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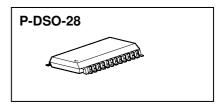
Smart High-Side Power Switch Four Channels: $4 \times 35 \text{m}\Omega$

Advanced Current Sense

Product Summary

Operating Voltage	$V_{bb(on)}$	5.040V		
	Active channels	one	four parallel	
On-state Resistance	R _{on}	$35 m\Omega$	$9m\Omega$	
Nominal load current	I _{L(NOM)}	5.4A	11.1A	
Current limitation	$I_{L(SCr)}$	40A	40A	

Package



General Description

- N channel vertical power MOSFET with charge pump, ground referenced CMOS compatible input and diagnostic feedback, monolithically integrated in Smart SIPMOS® technology.
- Providing embedded protective functions

Applications

- μC compatible high-side power switch with diagnostic feedback for 12V and 24V grounded loads
- · All types of resistive and capacitve loads
- Most suitable for loads with high inrush currents, so as lamps
- Replaces electromechanical relays, fuses and discrete circuits

Basic Functions

- Very low standby current
- Improved electromagnetic compatibility (EMC)
- CMOS compatible input
- Stable behaviour at undervoltage
- Wide operating voltage range

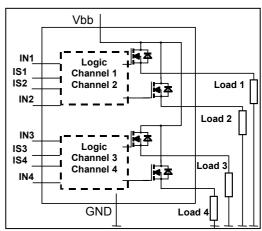
Protection Functions

- Short circuit protection
- Overload protection
- Current limitation
- Thermal shutdown
- Reverse battery protection with external resistor
- Overvoltage protection with external resistor (incl. load dump)
- · Loss of ground protection
- Electrostatic discharge protection (ESD)

Diagnostic Function

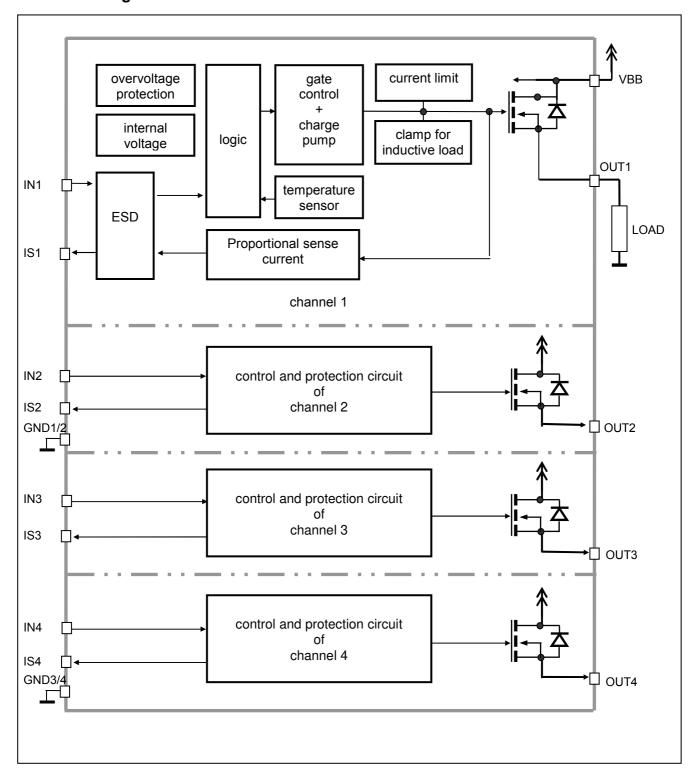
 Proportional load current sense (with defined fault signal during thermal shutdown and current limit)

Block Diagram





Functional diagram





Pin Definitions and Functions

Pin	Symbol	Function
1, 7, 8, 14, 15, 28	V _{bb}	Positive power supply voltage. Design the wiring for the simultaneous max. short circuit currents from channel 1 to 4 and also for low thermal resistance
4	IN1	Input 1,2, 3,4 activates channel 1,2,3,4 in case
3	IN2	of logic high signal
11	IN3	
10	IN4	
25,26,27	OUT1	Output 1,2,3,4 protected high-side power output
22,23,24	OUT2	of channel 1,23,4. Design the wiring for the
19,20,21	OUT3	max. short circuit current
16,17,18	OUT4	
5	IS1	Diagnostic feedback 1 4 of channel 1 to 4
6	IS2	Providing a sense current, proportional to the
12	IS3	load current
13	IS4	
2	GND1/2	Ground of chip 1 (channel 1,2)
9	GND3/4	Ground of chip 2 (channel 3,4)

