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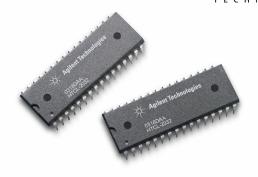
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## HCTL-2032, HCTL-2032-SC, HCTL-2032-SCT, HCTL-2022

Quadrature Decoder/ Counter Interface ICs

## **Data Sheet**



## Description

The HCTL-20XX-XX is CMOS ICs that perform the quadrature decoder, counter, and bus interface function. The HCTL-20XX-XX is designed to improve system performance in digital closed loop motion control systems and digital data input systems. It does this by shifting time intensive quadrature decoder functions to a cost effective hardware solution. The HCTL-20XX-XX consists of a quadrature decoder logic, a binary up/down state counter, and an 8-bit bus interface. The use of Schmitt-triggered CMOS inputs and input noise filters allows reliable operation in noisy environments. The HCTL-20XX-XX contains 32-bit counter and provides LSTLL compatible tri-state output buffers. Operation is specified for a temperature range from -40 to +100°C at clock frequencies up to 33MHz.

The HCTL-2032 and HCTL-2032-SC have dual-axis capability and index channel support. Both devices can be programmed as 4x/2x/1x count mode. The HCTL-2032 and HCTL2032-SC also provides quadrature decoder output signals and cascade signals for use with many standard computer ICs.

The HCTL-2022 has most of the HCTL-2032 features, but it can only supports single axis and fixed at 4x count mode. The HCTL-2022 doesn't provide decoder output and cascade signals.

#### Features

- Interfaces Encoder to Microprocessor
- 33 MHz Clock Operation
- Programmable Count Modes (1x, 2x or 4x)
- Single or Dual Axis Support
- Index Channel Support
- High Noise Immunity:
- Schmitt Trigger Inputs and Digital Noise Filter
- 32-Bit Binary Up/Down Counter
- Latched Outputs
- 8-Bit Tristate Interface
- 8, 16, 24, or 32-Bit Operating Modes
- Quadrature Decoder Output Signals, Up/Down and Count
- Cascade Output Signals, Up/Down and Count
- Substantially Reduced System Software
- 5V Operation (V<sub>DD</sub> V<sub>SS</sub>)
- TTL/CMOS Compatible I/O
- Operating Temperature: -40°C to 100°C
- 32-Pin PDIP, 32-Pin SOIC, 20-Pin PDIP

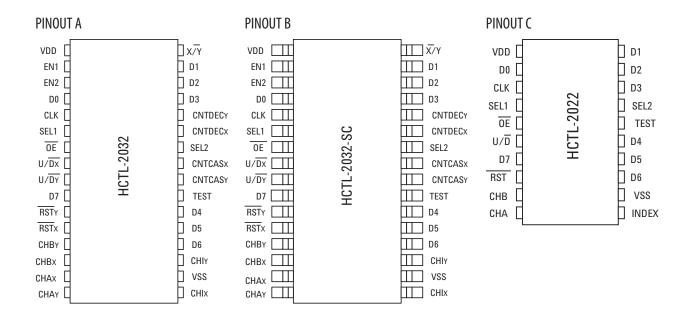
#### Applications

- Interface Quadrature Incremental Encoders to Microprocessors
- Interface Digital Potentiometers to Digital Data Input Buses

ESD WARNING: Standard CMOS handling precautions should be observed with the HCTL-2032 family ICs.

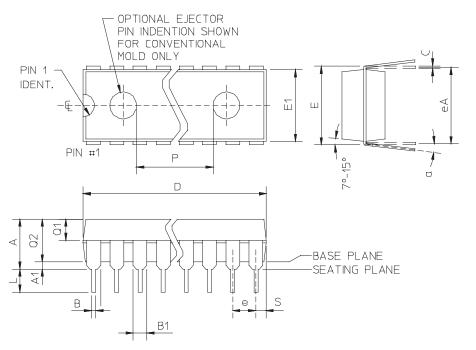
Devices

Part Number	Description	Package Drawing
HCTL-2032	32-bit counter, dual axis, decoder and cascade outputs, index channel support, programmable count modes, and 33 Mhz clock operation.	A
HCTL-2032-SC	All features of HCTL-2032.	В
HCTL-2022	Most of the HCTL-2032 features. The device supports single axis, and no decoder output and cascade signals. The programmable count mode is set to 4x internally.	C

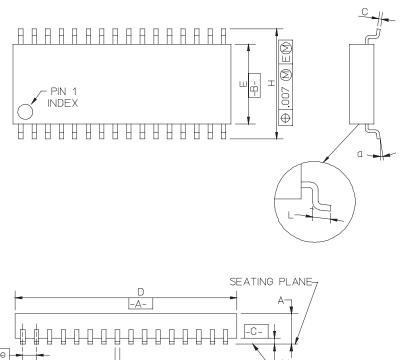


## Package Dimensions (dimensions in inches)

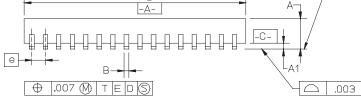
1) HCTL - 2032



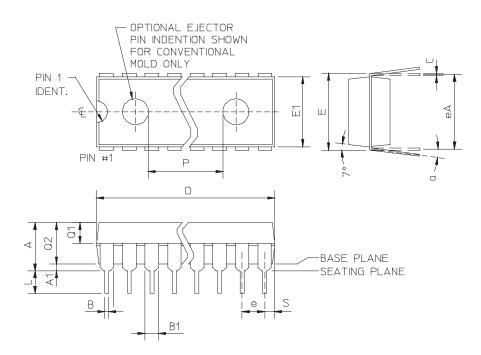
SYMBOL	MIN.	NOM.	MAX.
A	-	-	-
A1	-	-	-
В	.016	.018	.020
B1	.045	.050	.055
C	-	.010	-
D	1.640	1.650	1.660
E	.590	.610	.630
E1	.546	.550	.554
e	.1	ίοο τγρ	
eA		-	
L	.100	-	-
a	-	-	-
Q1	.066	.070	.074
02	-	-	-
S	-	_	-



