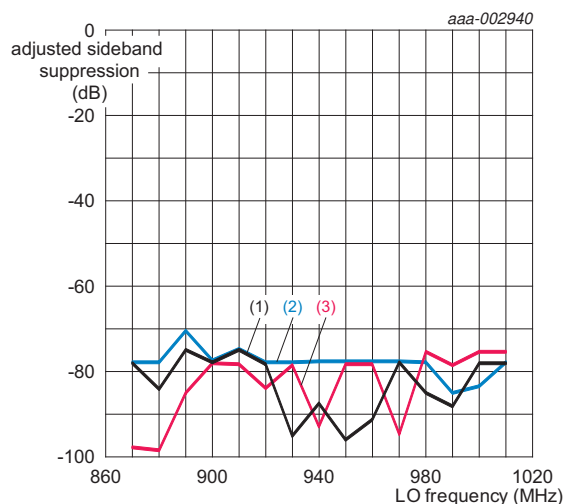
Parameters for the six following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ; LO = 0 dBm;  
IQ frequency = 5 MHz; IQ amplitude = 0.25 V (p-p) single-ended sine wave;  
 $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



Adjusted at 750 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

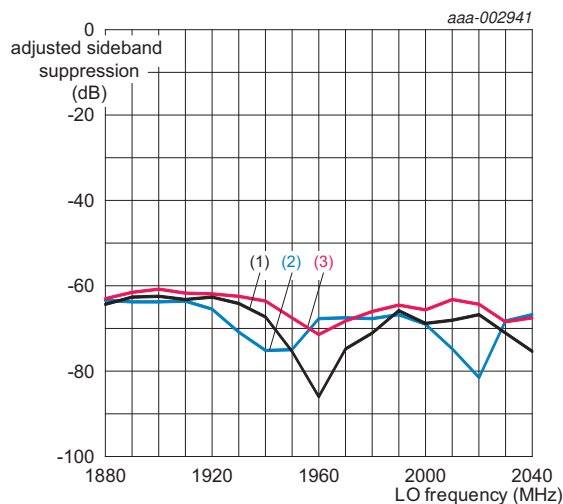
Fig 35. Adjusted CF versus  $f_{LO}$  and  $T_{mb}$  (750 LTE band)



Adjusted at 942.5 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

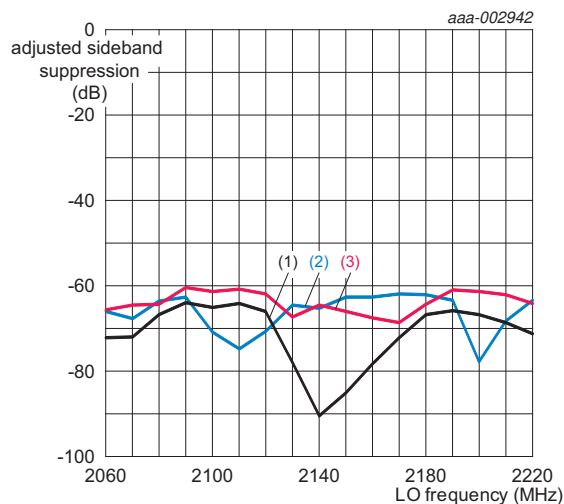
Fig 36. Adjusted CF versus  $f_{LO}$  and  $T_{mb}$  (GSM band)



Adjusted at 1840 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

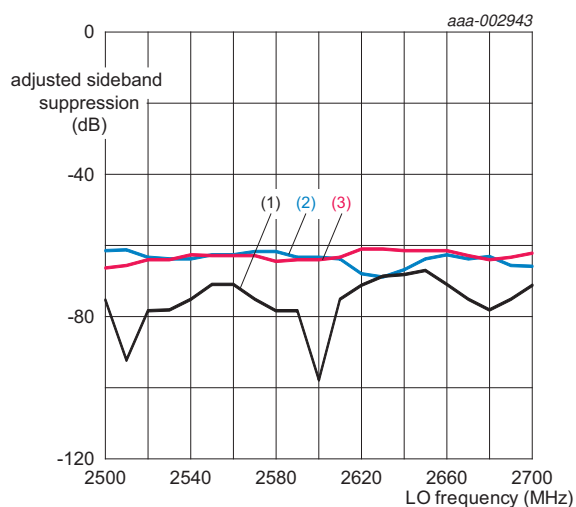
Fig 37. Adjusted CF versus  $f_{LO}$  and  $T_{mb}$  (PCS band)



Adjusted at 2140 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

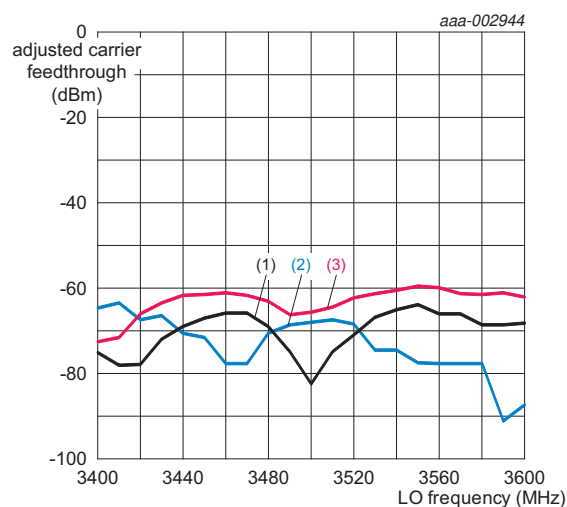
Fig 38. Adjusted CF versus  $f_{LO}$  and  $T_{mb}$  (UMTS band)



