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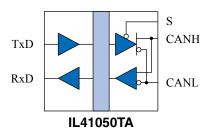






High-Speed, Low-Power Isolated CAN Transceiver

Functional Diagram



V _{DD2} (V)	$TxD^{(1)}$	S	CANH	CANL	Bus State	RxD
4.75 to 5.25	\rightarrow	Low ⁽²⁾	High	Low	Dominant	Low
4.75 to 5.25	X	High	$V_{\rm DD2}/2$	$V_{\rm DD2}/2$	Recessive	High
4.75 to 5.25	↑	X	$V_{\rm DD2}/2$	$V_{\rm DD2}/2$	Recessive	High
<2V (no pwr)	X	X	0 <v<2.5< td=""><td>0<v<2.5< td=""><td>Recessive</td><td>High</td></v<2.5<></td></v<2.5<>	0 <v<2.5< td=""><td>Recessive</td><td>High</td></v<2.5<>	Recessive	High
2 <v<sub>DD2<4.75</v<sub>	>2V	X	0 <v<2.5< td=""><td>0<v<2.5< td=""><td>Recessive</td><td>High</td></v<2.5<></td></v<2.5<>	0 <v<2.5< td=""><td>Recessive</td><td>High</td></v<2.5<>	Recessive	High

Table 1. Function table.

Notes:

- 1. TxD input is edge triggered: $\uparrow = \text{Logic Lo to Hi}, \downarrow = \text{Hi to Lo}$
- 2. Valid for logic state as described or open circuit
- X = don't care

Features

- 180 ns typical loop delay
- 70 mA maximum bus-side dynamic supply current
- 12 mA maximum quiescent recessive supply current
- 1 Mbps
- Fully compliant with the ISO 11898 CAN standard
- −55°C to +125°C operating temperature
- 3 V to 5.5 V power supplies
- >110-node fan-out
- 44000 year barrier life
- 2500 V_{RMS} isolation per UL 1577
- ±500 V CDM ESD
- 30 kV/µs transient immunity
- · Silent mode to disable transmitter
- Unpowered nodes do not disturb the bus
- Transmit data (TxD) dominant time-out function
- Edge triggered, non-volatile input improves noise performance
- Thermal shutdown protection
- Bus power short-circuit protection
- 0.15" and 0.3" and 16-pin JEDEC-standard SOIC packages
- UL 1577 recognized and IEC 61010-1 approved

Applications

- Factory automation
- Battery management systems
- · Noise-critical CAN
- DeviceNet
- Equipment covered under IEC 61010-1 Edition 3

Description

The IL41050TA is a galvanically isolated, CAN (Controller Area Network) transceiver, designed as the interface between the CAN protocol controller and the physical bus.

Quiescent and dynamic supply current is significantly lower than NVE's higher speed CAN (Controller Area Network) transceivers.

The IL41050 family provides isolated differential transmit capability to the bus and isolated differential receive capability to the CAN controller via NVE's patented* IsoLoop spintronic Giant Magnetoresistance (GMR) technology.

A unique ceramic/polymer composite barrier provides excellent isolation and virtually unlimited barrier life.

Advanced features facilitate reliable bus operation. Unpowered nodes do not disturb the bus, and a unique non-volatile programmable power-up feature prevents unstable nodes. The devices also have a hardware-selectable silent mode that disables the transmitter.

Designed for harsh CAN and DeviceNet environments, IL41050TA transceivers have transmit data dominant time-out, bus pin transient protection, a rugged Charged Device Model ESD rating, thermal shutdown protection, and short-circuit protection. Unique edgetriggered inputs improve noise performance.

www.IsoLoop.com

 $IsoLoop^{\$} is a registered trademark of NVE Corporation. \\ *U.S. Patent number 5,831,426; 6,300,617 and others.$

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Absolute Maximum Ratings(1)(2)

