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# IR3101 iMOTION ${ }^{\text {"m }}$ Series 1.6A, 500V 

## Description

IR3101 is a gate driver IC integrated half bridge FredFET designed for sub 250W (heat-sink-less) motor drive applications. The sleek and compact single-in-line package is optimized for electronic motor control in appliance applications such as fans and compressors for refrigerators. The IR3101 offers an extremely compact, high performance half-bridge inverter, in a single isolated package for a very simple design for twophase and three-phase motor drivers.
Proprietary HVIC and latch immune CMOS technologies, along with the HEXFET ${ }^{\circledR}$ power FredFET ${ }^{\circledR}$ technology (HEXFET ${ }^{\circledR}$ with ultra-fast recovery body diode characteristics), enable efficient and rugged single package construction. Propagation delays for the high and low side power FredFETs are matched thanks to the advance IC technology.

## Features

- Output power FredFets in half-bridge configuration
- High side gate drive designed for bootstrap operation
- Bootstrap diode integrated into package.
- Lower power level-shifting circuit
- Lower di/dt gate drive for better noise immunity
- Excellent latch immunity on all inputs and outputs
- ESD protection on all leads
- Isolation $1500 \mathrm{~V}_{\text {RMS }} \min$



## Absolute Maximum Ratings

Absolute maximum ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM. The thermal resistance and power dissipation are measured under board mounted and still air conditions.

| Parameters | Description | Max. Values | Units |
| :---: | :---: | :---: | :---: |
| $V_{D D}$ | High voltage supply | 500 | V |
| $V_{B}$ | High side floating supply | $\mathrm{V}_{0}+25$ | V |
| $\mathrm{P}_{\mathrm{D}}$ | Package power dissipation @ $\mathrm{T}_{\mathrm{C}} \leq 80^{\circ} \mathrm{C}$ (per die) | 5.8 | W |
| Rth ${ }_{\text {Jc }}$ | Thermal resistance, junction to case | 12 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Rth ${ }_{\text {JA }}$ | Thermal resistance, junction to ambient (note 1) | 85 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $V_{\text {ISO }}$ | Isolation Voltage (1 min) | 1500 | $V_{\text {RMS }}$ |
| $\mathrm{T}_{\mathrm{J}}$ | Junction temperature (Power Mosfet) | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{S}}$ | Storage temperature | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |
| TL | Lead temperature (soldering, 10 seconds) | 300 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{I}_{0}$ | Maximum current rating (note 2 ) | 1.6 | A |
| $\mathrm{I}_{0}$ | Continuous output current ( $\mathrm{V}_{1 N}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{CC}}=15 \mathrm{~V}$ ) | 1.3 | A |
|  |  | 2 | A |

Note 1: under normal operational conditions: both power devices working, no heatsink
Note 2: see figure $4, \mathrm{f}_{\mathrm{PWM}}=20 \mathrm{kHz}$

## Internal Electrical Schematic - I R3101



Figure 1: Internal connections

## Recommended Operating Conditions

For proper operation, the device should be used within the recommended conditions.

| Symbol | Definition | Min. | Max. | Units |
| :--- | :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{B}}$ | High side floating supply absolute voltage | $\mathrm{V}_{\mathrm{O}}+10$ | $\mathrm{~V}_{\mathrm{O}}+20 \mathrm{~V}$ | V |
| $\mathrm{~V}_{\mathrm{DD}}$ | High voltage supply | - | 450 | V |
| $\mathrm{~V}_{\mathrm{CC}}$ | Low side and logic fixed supply voltage | 10 | 20 | V |
| $\mathrm{~V}_{I N}$ | Logic input voltage | $\mathrm{V}_{S S}$ | $\mathrm{~V}_{\mathrm{CC}}$ | V |
| $\mathrm{V}_{\mathrm{SS}}$ | Logic ground | -5 | 5 | V |

