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# Integrated Load Switch with 500mA High Precision Current Limit 

## Features

■ 1.8 to 5.5 V Input Voltage Range
■ Typical $R_{\mathrm{DS}(\mathrm{ON})}=140 \mathrm{~m} \Omega @ \mathrm{~V}_{\mathrm{IN}}=5.5 \mathrm{~V}$
■ Typical $R_{D S(O N)}=160 \mathrm{~m} \Omega @ \mathrm{~V}_{\mathrm{IN}}=3.3 \mathrm{~V}$
■ Fixed 500 mA Current Limit (min)

- $5 \%$ Accurate Current Limit
- $72 \Omega$ (typ) Output Discharge Resistance
- ESD Protected, above 8 kV HBM and 2 kV CDM


## Applications

- PDAs
- Cell Phones
- GPS Devices
- MP3 Players
- Digital Cameras
- Peripheral Ports
- Notebook Computer


## General Description

The FPF2200-FPF2202 are low $\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}$ P-Channel MOSFET load switches with high precision current limit value. The input voltage range operates from 1.8 V to 5.5 V to fulfill today's Ultra Portable Device's supply requirement. Switch control is by a logic input (ON) capable of interfacing directly with low voltage control signal. On-chip pull-down is available for output quick discharge when switch is turned off.

For the FPF2201, if the constant current condition still persists after 30 ms , these parts will shut off the switch and pull the fault signal pin (FLAGB) low. The FPF2200 has an auto-restart feature which will turn the switch on again after 450 ms if the ON pin is still active. The FPF2201 do not have this auto-restart feature so the switch will remain off until the ON pin is cycled. For the FPF2202, a current limit condition will immediately pull the fault signal pin low and the part will remain in the constantcurrent mode until the switch current falls below the current limit. For the FPF2200 through FPF2202, the minimum current limit is 500 mA with $5 \%$ accuracy.


TOP

## Ordering Information

| Part | Current Limit <br> $(\mathbf{m A})$ | Current Limit <br> Blanking Time <br> $(\mathbf{m S})$ | Auto-Restart Time <br> (mS) | ON Pin <br> Activity |
| :---: | :---: | :---: | :---: | :---: |
| FPF2200 | 500 | 30 | 450 | Active HI |
| FPF2201 | 500 | 30 | NA | Active HI |
| FPF2202 | 500 | NA | NA | Active HI |

## Typical Application Circuit



## Functional Block Diagram



## Pin Configuration



MicroFET 2x2 6L
(BOTTOMVIEW)

## Pin Description

| Pin | Name | Function |
| :---: | :---: | :--- |
| 1 | N/C | No Connection |
| 2 | VIN | Supply Input: Input to the power switch and the supply voltage for the IC |
| 3 | V OUT | Switch Output: Output of the power switch |
| 4 | FLAGB | Fault Output: Active LO, open drain output which indicates an over current, supply <br> under voltage or over temperature state |
| 5 | GND | Ground |
| 6 | ON | ON/OFF Control Input |

## Absolute Maximum Ratings

| Parameter |  | Min. | Max. |
| :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IN}}, \mathrm{V}_{\text {OUT }}$ ON, FLAGB TO GND | -0.3 | 6 | Unit |
| Power Dissipation @ $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 1.2 | V |
| Operating and Storage Junction Temperature | -65 | 125 | ${ }^{\circ} \mathrm{C}$ |
| Thermal Resistance, Junction to Ambient |  |  | 86 |
| Electrostatic Discharge Protection | HBM | 8000 |  |

## Recommended Operating Range

| Parameter | Min. | Max. | Unit |
| :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IN}}$ | 1.8 | 5.5 | V |
| Ambient Operating Temperature, $\mathrm{T}_{\mathrm{A}}$ | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

## Electrical Characteristics

$\mathrm{V}_{1 \mathrm{~N}}=1.8$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}$ unless otherwise noted. Typical values are at $\mathrm{V}_{\mathbb{I N}}=3.3 \mathrm{~V}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Parameter | Symbol | Conditions | Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Basic Operation |  |  |  |  |  |  |
| Operating Voltage | $\mathrm{V}_{\text {IN }}$ |  | 1.8 |  | 5.5 | V |
| Quiescent Current | $\mathrm{I}_{\mathrm{Q}}$ | $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}, \mathrm{~V}_{\text {IN }}=\mathrm{V}_{\text {ON }}=1.8 \mathrm{~V}$ |  | 40 | 65 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}, \mathrm{~V}_{\text {IN }}=\mathrm{V}_{\text {ON }}=3.3 \mathrm{~V}$ |  | 45 | 75 |  |
|  |  | $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}, \mathrm{~V}_{\text {IN }}=\mathrm{V}_{\text {ON }}=5.5 \mathrm{~V}$ |  | 55 | 85 |  |
| $\mathrm{V}_{\text {IN }}$ Shutdown Current |  | $\begin{aligned} & \mathrm{V}_{\mathrm{ON}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{IN}}=5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{OUT}}=\text { short to } \\ & \mathrm{GND} \end{aligned}$ |  |  | 2.5 | $\mu \mathrm{A}$ |


