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## CAT5171

## 256-position I²C Compatible Digital Potentiometer (POT)

The CA5171 is a 256-position digital linear taper potentiometer ideally suited for replacing mechanical potentiometers and variable resistors.

The wiper settings are controlled through an $\mathrm{I}^{2} \mathrm{C}$-compatible digital interface. Upon power-up, the wiper assumes a midscale position and may be repositioned anytime after the power is stable. The device can be programmed to reset the wiper position to midscale or to go to a shutdown state during operation. An address input pin, AD0, allows the connection of two devices onto the same $\mathrm{I}^{2} \mathrm{C}$ bus.

The CAT5171 operates from 2.7 V to 5.5 V , while consuming less than $2 \mu \mathrm{~A}$. This low operating current, combined with a small package footprint, makes the CAT5171 ideal for battery-powered portable applications.

The CAT5171, designed as a pin for pin replacement for the AD5245, is offered in the 8-lead SOT23 package and operates over the $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ industrial temperature range.

## Features

- 256-position
- End-to-End Resistance: $50 \mathrm{k} \Omega, 100 \mathrm{k} \Omega$
- $I^{2} \mathrm{C}$ Compatible Interface
- Power-on Preset to Midscale
- Single Supply 2.7 V to 5.5 V
- Low Temperature Coefficient $100 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$
- Low Power, $\mathrm{I}_{\mathrm{DD}} 2 \mu \mathrm{~A}$ max
- Wide Operating Temperature $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
- SOT-23 8-lead ( $2.9 \mathrm{~mm} \times 3 \mathrm{~mm}$ ) Package
- These Devices are Pb-Free, Halogen Free/BFR Free and are RoHS Compliant


## Typical Applications

- Potentiometer Replacement
- Transducer Adjustment of Pressure, Temperature, Position, Chemical, and Optical Sensors
- RF Amplifier Biasing
- Gain Control and Offset Adjustment

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SOT23-8 TP, TB SUFFIX CASE 527AK


PIN CONNECTIONS


ORDERING INFORMATION
See detailed ordering and shipping information in the package dimensions section on page 2 of this data sheet.

## CAT5171



Figure 1. Functional Block Diagram

Table 1. ORDERING INFORMATION

| Part Number | Resistance | Temperature Range | Package | Shipping ${ }^{\dagger}$ |
| :---: | :---: | :---: | :---: | :---: |
| CAT5171TBI-50GT3 | $50 \mathrm{k} \Omega$ |  | SOT-23-8 | $3000 /$ Tape \& Reel |
|  | CAT5171TBI-00GT3 | $100 \mathrm{k} \Omega$ | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ | (Pb-Free) |

$\dagger$ For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

1. For detailed information and a breakdown of device nomenclature and numbering systems, please see the ON Semiconductor Device Nomenclature document, TND310/D, available at www.onsemi.com.

Table 2. PIN FUNCTION DESCRIPTION

| Pin No. | Pin Name |  |
| :---: | :---: | :--- |
| 1 | W | Resistor's Wiper Terminal |
| 2 | V $_{\text {DD }}$ | Positive Power Supply |
| 3 | GND | Digital Ground |
| 4 | SCL | Serial Clock Input |
| 5 | SDA | Serial Data Input |
| 6 | AD0 | I $^{2}$ C Address bit 0 input |
| 7 | B | Bottom Terminal of resistive element |
| 8 | A | Top Terminal of resistive element |

Table 3. ABSOLUTE MAXIMUM RATINGS (Note 2)

| Rating | Value | Unit |
| :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}$ to GND | -0.3 to 6.5 | V |
| $\mathrm{~V}_{\mathrm{A}}, \mathrm{V}_{\mathrm{B}}, \mathrm{V}_{\mathrm{W}}$ to GND | $\mathrm{V}_{\mathrm{DD}}$ |  |
| $\mathrm{I}_{\text {MAX }}$ | $\pm 20$ | mA |
| Digital Inputs and Output Voltage to GND | 0 to 6.5 | V |
| Operating Temperature Range | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Maximum Junction Temperature (TJMAX) | 150 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |
| Lead Temperature (Soldering, 10 sec) | 300 | ${ }^{\circ} \mathrm{C}$ |

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.
2. Maximum terminal current is bounded by the maximum current handling of the switches, maximum power dissipation of the package, and maximum applied voltage across any two of the $\mathrm{A}, \mathrm{B}$, and W terminals at a given resistance.

