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Features
■ 64-Kbit/256-Kbit ferroelectric random access memory (F-RAM) a Logically organized as $8 \mathrm{~K} \times 8$ (FM31276) $/ 32 \mathrm{~K} \times 8$ (FM31278)

- High-endurance 100 trillion ( $10^{14}$ ) read/writes
a 151-year data retention (See the Data Retention and Endurance table)
$\square$ NoDelay ${ }^{\text {TM }}$ writes
$\square$ Advanced high-reliability ferroelectric process
■ High Integration Device Replaces Multiple Parts
a Serial nonvolatile memory
a Real time clock (RTC)
口 Low voltage reset
a Watchdog timer
a Early power-fail warning/NMI
a Two 16-bit event counter
a Serial number with write-lock for security
- Real-time Clock/Calendar
a Backup current at $2 \mathrm{~V}: 1.15 \mu \mathrm{~A}$ at $+25^{\circ} \mathrm{C}$
$\square$ Seconds through centuries in BCD format
- Tracks leap years through 2099
- Uses standard 32.768 kHz crystal ( $6 \mathrm{pF} / 12.5 \mathrm{pF}$ )
■ Software calibration
a Supports battery or capacitor backup
- Processor Companion
$\square$ Active-low reset output for $V_{D D}$ and watchdog
a Programmable low- $V_{D D}$ reset trip point
a Manual reset filtered and debounced
$\square$ Programmable watchdog timer
- Dual Battery-backed event counter tracks system intrusions or other events
$\square$ Comparator for power-fail interrupt
a 64-bit programmable serial number with lock
- Fast 2-wire serial interface ( $\mathrm{I}^{2} \mathrm{C}$ )
- Up to $1-\mathrm{MHz}$ frequency
$\square$ Supports legacy timings for 100 kHz and 400 kHz
a RTC, Supervisor controlled via $I^{2} C$ interface
$\square$ Device select pins for up to 4 memory devices
■ Low power consumption
a 1.5 mA active current at 1 MHz
व $150 \mu \mathrm{~A}$ standby current
■ Operating voltage: $\mathrm{V}_{\mathrm{DD}}=4.0 \mathrm{~V}$ to 5.5 V
■ Industrial temperature: $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
■ 14-pin small outline integrated circuit (SOIC) package

■ Restriction of hazardous substances (RoHS) compliant
■ Underwriters laboratory (UL) recognized

## Functional Overview

The FM31276/FM31278 device integrates F-RAM memory with the most commonly needed functions for processor-based systems. Major features include nonvolatile memory, real time clock, low- $V_{D D}$ reset, watchdog timer, nonvolatile event counter, lockable 64-bit serial number area, and general purpose comparator that can be used for a power-fail (NMI) interrupt or any other purpose.
The FM31276/FM31278 is a 64-Kbit/256-Kbit nonvolatile memory employing an advanced ferroelectric process. A ferroelectric random access memory or F-RAM is nonvolatile and performs reads and writes similar to a RAM. This memory is truly nonvolatile rather than battery backed. It provides reliable data retention for 151 years while eliminating the complexities, overhead, and system-level reliability problems caused by other nonvolatile memories. The FM31276/FM31278 is capable of supporting $10^{14}$ read/write cycles, or 100 million times more write cycles than EEPROM.
The real time clock (RTC) provides time and date information in BCD format. It can be permanently powered from an external backup voltage source, either a battery or a capacitor. The timekeeper uses a common external 32.768 kHz crystal and provides a calibration mode that allows software adjustment of timekeeping accuracy.
The processor companion includes commonly needed CPU support functions. Supervisory functions include a reset output signal controlled by either a low $\mathrm{V}_{\mathrm{DD}}$ condition or a watchdog timeout. $\overline{R S T}$ goes active when $V_{D D}$ drops below a programmable threshold and remains active for 100 ms after $\mathrm{V}_{\mathrm{DD}}$ rises above the trip point. A programmable watchdog timer runs from 100 ms to 3 seconds. The watchdog timer is optional, but if enabled it will assert the reset signal for 100 ms if not restarted by the host before the timeout. A flag-bit indicates the source of the reset.
A comparator on PFI compares an external input pin to the onboard 1.2 V reference. This is useful for generating a power-fail interrupt (NMI) but can be used for any purpose. The family also includes a programmable 64-bit serial number that can be locked making it unalterable. Additionally it offers a dual battery-backed event counter that tracks the number of rising or falling edges detected on a dedicated input pin.
For a complete list of related documentation, click here.

## Logic Block Diagram



FM31276/FM31278

## Contents

Pinout ..... 4
Pin Definitions ..... 4
Overview ..... 5
Memory Architecture ..... 5
Processor Companion ..... 5
Processor Supervisor ..... 5
Manual Reset ..... 6
Reset Flags ..... 6
Early Power Fail Comparator ..... 6
Event Counter ..... 7
Serial Number ..... 7
Real-time Clock Operation ..... 7
Backup Power ..... 8
Trickle Charger ..... 8
Calibration ..... 9
Crystal Oscillator ..... 9
Layout Recommendations ..... 10
Register Map ..... 13
I2C Interface ..... 19
STOP Condition (P) ..... 19
START Condition (S) ..... 19
Data/Address Transfer ..... 19
Acknowledge / No-acknowledge ..... 19
Slave Address ..... 20
Addressing Overview - Memory ..... 20
Addressing Overview - RTC \& Companion ..... 20
Data Transfer ..... 20
Memory Operation ..... 21
Memory Write Operation ..... 21
Memory Read Operation ..... 21
RTC/Companion Write Operation ..... 23
RTC/Companion Read Operation ..... 23
Addressing FRAM Array in the FM31276/FM31278Family.23
Maximum Ratings ..... 24
Operating Range ..... 24
DC Electrical Characteristics ..... 24
Data Retention and Endurance ..... 26
Capacitance ..... 26
Thermal Resistance ..... 26
AC Test Loads and Waveforms ..... 26
AC Test Conditions ..... 26
Supervisor Timing ..... 27
AC Switching Characteristics ..... 28
Ordering Information ..... 29
Ordering Code Definitions ..... 29
Package Diagram ..... 30
Acronyms ..... 31
Document Conventions ..... 31
Units of Measure ..... 31
Document History Page ..... 32
Sales, Solutions, and Legal Information ..... 33
Worldwide Sales and Design Support ..... 33
Products ..... 33
PSoC® Solutions ..... 33
Cypress Developer Community ..... 33
Technical Support ..... 33

## Pinout

Figure 1. 14-pin SOIC pinout


## Pin Definitions

| Pin Name | I/O Type | Description |
| :---: | :---: | :---: |
| A1-A0 | Input | Device Select Address 1-0. These pins are used to select one of up to 4 devices of the same type on the same $\mathrm{I}^{2} \mathrm{C}$ bus. To select the device, the address value on the three pins must match the corresponding bits contained in the slave address. The address pins are pulled down internally. |
| SDA | Input/Output | Serial Data/Address. This is a bi-directional pin for the $\mathrm{I}^{2} \mathrm{C}$ interface. It is open-drain and is intended to be wire-OR'd with other devices on the $\mathrm{I}^{2} \mathrm{C}$ bus. The input buffer incorporates a Schmitt trigger for noise immunity and the output driver includes slope control for falling edges. An external pull-up resistor is required. |
| SCL | Input | Serial Clock. The serial clock pin for the $I^{2} \mathrm{C}$ interface. Data is clocked out of the device on the falling edge, and into the device on the rising edge. The SCL input also incorporates a Schmitt trigger input for noise immunity. |
| CNT1, CNT2 | Input | Event Counter Inputs. These battery-backed inputs increment counters when an edge is detected on the corresponding CNT pin. The polarity is programmable. These pins should not be left floating. Tie to ground if these pins are not used. |
| X1, X2 | Input/Output | 32.768 kHz crystal connection. When using an external oscillator, apply the clock to X1 and a DC mid-level to X2. These pins should be left unconnected if RTC is not used. |
| $\overline{\mathrm{RST}}$ | Input/Output | Reset. This active-low output is open drain with weak pull-up. It is also an input when used as a manual reset. This pin should be left floating if unused. |
| PFI | Input | Early Power-fail Input. Typically connected to an unregulated power supply to detect an early power failure. This pin must be tied to ground if unused. |
| CAL/PFO | Output | Calibration/Early Power-fail Output. In calibration mode, this pin supplies a 512 Hz square-wave output for clock calibration. In normal operation, this is the early power-fail output. |
| $\mathrm{V}_{\text {BAK }}$ | Power supply | Backup supply voltage. Connected to a 3 V battery or a large value capacitor. If no backup supply is used, this pin should be tied to ground and the VBC bit should be cleared in the RTC register OBh. The trickle charger is UL recognized and ensures no excessive current when using a lithium battery. |
| $\mathrm{V}_{\text {SS }}$ | Power supply | Ground for the device. Must be connected to the ground of the system. |
| $\mathrm{V}_{\mathrm{DD}}$ | Power supply | Power supply input to the device. |

## Overview

The FM31276/FM31278 device combines a serial nonvolatile RAM with a real time clock (RTC) and a processor companion. The companion is a highly integrated peripheral including a processor supervisor, a comparator used for early power-fail warning, nonvolatile event counters, and a 64-bit serial number. The FM31276/FM31278 integrates these complementary but distinct functions under a common interface in a single package. The product is organized as two logical devices. The first is a memory and the second is the companion which includes all the remaining functions. From the system perspective they appear to be two separate devices with unique IDs on the serial bus.
The memory is organized as a standalone nonvolatile $\mathrm{I}^{2} \mathrm{C}$ memory using standard device ID value. The real time clock and supervisor functions are accessed with a separate $I^{2} C$ device ID. This allows clock/calendar data to be read while maintaining the most recently used memory address. The clock and supervisor functions are controlled by 25 special function registers. The RTC and event counter circuits are maintained by the power source on the $\mathrm{V}_{\text {BAK }}$ pin, allowing them to operate from battery or backup capacitor power when $\mathrm{V}_{\mathrm{DD}}$ drops below a set threshold. Each functional block is described below.