| Parameter | Symbol | Conditions | Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {OUT }}$ Shutdown Current |  | $\begin{aligned} & \mathrm{V}_{\mathrm{ON}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{OUT}}=5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{IN}}=\text { short to } \\ & \mathrm{GND} \end{aligned}$ |  |  | 1 | $\mu \mathrm{A}$ |
| On-Resistance | $\mathrm{R}_{\mathrm{ON}}$ | $\mathrm{V}_{\text {IN }}=5.5 \mathrm{~V}, \mathrm{I}_{\text {OUT }}=200 \mathrm{~mA}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 140 | 185 | $\mathrm{m} \Omega$ |
|  |  | $\mathrm{V}_{\text {IN }}=3.3 \mathrm{~V}, \mathrm{I}_{\text {OUT }}=200 \mathrm{~mA}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 160 | 210 |  |
|  |  | $\mathrm{V}_{\text {IN }}=1.8 \mathrm{~V}, \mathrm{I}_{\text {OUT }}=200 \mathrm{~mA}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 230 | 300 |  |
|  |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=3.3 \mathrm{~V}, \mathrm{I}_{\text {OUT }}=200 \mathrm{~mA}, \\ & \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to } 85^{\circ} \mathrm{C} \end{aligned}$ | 90 |  | 265 |  |
| Output Discharge Resistance |  | $\mathrm{V}_{\mathrm{IN}}=3.3 \mathrm{~V}, \mathrm{~V}_{\text {ON }}=0 \mathrm{~V}, \mathrm{I}_{\text {OUT }}=10 \mathrm{~mA}$ |  | 72 | 105 | $\Omega$ |
| ON Input Logic High Voltage (ON) | $\mathrm{V}_{\mathrm{IH}}$ | $\mathrm{V}_{\mathrm{IN}}=1.8 \mathrm{~V}$ | 0.8 |  |  | V |
|  |  | $\mathrm{V}_{\mathrm{IN}}=5.5 \mathrm{~V}$ | 1.4 |  |  |  |
| ON Input Logic Low Voltage (OFF) | $\mathrm{V}_{\text {IL }}$ | $\mathrm{V}_{\mathrm{IN}}=1.8 \mathrm{~V}$ |  |  | 0.5 | V |
|  |  | $\mathrm{V}_{\text {IN }}=5.5 \mathrm{~V}$ |  |  | 1.0 |  |
| On Input Leakage |  | $\mathrm{V}_{\text {ON }}=\mathrm{V}_{\text {IN }}$ or GND | -1 |  | 1 | $\mu \mathrm{A}$ |
| FLAGB Output Logic Low Voltage |  | $\mathrm{V}_{\text {IN }}=5.5 \mathrm{~V}, \mathrm{I}_{\text {SINK }}=100 \mu \mathrm{~A}$ |  | 0.05 | 0.1 | V |
|  |  | $\mathrm{V}_{\text {IN }}=1.8 \mathrm{~V}, \mathrm{I}_{\text {SINK }}=100 \mu \mathrm{~A}$ |  | 0.12 | 0.25 |  |
| FLAGB Output High Leakage Current |  | $\mathrm{V}_{\mathrm{IN}}=5.5 \mathrm{~V}$, Switch on |  |  | 1 | $\mu \mathrm{A}$ |