Table 4. ELECTRICAL CHARACTERISTICS: $\mathbf{5 0} \mathbf{k \Omega}$ and $100 \mathbf{k \Omega}$ Versions
$\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to $5.5 \mathrm{~V} ; \mathrm{V}_{\mathrm{A}}=\mathrm{V}_{\mathrm{DD}} ; \mathrm{V}_{\mathrm{B}}=0 \mathrm{~V} ;-40^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{A}}<+85^{\circ} \mathrm{C}$; unless otherwise noted.

| Parameter | Test Conditions | Symbol | Min | Typ <br> $($ Note 3) | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |

DC CHARACTERISTICS - RHEOSTAT MODE

| Resistor Differential Nonlinearity (Note 4) | $\mathrm{R}_{\mathrm{WB}}, \mathrm{V}_{\mathrm{A}}=$ no connection | R-DNL | -1 | $\pm 0.1$ | +1 | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resistor Integral Nonlinearity (Note 4) | $\mathrm{R}_{\mathrm{WB}}, \mathrm{V}_{\mathrm{A}}=$ no connection | R-INL | -2 | $\pm 0.4$ | +2 | LSB |
| Nominal Resistor Tolerance (Note 5) | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | $\Delta \mathrm{R}_{\text {AB }}$ | -20 |  | +20 | \% |
| Resistance Temperature Coefficient | $\mathrm{V}_{\mathrm{AB}}=\mathrm{V}_{\mathrm{DD}}$, Wiper $=$ no connection | $\Delta \mathrm{R}_{\text {AB }} / \Delta \mathrm{T}$ |  | 100 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
| Wiper Resistance | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{W}}= \pm 3 \mathrm{~mA}$ | $\mathrm{R}_{\mathrm{W}}$ |  | 50 | 120 | $\Omega$ |
|  | $\mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V}, \mathrm{I}_{\mathrm{W}}= \pm 3 \mathrm{~mA}$ |  |  | 100 | 250 |  |

DC CHARACTERISTICS - POTENTIOMETER DIVIDER MODE

| Resolution |  | N |  |  | 8 | Bits |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: |
| Differential Nonlinearity (Note 6) |  | DNL | -1 | $\pm 0.1$ | +1 | LSB |
| Integral Nonlinearity (Note 6) |  | INL | -1 | $\pm 0.4$ | +1 | LSB |
| Voltage Divider Temperature Coefficient | Code $=0 \times 80$ | $\Delta \mathrm{~V}_{\text {W }} / \Delta T$ |  | 100 |  | $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$ |
| Full-Scale Error | Code $=0 \times F F$ | $\mathrm{~V}_{\text {WFSE }}$ | -3 | -1 | 0 | LSB |
| Zero-Scale Error | Code $=0 \times 00$ | $\mathrm{~V}_{\text {WZSE }}$ | 0 | 1 | 3 | LSB |