SYMBOL	MIN.	NOM.	MAX.
A	.090	.095	,100
A1	.004	.007	,010
В	.014	.016	.020
С	.006	.008	.0125
θ		)50 BSC	
D	.810	.818	.822
E	.292	.296	.299
L	.DZ4	.032	.040
Н	.405	.412	.419
	D°	4°	8°

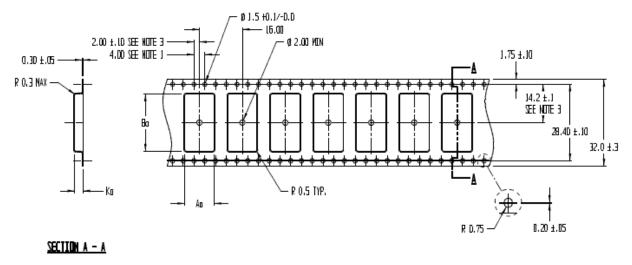


## 3) HCTL - 2022



SYMBOL	MIN.	NOM.	MAX.
A	-	-	.170
A1	.015	-	-
В	.015	.018	.022
B1	.055	.060	.065
С	.008	.010	.012
D	1.015	1.020	1.025
E	.300	-	.325
E1	.250	.260	.270
E	.1	00 BSC.	
eA	Ε.	00 BSC.	
L	.125	-	.135
α	0°	-	15°
P	.710	-	-
Q1	.060	.065	.070
02	-	.130	-
S	.055	.060	.065

## 4) HCTL-2032 – SCT (Tape and Reel Version of HCTL-2032-SC)



		10.00
80	_	

- Bo = 21,20 Ko = 3.10

Notes:

- 1. 10 Sprocket hole pitch cumulative tolerance 0.2
- 2. Camber in compliance with EIA 481
- 3. Pocket position relative to sprocket hole measured as true position of pocket, not pocket hole
- 4. All dimensions in mm

## **Operating Characteristics**

## **Table 1. Absolute Maximum Ratings**

(All voltages below are referenced to V<sub>SS</sub>)

Parameter	Symbol	Limits	Units
DC Supply Voltage	V <sub>DD</sub>	-0.3 to +6.0	V
Input Voltage	V <sub>IN</sub>	-0.3 to (V <sub>DD</sub> +0.3)	V
Storage Temperature	Ts	-55 to +150	°C
Operating Temperature [1]	T <sub>A</sub>	-40 to +100	°C

## Table 2. Recommended Operating Conditions

Parameter	Symbol	Limits	Units
DC Supply Voltage	V <sub>DD</sub>	4.5 to 5.5	V
Ambient Temperature <sup>[1]</sup>	T <sub>A</sub>	-40 to +100	°C

## Table 3. DC Characteristics $V_{DD}$ = 5V $\pm$ 5%; $T_A$ = -40 to 100°C

Symbol	Parameter	Condition	Min	Тур	Max	Unit
V <sub>IL</sub> <sup>[2]</sup>	Low-Level Input Voltage				1.5	V
V <sub>IH</sub> [2]	High-Level Input Voltage		3.5			V
V <sub>T+</sub>	Schmitt-Trigger Positive-Going Threshold			3.5	4.0	V
V <sub>T-</sub>	Schmitt-Trigger Negative-Going Thresh- old		1.0	1.5		V
V <sub>H</sub>	Schmitt-Trigger Hysteresis		1.0	2.0		V
I <sub>IN</sub>	Input Current	$V_{IN}=V_{SS}$ or $V_{DD}$	-10	1	+10	μΑ
V <sub>OH</sub> <sup>[2]</sup>	High-Level Output Voltage	l <sub>OH</sub> = -3.75 mA	2.4	4.5		V
V <sub>OL</sub> [2]	Low-Level Output Voltage	$I_{OL} = +3.75 \text{mA}$		0.2	0.4	V
I <sub>OZ</sub>	High-Z Output Leakage Current	$V_{O}=V_{SS} \text{ or } V_{DD}$	-10	1	+10	μA
I <sub>DD</sub>	Quiescent Supply Current	$V_{IN}$ =Vss or $V_{DD}$		1	10	μA
C <sub>IN</sub> [3]	Input Capacitance	Any Input		5		pF
C <sub>OUT</sub> <sup>[3]</sup>	Output Capacitance	Any Output		5		pF

Notes

1. Free Air

2. In general, for any V<sub>DD</sub> between the allowable limits (+4.5V to +5.5V),  $V_{IL} = 0.3V_{DD}$  and  $V_{IH} = 0.7V_{DD}$ ;

typical values are  $V_{OH} = V_{DD} - 0.5V$  and  $V_{OL} = V_{SS} + 0.2V$ 3. Including package capacitance