Parameter	Symbol	Min.	Тур.	Max.	Units	Test Conditions
Storage temperature	T_{s}	-55		150	°C	
Junction temperature	T_{J}	-55		150	°C	
Ambient operating temperature	T_A	-55		125	°C	
DC voltage at CANH and CANL pins	V_{CANH} , V_{CANL}	-45		45	V	0 V< V _{DD2} < 5.25 V; indefinite duration
Supply voltage	V_{DD1}, V_{DD2}	-0.3		7	V	
Digital input voltage	V_{TxD}, V_{S}	-0.3		$V_{DD} + 0.3$	V	
Digital output voltage	V_{RxD}	-0.3		$V_{DD} + 0.3$	V	
DC voltage at V _{REF}	$V_{ m REF}$	-0.3		$V_{DD} + 0.3$	V	
Transient voltage at CANH or CANL	$V_{trt(CAN)}$	-150		150	V	
Electrostatic discharge at all pins	V_{esd}	-4000		4000	V	Human body model
Electrostatic discharge at all pins	V_{esd}	-500		500	V	Machine model

Recommended Operating Conditions

Parameter	Symbol	Min.	Тур.	Max.	Units	Test Conditions
Supply voltage	$egin{array}{c} V_{ ext{DD1}} \ V_{ ext{DD2}} \end{array}$	3.0 4.75		5.5 5.25	V	
Junction temperature	T _J	-55		140	°C	
Input voltage at any bus terminal (separately or common mode)	$egin{array}{c} V_{ m CANH} \ V_{ m CANL} \end{array}$	-12		12	V	
High-level digital input voltage ⁽³⁾⁽⁴⁾	$ m V_{IH}$	2.0 2.4 2.0		$egin{array}{c} V_{ ext{DD1}} \ V_{ ext{DD1}} \ V_{ ext{DD2}} \end{array}$	V	$V_{DD1} = 3.3 \text{ V}$ $V_{DD1} = 5.0 \text{ V}$ $V_{DD2} = 5.0 \text{ V}$
Low-level digital input voltage ⁽³⁾⁽⁴⁾	$V_{\scriptscriptstyle \mathrm{IL}}$	0		0.8	V	
Digital output current (RxD)	I_{OH}	-8		8	mA	$V_{\rm DD1} = 3.3 \text{V to 5V}$
Ambient operating temperature	T_{A}	-55		125	°C	
Digital input signal rise and fall times	t_{IR}, t_{IF}			1	μs	

Insulation Specifications

Parameter		Symbol	Min.	Тур.	Max.	Units	Test Conditions
Creepage distance	IL41050TA-3E		4.0			******	
(external)	IL41050TAE		8.08			mm	
Total barrier thickne	ss (internal)		0.012	0.013		mm	
Barrier impedance				$> 10^{14} \parallel 7$		$\Omega \parallel pF$	
Leakage current				0.2		$\mu A_{ m RMS}$	$240 \text{ V}_{\text{RMS}}, 60 \text{ Hz}$
Barrier life				44000		Years at	60% confidence level
Darrier life				44000		100°C	activation energy

Thermal Characteristics

Parameter		Symbol	Min.	Тур.	Max.	Units	Test Conditions
Junction–ambient thermal resistance	IL41050TA-3E IL41050TAE	θ_{JA}		100 60		°C/W	Soldered to double-sided board;
Junction–case thermal resistance	IL41050TA-3E IL41050TAE	$\Psi_{\scriptscriptstyle JT}$		25 12		°C/W	free air
Power dissipation	IL41050TA-3E IL41050TAE	$P_{\scriptscriptstyle D}$			625 800	mW	





Safety and Approvals

IEC 61010-1 Approved

TUV Certificate Numbers: N1502812

Classification as reinforced Insulation:

Model	Package	Pollution Degree	Material Group	Max. Working Voltage
IL41050	SOIC (0.15" and 0.3")	II	III	$300 \mathrm{V_{RMS}}$

UL 1577 Recognized

Component Recognition Program File Number: E207481 Each part tested at 3000V_{RMS} (4240V_{PK}) for 1 second Each lot sample tested at 2500V_{RMS} (3530V_{PK}) for 1 minute

Soldering Profile

Per JEDEC J-STD-020C

Moisture Sensitivity Level: MSL=2

Notes:

- 1. Absolute Maximum specifications mean the device will not be damaged if operated under these conditions. It does not guarantee performance.
- 2. All voltages are with respect to network ground except differential I/O bus voltages.
- 3. The TxD input is edge sensitive. Voltage magnitude of the input signal is specified, but edge rate specifications must also be met.
- 4. The maximum time allowed for a logic transition at the TxD input is 1 μ s.



