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## Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832
Email \& Skype: info@chipsmall.com Web: www.chipsmall.com Address: A1208, Overseas Decoration Building, \#122 Zhenhua RD., Futian, Shenzhen, China

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 


#### Abstract

General Description The DS1859 dual, temperature-controlled, nonvolatile (NV) variable resistors with three monitors consists of two $50 \mathrm{k} \Omega$ or two $20 \mathrm{k} \Omega$, 256-position, linear, variable resistors; three analog monitor inputs (MON1, MON2, MON3); and a direct-to-digital temperature sensor. The device provides an ideal method for setting and tem-perature-compensating bias voltages and currents in control applications using minimal circuitry. The variable resistor settings are stored in EEPROM memory and can be accessed over the 2-wire serial bus.


## Applications

Optical Transceivers
Optical Transponders
Instrumentation and Industrial Controls
RF Power Amps
Diagnostic Monitoring

Typical Operating Circuit


Features

- SFF-8472 Compatible
- Five Monitored Channels (Temperature, Vcc, MON1, MON2, MON3)
- Three External Analog Inputs (MON1, MON2, MON3) That Support Internal and External Calibration
- Scalable Dynamic Range for External Analog Inputs
- Internal Direct-to-Digital Temperature Sensor
- Alarm and Warning Flags for All Monitored Channels
- Two 50k $\Omega$ or Two 20k $\Omega$, Linear, 256-Position, Nonvolatile Temperature-Controlled Variable Resistors
- Resistor Settings Changeable Every $2^{\circ} \mathrm{C}$
- Access to Monitoring and ID Information Configurable with Separate Device Addresses
- 2-Wire Serial Interface
- Two Buffers with TTL/CMOS-Compatible Inputs and Open-Drain Outputs
- Operates from a 3.3V or 5V Supply
- Operating Temperature Range of $-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$

Ordering Information

| PART | RESISTANCE | PIN-PACKAGE |
| :--- | :---: | :--- |
| DS1859B-020 | $20 \mathrm{k} \boldsymbol{\Omega}$ | 16 CSBGA |
| DS1859B-020 + | $20 \mathrm{k} \Omega$ | 16 CSBGA |
| DS1859B-050 | $50 \mathrm{k} \Omega$ | 16 CSBGA |
| DS1859B-050+ | $50 \mathrm{k} \Omega$ | 16 CSBGA |

+Denotes lead-free package.
Ordering Information continued at end of data sheet.

Pin Configurations


## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

## ABSOLUTE MAXIMUM RATINGS

Voltage Range on $\mathrm{V}_{\mathrm{CC}}$ Relative to Ground ...........-0.5V to +6.0 V
Voltage Range on Inputs Relative
to Ground* ...............................................-0.5V to V $\mathrm{CC}+0.5 \mathrm{~V}$
Voltage Range on Resistor Inputs Relative to Ground*
.-0.5 V to $\mathrm{V} \mathrm{CC}+0.5 \mathrm{~V}$
Current into Resistors
.5 mA
*Not to exceed 6.0V.

Operating Temperature Range ........................... $-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$ Programming Temperature Range ......................... $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$
Storage Temperature Range ............................. $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ Soldering Temperature .See IPC/JEDEC J-STD-020A

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## RECOMMENDED DC OPERATING CONDITIONS

( $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage | $V_{C C}$ | (Note 1) | 2.85 | 5.5 | V |
| Input Logic 1 (SDA, SCL, WPEN) | $\mathrm{V}_{\mathrm{IH}}$ | (Note 2) | $0.7 \times \mathrm{Vcc}$ | $V_{C C}+0.3$ | V |
| Input Logic 0 (SDA, SCL, WPEN) | $\mathrm{V}_{\text {IL }}$ | (Note 2) | -0.3 | $+0.3 \times \mathrm{VCC}$ | V |
| Resistor Inputs (L0, L1, H0, H1) |  |  | -0.3 | $V_{C C}+0.3$ | V |
| Resistor Current | IRES |  | -3 | +3 | mA |
| High-Z Resistor Current | IROFF |  | 0.001 | 0.1 | $\mu \mathrm{A}$ |
| Input Logic Levels (IN1, IN2) |  | Input logic 1 | 1.5 |  | V |
|  |  | Input logic 0 |  | 0.9 |  |

## DC ELECTRICAL CHARACTERISTICS

( $\mathrm{V}_{\mathrm{CC}}=2.85 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Current | IcC | (Note 3) |  | 1 | 2 | mA |
| Input Leakage | IIL |  | -200 |  | +200 | nA |
| Low-Level Output Voltage (SDA, OUT1, OUT2) | VOL1 | 3mA sink current | 0 |  | 0.4 | V |
|  | VOL2 | 6 mA sink current | 0 |  | 0.6 |  |
| Full-Scale Input (MON1, MON2, MON3) |  | At factory setting (Note 4) | 2.4875 | 2.5 | 2.5125 | V |
| Full-Scale VCC Monitor |  | At factory setting (Note 5) | 6.5208 | 6.5536 | 6.5864 | V |
| I/O Capacitance | $\mathrm{Cl}_{1 / \mathrm{O}}$ |  |  |  | 10 | pF |
| WPEN Pullup | RWPEN |  | 40 | 65 | 100 | $\mathrm{k} \Omega$ |
| Digital Power-On Reset | POD |  | 1.0 |  | 2.2 | V |
| Analog Power-On Reset | POA |  | 2.0 |  | 2.6 | V |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

## ANALOG RESISTOR CHARACTERISTICS

( $\mathrm{V} C \mathrm{CC}=2.85 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Position 00h Resistance (50k ) |  | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 0.65 | 1.0 | 1.35 | $\mathrm{k} \Omega$ |
| Position FFh Resistance (50k ) |  | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 40 | 50 | 60 | $\mathrm{k} \Omega$ |
| Position 00h Resistance (20k ) |  | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 0.20 | 0.40 | 0.55 | $\mathrm{k} \Omega$ |
| Position FFh Resistance (20k $\Omega$ ) |  | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 16 | 20 | 24 | k $\Omega$ |
| Absolute Linearity |  | (Note 6) | -2 |  | +2 | LSB |
| Relative Linearity |  | (Note 7) | -1 |  | +1 | LSB |
| Temperature Coefficient |  | (Note 8) |  | 50 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |

## ANALOG VOLTAGE MONITORING

( $\mathrm{V}_{\mathrm{CC}}=2.85 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Input Resolution | $\triangle \mathrm{VMON}$ |  | 610 |  | $\mu \mathrm{V}$ |
| Supply Resolution | $\Delta \mathrm{V}_{\mathrm{CC}}$ |  | 1.6 |  | mV |
| Input/Supply Accuracy (MON1, MON2, MON3, VCC) | Acc | At factory setting | 0.25 | 0.5 | $\begin{gathered} \text { \% FS } \\ \text { (full scale) } \end{gathered}$ |
| Update Rate for MON1, MON2, MON3, Temp, or Vcc | trame |  | 30 | 45 | ms |
| Input/Supply Offset (MON1, MON2, MON3, Vcc) | Vos | (Note 14) | 0 | 5 | LSB |

## DIGITAL THERMOMETER

( $\mathrm{V}_{\mathrm{CC}}=2.85 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$, unless otherwise noted.)


