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**2.5V SEQUENTIAL FLOW-CONTROL DEVICE**  
**48 BIT WIDE CONFIGURATION**  
 For use with 128Mb to 256Mb DDR SDRAM

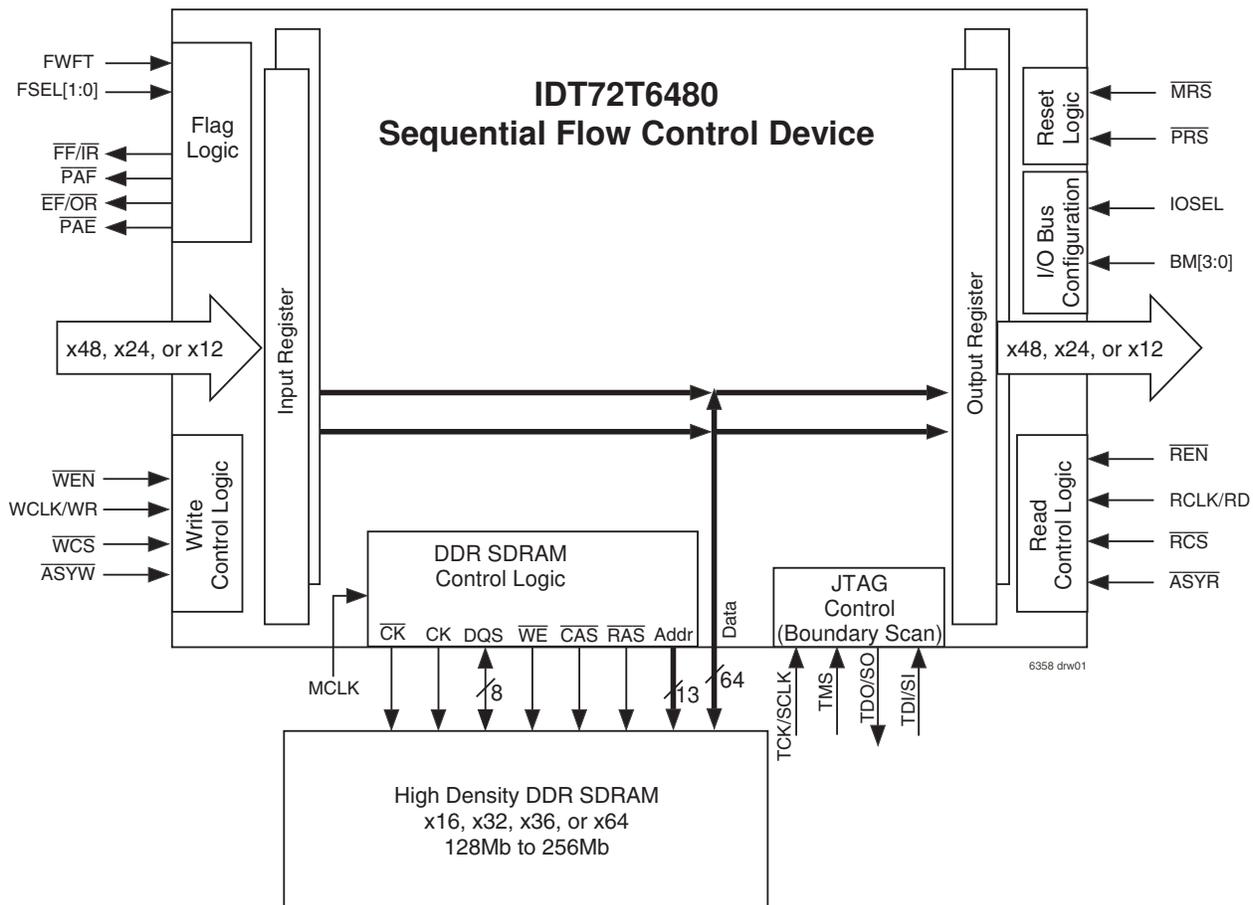
**IDT72T6480**

**FEATURES**

- Product to be used with single or multiple external DDR SDRAM to provide significant storage capability of up to 1Gb density
- 133MHz operation (7.5ns read/write cycle time)
- User selectable input and output port bus-sizing
  - x48in to x48out
  - x48in to x24out
  - x48in to x12out
  - x24in to x48out
  - x24in to x24out
  - x24in to x12out
  - x12in to x48out
  - x12in to x24out
  - x12in to x12out
- For other bus configurations see IDT72T6360 (x9, x18, or x36)
- 2.5V-LVTTL or 3.3V-LVTTL configured ports
- Independent and simultaneous read and write access
- User selectable synchronous/asynchronous read and write port timing

- IDT Standard mode or FWFT mode of operation
- Empty and full flags for monitoring memory status
- Programmable Almost-Empty and Almost-Full flags, each flag can default to one of four preselected offsets or serially programmed to a specific value
- Selectable synchronous/asynchronous timing modes for Almost-Empty and Almost-Full flags
- Master Reset clears all data and settings
- Partial Reset clears data, but retains programmable settings
- Depth expandable with multiple devices for densities greater than 1Gb
- Width expandable with multiple devices for bus widths greater than 36 bits
- JTAG functionality (Boundary Scan)
- Available in a 324-pin PBGA, 1mm pitch, 19mm x 19mm
- HIGH performance 0.18µm CMOS technology
- Industrial temperature range (-40°C to +85°C) is available
- Supports industry standard DDR specifications, including Samsung, Micron, and Infineon memories

**FUNCTIONAL BLOCK DIAGRAM**



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**COMMERCIAL AND INDUSTRIAL TEMPERATURE RANGES**

**FEBRUARY 2009**

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## DESCRIPTION

The IDT72T6480 sequential flow-control device is a device incorporating a seamless connection to external DDR SDRAM for significant storage capacity supporting high-speed applications. Both read and write ports of the sequential flow-control can operate independently at up to 133MHz. There is a user selectable correction feature that will correct any erroneous single data bit when reading from the SDRAM.

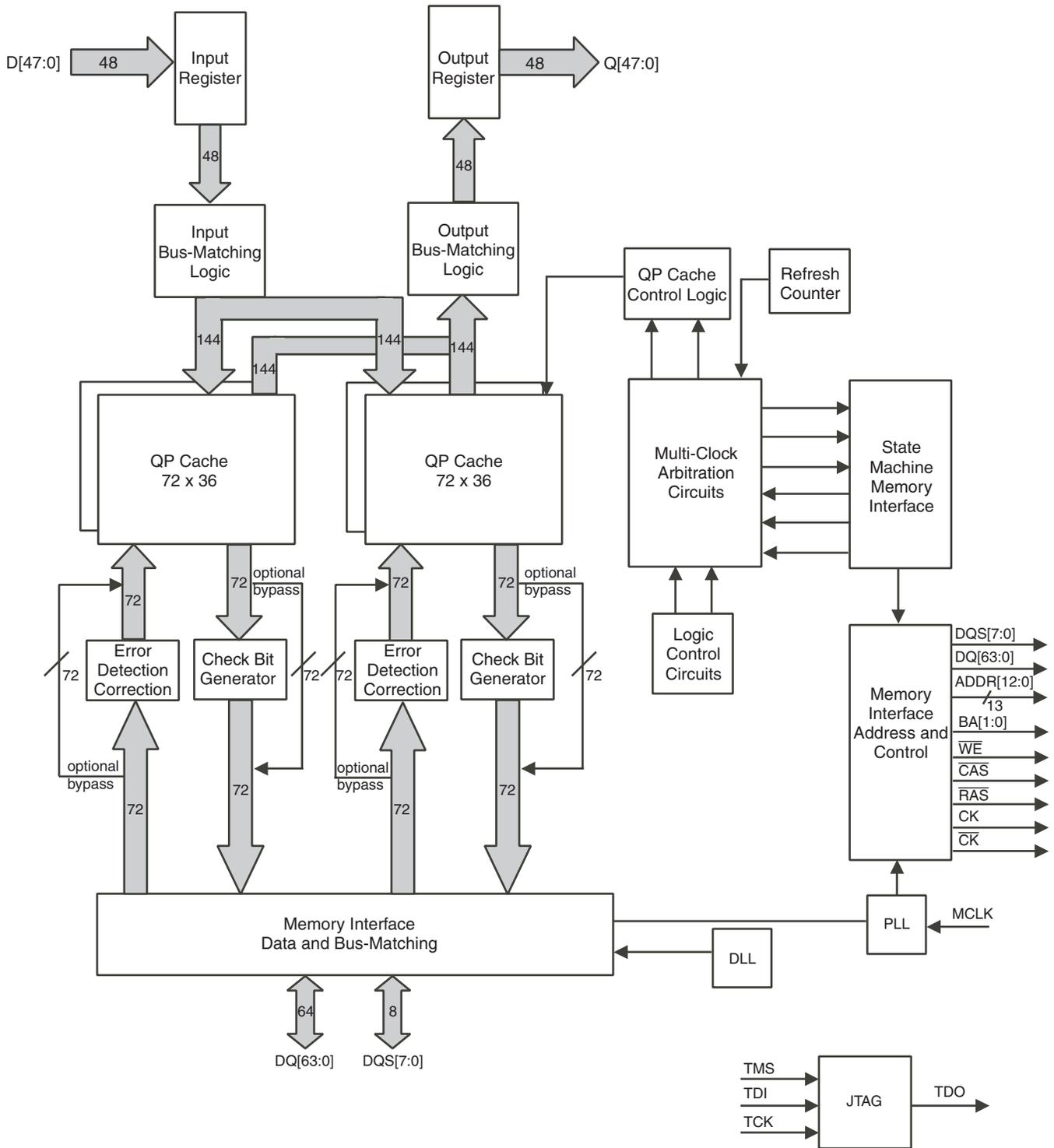
The independent read and write ports each has associated read and write clocks, enables, and chip selects. Both ports can operate either synchronously or asynchronously. Other features include bus-matching, programmable status flags with selectable synchronous/asynchronous timing modes, IDT Standard or FWFT mode timing, and JTAG boundary scan functionality.

The bus-matching feature will allow the inputs and outputs to be configured to x48, x24, or x12 bus width. There are four default offset values available

for the programmable flags ( $\overline{PAE}/\overline{PAF}$ ), as well as the option of serially programming the offsets to a specific value.

The device package is 19mm x 19mm 324-pin PBGA. It operates at a 2.5V core voltage with selectable 2.5V or 3.3V I/Os. The I/O interface to the SDRAM will be 2.5V SSTL\_2 only and not 3.3V tolerant. Both industrial and commercial temperature ranges will be offered.

The sequential flow-control device controls individual DDR SDRAM of either 128Mb or 256Mb. The device will support industry standard DDR specification memories (note DDR II is not supported), which include vendors such as Samsung, Micron, and Infineon. The data bus connected to the DDR SDRAM can be 16-bit, 32-bit, or 64-bits wide. The sequential flow-control device can independently control up to four separate external memories for a maximum of density of 1Gb (128MB). Depth expansion mode is available for applications that require more than 1Gb of storage memory.