## Memory Architecture

The FM31276/FM31278 device is available in memory size $64-\mathrm{Kbit} / 256-\mathrm{Kbit}$. The device uses two-byte addressing for the memory portion of the chip. This makes the device software compatible with its standalone memory counterparts, but makes them compatible within the entire family.
The memory array is logically organized as $8,192 \times 8$ bits / $32,768 \times 8$ bits and is accessed using an industry-standard $\mathrm{I}^{2} \mathrm{C}$ interface. The memory is based on F-RAM technology. Therefore it can be treated as RAM and is read or written at the speed of the $\mathrm{I}^{2} \mathrm{C}$ bus with no delays for write operations. It also offers effectively unlimited write endurance unlike other nonvolatile memory technologies. The $\mathrm{I}^{2} \mathrm{C}$ protocol is described on page 19 .

The memory array can be write-protected by software. Two bits in the processor companion area (WP1, WPO in register OBh) control the protection setting. Based on the setting, the protected addresses cannot be written and the $\mathrm{I}^{2} \mathrm{C}$ interface will not acknowledge any data to protected addresses. The special function registers containing these bits are described in detail below.
Table 1. Block Memory Write Protection

| WP1 | WP0 | Protected Address Range |
| :---: | :---: | :---: |
| 0 | 0 | None |
| 0 | 1 | Bottom $1 / 4$ |
| 1 | 0 | Bottom $1 / 2$ |
| 1 | 1 | Full array |

## Processor Companion

In addition to nonvolatile RAM, the FM31276/FM31278 incorporates a real time clock and highly integrated processor companion. The companion includes a low- $\mathrm{V}_{\mathrm{DD}}$ reset, a programmable watchdog timer, a battery-backed event counters, a comparator for early power-fail detection or other purposes, and a 64-bit serial number.

## Processor Supervisor

Supervisors provide a host processor two basic functions: detection of power supply fault conditions and a watchdog timer to escape a software lockup condition. The FM31276/FM31278 has a reset pin ( $\overline{\mathrm{RST}}$ ) to drive a processor reset input during power faults, power-up, and software lockups. It is an open drain output with a weak internal pull-up to $\mathrm{V}_{\mathrm{DD}}$. This allows other reset sources to be wire-OR'd to the RST pin. When $V_{D D}$ is above the programmed trip point, $\overline{\mathrm{RST}}$ output is pulled weakly to $\mathrm{V}_{\mathrm{DD}}$. If $\mathrm{V}_{\mathrm{DD}}$ drops below the reset trip point voltage level ( $\mathrm{V}_{\mathrm{TP}}$ ), the RST pin will be driven LOW. It will remain LOW until $V_{D D}$ falls too low for circuit operation which is the $\mathrm{V}_{\mathrm{RST}}$ level. When $\mathrm{V}_{\mathrm{DD}}$ rises again above $\mathrm{V}_{\mathrm{TP}}, \overline{\mathrm{RST}}$ continues to drive LOW for at least 100 $\mathrm{ms}\left(\mathrm{t}_{\mathrm{RPU}}\right)$ to ensure a robust system reset at a reliable $\mathrm{V}_{\mathrm{DD}}$ level. After $t_{\text {RPU }}$ has been met, the RST pin will return to the weak HIGH state. While RST is asserted, serial bus activity is locked out even if a transaction occurred as $\mathrm{V}_{\mathrm{DD}}$ dropped below $\mathrm{V}_{T P}$. $A$ memory operation started while $\mathrm{V}_{\mathrm{DD}}$ is above $\mathrm{V}_{\mathrm{TP}}$ will be completed internally.
Table 1 below shows how bit VTP controls the trip point of the low- $V_{D D}$ reset. They are located in register 0Bh, bits 1 and 0 . The reset pin will drive LOW when $V_{D D}$ is below the selected $V_{T P}$ voltage, and the $I^{2} \mathrm{C}$ interface and F-RAM array will be locked out. Note that the bit 1 location is a don't care. Figure 2 illustrates the reset operation in response to a low $\mathrm{V}_{\mathrm{DD}}$.
Table 2. VTP setting

| VTP Setting | VTP |
| :---: | :---: |
| 3.9 V | 0 |
| 4.4 V | 1 |

Figure 2. Low $\mathrm{V}_{\mathrm{DD}}$ Reset


A watchdog timer can also be used to drive an active reset signal. The watchdog is a free-running programmable timer. The timeout period can be software programmed from 100 ms to 3 seconds in 100 ms increments via a 5-bit nonvolatile register. All programmed settings are minimum values and vary with
temperature according to the operating specifications. The watchdog has two additional controls associated with its operation, a watchdog enable bit (WDE) and timer restart bits (WR). Both the enable bit must be set and the watchdog must timeout in order to drive RST active. If a reset event occurs, the timer will automatically restart on the rising edge of the reset pulse. If WDE = '0', the watchdog timer runs but a watchdog fault will not cause $\overline{\text { RST }}$ to be asserted LOW. The WTR flag will be set, indicating a watchdog fault. This setting is useful during software development if the developer does not want RST to drive. Note that setting the maximum timeout setting (11111b) disables the counter to save power. The second control is a nibble that restarts the timer preventing a reset. The timer should be restarted after changing the timeout value.
The watchdog timeout value is located in register 0Ah, bits 4:0, and the watchdog enable is bit 7 . The watchdog is restarted by writing the pattern 1010b to the lower nibble of register 09h. Writing this pattern will also cause the timer to load new timeout values. Writing other patterns to this address will not affect its operation. Note the watchdog timer is free-running. Prior to enabling it, users should restart the timer as described above. This assures that the full timeout period will be set immediately after enabling. The watchdog is disabled when $\mathrm{V}_{\mathrm{DD}}$ is below $\mathrm{V}_{T \mathrm{P}}$. The following table summarizes the watchdog bits. A block diagram follows.

| Watchdog Timeout | WDT(4:0) | OAh, bits 4:0 |
| :--- | :--- | :--- |
| Watchdog Enable | WDE | 0Ah, bit 7 |
| Watchdog Restart | WR(3:0) | 09h, bits 3:0 |

Watchdog Restart

Figure 3. Watchdog Timer


## Manual Reset

The $\overline{\text { RST }}$ is a bi-directional signal allowing the FM31276/FM31278 to filter and de-bounce a manual reset
switch. The RST input detects an external low condition and responds by driving the $\overline{\mathrm{RST}}$ signal LOW for 100 ms .
Figure 4. Manual Reset


Note The internal weak pull-up eliminates the need for additional external components.

## Reset Flags

In case of a reset condition, a flag bit will be set to indicate the source of the reset. A low- $V_{D D}$ reset is indicated by the POR flag, register 09h bit 6 . A watchdog reset is indicated by the WTR flag, register 09h bit 7. Note that the flags are internally set in response to reset sources, but they must be cleared by the user. When the register is read, it is possible that both flags are set if both have occurred since the user last cleared them.

## Early Power Fail Comparator

An early power fail warning can be provided to the processor well before $\mathrm{V}_{\mathrm{DD}}$ drops out of spec. The comparator is used to create a power fail interrupt (NMI). This can be accomplished by connecting the PFI pin to the unregulated power supply via a resistor divider. An application circuit is shown below.

Figure 5. Comparator as a Power-Fail Warning


The voltage on the PFI input pin is compared to an onboard 1.2 V reference. When the PFI input voltage drops below this threshold, the comparator will drive the CAL/PFO pin to a ILOW state. The comparator has 100 mV (max) of hysteresis to reduce noise sensitivity, only for a rising PFI signal. For a falling PFI edge, there is no hysteresis.

The comparator is a general purpose device and its application is not limited to the NMI function.

The comparator is not integrated into the special function registers except as it shares its output pin with the CAL output. When the RTC calibration mode is invoked by setting the CAL bit (register 00h, bit 2), the CAL/PFO output pin will be driven with a 512 Hz square wave and the comparator will be ignored. Since most users only invoke the calibration mode during production, this should have no impact on system operations using the comparator.

Note The maximum voltage on the comparator input PFI is limited to 3.75 V under normal operating conditions.

## Event Counter

The FM31276/FM31278 offers the user two battery-backed event counters. Input pins CNT1 and CNT2 are programmable edge detectors. Each clocks a 16-bit counter. When an edge occurs, the counters will increment their respective registers. Counter 1 is located in registers 0 Dh and 0Eh, Counter 2 is located in registers 0Fh and 10h. These register values can be read anytime $\mathrm{V}_{\mathrm{DD}}$ is above $\mathrm{V}_{\mathrm{TP}}$, and they will be incremented as long as a valid $\mathrm{V}_{\text {BAK }}$ power source is provided. To read, set the RC bit, register OCh bit 3 to 1 . This takes a snapshot of all four counter bytes allowing a stable value even if a count occurs during the read. The registers can be written by software allowing the counters to be cleared or initialized by the system. Counts are blocked during a write operation. The two counters can be cascaded to create a single 32-bit counter by setting the CC control bit (register 0Ch, bit 2). When cascaded, the CNT1 input will cause the counter to increment. CNT2 is not used in this mode and should be tied to ground.

Figure 6. Event Counter


The control bits for event counting are located in register 0Ch. Counter 1 Polarity is bit C1P, bit 0 ; Counter 2 Polarity is C2P, bit 1 ; the Cascade Control is CC, bit 2; and the Read Counter bit is RC, bit 3.
The polarity bits must be set prior to setting the counter value(s). If a polarity bit is changed, the counter may inadvertently
increment. If the counter pins are not being used, tie them to ground.

