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## FMS6404 Precision Composite Video Output with Sound Trap and Group Delay Compensation

## **Features**

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- 7.6MHz 5th-Order Composite Video Filter
- 14dB Notch at 4.425MHz to 4.6MHz for Sound Trap Capable of Handling Stereo
- 50dB Stopband Attenuation at 27MHz on CV Output
- > 0.5dB Flatness to 4.2MHz on CV Output
- Equalizer and Notch Filter for Driving RF Modulator with Group Delay of -180ns
- No External Frequency Selection Components or Clocks
- < 5ns Group Delay on CV Output</p>
- AC-Coupled Input
- AC- or DC-Coupled Output
- Capable of PAL Frequency for CV
- Continuous Time Low-Pass Filters
- <1.4% Differential Gain with 0.7° Differential Phase on CV Channel
- Integrated DC Restore Circuitry with Low Tilt

## Applications

- Cable Set-Top Boxes
- Satellite Set-Top Boxes
- DVD Players

## Description

The FMS6404 is a single composite video 5th-order Butterworth low-pass video filter optimized for minimum overshoot and flat group delay. The device contains an audio trap that removes video information in a spectral location of the subsequent RF audio carrier. The group delay compensation circuit pre-distorts the signal to compensate for the inherent receiver intermediate frequency (IF) filter's group delay distortion.

In a typical application, the composite video from the DAC is AC coupled into the filter. The CV input has DC-restore circuitry to clamp the DC input levels during video synchronization. The clamp pulse is derived from the CV channel.

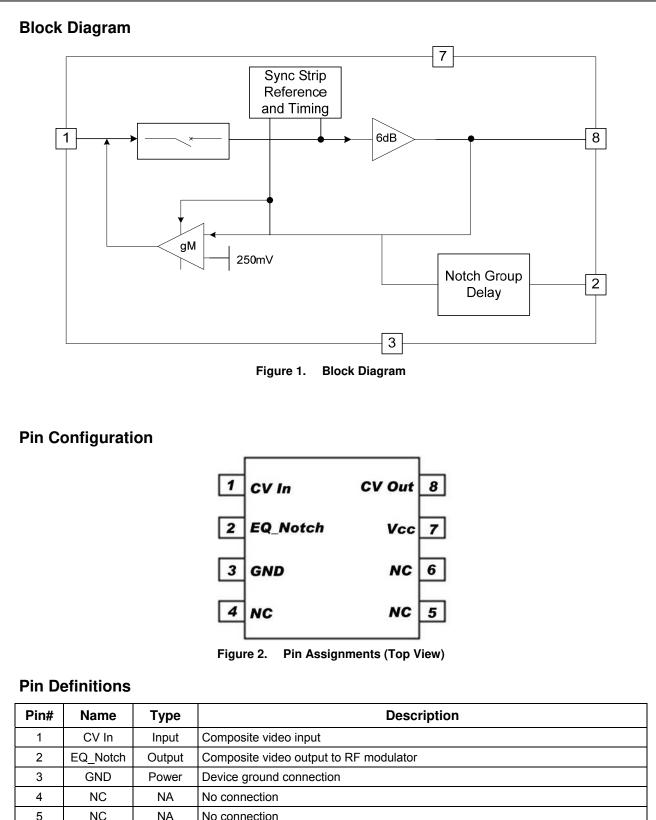
All outputs are capable of driving  $2V_{PP}$ , AC- or DCcoupled, into either a single or dual video load. A single video load consists of a series  $75\Omega$  impedance matching resistor connected to a terminated  $75\Omega$  line. This presents a total of  $150\Omega$  of loading to the part. A dual load would be two of these in parallel, which presents a total of  $75\Omega$  to the part. The gain of the CV signal is 6dB with  $1V_{PP}$  input levels. All video channels are clamped during synchronization to establish the appropriate output voltage reference levels.

