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# FDC6324L Integrated Load Switch

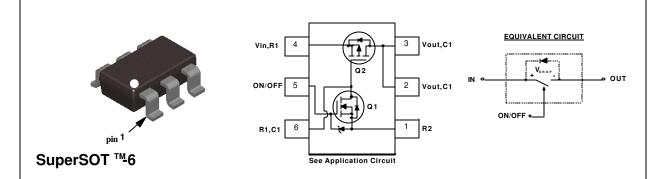
#### **General Description**

These Integrated Load Switches are produced using Fairchild's proprietary, high cell density, DMOS technology. This very high density process is especially tailored to minimize on-state resistance and provide superior switching performance. These devices are particularly suited for low voltage high side load switch application where low conduction loss and ease of driving are needed.

#### **Features**

- High density cell design for extremely low on-resistance.
- V<sub>ONOFF</sub> Zener protection for ESD ruggedness. >6KV Human Body Model.
- SuperSOT<sup>TM</sup>-6 package design using copper lead frame for superior thermal and electrical capabilities.





Absolute Operating Range T<sub>A</sub> = 25°C unless otherwise noted

Symbol	Parameter	FDC6324L	Units
V <sub>IN</sub>	Input Voltage Range	3 - 20	V
V <sub>ON/OFF</sub>	ON/OFF Voltage Range	1.5 - 8	V
I <sub>L</sub>	Load Current @ V <sub>DROP</sub> =0.5V - Continuous (Note 1)	1.5	A
	- Pulsed (Note 1 & 3)	2.5	
P <sub>D</sub>	Maximum Power Dissipation (Note 2a)	0.7	W
$T_J, T_{STG}$	Operating and Storage Temperature Range	-55 to 150	°C
ESD	Electrostatic Discharge Rating MIL-STD-883D Human Body Model (100pf/1500Ohm)	6	kV
THERMA	L CHARACTERISTICS		•
$R_{\theta JA}$	Thermal Resistance, Junction-to-Ambient (Note 2a)	180	°C/W
$R_{\theta JC}$	Thermal Resistance, Junction-to-Case (Note 2)	60	°C/W

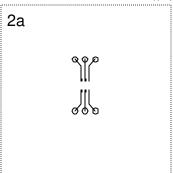
Electrical Characteristics (T <sub>A</sub> = 25°C unless otherwise noted)									
Symbol	Parameter	Conditions	Min	Тур	Max	Units			
OFF CHAI	RACTERISTICS								
I <sub>FL</sub>	Forward Leakage Current	V <sub>IN</sub> = 20 V, V <sub>ON/OFF</sub> = 0 V			1	μΑ			
I <sub>RL</sub>	Reverse Leakage Current	V <sub>IN</sub> = -20 V, V <sub>ON/OFF</sub> = 0 V			-1	μΑ			
ON CHAR	ACTERISTICS (Note 3)								
V <sub>IN</sub>	Input Voltage		3		20	V			
V <sub>ON/OFF</sub>	On/Off Voltage		1.5		8	V			
V <sub>DROP</sub>	Conduction Voltage Drop @ 1A	V <sub>IN</sub> = 10 V, V <sub>ON/OFF</sub> = 3.3V		0.135	0.2	V			
		V <sub>IN</sub> = 5 V, V <sub>ON/OFF</sub> = 3.3 V		0.215	0.3				
IL	Load Current	$V_{DROP} = 0.2 \text{ V}, V_{IN} = 10 \text{ V}, V_{ON/OFF} = 3.3 \text{ V}$	1			Α			
		V <sub>DROP</sub> = 0.3 V, V <sub>IN</sub> = 5 V, V <sub>ON/OFF</sub> = 3.3 V	1						