6358 drw02

Figure 1. Sequential Flow-Control Device Block Diagram

# PIN CONFIGURATION

A1 BALL PAD CORNER

A	GND	GND	DQ10	DQ8	DQ4	DQ1	$\overline{CK}$	A1	A5	A9	$\overline{WE}$	$\overline{RAS}$	DQ35	DQ36	DQ38	DQ40	GND	GND	
B	GND	GND	DQS1	DQ9	DQ5	DQ2	AVCC	A0	A4	A10	BA1	DQ32	DQS4	DQ39	DQ44	DQS5	GND	GND	
C	DQ14	DQ13	DQ11	DQ12	DQ6	DQ3	CK	A2	A6	A11	BA0	DQ34	DQ37	DQ41	DQ47	DQ45	DQ49	DQ50	
D	DQ16	DQ15	DQ17	DQ7	DQS0	DQ0	AVCC	A3	A7	A8	A12	$\overline{CAS}$	DQ33	DQ43	DQ46	VREF	DQ51	DQS6	
E	DQ19	DQ18	DQS2	DQ20	DQ21	AGND	AGND	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	DQ42	DQ54	DQ48	DQ52	DQ53
F	DQ23	DQ22	DQ24	DQ25	DQ26	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	DQ59	DQ58	DQ55	DQ56	DQ57
G	DQS3	DQ27	DQ28	DQ29	DQ30	Vcc	Vcc	GND	GND	GND	GND	Vcc	Vcc	Vcc	DQ62	DQ61	DQS7	DQ60	
H	MCLK	DQ31	D0	D1	D2	Vcc	GND	GND	GND	GND	GND	GND	VDDQ	VDDQ	Q2	Q1	DQ63	Q0	
J	D4	D3	D5	D6	D7	Vcc	GND	GND	GND	GND	GND	GND	VDDQ	VDDQ	Q6	Q5	Q3	Q4	
K	D11	D12	D10	D9	D8	Vcc	GND	GND	GND	GND	GND	GND	VDDQ	VDDQ	Q7	Q12	Q8	Q9	
L	D16	D17	D15	D14	D13	Vcc	Vcc	GND	GND	GND	GND	GND	VDDQ	VDDQ	Q13	Q11	Q10	Q14	
M	D21	D22	D20	D19	D18	Vcc	Vcc	Vcc	GND	GND	GND	VDDQ	VDDQ	VDDQ	Q28	Q15	Q16	Q17	
N	D26	D27	D25	D24	D23	Vcc	Vcc	Vcc	Vcc	Vcc	Vcc	VDDQ	VDDQ	Q23	Q29	Q19	Q20	Q18	
P	D31	D32	D30	D29	D28	FSEL1	IOSEL	Vcc	BM3	MIC0	JSEL	TDO/SO	Q47	Q32	Q33	Q31	Q25	Q21	
R	D34	D35	D33	D38	D45	FSEL0	IDEM	MTYPE0	BM2	MIC1	TDI/SI	$\overline{FF/IR}$	Q46	Q42	Q34	Q30	Q26	Q22	
T	D36	D37	D39	D40	D41	$\overline{ASYW}$	$\overline{WCS}$	MSPEED	BM1	TMS	$\overline{SWEN}$	$\overline{PAF}$	Q45	Q41	Q35	Q36	Q27	Q24	
U	GND	GND	D42	D43	D46	$\overline{PRS}$	FWFT	MTYPE1	BM0	MIC2	TCK/SCLK	$\overline{OE}$	$\overline{PAE}$	Q43	Q39	Q37	GND	GND	
V	GND	GND	D44	D47	$\overline{MRS}$	$\overline{ASYR}$	$\overline{WEN}$	WCLK/ WR	RCLK/ RD	$\overline{REN}$	$\overline{SREN}$	$\overline{RCS}$	$\overline{EF/OR}$	Q44	Q40	Q38	GND	GND	
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	

6358 drw03

PBGA (BB324-1, order code: BB)  
 TOP VIEW

## PIN DESCRIPTIONS

Symbol	Pin No. Location	Name	I/O TYPE	Description
<b>READ PORT INTERFACE</b>				
$\overline{\text{ASYR}}^{(1)}$	V6	Asynchronous Read Port	INPUT 3.3V or 2.5V LVTTTL	A HIGH on this input during master reset will select synchronous read operation for the output port. A LOW will select asynchronous operation. If asynchronous is selected the device must operate in IDT Standard mode and the read enable must be tied to GND.
$\overline{\text{EF/OR}}$	V13	Empty Flag/ Output Ready	OUTPUT 3.3V or 2.5V LVTTTL	In IDT Standard mode, the $\overline{\text{EF}}$ function is selected. $\overline{\text{EF}}$ indicates whether or not the device memory is empty. In FWFT mode, the $\overline{\text{OR}}$ function is selected. $\overline{\text{OR}}$ indicates whether or not there is valid data available at the outputs.
$\overline{\text{OE}}$	U12	Output Enable	INPUT 3.3V or 2.5V LVTTTL	Asynchronous three-state control of the data outputs. All data outputs Q[47:0] will be placed in high-impedance if this pin is HIGH. Conversely, all data outputs will be active when this pin is LOW.
$\overline{\text{PAE}}$	U13	Programmable Almost Empty Flag	OUTPUT 3.3V or 2.5V LVTTTL	This is the programmable almost empty flag that can be used as an early indicator for the empty boundary condition of the internal memory. $\overline{\text{PAE}}$ goes LOW if the number of words in the sequential flow-control device is less than offset n, which is stored in the empty offset register. $\overline{\text{PAE}}$ goes HIGH if the number of words in the sequential flow-control device is greater than or equal to the offset n.
Q[47:0]	See Pin No. table	Data Output Bus	OUTPUT 3.3V or 2.5V LVTTTL	Data outputs for a 48, 24, and 12-bit bus.
RCLK/ RD	V9	Read Clock/ Read Strobe	INPUT 3.3V or 2.5V LVTTTL	This is a dual function pin. If synchronous operation of the read port is selected, the rising edge of RCLK reads data from the sequential flow-control device when $\overline{\text{REN}}$ is enabled. If asynchronous operation of the read port is selected, a rising edge on RD reads data from the sequential flow-control device without the need of a free-running input read clock.
$\overline{\text{RCS}}$	V12	Read Chip Select	INPUT 3.3V or 2.5V LVTTTL	Synchronous three-state control of the data outputs. Provides another means of controlling the data outputs synchronous to RCLK. Can be regarded as a second output enable signal.
$\overline{\text{REN}}$	V10	Read Enable	INPUT 3.3V or 2.5V LVTTTL	$\overline{\text{REN}}$ enables RCLK for reading data from the sequential flow-control device. If asynchronous mode is selected on the read port, this signal should be tied to GND.
$\overline{\text{SREN}}$	V11	Serial Read Enable	INPUT 3.3V or 2.5V LVTTTL	When $\overline{\text{SREN}}$ is brought LOW before the rising edge of SCLK, the contents of the $\overline{\text{PAE}}$ and $\overline{\text{PAF}}$ offset registers are copied to a serial shift register. While $\overline{\text{SREN}}$ is maintained LOW, on each rising edge of SCLK, one bit of data is shifted out of this serial shift register through the SO output pin used only when JSEL = 0.
<b>WRITE PORT INTERFACE</b>				
$\overline{\text{ASYW}}^{(1)}$	T6	Asynchronous Write Port	INPUT 3.3V or 2.5V LVTTTL	A HIGH on this input during master reset will select synchronous write operation for the input port. A LOW will select asynchronous operation. If asynchronous is selected the device must operate in IDT Standard mode and the write enable must be tied to GND.
D[47:0]	See Pin No. table	Data Inputs	INPUT 3.3V or 2.5V LVTTTL	Data inputs for a 48, 24, and 12-bit bus.
$\overline{\text{FF/IR}}$	R12	Full Flag/ Input Ready	OUTPUT 3.3V or 2.5V LVTTTL	In IDT Standard mode, the $\overline{\text{FF}}$ function is selected. $\overline{\text{FF}}$ indicates whether or not the device memory is full. In FWFT mode, the $\overline{\text{IR}}$ function is selected. $\overline{\text{IR}}$ indicates whether or not there is space available for writing to the device memory.
$\overline{\text{PAF}}$	T12	Programmable Almost Full Flag	OUTPUT 3.3V or 2.5V LVTTTL	This is the programmable almost full flag that can be used as an early indicator for the full boundary condition of the internal memory. $\overline{\text{PAF}}$ goes HIGH if the number of free locations in the sequential flow-control device is more than offset m, which is stored in the full offset register. $\overline{\text{PAF}}$ goes LOW if the number of free locations in the sequential flow-control device is less than or equal to the offset m.
$\overline{\text{SWEN}}$	T11	Serial Write Enable	INPUT 3.3V or 2.5V LVTTTL	On each rising edge of SCLK when $\overline{\text{SWEN}}$ is LOW, data from the SI pin is serially loaded into the $\overline{\text{PAE}}$ and $\overline{\text{PAF}}$ registers used only when JSEL = 0.

## PIN DESCRIPTIONS (Continued)

Symbol	Pin No. Location	Name	I/O TYPE	Description
<b>WRITE PORT INTERFACE (Continued)</b>				
WCLK/WR	V8	Write Clock/ Write Strobe	INPUT 3.3V or 2.5V LVTTTL	This is a dual function pin. If synchronous operation of the write port is selected, the rising edge of WCLK writes data into the sequential flow-control device when $\overline{WEN}$ is enabled. If asynchronous operation of the write port is selected, a rising edge on WR writes data into the sequential flow-control device without the need of a free-running input write clock.
$\overline{WCS}$	T7	Write Chip Select	INPUT 3.3V or 2.5V LVTTTL	Synchronous three-state control of the data inputs. Provides a means of controlling the data inputs synchronous to WCLK. Typically used to avoid bus-contention when multiple devices are sharing the same input data bus.
$\overline{WEN}$	V7	Write Enable	INPUT 3.3V or 2.5V LVTTTL	$\overline{WEN}$ enables WCLK for writing data into the sequential flow-control device. If asynchronous mode is selected on the write port, this signal should be tied to GND.
<b>MEMORY INTERFACE</b>				
A[12:0]	See Pin No. table	Memory Address Bus	OUTPUT SSTL_2	Output address bus to be connected to the input address bus of the external memory to provide row and column address.
BA[1:0]	BA1-B11 BA0-C11	Memory Bank Address Input Bit	OUTPUT SSTL_2	Address bits to be connected to the external memory's BA inputs to determine which bank an ACTIVE, READ, WRITE, or PRECHARGE command is being applied.
CK	C7	Memory Clock	OUTPUT SSTL_2	Clock output to be connected to the external memory's input clock.
$\overline{CK}$	A7	Memory Clock Inverted	OUTPUT SSTL_2	Differential clock output to be connected to the external memory's differential input clock.
$\overline{CAS}$	D12	Memory Column Address Strobe	OUTPUT SSTL_2	Output enable signal to be connected to the external memory's $\overline{CAS}$ pin to activate and deactivate the column address strobe.
DQ[63:0]	See Pin No. table	Memory Data Bus	Bi-Directional SSTL_2	Input/output data bus for the external memory's data bus.
DQS[7:0]	See Pin No. table	Memory Data Strobe	Bi-Directional SSTL_2	Input/output data strobe to be connected to the external memory's data strobe.
$\overline{RAS}$	A12	Memory Row Address Strobe	OUTPUT SSTL_2	Output strobe signal to be connected to the external memory's $\overline{RAS}$ pin to activate and deactivate the row address strobe.
$\overline{WE}$	A11	Memory Write Enable	OUTPUT SSTL_2	Output strobe signal to be connected to the external memory's $\overline{WE}$ pin to activate and deactivate the write address strobe.
<b>CONTROL AND FEATURE INTERFACE</b>				
BM[3:0] <sup>(1)</sup>	See Pin No. table	Bus-Matching Bit	INPUT 3.3V or 2.5V LVTTTL	Selects the bus width of the read and write ports.
FSEL[1:0] <sup>(1)</sup>	FSEL1-P6 FSEL0-R6	Flag Select Bit	INPUT 3.3V or 2.5V LVTTTL	During master reset, these inputs will select one of four default values for the programmable flags $\overline{PAE}$ and $\overline{PAF}$ . The selected value will apply to both $\overline{PAE}$ and $\overline{PAF}$ offset.
FWFT <sup>(1)</sup>	U7	First Word Fall Through	INPUT 3.3V or 2.5V LVTTTL	During master reset, a HIGH on this input selects FWFT timing mode. A LOW selects IDT Standard timing mode.
IDEM <sup>(1)</sup>	R7	IDT Standard Mode Depth Expansion Mode Select	INPUT 3.3V or 2.5V LVTTTL	This select pin is used for depth expansion configuration in IDT Standard mode. If this pin is tied HIGH, then the $\overline{FF}/\overline{IR}$ signal will be inverted to provide a seamless depth expansion interface. If depth expansion in FWFT mode is desired, this pin should be tied to GND. If no depth expansion is used, this pin should be tied to GND.
IOSEL <sup>(1)</sup>	P7	I/O VDDQ Select	INPUT 3.3V or 2.5V LVTTTL	This input determines whether the inputs and outputs will tolerate a 2.5V or 3.3V voltage signals. If IOSEL is HIGH, then all I/Os will be 2.5V tolerant. If IOSEL is LOW, then all I/Os will be 3.3V tolerant. See table 15, for a list of affected I/O signals.