## RESISTOR TERMINALS

| Voltage Range (Note 7) |  | $\mathrm{V}_{\mathrm{A}, \mathrm{B}, \mathrm{W}}$ | GND |  | $\mathrm{V}_{\mathrm{DD}}$ | V |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Capacitance (Note 8) A, B | $\begin{gathered} \mathrm{f}=1 \mathrm{MHz}, \text { measured to GND, } \\ \text { Code }=0 \times 80 \end{gathered}$ | $\mathrm{C}_{\text {A, }}$ |  | 45 |  | pF |
| Capacitance (Note 8) W | $\begin{gathered} \mathrm{f}=1 \mathrm{MHz}, \text { measured to GND, } \\ \text { Code }=0 \times 80 \end{gathered}$ | $\mathrm{C}_{\text {w }}$ |  | 60 |  | pF |
| Common-Mode Leakage (Note 8) | $\mathrm{V}_{\mathrm{A}}=\mathrm{V}_{\mathrm{B}}=\mathrm{V}_{\mathrm{DD}} / 2$ | $\mathrm{I}_{\text {CM }}$ |  | 1 |  | nA |

DIGITAL INPUTS

| Input Logic High | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | $\mathrm{~V}_{\mathrm{IH}}$ | $0.7 \times \mathrm{V}_{\mathrm{DD}}$ |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Input Logic Low | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | $\mathrm{~V}_{\mathrm{IL}}$ |  |  |  |
| Input Logic High | $\mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V}$ | $\mathrm{~V}_{\mathrm{IH}}$ | $0.7 \times \mathrm{V}_{\mathrm{DD}}$ |  |  |
| Input Logic Low | $\mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V}$ | $\mathrm{~V}_{\mathrm{IL}}$ |  |  |  |
| Input Current | $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$ or 5 V | $\mathrm{I}_{\mathrm{IL}}$ |  | V |  |

## POWER SUPPLIES

| Power Supply Range |  | $\mathrm{V}_{\mathrm{DD} \text { RANGE }}$ | 2.7 |  | 5.5 | V |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Current | $\mathrm{V}_{\mathrm{IH}}=5 \mathrm{~V}$ or $\mathrm{V}_{\mathrm{IL}}=0 \mathrm{~V}$ | $\mathrm{I}_{\mathrm{DD}}$ |  | 0.3 | 2 | $\mu \mathrm{~A}$ |
| Power Dissipation (Note 8) | $\mathrm{V}_{\mathrm{IH}}=5 \mathrm{~V}$ or $\mathrm{V}_{\mathrm{IL}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}=5 \mathrm{~V}$ | $\mathrm{P}_{\mathrm{DISS}}$ |  |  | 0.2 | mW |
| Power Supply Sensitivity | $\Delta \mathrm{V}_{\mathrm{DD}}=+5 \mathrm{~V} \pm 10 \%$, Code $=$ Midscale | PSS |  |  | $\pm 0.05$ | $\% / \%$ |

DYNAMIC CHARACTERISTICS (Notes 8 and 10)

| Bandwidth -3 dB | $\mathrm{R}_{\mathrm{AB}}=50 \mathrm{k} \Omega / 100 \mathrm{k} \Omega$, Code $=0 \times 80$ | BW |  | $100 / 40$ |  | kHz |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Total Harmonic Distortion | $\mathrm{V}_{\mathrm{A}}=1 \mathrm{~V} \mathrm{rms}, \mathrm{~V}_{\mathrm{B}}=0 \mathrm{~V}$, <br> $\mathrm{f}=1 \mathrm{kHz}, \mathrm{R}_{\mathrm{AB}}=10 \mathrm{k} \Omega$ | $\mathrm{THD}_{\mathrm{W}}$ |  | 0.05 |  | $\%$ |
| $\mathrm{~V}_{\mathrm{W}}$ Settling Time $(50 \mathrm{k} \Omega / 100 \mathrm{k} \Omega)$ | $\mathrm{V}_{\mathrm{A}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{B}}=0 \mathrm{~V}, \pm 1 \mathrm{LSB}$ error band | ts |  | 2 |  | $\mu \mathrm{~s}$ |