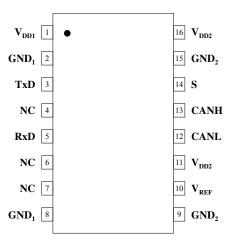
IL41050-3 Pin Connections (0.15" SOIC Package)

1	V_{DD1}	V _{DD1} power supply input
2	GND_1	V _{DD1} power supply ground return
3	TxD	Transmit Data input
4	RxD	Receive Data output
5	NC	No internal connection
6	NC	No internal connection
7	NC	No internal connection
8	NC	No internal connection
9	IsoRxD	Isolated RxD output. No connection should be made to this pin.
10	CANL	Low level CANbus line
11	V_{DD2}	V _{DD2} CAN I/O bus circuitry power supply input*
12	CANH	High level CANbus line
13	S	Mode select input. Leave open or set low for normal operation; set high for silent mode.
14	IsoTxD	Isolated TxD output. No connection should be made to this pin.
15	GND_2	V _{DD2} power supply ground return
16	V_{DD2}	V _{DD2} isolation power supply input*

V_{DD2} $\mathbf{V_{DD1}}$ GND_1 2 15 **GND**₂ TxD 314 IsoTxD RxD 4 13 **S** 12 CANH NC 5 NC 6 11 V_{DD2} 10 CANL NC 7 NC 8 9 IsoRxD

IL41050 Pin Connections (0.3" SOIC Package)

1	V_{DD1}	V _{DD1} power supply input
2	GND_1	V _{DD1} power supply ground return (pin 2 is internally connected to pin 8)
3	TxD	Transmit Data input
4	NC	No internal connection
5	RxD	Receive Data output
6	NC	No internal connection
7	NC	No internal connection
8	GND_1	V _{DD1} power supply ground return (pin 8 is internally connected to pin 2)
9	GND_2	V _{DD2} power supply ground return (pin 9 is internally connected to pin 15)
10	V_{REF}	Reference voltage output (nominally 50% of V _{DD2})
11	V_{DD2}	V _{DD2} CAN I/O bus circuitry power supply input*
12	CANL	Low level CANbus line
13	CANH	High level CANbus line
14	S	Mode select input. Leave open or set low for normal operation; set high for silent mode.
15	GND_2	V _{DD2} power supply ground return (pin 15 is internally connected to pin 9)
16	V_{DD2}	V _{DD2} isolation power supply input*



^{*}NOTE: Pin 11 is not internally connected to pin 16; both should be connected to the V_{DD2} power supply for normal operation.