Pin configuration

(top view)		
V_{bb}	1 •	28	V_{bb}
GND1/2	2	27	OUT1
IN2	3	26	OUT1
IN1	4	25	OUT1
IS1	5	24	OUT2
IS2	6	23	OUT2
V_{bb}	7	22	OUT2
V_{bb}	8	21	OUT3
GND3/4	9	20	OUT3
IN4	10	19	OUT3
IN3	11	18	OUT4
IS3	12	17	OUT4
IS4	13	16	OUT4
V_{bb}	14	15	V_{bb}



Maximum Ratings at $T_i = 25$ °C unless otherwise specified

Parameter	Symbol	Values	Unit
Supply voltage (overvoltage protection see page 6)	$V_{ m bb}$	40	V
Supply voltage for full short circuit protection ¹⁾ $T_{j,\text{start}} = -40 \dots + 150^{\circ}\text{C}$	$V_{ m bb}$	36	V
Load current (Short-circuit current, see page 6)	<i>I</i> _	/ _{L(lim)} 2	
Load dump protection ³⁾ $V_{\text{LoadDump}} = V_{\text{A}} + V_{\text{S}}$, $V_{\text{A}} = 13.5 \text{ V}$ $R_{\text{I}}^{(4)} = 2 \Omega$, $t_{\text{d}} = 400 \text{ ms}$; IN = low or high, each channel loaded with $R_{\text{L}} = 4.7 \Omega$,	V _{Load dump} 5)	60	V
Operating temperature range	T _j	-40+150	°C
Storage temperature range	T_{stg}	-55+150	
Power dissipation (DC) ⁶⁾ $T_a = 25^{\circ}\text{C}$:	P _{tot}	3.7	W
(all channels active) $T_a = 85$ °C:		1.9	
Maximal switchable inductance, single pulse $V_{bb} = 12V$, $T_{j,start} = 150^{\circ}C^{6}$,			
$I_{L} = 4.0 \text{ A}, E_{AS} = 0.8 \text{J}, 0 \Omega$ one channel:	Z_{L}	33	mH
$I_{L} = 6.0 \text{ A}, E_{AS} = 1.0 \text{J}, 0 \Omega$ two parallel channels:		37	
$I_{L} = 9.5 \text{ A}, E_{AS} = 1.5 \text{J}, 0 \Omega$ four parallel channels:		64	
see diagrams on page 10			
Electrostatic discharge capability (ESD) IN: (Human Body Model) IS: out to all other pins shorted: acc. MIL-STD883D, method 3015.7 and ESD assn. std. S5.1-1993	V _{ESD}	1.0 4.0 8.0	kV
$R=1.5k\Omega$; $C=100pF$			
Input voltage (DC)	V _{IN}	-10 +16	V
Current through input pin (DC)	I _{IN}	±0.3	mA
Current through sense pin (DC)	$I_{\rm IS}$	±0.3	
see internal circuit diagram page 9			

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¹⁾ Single pulse

Current limit is a protection function. Operation in current limitation is considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

Supply voltages higher than $V_{bb(AZ)}$ require an external current limit for the GND and status pins (a 75 Ω resistor for the GND connection is recommended.

 $^{^{4)}}$ $R_{\rm I}$ = internal resistance of the load dump test pulse generator

⁵⁾ V_{Load dump} is setup without the DUT connected to the generator per ISO 7637-1 and DIN 40839

⁶⁾ Device on 50mm*50mm*1.5mm epoxy PCB FR4 with 6cm² (one layer, 70μm thick) copper area for V_{bb} connection. PCB is vertical without blown air. See page 15



Thermal Characteristics

Parameter and Conditions		Symbol	Values		Unit	
			min	typ	Max	
Thermal resistance junction - soldering point ^{7)8),} junction - ambient ⁸⁾	each channel:	R _{thjs} R _{thja}			11	K/W
@ 6 cm ² cooling area	one channel active:			40		
	all channels active:			33		

