imall

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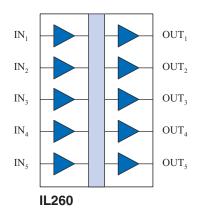
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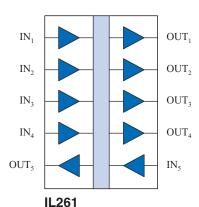


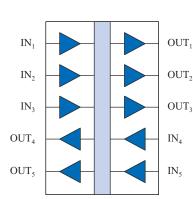


High Speed Five-Channel Digital Isolators

Functional Diagrams







IL262

Features

- High Speed: 110 Mbps
- 3 V to 5 V power supplies
- 1.5 mA/channel typical quiescent current
- -40°C to +85°C operating temperature
- 44000 year barrier life
- 2500 V_{RMS} isolation per UL 1577
- 2 ns typical pulse width distortion
- 100 ps typical pulse jitter
- 4 ns typical propagation delay skew
- 10 ns typical propagation delay
- 30 kV/µs typical common mode rejection
- Low EMC footprint
- 2 ns channel-to-channel skew
- 0.3" and 0.15" 16-pin SOIC packages
- UL 1577 recognized and IEC 61010-1 approved

Applications

- ADCs and DACs
- Multiplexed data transmission
- Board-to-board communication
- Peripheral interfaces
- Equipment covered under 61010-1 Edition 3
- 5 kV_{RMS} rated IEC 60601-1 medical applications

Description

NVE's IL260-Series five-channel high-speed digital isolators are CMOS devices manufactured with NVE's patented* IsoLoop[®] spintronic Giant Magnetoresistive (GMR) technology.

A unique ceramic/polymer composite barrier provides excellent isolation and virtually unlimited barrier life.

All transmit and receive channels operate at 110 Mbps over the full temperature and supply voltage range. The symmetric magnetic coupling barrier provides a typical propagation delay of only 10 ns and a pulse width distortion of 2 ns, achieving the best specifications of any isolator. The unique fifth channel can be is used to distribute isolated clocks or handshake signals to multiple delta-sigma A/D converters. High channel density makes these devices ideal for isolating ADCs and DACs, parallel buses and peripheral interfaces.

Typical transient immunity of 30 kV/ μ s is unsurpassed. Performance is specified over the temperature range of -40°C to +85°C without derating.

IL260-Series Isolators are available in 0.3" and 0.15" 16-pin SOIC packages. In the 0.15" packages, the five-channel devices provide the highest channel density available.

IsoLoop is a registered trademark of NVE Corporation. *U.S. Patent number 5,831,426; 6,300,617 and others.

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Absolute Maximum Ratings⁽¹⁾

Parameters	Symbol	Min.	Тур.	Max.	Units	Test Conditions
Storage Temperature	Ts	-55		150	°C	
Ambient Operating Temperature	T _A	-40		85	°C	
Supply Voltage	V_{DD1}, V_{DD2}	-0.5		7	V	
Input Voltage	VI	-0.5		$V_{DD} + 0.5$	V	
Output Voltage	Vo	-0.5		$V_{DD} + 0.5$	V	
Output Current Drive	Io	-10		10	mA	
Lead Solder Temperature				260	°C	10 sec.
ESD			2		kV	HBM

Recommended Operating Conditions

Parameters	Symbol	Min.	Тур.	Max.	Units	Test Conditions
Ambient Operating Temperature	T _A	-40		85	°C	
Supply Voltage	V_{DD1}, V_{DD2}	3.0		5.5	V	3.3/5.0 V Operation
Logic High Input Voltage	V _{IH}	2.4		V _{DD}	V	
Logic Low Input Voltage	V _{IL}	0		0.8	V	
Input Signal Rise and Fall Times	t_{IR}, t_{IF}			1	μs	

Insulation Specifications

Parameters		Symbol	Min.	Тур.	Max.	Units	Test Conditions
Creepage Distance	0.15" SOIC		4.0				
(external)	0.3" SOIC		8.08			mm	
Total Barrier Thickn	ess (internal)		0.012	0.013		mm	
Leakage Current ⁽⁵⁾				0.2		μA_{RMS}	240 V _{RMS}
Barrier Impedance ⁽⁵⁾				>10 ¹⁴ 7		$\Omega \parallel pF$	
Capacitance (Input-	Output) ⁽⁵⁾	C _{I-O}		5		pF	f = 1 MHz
· · · · · ·				44000		Years at	60% confidence level
Barrier Life				44000		100°C	activation energy