## PIN DESCRIPTIONS (Continued)

Symbol	Pin No. Location	Name	I/O TYPE	Description
<b>CONTROL AND FEATURE INTERFACE (Continued)</b>				
JSEL <sup>(1)</sup>	P11	JTAG Select	INPUT 3.3V or 2.5V LVTTTL	This pin selects whether the JTAG pins will be used for serial programming. If JSEL is HIGH, the JTAG pins will only be used for JTAG boundary-scan function. If JSEL is LOW, the JTAG function is disabled and the JTAG pins will be used for serial programming of the $\overline{\text{PAE}}/\overline{\text{PAF}}$ offset registers.
MIC[2:0] <sup>(1)</sup>	MIC2-U10 MIC1-R10 MIC0-P10	Memory Configuration	INPUT 3.3V or 2.5V LVTTTL	These signals enable the EDC feature of the device. See Table 8, MIC[2:0] Configurations for details.
MCLK	H1	Master Clock	INPUT 3.3V or 2.5V LVTTTL	33MHz reference clock used to generate CK and $\overline{\text{CK}}$ for external memory interface.
$\overline{\text{MRS}}$	V5	Master Reset	INPUT 3.3V or 2.5V LVTTTL	Master reset initializes the read and write pointers to zero and sets the output register to all zeros. All initialized settings for the device will be configured during master reset.
MSPEED <sup>(1)</sup>	T8	Memory Speed	INPUT 3.3V or 2.5V LVTTTL	This input select the speed of the external memory interfacing the sequential flow-control device. A LOW selects 133MHz, and HIGH selects 166MHz.
MTYPE <sup>(1)</sup> [1:0]	MTYPE1-U8 MTYPE0-R8	Memory Type [1:0]	INPUT 3.3V or 2.5V LVTTTL	These inputs select which type of external memory is interfacing the sequential flow-control device. See Table 14 for the list of selectable memories.
$\overline{\text{PRS}}$	U6	Partial Reset	INPUT 3.3V or 2.5V LVTTTL	Partial reset initializes the read and write pointers to zero and sets the output registers to all zeros. All existing configurations in the sequential flow-control device will not be affected. This includes the IDT Standard or FWFT mode timing, programmable flag settings, and bus width and data rate mode.
TCK/ SCLK	U11	JTAG Clock/ Serial Clock	INPUT 3.3V or 2.5V LVTTTL	This is a dual function pin. When the JSEL pin is HIGH, this is the clock input for JTAG boundary-scan function. One of four terminals required by IEEE Standard 1149.1-1990. Test operations of the device are synchronous to TCK. Data from TMS and TDI are sampled on the rising edge of TCK and outputs change on the falling edge of TCK. When the JSEL pin is LOW, this is the serial clock input for writing and reading the $\overline{\text{PAE}}/\overline{\text{PAF}}$ offset registers. On the rising edge of every SCLK when $\overline{\text{SWEN}}$ is LOW, one bit of data from the SI pin is shifted into the $\overline{\text{PAE}}$ and $\overline{\text{PAF}}$ offset registers. On the rising edge of each SCLK when SREN is LOW, one bit of data from the SO pin is shifted out of the $\overline{\text{PAE}}$ and $\overline{\text{PAF}}$ offset registers. If the JTAG or serial programming is not used this signal needs to be tied to GND.
TDI/SI	R11	JTAG Test Data Input/ Serial Input	INPUT 3.3V or 2.5V LVTTTL	This is a dual function pin. When the JSEL pin is HIGH, this is the JTAG test data input pin. One of four terminals required by IEEE Standard 1149.1-1990. During the JTAG boundary scan operation, test data serially loaded via the TDI on the rising edge of TCK to the Instruction Register, ID Register and Bypass Register. When the JSEL pin is LOW, this is the serial input pin for the $\overline{\text{PAE}}/\overline{\text{PAF}}$ offset registers. An internal pull-up resistor forces TDI/SI HIGH if left unconnected.
TDO/SO	P12	JTAG Test Data Output/Serial Output	OUTPUT 3.3V or 2.5V LVTTTL	This is a dual function pin. When the JSEL pin is HIGH, this is the JTAG test data output pin. One of four terminals required by IEEE Standard 1149.1-1990. During the JTAG boundary scan operation, test data serially loaded output via the TDO on the falling edge of TCK from either the Instruction Register, ID Register and Bypass Register. This output is high-impedance except when shifting, while in SHIFT-DR and SHIFT-IR controller states. When the JSEL pin is LOW, this is the serial data output pin for the $\overline{\text{PAE}}/\overline{\text{PAF}}$ offset registers.
TMS	T10	JTAG Mode Select	INPUT 3.3V or 2.5V LVTTTL	TMS is a serial input pin. One of four terminals required by IEEE Standard 1149.1-1990. TMS directs the device through its TAP controller states. An internal pull-up resistor forces TMS HIGH if left unconnected.

NOTE: 1. These pins should not change after master reset.

**Please see next page for Power & Ground pins and Pin Number Location Table.**

## PIN DESCRIPTIONS (Continued)

Symbol	Pin No. Location	Name	I/O TYPE	Description
<b>POWER AND GROUND SIGNALS</b>				
Vcc	See Pin No. table	Core Vcc and Output voltage for DDR SDRAM	Power	The core power supply pins for the device as well as to the external DDR SDRAM. Needs to be connected to a +2.5V Vcc power plane.
AVcc	B7, D7	Internal PLL Vcc	Power	The power supply pins for the internal PLL of the device. Needs to be connected to a +2.5V supply rail.
VDDQ	See Pin No. table	Output rail voltage for I/Os	Power	This pin is used to provide power to the output drivers. The nominal values are 2.5V or 3.3V, depending on the state of the IOSEL pin.
VREF	D16	Reference Voltage	Power	This is a voltage reference input to the SDRAM and must be connected to Vcc/2.
GND	See Pin table	Ground Pin	Ground	The ground pins for the device that must be connected to the ground plane.
AGND	E6, E7	Ground pin for analog circuit	Ground	The ground pins for the analog circuitry in the device that must be connected to the ground plane.

## PIN NUMBER LOCATION TABLE

Symbol	Name	I/O TYPE	Pin Number
A[12:0]	Memory Address Bus	OUTPUT SSTL_2	A12-D11, A11-C10, A10-B10, A9-A10, A8-D10, A7-D9, A6-C9, A5-A9, A4-B9, A3-D8, A2-C8, A1-A8, A0-B8
BM[3:0]	Bus-Matching	INPUT 3.3V or 2.5V LVTTTL	BM3-P9, BM2-R9, BM1-T9, BM0-U9
D[47:0]	Data Inputs	INPUT 3.3V or 2.5V LVTTTL	D47-V4, D46-U5, D45-R5, D44-V3, D43-U4, D42-U3, D41-T5, D40-T4, D39-T3, D38-R4, D37-T2, D36-T1, D35-R2, D34-R1, D33-R3, D32-P2, D31-P1, D30-P3, D29-P4, D28-P5, D27-N2, D26-N1, D25-N3, D24-N4, D23-N5, D22-M2, D21-M1, D20-M3, D19-M4, D18-M5, D17-L2, D16-L1, D15-L3, D14-L4, D13-L5, D12-K2, D11-K1, D10-K3, D9-K4, D8-K5, D7-J5, D6-J4, D5-J3, D4-J1, D3-J2, D2-H5, D1-H4, D0-H3
DQ[63:0]	Memory Data Bus	Bi-Directional SSTL_2	DQ63-H17, DQ62-G15, DQ61-G16, DQ60-G18, DQ59-F14, DQ58-F15, DQ57-F18, DQ56-F17, DQ55-F16, DQ54-E15, DQ53-E18, DQ52-E17, DQ51-D17, DQ50-C18, DQ49-C17, DQ48-E16, DQ47-C15, DQ46-D15, DQ45-C16, DQ44-B15, DQ43-D14, DQ42-E14, DQ41-C14, DQ40-A16, DQ39-B14, DQ38-A15, DQ37-C13, DQ36-A14, DQ35-A13, DQ34-C12, DQ33-D13, DQ32-B12, DQ31-H2, DQ30-G5, DQ29-G4, DQ28-G3, DQ27-G2, DQ26-F5, DQ25-F4, DQ24-F3, DQ23-F1, DQ22-F2, DQ21-E5, DQ20-E4, DQ19-E1, DQ18-E2, DQ17-D3, DQ16-D1, DQ15-D2, DQ14-C1, DQ13-C2, DQ12-C4, DQ11-C3, DQ10-A3, DQ9-B4, DQ8-A4, DQ7-D4, DQ6-C5, DQ5-B5, DQ4-A5, DQ3-C6, DQ2-B6, DQ1-A6, DQ0-D6,
DQS[7:0]	Memory Data Strobe	Bi-Directional SSTL_2	DQS7-G17, DQS6-D18, DQS5-B16, DQS4-B13, DQS3-G1, DQS2-E3, DQS1-B3, DQS0-D5
Q[47:0]	Data Outputs	Output 3.3V or 2.5V LVTTTL	Q47-P13, Q46-R13, Q45-T13, Q44-V14, Q43-U14, Q42-R14, Q41-T14, Q40-V15, Q39-U15, Q38-V16, Q37-U16, Q36-T16, Q35-T15, Q34-R15, Q33-P15, Q32-P14, Q31-P16, Q30-R16, Q29-N15, Q28-M15, Q27-T17, Q26-R17, Q25-P17, Q24-T18, Q23-N14, Q22-R18, Q21-P18, Q20-N17, Q19-N16, Q18-N18, Q17-M18, Q16-M17, Q15-M16, Q14-L18, Q13-L15, Q12-K16, Q11-L16, Q10-L17, Q9-K18, Q8-K17, Q7-K15, Q6-J15, Q5-J16, Q4-J18, Q3-J17, Q2-H15, Q1-H16, Q0-H18
Vcc	Core Vcc & Output voltage for DDR SDRAM	Power	E(8-13), F(6-13), G(6,7,12-14), H6, J6, K6, L(6,7), M(6-8), N(6-11), P8
VDDQ	Output rail voltage for I/Os	Power	H(13,14), J(13,14), K(13,14), L(13,14), M(12-14), N(12,13)
GND	Ground Pin	Ground	A(1,2,17,18), B(1,2,17,18), G(8-11), H(7-12), J(7-12), K(7-12), L(8-12), M(9-11), U(1,2,17,18), V(1,2,17,18)

## DETAILED DESCRIPTIONS

### SEQUENTIAL FLOW-CONTROL STRUCTURE

The IDT sequential flow-control (SFC) device is comprised of three interfaces: input port, output port, and memory interface. The input and output port can operate independently of each other with selectable bus widths of x12, x24, or x48 bits wide. The third interface, or memory interface, is connected directly to an external memory, which can be used to offload data entering the SFC device.