Electrical Characteristics

Parameter and Cond	ditions, each of the four channels	Symbol		Values	3	Unit
at T _j = -40+150°C, V_{bb}	= 12 V unless otherwise specified		min	typ	Max	
Load Switching Cap	pabilities and Characteristics					
On-state resistance (V_{bb} to OUT); IL = 5 A					
see diagram, page 11	each channel, $T_j = 25$ °C: $T_j = 150$ °C:	R _{ON}		30 55	35 64	mΩ
Nominal load current	one channel active: two parallel channels active: four parallel channels active:	/ _{L(NOM)}	5.0 6.7 10.5	5.4 7.4 11.1	 	A
Device on PCB8), $T_a = 8$	35°C, <i>T</i> _j ≤ 150°C					
Output current while O	GND disconnected, V _{IN} = 0,	I _{L(GNDhigh)}			1	mA
see diagram page 10; (not subject to production	on test - specified by design)					
Turn-on time ⁹⁾	IN \perp to 90% V_{OUT} :	<i>t</i> on		50	150	μs
Turn-off time	IN \square to 10% V_{OUT} :	$t_{ m off}$		120	250	
$R_{\rm L} = 12\Omega$						
Slew rate on 9)		d V/dt _{on}	0.2		0.9	V/µs
$V_{\rm OUT}$ rising from 10 to	o 30% of $V_{\rm bb}, R_{\rm L} = 12 \Omega$:					
Slew rate off 9) V _{OUT} falling from 70 t	to 40% of $V_{\rm bb}$, $R_{\rm L}=12\Omega$:	-d V/dt _{off}	0.1		0.9	V/µs

⁷⁾ Soldering point: upper side of solder edge of device pin 7,8. See page 16.

Device on 50mm*50mm*1.5mm epoxy PCB FR4 with 6cm² (one layer, 70μm thick) copper area for V_{bb} connection. PCB is vertical without blown air. See page 15

⁹⁾ See timing diagram on page 12.



operating randincters	0	perating	Parameters
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<u> </u>						
Operating voltage		$V_{ m bb(on)}$	5.0		40	V
Overvoltage protection ¹⁰⁾		$V_{\rm bb(AZ)}$	41	47	52	V
$I_{bb} = 40 \text{ mA}$						
Standby current ¹¹⁾	<i>T</i> _j =-4025°C:	I _{bb(off)}		10	25	μΑ
$V_{IN} = 0$; see diagram page 12	$T_{\rm j} = 150 {\rm ^{\circ}C}$:			40	80	
	<i>T</i> _i =125°C:				25	
(not subject to production test - specific	ed by design)					
Off-State output current	$T_{\rm j}$ =-4025°C:	I _{L(off)}		1	6	μΑ
(included in $I_{bb(off)}$) $V_{IN} = 0$; each	n channel; T_j =150°C:	, ,			15	
Operating current, $V_{IN} = 5V$,						
$I_{\text{GND}} = I_{\text{GND}1/2} + I_{\text{GND}3/4},$	one channel on:	<i>I</i> GND		1.6		mA
	four channels on:			6.0		

Protection Functions¹²⁾

Current limit, (see timing diagrams, page 13)					
	I _{L(lim)}	36	45	58	Α
Repetitive short circuit current limit,					
$T_{\rm j} = T_{\rm jt}$ each channel	I _{L(SCr)}		40		Α
two,three or four parallel channels	(,		40		
(see timing diagrams, page 13)					
Initial short circuit shutdown time $T_{j,start} = 25$ °C:	$t_{ m off(SC)}$		4		ms
(see timing diagrams on page 13)					
Output clamp (inductive load switch off) ¹³⁾					
at $V_{ON(CL)} = V_{bb} - V_{OUT}$, $I_{L} = 40 \text{ mA}$	$V_{ON(CL)}$	41	47	52	V
Thermal overload trip temperature	T_{jt}	150			°C
Thermal hysteresis	$\Delta T_{\rm jt}$		10		K

Reverse Battery¹⁴⁾ (not subject to production test - specified by design)

Reverse battery voltage	- $V_{ m bb}$	 	14	V
Drain-source diode voltage (V _{out} > V _{bb})				
$I_{L} = -2A$; Tj = +150°C:	$-V_{ON}$	 500		mV

Supply voltages higher than $V_{bb(AZ)}$ require an external current limit for the GND and status pins (a 75 Ω resistor for the GND connection is recommended). See also $V_{ON(CL)}$ in table of protection functions and circuit diagram on page 9.

Infineon technologies

¹¹⁾ Measured with load; for the whole device; all channels off.

¹²⁾ Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

If channels are connected in parallel, output clamp is usually accomplished by the channel with the lowest $V_{\rm ON(CL)}$.