Safety and Approvals

IEC61010-1

TUV Certificate Numbers:

N1502812, N1502812-101

Classification as reinforced insulation:

Model	Package	Pollution Degree	Material Group	Max. Working Voltage
IL260, IL261, IL262	0.3" 16-pin SOIC	II	III	300 V _{RMS}
IL260-3, IL261-3, IL262-3	0.15" 16-pin SOIC	II	III	150 V _{RMS}

UL 1577

Component Recognition Program File Number: **E207481** Each part tested at 3000 V_{RMS} (4240 V_{PK}) for 1 second Each lot sample tested at 2500 V_{RMS} (3530 V_{PK}) for 1 minute

Soldering Profile

Per JEDEC J-STD-020C, MSL=2



IL260 Pin Connections

1	IN ₁	Input 1
2	GND ₁	Ground*
3	IN ₂	Input 2
4	IN ₃	Input 3
5	IN ₄	Input 4
6	V _{DD1}	Supply Voltage 1
7	IN ₅	Input 5
8	GND ₁	Ground*
9	GND ₂	Ground*
10	OUT ₅	Output 5
11	OUT ₄	Output 4
12	OUT ₃	Output 3
13	OUT ₂	Output 2
14	OUT ₁	Output 1
15	GND ₂	Ground*
16	V _{DD2}	Supply Voltage 2

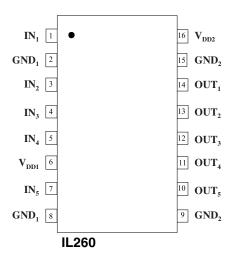
IL261 Pin Connections

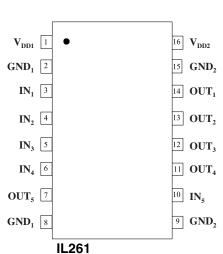
1	V _{DD1}	Supply Voltage 1
2	GND ₁	Ground*
3	IN ₁	Input 1
4	IN ₂	Input 2
5	IN ₃	Input 3
6	IN ₄	Input 4
7	OUT ₅	Output 5
8	GND ₁	Ground*
9	GND ₂	Ground*
10	IN ₅	Input 5
11	OUT ₄	Output 4
12	OUT ₃	Output 3
13	OUT ₂	Output 2
14	OUT ₁	Output 1
15	GND ₂	Ground*
16	V _{DD2}	Supply Voltage 2

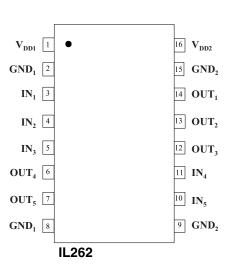
IL262 Pin Connections

1	V _{DD1}	Supply Voltage 1
2	GND ₁	Ground*
3	IN ₁	Input 1
4	IN ₂	Input 2
5	IN ₃	Input 3
6	OUT ₄	Output 4
7	OUT ₅	Output 5
8	GND ₁	Ground*
9	GND ₂	Ground*
10	IN ₅	Input 5
11	IN ₄	Input 4
12	OUT ₃	Output 3
13	OUT ₂	Output 2
14	OUT ₁	Output 1
15	GND ₂	Ground*
16	V _{DD2}	Supply Voltage 2

***NOTE:** Pins 2 and 8 are internally connected, as are pins 9 and 15.







IL260/IL261/IL262

3



IL260/IL261/IL262

3.3 Volt Electrical Specifications (T _{min} to T _{max})										
Parameters		Symbol	Min.	Тур.	Max.	Units	Test Conditions			
	IL260			300	400	μΑ				
Input Quiescent Current	IL261	I _{DD1}		1.5	2	mA				
_	IL262			3	4	mA				
	IL260			6.5	10	mA				
Output Quiescent Current	IL261	I _{DD2}		5	8	mA				
	IL262			3.5	6	mA				
Logic Input Current		Ii	-10		10	μΑ				
Logic High Output Voltage		V	V _{DD} -0.1	V _{DD}		V	$I_0 = -20 \ \mu A, V_I = V_{IH}$			
Logic High Output Voltage		V_{OH}	0.8 x V _{DD}	0.9 x V _{DD}		v	$I_0 = -4 \text{ mA}, V_I = V_{IH}$			
Logic Low Output Voltage		V		0	0.1	V	$I_0 = 20 \ \mu A, V_I = V_{IL}$			
Logic Low Output Voltage		V _{OL}		0.5	0.8	v	$I_0 = 4 \text{ mA}, V_I = V_{IL}$			