3. Typical specifications represent average readings at $+25^{\circ} \mathrm{C}$ and $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$.
4. Resistor position nonlinearity error R-INL is the deviation from an ideal value measured between the maximum resistance and the minimum resistance wiper positions. R-DNL measures the relative step change from ideal between successive tap positions. Parts are guaranteed monotonic.
5. $\mathrm{V}_{\mathrm{AB}}=\mathrm{V}_{\mathrm{DD}}$, Wiper $\left(\mathrm{V}_{\mathrm{W}}\right)=$ no connect.
6. INL and DNL are measured at VW with the digital potentiometer configured as a potentiometer divider similar to a voltage output D/A converter. $\mathrm{V}_{\mathrm{A}}=\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{B}}=0 \mathrm{~V}$. DNL specification limits of $\pm 1 \mathrm{LSB}$ maximum are guaranteed monotonic operating conditions.
7. Resistor terminals $\mathrm{A}, \mathrm{B}, \mathrm{W}$ have no limitations on polarity with respect to each other.
8. Guaranteed by design and not subject to production test.
9. Maximum terminal current is bounded by the maximum current handling of the switches, maximum power dissipation of the package, and maximum applied voltage across any two of the $\mathrm{A}, \mathrm{B}$, and W terminals at a given resistance.
10. All dynamic characteristics use $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

Table 5. CAPACITANCE
$\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{f}=1.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$

| Symbol | Test | Conditions | Max | Units |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{C}_{/ / \mathrm{O}}$ <br> (Note 11$)$ | Input/Output Capacitance (SDA, SCL) | $\mathrm{V}_{/ / \mathrm{O}}=0 \mathrm{~V}$ | 8 | pF |

Table 6. POWER UP TIMING (Notes 11 and 12)

| Symbol | Parameter | Max | Units |
| :---: | :--- | :---: | :---: |
| tPuR | Power-up to Read Operation | 1 |  |
| tpuw | Power-up to Write Operation | ms |  |

11. This parameter is tested initially and after a design or process change that affects the parameter.
12. tPUR and $\mathrm{t}_{\text {PUw }}$ are delays required from the time $\mathrm{V}_{\mathrm{CC}}$ is stable until the specified operation can be initiated.

Table 7. DIGITAL POTENTIOMETER TIMING

| Symbol | Parameter | Min | Max |
| :---: | :--- | :---: | :---: |
| tWRPO | Wiper Response Time After Power Supply Stable |  | 50 |
| $t_{\text {WR }}$ | Wiper Response Time: SCL falling edge after last bit of wiper position data byte to <br> wiper change | $\mu \mathrm{s}$ |  |

Table 8. A.C. CHARACTERISTICS
$\mathrm{V}_{\mathrm{DD}}=+2.7 \mathrm{~V}$ to $+5.5 \mathrm{~V},-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ unless otherwise specified.

| Symbol | Parameter | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {SCL }}$ | Clock Frequency |  |  | 400 | kHz |
| $\mathrm{t}_{\text {HIGH }}$ | Clock High Period | 600 |  |  | ns |
| tLow | Clock Low Period | 1300 |  |  | ns |
| tsu:STA | Start Condition Setup Time (for a Repeated Start Condition) | 600 |  |  | ns |
| $t_{\text {HD: }}$ STA | Start Condition Hold Time | 600 |  |  | ns |
| $\mathrm{t}_{\text {SU:DAT }}$ | Data in Setup Time | 100 |  |  | ns |
| $\mathrm{t}_{\mathrm{HD} \text { : } \mathrm{DAT}}$ | Data in Hold Time | 0 |  |  | ns |
| tsu:STO | Stop Condition Setup Time | 600 |  |  | ns |
| $\mathrm{t}_{\text {BUF }}$ | Time the bus must be free before a new transmission can start | 1300 |  |  | ns |
| $\mathrm{t}_{\mathrm{R}}$ | SDA and SCL Rise Time |  |  | 300 | ns |
| $\mathrm{t}_{\mathrm{F}}$ | SDA and SCL Fall Time |  |  | 300 | ns |
| $\mathrm{t}_{\mathrm{DH}}$ | Data Out Hold Time |  | 100 |  | ns |
| $\mathrm{T}_{1}$ | Noise Suppression Time Constant at SCL, SDA Inputs |  |  | 50 | ns |
| $\mathrm{t}_{\text {AA }}$ | SCL Low to SDA Data Out and ACK Out |  |  | 1 | $\mu \mathrm{s}$ |