Note 3: Care should be taken to avoid switching condition where the $\mathrm{V}_{\mathrm{O}}$ node flies inductively below COM by more than 5 V

## International

 Ior RectifierIR3101

## MOSFET Characteristics

$\mathrm{V}_{\mathrm{BIAS}}\left(\mathrm{V}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{B}}\right)=15 \mathrm{~V}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ unless otherwise specified. The $\mathrm{V}_{\mathrm{DD}}$ parameter is referenced to COM.

| Symbol | Definition | Min. | Typ | Max. | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {(BR) } \mathrm{DSS}}$ | Drain-to-Source breakdown voltage | 500 | - | - | V | $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}, \mathrm{I}_{\mathrm{D}}=250 \mu \mathrm{~A}$ |
| $\mathrm{I}_{\text {DSS }}$ | Drain-to-Source leakage current | - | - | 50 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{DS}}=500 \mathrm{~V}, \mathrm{~V}_{1 \mathrm{~N}}=0 \mathrm{~V}$ |
| $\mathrm{R}_{\mathrm{DS} \text { (on) }}$ | Static drain-to-source on resistance | - | 0.8 | 1.0 | $\Omega$ | $\mathrm{I}_{\mathrm{D}}=1.5 \mathrm{~A}$ |
| $\mathrm{V}_{\text {SD }}$ | Diode forward voltage | - | 0.82 | 0.9 | V | $\mathrm{I}_{\mathrm{D}}=1.5 \mathrm{~A}, \mathrm{~V}_{1 \mathrm{~N}}=0 \mathrm{~V}$ |
| $\mathrm{R}_{\mathrm{DS} \text { (on) }}$ | Static drain-to-source on resistance | - | 1.7 | 2.0 | $\Omega$ | $\mathrm{I}_{\mathrm{D}}=1.5 \mathrm{~A}, \mathrm{~T}_{\mathrm{J}}=125^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{\text {SD }}$ | Diode forward voltage | - | 0.70 | 0.79 | V | $\mathrm{I}_{\mathrm{D}}=1.5 \mathrm{~A}, \mathrm{~V}_{1 N}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{J}}=125^{\circ} \mathrm{C}$ |
| $\mathrm{E}_{\text {ON }}$ | Turn-On energy losses | - | 100 | 135 | $\mu \mathrm{J}$ | $\begin{aligned} & \mathrm{I}_{\mathrm{F}}=1.5 \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{OC}}=300 \mathrm{~V} \\ & \mathrm{di} / \mathrm{dt}=200 \mathrm{~A} / \mu \mathrm{s} \end{aligned}$ |
| E | Turn-Off energy losses | - | 5 | 10 | $\mu \mathrm{J}$ |  |
| $\mathrm{E}_{\text {REC }}$ | Body-Diode reverse recovery <br> Llosses | - | 10 | 20 | $\mu \mathrm{J}$ |  |
| $\mathrm{t}_{\text {RR }}$ | Reverse recovery time | - | 105 | 180 | ns |  |
| $\mathrm{E}_{\text {ON }}$ | Turn-On energy losses | - | 150 | 205 | $\mu \mathrm{J}$ | $\begin{aligned} & \mathrm{T}_{\mathrm{J}}=125^{\circ} \mathrm{C} \\ & \mathrm{I}_{\mathrm{F}}=1.5 \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{CC}}=300 \mathrm{~V} \\ & \mathrm{di} / \mathrm{dt}=200 \mathrm{~A} / \mu \mathrm{s} \end{aligned}$ |
| EOfF | Turn-Off energy losses | - | 10 | 17 | $\mu \mathrm{J}$ |  |
| $\mathrm{E}_{\text {REC }}$ | Body-Diode reverse recovery Llosses | - | 15 | 35 | $\mu \mathrm{J}$ |  |
| $\mathrm{t}_{\text {RR }}$ | Reverse recovery time | - | 130 | 230 | ns |  |
| $\mathrm{C}_{\text {oss }}$ | Output capacitance | - | - | 100 | pF | $\mathrm{V}_{1 \mathrm{~N}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}=30 \mathrm{~V}, \mathrm{f}=1 \mathrm{MHz}$ |