## **Related Resources**

- AN-6024 FMS6xxx Product Series Understanding Analog Video Signal Clamps, Bias, DC Restore, and AC or DC coupling Methods
- <u>AN-6041 PCB Layout Considerations for Video</u> <u>Filter / Drivers</u>

## **Ordering Information**

| Part Number | Operating<br>Temperature Range | Package   | Packing Method         |
|-------------|--------------------------------|---|------------------------|
| FMS6404CSX  | -40°C to +70°C                 | 8-Lead, Small-Outline Integrated Circuit (SOIC),<br>JEDEC MS-012, .150" Narrow Body | 2500 Units<br>per Reel |



| Pin# | Name     | Туре   | Description                            |
|------|----------|--------|--|
| 1    | CV In    | Input  | Composite video input                  |
| 2    | EQ_Notch | Output | Composite video output to RF modulator |
| 3    | GND      | Power  | Device ground connection               |
| 4    | NC       | NA     | No connection                          |
| 5    | NC       | NA     | No connection                          |
| 6    | NC       | NA     | No connection                          |
| 7    | Vcc      | Power  | Device power connection                |
| 8    | CV Out   | Output | Composite video output                 |

FMS6404 — Precision Composite Video Output with Sound Trap and Group Delay Compensation

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

| Symbol           | Parameter                             | Min. | Max.                 | Unit |
|------------------|---------------------------------------|------|----------------------|------|
| V <sub>CC</sub>  | DC Supply Voltage                     | -0.3 | 6.0                  | V    |
| V <sub>IO</sub>  | Analog and Digital I/O                | -0.3 | V <sub>CC</sub> +0.3 | V    |
| V <sub>OUT</sub> | Maximum Output Current, Do Not Exceed |      | 100                  | mA   |

## **Electrostatic Discharge Information**

| Symbol | Parameter                         | Min. | Unit |
|--------|-----------------------------------|------|------|
| ESD    | Human Body Model, JESD22-A114     | 8    | kV   |
| ESD    | Charged Device Model, JESD22-C101 | 2    | κv   |

## **Reliability Information**

| Symbol           | Parameter  | Min. | Тур. | Max. | Unit |
|------------------|--|------|------|------|------|
| TJ               | Junction Temperature   |      |      | +150 | °C   |
| T <sub>STG</sub> | Storage Temperature Range  | -65  |      | +150 | °C   |
| TL               | Lead Temperature (Soldering, 10 Seconds)                             |      |      | +300 | °C   |
| □JA              | Thermal Resistance, JEDEC Standard, Multilayer Test Board, Still Air |      | 90   |      | °C/W |

## **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

| Symbol         | Parameter                   | Min. | Тур. | Max. | Unit |
|----------------|-----------------------------|------|------|------|------|
| T <sub>A</sub> | Operating Temperature Range | 0    |      | +70  | °C   |
| Vcc            | Supply Voltage Range        | 4.75 | 5.00 | 5.25 | V    |

## **DC Electrical Characteristics**

 $T_A=25^{\circ}C$ ,  $V_{CC}=5.0V$ ,  $R_S=37.5\Omega$ , all inputs are AC-coupled with  $0.1\mu$ F, and all outputs are AC coupled with  $220\mu$ F into  $150\Omega$  load; unless otherwise noted.

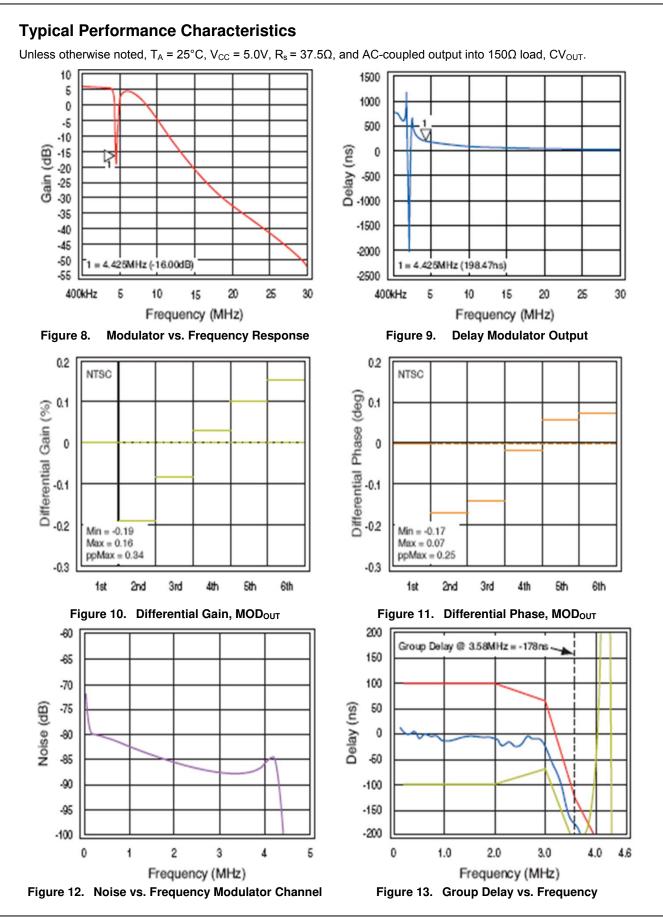
| Symbol          | Parameter                    | Condition                       | Min. | Тур. | Max. | Unit     |
|-----------------|------------------------------|---------------------------------|------|------|------|----------|
| V <sub>CC</sub> | Supply Voltage Range         | V <sub>S</sub> Range            | 4.75 | 5.00 | 5.25 | V        |
| lcc             | Quiescent Supply Current     | V <sub>S</sub> =+5.0V, No Load  | 50   | 70   | 90   | mA       |
| V <sub>IN</sub> | Video Input Voltage Range    | Referenced to GND if DC Coupled |      | 1.4  |      | $V_{PP}$ |
| PSRR            | Power Supply Rejection Ratio | DC                              |      | -50  |      | dB       |
| I <sub>SC</sub> | Output Short Circuit Current | CV, EQ_NOTCH to GND             |      | 85   |      | mA       |

