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## FAIRCHILD

## FDC6324L

## Integrated Load Switch

## General Description

These Integrated Load Switches are produced using Fairchild＇s proprietary，high cell density，DMOS technology．This very high density process is especially tailored to minimize on－state resistance and provide superior switching performance．These devices are particularly suited for low voltage high side load switch application where low conduction loss and ease of driving are needed．

## Features

－ $\mathrm{V}_{\mathrm{DROP}}=0.2 \mathrm{~V} @ \mathrm{~V}_{\mathrm{IN}}=12 \mathrm{~V}, \mathrm{I}_{\mathrm{L}}=1 \mathrm{~A}, \mathrm{~V}_{\mathrm{ON} / \mathrm{FFF}}=1.5$ to 8 V $\mathrm{V}_{\text {DROP }}=0.3 \mathrm{~V} @ \mathrm{~V}_{\mathrm{IN}_{\mathrm{N}}}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{L}}=1 \mathrm{~A}, \mathrm{~V}_{\text {ON／OFF }}=1.5$ to 8 V ．
－High density cell design for extremely low on－resistance．
－ $\mathrm{V}_{\text {ONOFF }}$ Zener protection for ESD ruggedness．$>6 \mathrm{KV}$ Human Body Model．
－SuperSOT ${ }^{T M}$－ 6 package design using copper lead frame for superior thermal and electrical capabilities．


SOT－23


SuperSOT ${ }^{\text {TM }} \mathbf{6}$


EQUIVALENT CIRCUIT


Absolute Operating Range $T_{A}=25^{\circ} \mathrm{C}$ unless otherwise noted

| Symbol | Parameter | FDC6324L | Units |
| :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IN }}$ | Input Voltage Range | 3－20 | V |
| $\mathrm{V}_{\text {ON／OFF }}$ | ON／OFF Voltage Range | 1．5－8 | V |
| $\mathrm{I}_{\mathrm{L}}$ | Load Current＠$V_{\text {DROP }}=0.5 \mathrm{~V}$－Continuous <br> －Pulsed | 1.5 | A |
|  |  | 2.5 |  |
| $\mathrm{P}_{\mathrm{D}}$ | Maximum Power Dissipation（Note 2a） | 0.7 | W |
| $\mathrm{T}_{\mathrm{J}}, \mathrm{T}_{\text {STG }}$ | Operating and Storage Temperature Range | －55 to 150 | ${ }^{\circ} \mathrm{C}$ |
| ESD | Electrostatic Discharge Rating MIL－STD－883D Human Body Model（100pf／1500Ohm） | 6 | kV |
| THERMAL CHARACTERISTICS |  |  |  |
| $\mathrm{R}_{\text {өJA }}$ | Thermal Resistance，Junction－to－Ambient（Note 2a） | 180 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{R}_{\text {өJC }}$ | Thermal Resistance，Junction－to－Case（Note 2） | 60 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |



Typical Electrical Characteristics $\left(T_{A}=25^{\circ} \mathrm{C}\right.$ unless otherwise noted)


Figure 1. $\mathrm{V}_{\text {DROP }}$ Versus $\mathrm{I}_{\mathrm{L}}$ at $\mathrm{V}_{\mathrm{IN}}=\mathbf{1 2 V}$.


Figure 3. $\mathrm{V}_{\text {DROP }}$ Versus $\mathrm{V}_{\text {IN }}$ at $\mathrm{I}_{\mathrm{L}}=1 \mathrm{~A}$.



Figure 2. $\mathrm{V}_{\mathrm{DROP}}$ Versus $\mathrm{I}_{\mathrm{L}}$ at $\mathrm{V}_{\mathrm{IN}}=5.0 \mathrm{~V}$.


Figure 4. $\mathrm{R}_{(\mathrm{ON})}$ Versus $\mathrm{I}_{\mathrm{L}}$ at $\mathrm{V}_{\mathrm{IN}}=5.0 \mathrm{~V}$.

Figure 5. On Resistance Variation with Input Voltage.

Typical Electrical Characteristics $\left(T_{A}=25^{\circ} \mathrm{C}\right.$ unless otherwise noted )


Figure 6. Switching Variation with R2 at Vin $=12 \mathrm{~V}$ and $\mathrm{R} 1=300 \mathrm{KOhm}$.


Figure 8. Switching Variation with R2 at Vin=3.3V and R1=300KOhm.


Figure 10. Vdrop Variation with Vin and R2.


Figure 7. Switching Variation with R2 at Vin=5V and R1=300KOhm.


Figure 9. \% of Current Overshoot Variation with Vin and R2.


Figure 11. Switching Waveforms.

Typical Electrical Characteristics $\left(T_{A}=25^{\circ} \mathrm{C}\right.$ unless otherwise noted)


Figure 12. Safe Operating Area.


Figure 13. Transient Thermal Response Curve.
Note: Thermal characterization performed on the conditions described in Note 2a. Transient thermal response will change depends on the circuit board

## FDC6324L Load Switch Application



## General Description

This device is particularly suited for computer peripheral switching applications where 20 V input and 1A output current capability are needed. This load switch integrates a small N-Channel Power MOSFET (Q1) which drives a large P-Channel Power MOSFET (Q2) in one tiny SuperSOT ${ }^{\text {TM }}-6$ package.
A load switch is usually configured for high side switching so that the load can be isolated from the active power source. A P-Channel Power MOSFET, because it does not require its drive voltage above the input voltage, is usually more cost effective than using an N -Channel device in this particular application. A large P-Channel Power MOSFET minimizes voltage drop. By using a small N -Channel device the driving stage is simplified.

## Component Values

| R1 | Typical $10 \mathrm{k}-1 \mathrm{M} \Omega$ |  |
| :--- | :--- | :--- |
| R2 | Typical $0-10 \mathrm{k} \Omega$ | (optional) |
| C1 | Typical 1000 pF | (optional) |

## Design Notes

- R1 is needed to turn off Q2.
- R2 can be used to soft start the switch in the case the output capacitance Co is small.
- R2 $\leq$ should be at least 10 times smaller than R1 to guarantee Q1 turns on.
- By using R1 and R2 a certain amount of current is lost from the input. This bias current loss is given by the equation
$I_{B I A S \_L O S S}=\frac{V i n}{R 1+R 2} \quad$ when the switch is ON. I $I_{\text {BIAS_LOSS }}$ can be minimized by large R1.
- R2 and $\mathrm{C}_{\mathrm{RSS}}$ of Q 2 make ramp for slow turn on. If excessive overshoot current occurs due to fast turn on, additional capacitance C 1 can be added externally to slow down the turn on.


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| :---: | :---: | :---: |
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| CROSSVOLT ${ }^{\text {™ }}$ | РОРтм | VCX ${ }^{\text {™ }}$ |
| $\mathrm{E}^{2} \mathrm{CMOS}^{\text {™ }}$ | PowerTrench ${ }^{\text {™ }}$ |  |
| FACT ${ }^{\text {m }}$ | QFET ${ }^{\text {TM }}$ |  |
| FACT Quiet Series ${ }^{\text {™ }}$ | QS ${ }^{\text {TM }}$ |  |
| FAST ${ }^{\circledR}$ | Quiet Series ${ }^{\text {TM }}$ |  |
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