## Serial Number

A memory location to write a 64-bit serial number is provided. It is a writeable nonvolatile memory block that can be locked by the user once the serial number is set. The 8 bytes of data and the lock bit are all accessed via the device ID for the Processor Companion. Therefore the serial number area is separate and distinct from the memory array. The serial number registers can be written an unlimited number of times, so these locations are general purpose memory. However, once the lock bit is set, the values cannot be altered and the lock cannot be removed. Once locked the serial number registers can still be read by the system.
The serial number is located in registers 11 h to 18 h . The lock bit is SNL (register 0Bh, bit 7). Setting the SNL bit to a '1' disables writes to the serial number registers, and the SNL bit cannot be cleared.

## Real-time Clock Operation

The real-time clock (RTC) is a timekeeping device that can be battery or capacitor backed for permanently-powered operation. It offers a software calibration feature that allows high accuracy.

The RTC consists of an oscillator, clock divider, and a register system for user access. It divides down the 32.768 kHz time-base and provides a minimum resolution of seconds ( 1 Hz ). Static registers provide the user with read/write access to the time values. It includes registers for seconds, minutes, hours, day-of-the-week, date, months, and years. A block diagram (Figure 7) illustrates the RTC function.

The user registers are synchronized with the timekeeper core using R and W bits in register 00h described below. Changing the $R$ bit from ' 0 ' to ' 1 ' transfers timekeeping information from the core into holding registers that can be read by the user. If a timekeeper update is pending when R is set, then the core will be updated prior to loading the user registers. The registers are frozen and will not be updated again until the $R$ bit is cleared to ' 0 '. $R$ is used for reading the time.
Setting the $W$ bit to ' 1 ' locks the user registers. Clearing it to ' 0 ' causes the values in the user registers to be loaded into the timekeeper core. W bit is used for writing new time values. Users should be certain not to load invalid values, such as FFh, to the timekeeping registers. Updates to the timekeeping core occur continuously except when locked.

Figure 7. Real-time Clock Core Block Diagram


## Backup Power

The real-time clock/calendar is intended to be permanently powered. When the primary system power fails, the voltage on the $V_{D D}$ pin will drop. When $V_{D D}$ is less than 2.5 V , the RTC (and event counters) will switch to the backup power supply on $\mathrm{V}_{\text {BAK }}$. The clock operates at extremely low current in order to maximize battery or capacitor life. However, an advantage of combining a clock function with F-RAM memory is that data is not lost regardless of the backup power source.
The $I_{\text {BAK }}$ current varies with temperature and voltage (see DC Electrical Characteristics table). The following graph shows $I_{\text {BAK }}$ as a function of $\mathrm{V}_{\mathrm{BAK}}$. These curves are useful for calculating backup time when a capacitor is used as the $\mathrm{V}_{\mathrm{BAK}}$ source.
Figure 8. $I_{B A K}$ vs. $V_{B A K}$ Voltage


The minimum $V_{B A K}$ voltage varies linearly with temperature. The user can expect the minimum $\mathrm{V}_{\mathrm{BAK}}$ voltage to be 1.23 V at $+85^{\circ} \mathrm{C}$ and 1.90 V at $-40^{\circ} \mathrm{C}$. The tested limit is 1.55 V at $+25^{\circ} \mathrm{C}$.
Note The minimum $\mathrm{V}_{\text {BAK }}$ voltage has been characterized at $-40^{\circ} \mathrm{C}$ and $+85^{\circ} \mathrm{C}$ but is not $100 \%$ tested.

Figure 9. $\mathrm{V}_{\mathrm{BAK}}(\mathrm{min}$.$) vs Temperature$


## Trickle Charger

To facilitate capacitor backup the $\mathrm{V}_{\text {BAK }}$ pin can optionally provide a trickle charge current. When the VBC bit (register OBh, bit 2) is set to ' 1 ', the $\mathrm{V}_{\text {BAK }}$ pin will source approximately $80 \mu \mathrm{~A}$ until $\mathrm{V}_{\text {BAK }}$ reaches 3.75 V . In 5 V systems, this charges the capacitor to $\mathrm{V}_{\mathrm{DD}}$ without an external diode and resistor charger and also prevents the user from exceeding the $\mathrm{V}_{\mathrm{BAK}}$ maximum voltage specification. There is a Fast Charge mode which is enabled by the FC bit (register 0Bh, bit 5). In this mode the trickle charger current is set to approximately 1 mA , allowing a large backup capacitor to charge more quickly.

In the case where no battery is used, the $V_{\text {BAK }}$ pin should be tied to $\mathrm{V}_{\mathrm{SS}}$. $\mathrm{V}_{\mathrm{BAK}}$ should not be tied to 5 V since the $\mathrm{V}_{\mathrm{BAK}}$ (max) specification will be exceeded. Be sure to turn off the trickle charger (VBC = '0'), otherwise charger current will be shunted to ground from $\mathrm{V}_{\mathrm{DD}}$.
Note Systems using lithium batteries should clear the VBC bit to ' 0 ' to prevent battery charging. The $\mathrm{V}_{\text {BAK }}$ circuitry includes an internal $1 \mathrm{~K} \Omega$ series resistor as a safety element. The trickle charger is UL Recognized.

## Calibration

When the CAL bit in the register 00h is set to ' 1 ', the clock enters calibration mode. In calibration mode, the CAL/PFO output pin is dedicated to the calibration function and the power fail output is temporarily unavailable. Calibration operates by applying a digital correction to the counter based on the frequency error. In this mode, the CAL/PFO pin is driven with a 512 Hz (nominal) square wave. Any measured deviation from 512 Hz translates into a timekeeping error. The user converts the measured error in ppm and writes the appropriate correction value to the calibration register. The correction factors are listed in the table below. Positive ppm errors require a negative adjustment that removes pulses. Negative ppm errors require a positive correction that adds pulses. Positive ppm adjustments have the CALS (sign) bit set to ' 1 ', whereas negative ppm adjustments have CALS = '0'. After calibration, the clock will have a maximum error of $\pm 2.17 \mathrm{ppm}$ or $\pm 0.09$ minutes per month at the calibrated temperature.

The calibration setting is stored in F-RAM so it is not lost should the backup source fail. It is accessed with bits $\operatorname{CAL}(4: 0)$ in register 01 h . This value can be written only when the CAL bit is set to a ' 1 '. To exit the calibration mode, the user must clear the CAL bit to a ' 0 '. When the CAL bit is ' 0 ', the CAL/PFO pin will revert to the power fail output function.

## Crystal Oscillator

The crystal oscillator is designed to use a $6 \mathrm{pF} / 12.5 \mathrm{pF}$ crystal without the need for external components, such as loading capacitors. The FM31276/FM31278 device has built-in loading capacitors that are optimized for use with 6 pF crystals, but which work well with 12.5 pF crystals. For either crystal, no additional external loading capacitors are required nor suggested.

If a 32.768 kHz crystal is not used, an external oscillator may be connected to the FM31276/FM31278. Apply the oscillator to the X1 pin. Its high and low voltage levels can be driven rail-to-rail or amplitudes as low as approximately 500 mV p-p. To ensure proper operation, a DC bias must be applied to the X2 pin. It should be centered between the high and low levels on the X1 pin. This can be accomplished with a voltage divider.

Figure 10. External Oscillator


In the example, R1 and R2 are chosen such that the X2 voltage is centered around the X1 oscillator drive levels. If you wish to avoid the DC current, you may choose to drive X1 with an external clock and X2 with an inverted clock using a CMOS inverter.

## Layout Recommendations

The X1 and X2 crystal pins employ very high impedance circuits and the oscillator connected to these pins can be upset by noise or extra loading. To reduce RTC clock errors from signal switching noise, a guard ring must be placed around these pads
and the guard ring grounded. SDA and SCL traces should be routed away from the X1 / X2 pads. The X1 and X2 trace lengths should be less than 5 mm . The use of a ground plane on the backside or inner board layer is preferred. See layout example. Red is the top layer, green is the bottom layer.

Figure 11. Layout Recommendations


Layout for Surface Mount Crystal (red $=$ top layer, green $=$ bottom layer $)$


Layout for Through Hole Crystal
(red = top layer, green = bottom layer)