- 1.  $V_{IN}$ =20V,  $V_{ONOFF}$ =8V,  $V_{DROP}$ =0.5V,  $T_A$ =25°C
- 2.  $R_{\text{BJA}}$  is the sum of the junction-to-case and case-to-ambient thermal resistance where the case thermal reference is defined as the solder mounting surface of the drain pins.  $R_{\text{BJC}}$  is guaranteed by design while  $\mathbf{R}_{\text{\tiny BCA}}$  is determined by the user's board design.

$$P_D(t) = \frac{T_T T_A}{R_{B,t}(t)} = \frac{T_{t-} T_A}{R_{B,t} + R_{B,t}(t)} = I_D^2(t) \times R_{DS(ON)@T}$$

 $P_D(t) = \frac{T_T T_A}{R_{0,J} k l} = \frac{T_T - T_A}{R_{0,J} k R_{0,C}(l)} = I_D^2(t) \times R_{DQON)@T_J}$ Typical  $R_{0,JA}$  for single device operation using the board layouts shown below on FR-4 PCB in astill air environment

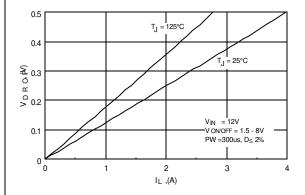
a. 180°C/W when mounted on a 2oz minimum copper pad.



Scale 1 : 1 on letter size paper

3. Pulse Test: Pulse Width ≤ 300µs, Duty Cycle≤ 2.0%

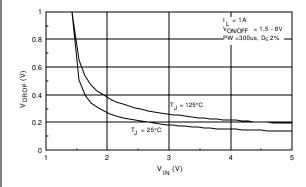
# Typical Electrical Characteristics (T $_{\!\scriptscriptstyle A}$ = $\,$ 25 $\,^{\rm o}\text{C}$ unless otherwise noted )



0.5
0.4
0.4
0.5
0.7  $J = 125^{\circ}C$   $T_{J} = 25^{\circ}C$   $T_{J} = 25^{\circ$ 

Figure 1.  $V_{DROP}$  Versus  $I_L$  at  $V_{IN}$ =12V.





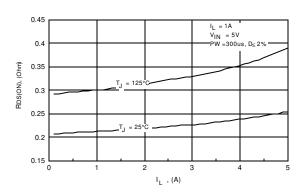


Figure 3.  $V_{\text{DROP}}$  Versus  $V_{\text{IN}}$  at  $I_{\text{L}}$ =1A.

Figure 4.  $R_{(ON)}$  Versus  $I_L$  at  $V_{IN}$ =5.0V.

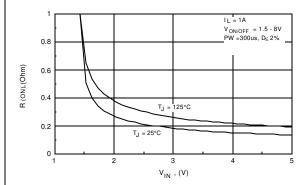
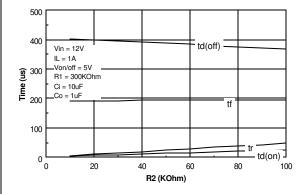


Figure 5. On Resistance Variation with Input Voltage.

# Typical Electrical Characteristics ( $T_A = 25$ °C unless otherwise noted )



500

Vin = 5V
IL = 1A

Vonvoff = 5V
R1 = 300KOhm
Ci = 10uF
Co = 1uF

100

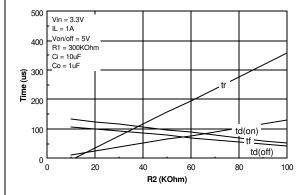
100

20
40
60
80
100

R2 (KOhm)

Figure 6. Switching Variation with R2 at Vin= 12V and R1= 300KOhm.

Figure 7. Switching Variation with R2 at Vin=5V and R1=300KOhm.



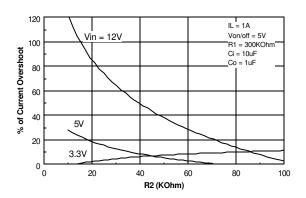
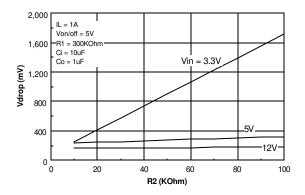


Figure 8. Switching Variation with R2 at Vin=3.3V and R1=300KOhm.