## **AC Electrical Characteristics**

 $T_A=25^{\circ}C$ ,  $V_{CC}=5.0V$ ,  $R_S=37.5\Omega$ , all inputs are AC-coupled with  $0.1\mu$ F, and all outputs are AC coupled with  $220\mu$ F into  $150\Omega$  load, unless otherwise noted.

| Symbol                  | Parameter                                       | Condition  | Min. | Тур. | Max. | Unit |
|-------------------------|---|--|------|------|------|------|
| $AV_{CV}$               | Low Frequency Gain CV <sub>OUT</sub>            | at 400kHz  | 5.8  | 6.0  | 6.2  | dB   |
| $AV_{EQ}$               | Low Frequency Gain<br>(EQ_NOTCH)                | at 400kHz  | 5.7  | 6.0  | 6.4  | dB   |
| $CV_{sync}$             | CV <sub>OUT</sub> Output Level<br>(During Sync) | Sync Present on CV <sub>IN</sub> (After 6dB Gain)  |      | 0.35 | 0.50 | V    |
| EQ <sub>sync</sub>      | EQ_NOTCH Output Level<br>(During Sync)          | Sync Present on CV <sub>IN</sub> (After 6dB Gain)  |      | 0.35 | 0.50 | V    |
| t <sub>CLAMP</sub>      | Clamp Response Time<br>CV <sub>OUT</sub>        | Settled to within 10mV   |      | 5    |      | ms   |
| f <sub>FLAT</sub>       | Gain Flatness to 4.2MHz<br>CV <sub>OUT</sub>    |  | -0.5 | 0    | 0.5  | dB   |
| f <sub>C</sub>          | -3dB Bandwidth                                  | CV <sub>OUT</sub> Channel  | 6.7  | 7.6  |      | MHz  |
| f <sub>SB</sub>         | Stopband Attenuation $CV_{OUT}$                 | at 27MHz   | 40   | 50   |      | dB   |
| dG                      | Differential Gain                               | CV <sub>OUT</sub>  |      | 1.4  | 3.0  | %    |
| dq                      | Differential Phase                              | CV <sub>OUT</sub>  |      | 0.7  | 1.5  | ٥    |
| THD                     | Output Distortion                               | V <sub>OUT</sub> =1.4V <sub>pp</sub> at 3.58MHz  |      | 0.3  |      | %    |
| X <sub>TALK</sub>       | Crosstalk                                       | V <sub>OUT</sub> =1.4V <sub>pp</sub> at 3.58MHz  |      | -50  |      | dB   |
| SNR                     | SNR CV <sub>OUT</sub> Channel                   | NTC-7 Weighting 4.2MHz Low-Pass<br>V <sub>IN</sub> =714mV, V <sub>OUT</sub> =1.428V <sub>PP</sub> /1.010Vrms | 70   | 75   |      | dB   |
|                         | SNR EQ_NOTCH Channel                            | NTC-7 Weighting 4.2MHz Low-Pass V <sub>IN</sub><br>=714mV V <sub>OUT</sub> =1.428Vpp/1.010Vrms               | 65   | 70   |      | dB   |
| t <sub>pd</sub>         | Propagation Delay                               | at 400kHz  |      | 112  |      | ns   |
| GD                      | Group Delay CVOUT                               | at 3.58MHz (Reference to 400KHz)   | -5   | 0    | 5    | ns   |
| t <sub>CLGCV</sub>      | Chroma-Luma Gain CV <sub>OUT</sub>              | f=3.58MHz (Reference to 400kHz)  | 98   | 100  | 102  | %    |
| t <sub>CLDCV</sub>      | Chroma-Luma Delay CV <sub>OUT</sub>             | f=3.58MHz (Reference to 400kHz)  | -10  | 0    | 10   | ns   |
| t <sub>GDEQ</sub>       | Group Delay EQ_NOTCH                            | f=3.58MHz (Reference to 400kHz)  | -195 | -180 | -165 | ns   |
| t <sub>CLGEQ</sub>      | Chroma-Luma Gain<br>EQ_NOTCH                    | f=3.58MHz (Reference to 400kHz)  | 95   | 100  | 105  | %    |
| t <sub>CLDEQ</sub>      | Chroma-Luma Delay<br>EQ_NOTCH                   | f=3.58MHz (Reference to 400kHz)  | -195 | -180 | -165 | ns   |
| $\text{dG}_{\text{EQ}}$ | Differential Gain                               | EQ_NOTCH Channel   |      | 0.3  | 1.0  | %    |
| $dq_{EQ}$               | Differential Phase                              | EQ_NOTCH Channel   |      | 0.30 | 0.75 | %    |
| MCF                     | Modulator Channel Flatness                      | EQ_NOTCH from 400kHz to 3.75MHz  | -0.5 | 0    | 0.5  | dB   |
| $AV_{PK}$               | Gain Peaking                                    | EQ_NOTCH from >3.75MHz to 4.2MHz   | -0.5 | 0    | 0.5  | dB   |
| Atten1                  | Notch Attenuation 1                             | EQ_NOTCH at 4.425MHz   | 14   |      |      | dB   |
| Atten2                  | Notch Attenuation 2                             | EQ_NOTCH at 4.5MHz   | 20   |      |      | dB   |
| Atten3                  | Notch Attenuation 3                             | EQ_NOTCH at 4.6MHz   | 14   |      |      | dB   |
| t <sub>PASS</sub>       | Passband Group Delay<br>EQ_NOTCH                | f=400kHz to f=3MHz   | -35  |      | 35   | ns   |