## Table 3. Digital Calibration Adjustments

| Positive Calibration for slow clocks: Calibration will achieve $\pm 2.17$ PPM after calibration |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Measured Frequency Range |  | Error Range (PPM) |  |  |
|  | Min | Max | Min | Max | Program Calibration Register to: |
| 0 | 512.0000 | 511.9989 | 0 | 2.17 | 000000 |
| 1 | 511.9989 | 511.9967 | 2.18 | 6.51 | 100001 |
| 2 | 511.9967 | 511.9944 | 6.52 | 10.85 | 100010 |
| 3 | 511.9944 | 511.9922 | 10.86 | 15.19 | 100011 |
| 4 | 511.9922 | 511.9900 | 15.20 | 19.53 | 100100 |
| 5 | 511.9900 | 511.9878 | 19.54 | 23.87 | 100101 |
| 6 | 511.9878 | 511.9856 | 23.88 | 28.21 | 100110 |
| 7 | 511.9856 | 511.9833 | 28.22 | 32.55 | 100111 |
| 8 | 511.9833 | 511.9811 | 32.56 | 36.89 | 101000 |
| 9 | 511.9811 | 511.9789 | 36.90 | 41.23 | 101001 |
| 10 | 511.9789 | 511.9767 | 41.24 | 45.57 | 101010 |
| 11 | 511.9767 | 511.9744 | 45.58 | 49.91 | 101011 |
| 12 | 511.9744 | 511.9722 | 49.92 | 54.25 | 101100 |
| 13 | 511.9722 | 511.9700 | 54.26 | 58.59 | 101101 |
| 14 | 511.9700 | 511.9678 | 58.60 | 62.93 | 101110 |
| 15 | 511.9678 | 511.9656 | 62.94 | 67.27 | 101111 |
| 16 | 511.9656 | 511.9633 | 67.28 | 71.61 | 110000 |
| 17 | 511.9633 | 511.9611 | 71.62 | 75.95 | 110001 |
| 18 | 511.9611 | 511.9589 | 75.96 | 80.29 | 110010 |
| 19 | 511.9589 | 511.9567 | 80.30 | 84.63 | 110011 |
| 20 | 511.9567 | 511.9544 | 84.64 | 88.97 | 110100 |
| 21 | 511.9544 | 511.9522 | 88.98 | 93.31 | 110101 |
| 22 | 511.9522 | 511.9500 | 93.32 | 97.65 | 110110 |
| 23 | 511.9500 | 511.9478 | 97.66 | 101.99 | 110111 |
| 24 | 511.9478 | 511.9456 | 102.00 | 106.33 | 111000 |
| 25 | 511.9456 | 511.9433 | 106.34 | 110.67 | 111001 |
| 26 | 511.9433 | 511.9411 | 110.68 | 115.01 | 111010 |
| 27 | 511.9411 | 511.9389 | 115.02 | 119.35 | 111011 |
| 28 | 511.9389 | 511.9367 | 119.36 | 123.69 | 111100 |
| 29 | 511.9367 | 511.9344 | 123.70 | 128.03 | 111101 |
| 30 | 511.9344 | 511.9322 | 128.04 | 132.37 | 111110 |
| 31 | 511.9322 | 511.9300 | 132.38 | 136.71 | 111111 |

Table 3. Digital Calibration Adjustments (continued)

| Negative Calibration for fast clocks: Calibration will achieve $\pm 2.17$ PPM after calibration |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Measured Frequency Range |  | Error Range (PPM) |  |  |
|  | Min | Max | Min | Max | Program Calibration Register to: |
| 0 | 512.0000 | 512.0011 | 0 | 2.17 | 000000 |
| 1 | 512.0011 | 512.0033 | 2.18 | 6.51 | 000001 |
| 2 | 512.0033 | 512.0056 | 6.52 | 10.85 | 000010 |
| 3 | 512.0056 | 512.0078 | 10.86 | 15.19 | 000011 |
| 4 | 512.0078 | 512.0100 | 15.20 | 19.53 | 000100 |
| 5 | 512.0100 | 512.0122 | 19.54 | 23.87 | 000101 |
| 6 | 512.0122 | 512.0144 | 23.88 | 28.21 | 000110 |
| 7 | 512.0144 | 512.0167 | 28.22 | 32.55 | 000111 |
| 8 | 512.0167 | 512.0189 | 32.56 | 36.89 | 001000 |
| 9 | 512.0189 | 512.0211 | 36.90 | 41.23 | 001001 |
| 10 | 512.0211 | 512.0233 | 41.24 | 45.57 | 001010 |
| 11 | 512.0233 | 512.0256 | 45.58 | 49.91 | 001011 |
| 12 | 512.0256 | 512.0278 | 49.92 | 54.25 | 001100 |
| 13 | 512.0278 | 512.0300 | 54.26 | 58.59 | 001101 |
| 14 | 512.0300 | 512.0322 | 58.60 | 62.93 | 001110 |
| 15 | 512.0322 | 512.0344 | 62.94 | 67.27 | 001111 |
| 16 | 512.0344 | 512.0367 | 67.28 | 71.61 | 010000 |
| 17 | 512.0367 | 512.0389 | 71.62 | 75.95 | 010001 |
| 18 | 512.0389 | 512.0411 | 75.96 | 80.29 | 010010 |
| 19 | 512.0411 | 512.0433 | 80.30 | 84.63 | 010011 |
| 20 | 512.0433 | 512.0456 | 84.64 | 88.97 | 010100 |
| 21 | 512.0456 | 512.0478 | 88.98 | 93.31 | 010101 |
| 22 | 512.0478 | 512.0500 | 93.32 | 97.65 | 010110 |
| 23 | 512.0500 | 512.0522 | 97.66 | 101.99 | 010111 |
| 24 | 512.0522 | 512.0544 | 102.00 | 106.33 | 011000 |
| 25 | 512.0544 | 512.0567 | 106.34 | 110.67 | 011001 |
| 26 | 512.0567 | 512.0589 | 110.68 | 115.01 | 011010 |
| 27 | 512.0589 | 512.0611 | 115.02 | 119.35 | 011011 |
| 28 | 512.0611 | 512.0633 | 119.36 | 123.69 | 011100 |
| 29 | 512.0633 | 512.0656 | 123.70 | 128.03 | 011101 |
| 30 | 512.0656 | 512.0678 | 128.04 | 132.37 | 011110 |
| 31 | 512.0678 | 512.0700 | 132.38 | 136.71 | 011111 |

## Register Map

The RTC and processor companion functions are accessed via 25 special function registers, which are mapped to a separate $\mathrm{I}^{2} \mathrm{C}$ device ID. The interface protocol is described on page 19. The registers contain timekeeping data, control bits, and information flags. A description of each register follows the summary table.
Table 4. Register Map Summary Table
Nonvolatile $=\square \quad$ Battery-backed $=\square$

| Address | Data |  |  |  |  |  |  |  | Function | Range |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |
| 18h | Serial Number Byte 7 |  |  |  |  |  |  |  | Serial Number 7 | FFh |
| 17h | Serial Number Byte 6 |  |  |  |  |  |  |  | Serial Number 6 | FFh |
| 16h | Serial Number Byte 5 |  |  |  |  |  |  |  | Serial Number 5 | FFh |
| 15h | Serial Number Byte 4 |  |  |  |  |  |  |  | Serial Number 4 | FFh |
| 14h | Serial Number Byte 3 |  |  |  |  |  |  |  | Serial Number 3 | FFh |
| 13h | Serial Number Byte 2 |  |  |  |  |  |  |  | Serial Number 2 | FFh |
| 12h | Serial Number Byte 1 |  |  |  |  |  |  |  | Serial Number 1 | FFh |
| 11h | Serial Number Byte 0 |  |  |  |  |  |  |  | Serial Number 0 | FFh |
| 10h | Counter 2 MSB |  |  |  |  |  |  |  | Event Counter 2 MSB | FFh |
| 0Fh | Counter 2 LSB |  |  |  |  |  |  |  | Event Counter 2 LSB | FFh |
| 0Eh | Counter 1 MSB |  |  |  |  |  |  |  | Event Counter 1 MSB | FFh |
| 0Dh | Counter 1 LSB |  |  |  |  |  |  |  | Event Counter 1 LSB | FFh |
| 0Ch |  |  |  |  | RC | CC | C2P | C1P | Event Count Control |  |
| 0Bh | SNL | - | FC | WP1 | WP0 | VBC | - | VTP | Companion Control |  |
| 0Ah | WDE | - | - | WDT4 | WDT3 | WDT2 | WDT1 | WDT0 | Watchdog Control |  |
| 09h | WTR | POR | LB | - | WR3 | WR2 | WR1 | WR0 | Watchdog Restart/Flags |  |
| 08h | 10 years |  |  |  | years |  |  |  | Years | 00-99 |
| 07h | 0 | 0 | 0 | $\begin{gathered} 10 \\ \text { months } \end{gathered}$ | months |  |  |  | Month | 01-12 |
| 06h | 0 | 0 | 10 date |  | date |  |  |  | Date | 01-31 |
| 05h | 0 | 0 | 0 | 0 | 0 |  | day |  | Day | 01-07 |
| 04h | 0 | 0 | 10 hours |  | hours |  |  |  | Hours | 00-23 |
| 03h | 0 | 10 minutes |  |  | minutes |  |  |  | Minutes | 00-59 |
| 02h | 0 | 10 seconds |  |  | seconds |  |  |  | Seconds | 00-59 |
| 01h | $\overline{\text { OSCEN }}$ | reserved | CALS | CAL4 | CAL3 | CAL2 | CAL1 | CALO | CAL Control |  |
| 00h | reserved | CF | reserved | reserved | reserved | CAL | W | R | RTC Control |  |

Note When the device is first powered up and programmed, all timekeeping registers must be written because the battery-backed register values cannot be guaranteed. The table below shows the default values of the non-volatile registers. All other register values should be treated as unknown.