Requires a 75 Ohm resistor in the GND connection. The reverse load current through the intrinsic drain-source diode has to be limited by the connected load. Note that the power dissipation is higher compared to normal operating conditions due to the voltage drop across the intrinsic drain-source diode. The temperature protection and sense functionality is not active during reverse current operation! Input and Status currents have to be limited (see max. ratings page 4 and circuit page 9).



Input	15)
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Input resistance (see circuit page 9)		R_{I}	2.5	3.5	6.0	kΩ
Input turn-on threshold voltage		$V_{IN(T+)}$	1.7		3.2	V
Input turn-off threshold voltage		$V_{IN(T-)}$	1.5			V
Input threshold hysteresis		$\Delta V_{\text{IN(T)}}$		0.3		V
Off state input current	$V_{IN} = 0.4 \text{ V}$:	I _{IN(off)}	1		35	μA
On state input current	$V_{IN} = 5 \text{ V}$:	I _{IN(on)}	20	50	90	μΑ

Diagnostic Characteristics

Diagnostic Characteristics					
Current sense ratio, static on-condition, $k_{\text{ILIS}} = l_{\text{L}} : l_{\text{IS}}$	<i>k</i> _{ILIS}		5 300		
$I_{L} = 10 \text{ A}$: $I_{L} = 2 \text{ A}$: $I_{L} = 1 \text{ A}$: $I_{L} = 0.5 \text{ A}$:		4575 4100 4200 3580	5300 5300 5300 5800	6000 6300 6600 8080	
Sense signal in case of fault-conditions ¹⁶⁾	V_{fault}	5.4	6.3	7.5	V
Sense signal delay after thermal shutdown ¹⁷⁾	t _{delay(fault)}			1	ms
(not subject to production test - specified by design)					
Sense current saturation	I _{IS,lim}	4			mA
Current sense output voltage limitation $I_{IS} = 0$, $I_{L} = 5$ A:	V _{IS(lim)}	5.4	6.3	7.5	V
Current sense leakage/offset current					
$V_{IN}=0, V_{IS}=0, I_{L}=0$:	I _{IS(LL)}			1	μΑ
$V_{IN}=5 \text{ V}, \ V_{IS}=0, \ I_{L}=0$:	I _{IS(LH)}		2.5		•
Current sense settling time to $I_{\text{IS static}}\pm 10\%$ after positive input slope, $I_{\text{L}}=0$ 5 A, (not subject to production test - specified by design)	$t_{son(IS)}$			300	μs

 $^{^{\}rm 15)}\,$ If ground resistors ${\rm R}_{\rm GND}$ are used, add the voltage drop across these resistors.

¹⁶⁾ In the case of current limitation or thermal shutdown the sense signal is no longer a current proportional to the load current, but a fixed voltage of typ. 6 V.

In the case of thermal shutdown the V_{fault} signal remains for $t_{delay(fault)}$ longer than the restart of the switch (see diagram on page 14).



Truth Table

	Input level	Output level	Current Sense IIS
Normal	L	L	0
Operation	Н	Н	nominal
Current- Limitation ¹⁸⁾	Н	Н	V _{fault}
Short circuit to GND	L H	L	0 V _{fault}
Overtemperature	L H	L L	0 V _{fault}
Short circuit to Vbb	L H	H H	0 <nominal<sup>19)</nominal<sup>
Open load	L H	Z H	0
Negative output Voltage clamp	L	L	0

L = "Low" Level

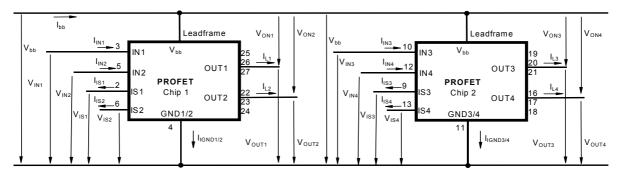
X = don't care

Z = high impedance, potential depends on external circuit

H = "High" Level $V_{fault} = 6V$ typ, constant voltage independent of external used sense resistor.

Parallel switching of channels is possible by connecting the inputs and outputs in parallel. The current sense outputs have to be connected with a single sense resistor.

Terms



Leadframe (V_{bb}) is connected to pin 1, 7, 8, 14, 15, 28.

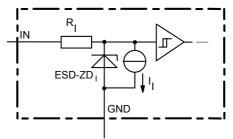
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¹⁸⁾ Current limitation is only possible while the device is switched on.