Switching Specifications ($V_{DD} = 3.3 \text{ V}$)										
Maximum Data Rate		100	110		Mbps	$C_L = 15 \text{ pF}$				
Minimum Pulse Width ⁽⁷⁾	PW	10			ns	50% Points, V _o				
Propagation Delay Input to Output (High to Low)	t _{PHL}		12	18	ns	$C_L = 15 \text{ pF}$				
Propagation Delay Input to Output (Low to High)	t _{PLH}		12	18	ns	$C_L = 15 \text{ pF}$				
Pulse Width Distortion $ t_{PHL}-t_{PLH} ^{(2)}$	PWD		2	3	ns	$C_{L} = 15 \text{ pF}$				
Propagation Delay Skew ⁽³⁾	t _{PSK}		4	6	ns	$C_L = 15 \text{ pF}$				
Output Rise Time (10%–90%)	t _R		2	4	ns	$C_L = 15 \text{ pF}$				
Output Fall Time (10%–90%)	t _F		2	4	ns	$C_L = 15 \text{ pF}$				
Common Mode Transient Immunity (Output Logic High to Logic Low) ⁽⁴⁾	$ CM_H , CM_L $	20	30		kV/μs	$V_{\rm CN} = 300 \text{ V}$				
Channel-to-Channel Skew			2	3	ns	$C_{L} = 15 \text{ pF}$				
Dynamic Power Consumption ⁽⁶⁾			140	240	µA/MHz	per channel				

Magnetic Field Immunity ⁽⁸⁾ ($V_{DD2} = 3V, 3V < V_{DD1} < 5.5V$)										
Power Frequency Magnetic Immunity	H_{PF}	1000	1500		A/m	50Hz/60Hz				
Pulse Magnetic Field Immunity	H _{PM}	1800	2000		A/m	$t_p = 8\mu s$				
Damped Oscillatory Magnetic Field	Hosc	1800	2000		A/m	0.1Hz – 1MHz				
Cross-axis Immunity Multiplier ⁽⁹⁾	K _X		2.5							

4

IL260/IL261/IL262



	5 Volt Electrical Specifications (T _{min} to T _{max})										
Parameters		Symbol	Min.	Тур.	Max.	Units	Test Conditions				
	IL260			350	500	μΑ					
Input Quiescent Current	IL261	I _{DD1}		2	3	mA					
_	IL262			4	6	mA					
	IL260			10	15	mA					
Output Quiescent Current	IL261	I _{DD2}		7.5	12	mA					
_	IL262			5	9	mA					
Logic Input Current		Ii	-10		10	μΑ					
Lagia High Output Valtage		V	V _{DD} -0.1	V _{DD}		v	$I_0 = -20 \ \mu A, V_I = V_{IH}$				
Logic High Output Voltage		V _{OH}	0.8 x V _{DD}	0.9 x V _{DD}		v	$I_0 = -4 \text{ mA}, V_1 = V_{1H}$				
		V		0	0.1	V	$I_0 = 20 \ \mu A, V_I = V_{IL}$				
Logic Low Output Voltage		V _{OL}		0.5	0.8	v	$I_0 = 4 \text{ mA}, V_1 = V_{1L}$				

Switching Specifications ($V_{DD} = 5 \text{ V}$)									
Maximum Data Rate		100	110		Mbps	$C_L = 15 \text{ pF}$			
Minimum Pulse Width ⁽⁷⁾	PW	10			ns	50% Points, Vo			
Propagation Delay Input to Output (High to Low)	t _{PHL}		10	15	ns	$C_L = 15 \text{ pF}$			
Propagation Delay Input to Output (Low to High)	t _{PLH}		10	15	ns	$C_L = 15 \text{ pF}$			
Pulse Width Distortion $ t_{PHL} - t_{PLH} ^{(2)}$	PWD		2	3	ns	$C_L = 15 \text{ pF}$			
Pulse Jitter ⁽¹⁰⁾	t _J		100		ps	$C_L = 15 \text{ pF}$			
Propagation Delay Skew ⁽³⁾	t _{PSK}		4	6	ns	$C_L = 15 \text{ pF}$			
Output Rise Time (10%–90%)	t _R		1	3	ns	$C_L = 15 \text{ pF}$			
Output Fall Time (10%–90%)	t _F		1	3	ns	$C_L = 15 \text{ pF}$			
Common Mode Transient Immunity (Output Logic High to Logic Low) ⁽⁴⁾	$ CM_{\rm H} , CM_{\rm L} $	20	30		kV/μs	$V_{CN} = 300 V$			
Channel-to-Channel Skew			2	3	ns	$C_L = 15 \text{ pF}$			
Dynamic Power Consumption ⁽⁶⁾			200	340	µA/MHz	per channel			