## **Functional Pin Description**

## Table 4. Functional Pin Descriptions

	Pin								
Symbol	HCTL 2032/ 2032-SC	HCTL 2022	_ Description						
V <sub>DD</sub>	1	1	Power Supply						
V <sub>SS</sub>	18	12	Ground						
CLK	5	3	CLK is a Schmitt	trigge:	r input for th	ne external o	clock signal		
CHA <sub>X</sub>	15	10	CHA <sub>X</sub> , CHA <sub>Y</sub> , CH						
CHAY	16	NC	from a quadratu					•	
CHB <sub>X</sub>	14	9	channels, A and are the 1st axis			-	•	required. CH	$A_X$ and $CHB_X$
CHBY	13	NC	are the 15t axis			ure the 2nd	a axis.		
CHIX	17	11	$CHI_X$ and $CHI_Y$ a	re Schn	nitt-trigger i	nputs that a	accept the o	outputs of In	dex channel
CHIY	19	NC	from an increm	ental op	otical shaft e	ncoder.			
RSTN <sub>X</sub>	12	8	This active low	Schmit	t-trigger inp	out clears th	e internal j	position cou	nter and the
RSTN <sub>Y</sub>	11	NC	position latch. It			-			
			respect to any o to reset the 2nd		-	$RST_X$ is to re	eset the 1st	axis counter	and RST <sub>Y</sub> / is
OEN	7	5				c tha tri stat		ufforg The O	E/ CEL 1 and
UEIN	/	5	This CMOS activ SEL2 inputs are						
			to control the lo	•			0		
SEL <sub>1</sub>	6	4	These CMOS in	outs dire	ectly contro	Is which dat	ta byte fron	n the positio	n latch is en-
SEL <sub>2</sub>	26	17	abled into the 8		•	buffer. As in	n OÉ/ above	e, SEL <sub>1</sub> and S	EL <sub>2</sub> also con-
			trol the internal	inhibit	logic.				
						BYTE SEL			
			SEL1 S	EL2	MSB	2ND	3RD	LSB	
			0	1	D4				
			1	1		D3			
			0	0			D2	D1	
				0				D1	
EN <sub>1</sub>	2	NC	These CMOS con			high or low	to activate	the selected	count mode
EN <sub>2</sub>	3	NC	before the deco	aing be	egins.				
						Count Mode	es	7	
			EN1	EN2	4x	2x	1x	1	
			0	0		Illegal Mode	9		
			1	0	On			4	
			0	<u>1</u> 1		On	On	-	
				•				<u></u>	
X/Y	32	NC	Select the 1 <sup>st</sup> o			e read. Low	bit enable	s the 1 <sup>st</sup> axi	s data, while
	27	NC	high bit enables			compostible			
CNTDECX	27 28		A pulse is prese er (4x/2x/1x) ha						
CNTDECY	28	NC	is for 2 <sup>nd</sup> axis.						
U/Dx	8	6	This LSTTL-com	patible	output allow	vs the user t	o determin	e whether th	e IC is count-
U/Dy	9	NC	ing up or down	and is i	ntended to	be used wit	h the CNTD	EC and CNT	CAS outputs.
			The proper sign				el) will be	present befo	ore the rising
			edge of the CN	IDEC an	IU CIVICAS (	outputs.			

		_	
CNTCAS <sub>X</sub>	25	NC	A pulse is presented on this LSTTL-compatible output when the HCTL-2032 / 2032-
			SC internal counter overflows or underflows. The rising edge on this waveform may
CNTCASY	24	NC	be used to trigger an external counter.
TEST	23	16	This pin is used for internal testing. Tied it to ground or leave it floating for normal
			operation.
D0	4	2	These LSTTL-compatible tri-state outputs form an 8-bit output ports through which
D1	31	20	the contents of the 32-bit position latch may be read in 4 sequential bytes. The MSB
D2	30	19	<ul> <li>is read first followed by the rest of the bytes with the LSB is read last.</li> </ul>
D3	29	18	
D4	22	15	
D5	21	14	_
D6	20	13	-
D7	10	7	-