**Typical Performance Characteristics** Unless otherwise noted,  $T_A = 25^{\circ}$ C,  $V_{CC} = 5.0$ V,  $R_s = 37.5\Omega$ , and AC-coupled output into 150 $\Omega$  load, CV<sub>OUT</sub>. 10 140 0 Ż 120 P 100 -10 Delay (ns) Gain (dB) -20 80 Mkr Frequency Gain -30 60 400kHz 6dB Ref 3 -40 6.53MHz -1dB BW 40 1 7.87MHz -3dB BW 2 20 -50 27MHz -44.66dB 3 = 8.2MHz (111.35ns) = 50.66dB = Gain, - Gain, -60 0 400kHz 5 15 20 25 30 5 15 30 10 400kHz 10 20 25 Frequency (MHz) Frequency (MHz) Figure 3. Frequency Response Figure 4. Group Delay vs. Frequency 2.0 0.8 Min = -0.00 Min = -0.01 NTSC NTSC Max = 1.17Max = 0.59 ppMax = 1.16 Differential Phase (deg) ppMax = 0.60 1.5 0.6 Differential Gain (%) 1.0 0.4 0.5 0.2 0 0 -0.5 -0.2 1st 2nd 3rd 4th Sth 6th 1st 2nd 3rd 4th 5th 6th Figure 5. **Differential Gain** Figure 6. **Differential Phase** -60 -65 -70 -75 Noise (dB) -80 -85 -90 -95 -100 -105 -110 1.0 0 2.0 3.0 4.0 5.0 Frequency (MHz) Figure 7. Noise vs. Frequency



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## **Applications Information**

#### **Layout Considerations**

General layout and supply bypassing play a major role in high-frequency performance and thermal characteristics. Fairchild offers a four-layer board with full power and ground planes board to guide layout and aid device evaluation. Following this layout configuration provides optimum performance and thermal characteristics for the device. For best results, follow the steps and recommended routing rules below.