Operating Specifications

Electrical Specif	ications (T _{min} to T _m	ax and V _{DD1} , V	$_{\rm DD2} = 4.75 \text{ V to}$	5.25 V unless	otherwise st	ated)
Parameter	Symbol	Min.	Тур.	Max.	Units	Test Conditions
Power Supply Current		•				1
11 v			1.75	2.0		$dr = 0 \text{ bps}; V_{DD1} = 5 \text{ V}$
Quiescent supply current (recessive)	IQ_{VDD1}	1	1.75	3.0	mA	dr = 0 bps;
	(VDD)	0.7	1.4	2.0		$V_{DD1} = 3.3 \text{ V}$
		1.2	2.0	2.2		$dr = 1 \text{ Mbps}, R_L = 60\Omega;$
B : 1 ((1 : 0)		1.2	2.0	3.2		$V_{DD1} = 5 \text{ V}$
Dynamic supply current (dominant)	I_{VDD1}	0.0	1.6	2.2	mA	$dr = 1 \text{ Mbps}, R_L = 60\Omega;$
		0.9	1.6	2.2		$V_{DD1} = 3.3 \text{ V}$
Quiescent supply current (recessive)	IQ_{VDD2}	3.5	7	12	A	0 bps
Dynamic supply current (dominant)	I_{VDD2}	26	52	70	mA	1 Mbps, $R_L = 60\Omega$
Transmitter Data input (TxD) ⁽¹⁾						-
High level input voltage ↑	$V_{ m IH}$	2.4		5.25	V	$V_{DD1} = 5 \text{ V}$; recessive
High level input voltage ↑	$V_{ m IH}$	2.0		3.6	V	$V_{\rm DD1} = 3.3 \text{ V}$; recessive
Low level input voltage ↓	V_{IL}	-0.3		0.8	V	Output dominant
TxD input rise and fall time ⁽²⁾	$t_{\rm r}$			1	μs	10% to 90%tr
High level input current	I _{IH}	-10		10	μA	$V_{TxD} = V_{DD1}$
Low level input current	$I_{\rm IL}$	10		10	μA	$V_{TxD} = 0 V$
Mode select input (S)		•			•	1.10
High level input voltage	V_{IH}	2.0		$V_{DD2} + 0.3$	V	Silent mode
Low level input voltage	V_{IL}	-0.3		0.8	V	High-speed mode
High level input current	$I_{ m IH}$	20	30	50	μA	$V_S = 2 V$
Low level input current	$I_{\rm IL}$	15	30	45	μA	$V_S = 0 \text{ V}$
Receiver Data output (RxD)	-iL					1 '3 * '
High level output current	I_{OH}	-2	-8.5	-20	mA	$V_{RxD} = 0.8 V_{DD1}$
Low level output current	I _{OL}	2	8.5	20	mA	$V_{RxD} = 0.45 \text{ V}$
Failsafe supply voltage ⁽⁴⁾	V _{DD2}	3.6		3.9	V	· KAD
Reference Voltage output (V _{REF})	· DD2				·	
Reference Voltage output	V_{REF}	$0.45~\mathrm{V_{DD2}}$	0.5 V _{DD2}	$0.55 V_{DD2}$	V	$-50 \mu A < I_{VREF} < +50 \mu A$
Bus lines (CANH and CANL)	· KEI	THE HODE	5 to 1 DD2	STOCK DD2		V PH - VKEI
Recessive voltage at CANH pin	V _{O(reces)} CANH	2.0	2.5	3.0	V	$V_{TxD} = V_{DD1}$, no load
Recessive voltage at CANL pin	V _{O(reces)} CANL	2.0	2.5	3.0	V	$V_{TxD} = V_{DD1}$, no load
			2.0			$-27V < V_{CANH} < +32V;$
Recessive current at CANH pin	I _{O(reces)} CANH	-2.5		+2.5	mA	$0V < V_{DD2} < 5.25V$
D	7 01377					$-27V < V_{CANL} < +32V;$
Recessive current at CANL pin	I _{O(reces)} CANL	-2.5		+2.5	mA	$0 \text{ V} < V_{DD2} < 5.25 \text{ V}$
Dominant voltage at CANH pin	V _{O(dom)} CANH	3.0	3.6	4.25	V	$V_{TxD} = 0 V$
Dominant voltage at CANL pin	V _{O(dom)} CANL	0.5	1.4	1.75	V	$V_{TxD} = 0 V$
	O(dolli)					$V_{TxD} = 0 \text{ V}$; dominant
Differential bus input voltage		1.5	2.25	3.0	V	$42.5 \Omega < R_L < 60 \Omega$
(V _{CANH} – V _{CANL})	$V_{i(dif)(bus)}$					$V_{TxD} = V_{DD1};$
(CANII CANL)		-120	0	+50	mV	recessive; no load
Short-circuit output current at CANH	I _{O(sc)} CANH	-45	-70	-95	mA	$V_{\text{CANH}} = 0 \text{ V}, V_{\text{TxD}} = 0$
Short-circuit output current at CANL	I _{O(sc)} CANL	45	70	120	mA	$V_{\text{CANL}} = 36 \text{ V}, V_{\text{TxD}} = 0$
•						$-5 \text{ V} < \text{V}_{\text{CANL}} < +10 \text{ V};$
Differential receiver threshold voltage	$V_{i(dif)(th)}$	0.5	0.7	0.9	V	$-5 \text{ V} < \text{V}_{\text{CANH}} < +10 \text{ V}$
Differential receiver input voltage				1.00		$-5 \text{ V} < \text{V}_{\text{CANL}} < +10 \text{ V};$
hysteresis	$V_{i(dif)(hys)}$	50	70	100	mV	$-5 \text{ V} < \text{V}_{\text{CANH}} < +10 \text{ V}$
Common Mode input resistance at	- D	1.7	2.7	2-		- CANII - V
CANH	$R_{i(CM)(CANH)}$	15	25	37	kΩ	
Common Mode input resistance at	D	1.5	2.5	27	1.0	
CANL	$R_{i(CM)(CANL)}$	15	25	37	kΩ	
Matching between Common Mode	n	-3		12	0./	X7 X7
	$R_{i(CM)(m)}$		0	+3	%	$V_{CANL} = V_{CANH}$