| Protections |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Current Limit | ILIM | $\mathrm{V}_{\text {IN }}=3.3 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=3.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | 504 | 530 | 557 | mA |
| Thermal Shutdown |  | Shutdown Threshold |  | 140 |  | ${ }^{\circ} \mathrm{C}$ |
|  |  | Return from Shutdown |  | 130 |  |  |
|  |  | Hysteresis |  | 10 |  |  |
| Under Voltage Shutdown | UVLO | $\mathrm{V}_{\text {IN }}$ increasing | 1.55 | 1.65 | 1.75 | V |
| Under Voltage Shutdown Hysteresis |  |  |  | 50 |  | mV |
| Dynamic |  |  |  |  |  |  |
| Turn On Time | $\mathrm{t}_{\mathrm{ON}}$ | $\mathrm{R}_{\mathrm{L}}=500 \Omega, \mathrm{C}_{\mathrm{L}}=0.001 \mathrm{uF}$ |  | 70 |  | $\mu \mathrm{S}$ |
| Turn Off Time | toff | $\mathrm{R}_{\mathrm{L}}=500 \Omega, \mathrm{C}_{\mathrm{L}}=0.001 \mathrm{uF}$ |  | 600 |  | nS |
| $\mathrm{V}_{\text {OUT }}$ Rise Time | $\mathrm{t}_{\text {RISE }}$ | $\mathrm{R}_{\mathrm{L}}=500 \Omega, \mathrm{C}_{\mathrm{L}}=0.001 \mathrm{uF}$ |  | 40 |  | $\mu \mathrm{S}$ |
| $\mathrm{V}_{\text {OUT }}$ Fall Time | $\mathrm{t}_{\text {FALL }}$ | $\mathrm{R}_{\mathrm{L}}=500 \Omega, \mathrm{C}_{\mathrm{L}}=0.001 \mathrm{uF}$ |  | 100 |  | nS |
| Over Current Blanking Time | $\mathrm{t}_{\text {bLANK }}$ | FPF2200, FPF2201 | 15 | 30 | 60 | mS |
| Auto-Restart Time | $\mathrm{t}_{\text {RSTRT }}$ | FPF2200 | 225 | 450 | 900 | mS |
| Current Limit Response Time |  | $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{ON}}=3.3 \mathrm{~V}$. Over-Current Condition: $\mathrm{R}_{\mathrm{LOAD}}=1.55 \Omega$ |  | 5 |  | $\mu \mathrm{S}$ |

## Typical Characteristics



Figure 1. Quiescent Current vs. Input Voltage


Figure 3. $\mathrm{V}_{\mathrm{ON}}$ vs. Input Voltage


Figure 5. V $\mathrm{ON}_{\mathrm{ON}}$ Low Voltage vs. Temperature


Figure 2. Quiescent Current vs. Temperature


Figure 4. $\mathrm{V}_{\mathrm{ON}}$ High Voltage vs. Temperature


Figure 6. On Pin Current vs. Temperature

## Typical Characteristics



Figure 7. Current Limit vs. Output Voltage


Figure 9. Ron vs. Input Voltage


Figure 11. $\mathrm{T}_{\mathrm{ON}} / \mathrm{T}_{\text {OFF }}$ vs. Temperature


Figure 8. Current Limit vs. Temperature


Figure 10. $\mathrm{R}_{\mathrm{ON}}$ vs. Temperature


Figure 12. $\mathrm{T}_{\text {RISE }} / \mathrm{T}_{\text {FALL }}$ vs. Temperature

## Typical Characteristics



Figure 13. $\mathrm{T}_{\text {BLANK }}$ vs Temperature


Figure 15. Toff Response


Figure 17. Current Limit Response (Output is loaded with $1.2 \Omega$ resistor and $\mathrm{C}_{\mathrm{OUT}}=10 \mu \mathrm{~F}$ )


Figure 14. $\mathrm{T}_{\mathrm{ON}}$ Response


Figure 16. TBLANK Response


Figure 18. Current Limit Response (Output is loaded with $1.2 \Omega$ resistor and $\mathrm{C}_{\mathrm{OUT}}=100 \mu \mathrm{~F}$ )

## Typical Characteristics



Figure 19. Current Limit Response (Switch is powered into a short - Input and enable pins are tied together)


Figure 20. Current Limit Response
(Output shorted to GND while the switch is in normal operation)

## Description of Operation

The FPF2200, FPF2201, and FPF2202 are state of the art High Precision Current Limit switches designed to meet USB OTG (On-The-Go) applications with optimum current for a safe design practice. The core of each device is a $0.16 \Omega$ P-channel MOSFET and a controller capable of functioning over an input operating range of $1.8-5.5 \mathrm{~V}$. The controller protects or offers current limiting, UVLO(undervoltage lockout) and thermal shutdown protection. The minimum current limit value is set to 500 mA allowing to draw as much as 500 mA from the USB port.

