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FAN7387V Ballast Control IC for Compact Fluorescent Lamp

Features

- Integrated Half-Bridge MOSFET
- Internal Clock Using RCT

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- Enable External Sync Function Using RCT
- Dead-Time Control by using Resistor
- Shut Down (Disable Mode)
- Internal Shunt Regulator
- UVLO Function High and Low Side

Applications

Compact Fluorescent Lamp Ballast

Description

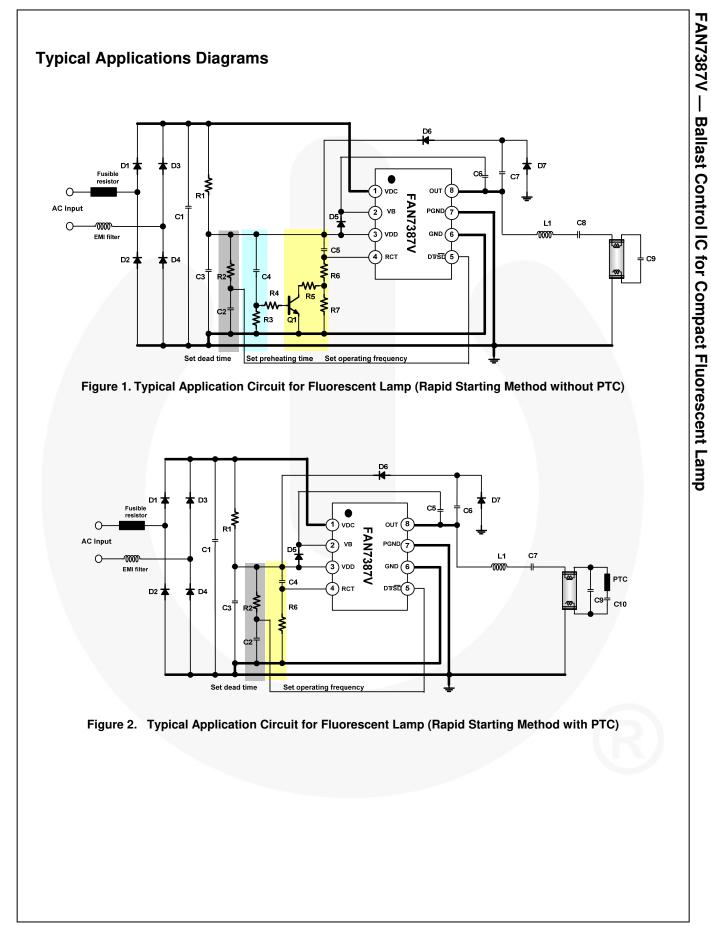
The FAN7387V, developed using Fairchild's unique highvoltage process and system-in-package (SiP) concept, is a ballast-control integrated circuit (IC) for a compact fluorescent lamp (CFL). The FAN7387V has a simple oscillating circuit using an external resistor and capacitor so the frequency variation is stable across the temperature range. FAN7387V has a external pin for dead time control and shutdown. By using this resistor, a designer can choose the optimum dead time to reduce the power loss on internal switching devices (MOSFETs).