Adjusted at 2600 MHz and after nulling  $T_{mb}$  at 25  $^{\circ}\text{C}$

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

**Fig 39. Adjusted CF versus  $f_{lo}$  and  $T_{mb}$  (2.6 GHz LTE band)**

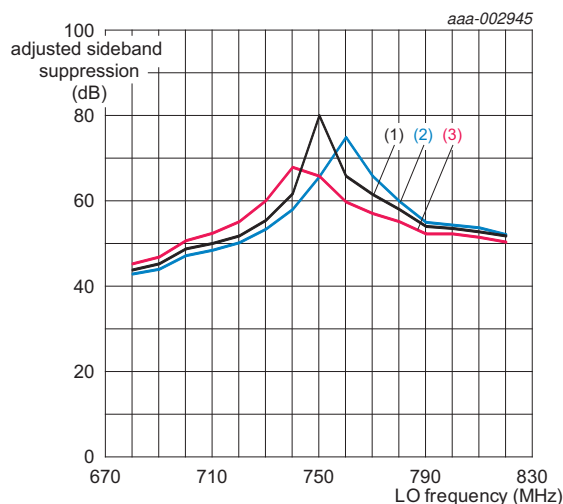


Adjusted at 3500 MHz and after nulling  $T_{mb}$  at 25  $^{\circ}\text{C}$

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

**Fig 40. Adjusted CF versus  $f_{lo}$  and  $T_{mb}$  (Wi MAX/LTE band)**

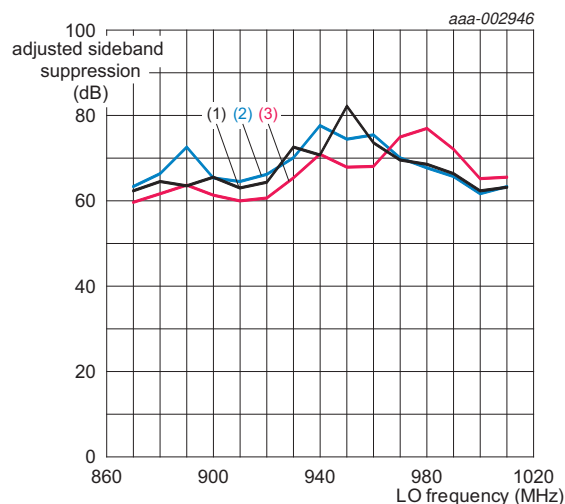
Parameters for the six following drawings:  $V_{CC} = 5\text{ V}$ ;  $T_{mb} = 25\text{ }^{\circ}\text{C}$ ;  $LO = 0\text{ dBm}$ ;  
 IQ frequency = 5 MHz; IQ amplitude = 0.25 V (p-p) single-ended sine wave;  
 $V_{i(cm)} = 0.5\text{ V}$ ; broadband output match; unless otherwise specified.



Adjusted at 750 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

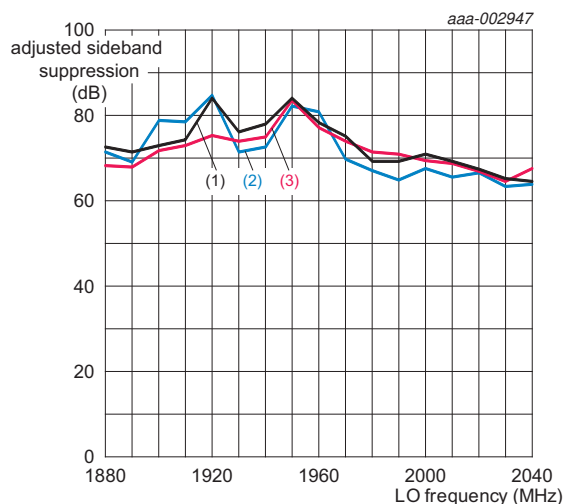
**Fig 41. Adjusted SBS versus  $f_{LO}$  and  $T_{mb}$  (750 LTE band)**



Adjusted at 942.5 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

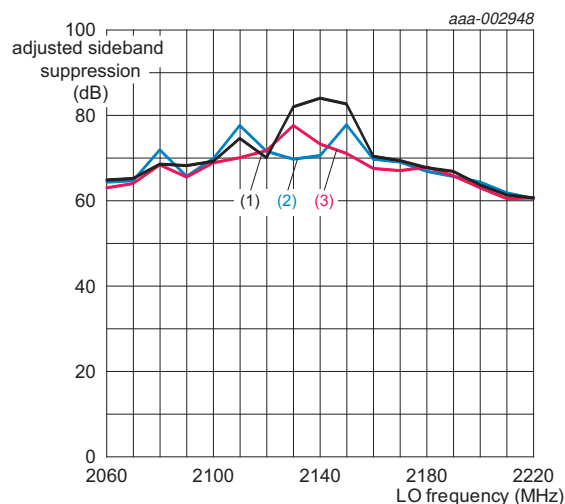
**Fig 42. Adjusted SBS versus  $f_{LO}$  and  $T_{mb}$  (GSM900 band)**



Adjusted at 1840 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

**Fig 43. Adjusted SBS versus  $f_{LO}$  and  $T_{mb}$  (PCS band)**



Adjusted at 2140 MHz and after nulling  $T_{mb}$  at 25 °C

- (1)  $T_{mb} = +25\text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40\text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85\text{ }^{\circ}\text{C}$ .

**Fig 44. Adjusted SBS versus  $f_{LO}$  and  $T_{mb}$  (UMTS band)**