## On/Off Control

The ON pin is active high, and controls the state of the switch. Applying a continuous high signal will hold the switch in the ON state. The switch will move into the OFF state when the active high is removed, or if a fault is encountered. For all versions, an undervoltage on $\mathrm{V}_{\text {IN }}$ or a junction temperature in excess of $140^{\circ} \mathrm{C}$ overrides the ON control to turn off the switch. In addition, excessive currents will cause the switch to turn off in the FPF2200 and FPF2201. The FPF2200 has an Auto-Restart feature which will automatically turn the switch ON again after 450 ms . For the FPF2201, the ON pin must be toggled to turn-on the switch again. The FPF2202 does not turn off in response to an over current condition but instead remains operating in a constant current mode so long as ON is active and the thermal shutdown or UVLO have not activated.

## Fault Reporting

Upon the detection of an over-current condition, an input UVLO, or an over-temperature condition, the FLAGB signals the fault mode by activating LO. In the event of an over-current condition for the FPF2200 and FPF2201, the FLAGB goes LO at the end of the blanking time while FLAGB goes LO immediately for the FPF2202. If the over-current condition lasts longer than blanking time, FLAGB remains LO through the Auto-Restart Time for the FPF2200 while for the FPF2201, FLAGB is latched LO and ON must be toggled to release it. With the FPF2202, FLAGB is LO during the faults and immediately returns HI at the end of the fault condition. FLAGB is an open-drain MOSFET which requires a pull-up resistor between VIN and FLAGB. During shutdown, the pull-down on FLAGB is disabled to reduce current draw from the supply. A $100 \mathrm{~K} \Omega$ pull up resistor is recommended to be used in the application.

## Current Limiting

The current limit ensures that the current through the switch doesn't exceed a maximum value while not limiting at less than a minimum value. The minimum current at which the parts will limit is set to 500 mA . The FPF2200 and FPF2201 have a blanking time of 30 ms (nominal) during which the switch will act as a constant current source. At the end of the blanking time, the switch will be turned-off. The FPF2202 has no current limit blanking period so it will remain in a constant current state until the ON pin is deactivated or the thermal shutdown turns-off the switch.

## Undervoltage Lockout (UVLO)

The undervoltage lockout turns-off the switch if the input voltage drops below the undervoltage lockout threshold. With the ON pin active the input voltage rising above the undervoltage lockout threshold will cause a controlled turn-on of the switch which limits current over-shoots.

## Output Discharge Resistor

The FPF2200/1/2 family contains an $80 \Omega$ on-chip load resistor for quick output discharge when the switch is turned off. This features become more attractive when application requires large output capacitor to be discharged when switch turns-off. However, $\mathrm{V}_{\text {OUT }}$ pin should not be connected directly to the battery source due to the discharge mechanism of the load switch.

## Thermal Shutdown

The thermal shutdown protects the die from internally or externally generated excessive temperatures. During an overtemperature condition the FLAGB is activated and the switch is turned-off. The switch automatically turns-on again if temperature of the die drops below the threshold temperature.

## Application Information

## Input Capacitor

To limit the voltage drop on the input supply caused by transient in-rush currents when the switch is turned on into a discharged load capacitor or a short-circuit, a capacitor is recommended to be placed between $\mathrm{V}_{\mathrm{IN}}$ and GND. A $1 u F$ ceramic capacitor, $\mathrm{C}_{\mathrm{IN}}$, placed close to the pins is usually sufficient. Higher values of $\mathrm{C}_{\mathrm{IN}}$ can be used to further reduce the voltage drop.

## Output Capacitor

A 0.1 uF capacitor $\mathrm{C}_{\text {OUT }}$, should be placed between $\mathrm{V}_{\text {OUT }}$ and GND. This capacitor will prevent parasitic board inductances from forcing $\mathrm{V}_{\text {OUT }}$ below GND when the switch turns-off. For the FPF2200 and FPF2201, the total output capacitance needs to be kept below a maximum value, $\mathrm{C}_{\text {OUT }}(\max )$, to prevent the part from registering an over-current condition and turning-off the switch. The maximum output capacitance can be determined from the following formula:

$$
\mathrm{C}_{\text {OUT (Max) }}=\frac{\mathrm{I}_{\text {LIM (Max) }} \mathrm{X}_{\mathrm{BLANK} \text { (Min) }}}{\mathrm{V}_{\text {IN }}}
$$

## Power Dissipation

During normal on-state operation, the power dissipated in the device will depend upon the level at which the current limit is set. The maximum allowed setting for the current limit is 500 mA and will result in a power dissipation of:

$$
\mathrm{P}=\left(\mathrm{I}_{\mathrm{LIM}}\right)^{2} * \mathrm{R}_{\mathrm{ON}}=(0.5)^{2} * 0.16=40 \mathrm{~mW}
$$