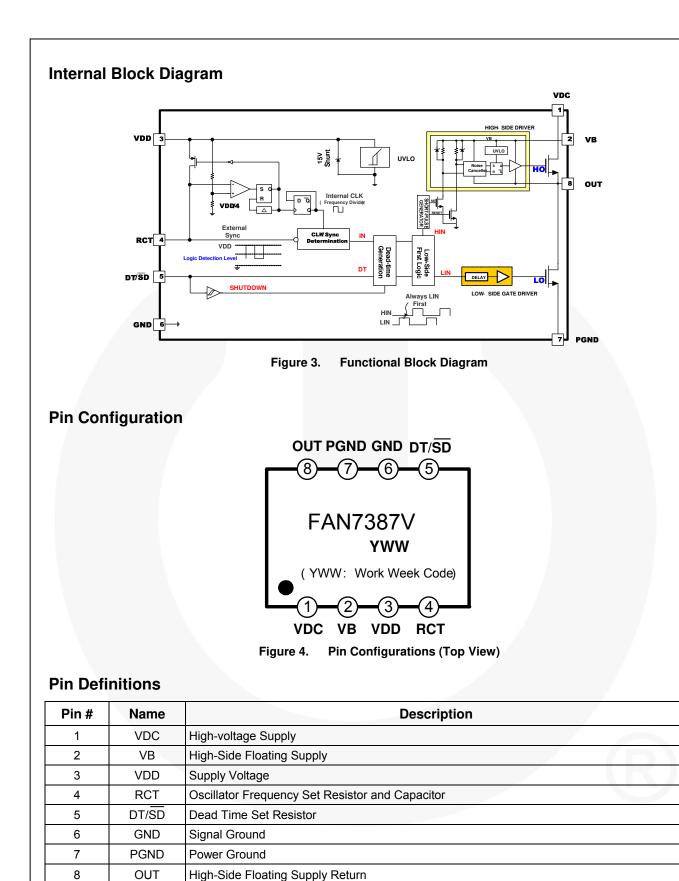
8-DIP



Ordering Information

Part Number	Operating Temperature	Package	Packing Method
FAN7387VN	-40 to +125°C	8-Lead, Dual-In-line Package (DIP)	Tube





FAN7387V — Ballast Control IC for Compact Fluorescent Lamp

Absolute Maximum Ratings

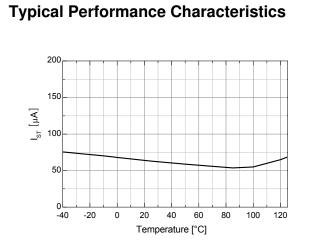
Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. $T_A=25^{\circ}C$ unless otherwise specified.

Symbol	Parameter	Min.	Тур.	Max.	Unit
VB	High-Side Floating Supply	-0.3		465.0	V
V _{OUT}	High-Side Floating Supply Return	-0.3		440.0	V
V _{RCT}	RCT Pins Input Voltage		V _{DD}		V
I _{CL}	Clamping Current Level ⁽¹⁾			25	mA
dV _{OUT} /dt	Allowable Offset Voltage Slew Rate		50		V/ns
T _A	Operating Temperature Range	-40		+125	°C
T _{STG}	Storage Temperature Range	-65		+150	°C
P _D	Power Dissipation		2.1		W
Θ_{JA}	Thermal Resistance (Junction-to-Air)		70		°C/W

Note:

1. Do not supply a low-impedance voltage source to the internal clamping Zener diode between the GND and the VDD pin of this device.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit	
High Vol	tage Supply Section						
V _{DC}	High Voltage Supply Voltage		440			V	
Low-Side	Supply Characteristics (V _{DD})			<u> </u>			
VDD _{UV+}	V _{DD} UVLO Positive-Going Threshold	V _{DD} Increasing	9	11	13		
VDD _{UV-}	V _{DD} UVLO Negative-Going Threshold	V _{DD} Decreasing	7.8	8.8	9.8		
VDD _{UHY}	V _{DD} -Side UVLO Hysteresis			2.2		V	
V _{CL}	Supply Camping Voltage	I _{DD} =10mA	14.4	15.4			
I _{ST}	Startup Supply Current	V _{DD} =9V		60	90	†	
I _{QDD}	Low-Side Quiescent Supply Current	R _{DT} =100kΩ		230	380	μA	
I _{DD}	Dynamic Operating Supply Current	20kHz, C _L =1nF	6	0.6		mA	
	e Supply Characteristics (VB-V _{OUT})			1 1			
V _{HSUV+}	High-Side UVLO Positive-Going Threshold	V _B -V _{OUT} Increasing	8	9	10		
V _{HSUV-}	High-Side UVLO Negative-Going Threshold	V _B -V _{OUT} Decreasing	7.5	8.5	9.5	V	
V _{HSUHY}	V _{BS} Supply UVLO Hysteresis			0.5			
I _{QHS}	High-Side Quiescent Supply Current			50	90		
I _{PBS}	High-Side Dynamic Operating Supply Current	20kHz, C _L =1nF		130	180	μA	
	r Characteristics						
f _{osc}	Oscillation Frequency	R _T =50kΩ, C _T =330pF	18	20	22	kHz	
D	Duty Cycle	Running Mode	47.5	49.0		%	
V _{RCT+}	Upper Threshold Voltage of RCT	Running Mode		V _{DD}			
V _{RCT-}	Lower Threshold Voltage of RCT	Running Mode		V _{DD} /4			
VIH	Logic "1" Input Voltage of RCT	Running Mode		3/4 V _{DD}		- V	
VIL	Logic "0" Input Voltage of RCT	Running Mode			3/5 V _{DD}		
t _D	Dead Time	R_{DT} =100k Ω	440	540	640		
t _{DMIN}	Minimum Dead Time	V _{DT/SD} =V _{DD}	280	400	520	ns	
Protectio	n Characteristics				1		
V _{SD+}	Shutdown "1" Input Voltage		2.5				
V _{SD-}	Shutdown "0" Input Voltage	V _{SD/DT} =0 After Run Mode			1	- V μΑ	
I _{SD}	Shutdown Current				350		
t _{SD}	Shutdown Propagation Delay			180		Ns	
Internal I	MOSFET Section	-			1		
I _{LKMOS}	Internal MOSFET Leakage Current	V _{DS} =400V		<u> </u>	50	μA	
R _{ON}	Static Drain-Source On-Resistance	V _{GS} =10V, I _D =190mA		4.6	6.0	Ω	
I _S	Maximum Continuous Drain-Source Diode Forward Current			0.38		A	
I _{SM}	Maximum Pulsed Continuous Drain-Source Diode Forward Current			3.04		Α	
V_{SD}	Drain-Source Diode Forward Voltage	V _{GS} =0V, I _S =0.38A		1	1.2	V	