Low ohmic short to $V_{\rm bb}$ may reduce the output current $I_{\rm L}$ and therefore also the sense current $I_{\rm IS}$.



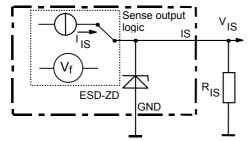
Input circuit (ESD protection), IN1 to IN4



The use of ESD zener diodes as voltage clamp at DC conditions is not recommended.

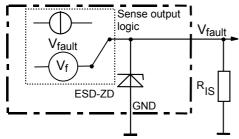
Sense output

Normal operation: $I_S = I_L / k_{ILIS}$ $V_{IS} = I_S * R_{IS}$; $R_{IS} = 1 k\Omega$ nominal $R_{IS} > 500\Omega$



ESD-Zener diode: V_{ESD} = 6.1 V typ., max 14 mA;

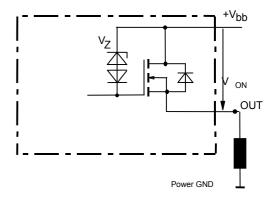
Operation under fault condition so as thermal shut down or current limitation



 $V_{fault} = 6V typ$

V_{fault} < V_{ESD} under all conditions

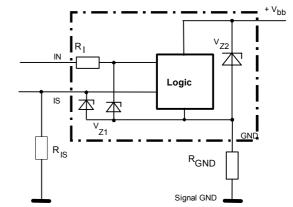
Overvoltage output clamp, OUT1 or OUT2



 V_{ON} clamped to $V_{ON(CL)} = 47 \text{ V typ.}$

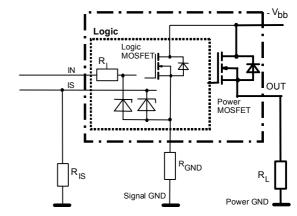
Overvoltage protection of logic part

GND1/2 or GND3/4



 V_{Z1} = 6.1 V typ., V_{Z2} = 47 V typ., R_I = 3.5 k Ω typ., R_{GND} = 75 Ω

Reverse battery protection

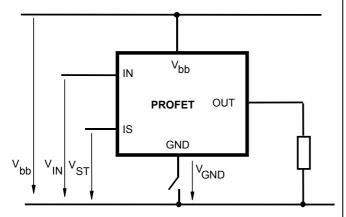


 $R_{\text{GND}} = 75 \,\Omega$, $R_{\text{I}} = 3.5 \,\text{k}\Omega$ typ,

Temperature protection and sense functionality is not active during inverse current operation.

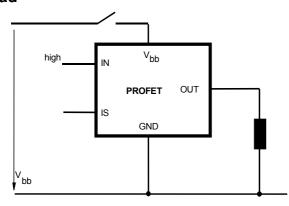


GND disconnect



Any kind of load. In case of IN = high is $V_{OUT} \approx V_{IN} - V_{IN}(T_+)$. Due to $V_{GND} > 0$, no $V_{ST} = low$ signal available.

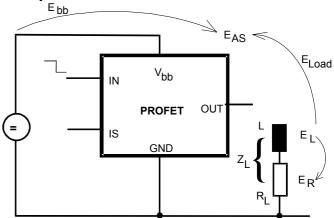
V_{bb} disconnect with energized inductive load



For inductive load currents up to the limits defined by Z_L (max. ratings and diagram on page 10) each switch is protected against loss of V_{bb} .

Consider at your PCB layout that in the case of Vbb disconnection with energized inductive load all the load current flows through the GND connection.

Inductive load switch-off energy dissipation



Energy stored in load inductance:

$$E_{L} = \frac{1}{2} \cdot L \cdot I_{L}^{2}$$

While demagnetizing load inductance, the energy dissipated in PROFET is

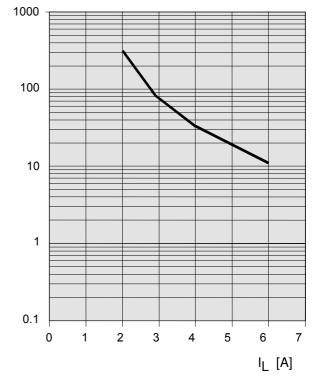
$$E_{AS} = E_{bb} + E_L - E_R = \int V_{ON(CL)} \cdot i_L(t) dt$$

with an approximate solution for $R_L > 0 \Omega$:

$$E_{AS} = \frac{I_L \cdot L}{2 \cdot R_L} \left(V_{bb} + |V_{OUT(CL)}| \right) ln \left(1 + \frac{I_L \cdot R_L}{|V_{OUT(CL)}|} \right)$$