If the part goes into current limit, the maximum power dissipation will occur when the output is shorted to ground. For the FPF2200, the power dissipation will scale by the AutoRestart Time, $\mathrm{t}_{\text {RSTRT, }}$, and the Over Current Blanking Time, $t_{\text {BLANK }}$, so that the maximum power dissipated is:

$$
\begin{aligned}
\mathrm{P}_{\text {(Max) }} & =\frac{\mathrm{t}_{\text {BLANK }}}{\mathrm{t}_{\text {BLANK }}+\mathrm{t}_{\text {RSTRT }}} * \mathrm{~V}_{\text {IN (Max) }}{ }^{*} \mathrm{I}_{\mathrm{LIM}(\text { Max })} \\
& =\frac{30}{30+450} * 5.5 * 0.5=0.17 \mathrm{~W}
\end{aligned}
$$

Note this is below the maximum package power dissipation, and the thermal shutdown feature will act as additional safety to protect the part from damage due to excessive heating. The junction temperature is only able to increase to the thermal shutdown threshold. Once this temperature has been reached, toggling ON will not turn-on the switch until the junction temperature drops. For the FPF2202, a short on the output will cause the part to operate in a constant current state dissipating a worst case power of:

$$
P_{(\operatorname{Max})}=V_{I N(M A X)}{ }^{*} \operatorname{I}_{\operatorname{LIM}(\operatorname{MAX})}=5.5 * 0.557=3.064 W
$$

This large amount of power will activate the thermal shutdown and the part will cycle in and out of thermal shutdown so long as the ON pin is active and the short is present.

## PCB Layout Recommendations

For best performance, all traces should be as short as possible. To be more effective, the input and output capacitors should be placed close to the device to minimize the effects that parasitic trace inductances may have on normal and short-circuit operation. Using wide traces for $\mathrm{V}_{\text {IN }}, \mathrm{V}_{\text {OUT }}$ and GND will help minimize parasitic electrical effects along with minimizing the case to ambient thermal impedance.

## Improving Thermal Performance

An improper layout could result in higher junction temperature and triggering the thermal shutdown protection feature. This concern applies when the switch is set at higher current limit value and an over-current condition occurs. In this case, the power dissipation of the switch, from the formula below, could exceed the maximum absolute power dissipation of 1.2 W .

$$
\mathrm{PD}=\left(\mathrm{V}_{\mathrm{IN}}-\mathrm{V}_{\mathrm{OUT}}\right) \times \mathrm{I}_{\mathrm{LIM}} \text { (Max) }
$$

The following techniques have been identified to improve the thermal performance of this family of devices. These techniques are listed in order of the significance of their impact.

1. Thermal performance of the load switch can be improved by connecting pin7 of the DAP (Die Attach Pad) to the GND plane of the PCB.
2. Embedding two exposed through-hole vias into the DAP (pin7) provides a path for heat to transfer to the back GND plane of the PCB. A drill size of Round, 14 mils ( 0.35 mm ) with 1-ounce copper plating is recommended to result in appropriate solder reflow. A smaller size hole prevents the solder from penetrating into the via, resulting in device lift-up. Similarly, a larger via-hole consumes excessive solder, and may result in voiding of the DAP.


Figure 21: Two through hole open vias embedded in DAP
3. The $\mathrm{V}_{\mathrm{IN}}, \mathrm{V}_{\text {OUT }}$ and GND pins will dissipate most of the heat generated during a high load current condition. The layout suggested in Figure 23 provides each pin with adequate copper so that heat may be transferred as efficiently as possible out of the device. The low-power FLAGB and ON pin traces may be laid-out diagonally from the device to maximize the area available to the ground pad. Placing the input and output capacitors as close to the device as possible also contributes to heat dissipation, particularly during high load currents.


Figure 22: X-Ray result (bottom view with $45^{\circ}$ angle)


Figure 23: Proper layout of output and ground copper area

## FPF22XX Demo Board

FPF22XX Demo board has components and circuitry to demonstrate FPF2223/4/5 load switches functions and features. R4 resistor with $0 \Omega$ value is used for measuring the output current. Load current can be scoped by removing the R4 resistor and soldering a current loop to the R4 footprint. Thermal performance of the board is improved using a few techniques recommended in the layout recommendations section of datasheet. R3 resistor should be left open for FPF220X family.


Figure 24: Top, SST, and AST Layers


Figure 25: Bottom and ASB Layers
Figure 26: Zoom in to Top layer



#### Abstract

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