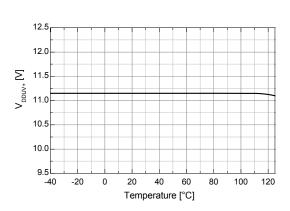


Figure 5. Startup Current vs. Temperature

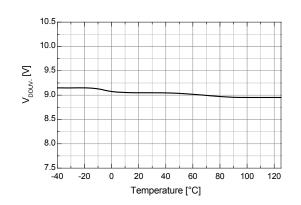


Figure 7. V_{DD} UVLO- vs. Temperature

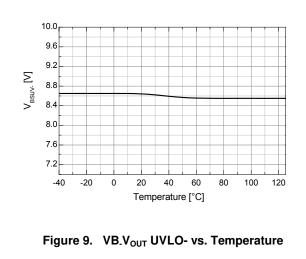


Figure 6. V_{DD} UVLO+ vs. Temperature

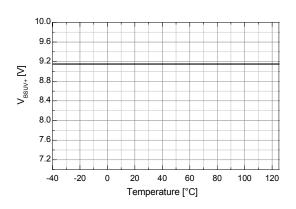
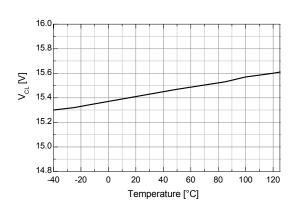
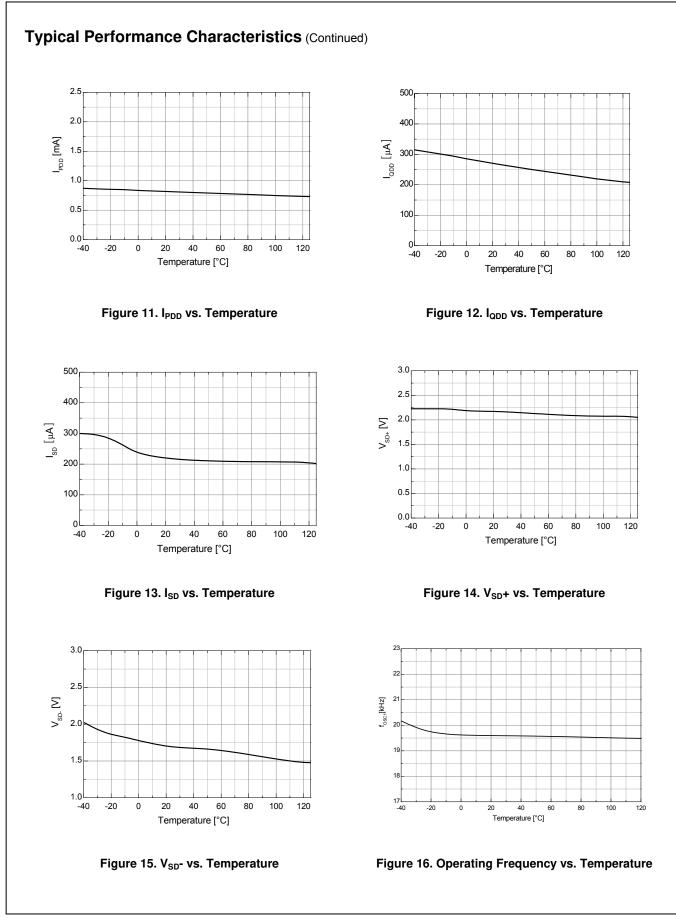
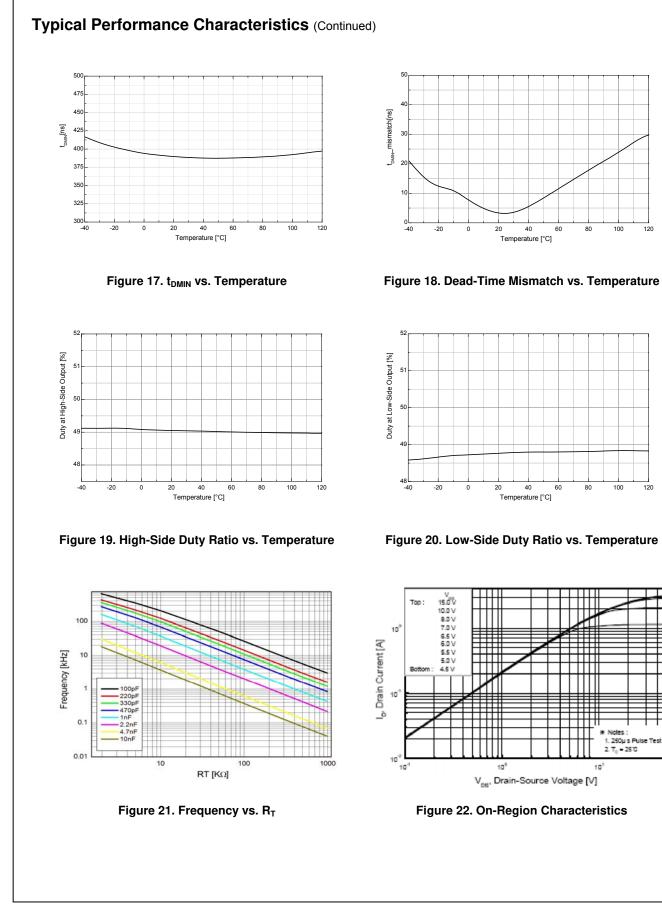