Electrical Specifications (T_{min} to T_{max} and V_{DD1} , V_{DD2} = 4.5 V to 5.5 V unless otherwise stated)								
Parameter	Symbol	Min.	Тур.	Max.	Units	Test Conditions		
Bus lines (cont)								
Differential input resistance	$R_{i(diff)}$	25	50	75	kΩ			
Input capacitance, CANH	$C_{i(CANH)}$		7.5	20	pF	$V_{TxD} = V_{DD1}$		
Input capacitance, CANL	$C_{i(CANL)}$		7.5	20	pF	$V_{TxD} = V_{DD1}$		
Differential input capacitance	$C_{i(dif)}$		3.75	10	pF	$V_{TxD} = V_{DD1}$		
Input leakage current at CANH	I _{LI(CANH)}	100	170	250	μΑ	$V_{CANH} = 5 \text{ V}, V_{DD2} = 0$		
Input leakage current at CANL	I _{LI(CANL)}	100	170	250	μΑ	$V_{CANL} = 5 \text{ V}, V_{DD2} = 0$		
Thermal Shutdown								
Shutdown junction temperature	$T_{i(SD)}$	155	165	180	°C			

Timing C	Timing Characteristics (60 Ω / 100 pF bus loading; 20 pF RxD load; see Fig. 1)									
Parameter	Symbol	Min.	Тур.	Max.	Units	Test Conditions				
TxD to bus active delay	t	44	93	160	ns	$V_S = 0 \text{ V}; V_{DD1} = 5 \text{ V}$				
TXD to bus active delay	$t_{d(TxD\text{-}BUSon)}$	36	96	128	113	$V_S = 0 \text{ V}; V_{DD1} = 3.3 \text{ V}$				
TxD to bus inactive delay	4	34	68	110	ns	$V_S = 0 \ V; \ V_{DD1} = 5 \ V$				
1xD to bus mactive delay	$t_{d(TxD-BUSoff)}$	37	71	113	115	$V_S = 0 \text{ V}; V_{DD1} = 3.3 \text{ V}$				
Bus active to RxD delay	+	29	63	125	nc	$V_S = 0 \ V; \ V_{DD1} = 5 \ V$				
Bus active to KXD delay	$t_{d(BUSon-RxD)}$	32	66	128	ns	$V_S = 0 \text{ V}; V_{DD1} = 3.3 \text{ V}$				
Bus inactive to RxD delay	+	69	108	170	ns	$V_S = 0 \text{ V}; V_{DD1} = 5 \text{ V}$				
Bus mactive to KXD delay	$t_{d(BUSoff-RxD)}$	72	111	173	115	$V_S = 0 \text{ V}; V_{DD1} = 3.3 \text{ V}$				
Loop delay low-to-high or high-to-low	т	74	180	250	nc	$V_S = 0 V$; "Typ." at				
Loop delay low-to-nigh of high-to-low	T_{LOOP}	/4	160	230	ns	25°C and nominal loads				
TxD dominant time for timeout	Т	250	457	765	ше	$V_{TxD} = 0 V$				
1xD dominant time for timeout	$T_{\text{dom(TxD)}}$	230	437	703	μs	$3.0 \text{ V} > \text{V}_{\text{DD1}} < 5.5 \text{ V}$				

Magnetic Field Immunity(3)

$V_{DD1} = 5 \text{ V}, V_{DD2} = 5 \text{ V}$								
Power frequency magnetic immunity	H_{PF}	2,500	3,000		A/m	50 Hz/60 Hz		
Pulse magnetic field immunity	H_{PM}	3,000	3,500		A/m	$t_p = 8 \mu s$		
Cross-axis immunity multiplier	K_X		1.8			Figure 1		
	1	$V_{\rm DD1} = 3.3 \text{ V}, \text{ V}$	$V_{\rm DD2} = 5 \text{ V}$					
Power frequency magnetic immunity	H_{PF}	1,000	1,500		A/m	50 Hz/60 Hz		
Pulse magnetic field immunity	H_{PM}	1,800	2,000		A/m	$t_p = 8 \mu s$		
Cross-axis immunity multiplier	K_X		1.5			Figure 1		