## NONVOLATILE MEMORY CHARACTERISTICS

$\left(\mathrm{V}_{\mathrm{CC}}=2.85 \mathrm{~V}\right.$ to 5.5 V )

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EEPROM Writes |  | $+70^{\circ} \mathrm{C}$ | 50,000 |  |  |  |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

## AC ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}_{\mathrm{CC}}=2.85 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+95^{\circ} \mathrm{C}$, unless otherwise noted. See Figure 6.)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| SCL Clock Frequency | fSCL | Fast mode (Note 9) | 0 | 400 | kHz |
|  |  | Standard mode (Note 9) | 0 | 100 |  |
| Bus Free Time Between STOP and START Condition | tBUF | Fast mode (Note 9) | 1.3 |  | $\mu \mathrm{s}$ |
|  |  | Standard mode (Note 9) | 4.7 |  |  |
| Hold Time (Repeated) START Condition | thD:STA | Fast mode (Notes 9, 10) | 0.6 |  | $\mu \mathrm{s}$ |
|  |  | Standard mode (Notes 9, 10) | 4.0 |  |  |
| LOW Period of SCL Clock | tıow | Fast mode (Note 9) | 1.3 |  | $\mu \mathrm{s}$ |
|  |  | Standard mode (Note 9) | 4.7 |  |  |
| HIGH Period of SCL Clock | thigh | Fast mode (Note 9) | 0.6 |  | $\mu \mathrm{S}$ |
|  |  | Standard mode (Note 9) | 4.0 |  |  |
| Data Hold Time | thD:DAT | Fast mode (Notes 9, 11, 12) | 0 | 0.9 | $\mu \mathrm{s}$ |
|  |  | Standard mode (Notes 9, 11, 12) | 0 |  |  |
| Data Setup Time | tsu:DAT | Fast mode (Note 9) | 100 |  | ns |
|  |  | Standard mode (Note 9) | 250 |  |  |
| START Setup Time | tSU:STA | Fast mode (Note 9) | 0.6 |  | $\mu \mathrm{S}$ |
|  |  | Standard mode (Note 9) | 4.7 |  |  |
| Rise Time of Both SDA and SCL Signals | tR | Fast mode (Note 13) | $20+0.1 C_{B}$ | 300 | ns |
|  |  | Standard mode (Note 13) | $20+0.1 C_{B}$ | 1000 |  |
| Fall Time of Both SDA and SCL Signals | ${ }_{\text {t }}$ | Fast mode (Note 13) | $20+0.1 C_{B}$ | 300 | ns |
|  |  | Standard mode (Note 13) | $20+0.1 C_{B}$ | 300 |  |
| Setup Time for STOP Condition | tSU:STO | Fast mode | 0.6 |  | $\mu \mathrm{S}$ |
|  |  | Standard mode | 4.0 |  |  |
| Capacitive Load for Each Bus Line | СВ | (Note 13) |  | 400 | pF |
| EEPROM Write Time | tw | (Note 14) | 10 | 20 | ms |

Note 1: All voltages are referenced to ground
Note 2: I/O pins of fast-mode devices must not obstruct the SDA and SCL lines if $\mathrm{V}_{\mathrm{Cc}}$ is switched off.
Note 3: SDA and SCL are connected to $\mathrm{V}_{\mathrm{CC}}$ and all other input signals are connected to well-defined logic levels.
Note 4: Full Scale is user programmable. The maximum voltage that the MON inputs read is approximately Full Scale, even if the voltage on the inputs is greater than Full Scale.
Note 5: This voltage defines the maximum range of the analog-to-digital converter voltage, not the maximum VCC voltage.
Note 6: Absolute linearity is the difference of measured value from expected value at DAC position. The expected value is a straight line from measured minimum position to measured maximum position.
Note 7: Relative linearity is the deviation of an LSB DAC setting change vs. the expected LSB change. The expected LSB change is the slope of the straight line from measured minimum position to measured maximum position.
Note 8: See the Typical Operating Characteristics.
Note 9: A fast-mode device can be used in a standard-mode system, but the requirement tsu:DAT $>250 \mathrm{~ns}$ must then be met. This is automatically the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line trMAX + tsu:DAT $=1000 \mathrm{~ns}+250 \mathrm{~ns}=1250 \mathrm{~ns}$ before the SCL line is released.

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Note 10: After this period, the first clock pulse is generated.
Note 11: The maximum thD:DAT only has to be met if the device does not stretch the LOW period (tLow) of the SCL signal.
Note 12: A device must internally provide a hold time of at least 300ns for the SDA signal (see the $\mathrm{V}_{\mathrm{IH}}$ MIN of the SCL signal) to bridge the undefined region of the falling edge of SCL.
Note 13: $\mathrm{C}_{\mathrm{B}}$-total capacitance of one bus line, timing referenced to $0.9 \times \mathrm{V}_{\mathrm{CC}}$ and $0.1 \times \mathrm{V}_{\mathrm{CC}}$.
Note 14: Guaranteed by design.
$\left(\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, for both $50 \mathrm{k} \Omega$ and $20 \mathrm{k} \Omega$ versions, unless otherwise noted.)


## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Typical Operating Characteristics (continued)
$\left(\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, for both $50 \mathrm{k} \Omega$ and $20 \mathrm{k} \Omega$ versions, unless otherwise noted.)


RESISTANCE
vs. POWER-UP VOLTAGE


RESISTOR 1 INL (LSB)


RESISTOR 1 DNL (LSB)


RESISTANCE
vs. POWER-UP VOLTAGE



# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

## Typical Operating Characteristics (continued)

$\left(\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, for both $50 \mathrm{k} \Omega$ and $20 \mathrm{k} \Omega$ versions, unless otherwise noted.)


LSB ERROR vs. FULL-SCALE INPUT


POSITION FFh RESISTANCE
vs. TEMPERATURE


POSITION FFh RESISTANCE
vs. TEMPERATURE


TEMPERATURE COEFFICIENT vs. SETTING


LSB ERROR vs. FULL-SCALE INPUT


## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Pin Description

| PIN | BALL | NAME |  |
| :---: | :---: | :---: | :--- |
| 1 | B2 | SDA | 2-Wire Serial Data I/O Pin. Transfers serial data to and from the device. |
| 2 | A2 | SCL | 2-Wire Serial Clock Input. Clocks data into and out of the device. |
| 3 | C3 | OUT1 | Open-Drain Buffer Output |
| 4 | A1 | IN1 | TTL/CMOS-Compatible Input to Buffer |
| 5 | B1 | OUT2 | Open-Drain Buffer Output |
| 6 | C2 | IN2 | TTL/CMOS-Compatible Input to Buffer |
| 7 | C1 | WPEN | Write Protect Enable. The device is not write protected if WPEN is connected to ground. This pin has <br> an internal pullup (RwPEN). See Table 6. |
| 8 | D1 | GND | Ground |
| 9 | D3 | MON1 | External Analog Input |
| 10 | D4 | MON2 | External Analog Input |
| 11 | C4 | MON3 | External Analog Input |
| 12 | D2 | L0 | Low-End Resistor 0 Terminal. It is not required that the low-end terminals be connected to a potential <br> less than the high-end terminals of the corresponding resistor. Voltage applied to any of the resistor <br> terminals cannot exceed the power-supply voltage, VCC, or go below ground. |
| 13 | B3 | H0 | High-End Resistor 0 Terminal. It is not required that the high-end terminals be connected to a <br> potential greater than the low-end terminals of the corresponding resistor. Voltage applied to any of <br> the resistor terminals cannot exceed the power-supply voltage, VCC, or go below ground. |
| 14 | B4 | L1 | Low-End Resistor 1 Terminal |
| 15 | A4 | H1 | High-End Resistor 1 Terminal |
| 16 | A3 | VCC | Supply Voltage |