### WRITING AND READING FROM THE SEQUENTIAL FLOW-CONTROL DEVICE

Writing into the SFC device is accomplished by setting the write enable signal ( $\overline{WEN}$ ) and write chip select ( $\overline{WCS}$ ) low with a free running write clock (WCLK). Data will be written on the rising edge of every WCLK into the Quad-Port (QP) cache of the SFC device. The internal state machine of the device will determine whether to send the data to the DDR SDRAM or send it directly through to the output bus, depending on when the data is to be accessed. This provides “data coherency” and minimizes the path that the data has to travel.

Reading from the SFC device is accomplished by setting the read enable signal ( $\overline{REN}$ ) and read chip select ( $\overline{RCS}$ ) low with a free running read clock (RCLK). Data will be sent to the output bus on the rising edge of every RCLK. This data will be accessed either from the QP cache or the external DDR SDRAM.

### EXTERNAL MEMORY SELECTION

The DDR SDRAM interface of the SFC device can support DDR SDRAM with standard DDR I specifications. The SFC device can support any external memory within the following characteristics:

- Bus width: 16-bit or 32-bit wide
- Speed: 133MHz or 166MHz
- Density: 128Mb or 256Mb

Table 1 lists the DDR SDRAM minimum specifications that are required to meet the sequential flow-control device requirements. Table 2 lists the memory vendors and associated part numbers of DDR SDRAMs that have been validated by IDT to meet the requirements for the DDR SDRAM interface.

**TABLE 1 – DDR SDRAM MINIMUM SPECIFICATIONS**

DDR SDRAM Minimum Specifications					
Symbol		Parameter	16-bit DDR SDRAM	32-bit DDR SDRAM	Units
tCK	CL = 2.5	Clock cycle time	6	n/a	ns
	CL = 3.0		n/a	6	
tRFC		Auto refresh command period	75	63	ns
tRCD		Active to read/write delay	20	n/a	ns
tRP		Precharge command period	20	18	ns
tWR		Write recovery time	15	1.5	ns
tRCDRD		Active to read delay	n/a	18	ns
tRCDWR		Active to write delay	n/a	9	ns

**NOTE:**

1. These are the minimum specifications that the DDR SDRAM must meet.

**TABLE 2 – SUPPORTED MEMORY VENDORS**

Density	Bus Width	Vendor	Part#
128Mb	32	Samsung	K4D263238"X"-GC45
256Mb	16	Samsung	K4H561638"X"-TCLB3
			K4H561638"X"-GCLB3
256Mb	16	Micron	MT46V16M16TG-6T
			MT46V16M16TG-75
256Mb	16	Infineon	HYB25D256160BTL-6
			HYB25D256160BTL-7
256Mb	32	Samsung	K4D553238"X"-JC50

**NOTES:**

1. The part numbers listed above include packages that are recommended and validated by IDT. Other packages (such as lead free PCB, FBGA, etc.) may also be used but have not been validated by IDT.
2. The letter "X" for Samsung memory part numbers denotes the latest die revision for that particular device. Check with Samsung for the latest updated part number.

**EXTERNAL MEMORY CONFIGURATIONS**

The DDR SDRAM interface of the sequential flow-control (SFC) device has a 64-bit output data bus that provides up to four (16-bit SDRAM) external DDR SDRAM connections. For multiple memory connections, they must be of the same density configuration and speed grade. For example, two device connected cannot consist of one 128Mb and one 256Mb memory nor two 128Mb with one at 100MHz and the other at 133MHz. Below is a summary of the possible configurations:

- One 16-bit device connecting a x16 interface to the DDR SDRAM
- One 32-bit device connecting a x32 interface to the DDR SDRAM

- Two 16-bit devices connecting a x32 interface to the DDR SDRAM
- Two 32-bit devices connecting a x36 interface to the DDR SDRAM
- Two 32-bit devices connecting a x64 interface to the DDR SDRAM
- Three 16-bit devices connecting a x36 interface to the DDR SDRAM
- Four 16-bit devices connecting a x64 interface to the DDR SDRAM

These various configurations determine the storage density of the SFC device. The storage density can range from a minimum of 128Mb to a maximum of 1Gb. Table 3 lists the possible ways to connect the DDR SDRAMs and the number of chipset solutions to obtain the various storage densities.

**TABLE 3 – TOTAL POSSIBLE EXTERNAL MEMORY CONFIGURATIONS**

Two Chip Solution <sup>(1)</sup>	Three Chip Solution <sup>(1)</sup>	Four Chip Solution <sup>(1)</sup>	Five Chip Solution <sup>(1)</sup>
<sup>(2)</sup> Configuration 1, 2	<sup>(2)</sup> Configurations 3, 4, 5	<sup>(2)</sup> Configuration 6	<sup>(2)</sup> Configuration 7
1 x 128Mb [4M x 32] Total memory: 128Mb	2 x 128Mb [4M x 32] Total memory: 256Mb	N/A	N/A
1 x 256Mb [8M x 32] Total memory: 256Mb	2 x 256Mb [8M x 32] Total memory: 512Mb	N/A	N/A
1 x 128Mb [8M x 16] Total memory: 128Mb	2 x 128Mb [8M x 16] Total memory: 256Mb	3 x 128Mb [8M x 16] Total memory: 384Mb	4 x 128Mb [8M x 16] Total memory: 512Mb
1 x 256Mb [16M x 16] Total memory: 256Mb	2 x 256Mb [16M x 16] Total memory: 512Mb	3 x 256Mb [16M x 16] Total memory: 768Mb	4 x 256Mb [16M x 16] Total memory: 1Gb

**NOTES:**

1. The chip solution number includes the sequential flow-control device and external DDR SDRAM
2. See Figure 2a-2g for the 7 different configurations referenced in the table above.

**CONNECTING THE DDR SDRAM**

Below are the various chipset solution configurations available to the sequential flow-control device (see Figure 2a-2g). The external memory interface is designed to seamlessly connect one or more DDR SDRAMs. The output signal names should be connected directly to its corresponding input signal on the DDR SDRAM. There are three signals on the DDR SDRAM that must be tied to a static state. CKE,  $\overline{CS}$ , and DM. Table 4 outlines how to connect the many interface pins to the DDR SDRAM(s). Figure 3 and 4 are some examples of the memory interface connections for various density configurations. For information on DDR SDRAM layout recommendations, please see IDT application note AN-423.

**DDR SDRAM:** 128Mb [4Mb x 32] or 256Mb [8Mb x 32]  
**Total Memory Density:** 128Mb or 256Mb  
**Useable Memory<sup>(2)</sup>:** 108Mb or 252Mb

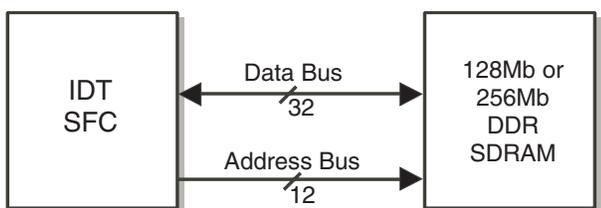


Figure 2a. Configuration 1 - Two Chip Solution

**DDR SDRAM:** 256Mb [16Mb x 16]  
**Total Memory Density:** 256Mb  
**Useable Memory<sup>(2)</sup>:** 216Mb

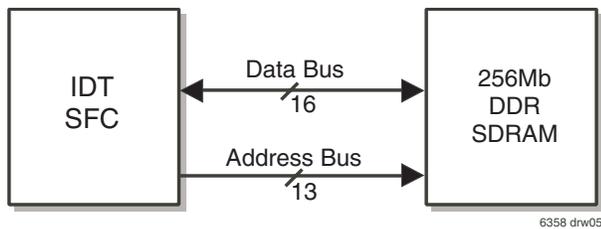


Figure 2b<sup>(1)</sup>. Configuration 2 - Two Chip Solution

**DDR SDRAM:** 128Mb [4Mb x 32] or 256Mb [8Mb x 32]  
**Total Memory Density:** 256Mb or 512Mb  
**Useable Memory<sup>(2)</sup>:** 216Mb or 504Mb

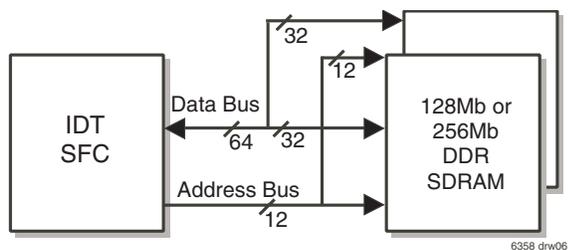


Figure 2c. Configuration 3 - Three Chip Solution

**NOTES:**

1. • 12-bit address bus for 8Mb x16  
• 13-bit address bus for 16Mb x16
2. Refer to Total Available Memory Usage section for details.

**DDR SDRAM:** 128Mb [4Mb x 32] or 256Mb [8Mb x 32]  
**Total Memory Density:** 256Mb or 512Mb  
**Useable Memory<sup>(2)</sup>:** 108Mb or 252Mb

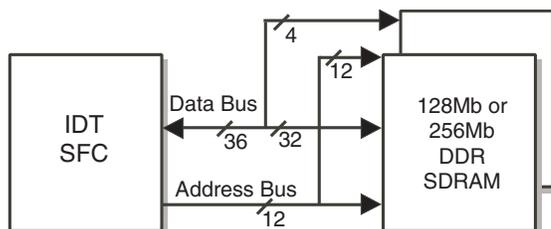


Figure 2d. Configuration 4 - Three Chip Solution

**DDR SDRAM:** 256Mb [16Mb x 16]  
**Total Memory Density:** 512Mb  
**Useable Memory<sup>(2)</sup>:** 504Mb

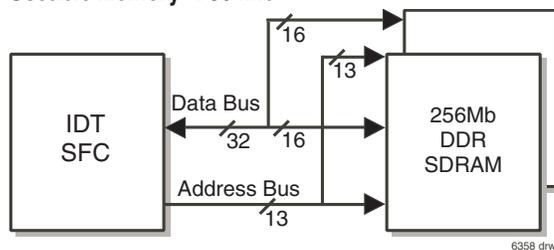


Figure 2e<sup>(1)</sup>. Configuration 5 - Three Chip Solution

**DDR SDRAM:** 256Mb [16Mb x 16]  
**Total Memory Density:** 768Mb  
**Useable Memory<sup>(2)</sup>:** 567Mb

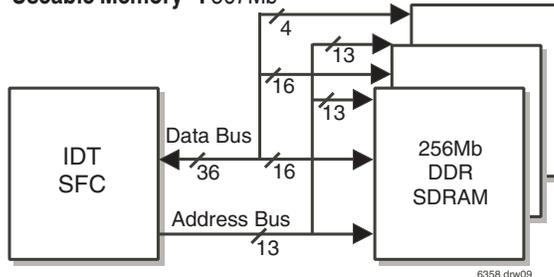


Figure 2f<sup>(1)</sup>. Configuration 6 - Four Chip Solution

**DDR SDRAM:** 256Mb [16Mb x 16]  
**Total Memory Density:** 1Gb  
**Useable Memory<sup>(2)</sup>:** 1008Mb

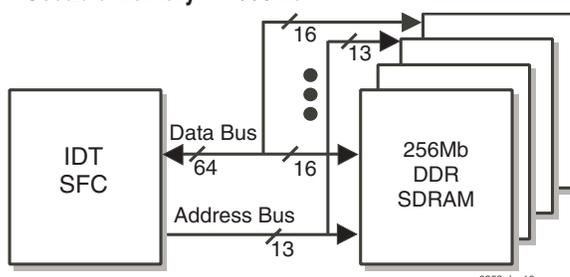
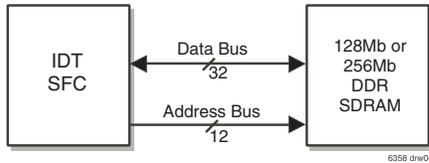