## Table 5. Default Register Values

| Address | Hex Value |
| :--- | :--- |
| 18 h | $0 \times 00$ |
| 17 h | $0 \times 00$ |
| 16 h | $0 \times 00$ |
| 15 h | $0 \times 00$ |
| 14 h | $0 \times 00$ |
| 13 h | $0 \times 00$ |
| 1 h | $0 \times 00$ |
| 11 h | $0 \times 00$ |
| 0 Bh | $0 \times 00$ | | Address | Hex Value |
| :--- | :--- | :--- |
| 0 Ah | $0 \times 1 \mathrm{~F}$ |
| 07 h | $0 \times 00$ |
| 06 h | $0 \times 01$ |
| 05 h | $0 \times 01$ |
| 04 h | $0 \times 01$ |
| 03 h | $0 \times 00$ |
| 02 h | $0 \times 01$ |
| 01 h | $0 \times 00$ |

Table 6. Register Description

| Address | Description |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 18h | Serial Number Byte 7 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 63 | SN. 62 | SN. 61 | SN. 60 | SN. 59 | SN. 58 | SN. 57 | SN. 56 |
|  | Upper byte of the serial number. Read/write when SNL = '0', read-only when SNL = '1'. Nonvolatile. |  |  |  |  |  |  |  |
| 17h | Serial Number Byte 6 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 55 | SN. 54 | SN. 53 | SN. 52 | SN. 51 | SN. 50 | SN. 49 | SN. 48 |
| 16h | Byte 6 of the serial number. Read/write when SNL = '0', read-only when SNL = '1'. Nonvolatile. |  |  |  |  |  |  |  |
|  | Serial Number Byte 5 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 47 | SN. 46 | SN. 45 | SN. 44 | SN. 43 | SN. 42 | SN. 41 | SN. 40 |
|  | Byte 5 of the serial number. Read/write when $\mathrm{SNL}=$ ' 0 ', read-only when $\mathrm{SNL}=$ ' 1 '. Nonvolatile. |  |  |  |  |  |  |  |
| 15h | Serial Number Byte 4 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 39 | SN. 38 | SN. 37 | SN. 36 | SN. 35 | SN. 34 | SN. 33 | SN. 32 |
|  | Byte 4 of the serial number. Read/write when $\mathrm{SNL}=$ '0', read-only when SNL = ' 1 '. Nonvolatile. |  |  |  |  |  |  |  |
| 14h | Serial Number Byte 3 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 31 | SN. 30 | SN. 29 | SN. 28 | SN. 27 | SN. 26 | SN. 25 | SN. 24 |
|  | Byte 3 of the serial number. Read/write when SNL = '0', read-only when SNL = '1'. Nonvolatile. |  |  |  |  |  |  |  |
| 13h | Serial Number Byte 2 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 23 | SN. 22 | SN. 21 | SN. 20 | SN. 19 | SN. 18 | SN. 17 | SN. 16 |
|  | Byte 2 of the serial number. Read/write when SNL = ' 0 ', read-only when SNL = ' 1 '. Nonvolatile. |  |  |  |  |  |  |  |
| 12h | Serial Number Byte 1 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 15 | SN. 14 | SN. 13 | SN. 12 | SN. 11 | SN. 10 | SN. 9 | SN. 8 |
|  | Byte 1 of the serial number. Read/write when SNL = ' 0 ', read-only when SNL = ' 1 '. Nonvolatile. |  |  |  |  |  |  |  |
| 11h | Serial Number Byte 0 |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SN. 7 | SN. 6 | SN. 5 | SN. 4 | SN. 3 | SN. 2 | SN. 1 | SN. 0 |
|  | LSB of the serial number. Read/write when SNL = '0', read-only when SNL = ' 1 '. Nonvolatile. |  |  |  |  |  |  |  |
| 10h | Counter 2 MSB |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | C2.15 | C2.14 | C2.13 | C2.12 | C2.11 | C2.10 | C2.9 | C2.8 |
|  | Event Counter 2 MSB. Increments on overflows from Counter 2 LSB. Battery-backed, read/write. |  |  |  |  |  |  |  |

Table 6. Register Description (continued)

| Address | Description |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OFh | Counter 2 LSB |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | C2.7 | C2.6 | C2.5 | C2. 4 | C2.3 | C2.2 | C2.1 | C2.0 |
|  | Event Counter 2 LSB. Increments on programmed edge event on CNT2 input or overflows from Counter 1 MSB when CC = '1'. Battery-backed, read/write. |  |  |  |  |  |  |  |
| OEh | Counter 1 MSB |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | C1.15 | C1.14 | C1.13 | C1.12 | C1.11 | C1.10 | C1.9 | C1.8 |
|  | Event Counter 1MSB. Increments on overflows from Counter 1 LSB. Battery-backed, read/write. |  |  |  |  |  |  |  |
| ODh | Counter 1 LSB |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | C1.7 | C1.6 | C1.5 | C1.4 | C1.3 | C1.2 | C1.1 | C1.0 |
|  | Event Counter 1 LSB. Increments on programmed edge event on CNT1 input. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 0Ch | Event Counter Control |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | - | - | - | - | RC | CC | C2P | C1P |
| RC | Read Counter. Setting this bit to ' 1 ' takes a snapshot of the four counters bytes allowing the system to read the values without missing count events. The RC bit will be automatically cleared. |  |  |  |  |  |  |  |
| CC | Counter Cascade. When $C C=$ ' 0 ', the event counters operate independently according to the edge programmed by C1P and C2P respectively. When CC = ' 1 ', the counters are cascaded to create one 32-bit counter. The registers of Counter 2 represent the most significant 16-bits of the counter and CNT1 is the controlling input. Bit C2P is don't care when CC = ' 1 '. Battery-backed, read/write. |  |  |  |  |  |  |  |
| C2P | CNT2 detects falling edges when C2P = ' 0 ', rising edges when $\mathrm{C} 2 \mathrm{P}=$ ' 1 '. C2P is "don't care" when $\mathrm{CC}=$ ' 1 '. The value of Event Counter 2 may inadvertently increment if C2P is changed. Battery-backed, read/write. |  |  |  |  |  |  |  |
| C1P | CNT1 detects falling edges when $\mathrm{C} 1 \mathrm{P}=$ ' 0 ', rising edges when $\mathrm{C} 1 \mathrm{P}=$ ' 1 '. The value of Event Counter 1 may inadvertently increment if C 1 P is changed. Battery-backed, read/write. |  |  |  |  |  |  |  |
| OBh | Companion Control |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | SNL | - | FC | WP1 | WP0 | VBC | VTP1 | VTP0 |
| SNL | Serial Number Lock: Setting to a ' 1 ' makes registers 11 h to 18 h and SNL permanently read-only. SNL cannot be cleared once set to '1'. Nonvolatile, read/write. |  |  |  |  |  |  |  |
| FC | Fast Charge: Setting FC to '1' (and VBC = '1') causes a $\sim 1 \mathrm{~mA}$ trickle charge current to be supplied on $\mathrm{V}_{\text {BAK }}$. Clearing VBC to '0' disables the charge current. Nonvolatile, read/write. |  |  |  |  |  |  |  |
| WP(1:0) | Write Protect. These bits control the write protection of the memory array. Nonvolatile, read/write. <br> Write protect address <br> WP1 <br> WP0 |  |  |  |  |  |  |  |
|  | None$0$$0$ |  |  |  |  |  |  |  |
|  | Bottom 1/4 |  | 0 |  |  |  |  |  |
|  | Bottom 1/2 |  |  |  |  |  |  |  |
|  | Full array |  |  |  |  |  |  |  |
| VBC | $V_{\text {BAK }}$ Charger Control. Setting VBC to ' 1 ' (and FC = ' 0 ') causes a $80 \mu \mathrm{~A}$ ( 1 mA if $\mathrm{FC}=$ ' 1 ') trickle charge current to be supplied on $V_{\text {BAK }}$. Clearing VBC to ' 0 ' disables the charge current. Nonvolatile, read/write. |  |  |  |  |  |  |  |

Table 6. Register Description (continued)

| Address | Description |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VTP | VTP Select. These bits control the reset trip point for the low $\mathrm{V}_{\mathrm{DD}}$ reset function. Nonvolatile, read/write. <br> Trip Volatge <br> VTP |  |  |  |  |  |  |
| OAh | Watchdog Control |  |  |  |  |  |  |
|  | D7 ${ }^{\text {d }}$ | D5 | D4 | D3 | D2 | D1 | D0 |
|  | WDE - | - | WDT4 | WDT3 | WDT2 | WDT1 | WDT0 |
| WDE | Watchdog Enable. When WDE = ' 1 ', a watchdog timer fault will cause the $\overline{\text { RST }}$ signal to go active. When WDE = '0' the timer runs but has no effect on RST, however the WTR flag will be set when a fault occurs. Note as the timer is free-running, users should restart the timer using $\mathrm{WR}(3: 0)$ prior to setting $\mathrm{WDE}=$ ' 1 '. This assures a full watchdog timeout interval occurs. Nonvolatile, read/write. |  |  |  |  |  |  |
| WDT(4:0) | Watchdog Timeout. Indicates the minimum watchdog timeout interval with 100 ms resolution. New watchdog timeouts are loaded when the timer is restarted by writing the 1010b pattern to WR(3:0). Nonvolatile, read/write. |  |  |  |  |  |  |
|  | Watchdog Timeout |  | WDT3 | WDT2 | WDT1 | WDT0 |  |
|  | Invalid - default 100 ms | 0 | 0 | 0 | 0 | 0 |  |
|  | 100 ms | 0 | 0 | 0 | 0 | 1 |  |
|  | 200 ms | 0 | 0 | 0 | 1 | 0 |  |
|  | 300 ms | 0 | 0 | 0 | 1 | 1 |  |
|  | 2000 ms | 1 | 0 | 1 | 0 | 0 |  |
|  | 2100 ms | 1 | 0 | 1 | 0 |  |  |
|  | 2200 ms | 1 | 0 | 1 | 1 | 10 |  |
|  | 2900 ms | 1 | 1 | 1 | 0 | 1 |  |
|  | 3000 ms | 1 | 1 | 1 | 1 | 0 |  |
|  | Disable Counter | 1 | 1 | 1 | 1 | 1 |  |
| 09h | Watchdog Restart and Flags |  |  |  |  |  |  |
|  | D7 $\quad$ D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | WTR POR | LB | - | WR3 | WR2 | WR1 | WR0 |
| WTR | Watchdog Timer Reset Flag: When a watchdog timer fault occurs, the WTR bit will be set to ' 1 '. It must be cleared by the user. Note that both WTR and POR could be set if both reset sources have occurred since the flags were cleared by the user. Battery-backed. Read/Write (internally set, user can clear bit). |  |  |  |  |  |  |
| POR | Power-on Reset Flag: When the $\overline{R S T}$ pin is activated by $\mathrm{V}_{\mathrm{DD}}<\mathrm{V}_{\mathrm{TP}}$, the POR bit will be set to ' 1 '. It must be cleared by the user. Note that both WTR and POR could be set if both reset sources have occurred since the flags were cleared by the user. Battery-backed. Read/Write (internally set, user can clear bit). |  |  |  |  |  |  |
| LB | Low Backup Flag: On power up, if the $\mathrm{V}_{\text {BAK }}$ source is below the minimum voltage to operate the RTC and event counters, this bit will be set to ' 1 '. The user should clear it to ' 0 ' when initializing the system. Battery-backed. Read/Write (internally set, user can clear bit). |  |  |  |  |  |  |