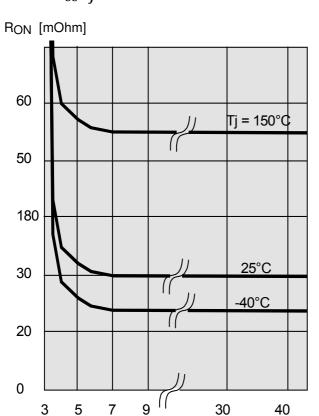
Maximum allowable load inductance for a single switch off (one channel)⁶⁾

$$L = f(I_L)$$
; T_{j,start} = 150°C, V_{bb} = 12 V, R_L = 0 Ω
Z_L [mH]



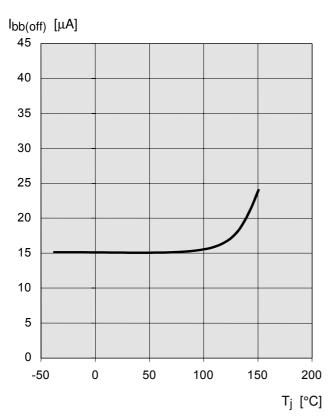


Typ. on-state resistance $R_{ON} = f(V_{bb}, T_j)$; $I_L = 2 \text{ A}$, IN = high



Typ. standby current

 $I_{bb(off)} = f(T_j)$; $V_{bb} = 9...34 \text{ V}$, IN1,2,3,4 = low

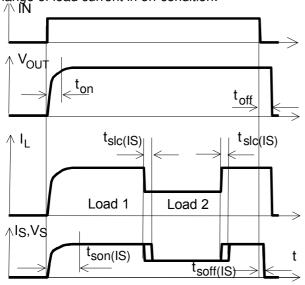


V_{bb} [V]



Functionality diagrams

All diagrams are shown for chip 1 (channel 1/2). For chip 2 (channel 3/4) the diagrams are valid too. The channels 1 and 2, respectively 3 and 4, are symmetric and consequently the diagrams are valid for each channel as well as for permuted channels



The sense signal is not valid during settling time after turn on or change of load current.

Figure 1b: V_{bb} turn on:

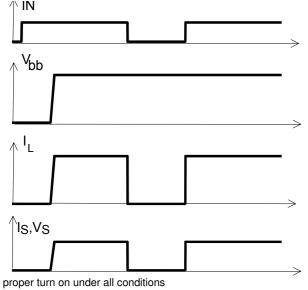
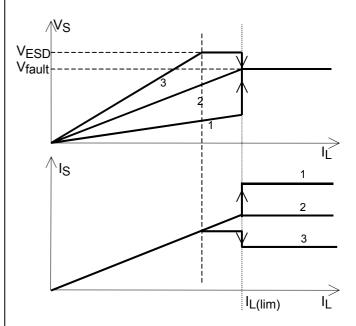


Figure 1c: Behaviour of sense output: Sense current (I_S) and sense voltage (V_S) as function of load current dependent on the sense resistor

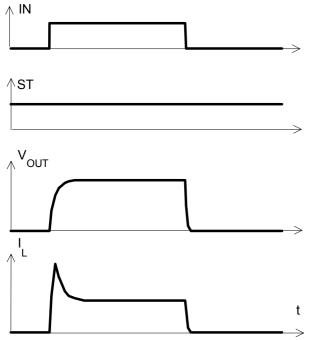
Shown is V_S and I_S for three different sense resistors. Curve 1 refers to a low resistor, curve 2 to a medium-sized resistor and curve 3 to a big resistor. Note, that the sense resistor may not fall short of a minimum value of 500Ω .



$$\begin{split} I_S &= I_L \, / \, k_{ILIS} \\ V_{IS} &= I_S * \, R_{IS}; \; R_{IS} = 1 \, \, k\Omega \; nominal \\ R_{IS} &> 500\Omega \end{split}$$



Figure 2a: Switching a lamp:



The initial peak current should be limited by the lamp and not by the current limit of the device.