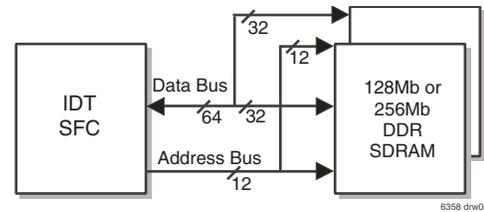
Figure 2g<sup>(1)</sup>. Configuration 7 - Five Chip Solution

TABLE 4 – SFC TO DDR SDRAM INTERFACE CONNECTIONS



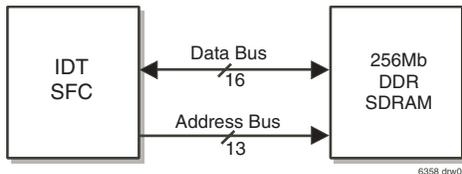
CONFIGURATION 1

SFC Outputs	DDR SDRAM
DQ[31:0]	DQ[31:0]
DQS[3:0]	DQS[3:0]
A[11:0]	A[11:0]
CK, $\overline{CK}$	CK, $\overline{CK}$
$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$
BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>	
CKE → Vcc	
$\overline{CS}$ → GND	
DM[3:0] → GND	
<b>SFC Hard wired pins</b>	
DQ[63:32] → Vcc	
DQS[7:4] → Vcc	
A12 → Vcc	



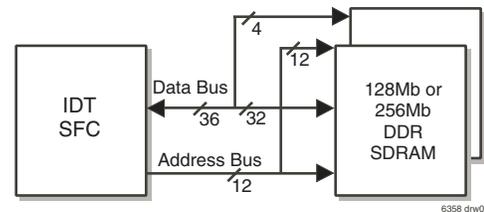
CONFIGURATION 3

SFC Outputs	DDR SDRAM #1	DDR SDRAM #2
DQ[31:0]	DQ[31:0]	--
DQ[63:32]	--	DQ[31:0]
DQS[3:0]	DQS[3:0]	--
DQS[7:4]	--	DQS[3:0]
A[11:0]	A[11:0]	A[11:0]
CK, $\overline{CK}$	CK, $\overline{CK}$	CK, $\overline{CK}$
$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$
BA[1:0]	BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>		
CKE → Vcc		
$\overline{CS}$ → GND		
DM[3:0] → GND		
<b>SFC Hard wired pins</b>		
A12 → Vcc		



CONFIGURATION 2

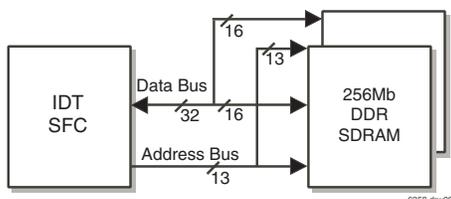
SFC Outputs	DDR SDRAM
DQ[15:0]	DQ[15:0]
DQS0	LDQS
DQS1	UDQS
A[12:0]	A[12:0]
CK, $\overline{CK}$	CK, $\overline{CK}$
$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$
BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>	
CKE → Vcc	
$\overline{CS}$ → GND	
LDM → GND	
UDM → GND	
<b>SFC Hard wired pins</b>	
DQ[63:16] → Vcc	
DQS[7:2] → Vcc	



CONFIGURATION 4

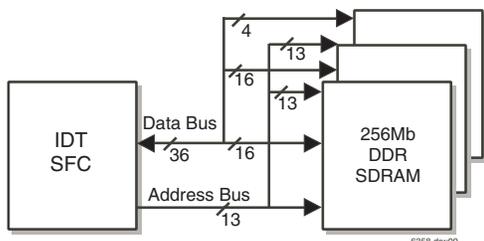
SFC Outputs	DDR SDRAM #1	DDR SDRAM #2
DQ[31:0]	DQ[31:0]	--
DQ[35:32]	--	DQ[3:0]
DQ[63:36]	--	--
DQS[3:0]	DQS[3:0]	--
DQS[7:4]	--	DQS[3:0]
A[11:0]	A[11:0]	A[11:0]
CK, $\overline{CK}$	CK, $\overline{CK}$	CK, $\overline{CK}$
$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$
BA[1:0]	BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>		
CKE → Vcc		
$\overline{CS}$ → GND		
DM[3:0] → GND		
DQ[31:4] → Vcc		
<b>SFC Hard wired pins</b>		
A12 → Vcc		

**TABLE 4 – SFC TO DDR SDRAM INTERFACE CONNECTIONS(Continued)**



**CONFIGURATION 5**

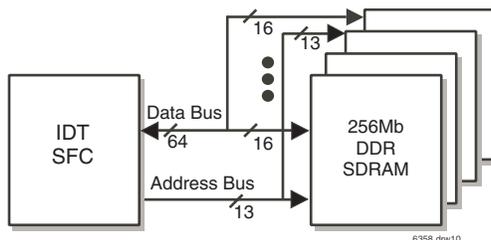
SFC Outputs	DDR SDRAM #1	DDR SDRAM #2
DQ[15:0]	DQ[15:0]	--
DQ[31:16]	--	DQ[15:0]
DQS0	LDQS	--
DQS1	UDQS	--
DQS2	--	LDQS
DQS3	--	UDQS
A[12:0]	A[12:0]	A[12:0]
CK, $\overline{CK}$	CK, $\overline{CK}$	CK, $\overline{CK}$
$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$
BA[1:0]	BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>		
CKE → Vcc		
$\overline{CS}$ → GND		
LDM → GND		
UDM → GND		
<b>SFC Hard wired pins</b>		
DQ[63:32] → Vcc		
DQS[7:4] → Vcc		



**CONFIGURATION 6**

SFC Outputs	DDR SDRAM #1	DDR SDRAM #2	DDR SDRAM #3
DQ[15:0]	DQ[15:0]	--	--
DQ[31:16]	--	DQ[15:0]	--
DQ[35:32]	--	--	DQ[3:0]
DQS0	LDQS	--	--
DQS1	UDQS	--	--
DQS2	--	LDQS	--
DQS3	--	UDQS	--
DQS4	--	--	LDQS
DQS5	--	--	UDQS
A[12:0]	A[12:0]	A[12:0]	A[12:0]
CK, $\overline{CK}$	CK, $\overline{CK}$	CK, $\overline{CK}$	CK, $\overline{CK}$
$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$	$\overline{RAS}$ , $\overline{CAS}$
BA[1:0]	BA[1:0]	BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$	$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>			
CKE → Vcc			
$\overline{CS}$ → GND			
LDM → GND			
UDM → GND			
DQ[15:4] → Vcc			
<b>SFC Hard wired pins</b>			
DQ[63:36] → Vcc			
DQS[7:6] → Vcc			

TABLE 4 – SFC TO DDR SDRAM INTERFACE CONNECTIONS(Continued)



**CONFIGURATION 7**

SFC Outputs	DDR SDRAM #1	DDR SDRAM #2	DDR SDRAM #3	DDR SDRAM #4
DQ[15:0]	DQ[15:0]	--	--	--
DQ[31:16]	--	DQ[15:0]	--	--
DQ[47:32]	--	--	DQ[15:0]	--
DQ[63:48]	--	--	--	DQ[15:0]
DQS0	LDQS	--	--	--
DQS1	UDQS	--	--	--
DQS2	--	LDQS	--	--
DQS3	--	UDQS	--	--
DQS4	--	--	LDQS	--
DQS5	--	--	UDQS	--
DQS6	--	--	--	LDQS
DQS7	--	--	--	UDQS
A[12:0]	A[12:0]	A[12:0]	A[12:0]	A[12:0]
CK, $\overline{CK}$				
$\overline{RAS}$ , $\overline{CAS}$				
BA[1:0]	BA[1:0]	BA[1:0]	BA[1:0]	BA[1:0]
$\overline{WE}$	$\overline{WE}$	$\overline{WE}$	$\overline{WE}$	$\overline{WE}$
<b>DDR SDRAM Hard wired pins</b>				
CKE → Vcc				
$\overline{CS}$ → GND				
LDM → GND				
UDM → GND				

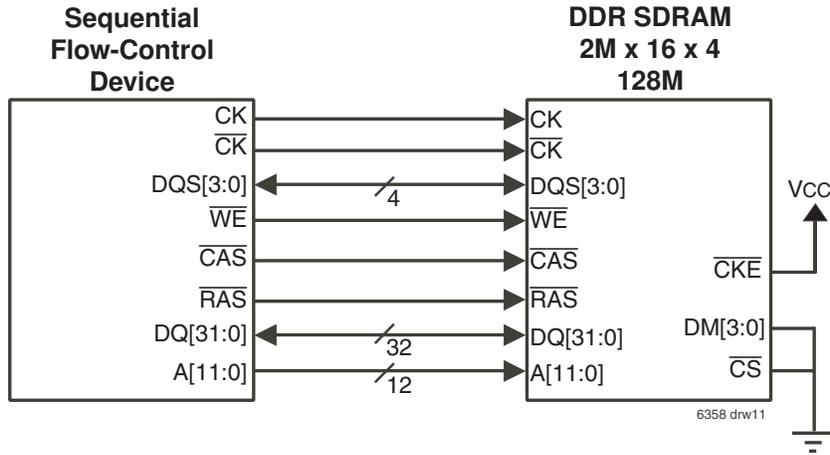


Figure 3. Memory Interface Connection (Single Chip)

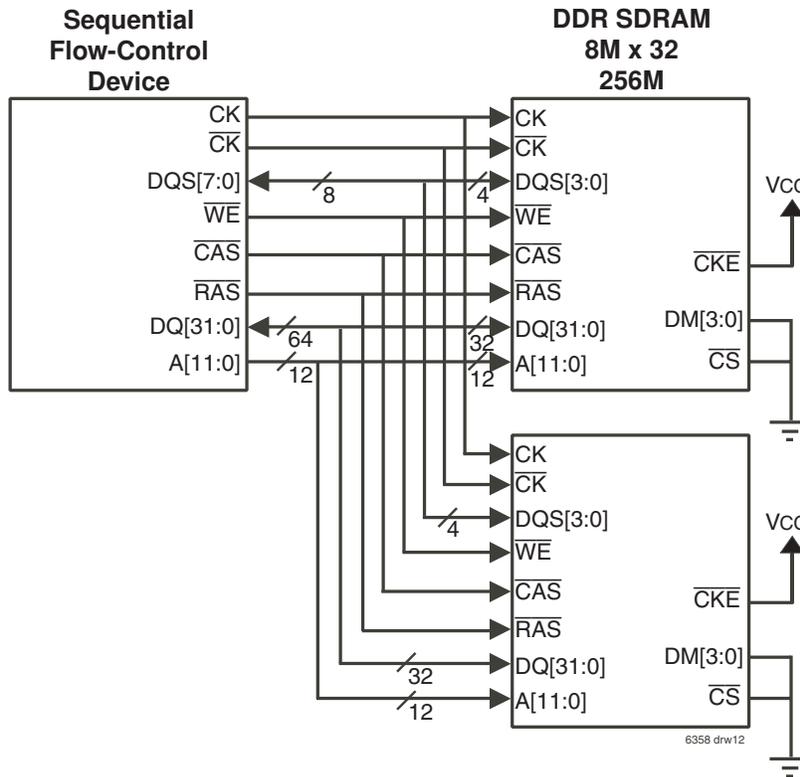


Figure 4. Memory Interface Connection (Two Chip)

**TOTAL AVAILABLE MEMORY USAGE**

The sequential flow-control (SFC) is designed to efficiently use as much of the DDR SDRAM memory as possible, but due to the discontinuity between the SFC bus width (x48) and the DDR SDRAM interface (x16 or x32), some columns in a row of the SDRAM will not be used. As a result, the total usable memory will be slightly less than the total available memory in the SDRAM. Table 5 outlines the total usable memory for the various configurations depending on

whether or not the Error Detection and Correction (EDC) feature is selected. If the EDC feature is selected, 8 syndrome bits will be generated per every 64 bits of data. Therefore every third write burst to the SDRAM will send out the 8 syndrome bits, resulting in 24 unused bits in the column. Therefore, using the EDC feature, there will be significantly less usable memory of data storage. The EDC function is described in the Error Detection and Correction section of this datasheet.