Table 6. Register Description (continued)

| Address | Description |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WR(3:0) | Watchdog Restart: Writing a pattern 1010b to WR(3:0) restarts the watchdog timer. The upper nibble contents do not affect this operation. Writing any pattern other than 1010b to WR(3:0) has no effect on the timer. This allows users to clear the WTR, POR, and LB flags without affecting the watchdog timer. Battery-backed, Write-only. |  |  |  |  |  |  |  |
| 08h | Timekeeping - Years |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | DO |
|  | 10 year. 3 | 10 year. 2 | 10 year. 1 | 10 year. 0 | Year. 3 | Year. 2 | Year. 1 | Year. 0 |
|  | Contains the lower two BCD digits of the year. Lower nibble contains the value for years; upper nibble contains the value for 10 s of years. Each nibble operates from 0 to 9 . The range for the register is $0-99$. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 07h | Timekeeping - Months |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | 0 | 0 | 0 | 10 Month | Month. 3 | Month. 2 | Month. 1 | Month. 0 |
|  | Contains the BCD digits for the month. Lower nibble contains the lower digit and operates from 0 to 9 ; upper nibble (one bit) contains the upper digit and operates from 0 to 1 . The range for the register is $1-12$. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 06h | Timekeeping - Date of the month |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | 0 | 0 | 10 date. 1 | 10 date. 0 | Date. 3 | Date. 2 | Date. 1 | Date. 0 |
|  | Contains the BCD digits for the date of the month. Lower nibble contains the lower digit and operates from 0 to 9 ; upper nibble contains the upper digit and operates from 0 to 3 . The range for the register is $1-31$. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 05h | Timekeeping - Day of the week |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | 0 | 0 | 0 | 0 | 0 | Day. 2 | Day. 1 | Day. 0 |
|  | Lower nibble contains a value that correlates to day of the week. Day of the week is a ring counter that counts from 1 to 7 then returns to 1 . The user must assign meaning to the day value, as the day is not integrated with the date. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 04h | Timekeeping - Hours |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | 0 | 0 | 10 hours. 1 | 10 hours. 0 | Hours. 3 | Hours. 2 | Hours. 1 | Hours. 0 |
|  | Contains the BCD value of hours in 24-hour format. Lower nibble contains the lower digit and operates from 0 to 9 ; upper nibble (two bits) contains the upper digit and operates from 0 to 2 . The range for the register is $0-23$. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 03h | Timekeeping - Minutes |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | 0 | 10 min .2 | 10 min .1 | 10 min .0 | Min. 3 | Min. 2 | Min. 1 | Min. 0 |
|  | Contains the BCD value of minutes. Lower nibble contains the lower digit and operates from 0 to 9 ; upper nibble contains the upper minutes digit and operates from 0 to 5 . The range for the register is $0-59$. Battery-backed, read/write. |  |  |  |  |  |  |  |

Table 6. Register Description (continued)

| Address | Description |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 02h | Timekeeping - Seconds |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | 0 | 10 sec .2 | 10 sec .1 | 10 sec .0 | Seconds. 3 | Seconds. 2 | Seconds. 1 | Seconds. 0 |
|  | Contains the BCD value of seconds. Lower nibble contains the lower digit and operates from 0 to 9 ; upper nibble contains the upper digit and operates from 0 to 5 . The range for the register is $0-59$. Battery-backed, read/write. |  |  |  |  |  |  |  |
| 01h | CAL/Control |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | OSCEN | Reserved | CALS | CAL. 4 | CAL. 3 | CAL. 2 | CAL. 1 | CAL. 0 |
| OSCEN | Oscillator Enable. When set to ' 1 ', the oscillator is halted. When set to ' 0 ', the oscillator runs. Disabling the oscillator can save battery power during storage. On a power-up without battery, this bit is set to ' 1 '. Battery-backed, read/write. |  |  |  |  |  |  |  |
| Reserved | Reserved bits. Do not use. Should remain set to '0'. |  |  |  |  |  |  |  |
| CALS | Calibration Sign: Determines if the calibration adjustment is applied as an addition to or as a subtraction from the time-base. This bit can be written only when CAL = ' 1 '. Nonvolatile, read/write. |  |  |  |  |  |  |  |
| CAL(4:0) | Calibration Setting: These five bits control the calibration of the clock. These bits can be written only when CAL = '1'. Nonvolatile, read/write. |  |  |  |  |  |  |  |
| 00h | RTC Control |  |  |  |  |  |  |  |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | Reserved | CF | Reserved | Reserved | Reserved | CAL | W | R |
| CF | Century Overflow Flag. This bit is set to a ' 1 ' when the values in the years register overflows from 99 to 00 . This indicates a new century, such as going from 1999 to 2000 or 2099 to 2100 . The user should record the new century information as needed. This bit is cleared to ' 0 ' when the Flag register is read. It is read-only for the user. Battery-backed. |  |  |  |  |  |  |  |
| CAL | Calibration Setting. When set to ' 1 ', the clock enters calibration mode. When CAL is set to ' 0 ', the clock operates normally, and the CAL/PFO pin is controlled by the power fail comparator. Battery-backed, read/write. |  |  |  |  |  |  |  |
| W | Write Time. Setting the $W$ bit to ' 1 ' freezes the clock. The user can then write the timekeeping registers with updated values. Resetting the W bit to '0' causes the contents of the time registers to be transferred to the timekeeping counters and restarts the clock. Battery-backed, read/write. |  |  |  |  |  |  |  |
| R | Read Time. Setting the R bit to ' 1 ' copies a static image of the timekeeping core and place it into the user registers. The user can then read them without concerns over changing values causing system errors. The R bit going from ' 0 ' to ' 1 ' causes the timekeeping capture, so the bit must be returned to ' 0 ' prior to reading again. Battery-backed, read/write. |  |  |  |  |  |  |  |
| Reserved | Reserved bits. Do not use. Should remain set to ' 0 '. |  |  |  |  |  |  |  |

## $\mathrm{I}^{2} \mathrm{C}$ Interface

The FM31276/FM31278 employs an industry standard $\mathrm{I}^{2} \mathrm{C}$ bus that is familiar to many users. This product is unique since it incorporates two logical devices in one chip. Each logical device can be accessed individually. Although monolithic, it appears to the system software to be two separate products. One is a memory device. It has a Slave Address (Slave ID = 1010b) that operates the same as a stand-alone memory device. The second device is a real-time clock and processor companion which have a unique Slave Address (Slave ID = 1101b).

By convention, any device that is sending data onto the bus is the transmitter while the target device for this data is the receiver. The device that is controlling the bus is the master. The master is responsible for generating the clock signal for all operations. Any device on the bus that is being controlled is a slave. The FM31276/FM31278 is always a slave device.

The bus protocol is controlled by transition states in the SDA and SCL signals. There are four conditions including START, STOP, data bit, or acknowledge. Figure 12 and Figure 13 illustrates the signal conditions that specify the four states. Detailed timing diagrams are shown in the electrical specifications section.

## STOP Condition (P)

A STOP condition is indicated when the bus master drives SDA from LOW to HIGH while the SCL signal is HIGH. All operations using the FM31276/FM31278 should end with a STOP condition. If an operation is in progress when a STOP is asserted, the operation will be aborted. The master must have control of SDA in order to assert a STOP condition.

## START Condition (S)

A START condition is indicated when the bus master drives SDA from HIGH to LOW while the SCL signal is HIGH. All commands should be preceded by a START condition. An operation in progress can be aborted by asserting a START condition at any time. Aborting an operation using the START condition will ready the FM31276/FM31278 for a new operation.

If during operation the power supply drops below the specified $\mathrm{V}_{\mathrm{TP}}$ minimum, any $\mathrm{I}^{2} \mathrm{C}$ transaction in progress will be aborted and the system should issue a START condition prior to performing another operation.

Figure 12. START and STOP Conditions


Figure 13. Data Transfer on the $I^{2} C$ Bus


## Data/Address Transfer

All data transfers (including addresses) take place while the SCL signal is HIGH. Except under the three conditions described above, the SDA signal should not change while SCL is HIGH.

## Acknowledge / No-acknowledge

The acknowledge takes place after the 8th data bit has been transferred in any transaction. During this state the transmitter should release the SDA bus to allow the receiver to drive it. The receiver drives the SDA signal LOW to acknowledge receipt of
the byte. If the receiver does not drive SDA LOW, the condition is a no-acknowledge and the operation is aborted.

The receiver would fail to acknowledge for two distinct reasons. First is that a byte transfer fails. In this case, the no-acknowledge ceases the current operation so that the device can be addressed again. This allows the last byte to be recovered in the event of a communication error.

Second and most common, the receiver does not acknowledge to deliberately end an operation. For example, during a read
operation, the FM31276/FM31278 will continue to place data onto the bus as long as the receiver sends acknowledges (and clocks). When a read operation is complete and no more data is needed, the receiver must not acknowledge the last byte. If the
receiver acknowledges the last byte, this will cause the FM31276/FM31278 to attempt to drive the bus on the next clock while the master is sending a new command such as STOP.