Notes:

- 1. The TxD input is edge sensitive. Voltage magnitude of the input signal is specified, but edge rate specifications must also be met.
- 2. The maximum time allowed for a logic transition at the TxD input is 1 μ s.
- 3. Uniform magnetic field applied across the pins of the device. Cross-axis multiplier effective when field is applied perpendicular to the pins.
- 4. If V_{DD2} falls below the specified failsafe supply voltage, RxD will go High.



Timing Test Circuit

Timing parameters are measured with $60 \Omega / 100 \text{ pF}$ bus line loading and 20 pF on RxD as shown in Figure 1 below:

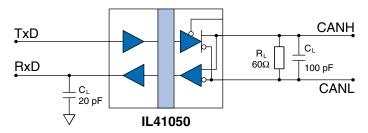


Figure 1. Timing characteristics test circuit.

Block Diagram

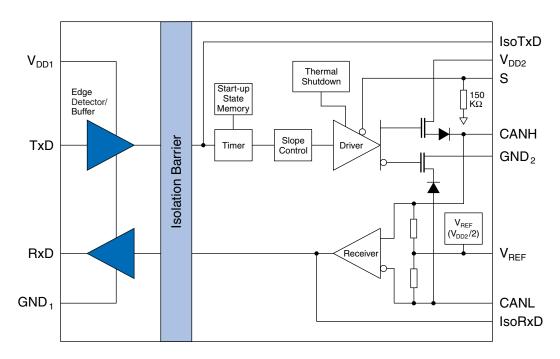


Figure 2. IL41050TA detailed functional diagram.



Application Information

As Figure 3 shows, the IL41050 can provide isolation and level shifting between a 5 volt CAN bus and a 3 volt microcontroller:

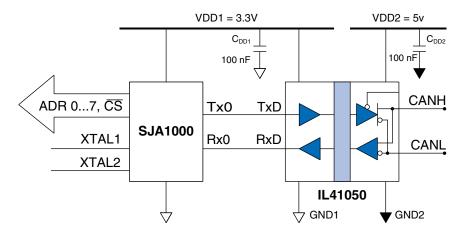


Figure 3. Isolated CAN node using the IL41050 and an SJA1000 MCU.

Bus-Side Power Supply Pins

Both V_{DD2} power supply inputs (pins 11 and 16) must be connected to the bus-side power supply. Pin 11 powers the bus side of the CAN I/O circuitry, while pin 16 powers the bus-side isolation circuitry. For testing purposes, they are not internally connected, but the part will not operate without both pins powered, and operation without both pins powered can cause damage.

Power Supply Decoupling

Both V_{DD1} and V_{DD2} must be bypassed with 100 nF ceramic capacitors. These supply the dynamic current required for the isolator switching and should be placed as close as possible to V_{DD} and their respective ground return pins.

Input Configurations

The TxD input should not be left open as the state will be indeterminate. If connected to an open-drain or open collector output, a pull-up resistor (typically 16 k Ω) should be connected from the input to V_{DD1} .

The Mode Select ("S") input has a nominal 150 k Ω internal pull-down resistor. It can be left open or set low for normal operation.

Dominant Mode Time-out and Failsafe Receiver Functions

CAN bus latch up is prevented by an integrated Dominant mode timeout function. If the TxD pin is forced permanently low by hardware or software application failure, the time-out returns the RxD output to the high state no more than 765 µs after TxD is asserted dominant. The timer is triggered by a negative edge on TxD. If the duration of the low is longer than the internal timer value, the transmitter is disabled, driving the bus to the recessive state. The timer is reset by a positive edge on pin TxD.

If power is lost on Vdd2, the IL41050 asserts the RxD output high when the supply voltage falls below 3.8 V. RxD will return to normal operation when Vdd2 rises above approximately 4.2 V.