Figure 9. % of Current Overshoot Variation with Vin and R2.



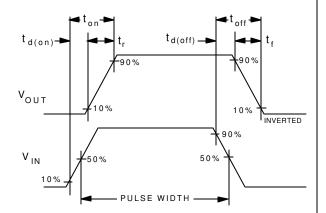


Figure 10. Vdrop Variation with Vin and R2.

Figure 11. Switching Waveforms.

# Typical Electrical Characteristics ( $T_A = 25$ °C unless otherwise noted )

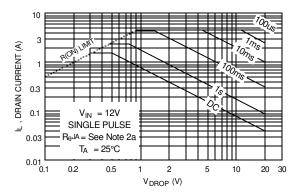


Figure 12. Safe Operating Area.

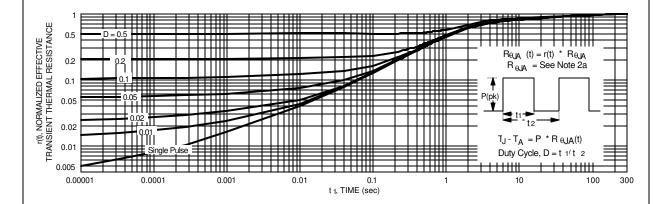
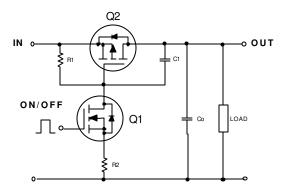


Figure 13. Transient Thermal Response Curve.

Note: Thermal characterization performed on the conditions described in Note 2a. Transient thermal response will change depends on the circuit board

## FDC6324L Load Switch Application

#### **APPLICATION CIRCUIT**



#### **General Description**

This device is particularly suited for computer peripheral switching applications where 20V input and 1A output current capability are needed. This load switch integrates a small N-Channel Power MOSFET (Q1) which drives a large P-Channel Power MOSFET (Q2) in one tiny SuperSOT<sup>TM</sup>-6 package.

A load switchis usually configured for high side switching so that the load can be isolated from the active power source. A P-Channel Power MOSFET, because it does not require its drive voltage above the input voltage, is usually more cost effective than using an N-Channel device in this particular application. A large P-Channel Power MOSFET minimizes voltage drop. By using a small N-Channel device the driving stage is simplified.

### **Component Values**

R1 Typical  $10k - 1M\Omega$ 

 $\begin{array}{lll} R2 & Typical & 0 - 10k\Omega & \text{(optional)} \\ C1 & Typical & 1000pF & \text{(optional)} \end{array}$ 

#### **Design Notes**

- R1 is needed to turn off Q2.
- R2 can be used to soft start the switch in the case the output capacitance Co is small.
- $R2 \le$  should be at least 10 times smaller than R1 to guarantee Q1 turns on.
- By using R1 and R2 a certain amount of current is lost from the input. This bias current loss is given by the equation

 $I_{BIAS\_LOSS} = \frac{Vin}{R1 + R2}$  when the switch is ON.  $I_{BIAS\_LOSS}$  can be minimized by large R1.

• R2 and C<sub>RSS</sub> of Q2 make ramp for slow turn on. If excessive overshoot current occurs due to fast turn on, additional capacitance C1 can be added externally to slow down the turn on.

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 $\begin{array}{lll} \mathsf{FAST}^{\circledast} & \mathsf{Quiet\,Series^{TM}} \\ \mathsf{FASTr^{TM}} & \mathsf{SuperSOT^{TM}\text{-}3} \\ \mathsf{GTO^{TM}} & \mathsf{SuperSOT^{TM}\text{-}6} \\ \mathsf{HiSeC^{TM}} & \mathsf{SuperSOT^{TM}\text{-}8} \end{array}$ 

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