Figure 14. Acknowledge on the $\mathrm{I}^{2} \mathrm{C}$ Bus


## Slave Address

The first byte that the FM31276/FM31278 expects after a START condition is the slave address. As shown in Figure 15 and Figure 16 , the slave address contains the device type or slave ID, the device select address bits, and a bit that specifies if the transaction is a read or a write.

The FM31276/FM31278 has two Slave Addresses (Slave IDs) associated with two logical devices. Bits 7-4 are the device type (slave ID) and should be set to 1010b for the memory device. The other logical device within the FM31276/FM31278 is the real-time clock and companion. Bits 7-4 are the device type (slave ID) and should be set to 1101b for the RTC and companion. A bus transaction with this slave address will not affect the memory in any way. The figures below illustrate the two Slave Addresses.

Bits 2-1 are the device select address bits. They must match the corresponding value on the external address pins to select the device. Up to four FM31276/FM31278 devices can reside on the same $\mathrm{I}^{2} \mathrm{C}$ bus by assigning a different address to each. Bit 0 is the read/write bit ( $R / \bar{W}$ ). $R / \bar{W}=$ ' 1 ' indicates a read operation and $R / \bar{W}=$ ' 0 ' indicates a write operation.

Figure 15. Memory Slave Device Address


Figure 16. Companion Slave Device Address


## Addressing Overview - Memory

After the FM31276/FM31278 (as receiver) acknowledges the slave address, the master can place the memory address on the bus for a write operation. The address requires two bytes. The complete 15 -bit address is latched internally. Each access causes the latched address value to be incremented automatically. The current address is the value that is held in the latch; either a newly written value or the address following the last access. The current address will be held for as long as $\mathrm{V}_{\mathrm{DD}}>\mathrm{V}_{T P}$ or until a new value is written. Reads always use the current address. A random read address can be loaded by beginning a write operation as explained below.

After transmission of each data byte, just prior to the acknowledge, the FM31276/FM31278 increments the internal address latch. This allows the next sequential byte to be accessed with no additional addressing. After the last address (7FFFh) is reached, the address latch will roll over to 0000h. There is no limit to the number of bytes that can be accessed with a single read or write operation.

## Addressing Overview - RTC \& Companion

The RTC and Processor Companion operate in a similar manner to the memory, except that it uses only one byte of address. Addresses 00 h to 18 h correspond to special function registers. Attempting to load addresses above 18 h is an illegal condition; the FM31276/FM31278 will return a NACK and abort the $\mathrm{I}^{2} \mathrm{C}$ transaction.

## Data Transfer

After the address bytes have been transmitted, data transfer between the bus master and the FM31276/FM31278 can begin. For a read operation the FM31276/FM31278 will place 8 data bits on the bus then wait for an acknowledge from the master. If the acknowledge occurs, the FM31276/FM31278 will transfer the next sequential byte. If the acknowledge is not sent, the FM31276/FM31278 will end the read operation. For a write
operation, the FM31276/FM31278 will accept 8 data bits from the master then send an acknowledge. All data transfer occurs MSB (most significant bit) first.

## Memory Operation

The FM31276/FM31278 is designed to operate in a manner very similar to other $I^{2} \mathrm{C}$ interface memory products. The major differences result from the higher performance write capability of F-RAM technology. These improvements result in some differences between the FM31276/FM31278 and a similar configuration EEPROM during writes. The complete operation for both writes and reads is explained below.

The memory address for FM31276 range from 0x0000 to $0 \times 1$ FFFF, and for FM31278, they range from $0 \times 0000$ to $0 \times 7 F F F$. Memory functionality is described with respect to FM31278 in the following sections.

## Memory Write Operation

All writes begin with a slave address, then a memory address. The bus master indicates a write operation by setting the LSB of the slave address ( $\mathrm{R} / \overline{\mathrm{W}}$ bit) to a ' 0 '. After addressing, the bus
master sends each byte of data to the memory and the memory generates an acknowledge condition. Any number of sequential bytes may be written. If the end of the address range is reached internally, the address counter will wrap from 7FFFh to 0000h.
Unlike other nonvolatile memory technologies, there is no effective write delay with F-RAM. Since the read and write access times of the underlying memory are the same, the user experiences no delay through the bus. The entire memory cycle occurs in less time than a single bus clock. Therefore, any operation including read or write can occur immediately following a write. Acknowledge polling, a technique used with EEPROMs to determine if a write is complete is unnecessary and will always return a ready condition.
Internally, an actual memory write occurs after the 8th data bit is transferred. It will be complete before the acknowledge is sent. Therefore, if the user desires to abort a write without altering the memory contents, this should be done using START or STOP condition prior to the 8th data bit. The FM31276/FM31278 uses no page buffering.
Figure 17 and Figure 18 below illustrate a single-byte and multiple-byte write cycles.

Figure 17. Single-Byte Write


Figure 18. Multi-Byte Write


## Memory Read Operation

There are two basic types of read operations. They are current address read and selective address read. In a current address read, the FM31276/FM31278 uses the internal address latch to supply the address. In a selective read, the user performs a procedure to set the address to a specific value.

## Current Address \& Sequential Read

As mentioned above the FM31276/FM31278 uses an internal latch to supply the address for a read operation. A current address read uses the existing value in the address latch as a
starting place for the read operation. The system reads from the address immediately following that of the last operation.

To perform a current address read, the bus master supplies a slave address with the LSB set to a '1'. This indicates that a read operation is requested. After receiving the complete slave address, the FM31276/FM31278 will begin shifting out data from the current address on the next clock. The current address is the value held in the internal address latch.

Beginning with the current address, the bus master can read any number of bytes. Thus, a sequential read is simply a current
address read with multiple byte transfers. After each byte the internal address counter will be incremented.

Note Each time the bus master acknowledges a byte, this indicates that the FM31276/FM31278 should read out the next sequential byte.
There are four ways to properly terminate a read operation. Failing to properly terminate the read will most likely create a bus contention as the FM31276/FM31278 attempts to read out additional data onto the bus. The four valid methods are:

1. The bus master issues a no-acknowledge in the 9th clock cycle and a STOP in the 10th clock cycle. This is illustrated in the diagrams below. This is preferred.
2. The bus master issues a no-acknowledge in the 9th clock cycle and a START in the 10th.
3. The bus master issues a STOP in the 9th clock cycle.
4. The bus master issues a START in the 9th clock cycle.

If the internal address reaches 7FFFh, it will wrap around to 0000h on the next read cycle. Figure 19 and Figure 20 below show the proper operation for current address reads.

Figure 19. Current Address Read


Figure 20. Sequential Read


## Selective (Random) Read

There is a simple technique that allows a user to select a random address location as the starting point for a read operation. This involves using the first three bytes of a write operation to set the internal address followed by subsequent read operations.
To perform a selective read, the bus master sends out the slave address with the LSB (R/W) set to ' 0 '. This specifies a write
operation. According to the write protocol, the bus master then sends the address bytes that are loaded into the internal address latch. After the FM31276/FM31278 acknowledges the address, the bus master issues a START condition. This simultaneously aborts the write operation and allows the read command to be issued with the slave address LSB set to a '1'. The operation is now a current address read.

Figure 21. Selective (Random) Read


## RTC/Companion Write Operation

All RTC and Companion writes operate in a similar manner to memory writes. The distinction is that a different device ID is used and only one byte address is needed instead of two byte address. Figure 22 illustrates a single byte write to this device.

Note Although not required, it is recommended that A5-A7 in the register address byte are zeros in order to preserve compatibility with future devices.

Figure 22. Single Byte Write


## RTC/Companion Read Operation

As with writes, a read operation begins with the Slave Address. To perform a register read, the bus master supplies a Slave Address with the LSB set to ' 1 '. This indicates that a read operation is requested. After receiving the complete Slave Address, the FM31276/FM31278 will begin shifting data out from the current register address on the next clock. Auto-increment operates for the special function registers as with the memory address. A current address read for the registers look exactly like the memory except that the device ID is different.

The FM31276/FM31278 contains two separate address registers, one for the memory address and the other for the register address. This allows the contents of one address
register to be modified without affecting the current address of the other register. For example, this would allow an interrupted read to the memory while still providing fast access to an RTC register. A subsequent memory read will then continue from the memory address where it previously left off, without requiring the load of a new memory address. However, a write sequence always requires an address to be supplied.

## Addressing FRAM Array in the FM31276/FM31278 Family

The FM31276/FM31278 family includes 64-Kbit and 256-Kbit memory densities. The following 2-byte address field is shown for each density.

| Part Number | $\mathbf{1}^{\text {st }}$ Address Byte |  |  |  |  | Address Byte |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FM31276 | X | X | X | A12 | A11 | A10 | A9 | A8 | A7 | A6 | A5 | A4 | A3 | A2 | A1 | A0 |
| FM31278 | X | A14 | A13 | A12 | A11 | A10 | A9 | A8 | A7 | A6 | A5 | A4 | A3 | A2 | A1 | A0 |

## Maximum Ratings

Exceeding maximum ratings may shorten the useful life of the device. These user guidelines are not tested.
Storage temperature $\qquad$ $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$

Maximum junction temperature $95^{\circ} \mathrm{C}$

Supply voltage on $\mathrm{V}_{\mathrm{DD}}$ relative to $\mathrm{V}_{\mathrm{SS}} \ldots . . . . . .-1.0 \mathrm{~V}$ to +7.0 V Input voltage $\qquad$ -1.0 V to +7.0 V and $\mathrm{V}_{\mathrm{IN}}<\mathrm{V}_{\mathrm{DD}}+1.0 \mathrm{~V}$
Backup supply voltage $\qquad$ -1.0 V to +4.5 V

DC voltage applied to outputs
in High-Z state $\qquad$ -0.5 V to $\mathrm{V}_{\mathrm{DD}}+0.5 \mathrm{~V}$
Transient voltage (<20 ns) on any pin to ground potential $\qquad$ -2.0 V to $\mathrm{V}_{\mathrm{DD}}+2.0 \mathrm{~V}$