Figure 2. Differential Non-Linearity,
$V_{D D}=5.6 \mathrm{~V}$


Figure 4. Wiper Resistance at Room Temperature


Figure 3. Integral Non-Linearity,
$\mathrm{V}_{\mathrm{DD}}=5.6 \mathrm{~V}$


Figure 5. Wiper Voltage


Figure 6. Standby Current

TYPICAL CHARACTERISTICS


Figure 7. Change in End-to-End Resistance


Figure 9. Gain vs. Bandwidth (Tap 0x80)


Figure 8. End-to-End Resistance vs. Temperature


Figure 10. PSRR

## BASIC OPERATION

The CAT5171 is a 256 -position digitally controlled potentiometer. When power is first applied, the wiper assumes a mid-scale position. Once the power supply is
stable, the wiper may be repositioned via the $\mathrm{I}^{2} \mathrm{C}$ compatible interface.

## PROGRAMMING: VARIABLE RESISTOR

## Rheostat Mode

The resistance between terminals A and $\mathrm{B}, \mathrm{R}_{\mathrm{AB}}$, has a nominal value of $50 \mathrm{k} \Omega$ or $100 \mathrm{k} \Omega$ and has 256 contact points accessed by the wiper terminal, plus the B terminal contact. Data in the 8 -bit Wiper register is decoded to select one of these 256 possible settings.

The wiper's first connection is at the B terminal, corresponding to control position $0 \times 00$. Ideally this would present a $0 \Omega$ between the Wiper and $B$, but just as with a mechanical rheostat there is a small amount of contact resistance to be considered, there is a wiper resistance comprised of the $\mathrm{R}_{\mathrm{ON}}$ of the FET switch connecting the wiper output with its respective contact point. In CAT5171 this 'contact' resistance is typically $50 \Omega$. Thus a connection setting of $0 x 00$ yields a minimum resistance of $50 \Omega$ between terminals W and B.

For a $100 \mathrm{k} \Omega$ device, the second connection, or the first tap point, corresponds to $441 \Omega\left(\mathrm{R}_{\mathrm{WB}}=\mathrm{R}_{\mathrm{AB}} / 256+\mathrm{R}_{\mathrm{W}}=390.6\right.$ $+50 \Omega)$ for data $0 \times 01$. The third connection is the next tap point, is $831 \Omega(2 \times 390.6+50 \Omega)$ for data $0 \times 02$, and so on. Figure 11 shows a simplified equivalent circuit where the last resistor string will not be accessed; therefore, there is 1 LSB less of the nominal resistance at full scale in addition to the wiper resistance.


Figure 11. CAT5171 Equivalent Digital POT Circuit
The equation for determining the digitally programmed output resistance between W and B is

$$
\begin{equation*}
R_{W B}=\frac{D}{256} R_{A B}+R_{W} \tag{eq.1}
\end{equation*}
$$

where D is the decimal equivalent of the binary code loaded in the 8-bit Wiper register, $\mathrm{R}_{\mathrm{AB}}$ is the end-to-end resistance,
and $\mathrm{R}_{\mathrm{W}}$ is the wiper resistance contributed by the on resistance of the internal switch.
In summary, if $R_{A B}=100 \mathrm{k} \Omega$ and the A terminal is open circuited, the following output resistance $\mathrm{R}_{\mathrm{WB}}$ will be set for the indicated Wiper register codes:

Table 9. CODES AND CORRESPONDING R ${ }_{\text {WB }}$ RESISTANCE FOR $R_{A B}=100 \mathrm{k} \Omega, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$

| $\mathbf{D}$ (Dec.) | $\mathbf{R}_{\text {WB }}(\boldsymbol{\Omega})$ | Output State |
| :---: | :---: | :--- |
| 255 | 99,559 | Full Scale $\left(\mathrm{R}_{\text {AB }}-1 \mathrm{LSB}+\mathrm{R}_{\mathrm{W}}\right)$ |
| 128 | 50,050 | Midscale |
| 1 | 441 | 1 LSB |
| 0 | 50 | Zero Scale <br> (Wiper Contact Resistance) |

Be aware that in the zero-scale position, the wiper resistance of $50 \Omega$ is still present. Current flow between W and B in this condition should be limited to a maximum pulsed current of no more than 20 mA . Failure to heed this restriction can cause degradation or possible destruction of the internal switch contact.

Similar to the mechanical potentiometer, the resistance of the digital POT between the wiper W and terminal A also produces a digitally controlled complementary resistance $\mathrm{R}_{\mathrm{WA}}$. When these terminals are used, the B terminal can be opened. Setting the resistance value for $\mathrm{R}_{\mathrm{WA}}$ starts at a maximum value of resistance and decreases as the data loaded in the latch increases in value. The general equation for this operation is

$$
\begin{equation*}
R_{W A}(D)=\frac{256-D}{256} R_{A B}+R_{W} \tag{eq.2}
\end{equation*}
$$

For $\mathrm{R}_{\mathrm{AB}}=100 \mathrm{k} \Omega$ and the B terminal open circuited, the following output resistance $R_{W A}$ will be set for the indicated Wiper register codes.

Table 10. CODES AND CORRESPONDING RWA RESISTANCE FOR $R_{A B}=100 \mathrm{k} \Omega, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$

| $\mathbf{D}$ (Dec.) | $\mathbf{R}_{\text {WA }}(\boldsymbol{\Omega})$ | Output State |
| :---: | :---: | :--- |
| 255 | 441 | Full Scale |
| 128 | 50,050 | Midscale |
| 1 | 99,659 | 1 LSB |
| 0 | 100,050 | Zero Scale |

Typical device to device resistance matching is lot dependent and may vary by up to $\pm 20 \%$.

## ESD Protection



Figure 12. ESD Protection Networks

## Terminal Voltage Operating Range

The CAT5171 V ${ }_{\text {DD }}$ and GND power supply define the limits for proper 3-terminal digital potentiometer operation. Signals or potentials applied to terminals A, B or the wiper must remain inside the span of $\mathrm{V}_{\mathrm{DD}}$ and GND. Signals which attempt to go outside these boundaries will be clamped by the internal forward biased diodes.


Figure 13.

## Power-up Sequence

Because ESD protection diodes limit the voltage compliance at terminals A, B, and W (see Figure 12), it is recommended that $\mathrm{V}_{\mathrm{DD}} / \mathrm{GND}$ be powered before applying any voltage to terminals $\mathrm{A}, \mathrm{B}$, and W . The ideal power-up sequence is: $\mathrm{GND}, \mathrm{V}_{\mathrm{DD}}$, digital inputs, and then $\mathrm{V}_{\mathrm{A} / \mathrm{B} / \mathrm{W}}$. The order of powering $\mathrm{V}_{\mathrm{A}}, \mathrm{V}_{\mathrm{B}}, \mathrm{V}_{\mathrm{W}}$, and the digital inputs is not important as long as they are powered after $\mathrm{V}_{\mathrm{DD}} / \mathrm{GND}$.

## Power Supply Bypassing

Good design practice employs compact, minimum lead length layout design. Leads should be as direct as possible. It is also recommended to bypass the power supplies with quality low ESR Ceramic chip capacitors of $0.01 \mu \mathrm{~F}$ to $0.1 \mu \mathrm{~F}$. Low ESR $1 \mu \mathrm{~F}$ to $10 \mu \mathrm{~F}$ tantalum or electrolytic capacitors can also be applied at the supplies to suppress transient disturbances and low frequency ripple. As a further precaution digital ground should be joined remotely to the analog ground at one point to minimize the ground bounce.