Magnetic Field Immunity ⁽⁸⁾ ($V_{DD2} = 5V, 3V < V_{DD1} < 5.5V$)						
Power Frequency Magnetic Immunity	H_{PF}	2800	3500		A/m	50Hz/60Hz
Pulse Magnetic Field Immunity	H _{PM}	4000	4500		A/m	$t_p = 8\mu s$
Damped Oscillatory Magnetic Field	Hosc	4000	4500		A/m	0.1Hz – 1MHz
Cross-axis Immunity Multiplier ⁽⁹⁾	K _X		2.5			

Notes (apply to both 3.3 V and 5 V specifications):

- 1. Absolute maximum means the device will not be damaged if operated under these conditions. It does not guarantee performance.
- 2. PWD is defined as $|t_{PHL} t_{PLH}|$. %PWD is equal to PWD divided by pulse width.
- 3. t_{PSK} is the magnitude of the worst-case difference in t_{PHL} and/or t_{PLH} between devices at 25°C.
- 4. CM_{H} is the maximum common mode voltage slew rate that can be sustained while maintaining $V_0 > 0.8 V_{DD2}$. CM_L is the maximum common mode input voltage that can be sustained while maintaining $V_0 < 0.8 V$. The common mode voltage slew rates apply to both rising and falling common mode voltage edges.
- 5. Device is considered a two terminal device: pins 1-8 shorted and pins 9-16 shorted.
- 6. Dynamic power consumption numbers are calculated per channel and are supplied by the channel's input side power supply.
- 7. Minimum pulse width is the minimum value at which specified PWD is guaranteed.
- 8. The relevant test and measurement methods are given in the Electromagnetic Compatibility section on p. 6.
- 9. External magnetic field immunity is improved by this factor if the field direction is "end-to-end" rather than to "pin-to-pin" (see diagram on p. 6).
- 10. 66,535-bit pseudo-random binary signal (PRBS) NRZ bit pattern with no more than five consecutive 1s or 0s; 800 ps transition time.



Application Information

Electrostatic Discharge Sensitivity

This product has been tested for electrostatic sensitivity to the limits stated in the specifications. However, NVE recommends that all integrated circuits be handled with appropriate care to avoid damage. Damage caused by inappropriate handling or storage could range from performance degradation to complete failure.

Electromagnetic Compatibility

IsoLoop Isolators have the lowest EMC footprint of any isolation technology. There are no internal clocks or carriers. IsoLoop Isolators' Wheatstone bridge configuration and differential magnetic field signaling ensure excellent EMC performance against all relevant standards.

These isolators are fully compliant with generic EMC standards EN50081, EN50082-1 and the umbrella line-voltage standard for Information Technology Equipment (ITE) EN61000. NVE has completed compliance tests in the categories below:

EN50081-1

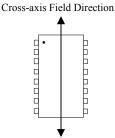
Residential, Commercial & Light Industrial Methods EN55022, EN55014

EN50082-2: Industrial Environment

Methods EN61000-4-2 (ESD), EN61000-4-3 (Electromagnetic Field Immunity), EN61000-4-4 (Electrical Transient Immunity), EN61000-4-6 (RFI Immunity), EN61000-4-8 (Power Frequency Magnetic Field Immunity), EN61000-4-9 (Pulsed Magnetic Field), EN61000-4-10 (Damped Oscillatory Magnetic Field) ENV50204

Radiated Field from Digital Telephones (Immunity Test)

Immunity to external magnetic fields is even higher if the field direction is "end-to-end" rather than to "pin-to-pin" as shown in the diagram below:



Dynamic Power Consumption

IsoLoop Isolators achieve their low power consumption from the way they transmit data across the isolation barrier. By detecting the edge transitions of the input logic signal and converting these to narrow current pulses, a magnetic field is created around the GMR Wheatstone bridge. Depending on the direction of the magnetic field, the bridge causes the output comparator to switch following the input logic signal. Since the current pulses are narrow, about 2.5 ns, the power consumption is independent of mark-to-space ratio and solely dependent on frequency. This has obvious advantages over optocouplers, which have power consumption heavily dependent on mark-to-space ratio.