## Detailed Description

The user can read the registers that monitor the Vcc, MON1, MON2, MON3, and temperature analog signals. After each signal conversion, a corresponding bit is set that can be monitored to verify that a conversion has occurred. The signals also have alarm and warning flags that notify the user when the signals go above or below the user-defined value. Interrupts can also be set for each signal.
The position values of each resistor can be independently programmed. The user can assign a unique value to each resistor for every $2^{\circ} \mathrm{C}$ increment over the $-40^{\circ} \mathrm{C}$ to $+102^{\circ} \mathrm{C}$ range.

Two buffers are provided to convert logic-level inputs into open-drain outputs. Typically, these buffers are used to implement transmit (Tx) fault and loss-of-signal (LOS) functionality. Additionally, OUT1 can be asserted in the event that one or more of the monitored values go beyond user-defined limits.

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors



Figure 1. Block Diagram

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

Table 1. Scales for Monitor Channels at Factory Setting

| SIGNAL | +FS <br> SIGNAL | +FS <br> (hex) | -FS <br> SIGNAL | -FS <br> (hex) |
| :---: | :---: | :---: | :---: | :---: |
| Temperature | $127.984^{\circ} \mathrm{C}$ | $7 F F C$ | $-128^{\circ} \mathrm{C}$ | 8000 |
| VCC $^{\text {7C }}$ | 6.5528 V | FFF8 | 0 V | 0000 |
| MON1 | 2.4997 V | FFF8 | OV | 0000 |
| MON2 | 2.4997 V | FFF8 | OV | 0000 |
| MON3 | 2.4997 V | FFF8 | OV | 0000 |

Table 2. Signal Comparison

| SIGNAL | FORMAT |
| :---: | :---: |
| VCC $_{\text {CO }}$ | Unsigned |
| MON1 | Unsigned |
| MON2 | Unsigned |
| MON3 | Unsigned |
| Temperature | Two's complement |

## Monitored Signals

Each signal (Vcc, MON1, MON2, MON3, and temperature) is available as a 16 -bit value with 12 -bit accuracy (left-justified) over the serial bus. See Table 1 for signal scales and Table 2 for signal format. The four LSBs should be masked when calculating the value.
For the $20 \mathrm{k} \Omega$ version, the 3 LSBs are internally masked with 0 s.
The signals are updated every frame rate (trame) in a round-robin fashion.
The comparison of all five signals with the high and low user-defined values are done automatically. The corresponding flags are set to 1 within a specified time of the occurrence of an out-of-limit condition.

## Calculating Signal Values

The LSB $=100 \mu \mathrm{~V}$ for V CC , and the $\mathrm{LSB}=38.147 \mu \mathrm{~V}$ for the MON signals when using factory default settings.

## Monitor/Vcc Bit Weights

| MSB | $2^{15}$ | $2^{14}$ | $2^{13}$ | $2^{12}$ | $2^{11}$ | $2^{10}$ | $2^{9}$ | $2^{8}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LSB | $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |

## Vcc Conversion Examples

| MSB (BIN) | LSB (BIN) | VOLTAGE (V) |
| :---: | :---: | :---: |
| 10000000 | 10000000 | 3.29 |
| 11000000 | 11111000 | 4.94 |

Table 3. Look-up Table Address for Corresponding Temperature Values

| TEMPERATURE <br> $\left({ }^{\circ} \mathbf{C}\right)$ | CORRESPONDING LOOK-UP <br> TABLE ADDRESS |
| :---: | :---: |
| $<-40$ | 80 h |
| -40 | 80 h |
| -38 | 81 h |
| -36 | 82 h |
| -34 | 83 h |
| - | - |
| +98 | C 5 h |
| +100 | C 6 h |
| +102 | C 7 h |
| $>+102$ | C 7 h |

## Monitor Conversion Example

| MSB (BIN) | LSB (BIN) | VOLTAGE (V) |
| :---: | :---: | :---: |
| 11000000 | 00000000 | 1.875 |
| 10000000 | 10000000 | 1.255 |

To calculate $\mathrm{V}_{\mathrm{C}}$, convert the unsigned 16 -bit value to decimal and multiply by $100 \mu \mathrm{~V}$.
To calculate MON1, MON2, or MON3, convert the unsigned 16 -bit value to decimal and multiply by $38.147 \mu \mathrm{~V}$.
To calculate the temperature, treat the two's complement value binary number as an unsigned binary number, then convert to decimal and divide by 256. If the result is greater than or equal to 128, subtract 256 from the result.
Temperature: high byte: $-128^{\circ} \mathrm{C}$ to $+127^{\circ} \mathrm{C}$ signed; low byte: $1 / 256^{\circ} \mathrm{C}$.

## Temperature Bit Weights

| $S$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $2^{-1}$ | $2^{-2}$ | $2^{-3}$ | $2^{-4}$ | $2^{-5}$ | $2^{-6}$ | $2^{-7}$ | $2^{-8}$ |

Temperature Conversion Examples

| MSB (BIN) | LSB (BIN) | TEMPERATURE ( ${ }^{\circ} \mathbf{C}$ ) |
| :---: | :---: | :---: |
| 01000000 | 00000000 | 64 |
| 01000000 | 00001111 | 64.059 |
| 01011111 | 00000000 | 95 |
| 11110110 | 00000000 | -10 |
| 11011000 | 00000000 | -40 |

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

Table 4. ADEN Address Configuration

| ADEN <br> (ADDRESS <br> ENABLE) | NO. OF SEPARATE <br> DEVICE <br> ADDRESSES | ADDITIONAL <br> INFORMATION |
| :---: | :---: | :---: |
| 0 | 2 | See Figure 2 |
| 1 | 1 (Main Device only) | See Figure 3 |

Table 5. ADEN and ADFIX Bits

| ADEN | ADFIX | AUXILIARY <br> ADDRESS | MAIN ADDRESS |
| :---: | :---: | :---: | :---: |
| 0 | 0 | A0h | A2h |
| 0 | 1 | A0h | EEPROM <br> (Table 01, 8Ch) |
| 1 | 0 | N/A | A2h |
| 1 | 1 | N/A | EEPROM <br> (Table 01, 8Ch) |



Figure 2. Memory Organization, $A D E N=0$

## Variable Resistors

The value of each variable resistor is determined by a temperature-addressed look-up table, which can assign a unique value ( 00 h to FFh ) to each resistor for every $2^{\circ} \mathrm{C}$ increment over the $-40^{\circ} \mathrm{C}$ to $+102^{\circ} \mathrm{C}$ range (see Table 3). See the Temperature Conversion section for more information.
The variable resistors can also be used in manual mode. If the TEN bit equals 0 , the resistors are in manual mode and the temperature indexing is disabled. The user sets the resistors in manual mode by writing to addresses 82h and 83h in Table 01 to control resistors 0 and 1, respectively.