Figure 14. Power Supply Bypassing

## $I^{2} \mathrm{C}$ BUS PROTOCOL

The following defines the features of the $\mathrm{I}^{2} \mathrm{C}$ bus protocol:

1. Data transfer may be initiated only when the bus is not busy.
2. During a data transfer, the data line must remain stable whenever the clock line is high. Any changes in the data line while the clock is high will be interpreted as a START or STOP condition.

The device controlling the transfer is a master, typically a processor or controller, and the device being controlled is the slave. The master will always initiate data transfers and provide the clock for both transmit and receive operations. Therefore, the CAT5171 will be considered a slave device in all applications.

## START Condition

The START condition precedes all commands to the device, and is defined as a high to low transition of SDA when SCL is high. The CAT5171 monitors the SDA and SCL lines and will not respond until this condition is met.

## STOP Condition

A low to high transition of SDA when SCL is high determines the STOP condition. All operations must end with a STOP condition.

## Device Addressing

The bus Master begins a transmission by sending a START condition. The Master then sends the address of the particular slave device it is requesting. The six most significant bits of the 8-bit slave address are fixed as 010110 for the CAT5171. The next bit (AD0) is the device least significant address bit and defines which device the Master is accessing. Up to two devices may be individually addressed by the system. Typically, $+5 \mathrm{~V}\left(\mathrm{~V}_{\mathrm{DD}}\right)$ or ground
is hard-wired to the AD 0 pin to establish the device's address.

After the Master sends a START condition and the slave address byte, the CAT5171 monitors the bus and responds with an acknowledge (on the SDA line) when its address matches the transmitted slave address.

## Acknowledge

After a successful data transfer, each receiving device is required to generate an acknowledge. The Acknowledging device pulls down the SDA line during the ninth clock cycle, signaling that it received the 8 bits of data.
The CAT5171 responds with an acknowledge after receiving a START condition and its slave address. If the device has been selected along with a write operation, it responds with an acknowledge after receiving each 8-bit byte.

When the CAT5171 is in a READ mode it transmits 8 bits of data, releases the SDA line, and monitors the line for an acknowledge. Once it receives this acknowledge, the CAT5171 will continue to transmit data. If no acknowledge is sent by the Master, the device terminates data transmission and waits for a STOP condition.

## Write Operation

In the Write mode, the Master device sends the START condition and the slave address information to the Slave device. After the Slave generates an acknowledge, the Master sends the instruction byte. After receiving another acknowledge from the Slave, the Master device transmits the data to be written into the wiper register. The CAT5171 acknowledges once more and the Master generates the STOP condition.


Figure 15. Bus Timing Diagram

## CAT5171



Figure 16. Start/Stop Condition


Figure 17. Acknowledge Condition

## INSTRUCTION AND REGISTER DESCRIPTION

## Slave Address Byte

The first byte sent to the CAT5171 from the master/processor is called the Slave Address Byte. The most significant six bits of the slave address are a device type identifier. For the CAT5171, these bits are fixed at 010110.

The next bit, AD0, is the first bit of the internal slave address and must match the physical device address which is defined by the state of the AD0 input pin for the CAT5171 to successfully continue the command sequence. Only the device which slave address matches the incoming device address sent by the master executes the instruction. The AD0 input can be actively driven by CMOS input signals or tied to the supply voltage or ground.

The next bit, $\mathrm{R} / \overline{\mathrm{W}}$, indicates whether this command corresponds to a Write or Read instruction. To write into the Wiper control register, $\mathrm{R} / \overline{\mathrm{W}}$ bit is set to a logic low; while a read from the wiper register is done with the bit high.