Power Supply Decoupling

Both power supplies to these devices should be decoupled with low ESR 47 nF ceramic capacitors. Ground planes for both GND_1 and GND_2 are highly recommended for data rates above 10 Mbps. Capacitors must be located as close as possible to the V_{DD} pins.

Signal Status on Start-up and Shut Down

To minimize power dissipation, input signals are differentiated and then latched on the output side of the isolation barrier to reconstruct the signal. This could result in an ambiguous output state depending on power up, shutdown and power loss sequencing. Therefore, the designer should consider including an initialization signal in the start-up circuit. Initialization consists of toggling the input either high then low, or low then high.

Data Transmission Rates

The reliability of a transmission system is directly related to the accuracy and quality of the transmitted digital information. For a digital system, those parameters which determine the limits of the data transmission are pulse width distortion and propagation delay skew.

Propagation delay is the time taken for the signal to travel through the device. This is usually different when sending a low-to-high than when sending a high-to-low signal. This difference, or error, is called pulse width distortion (PWD) and is usually in nanoseconds. It may also be expressed as a percentage:

$$PWD\% = \frac{Maximum Pulse Width Distortion (ns)}{Signal Pulse Width (ns)} \times 100\%$$

For example, with data rates of 12.5 Mbps:

$$PWD\% = \frac{3 \text{ ns}}{80 \text{ ns}} \times 100\% = 3.75\%$$

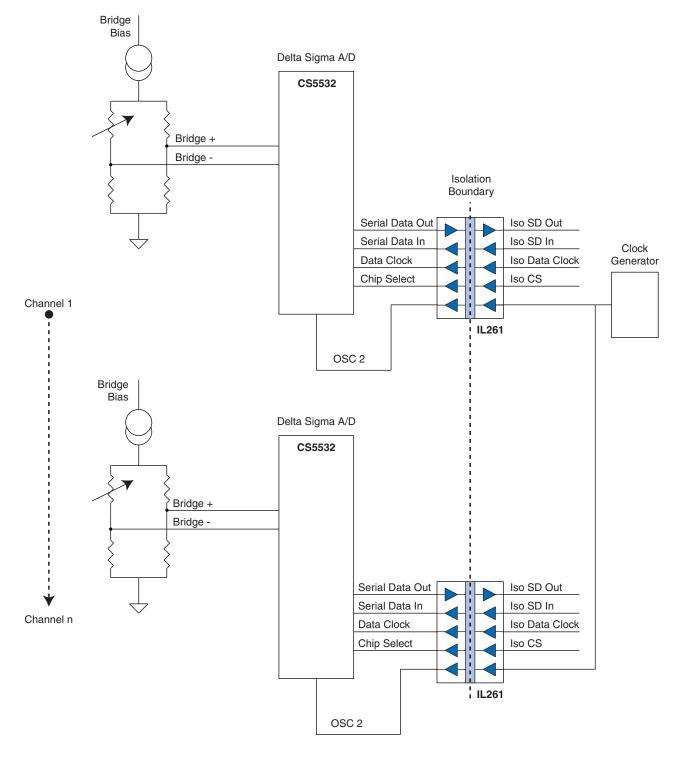
This figure is almost **three times** better than any available optocoupler with the same temperature range, and **two times** better than any optocoupler regardless of published temperature range. IsoLoop isolators exceed the 10% maximum PWD recommended by PROFIBUS, and will run to nearly 35 Mb within the 10% limit.

Propagation delay skew is the signal propagation difference between two or more channels. This becomes significant in clocked systems because it is undesirable for the clock pulse to arrive before the data has settled. Short propagation delay skew is therefore especially critical in high data rate parallel systems for establishing and maintaining accuracy and repeatability. Worstcase channel-to-channel skew in IL260-Series Isolators is only 3 ns, which is **ten times** better than any optocoupler. IL260-Series Isolators have a maximum propagation delay skew of 6 ns, which is **five times** better than any optocoupler.