## Switching Characteristics

## Table 5. Switching Characteristics

Max/Min specifications at  $V_{DD}=5V\pm5\%; T_A=-40$  to 100°C,  $C_L=40~pf$ 

Symb	ol	Description	Min.	Max.	Units
1	t <sub>CLK</sub>	Clock Period		33	MHz
2	t <sub>CHH</sub>	Pulse width, clock high	1/f		ns
3	t <sub>CD</sub>	Delay time, rising edge of clock to valid, updated count information on D0-7		31	ns
4	t <sub>ODE</sub>	Delay time, OEN fall to valid data		29	ns
5	t <sub>ODZ</sub>	Delay time, OEN rise to Hi-Z state on D0-7		29	ns
6	t <sub>SDV</sub>	Delay time, SEL0~SEL1 valid to stable, selected data byte		29	ns
		(delay to High Byte = delay to Low Byte)			
7	t <sub>XNYDV</sub>	Delay time, XNY valid to stable, selected data byte.		29	ns
8	t <sub>CLH</sub>	Pulse width, clock low	15		ns
9	tss	Setup time, SEL1~SEL2 before clock fall	12		ns
10	tos	Setup time, OEN before clock fall	12		ns
11	t <sub>XNYS</sub>	Setup time, XNY before clock fall	12		ns
12	t <sub>SH</sub>	Hold time, SEL1~SEL2 after clock fall	0		ns
13	t <sub>OH</sub>	Hold time, OEN after clock fall	0		ns
14	t <sub>XNYH</sub>	Hold time, XNY after clock fall	0		ns
15	t <sub>RST</sub>	Pulse width, RSTNX~RSTNY low	10		ns
16	t <sub>DCD</sub>	Hold time, last position count stable on D0-7 after clock rise	2		ns
17	t <sub>DSD</sub>	Hold time, last data byte stable after next SEL state change	2		ns
18	t <sub>DOD</sub>	Hold time, data byte stable after OEN rise	2		ns
19	t <sub>DXNYD</sub>	Hold time, data byte stable after XNY change	2		ns
20	t <sub>UDDX</sub>	Delay time, U/DNX valid after clock rise	4	29	ns
21	t <sub>UDDY</sub>	Delay time, U/DNY valid after clock rise	4	29	ns
22	t <sub>CHXD</sub>	Delay time, CNTDECX or CNTCASX high after clock rise	4	31	ns
23	t <sub>CHYD</sub>	Delay time, CNTDECY or CNTCASY high after clock rise	4	31	ns
24	t <sub>CLXD</sub>	Delay time, CNTDECX or CNTCASX low after clock fall	4	31	ns
25	t <sub>CLYD</sub>	Delay time, CNTDECY or CNTCASY low after clock fall	4	31	ns
26	t <sub>UDXH</sub>	Hold time, U/DNX stable after clock rise	2		ns
27	t <sub>UDYH</sub>	Hold time, U/DNY stable after clock rise	2		ns
28	t <sub>UDCXS</sub>	Setup time, U/DNX valid before CNTDECX or CNTCASX rise	Note 1		ns
29	t <sub>UDCYS</sub>	Setup time, U/DNY valid before CNTDECY or CNTCASY rise	Note 1		ns
30	t <sub>UDCXH</sub>	Hold time, U/DNX stable after CNTDECX or CNTCASX rise	Note 2		ns
31	tudcyh	Hold time, U/DNY stable after CNTDECY or CNTCASY: rise	Note 2		ns

#### Notes

1. tclk - max delay (item 20/21) + min delay (item 22/23) 2. tclk - max delay (item 22/23) + min delay (item 20/21)

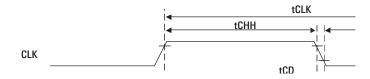
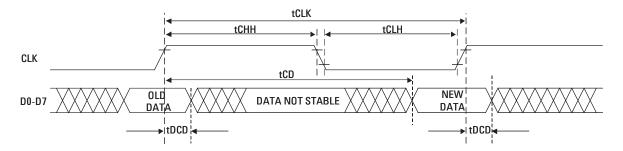
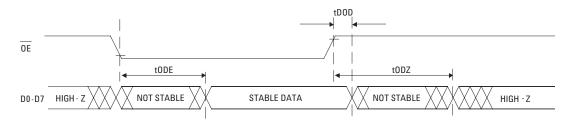


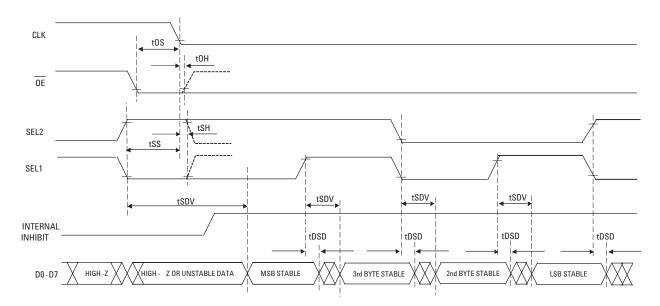
Figure 1. Reset Waveform











#### Figure 4. Bus Control Timing

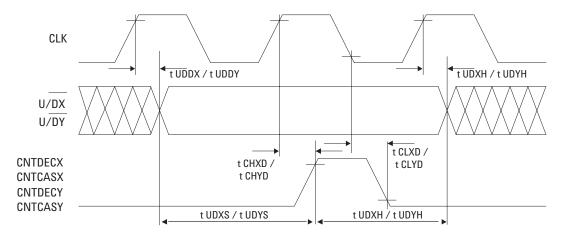


Figure 5. Decoder, Casade Output Timing

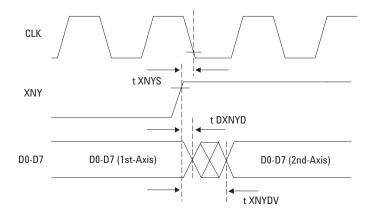


Figure 6. Output Data from 1<sup>st</sup>-axis and 2<sup>nd</sup>-axis

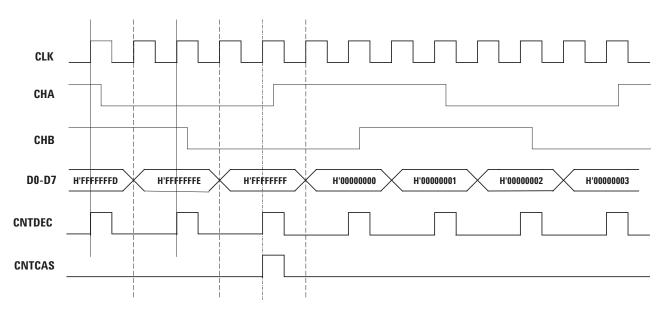


Figure 7. Quadrature decoder for 1<sup>st</sup>-axis/ 2<sup>nd</sup>-axis (4x count mode)