Figure 8. VB.V_{OUT} UVLO+ vs. Temperature











FAN7387V — Ballast Control IC for Compact Fluorescent Lamp

1.4

Notes

1.0

1. V_{cis} = 0V 2. 250u s Puíse Tr

Notes:

150

1. V₀₈ = 10 V 2. L = 0.19 A

200

1.2

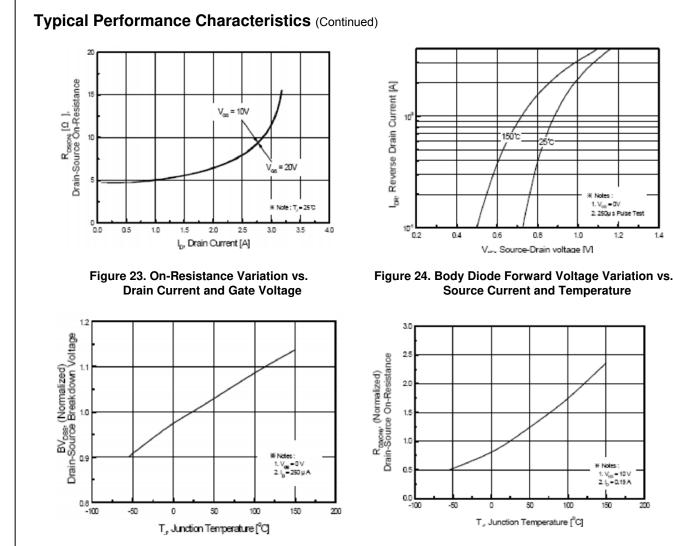


Figure 25. Breakdown Voltage Variation vs. Temperature

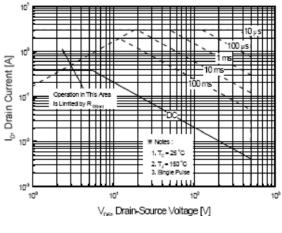
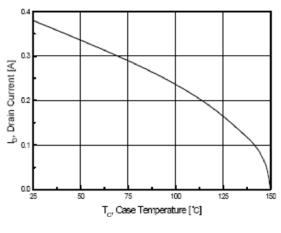
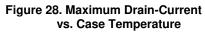


Figure 27. Maximum Safe Operating Area

Figure 26. On-Resistance Variation vs. Temperature





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Application Information

1. Under-Voltage Lockout (UVLO) Function

FAN7387V has a UVLO circuit for a low-side and highside block. When V_{DD} reaches to the VDD_{UV}+, the UVLO circuit is released and the FAN7387V operates normally. At UVLO condition, the FAN7387V has a low supply current of less than 130µA. Once UVLO is released, FAN7387V operates normally until V_{DD} goes below VDD_{UV}-, the UVLO hysteresis. FAN7387V also has a high-side gate driver. The supply for the high-side driver is applied between VB and V_{OUT} . To prevent malfunction at low supply voltage between VB and V_{OUT} , FAN7387V provides an additional UVLO circuit. If VB-V_{OUT} is under V_{HSUV} +, the driver holds LOW state to turn off the highside switch. Once the voltage of VB-V_{OUT} is higher than V_{HSUV+} , after VB-V_{OUT} exceeds V_{HSUV} -, the operation of driver resumes.