Memory Description
Main and auxiliary memories can be accessed by two separate device addresses. The Main Device address is A2h (or value in Table 01 byte 8Ch, when ADFIX = 1) and the Auxiliary Device address is AOh. A user option is provided to respond to one or two device addresses. This feature can be used to save component count in SFF applications (Main Device address can be used) or other applications where both GBIC (Auxiliary Device address can be used) and monitoring functions are implemented and two device addresses are needed. The memory blocks are enabled with the corresponding device address. Memory space from 80h and

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors



Figure 3. Memory Organization, $A D E N=1$
above is accessible only through the Main Device address. This memory is organized as three tables. The desired table can be selected by the contents of memory location 7Fh, Main Device. The Auxiliary Device address has no access to the tables, but the Auxiliary Device address can be mapped into the Main Device's memory space as a fourth table. Device addresses are programmable with two control bits in EEPROM.

Table 6. Main Device

| WPEN | MPEN | PROTECT MAIN |
| :---: | :---: | :---: |
| 0 | $X$ | No |
| $X$ | 0 | No |
| 1 | 1 | Yes |

Table 7. Auxiliary Device

| APEN | WPEN | PROTECT AUXILIARY |
| :---: | :---: | :---: |
| 0 | $X$ | No |
| 1 | $X$ | Yes |

ADEN configures memory access to respond to different device addresses (see Tables 4 and 5).
The default device address for EEPROM-generated addresses is A2h.
If the ADEN bit is 1, additional 128 bytes of EEPROM are accessible through the Main Device, selected as Table 00 (see Figure 3). In this configuration, the Auxiliary Device is not accessible. APEN controls the protection of Table 00 regardless of ADEN's setting.
ADFIX (address fixed) determines whether the Main Device address is determined by an EEPROM byte (Table 01, byte 8Ch, when ADFIX = 1). There can be up to 128 devices sharing a common 2 -wire bus, with each device having its own unique device address.

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

## Register Map

A description of the registers is below. The registers are read only (R) or read/write (R/W). The R/W registers are writable only if write protect has not been asserted (see the Memory Description section).

Bytes designated as "Reserved" have been set aside for added functionality in future revisions of this device.

Auxiliary Device

| MEMORY LOCATION <br> (hex) | EEPROM/SRAM | R/w | DEFAULT SETTING <br> (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 00 to 7F | EEPROM | R/W | 00 | Standards Data | - |

Main Device

| $\begin{aligned} & \text { MEMORY } \\ & \text { LOCATION } \\ & \text { (hex) } \end{aligned}$ | EEPROM/ SRAM | R/W | DEFAULT SETTING (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 00 to 01 | EEPROM | R/W | 00 | TMPlimhi (MSB to LSB) | Contains upper limit settings for temperature. If the limit is violated, an alarm flag in Main Device byte 70h is set. |
| 02 to 03 | EEPROM | R/W | 00 | TMPlimlo (MSB to LSB) | Contains lower limit settings for temperature. If the limit is violated, an alarm flag in Main Device byte 70h is set. |
| 04 to 05 | EEPROM | R/W | 00 | TMPwrnhi (MSB to LSB) | Contains upper limit settings for temperature. If the limit is violated, a warning flag in Main Device byte 74h is set. |
| 06 to 07 | EEPROM | R/W | 00 | TMPwrnlo (MSB to LSB) | Contains lower limit settings for temperature. If the limit is violated, a warning flag in Main Device byte 74h is set. |
| 08 to 09 | EEPROM | R/W | 00 | Vcclimhi (MSB to LSB) | Contains upper limit settings for $\mathrm{V}_{\mathrm{CC}}$. If the limit is violated, an alarm flag in Main Device byte 70 h is set. |
| OA to OB | EEPROM | R/W | 00 | Vcclimlo (MSB to LSB) | Contains lower limit settings for $V_{C C}$. If the limit is violated, an alarm flag in Main Device byte 70 h is set. |
| OC to OD | EEPROM | R/W | 00 | Vccwrnhi (MSB to LSB) | Contains upper limit settings for $\mathrm{V}_{\mathrm{Cc}}$. If the limit is violated, a warning flag in Main Device byte 74 h is set. |
| OE to OF | EEPROM | R/W | 00 | Vccwrnlo (MSB to LSB) | Contains lower limit settings for $V_{C C}$. If the limit is violated, a warning flag in Main Device byte 74 h is set. |
| 10 to 11 | EEPROM | R/W | 00 | MON1limhi (MSB to LSB) | Contains upper limit settings for MON1. If the limit is violated, an alarm flag in Main Device byte 70 h is set. |

Note: SRAM defaults are power-on defaults. EEPROM defaults are factory defaults.

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Main Device (continued)

| $\begin{aligned} & \text { MEMORY } \\ & \text { LOCATION } \\ & \text { (hex) } \end{aligned}$ | EEPROM/ SRAM | R/W | DEFAULT SETTING (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 12 to 13 | EEPROM | R/W | 00 | MON1limlo (MSB to LSB) | Contains lower limit settings for MON1. If the limit is violated, an alarm flag in Main Device byte 70h is set. |
| 14 to 15 | EEPROM | R/W | 00 | MON1wrnhi (MSB to LSB) | Contains upper limit settings for MON1. If the limit is violated, a warning flag in Main Device byte 74 h is set. |
| 16 to 17 | EEPROM | R/W | 00 | MON1wrnlo (MSB to LSB) | Contains lower limit settings for MON1. If the limit is violated, a warning flag in Main Device byte 74 h is set. |
| 18 to 19 | EEPROM | R/W | 00 | MON2limhi (MSB to LSB) | Contains upper limit settings for MON2. If the limit is violated, an alarm flag in Main Device byte 70 h is set. |
| 1 A to 1B | EEPROM | R/W | 00 | MON2limlo (MSB to LSB) | Contains lower limit settings for MON2. If the limit is violated, an alarm flag in Main Device byte 70h is set. |
| 1 C to 1D | EEPROM | R/W | 00 | MON2wrnhi (MSB to LSB) | Contains upper limit settings for MON2. If the limit is violated, a warning flag in Main Device byte 74 h is set. |
| 1E to 1F | EEPROM | R/W | 00 | MON2wrnlo (MSB to LSB) | Contains lower limit settings for MON2. If the limit is violated, a warning flag in Main Device byte 74 h is set. |
| 20 to 21 | EEPROM | R/W | 00 | MON3limhi (MSB to LSB) | Contains upper limit settings for MON3. If the limit is violated, an alarm flag in Main Device byte 71 h is set. |
| 22 to 23 | EEPROM | R/W | 00 | MON3limlo (MSB to LSB) | Contains lower limit settings for MON3. If the limit is violated, an alarm flag in Main Device byte 71 h is set. |
| 24 to 25 | EEPROM | R/W | 00 | MON3wrnhi (MSB to LSB) | Contains upper limit settings for MON3. If the limit is violated, a warning flag in Main Device byte 75 h is set. |
| 26 to 27 | EEPROM | R/W | 00 | MON3wrnlo (MSB to LSB) | Contains lower limit settings for MON3. If the limit is violated, a warning flag in Main Device byte 75 h is set. |
| 28 to 37 | EEPROM | - | - | Reserved | - |
| 38 to 5F | EEPROM | R/W | - | Memory | - |
| 60 to 61 | SRAM | R | - | Measured TMP <br> (MSB to LSB) | Digitized measured value for temperature. See Table 1. |
| 62 to 63 | SRAM | R | - | Measured VCC (MSB to LSB) | Digitized measured value for VCC. See Table 1. |
| 64 to 65 | SRAM | R | - | Measured MON1 <br> (MSB to LSB) | Digitized measured value for MON1. See Table 1. |
| 66 to 67 | SRAM | R | - | Measured MON2 <br> (MSB to LSB) | Digitized measured value for MON2. See Table 1. |