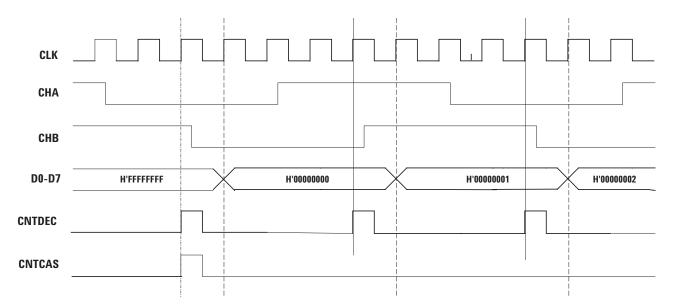


Figure 8. Quadrature decoder for 1<sup>st</sup>-axis/ 2<sup>nd</sup>-axis (2x count mode)

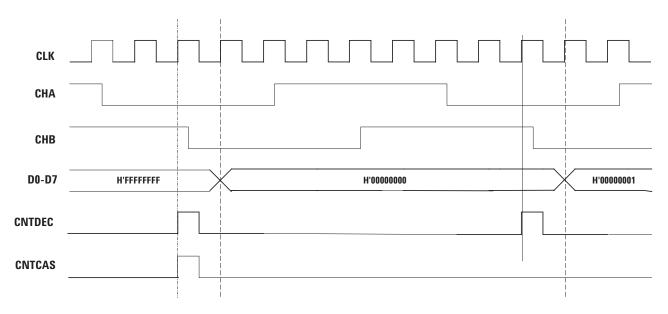


Figure 9. Quadrature decoder for 1<sup>st</sup>-axis/ 2<sup>nd</sup>-axis (1x count mode)

## Operation

A block diagram of the HCTL-20XX-XX family is shown in Figure 10. The operation of each major function is described in the following sections.

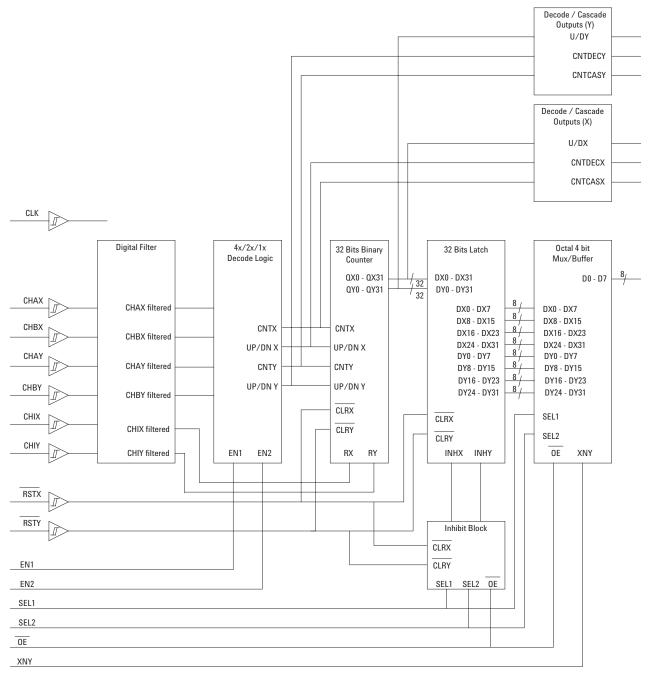


Figure 10. Simplified Logic Diagram

## **Digital Noise Filter**

The digital noise filter section is responsible for rejecting noise on the incoming quadrature signals. The input section uses two techniques to implement improved noise rejection. Schmitt-trigger inputs and a three-clock-cycle delay filter combine to reject low level noise and large, short duration noise spikes that typically occur in motor system applications. Both common mode and differential mode noise are rejected. The user benefits from these techniques by improved integrity of the data in the counter. False counts triggered by noise are avoided.

Figure 11 shows the simplified schematic of the input section. The signals are first passed through a Schmitt-trigger buffer to address the problem of input signals with slow rise times and low-level noise (approximately < 1V). The cleaned up signals are then passed to a four-bit delay filter. The signals on each channel are sampled on rising clock edges. A time history of the signals is stored in the four-bit shift register. Any change on the input is tested for a stable level being present for three consecutive rising clock edges. Therefore, the filtered output waveforms can change only after an input level has the same value for three consecutive rising clock edges.

Refer to Figure 12, which shows the timing diagram. The result of this circuitry is that short noise spikes between rising clock edges are ignored and pulses shorter than two clock periods are rejected.