**TABLE 5 – TOTAL USEABLE MEMORY BASED ON VARIOUS CONFIGURATIONS**

	Total DDR SDRAM Density	Total Usable Memory (EDC off)	Total Usable Memory (EDC on)
<b>Configuration 1</b>			
1 x [4Mb x 32]	128Mb	108Mb	72Mb
1 x [8Mb x 32]	256Mb	252Mb	144Mb
<b>Configuration 2</b>			
1 x [16Mb x 16]	256Mb	216Mb	144Mb
<b>Configuration 3</b>			
2 x [4Mb x 32]	256Mb	216Mb	144Mb
2 x [8Mb x 32]	512Mb	504Mb	288Mb
<b>Configuration 4</b>			
2 x [4Mb x 32]	256Mb	122Mb	108Mb
2 x [8Mb x 32]	512Mb	284Mb	252Mb
<b>Configuration 5</b>			
2 x [8Mb x 16]	256Mb	252Mb	144Mb
<b>Configuration 6</b>			
3 x [8Mb x 16]	384Mb	284Mb	252Mb
3 x [16Mb x 16]	768Mb	567Mb	504Mb
<b>Configuration 7</b>			
4 x [16Mb x 16]	1Gb	1008Mb	576Mb

**MAXIMUM I/O OPERATING FREQUENCY**

The sequential flow-control (SFC) device is designed to operate at the maximum frequency of 133MHz. There are certain configurations however, that can increase or decrease the maximum frequency of the input and output ports. In some configurations (e.g. x24 I/O width), the I/O speeds can run up to

166MHz. The main factors that determine the usable memory are the I/O bus-width of the SFC, the density and number of DDR SDRAMs connected, and whether or not EDC is used. Tables 6 and 7 lists the maximum frequency for the input and output ports of the SFC based on the various configurations.

**TABLE 6 – IDT72T6480 MAXIMUM FREQUENCY  
BASED ON 166MHz DDR SDRAM**

	Bus-Width x48		Bus-Width x24		Bus-Width x12	
	EDC On	EDC Off	EDC On	EDC Off	EDC On	EDC Off
<b>Configuration 1</b>	50MHz	66MHz	100MHz	133MHz	166MHz	166MHz
<b>Configuration 2</b>	33MHz	33MHz	66MHz	66MHz	133MHz	133MHz
<b>Configuration 3</b>	100MHz	133MHz	166MHz	166MHz	166MHz	166MHz
<b>Configuration 4</b>	66MHz	66MHz	133MHz	133MHz	166MHz	166MHz
<b>Configuration 5</b>	50MHz	66MHz	100MHz	133MHz	166MHz	166MHz
<b>Configuration 6</b>	66MHz	66MHz	133MHz	133MHz	166MHz	166MHz
<b>Configuration 7</b>	100MHz	133MHz	166MHz	166MHz	166MHz	166MHz

**TABLE 7 – IDT72T6480 MAXIMUM FREQUENCY  
BASED ON 133MHz DDR SDRAM**

	Bus-Width x48		Bus-Width x24		Bus-Width x12	
	EDC On	EDC Off	EDC On	EDC Off	EDC On	EDC Off
<b>Configuration 1</b>	33MHz	50MHz	66MHz	100MHz	133MHz	166MHz
<b>Configuration 2</b>	25MHz	25MHz	50MHz	66MHz	100MHz	133MHz
<b>Configuration 3</b>	66MHz	100MHz	133MHz	166MHz	166MHz	166MHz
<b>Configuration 4</b>	50MHz	50MHz	100MHz	100MHz	166MHz	166MHz
<b>Configuration 5</b>	33MHz	50MHz	66MHz	100MHz	133MHz	166MHz
<b>Configuration 6</b>	50MHz	50MHz	100MHz	100MHz	166MHz	166MHz
<b>Configuration 7</b>	66MHz	100MHz	133MHz	166MHz	166MHz	166MHz

**ERROR DETECTION AND CORRECTION**

The Error Detection and Correction (EDC) feature is available to ensure data integrity between the DDR SDRAM interface and the SFC. The EDC corrects all single bit hard and soft errors that are accessed from the DDR SDRAM. Multiple bit errors are not detected nor corrected.

The EDC logic blocks consist of a check bit generator and error detection correction logic. When the EDC is enabled, the check bit generator will generate 8 syndrome bits on the 8-byte boundary. The 8 syndrome bits are written into the DDR SDRAM along with the data. The SFC will burst write two cycles for data, and one cycle for syndrome bits. In order to minimize overhead and

increase throughput, not all memory in the DDR SDRAM is utilized. Table 5 lists the total usable memory for all 7 configurations when the EDC is enabled.

When a read operation is performed, the syndrome bits will be transferred to the error detection correction logic block and decoded to determine whether there are any single bit errors on the data. Single bit errors will be corrected and data is passed through to the QP cache.

The EDC is enabled using the MIC[2:0] pins. When the EDC is enabled, the dynamics of the total usable memory in the DDR SDRAM and the SFC operating speed will vary, listed in Tables 6 and 7. Table 8 shows how to enable the EDC feature for the 7 configurations

**TABLE 8 – MIC[2:0] CONFIGURATIONS**

	EDC Off	EDC On
Configuration 1	MIC [2:0] = 000	MIC [2:0] = 010
Configuration 2	MIC [2:0] = 001	MIC [2:0] = 011
Configuration 3	MIC [2:0] = 111	MIC [2:0] = 101
Configuration 4	MIC [2:0] = 100	MIC [2:0] = 110
Configuration 5	MIC [2:0] = 000	MIC [2:0] = 010
Configuration 6	MIC [2:0] = 100	MIC [2:0] = 110
Configuration 7	MIC [2:0] = 111	MIC [2:0] = 101

**SETTING THE MEMORY INTERFACE SIGNALS**

The configurations listed in Figure 2a-2g can be programmed into the sequential flow-control device by using the MIC[2:0], MTYPE[1:0], and

MSPEED. For information about these signals, please refer to the Signal Description section. Table 9 is a list that shows the settings for the different configurations.

**TABLE 9 – MEMORY CONFIGURATIONS SETTINGS**

	MIC[2:0]	MTYPE[1:0]	MSPEED
<b>Configuration 1</b>	000 - EDC Off 010 - EDC On	00 - (4Mb x 32) 10 - (8Mb x 32)	0 - 133MHz 1 - 166MHz
<b>Configuration 2</b>	001 - EDC Off 011 - EDC On	— 11 - (16Mb x 16)	0 - 133MHz 1 - 166MHz
<b>Configuration 3</b>	111 - EDC Off 101 - EDC On	00 - (4Mb x 32) 10 - (8Mb x 32)	0 - 133MHz 1 - 166MHz
<b>Configuration 4</b>	110 - EDC Off 100 - EDC On	00 - (4Mb x 32) 10 - (8Mb x 32)	0 - 133MHz 1 - 166MHz
<b>Configuration 5</b>	000 - EDC Off 010 - EDC On	— 11 - (16Mb x 16)	0 - 133MHz 1 - 166MHz
<b>Configuration 6</b>	110 - EDC Off 100 - EDC On	— 11 - (16Mb x 16)	0 - 133MHz 1 - 166MHz
<b>Configuration 7</b>	111 - EDC Off 101 - EDC On	— 11 - (16Mb x 16)	0 - 133MHz 1 - 166MHz

**TABLE 10 – DEVICE CONFIGURATION**

Signal Pins	Static State	Configuration
$\overline{\text{ASYR}}$	0	Read port configured in asynchronous mode
	1	Read port configured in synchronous mode
$\overline{\text{ASYW}}$	0	Write port configured in asynchronous mode
	1	Write port configured in synchronous mode
BM[3:0]	—	See Table 13 - Bus-Matching Configurations
FSEL[1:0]	00	Programmable flag register offset value = 127
	01	Programmable flag register offset value = 1,023
	10	Programmable flag register offset value = 4,095
	11	Programmable flag register offset value = 16,383
FWFT	0	IDT Standard mode
	1	FWFT mode
IDEM	0	Depth expansion in FWFT mode
	1	Depth expansion in IDT Standard mode
IOSEL	0	I/O voltage set to 3.3V levels
	1	I/O voltage set to 2.5V levels
JSEL	0	JTAG function is disabled
	1	JTAG function is enabled
MIC[2:0]	—	See Table 8 - MIC[2:0] Configurations for description
MSPEED	0	External memory interface clocks set to 133MHz
	1	External memory interface clocks set to 166MHz
MTYPE[1:0]	00	External memory configuration is: 4M x 32
	01	Not used
	10	External memory configuration is: 8M x 32
	11	External memory configuration is: 16M x 16

**TABLE 11 – DEFAULT PROGRAMMABLE FLAG OFFSETS**

FSEL1	FSEL0	Offset n,m
0	0	127
0	1	1,023
1	0	4,095
1	1	16,383

**TABLE 12 – NUMBER OF BITS REQUIRED FOR OFFSET REGISTERS**

Write Port Bus-Width	x48		x24		x12	
	EDC On	EDC Off	EDC On	EDC Off	EDC On	EDC Off
Configuration 1 (128Mb)	21	22	22	23	23	24
Configuration 1 (256Mb)	22	23	23	24	24	25
Configuration 2 (256Mb)	22	23	23	24	24	25
Configuration 3 (256Mb)	22	23	23	24	24	25
Configuration 3 (512Mb)	23	24	24	25	25	26
Configuration 4 (256Mb)	22	22	23	23	24	24
Configuration 4 (512Mb)	23	23	24	24	25	25
Configuration 5 (512Mb)	23	24	24	25	25	26
Configuration 6 (768Mb)	24	24	25	25	26	26
Configuration 7 (1Gb)	24	25	25	26	26	27

**FUNCTIONAL DESCRIPTIONS**

**MASTER RESET AND DEVICE CONFIGURATION**

During master reset the sequential flow-control configuration and settings are determined, this includes the following:

1. Synchronous or Asynchronous read and write port operation
2. Bus-width configuration
3. Default offset register values
4. IDT standard or first word fall through (FWFT) timing mode
5. Depth expansion in IDT standard or FWFT mode
6. I/O voltage set to 2.5V or 3.3V levels
7. JTAG function enabled or disabled
8. Configuration of the external memory interface

The state of the configuration inputs during master reset will determine which of the above modes are selected. A master reset comprises of pulsing the  $\overline{\text{MRS}}$  input pin from high to low for a period of time ( $t_{RS}$ ) with the configuration inputs held in their respective states. Table 10, Device Configuration summarizes the configuration modes available during master reset. These signals are described in detail in the signal description section.