Package power dissipation capability ( $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ ) $\qquad$ 1.0 W

Surface mount lead soldering temperature (3 seconds) $\qquad$ $+260^{\circ} \mathrm{C}$

DC output current
(1 output at a time, 1s duration) .................................. 15 mA
Electrostatic Discharge Voltage
Human Body Model (AEC-Q100-002 Rev. D) ...................... 2 kV
Charged Device Model (AEC-Q100-011 Rev. B) ............. 1.25 kV
Machine Model (AEC-Q100-003 Rev. E) ............................. 100 V
Latch-up current $\qquad$ $> \pm 100 \mathrm{~mA}$

Note PFI input voltage must not exceed 4.5 V . The $\mathrm{V}_{\mathrm{IN}}<$ $\mathrm{V}_{\mathrm{DD}}+1.0 \mathrm{~V}$ " restriction does not apply to the SCL and SDA inputs which do not employ a diode to $\mathrm{V}_{\mathrm{DD}}$.
Operating Range

| Range | Ambient Temperature ( $\mathbf{T}_{\mathbf{A}}$ ) | $\mathbf{V}_{\mathbf{D D}}$ |
| :--- | :---: | :---: |
| Industrial | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 4.0 V to 5.5 V |

## DC Electrical Characteristics

## Over the Operating Range

| Parameter | Description | Test Conditions |  | Min | Typ ${ }^{[1]}$ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{D D}{ }^{[2]}$ | Power supply |  |  | 4.0 | - | 5.5 | V |
| IDD | Average V ${ }_{\text {DD }}$ current | SCL toggling between $\mathrm{V}_{\mathrm{DD}}-0.3 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{SS}}$, other inputs $\mathrm{V}_{\mathrm{SS}}$ or $\mathrm{V}_{\mathrm{DD}}-0.3 \mathrm{~V}$. | $\mathrm{f}_{\mathrm{SCL}}=100 \mathrm{kHz}$ | - | - | 500 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{f}_{\mathrm{SCL}}=400 \mathrm{kHz}$ | - | - | 900 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{f}_{\mathrm{SCL}}=1 \mathrm{MHz}$ | - | - | 1500 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {SB }}$ | $\mathrm{V}_{\mathrm{DD}}$ standby current | $S C L=S D A=V_{D D}$. All other inputs $V_{S S}$ or $\mathrm{V}_{\mathrm{DD}}$. Stop command issued. |  | - | - | 150 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\mathrm{BAK}}{ }^{[3]}$ | RTC backup voltage |  | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 1.55 | - | 3.75 | V |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+25^{\circ} \mathrm{C}$ | 1.90 | - | 3.75 | V |
| $\mathrm{I}_{\text {BAK }}$ | RTC backup current | $\mathrm{V}_{\mathrm{BAK}}=3.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}<$ 2.4 V , oscillator running, CNT1, CNT2 at $\mathrm{V}_{\mathrm{BAK}}$. | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{BAK}}=3.0 \mathrm{~V}$ | - | - | 1.4 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{BAK}}=3.0 \mathrm{~V}$ | - | - | 2.1 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{BAK}}=2.0 \mathrm{~V}$ | - | - | 1.15 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=+85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{BAK}}=2.0 \mathrm{~V}$ | - | - | 1.75 | $\mu \mathrm{A}$ |
| $I_{\text {BAKTC }}{ }^{[4]}$ | Trickle charge current with $\mathrm{V}_{\mathrm{BAK}}=0 \mathrm{~V}$ |  | Fast Charge Off (FC = '0') | 50 | - | 120 | $\mu \mathrm{A}$ |
|  |  |  | Fast Charge On (FC = '1') | 200 | - | 2500 | $\mu \mathrm{A}$ |

[^0]
## DC Electrical Characteristics (continued)

Over the Operating Range

| Parameter | Description | Test Conditions |  | Min | Typ ${ }^{[1]}$ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {TP1 }}$ | $\mathrm{V}_{\mathrm{DD}}$ trip point voltage, $\text { VTP = } 0$ | $\overline{\mathrm{RST}}$ is asserted active when $\mathrm{V}_{\mathrm{DD}}<\mathrm{V}_{T P}$. |  | 3.75 | 3.90 | 4.00 | V |
| $\mathrm{V}_{\text {TP2 }}$ | $\mathrm{V}_{\mathrm{DD}}$ trip point voltage, $\text { VTP = } 1$ | $\overline{\mathrm{RST}}$ is asserted active when $\mathrm{V}_{\mathrm{DD}}<\mathrm{V}_{\mathrm{TP}}$. |  | 4.20 | 4.00 | 4.50 | V |
| $\mathrm{V}_{\mathrm{RST}}{ }^{[5]}$ | $\mathrm{V}_{\mathrm{DD}}$ for valid $\overline{\mathrm{RST}}$ | $\mathrm{I}_{\mathrm{OL}}=80 \mu \mathrm{~A}$ at $\mathrm{V}_{\mathrm{OL}}$ | $\mathrm{V}_{\text {BAK }}>\mathrm{V}_{\text {BAK }}$ min | 0 | - | - | V |
|  |  |  | $\mathrm{V}_{\text {BAK }}<\mathrm{V}_{\text {BAK }}$ min | 1.6 | - | - | V |
| ${ }^{\text {LI }}$ | Input leakage current | $\mathrm{V}_{\mathrm{SS}} \leq \mathrm{V}_{\mathrm{IN}} \leq \mathrm{V}_{\mathrm{DD}}$. Does not apply to $\mathrm{A} 0, \mathrm{~A} 1, \mathrm{X} 1, \mathrm{PFI}$, RST, or X2 |  | - | - | $\pm 1$ | $\mu \mathrm{A}$ |
| LLO | Output leakage current | $\mathrm{V}_{\mathrm{SS}} \leq \mathrm{V}_{\text {OUT }} \leq \mathrm{V}_{\mathrm{DD}}$. Does not apply to $\overline{\mathrm{RST}}, \mathrm{X} 1$, or X 2 |  | - | - | $\pm 1$ | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\mathrm{IL}}{ }^{[6]}$ | Input LOW voltage |  | All inputs except as listed below | -0.3 | - | $0.3 \times \mathrm{V}_{\mathrm{DD}}$ | V |
|  |  |  | CNT1, CNT2 battery-backed ( $\mathrm{V}_{\mathrm{DD}}<2.5 \mathrm{~V}$ ) | -0.3 | - | 0.5 | V |
|  |  |  | CNT1, CNT2 ( $\left.\mathrm{V}_{\mathrm{DD}}>2.5 \mathrm{~V}\right)$ | -0.3 | - | 0.8 | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH voltage |  | All inputs except as listed below | $0.7 \times \mathrm{V}_{\mathrm{DD}}$ | - | $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
|  |  |  | CNT1, CNT2 battery-backed $\left(\mathrm{V}_{\mathrm{DD}}<2.5 \mathrm{~V}\right)$ | $\mathrm{V}_{\mathrm{BAK}}-0.5$ | - | $\mathrm{V}_{\mathrm{BAK}}+0.3$ | V |
|  |  |  | CNT1, CNT2 ( $\left.\mathrm{V}_{\mathrm{DD}}>2.5 \mathrm{~V}\right)$ | $0.7 \times V_{D D}$ | - | $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
|  |  |  | PFI (comparator input) | - | - | 3.75 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output HIGH voltage | $\mathrm{I}_{\mathrm{OH}}=-2 \mathrm{~mA}$ |  | 2.4 | - | - | V |
| $\mathrm{V}_{\mathrm{OL}}$ | Output LOW voltage | $\mathrm{l}_{\mathrm{OL}}=3 \mathrm{~mA}$ |  | - | - | 0.4 | V |
| $\mathrm{R}_{\text {RST }}$ | Pull-up resistance for RST inactive |  |  | 50 | - | 400 | $\mathrm{k} \Omega$ |
| $\mathrm{R}_{\text {in }}$ | Input resistance (A1-A0) | For $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {IL }}($ Max $)$ |  | 20 | - | - | $\mathrm{k} \Omega$ |
|  |  | For $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{IH}}(\mathrm{Min})$ |  | 1 | - | - | $\mathrm{M} \Omega$ |
| $\mathrm{V}_{\mathrm{PFI}}$ | Power fail input reference voltage |  |  | 1.140 | 1.20 | 1.225 | V |
| $\mathrm{V}_{\mathrm{HYS}}$ | Power fail input (PFI) hysteresis (rising) |  |  | - | - | 100 | mV |

## Notes

5. The minimum $\mathrm{V}_{\mathrm{DD}}$ to guarantee the level of $\overline{\mathrm{RST}}$ remains a valid $\mathrm{V}_{\mathrm{L}}$ level.
6. Includes $\overline{\mathrm{RST}}$ input detection of external reset condition to trigger driving of $\overline{\mathrm{RST}}$ signal by FM31276/FM31278.

[^0]:    Notes

    1. Typical values are at $25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DD}}(\mathrm{typ})$. Not $100 \%$ tested.
    2. Full complete operation. Supervisory circuits, RTC, etc operate to lower voltages as specified.
    3. The $\mathrm{V}_{\mathrm{BAK}}$ trickle charger automatically regulates the maximum voltage on this pin for capacitor backup applications.
    4. $\mathrm{V}_{\mathrm{BAK}}$ will source current when trickle charge is enabled (VBC bit $={ }^{\prime} 1$ '), $\mathrm{V}_{\mathrm{DD}}>\mathrm{V}_{\mathrm{BAK}}$, and $\mathrm{V}_{\mathrm{BAK}}<\mathrm{V}_{\mathrm{BAK}}$ max.