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

Main Device (continued)

| MEMORY <br> LOCATION (hex) | EEPROM/ SRAM | R/W | DEFAULT SETTING (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 68 to 69 | SRAM | R | - | Measured MON3 (MSB to LSB) | Digitized measured value for MON3. See Table 1. |
| 6A to 6D | SRAM | - | - | Reserved | - |
| 6E | SRAM | - | - | Logic states | - |
| Bit 7 | - | R | X | HIZSTA | Resistor status bit. A high indicates that both resistors are in high-impedance mode. A low indicates that both resistors are operating normally. |
| 6 | - | R/W | 0 | HIZCO | Resistor control bit. Setting this bit high causes both resistors to go into a highimpedance state. |
| 5 | - | - | X | X |  |
| 4 | - | - | X | X | - |
| 3 | - | - | X | X | - |
| 2 | - | R | X | TXF | This status bit is high when OUT1 is high, assuming there is an external pullup resistor on OUT1. |
| 1 | - | R | X | RXL | This status bit is high when OUT2 is high, assuming there is an external pullup resistor on OUT2. |
| 0 | - | R | X | RDYB | This status bit goes high when $\mathrm{V}_{\mathrm{CC}}$ has fallen below the POA level. |
| 6 F | SRAM | - | - | Conversion updates | - |
| Bit 7 | - | R/W | 0 | TAU | This bit goes high after a temperature and address update has occurred for the corresponding measurement in bytes 60h to 61 h . This bit can be written to a 0 by the user and monitored to verify that a conversion has occurred. |
| 6 | - | R/W | 0 | VccU | This bit goes high after a $\mathrm{V}_{\mathrm{CC}}$ update has occurred for the corresponding measurement in bytes 62 h to 63 h . This bit can be written to a 0 by the user and monitored to verify that a conversion has occurred. |
| 5 | - | R/W | 0 | MON1U | This bit goes high after a MON1 update has occurred for the corresponding measurement in bytes 64 h to 65 h . This bit can be written to a 0 by the user and monitored to verify that a conversion has occurred. |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Main Device (continued)

| $\begin{aligned} & \text { MEMORY } \\ & \text { LOCATION } \\ & \text { (hex) } \end{aligned}$ | $\begin{aligned} & \text { EEPROM/ } \\ & \text { SRAM } \end{aligned}$ | R/W | $\begin{aligned} & \text { DEFAULT } \\ & \text { SETTING } \\ & \text { (hex) } \end{aligned}$ | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 4 | - | R/W | 0 | MON2U | This bit goes high after a MON2 update has occurred for the corresponding measurement in bytes 66h to 67h. This bit can be written to a 0 by the user and monitored to verify that a conversion has occurred. |
| 3 | - | - | 0 | MON3U | This bit goes high after a MON3 update has occurred for the corresponding measurement in bytes 68h to 69h. This bit can be written to a 0 by the user and monitored to verify that a conversion has occurred. |
| 2 | - | - | 0 | - | - |
| 1 | - | - | 0 | - | - |
| 0 | - | - | 0 | - | - |
| 70 | SRAM | R |  | Alarm flags | - |
| Bit 7 | - | - | - | TMPhi | This alarm flag goes high when the upper limit of the temperature setting is violated. |
| 6 | - | - | - | TMPlo | This alarm flag goes high when the lower limit of the temperature setting is violated. |
| 5 | - | - | - | Vcchi | This alarm flag goes high when the upper limit of the $\mathrm{V}_{\mathrm{CC}}$ setting is violated. |
| 4 | - | - | - | Vcclo | This alarm flag goes high when the lower limit of the $\mathrm{V}_{\mathrm{CC}}$ setting is violated. |
| 3 | - | - | - | MON1hi | This alarm flag goes high when the upper limit of the MON1 setting is violated. |
| 2 | - | - | - | MON1Io | This alarm flag goes high when the lower limit of the MON1 setting is violated. |
| 1 | - | - | - | MON2hi | This alarm flag goes high when the upper limit of the MON2 setting is violated. |
| 0 | - | - | - | MON2lo | This alarm flag goes high when the lower limit of the MON2 setting is violated. |
| 71 | SRAM | R | - | Alarm flags | - |
| Bit 7 | - | - | - | MON3hi | This alarm flag goes high when the upper limit of the MON3 setting is violated. |
| 6 | - | - | - | MON3Io | This alarm flag goes high when the lower limit of the MON3 setting is violated. |
| 5 | - | - | - | X | - |
| 4 | - | - | - | X | - |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Main Device (continued)

| $\begin{aligned} & \text { MEMORY } \\ & \text { LOCATION } \\ & \text { (hex) } \end{aligned}$ | EEPROM/ SRAM | R/W | DEFAULT <br> SETTING (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 3 | - | - | - | $X$ | - |
| 2 | - | - | - | X | - |
| 1 | - | - | - | X | - |
| 0 | - | - | - | MINT | A mask of all flags located in Table 01 byte 88 h determines the value of MINT. MINT is maskable to 0 if no interrupt is desired by setting Table 01 byte 88 h to 0 . |
| 72 to 73 | SRAM | - | - | Reserved | - |
| 74 | SRAM | R | - | Warning flags | - |
| Bit 7 | - | - | - | TMPhi | This warning flag goes high when the upper limit of the temperature setting is violated. |
| 6 | - | - | - | TMPlo | This warning flag goes high when the lower limit of the temperature setting is violated. |
| 5 | - | - | - | Vcchi | This warning flag goes high when the upper limit of the $V_{C C}$ setting is violated. |
| 4 | - | - | - | Vcclo | This warning flag goes high when the lower limit of the $\mathrm{V}_{\mathrm{CC}}$ setting is violated. |
| 3 | - | - | - | MON1hi | This warning flag goes high when the upper limit of the MON1 setting is violated. |
| 2 | - | - | - | MON1lo | This warning flag goes high when the lower limit of the MON1 setting is violated. |
| 1 | - | - | - | MON2hi | This warning flag goes high when the upper limit of the MON2 setting is violated. |
| 0 | - | - | - | MON2lo | This warning flag goes high when the lower limit of the MON2 setting is violated. |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Main Device (continued)