**PROGRAMMABLE ALMOST EMPTY/ALMOST FULL FLAGS**

The SFC has a set of programmable flags ( $\overline{\text{PAE}}$ / $\overline{\text{PAF}}$ ) that can be used as an early indicator for the empty and full boundary conditions. These flags have an offset value (n, m) that will determine the almost empty and almost full boundary conditions. There are four default offset values selectable during master reset, these values are shown in Table 11, Default Programmable Flag Offsets.

Offset values can also be programmed using the serial programming pins (SCLK, SI, and SWEN). The SFC has two internal offset registers that are used to store the specific offset value, one for the  $\overline{\text{PAE}}$  and one for the  $\overline{\text{PAF}}$ . The total number of bits (shown in Table 12, Number of Bits Required for Offset Registers) must be completely programmed to the offset registers. The serial programming sequence begins by writing data into the  $\overline{\text{PAE}}$  register followed by the  $\overline{\text{PAF}}$  register. See Figure 29, *Serial Loading of Programmable Flag Registers* for the associated timing diagram. The total number of bits required to program the offset registers will vary depending on the type of configuration that is shown in Figure 2a-2g, the bus-width selected, and whether EDC is used.

The values of n, m are used such that the  $\overline{\text{PAE}}$  will become active (LOW) when there are at least one to n words written in the device. Similarly  $\overline{\text{PAF}}$  will become active (LOW) when there are at least D – M words or more in the device, where D is the density of the SFC.

## SIGNAL DESCRIPTIONS

### INPUTS

#### DATA INPUTS (D<sub>0</sub> - D<sub>47</sub>)

Data inputs for 48-bit wide data (D<sub>0</sub> - D<sub>47</sub>), data inputs for 24-bit wide data (D<sub>0</sub> - D<sub>23</sub>) or data inputs for 12-bit wide data (D<sub>0</sub> - D<sub>11</sub>).

### CONTROLS

#### MASTER RESET ( $\overline{MRS}$ )

A Master Reset is accomplished whenever the  $\overline{MRS}$  input is toggled LOW then HIGH. This operation sets the internal read and write pointers to the first location of the RAM array.  $\overline{PAE}$  will go LOW,  $\overline{PAF}$  will go HIGH.

If FWFT is LOW during Master Reset then the IDT Standard mode, along with  $\overline{EF}$  and  $\overline{FF}$  are selected.  $\overline{EF}$  will go LOW and  $\overline{FF}$  will go HIGH. If FWFT is HIGH, then the First Word Fall Through mode (FWFT), along with  $\overline{IR}$  and  $\overline{OR}$ , are selected.  $\overline{OR}$  will go HIGH and  $\overline{IR}$  will go LOW.

All configuration control signals must be set prior to the LOW to HIGH transition of  $\overline{MRS}$ .

During a Master Reset, the output register is initialized to all zeroes. A Master Reset is required after power up, before a write operation can take place.  $\overline{MRS}$  is an asynchronous function.

See Figure 6, *Master Reset and Initialization*, for the relevant timing diagram.

#### PARTIAL RESET ( $\overline{PRS}$ )

A Partial Reset is accomplished whenever the  $\overline{PRS}$  input is toggled LOW then HIGH. As in the case of the Master Reset, the internal read and write pointers are set to the first location of the RAM array,  $\overline{PAE}$  goes LOW, and  $\overline{PAF}$  goes HIGH.

Whichever mode is active at the time of Partial Reset, IDT Standard mode or First Word Fall Through, that mode will remain selected. If the IDT Standard mode is active, then  $\overline{FF}$  will go HIGH and  $\overline{EF}$  will go LOW. If the First Word Fall Through mode is active, then  $\overline{OR}$  will go HIGH, and  $\overline{IR}$  will go LOW.

Following Partial Reset, all values held in the offset registers remain unchanged. The output register is initialized to all zeroes.  $\overline{PRS}$  is asynchronous.

A Partial Reset is useful for resetting the device during the course of operation, when reprogramming programmable flag offset settings may not be convenient.

See Figure 7, *Partial Reset*, for the relevant timing diagram.

#### ASYNCHRONOUS WRITE ( $\overline{ASYW}$ )

The write port can be configured for either synchronous or asynchronous mode of operation. If during Master Reset the  $\overline{ASYW}$  input is LOW, then asynchronous operation of the write port will be selected. During asynchronous operation of the write port the WCLK input becomes WR input, this is the asynchronous write strobe input. A rising edge on WR will write data present on the data inputs into the sequential flow-control device (SFC). ( $\overline{WEN}$  must be LOW when using the write port in asynchronous mode).

When the write port is configured for asynchronous operation the device must be operating on IDT standard mode, FWFT mode is not permissible. The full flag ( $\overline{FF}$ ) and programmable almost full flag ( $\overline{PAF}$ ) operates in an asynchronous manner, that is, the full flag and  $\overline{PAF}$  flag will be updated based in both a write operation and read operation. Note, if asynchronous mode is selected, FWFT is not permissible. Refer to Figure 24, *Asynchronous Write and  $\overline{PAE}$  flag – IDT Standard mode* and Figure 25, *Asynchronous Write and  $\overline{PAF}$  flag – IDT Standard mode* for relevant timing and operational waveforms.

#### ASYNCHRONOUS READ ( $\overline{ASYR}$ )

The read port can be configured for either synchronous or asynchronous mode of operation. If during a Master Reset the  $\overline{ASYR}$  input is LOW, then

asynchronous operation of the read port will be selected. During asynchronous operation of the read port the RCLK input becomes RD input, this is the asynchronous read strobe input. A rising edge on RD will read data from the SFC via the output register and data output port. ( $\overline{REN}$  must be tied LOW during asynchronous operation of the read port).

The  $\overline{OE}$  input provides three-state control of the Q<sub>n</sub> output bus, in an asynchronous manner.

When the read port is configured for asynchronous operation the device must be operating on IDT standard mode, FWFT mode is not permissible if the read port is asynchronous. The Empty Flag ( $\overline{EF}$ ) and programmable almost empty flag ( $\overline{PAE}$ ) operates in an asynchronous manner, that is, the empty flag and  $\overline{PAE}$  will be updated based on both a read operation and a write operation. Refer to Figure 23, *Asynchronous Read and  $\overline{PAF}$  flag – IDT Standard mode*, Figure 26, *Asynchronous Empty Boundary – IDT Standard mode*, Figure 27, *Asynchronous Full Boundary – IDT Standard mode*, and Figure 28, *Asynchronous Read and  $\overline{PAE}$  flag – IDT Standard mode*, for relevant timing and operational waveforms.

#### FIRST WORD FALL THROUGH (FWFT)

During Master Reset, the state of the FWFT input determines whether the device will operate in IDT standard mode or First Word Fall Through (FWFT) mode.

If, at the time of Master Reset, FWFT is LOW, then IDT Standard mode will be selected. This mode uses the Empty Flag ( $\overline{EF}$ ) to indicate whether or not there are any words present in the SFC. It also uses the Full Flag function ( $\overline{FF}$ ) to indicate whether or not the SFC has any free space for writing. In IDT Standard mode, every word read from the SFC, including the first, must be requested using the Read Enable ( $\overline{REN}$ ) and RCLK.

If, at the time of Master Reset, FWFT is HIGH, then FWFT mode will be selected. This mode uses Output Ready ( $\overline{OR}$ ) to indicate whether or not there is valid data at the data outputs (Q<sub>n</sub>). It also uses Input Ready ( $\overline{IR}$ ) to indicate whether or not the SFC has any free space for writing. In the FWFT mode, the first word written to an empty SFC goes directly to Q<sub>n</sub> after three RCLK rising edges,  $\overline{REN} = \text{LOW}$  is not necessary. Subsequent words must be accessed using the Read Enable ( $\overline{REN}$ ) and RCLK.

#### WRITE STROBE AND WRITE CLOCK (WR/WCLK)

If synchronous operation of the write port has been selected via  $\overline{ASYW}$ , this input behaves as WCLK.

A write cycle is initiated on the rising edge of the WCLK input. Data setup and hold times must be met with respect to the LOW-to-HIGH transition of the WCLK. It is permissible to stop the WCLK. Note that while WCLK is idle, the  $\overline{FF}/\overline{IR}$ , and  $\overline{PAF}$  flags will not be updated. The Write and Read Clocks can either be independent or coincident.

If asynchronous operation has been selected this input is WR (write strobe). Data is asynchronously written into the SFC via the D<sub>n</sub> inputs whenever there is a rising edge on WR. In this mode the  $\overline{WEN}$  input must be LOW.

#### WRITE ENABLE ( $\overline{WEN}$ )

When the  $\overline{WEN}$  input is LOW, data may be loaded into the SFC on the rising edge of every WCLK cycle if the device is not full. Data is stored in the RAM array sequentially and independently of any ongoing read operation.

When  $\overline{WEN}$  is HIGH, no new data is written in the SFC.

To prevent data overflow in the IDT Standard mode,  $\overline{FF}$  will go LOW, inhibiting further write operations. Upon the completion of a valid read cycle,  $\overline{FF}$  will go HIGH allowing a write to occur. The  $\overline{FF}$  is updated by two WCLK cycles + t<sub>SKEW</sub> after the RCLK cycle.

To prevent data overflow in the FWFT mode,  $\overline{IR}$  will go HIGH, inhibiting further write operations. Upon the completion of a valid read cycle,  $\overline{IR}$  will go LOW allowing a write to occur. The  $\overline{IR}$  flag is updated by two WCLK cycles + tSKEW after the valid RCLK cycle.

$\overline{WEN}$  is ignored when the SFC is full in either FWFT or IDT Standard mode.

If asynchronous operation of the write port has been selected, then  $\overline{WEN}$  must be held active.

### READ STROBE AND READ CLOCK (RD/RCLK)

If synchronous operation of the read port has been selected via  $\overline{ASYR}$ , this input behaves as RCLK. A read cycle is initiated on the rising edge of the RCLK input. Data can be read on the outputs, on the rising edge of the RCLK input. It is permissible to stop the RCLK. Note that while RCLK is idle, the  $\overline{EF}/\overline{OR}$  and  $\overline{PAE}$  flags will not be updated. The Write and Read Clocks can be independent or coincident.

If asynchronous operation has been selected this input is RD (Read Strobe). Data is asynchronously read from the SFC whenever there is a rising edge on RD. In this mode the  $\overline{REN}$  and  $\overline{RCS}$  inputs must be tied LOW. The  $\overline{OE}$  input is used to provide asynchronous control of the three-state Qn outputs.

### WRITE CHIP SELECT ( $\overline{WCS}$ )

The  $\overline{WCS}$  disables all Write data operations (data only) if it is held HIGH. To perform normal operations on the write port, the  $\overline{WCS}$  must be enabled, held LOW.

### READ ENABLE ( $\overline{REN}$ )

When Read Enable is LOW, data is loaded from the RAM array into the output register on the rising edge of every RCLK cycle if the device is not empty.

When the  $\overline{REN}$  input is HIGH, the output register holds the previous data and then no new data is loaded into the output register. The data outputs Q0-Qn maintain the previous data value.