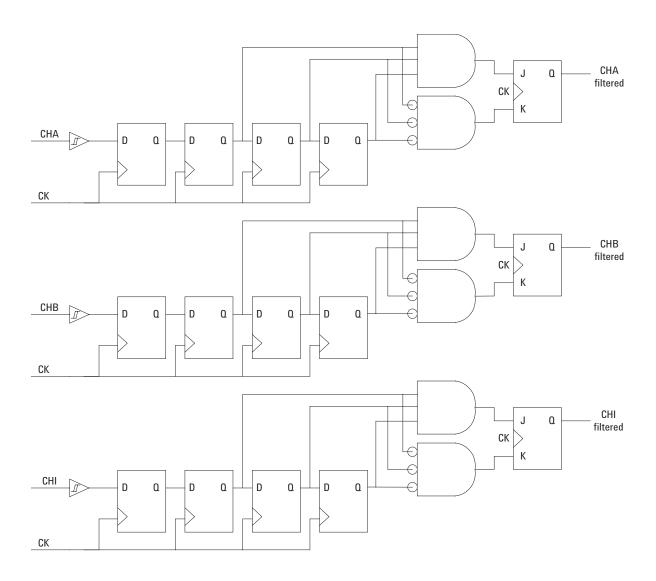


Figure 11. Simplified Digital Noise Filter Logic

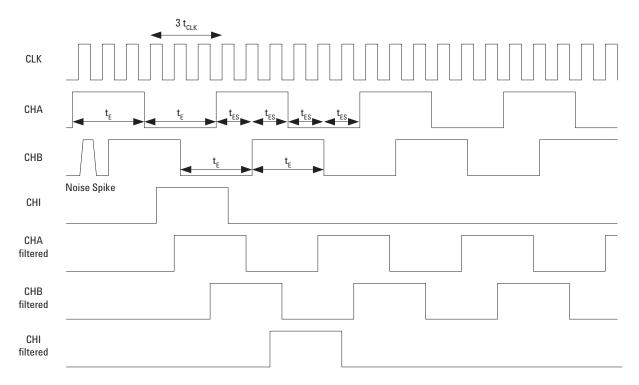
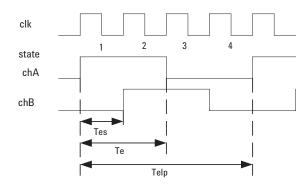


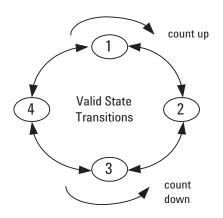
Figure 12. Signal Propagation through Digital Noise Filter

#### **Quadrature Decoder**

The quadrature decoder decodes the incoming filtered signals into count information. This circuitry multiplies the resolution of the input signals by a factor of one, two or four (1X, 2X, 4X decoding) depending on the resolution mode. When using an encoder for motion sensing, the user benefits from the selectable resolution by being able to provide better system control.

The quadrature decoder samples the outputs of the CHA and CHB filters. Based on the past binary state of the two signals and the present state, it outputs a count signal and a direction signal to the integral position counter. Figure 13 shows the quadrature states of Channel A and Channel B signals. The?4x decoder mode will output a count signal for every state transition (count up and count down). Figure 14 shows the valid state transitions for 2x and 1x decoder modes. The 2x/1x decoder will output a count signal at respective state transition, depending on the counting direction. Channel A leading channel B results in counting up. Channel B leading channel A results in counting down. Illegal state transitions, caused by faulty encoders or noise severe enough to pass through the filter, will produce an erroneous count.





			4X Decoder
CHA	CHB	STATE	(Count Up & Count Down)
1	0	1	Pulse
1	1	2	Pulse
0	1	3	Pulse
0	0	4	Pulse

Figure 13. 4x Decoder Mode

			2x	2x	1x	1x
CHA	CHB	STATE	Count Up	Count Down	Count Up	Count Down
1	0	1	Pulse	-	Pulse	-
1	1	2	-	Pulse	-	Pulse
0	1	3	Pulse	-	-	-
0	0	4	-	Pulse	-	-

Figure 14. 2x and 1x Decoder Modes

#### **Design Considerations**

The designer should be aware that the operation of the digital filter places a timing constraint on the relationship between incoming guadrature signals and the external clock. Figure 12 shows the timing waveform with an incremental encoder input. Since an input has to be stable for three rising clock edges, the encoder pulse width (t<sub>F</sub> - low or high) has to be greater than three clock periods (3t<sub>CLK</sub>). This guarantees that the asynchronous input will be stable during three consecutive rising clock edges. A realistic design also has to take into account finite rise time of the waveforms, asymmetry of the waveforms, and noise. In the presence of large amounts of noise, t<sub>F</sub> should be much greater than 3t<sub>CLK</sub> to allow for the interruption of the consecutive level sampling by the threebit delay filter. It should be noted that a change on the inputs that is qualified by the filter will internally propagate in a maximum of seven clock periods.

The quadrature decoder circuitry imposes a second timing constraint between the external clock and the input signals. There must be at least one clock period between consecutive quadrature states. As shown in Figure 13, a quadrature state is defined by consecutive edges on both channels. Therefore, t<sub>ES</sub> (encoder state period) > t<sub>CLK</sub>. The designer must account for deviations from the nominal 90 degree phasing of input signals to guarantee that t<sub>ES</sub> > t<sub>CLK</sub>.

#### **Position Counter**

This section consists of a 32-bit (HCTL-20XX-XX) binary up/down counter which counts on rising clock edges as explained in the Quadrature Decoder Section. All 32 bits of data are passed to the position data latch. The system can use this count data in several ways:

- A. System total range is 32 bits, so the count represents "absolute" position.
- B. The system is cyclic with 32 bits of count per cycle. RST/ is used to reset the counter every cycle and the system uses the data to interpolate within the cycle.
- C. System count is >8, 16, 24, or 32 bits, so the count data is used as a relative or incremental position input for a system software computation of absolute position. In this case counter rollover occurs. In order to prevent loss of position information, the processor must read the outputs of the IC before the count increments one-half of the maximum count capability. Two'scomplement arithmetic is normally used to compute position from these periodic position updates.
- D. The system count is >32 bits so the HCTL-2032 / 2032-SC can be cascaded with other standard counter ICs to give absolute position.