| $\begin{aligned} & \text { MEMORY } \\ & \text { LOCATION } \\ & \text { (hex) } \end{aligned}$ | $\begin{aligned} & \text { EEPROM/ } \\ & \text { SRAM } \end{aligned}$ | R/W | $\begin{aligned} & \text { DEFAULT } \\ & \text { SETTING } \\ & \text { (hex) } \end{aligned}$ | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 75 | SRAM | R | - | Warning Flags | - |
| Bit 7 | - | - | 0 | MON3hi | This warning flag goes high when the upper limit of the MON3 setting is violated. |
| 6 | - | - | 0 | MON3Io | This warning flag goes high when the lower limit of the MON3 setting is violated. |
| 5 | - | - | 0 | X | - |
| 4 | - | - | 0 | X | - |
| 3 | - | - | 0 | X | - |
| 2 | - | - | 0 | X | - |
| 1 | - | - | 0 | X | - |
| 76 to 7E | SRAM | - | - | Reserved | - |
| 7F | SRAM | R/W | - | Table select | - |
| Bit 7 | - | - | 0 | X | - |
| 6 | - | - | 0 | X | - |
| 5 | - | - | 0 | X | - |
| 4 | - | - | 0 | X | - |
| 3 | - | - | 0 | X | - |
| 2 | - | -- | 0 | X | - |
| 1 | - | - | 0 | Table select bits | Set bits $=00$ to select Table 00, set bits $=01$ to select Table 01, set bits $=10$ to select Table 02, set bits = 11 to select Table 03. |
| 0 | - | - | 0 |  |  |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

## Table 01h

| MEMORY <br> LOCATION <br> (hex) | EEPROM <br> SRAM | R/W | DEFAULT <br> SETTING <br> (hex) | NAME OF <br> LOCATION | Mode |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 80 | SRAM | R/W |  | FUNCTION |  |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Table 01h (continued)

| MEMORY LOCATION (hex) | EEPROM/ SRAM | R/W | DEFAULT <br> SETTING (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 5 | - | - | 0 | ADEN | Controls if the device responds to one or two device addresses (see the Memory Description section and Table 5). |
| 4 | - | - | 0 | ADFIX | Controls the means by which Main and Auxiliary Device addresses are set (see the Memory Description section and Table 5). |
| 3 | - | - | 0 | APEN | Controls auxiliary write protect. See the Memory Description section. |
| 2 | - | - | 0 | MPEN | Controls main write protect. See the Memory Description section. |
| 1 | - | - | 0 | INV1 | Configures buffer 1 with OUT1 $=$ MINT + (INV1 [XOR] IN1). |
| 0 | - | - | 0 | INV2 | Configures buffer 2 with OUT2 = INV2 [XOR] IN2. |
| 8A to 8B | EEPROM | - | - | Reserved |  |
| 8C | EEPROM | R/W | A2 | Device address | Contains Main Device address if the bit ADFIX = 1. If ADFIX $=0$, then address A2h is used. |
| 8D | EEPROM | - | - | Reserved | - |
| 8E | EEPROM | R/W |  |  | Contains bits used to perform right shift operations on the A/D output converter. See the Right Shift A/D Conversion Result section. |
| 7 | - | - | 0 | - |  |
| 6 | - | - | 0 | $\mathrm{MON1}_{2}$ | Right Shift Control MSB |
| 5 | - | - | 0 | MON1 ${ }_{1}$ |  |
| 4 | - | - | 0 | MON10 | Right Shift Control LSB |
| 3 | - | - | 0 | - |  |
| 2 | - | - | 0 | MON22 | Right Shift Control MSB |
| 1 | - | - | 0 | MON21 |  |
| 0 | - | - | 0 | MON20 | Right Shift Control LSB |
| 8F | EEPROM | R/W |  |  | Contains bits used to perform right shift operations on the A/D output converter. See the Right Shift A/D Conversion Result section. |
| 7 | - | - | 0 | - |  |
| 6 | - | - | 0 | MON32 | Right Shift Control MSB |
| 5 | - | - | 0 | MON31 |  |
| 4 | - | - | 0 | MON30 | Right Shift Control LSB |
| 3 | - | - | 0 | - |  |
| 2 | - | - | 0 | - |  |
| 1 | - | - | 0 | - |  |
| 0 | - | - | 0 | - |  |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Table 01h (continued)

| MEMORY <br> LOCATION <br> (hex) | EEPROM/ SRAM | R/W | DEFAULT SETTING (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 90 to 91 | EEPROM | - | 0 | Reserved |  |
| 92 to 93 | EEPROM | R/W | Factory Programmed | Gain Cal Vcc | Gain registers for internal calibration. See the Internal Calibration section. |
| 94 to 95 | EEPROM | R/W |  | Gain Cal Mon1 |  |
| 96 to 97 | EEPROM | R/W |  | Gain Cal Mon2 |  |
| 98 to 99 | EEPROM | R/W |  | Gain Cal Mon3 |  |
| 9A to 9F | EEPROM | - |  | Reserved |  |
| A0 to A1 | EEPROM | - |  | Reserved |  |
| A2 to A3 | EEPROM | R/W |  | Offset Cal VCC | Offset registers for internal calibration. See the Internal Calibration section. |
| A4 to A5 | EEPROM | R/W |  | Offset Cal Mon1 |  |
| A6 to A7 | EEPROM | R/W |  | Offset Cal Mon2 |  |
| A8 to A9 | EEPROM | R/W |  | Offset Cal Mon3 |  |
| AA to AD | EEPROM | - |  | Reserved |  |
| AE to AF | EEPROM | R/W |  | Offset Cal Tmp | Offset calibration for temperature calibrated at factory. |

## Table 02h

| MEMORY <br> LOCATION <br> (hex) | EEPROM/ <br> SRAM | R/W | DEFAULT <br> SETTING <br> (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 80 to C7 | EEPROM | R/W | FF | Resistor 0 Temp LUT | Look-up table for Resistor 0. |
| F0 to F7 | EEPROM | - | - | Reserved | - |
| F8 to FF | EEPROM | R | Factory <br> Programmed | Resistor 0 Cal Constants | Calibration constants for Resistor 0. <br> (See Table 8) |

## Table 03h

| MEMORY <br> LOCATION <br> (hex) | EEPROM/ <br> SRAM | R/W | DEFAULT <br> SETTING <br> (hex) | NAME OF LOCATION | FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 80 to C7 | EEPROM | R/W | FF | Resistor 1 Temp LUT | Look-up table for Resistor 1. |
| F0 to F7 | EEPROM | - | - | Reserved | - |
| F8 to FF | EEPROM | R | Factory <br> Programmed | Resistor 1 Cal Constants | Calibration constants for Resistor 1. <br> (See Table 8) |

## Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors

Programming the Look-up Table (LUT)
The following equation can be used to determine which resistor position setting, OOh to FFh, should be written in the LUT to achieve a given resistance at a specific temperature.
$\operatorname{pos}(\alpha, \mathrm{R}, \mathrm{C})=\frac{\mathrm{R}-\mathrm{u} \cdot\left[1+\mathrm{v} \cdot(\mathrm{C}-25)+\mathrm{w} \cdot(\mathrm{C}-25)^{2}\right]}{(\mathrm{x}) \cdot\left[1+\mathrm{y} \cdot(\mathrm{C}-25)+\mathrm{z} \cdot(\mathrm{C}-25)^{2}\right]}-\alpha$
$\alpha=3.852357$ for the $20 \mathrm{k} \Omega$ resistor
$\alpha=4.5680475$ for the $50 \mathrm{k} \Omega$ resistor
$R=$ the resistance desired at the output terminal
$\mathrm{C}=$ temperature in degrees Celsius
$\mathrm{u}, \mathrm{v}, \mathrm{w}, \mathrm{x}_{1}, \mathrm{x}_{0}, \mathrm{y}$, and z are calculated values found in the corresponding look-up tables. The variable $x$ from the equation above is separated into $x_{1}$ (the MSB of x ) and $x_{0}$ (the LSB of $x$ ). Their addresses and LSB values are given below. Resistor 0 variables are found in Table 1, and Resistor 1 variables are found in Table 2.
When shipped from the factory, all other memory locations in the LUTs are programmed to FFh.

## Table 8. Calibration Constants

| ADDRESS (Hex) | VARIABLE | LSB |
| :---: | :---: | :---: |
| F8 | $u$ | $2^{0}$ |
| F9 | v | $20 \mathrm{E}-6$ |
| FA | w | $100 \mathrm{E}-9$ |
| FB | $\mathrm{x}_{1}$ | $2^{1}$ |
| FC | $\mathrm{x}_{0}$ | $2^{-7}$ |
| FD | y | $2 \mathrm{E}-6$ (signed) |
| FE | z | $10 \mathrm{E}-9$ |
| FF | Reserved | - |

## Internal Calibration

The DS1859 has two methods for scaling an analog input to a digital result. The two methods are gain and offset. Each of the inputs (Vcc, MON1, MON2, and MON3) has a unique register for the gain and the offset found in Table 01h, 92h to 99h, and A2h to A9h.
To scale the gain and offset of the converter for a specific input, you must first know the relationship between the analog input and the expected digital result. The input that would produce a digital result of all zeros is the null value (normally this input is GND). The input that would produce a digital result of all ones is the fullscale (FS) value. The FS value is also found by multiplying an all-ones digital answer by the weighted LSB (e.g., since the digital reading is a 16 -bit register, let us


Figure 4. Look-Up Table Hysteresis
assume that the LSB of the lowest weighted bit is $50 \mu \mathrm{~V}$, then the FS value is $65,535 \times 50 \mu \mathrm{~V}=3.27675 \mathrm{~V}$ ).
A binary search is used to scale the gain of the converter. This requires forcing two known voltages to the input pin. It is preferred that one of the forced voltages is the null input and the other is $90 \%$ of FS . Since the LSB of the least significant bit in the digital reading register is known, the expected digital results are also known for both inputs (null/LSB = CNT1 and 90\%FS/ LSB = CNT2).
The user might not directly force a voltage on the input. Instead they have a circuit that transforms light, frequency, power, or current to a voltage that is the input to the DS1859. In this situation, the user does not need to know the relationship of voltage to expected digital result but instead knows the relationship of light, frequency, power, or current to the expected digital result.

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

An explanation of the binary search used to scale the gain is best served with the following example pseudocode:
/* Assume that the Null input is 0.5 V . */
$/ *$ In addition, the requirement for LSB is $50 \mu \mathrm{~V}$. */

$$
\begin{array}{ll}
\text { FS }=65535 \times 50 \mathrm{E}-6 ; & / * 3.27675 * / \\
\text { CNT1 }=0.5 / 50 \mathrm{E}-6 ; & / * 10000 * / \\
\text { CNT2 }=0.90 \times \text { FS } / 50 \mathrm{E}-6 ; & / * 58981.5 * /
\end{array}
$$

$l^{*}$ Thus the null input 0.5 V and the $90 \%$ of FS input is 2.949075V. */

Set the trim-offset-register to zero;
Set Right-Shift register to zero (typically zero. See Right-Shifting section);
gain_result = Oh;
Clamp $=$ FFF8h/2^(Right_Shift_Register);
For $\mathrm{n}=15$ down to 0
begin
gain_result = gain_result $+2^{\wedge}$ n;
Force the $90 \%$ FS input (2.949075V);
Meas2 $=$ read the digital result from the part;
If Meas2 >= Clamp then
gain_result = gain_result $-2^{\wedge} n$;
Else
Force the null input ( 0.5 V );
Meas1 = read the digital result from the part;
if (Meas2 - Meas1) > (CNT2 -
CNT1) then
gain_result = gain_result - $2^{\wedge}$ n;
end;
Set the gain register to gain_result;
The gain register is now set and the resolution of the conversion will best match the expected LSB. The next step is to calibrate the offset of the DS1859. With the correct gain value written to the gain register, again force the null input to the pin. Read the digital result from the part (Meas1). The offset value is equal to the negative value of Meas1.

$$
\text { Offset_Register }=\left[4000 \mathrm{~h}-\frac{\text { Meas1 }}{2}\right] \times \mathrm{OR}[4000 \mathrm{~h}]
$$

The calculated offset is now written to the DS1859 and the gain and offset scaling is now complete.

## Right-Shifting A/D Conversion Result (Scalable Dynamic Ranging)

The right-shifting method is used to regain some of the lost ADC range of a calibrated system. If a system is calibrated such that the maximum expected input results in a digital output value of less than 7FFFh (1/2 FS), then it is a candidate for using the right-shifting method.
If the maximum desired digital output is less than 7FFFh, then the calibrated system is using less than $1 / 2$ of the ADC's range. Similarly, if the maximum desired digital output is less than 1FFFh, then the calibrated system is only using $1 / 8$ of the ADC's range. For example, if using a zero for the right-shift during internal calibration and the maximum expected input results in a maximum digital output less than 1FFCh, only $1 / 8$ of the ADC's range is used. If left like this, the three MS bits of the ADC will never be used. In this example, a value of 3 for the rightshifting will maximize the ADC range. No resolution is lost since this is a 12-bit converter that is left justified. The value can be right-shifted four times without losing resolution. Table 9 shows when the right-shifting method can be used.