## **Recommended Routing / Layout Rules**

- Do not run analog and digital signals in parallel.
- Use separate analog and digital power planes to supply power.
- Traces must run on top of the ground plane.
- No trace should run over ground/power splits.
- Avoid routing at 90-degree angles.
- Minimize clock and video data trace length differences.
- Include 10µF and 0.1µF ceramic power supply bypass capacitors.
- Place the 0.1µF capacitor within 2.54mm (0.1in) of the device power pin.
- Place the 10µF capacitor within 19.05mm (0.75in) of the device power pin.
- For multi-layer boards, use a large ground plane to help dissipate heat.
- For two-layer boards, use a ground plane that extends beyond the device body at least 12.7mm (0.5in) on all sides. Include a metal paddle under the device on the top layer.
- Minimize all trace lengths to reduce series inductance.

#### **Output Considerations**

The outputs are DC offset from the input by 150mV; therefore,  $V_{OUT} = 2 \cdot V_{IN}$  DC + 150mV. This offset is required for optimal performance from the output driver and is held at the minimum value to decrease the standing DC current into the load. Since the FMS6404 has a 2 x (6dB) gain, the output is typically connected via a 75 $\Omega$ -series back-matching resistor, followed by the 75 $\Omega$  video cable. Due to the inherent divide-by-two of this configuration, the blanking level at the load of the video signal is always less than 1V. When AC-coupling the output, ensure that the coupling capacitor passes the lowest frequency content in the video signal and that line time distortion (video tilt) is kept as low as possible.

The selection of the coupling capacitor is a function of the subsequent circuit input impedance and the leakage current of the input being driven. To obtain the highest quality output video signal, the series termination resistor must be placed as close to the device output pin as possible. This greatly reduces the parasitic capacitance and inductance effect on the output driver. The distance from the device pin to the series termination resistor should be no greater than 2.54mm (0.1in).

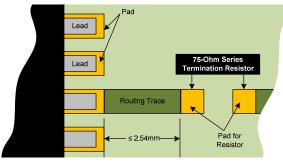


Figure 14. Termination Resistor Placement

#### **Thermal Considerations**

Since the interior of most systems, such as set-top boxes, TVs, and DVD players; is at +70°C; consideration must be given to providing an adequate heat sink for the device package for maximum heat dissipation. When designing a system board, determine how much power each device dissipates. Ensure that devices of high power are not placed in the same location, such as directly above (top plane) or below (bottom plane) each other on the PCB.

#### **PCB Thermal Layout Considerations**

- Understand the system power requirements and environmental conditions.
- Maximize thermal performance of the PCB.
- Consider using 70µm of copper for high-power designs.
- Make the PCB as thin as possible by reducing FR4 thickness.
- Use vias in power pad to tie adjacent layers together.
- Remember that baseline temperature is a function of board area, not copper thickness.
- Modeling techniques provide a first-order approximation.

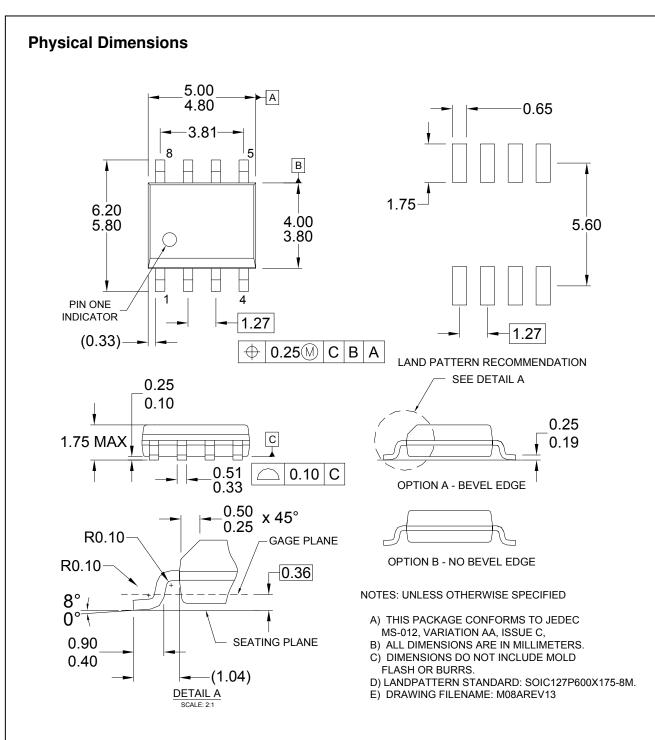


Figure 15. 8-Lead, Small-Outline Integrated Circuit (SOIC), JEDEC MS-012, .150" Narrow Body

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