## **Position Data Latch**

The position data latch is a 32-bit latch which captures the position counter output data on each rising clock edge, except when its inputs are disabled by the inhibit logic section during four-byte read operations. The output data is passed to the bus interface section. When active, a signal from the inhibit logic section prevents new data from being captured by the latch, keeping the data stable while successive reads are made through the bus section. The latch is automatically re-enabled at the end of these reads. The latch is cleared to 0 asynchronously by the RST signal.

#### **Inhibit Logic**

The Inhibit Logic Section samples the OE, SEL1 and SEL2 signals on the falling edge of the clock and, in response to certain conditions (see Figure 15), inhibits the position data latch. The RST signal asynchronously clears the inhibit logic, enabling the latch.

#### **Bus Interface**

The bus interface section consists of a 32 to 8 line multiplexer and an 8-bit, three-state output buffer. The multiplexer allows independent access to the low and high bytes of the position data latch. The SEL1, SEL2 and OE signals determine which byte is output and whether or not the output bus is in the high-Z state. In the HCTL-20XX-XX, the data latch is 32 bit wide.

## Quadrature Decoder Output (HCTL-2032 / 2032-SC only)

The quadrature decoder output section consists of count and up/down outputs derived from the 4x/2x/1x decoder mode of the HCTL-2032 / 2032-SC. When the decoder has detected a count, a pulse, one-half clock cycle long, will be output on the CNT<sub>DCDR</sub> pin. This output will occur during the clock cycle in which the internal counter is updated. The U/D pin will be set to the proper voltage level one clock cycle before the rising edge of the CNT-DCDR pulse, and held one clock cycle after the rising edge of the CNT<sub>DCDR</sub> pulse. These outputs are not affected by the inhibit logic.

## Cascade Output (HCTL-2032 / 2032-SC only)

The cascade output also consists of count and up/down outputs. When the HCTL-2032 / 2032-SC internal counter overflows or underflows, a pulse, one-half clock cycle long, will be output on the  $CNT_{CAS}$  pin. This output will occur during the clock cycle in which the internal counter is updated. The U/D pin will be set to the proper voltage level one clock cycle before the rising edge of the  $CNT_{CAS}$  pulse, and held one clock cycle after the rising edge of the  $CNT_{CAS}$  pulse. These outputs are not affected by the inhibit logic.

Step	SEL1	SEL2	OE	CLK	Inhibit Signal	Action
1	L	Н	L		1	Set inhibit; Read MSB
2	Н	Н	L		1	Read <sup>2nd</sup> Byte
3	L	L	L		1	Read 3 <sup>rd</sup> Byte
4	Н	L	L		1	Read LSB
5	Х	Х	Н		0	Completes inhibit logic reset

Figure 15. Four Bytes Read Sequence

#### Cascade Considerations (HCTL-2032 / 2032-SC only)

The HCTL-2032 / 2032-SC 's cascading system allows for position reads of more than four bytes. These reads can be accomplished by latching all the bytes and then reading the bytes sequentially over the 8-bit bus. It is assumed here that, externally, a counter followed by a latch is used to count any count that exceeds 32 bits. This configuration is compatible with the HCTL-2032 / 2032-SC internal counter/latch combination.

Consider the sequence of events for a read cycle that starts as the HCTL-2032 / 2032-SC 's internal counter rolls over. On the rising clock edge, count data is updated in the internal counter, rolling it over. A count-cascade pulse (CNT<sub>CAS</sub>) will be generated with some delay after the rising clock edge (t<sub>CHD</sub>). There will be additional propagation delays through the external counters and registers. Meanwhile, with SEL and OE low to start the read, the internal latches are inhibited at the falling edge and do not update again till the inhibit is reset.

If the CNT<sub>CAS</sub> pulse now toggles the external counter and this count gets latched a major count error will occur. The count error is because the external latches get updated when the internal latch is inhibited.

Valid data can be ensured by latching the external counter data when the high byte read is started (SEL and OE low). This latched external byte corresponds to the count in the inhibited internal latch. The cascade pulse that occurs during the clock cycle when the read begins gets counted by the external counter and is not lost.

For example, suppose the HCTL-2032 / 2032-SC count is at FFFFFFF and an external counter is at F0h, with the count going up. A count occurring in the HCTL-2032 / 2032-SC will cause the counter to roll over and a cascade pulse will be generated. A read starting on this clock cycle will show FFFFFFF from the HCTL-2032 / 2032-SC. The external latch should read F0h, but if the host latches the count after the cascade signal propagates through, the external latch will read F1h.

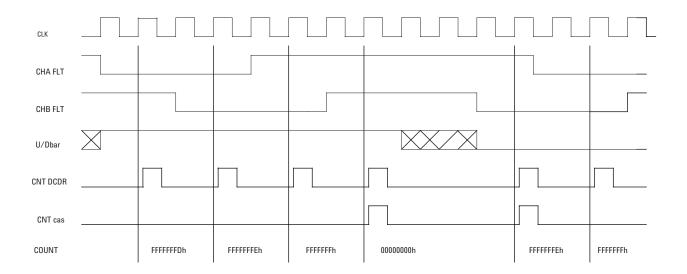
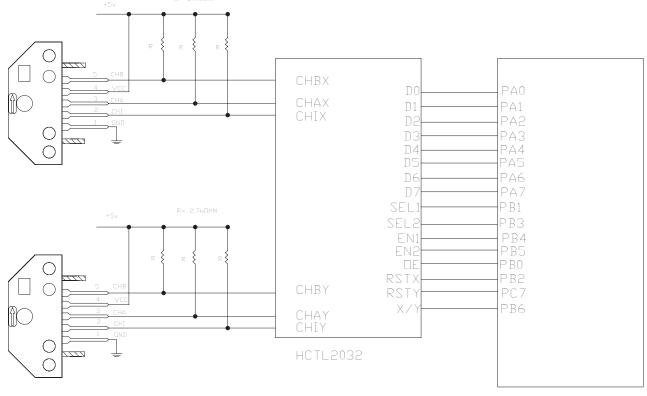