## Table 9. Right Shifting

| OUTPUT RANGE USED <br> WITH ZERO RIGHT-SHIFTS | NUMBER OF <br> RIGHT-SHIFTS NEEDED |
| :---: | :---: |
| Oh .. FFFFh | 0 |
| Oh .. 7FFFh | 1 |
| Oh .. 3FFFh | 2 |
| Oh .. 1FFFh | 3 |
| Oh .. OFFFh | 4 |

Memory Protection
Memory access from either device address can be either read/write or read only. Write protection is accomplished by a combination of control bits in EEPROM (APEN and MPEN in configuration register 89h) and a write-protect enable (WPEN) pin. Since the WPEN pin is often not accessible from outside the module, this scheme effectively allows the module to be locked by the manufacturer to prevent accidental writes by the end user.
Separate write protection is provided for the Auxiliary and Main Device address through distinct bits APEN and MPEN. APEN and MPEN are bits from configuration register 89h, Table 01. Due to the location, the APEN and MPEN bits can only be written through the

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

Main Device address. The control of write privileges through the Auxiliary Device address depends on the value of APEN. Care should be taken with the setting of MPEN, once set to a 1, assuming WPEN is high. Access through the Main Device is thereafter denied unless WPEN is taken to a low level. By this means, inadvertent end-user write access can be denied.
Main Device address space 60h to 7Fh is SRAM and is not write protected by APEN, MPEN, or WPEN. For example, the user may reset flags set by the device. Note that in single device mode (ADEN bit = 1), APEN determines the protection level of Table 00, independent of WPEN.
The write-protect operation, for both Main and Auxiliary Devices, is summarized in Tables 6 and 7.

## Temperature Conversion

The direct-to-digital temperature sensor measures temperature through the use of an on-chip temperature measurement technique with an operating range from $-40^{\circ} \mathrm{C}$ to $+102^{\circ} \mathrm{C}$. Temperature conversions are initiated upon power-up, and the most recent conversion is stored in memory locations 60h and 61h of the Main Device, which are updated every tframe. Temperature conversions do not occur during an active read or write to memory.
The value of each resistor is determined by the tempera-ture-addressed look-up table. The look-up table assigns a unique value to each resistor for every $2^{\circ} \mathrm{C}$ increment with a $1^{\circ} \mathrm{C}$ hysteresis at a temperature transition over the operating temperature range (see Figure 4).

Power-Up and Low-Voltage Operation During power-up, the device is inactive until VCC exceeds the digital power-on-reset voltage (POD). At this voltage, the digital circuitry, which includes the 2 -wire interface, becomes functional. However, EEPROMbacked registers/settings cannot be internally read (recalled into shadow SRAM) until VCC exceeds the analog power-on-reset voltage (POA), at which time the remainder of the device becomes fully functional. Once VCc exceeds POA, the RDYB bit in byte 6Eh of the Main Device memory is timed to go from a 1 to a 0 and indicates when analog-to-digital conversions begin. If VCC ever dips below POA, the RDYB bit reads as a 1 again. Once a device exceeds POA and the EEPROM is recalled, the values remain active (recalled) until VCc falls below POD.

For 2-wire device addresses sourced from EEPROM (ADFIX = 1), the device address defaults to A2h until VCC exceeds POA and the EEPROM values are recalled. The Auxiliary Device (AOh) is always available within this volt-
age window (between POD and the EEPROM recall) regardless of the programmed state of ADEN.
Furthermore, as the device powers up, the Vcclo alarm flag (bit 4 of 70 h in Main Device) defaults to a 1 until the first $V_{C C}$ analog-to-digital conversion occurs and sets or clears the flag accordingly.

## 2-Wire Operation

Clock and Data Transitions: The SDA pin is normally pulled high with an external resistor or device. Data on the SDA pin may only change during SCL-low time periods. Data changes during SCL-high periods will indicate a START or STOP condition depending on the conditions discussed below. See the timing diagrams in Figures 5 and 6 for further details.
START Condition: A high-to-low transition of SDA with SCL high is a START condition that must precede any other command. See the timing diagrams in Figures 5 and 6 for further details.
STOP Condition: A low-to-high transition of SDA with SCL high is a STOP condition. After a read or write sequence, the stop command places the DS1859 into a low-power mode. See the timing diagrams in Figures 5 and 6 for further details.
Acknowledge: All address and data bytes are transmitted through a serial protocol. The DS1859 pulls the SDA line low during the ninth clock pulse to acknowledge that it has received each word.
Standby Mode: The DS1859 features a low-power mode that is automatically enabled after power-on, after a STOP command, and after the completion of all internal operations.
Device Addressing: The DS1859 must receive an 8-bit device address following a START condition to enable a specific device for a read or write operation. The address is clocked into this part MSB to LSB. The address byte consists of either A2h or the value in Table 01 8Ch for the Main Device or A0h for the Auxiliary Device, then the R/W bit. This byte must match the address programmed into Table 01 8Ch or A0h (for the Auxiliary Device). If a device address match occurs, this part will output a zero for one clock cycle as an acknowledge and the corresponding block of memory is enabled (see the Memory Organization section). If the R/W bit is high, a read operation is initiated. If the R/W is low, a write operation is initiated (see the Memory Organization section). If the address does not match, this part returns to a low-power mode.

# Dual, Temperature-Controlled Resistors with Internally Calibrated Monitors 

## Write Operations

After receiving a matching address byte with the R/W bit set low, if there is no write protect, the device goes into the write mode of operation (see the Memory Organization section). The master must transmit an 8bit EEPROM memory address to the device to define the address where the data is to be written. After the byte has been received, the DS1859 transmits a zero for one clock cycle to acknowledge the address has been received. The master must then transmit an 8-bit data word to be written into this address. The DS1859 again transmits a zero for one clock cycle to acknowledge the receipt of the data. At this point, the master must terminate the write operation with a STOP condition. The DS1859 then enters an internally timed write process $t_{w}$ to the EEPROM memory. All inputs are disabled during this byte write cycle.

Page Write
The DS1859 is capable of an 8-byte page write. A page is any 8 -byte block of memory starting with an address evenly divisible by eight and ending with the starting address plus seven. For example, addresses 00h through 07h constitute one page. Other pages would be addresses 08h through 0Fh, 10h through 17h, 18h through 1Fh, etc.
A page write is initiated the same way as a byte write, but the master does not send a STOP condition after the first byte. Instead, after the slave acknowledges the data byte has been received, the master can send up to seven more bytes using the same nine-clock
sequence. The master must terminate the write cycle with a STOP condition or the data clocked into the DS1859 will not be latched into permanent memory.
The address counter rolls on a page during a write. The counter does not count through the entire address space as during a read. For example, if the starting address is 06h and 4 bytes are written, the first byte goes into address 06h. The second goes into address 07h. The third goes into address 00h (not 08h). The fourth goes into address 01h. If more than 9 bytes or more are written before a STOP condition is sent, the first bytes sent are overwritten. Only the last 8 bytes of data are written to the page.
Acknowledge Polling: Once the internally timed write has started and the DS1859 inputs are disabled, acknowledge polling can be initiated. The process involves transmitting a START condition followed by the device address. The R/W bit signifies the type of operation that is desired. The read or write sequence will only be allowed to proceed if the internal write cycle has completed and the DS1859 responds with a zero.

## Read Operations

After receiving a matching address byte with the R/W bit set high, the device goes into the read mode of operation. There are three read operations: current address read, random read, and sequential address read.

## Current Address Read

The DS1859 has an internal address register that maintains the address used during the last read or write


Figure 5. 2-Wire Data Transfer Protocol