## Wiper Control

The CAT5171 contains one 8-bit Wiper Control Register (WCR). The Wiper Control Register output is decoded to select one of 256 switches along its resistor array. The contents of the WCR may be written by the host via Write instruction.

The Wiper Control Register is a volatile register that loses its contents when the CAT5171 is powered-down. Upon
power-up, the wiper is set to midscale and may be repositioned anytime after the power has become stable.

## Instructions

Write and Read instructions are respectively three and two bytes in length. The basic sequence of the two instructions is illustrated in Table 11 and 12.

In write mode, the second byte is the instruction byte. The first bit (MSB) of the instruction byte is a don't care. The second MSB, RS, is the midscale reset. A logic high on this bit moves the wiper to the center tap. The third MSB, SD, is a shutdown bit. A logic high causes an open circuit at terminal A , and short the wiper terminal W to terminal B . The "shutdown" operation does not change the contents of the wiper register. When the shutdown bit, SD, goes back to a logic low, the previous wiper position is restored. Also during shutdown, new settings can be programmed. As soon as the device is returned from shutdown, the wiper position is set according to the wiper register value.

## Two CAT5171 on a Single Bus

When needed, it is possible to connect two CAT5171 potentiometers on the same $\mathrm{I}^{2} \mathrm{C}$ bus and be able to address each one independently. Each device can be set to a unique address by using the AD 0 input pin. One device AD 0 pin is connected to ground, and the other device AD 0 pin is tied to the supply voltage.

Table 11. Write



Table 12. READ

| S | 0 | 1 | 0 | 1 | 1 | 0 | ADO | R | A | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | A | P |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Slave Address Byte |  |  |  |  |  |  |  |  | Data Byte |  |  |  |  |  |  |  |  |  |
| SDA | 0 | 1 | 0 | 1 | 1 | 0 | ADO | R/W |  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |
| $\left.\begin{aligned} & S_{1}^{T} \\ & A \\ & A_{1} \\ & T \end{aligned} \right\rvert\,$ |  |  |  |  |  |  |  |  | C |  |  |  | Dat |  |  |  |  | N A C K | S T O P |

## Legend

$S=$ Start
$\mathrm{P}=$ Stop
A = Acknowledge
$\mathrm{AD} 0=$ Address bit 0 , needed when using two potentiometers on the same $\mathrm{I}^{2} \mathrm{C}$ bus.
$\mathrm{D}=$ Data bit
$\mathrm{R}=$ Read (bit is 1 for Read instruction)
$\overline{\mathrm{W}}=$ Write (bit is 0 for Write instruction)
$\mathrm{RS}=$ When the bit is 1 , the wiper position is moved to mid-scale $0 \times 80$

SD = Shut Down:
0: normal operation 1: wiper is parked at B terminal and terminal A is open circuit.
$\mathrm{X}=$ Don't Care

## PACKAGE DIMENSIONS

SOT-23, 8 Lead
CASE 527AK
ISSUE A


TOP VIEW

| SYMBOL | MIN | NOM | MAX |  |
| :---: | :---: | :---: | :---: | :---: |
| A | 0.90 |  | 1.45 |  |
| A1 | 0.00 |  | 0.15 |  |
| A2 | 0.90 | 1.10 | 1.30 |  |
| A3 | 0.60 |  | 0.80 |  |
| b | 0.28 |  | 0.38 |  |
| c | 0.08 | 2.90 BSC |  |  |
| D | 2.80 BSC |  |  |  |
| E | 0.60 BSC |  |  |  |
| E1 | 0.65 BSC |  |  |  |
| e | 0.60 REF |  |  |  |
| L | 0.30 | 0.60 |  |  |
| L1 | 0.25 REF |  |  |  |
| L2 |  |  |  |  |
| $\theta$ | $0^{\circ}$ |  |  |  |



SIDE VIEW

## Notes:

(1) All dimensions in millimeters. Angles in degrees.
(2) Complies with JEDEC standard MO-178.

END VIEW


END VIEW


#### Abstract

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