Figure 16. Decode and Casade Output Diagram (4x)

## Interfacing the HCTL-2032 to an Atmel AVR 90S8535

The circuit shown in Figure 17 shows the connections between an HCTL-2032 and an Atmel AVR controller. Data lines D0-D7 are connected to the Atmel AVR bus port. The 8 MHz oscillators clock the Atmel AVR, whereas the external 33 MHz oscillators clock the HCTL-2032. Figure 18 illustrates the program that interfaces with an Atmel AVR 90S8535.



AT90S8535

Figure 17. An HCTL-2032-to-Atmel AVR Interface

Set Portb.4	'EN1=1	
Reset Portb.5	'EN2=0	
Reset Portb.6	'Select X-axis	
Result_new = 0		
$Result_old_x = 0$		
Result_old_y = 0		
Do		
Set Portb.0	'Disable OE	
Waitms 25		
Reset Portb.1	'SEL1=0 (MSB)	
Set Portb.3	SEL1=0 (MSB) (SEL2=1 (MSB)	
Reset Portb.0	'Enable OE	
Reset Fortb.o		
Gosub Get_hi	'Get MSB	
Set Portb.1	'SEL1=1 (2nd Byte)	
Set Portb.3	'SEL2=1 (2nd Byte)	
Gosub Get_2nd	'Get 2nd Byte	
Reset Portb.1	'SEL1=0 (3rd Byte)	
Reset Portb.3	'SEL2=0 (3rd Byte)	
Gosub Get_3rd	'Get 3rd Byte	
Set Portb.1	'SEL1=1 (LSB)	
Reset Portb.3	(SEL2=0 (LSB)	
Gosub Get_lo	'Get LSB	
Set Portb.0	'Disable OE	
Waitms 25		
Mult = 1	(Assign LCD	
Temp = Result_lo * Mult Result = Temp	'Assign LSB	
Mult = Mult $*$ 256		
Temp = Result_3rd * Mult	'Assign 3rd Byte	
Result = Result + Temp		
Mult = Mult * 256		
Temp = Result_2nd * Mult	'Assign 2nd Byte	
Result = Result + Temp		
Mult = Mult * 256		
Temp = Result_hi * Mult	'Assign MSB	
Result = Result + Temp	-	
Pagult 22 hits Count Data		
Result = 32-bits Count Data '		
Loop		

Figure 18. Typical Program for Reading HCTL-2032 with Atmel AVR

Cat hi		
Get_hi:	Cat aureant data	
Hi_old = Pina	'Get current data	
Hi_new = Pina	'Get 2nd Data	
If Hi_new = Hi_old Then		
Result_hi = Hi_new	'Get stable data	
Return		
Else		
Goto Get_2nd		
End If		
Get_2nd:		
2nd_old = Pina	'Get current data	
2nd_new = Pina	'Get 2nd Data	
If 2nd_new = 2nd_old Then		
Result_2nd = 2nd_new	'Get stable data	
Return		
Else		
Goto Get_2nd		
End If		
Get_3rd:		
3rd_old = Pina	'Get current data	
3rd_new = Pina	'Get 2nd Data	
If 3rd_new = 3rd_old Then		
Result_3rd = 3rd_new	'Get stable data	
Return	Get stable data	
Else		
Goto Get_3rd		
End If		
Get_lo:		
Lo_old = Pina	'Get current data	
Lo_new = Pina	'Get 2nd Data	
If Lo_new = Lo_old Then		
Result_lo = Lo_new	'Get stable data	
Return		
Else		
Cata Catala		
Goto Get_lo		
Goto Get_lo End lf		

Figure 18 Cont. Typical Program for Reading HCTL-2032 with Atmel AVR

## Actions

- 1. At first, Port B4, B5, and B6 are setup for 4X encoding and X/Y axis selection.
- 2. The HCTL-2032 detects that OE/ are low on the next falling edge of the CLK and asserts the internal inhibit signal. Data can be read without regard for the phase of the CLK.
- 3. SEL1 and SEL2 are setup to select the appropriate bytes. The "Get\_hi" subroutine is called and the data is read into the AVR.
- 4. Step 3 is repeated by changing the SEL1 and SEL2 combinations and specific subroutine is called to read in the appropriate data.
- 5. The HCTL-2032 detects OE/ high on the next falling edge of the CLK. The program set OE/ high by writing the correct value to the respective Port. This causes the data lines to be tristated. On the next rising CLK edge new data is transferred from the counter to the position data latch.
- 6. For displaying purposes, the data is arranged in 32bit data by shifting the MSB to the left through multiplication.

## **Ordering Information**

HCTL - 20 XX - XX							
	3	2 E	Blank	32-PDIP Package			
	32	2 5	SC	32-SOIC Package			
	32	2 5	SCT	32-SOIC Package in Tape and Reel (1000 Pcs / Reel)			
	2	2 E	Blank	20-PDIP Package			
	1						

For product information and a complete list of distributors, please go to our web site: www.avagotech.com