Figure 2b: Switching a lamp with current limit: The behaviour of IS and VS is shown for a resistor, which refers to curve 1 in figure 1c

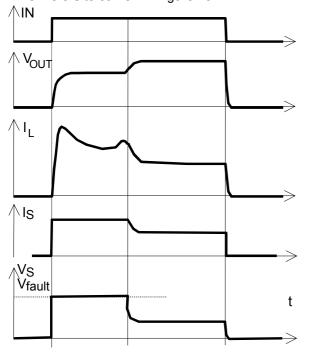
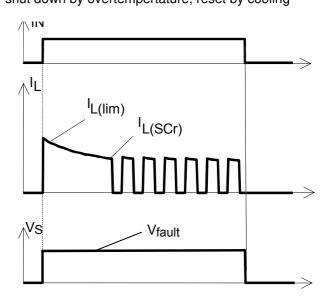


Figure 3a: Short circuit: shut down by overtempertature, reset by cooling



Heating up may require several milliseconds, depending on external conditions

 $I_{L(lim')} = 45 \text{ A typ.}$ increases with decreasing temperature.

Figure 3b: Turn on into short circuit: shut down by overtemperature, restart by cooling (two parallel switched channels 1 and 2)

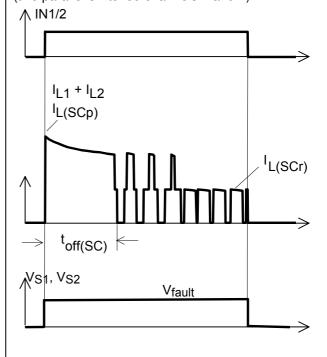




Figure 4a: Overtemperature:

Reset if $T_j < T_{jt}$

The behaviour of IS and VS is shown for a resistor, which refers to curve 1 in figure 1c

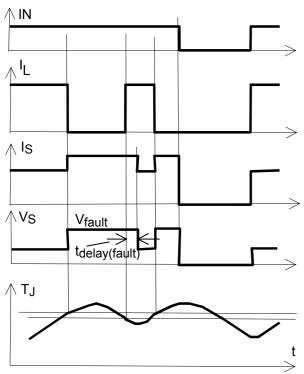


Figure 6a: Current sense versus load current:

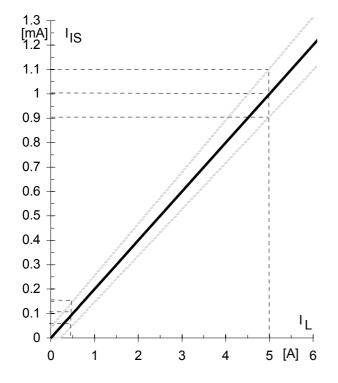
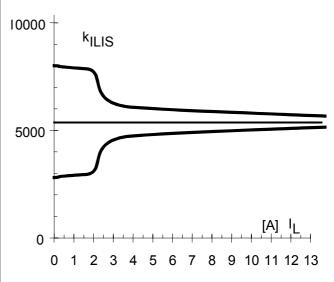


Figure 6b: Current sense ratio²⁰⁾:



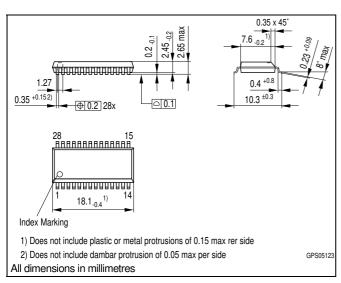
²⁰⁾ This range for the current sense ratio refers to all devices. The accuracy of the $k_{\scriptscriptstyle \rm ILIS}$ can be raised at least by a factor of two by calibrating the value of $k_{\scriptscriptstyle \rm ILIS}$ for every single device.



Package and Ordering Code

Standard: P-DSO-28-16

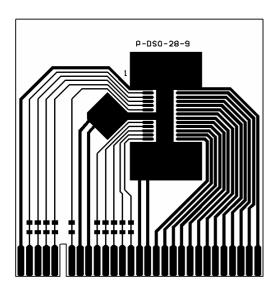
Sales Code	BTS 737 S3
Ordering Code	Q67060-S6133



Definition of soldering point with temperature T_s : upper side of solder edge of device pin 15.



Printed circuit board (FR4, 1.5mm thick, one layer 70 μ m, 6cm² active heatsink area) as a reference for max. power dissipation P_{tot}, nominal load current I_{L(NOM)} and thermal resistance R_{thja}



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