Application Diagram—Multi-Channel Delta-Sigma A/D Converter

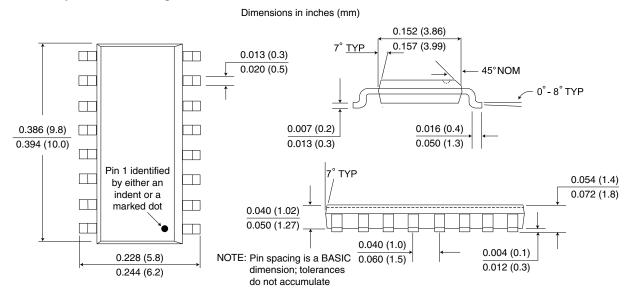
In a typical single-channel delta-sigma ADC, the system clock is located on the isolated side of the system and only four channels of isolation are required. With multiple ADCs configured in a channel-to-channel isolation configuration, however, clock jitter and edge placement accuracy of the system clock must be matched between ADCs. The best solution is to use a single clock on the system side and distribute the clock to each ADC. The five-channel IL261 is ideal, with the fifth channel used to distribute a single, isolated clock to multiple ADCs as shown below:



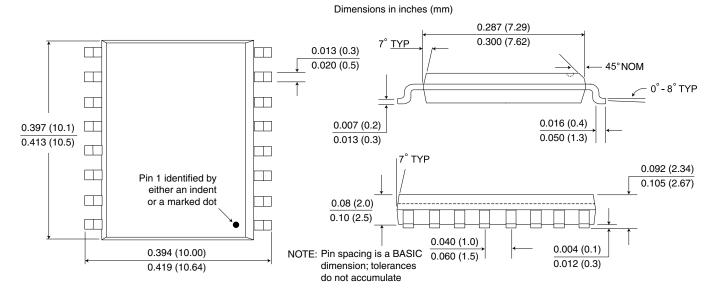


Package Drawings, Dimensions and Specifications

0.15" 16-pin SOIC Package

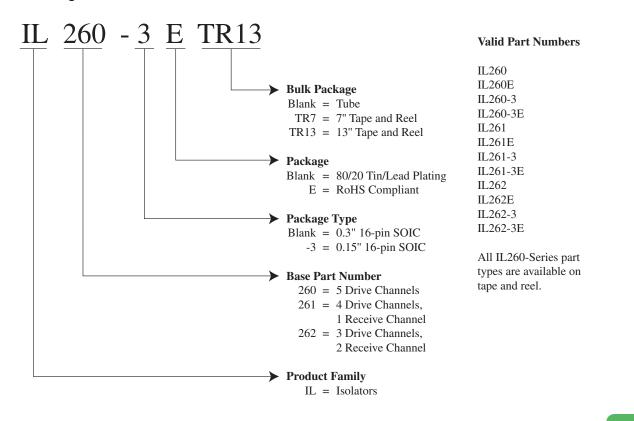


0.3" 16-pin SOIC Package





Ordering Information and Valid Part Numbers







IL260/IL261/IL262

Revision History

ISB-DS-001-IL260/1-N	Detailed isolation and barrier specifications.			
December 2012				
	Cosmetic changes.			
ISB-DS-001-IL260/1-M	Change			
	• Tightened typical output quiescent supply specs.			
ISB-DS-001-IL260/1/2-L	Change			
	• Update terms and conditions.			
ISB-DS-001-IL260/1/2-K	Change			
	• Added clarification of internal ground connections.			
ISB-DS-001-IL260/1/2-J	Change			
	• Relaxed Vdd1 quiescent current specification to 500µA.			
ISB-DS-001-IL260/1/2-I	Change			
	• Added typical jitter specification at 5V.			
ISB-DS-001-IL260/1/2-H	Change			
	• Added EMC details.			
ISB-DS-001-IL260/1/2-G	Change			
	• Added magnetic field immunity and electromagnetic compatibility specifications.			
ISB-DS-001-IL260/1/2-F	Change			
	Added IL262 configuration			
	• Added note on package drawings that pin-spacing tolerances are non accumulating.			
	• Changed ordering information to reflect that devices are fully RoHS compliant with no exemptions.			
ISB-DS-001-IL260/1-E	Change			
	Eliminated soldering profile chart			



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An ISO 9001 Certified Company

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