In the IDT Standard mode, every word accessed at Qn, including the first word written to an empty cache, must be requested using  $\overline{REN}$  provided that  $\overline{RCS}$  is LOW. When the last word has been read from the SFC, the Empty Flag ( $\overline{EF}$ ) will go LOW, inhibiting further read operations.  $\overline{REN}$  is ignored when the SFC is empty. Once a write is performed,  $\overline{EF}$  will go HIGH allowing a read to occur. The  $\overline{EF}$  flag is updated by two RCLK cycles + tSKEW after the valid WCLK cycle. Both  $\overline{RCS}$  and  $\overline{REN}$  must be active, LOW for data to be read out on the rising edge of RCLK.

In the FWFT mode, the first word written to an empty SFC automatically goes to the outputs Qn, on the third valid LOW-to-HIGH transition of RCLK + tSKEW after the first write.  $\overline{REN}$  and  $\overline{RCS}$  do not need to be asserted LOW for the First Word to fall through to the output register. In order to access all other words, a read must be executed using  $\overline{REN}$  and  $\overline{RCS}$ . The RCLK LOW-to-HIGH transition after the last word has been read from the SFC, Output Ready ( $\overline{OR}$ ) will go HIGH with a true read (RCLK with  $\overline{REN} = \text{LOW}; \overline{RCS} = \text{LOW}$ ), inhibiting further read operations.  $\overline{REN}$  is ignored when the SFC is empty.

If asynchronous operation of the Read port has been selected, then  $\overline{REN}$  must be held active, (LOW).

### OUTPUT ENABLE ( $\overline{OE}$ )

When Output Enable is enabled (LOW), the parallel output buffers receive data from the output register. When  $\overline{OE}$  is HIGH, the output data bus (Qn) goes into a high impedance state. During Master or a Partial Reset the  $\overline{OE}$  is the only input that can place the output bus Qn, into High-Impedance. During Reset the  $\overline{RCS}$  input can be HIGH or LOW, it has no effect on the Qn outputs.

### READ CHIP SELECT ( $\overline{RCS}$ )

The Read Chip Select input provides synchronous control of the Read output port. When  $\overline{RCS}$  goes LOW, the next rising edge of RCLK causes the Qn outputs to go to the Low-Impedance state. When  $\overline{RCS}$  goes HIGH, the next RCLK rising edge causes the Qn outputs to return to HIGH-Z. During a Master or Partial Reset the  $\overline{RCS}$  input has no effect on the Qn output bus.  $\overline{OE}$  is the only input that provides High-Impedance control of the Qn outputs. If  $\overline{OE}$  is LOW the Qn data outputs will be Low-Impedance regardless of  $\overline{RCS}$  until the first rising edge of RCLK after a Reset is complete. Then if  $\overline{RCS}$  is HIGH the data outputs will go to High-Impedance.

The  $\overline{RCS}$  input does not effect the operation of the flags. For example, when the first word is written to an empty SFC, the  $\overline{EF}$  will still go from LOW to HIGH based on a rising edge of RCLK, regardless of the state of the  $\overline{RCS}$  input.

Also, when operating the SFC in FWFT mode the first word written to an empty SFC will still be clocked through to the output register based on RCLK, regardless of the state of  $\overline{RCS}$ . For this reason the user must take care when a data word is written to an empty SFC in FWFT mode. If  $\overline{RCS}$  is disabled when an empty SFC is written into, the first word will fall through to the output register, but will not be available on the Qn outputs which are in HIGH-Z. The user must take  $\overline{RCS}$  active LOW to access this first word, place the output bus in LOW-Z.  $\overline{REN}$  must remain disabled HIGH for at least one cycle after  $\overline{RCS}$  has gone LOW. A rising edge of RCLK with  $\overline{RCS}$  and  $\overline{REN}$  active LOW, will read out the next word. Care must be taken so as not to lose the first word written to an empty SFC when  $\overline{RCS}$  is HIGH. See Figure 15 for *Read Chip Select*. If asynchronous operation of the Read port has been selected, then  $\overline{RCS}$  must be held active, (tied LOW).  $\overline{OE}$  provides three-state control of Qn.

### BUS-MATCHING (BM[3:0])

These pins are used to define the input and output bus widths. During Master Reset, the state of these pins is used to configure the device bus sizes. All flags will operate on the word/byte size boundary as defined by the selection of bus width. See Figures 17-20 for *Bus-Matching Configurations*. See Table 13, Bus-Matching Configurations for the available configurations.

**TABLE 13 – BUS-MATCHINGS**

BM3	BM2	BM1	BM0	Read Bus Width	Write Bus Width
1	0	0	0	x48	x48
1	0	0	1	x24	x48
1	1	0	1	x12	x48
1	0	1	1	x48	x24
1	1	1	1	x48	x12
0	0	0	1	x24	x24
0	1	0	1	x12	x24
0	0	1	1	x24	x12
0	1	1	1	x12	x12

### FLAG SELECT (FSEL[1:0])

During master reset, these inputs will select one of four default values for the programmable flags  $\overline{PAE}$  and  $\overline{PAF}$ . The selected value (listed in Table 14 - MTYPE[1:0] Configurations) will apply to both  $\overline{PAE}$  and  $\overline{PAF}$  offset.

### MEMORY CONFIGURATION (MIC[2:0])

These signals enable the EDC feature of the device. See Table 8, MIC[2:0] Configurations for more information.

### MEMORY SPEED (MSPEED)

This pin is used to determine the memory interface clock speed (CK and  $\overline{CK}$ ) for the external memory used. If MSPEED is HIGH, external memory CK and  $\overline{CK}$  will be operating at 166MHz. If MSPEED is LOW, then the external memory CK and  $\overline{CK}$  will be operating at 133MHz.

### MASTER CLOCK (MCLK)

33MHz reference clock used to generate CK and  $\overline{CK}$  for external memory interface.

### MEMORY TYPE (MTYPE[1:0])

These signals select the density configuration of the external DDR SDRAM used. See Table 14, MTYPE[1:0] Configurations for selection of the memory density configuration.

**TABLE 14 – MTYPE[1:0] CONFIGURATIONS**

	Density Configurations			
	4M x 32	8M x 32	Reserved	16M x 16
MTYPE0	0	0	1	1
MTYPE1	0	1	0	1

### DEPTH EXPANSION MODE SELECT (IDEM)

This select pin is used for depth expansion configuration in IDT Standard mode. If this pin is tied HIGH, then the  $\overline{FF}/\overline{IR}$  signal will be inverted to provide a seamless depth expansion interface. If this pin is tied LOW, the depth expansion in IDT Standard mode will be deactivated. For details on depth expansion configuration, see Figure 34, *Depth Expansion Configuration in IDT Standard Mode* and Figure 35, *Depth Expansion Configuration in FWFT Mode*.

### SERIAL READ ENABLE ( $\overline{SREN}$ )

The serial read enable input is an enable used for reading the value of the programmable offset registers. By setting the JSEL pin to LOW, the serial data output (SO) and serial clock (SCLK) signals can be used with  $\overline{SREN}$  to program the offset registers. When  $\overline{SREN}$  is LOW, data at the SO can be read from the offset register, one bit for each LOW-to-HIGH transition of SCLK. When serial read enable is HIGH, the reading of the offset registers will stop.  $\overline{SREN}$  must be kept LOW in order to read the entire contents of the scan out register. If at any point  $\overline{SREN}$  is toggled HIGH, the read pointer of the offset registers will reset to the first location. The next time  $\overline{SREN}$  is enabled the first contents in the offset register will be read back. Serial read enable functions the same way in both IDT Standard and FWFT modes. See Figure 30, *Reading of Programmable Flag Registers*, for the timing diagram.

### SERIAL WRITE ENABLE ( $\overline{SWEN}$ )

The serial write enable input is an enable used for serial programming of the programmable offset registers. By setting the JSEL pin to LOW, the serial input (SI) and serial clock (SCLK) signals can be used with  $\overline{SWEN}$  to program the offset registers. When  $\overline{SWEN}$  is LOW, data at the SI input are loaded into the offset register, one bit for each LOW-to-HIGH transition of SCLK. When  $\overline{SWEN}$  is HIGH, the offset registers retain the previous settings and no offsets are loaded. Serial write enable functions the same way in both Standard IDT and FWFT

modes. See Figure 29, *Serial Loading of Programmable Flag Registers*, for the timing diagram.

### I/O VDDQ SELECT (IOSEL)

This input determines whether the inputs and outputs will tolerate a 2.5V or 3.3V voltage signals. If IOSEL is HIGH, then all I/Os will be 2.5V levels. If IOSEL is LOW, then all I/Os will be 3.3V levels. See Table 15, Parameters affected by I/O selection for a list of affected I/O signals.

**TABLE 15 – PARAMETERS AFFECTED BY I/O SELECTION**

SFC I/O affected by I/O selection			DDR SDRAM I/O - NOT affected <sup>(1)</sup>	
$\overline{ASYR}$	MIC[2:0]	$\overline{RCS}$	A[12:0]	DQ[63:0]
$\overline{ASYW}$	MCLK	$\overline{REN}$	BA[1:0]	DQS[7:0]
BM[3:0]	MRS	$\overline{SREN}$	CK	RAS
D[47:0]	MSPEED	SWEN	$\overline{CK}$	WE
$\overline{EF}/\overline{OR}$	MTYPE[1:0]	TCK/SCLK	CAS	
$\overline{FF}/\overline{IR}$	$\overline{OE}$	TDI/SI		
FSEL[1:0]	PAE	TDO/SO		
FWFT	PAF	TMS		
IDEM	PRS	WCLK/WR		
IOSEL	Q[47:0]	$\overline{WCS}$		
JSEL	RCLK/RD	WEN		

**NOTE:**

1. I/O to DDR SDRAM is not affected by I/O voltage selection

### JTAG SELECT (JSEL)

This input determines whether the JTAG port will be activated or deactivated. If JSEL is HIGH, then the JTAG port is activated and the associated JTAG pins (TCK, TDI, TDO, TMS) are used for the boundary-scan function. If JSEL is LOW, the JTAG port is disabled and the serial programming pins (SCLK, SI, SO) will be used to program and read the offset register values for PAE and PAF. See Figure 29 and 30, *Serial Loading and Reading of Programmable Registers* for information on how to program the registers.

## OUTPUTS

### FULL FLAG/INPUT READY ( $\overline{FF}/\overline{IR}$ )

This is a dual purpose pin. In IDT Standard mode, the Full Flag ( $\overline{FF}$ ) function is selected. When the SFC is full,  $\overline{FF}$  will go LOW, inhibiting further write operations. When  $\overline{FF}$  is HIGH, the SFC is not full. If no reads are performed after a reset (either MRS or PRS),  $\overline{FF}$  will go LOW See Figure 12, *Full Boundary - IDT Standard Mode*, for the relevant timing information.

In FWFT mode, the Input Ready ( $\overline{IR}$ ) function is selected.  $\overline{IR}$  goes LOW when memory space is available for writing in data. When there is no longer any free space left,  $\overline{IR}$  goes HIGH, inhibiting further write operations. If no reads are performed after a reset (either MRS or PRS),  $\overline{IR}$  will go HIGH see Figure 9 *Write First Word Cycles - FWFT Mode*, for the relevant timing information.

The  $\overline{IR}$  status not only measures the contents of the SFC memory, but also counts the presence of a word in the output register. Thus, in FWFT mode, the total number of writes necessary to de-assert  $\overline{IR}$  is one greater than needed to assert  $\overline{FF}$  in IDT Standard mode.

$\overline{FF}/\overline{IR}$  is synchronous and updated on the rising edge of WCLK.  $\overline{FF}/\overline{IR}$  are double register-buffered outputs.