2. Oscillator

The running frequency is determined by an external timing resistor (R_T) and timing capacitor (C_T). The charge time of capacitor C_T from 1/4 V_{DD} to V_{DD} determines the running frequency of gate driver output (V_{OUT}).

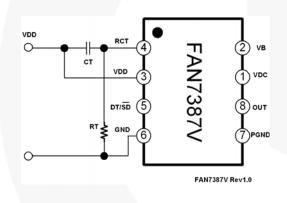


Figure 29. Typical Connection Method

Figure 30 shows the typical waveforms of RCT and internal signals (LO and HO) of IC. From the circuit analysis, the discharging time of RCT, t, is given by:

$$V_{RCT}(t) = V_{DD} \times In\left(\frac{-t}{R_{T} \bullet C_{T}}\right)$$
(1)

From Equation 1, it is possible to calculate the discharging time, t, from V_{DD} to one quarter (1/4) of V_{DD} by substituting $V_{RCT(t)}$ with 1/4 V_{DD} .

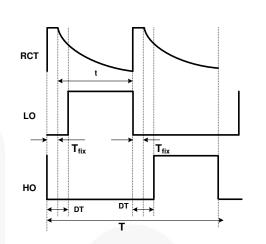


Figure 30. Typical Waveforms of RCT and Internal Signal (LO, HO) of IC

$$t = 1.38 \bullet R_T \bullet C_T \tag{2}$$

The running frequency of IC is determined by 1/t and is approximately given as:

$$f_{running} = \frac{1}{t} = \frac{1}{2(t + t_{fix})}$$
 (3)

where t is the discharging time of the RCT voltage and t_{fix} is constant value about 450ns of IC.

3. Programming Dead Time Control / Shutdown

A multi-function pin controls dead time using an external resistor (R_{DT}) and protects abnormal condition using an external switch. This pin should be connected to an external capacitor to maintain stable operation.

If the voltage of DT/SD is decreased to under 1V by an external switch, such as the TR or MOSFET, the FAN7387V enters shutdown mode. In this mode, the FAN7387V doesn't have any output signal.

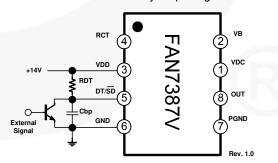
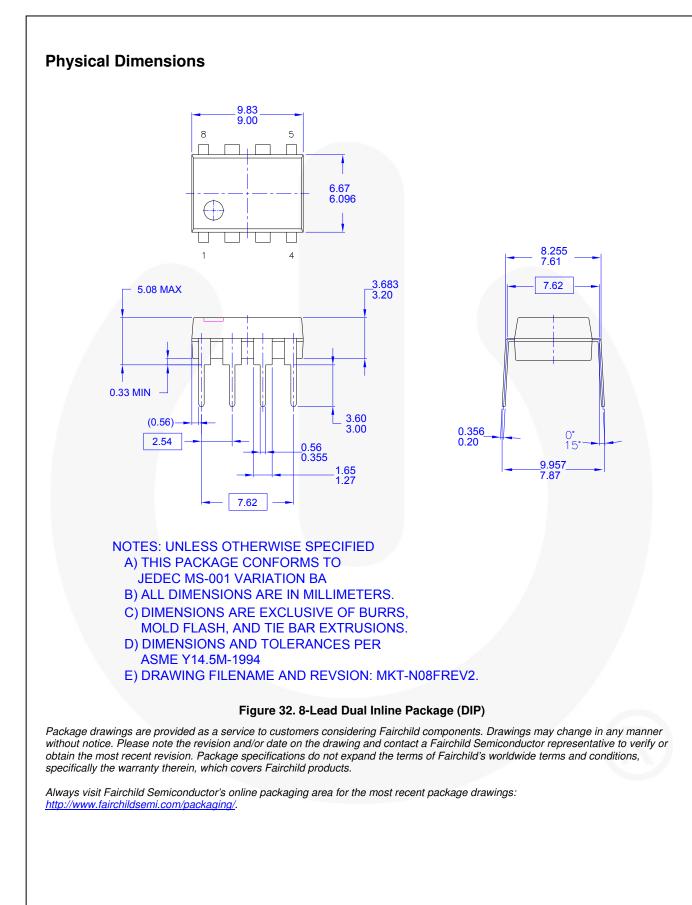


Figure 31. External Shutdown Circuit



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Rev. 161

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