Programmable Power-Up

A unique non-volatile programmable power-up feature prevents unstable nodes. A state that needs to be present at node power up can be programmed at the last power down. For example if a CAN node is required to "pulse" dominant at power up, TxD can be sent low by the controller immediately prior to power down. When power is resumed, the node will immediately go dominant allowing self-check code in the microcontroller to verify node operation. If desired, the node can also power up silently by presetting the TxD line high at power down. At the next power on, the IL41050 will remain silent, awaiting a dominant state from the bus.

The microcontroller can check that the CAN node powered down correctly before applying power at the next "power on" request. If the node powered down as intended, RxD will be set high and stored in the IL41050's non-volatile memory. The level stored in the RxD bit can be read before isolated node power is enabled, avoiding possible CAN bus disruption due to an unstable node.





Replacing Non-Isolated Transceivers

The IL41050 is designed to replace common non-isolated CAN transceivers such as the Philips/NXP TJA1050 with minimal circuit changes. Some notable differences:

- Some non-isolated CAN transceivers have internal TxD pull-up resistors, but the IL41050 TxD input should not be left open. If connected to an open-drain or open collector output, a pull-up resistor (typically 16 k Ω) should be connected from the input to V_{DDL} .
- Initialization behavior varies between CAN transceivers. To ensure the desired power-up state, the IL41050 should be initialized with a TxD pulse (low-to-high for recessive initialization), or shut down the transceiver in the desired power-up state (the "programmable power-up feature").
- Many non-isolated CAN transceivers have a V_{REF} output. Such a reference is available on the IL41050 wide-body version.

The VREF Output

V_{REF} is a reference voltage output used to drive bus threshold comparators in some legacy systems and is provided on the IL41050 wide-body version. The output is half of the bus supply $\pm 10\%$ (i.e., $0.45 \text{ V}_{DD2} < \text{V}_{REF} < 0.55 \text{ V}_{DD2}$), and can drive up to $50 \mu A$.

IsoRxD / IsoTxD Outputs

The IsoRxD and IsoTxD outputs are isolated versions of the RxD and TxD signals. These outputs are provided for troubleshooting on the narrow-body version, but normally no connections should be made to the pins.

The Isolation Advantage

Battery fire caused by over or under charging of individual lithium ion cells is a major concern in multi-cell high voltage electric and hybrid vehicle batteries. To combat this, each cell is monitored for current flow, cell voltage, and in some advanced batteries, magnetic susceptibility. The IL41050 allows seamless connection of the monitoring electronics of every cell to a common CAN bus by electrically isolating inputs from outputs, effectively isolating each cell from all other cells. Cell status is then monitored via the CAN controller in the Battery Management System (BMS).

Another major advantage of isolation is the tremendous increase in noise immunity it affords the CAN node, even if the power source is a battery. Inductive drives and inverters can produce transient swings in excess of 20 kV/µs. The traditional, non-isolated CAN node provides some protection due to differential signaling and symmetrical driver/receiver pairs, but the IL41050 typically provides more than twice the dV/dt protection of a traditional CAN node.





Electrostatic Discharge Sensitivity

This product has been tested for electrostatic sensitivity to the limits stated in the specifications. However, NVE recommends that all integrated circuits be handled with appropriate care to avoid damage. Damage caused by inappropriate handling or storage could range from performance degradation to complete failure.

Electromagnetic Compatibility

The IL41050 is fully compliant with generic EMC standards EN50081, EN50082-1 and the umbrella line-voltage standard for Information Technology Equipment (ITE) EN61000. The IsoLoop Isolator's Wheatstone bridge configuration and differential magnetic field signaling ensure excellent EMC performance against all relevant standards. NVE conducted compliance tests in the categories below:

EN50081-1

Residential, Commercial & Light Industrial Methods EN55022, EN55014

EN50082-2: Industrial Environment

Methods EN61000-4-2 (ESD), EN61000-4-3 (Electromagnetic Field Immunity), EN61000-4-4 (Electrical Transient Immunity), EN61000-4-6 (RFI Immunity), EN61000-4-8 (Power Frequency Magnetic Field Immunity), EN61000-4-9 (Pulsed Magnetic Field), EN61000-4-10 (Damped Oscillatory Magnetic Field)

ENV50204

Radiated Field from Digital Telephones (Immunity Test)

Immunity to external magnetic fields is higher if the field direction is "end-to-end" (rather than to "pin-to-pin") as shown in the diagram below:

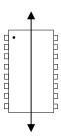
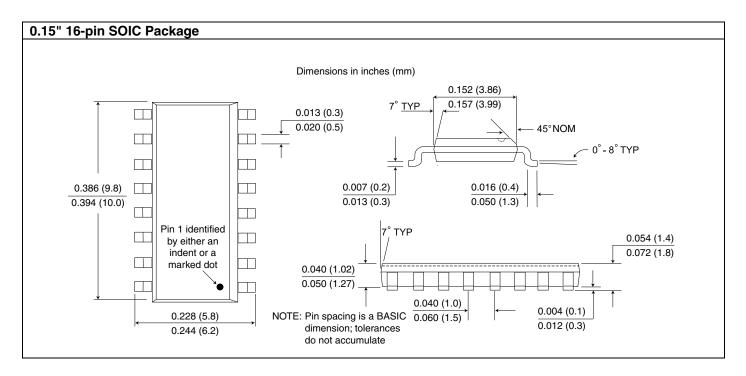
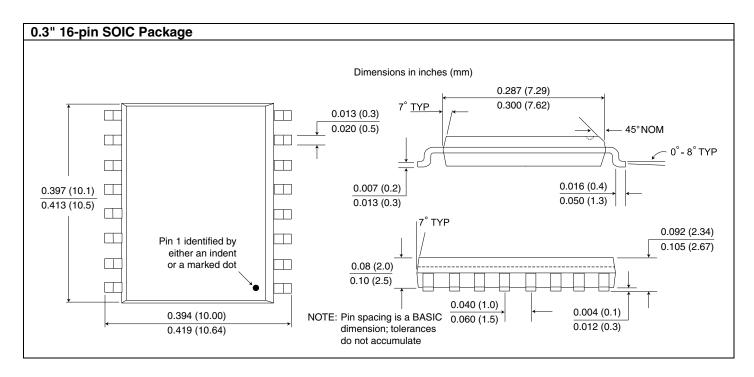


Figure 4. Orientation for high field immunity.



Package Drawings, Dimensions and Specifications

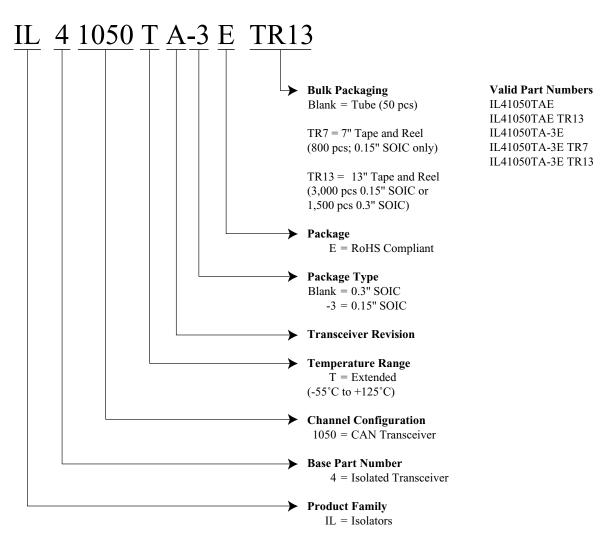








Ordering Information and Valid Part Numbers









Revision History

ISB-DS-001-IL41050TA-F January 2013

Change

- Added thermal characteristics (p. 2).
- · Cosmetic changes.

ISB-DS-001-IL41050TA-E December 2012

Changes

- UL 1577 recognition and IEC 61010-1 approval.
- Detailed isolation and barrier specifications.
- Style and cosmetic changes.

ISB-DS-001-IL41050TA-D October 2012

Changes

- Changed title to highlight speed.
- Added block diagram (detailed functional diagram).
- Rearranged and repaginated.

ISB-DS-001-IL41050TA-C July 2012

Changes

- Tightened and clarified typical loop delay specification.
- Clarified IsoRxD / IsoTxD outputs on narrow-body package.

ISB-DS-001-IL41050TA-B July 2012

Changes

- Specified timing characteristics test conditions and added test circuit (p. 5).
- More detailed application diagram (p. 6).
- Misc. cosmetic changes.

ISB-DS-001-IL41050TA-A May 2012

Changes

• Initial release.

ISB-DS-001-IL41050TA-Preview February 2012

Changes

• Released product preview.





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