## Driver IC Characteristic

| Symbol | Definition | Min. | Typ. | Max. | Unit s | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {CCUV }+}$ | $V_{C C}$ supply undervoltage positive going threshold | 8.0 | 8.9 | 9.8 | V |  |
| $\mathrm{V}_{\text {ccuv }}$. | $V_{C C}$ supply undervoltage negative going threshold | 7.4 | 8.2 | 9.0 | V |  |
| $\mathrm{V}_{\text {CCUVH }}$ | $V_{C C}$ supply undervoltage lockout hysteresis | 0.3 | 0.7 | - | V |  |
| $\mathrm{V}_{\mathrm{BSUV}+}$ | $V_{B S}$ supply undervoltage positive going threshold | 8.0 | 8.9 | 9.8 | V |  |
| $V_{\text {BSUV- }}$ | $V_{B S}$ supply undervoltage negative going threshold | 7.4 | 8.2 | 9.0 | V |  |
| $\mathrm{V}_{\text {BSUVH }}$ | $V_{B S}$ supply undervoltage lockout hysteresis | 0.3 | 0.7 | - | V |  |
| $\mathrm{V}_{\text {IH }}$ | Logic "1" input voltage for HIN \& LIN | 2.9 | - | - | V | $V_{C C}=10 \mathrm{~V}$ to 20 V |
| VIL | Logic "0" input voltage for HIN \& LIN | - | - | 0.8 | V | $V_{C C}=10 \mathrm{~V}$ to 20 V |
| $\mathrm{I}_{\text {IN+}}$ | Logic "1" input bias current | - | 5 | 20 | $\mu \mathrm{A}$ | $H_{I N}, L_{\text {IN }}=5 \mathrm{~V}$ |
| $\mathrm{I}_{\text {IN- }}$ | Logic "0" input bias current | - | 1 | 2 | $\mu \mathrm{A}$ | $\mathrm{H}_{\text {IN }}, \mathrm{L}_{\text {IN }}=0 \mathrm{~V}$ |
| MT | Delay Matching HS \& LS turn on/ turn off | - | 0 | 30 | ns |  |



| $\mathbf{H}_{\mathbf{I N}}$ | $\mathbf{L}_{\mathbf{N}}$ | $\mathbf{V}_{\mathbf{O}}$ |
| :---: | :---: | :---: |
| 0 | 1 | 0 |
| 1 | 0 | $\mathrm{~V}_{\mathrm{DD}}$ |
| 1 | 1 | Shoot-Through <br> condition |
| X | X | X |

Figure 2: Driver input/output relation

International

## Module Pin-Out Description

| Pin | Symbol | Lead Definitions |
| :---: | :---: | :--- |
| 1 | $\mathrm{~V}_{C C}$ | Logic and internal gate drive supply |
| 2 | $\mathrm{H}_{I N}$ | Logic input for high side gate output |
| 3 | $\mathrm{~L}_{\mathrm{IN}}$ | Logic input for low side gate output |
| 4 |  | Not Connected |
| 5 | $\mathrm{~V}_{S S}$ | Logic Ground |
| 6 | COM | Low side MOSFET gate return |
| 7 |  | Not Connected |
| 8 | $\mathrm{~V}_{\mathrm{B}}$ | High side gate drive floating supply |
| 9 | $\mathrm{~V}_{\mathrm{O}}$ | Half bridge output |
| 10 |  | Not Connected |
| 11 | $\mathrm{~V}_{\mathrm{DD}}$ | High voltage supply |

## Typical Application Connection I R3101



1. Electrolytic bus capacitors should be mounted as close to the module bus terminals as possible to reduce ringing and EMI problems. Additional high frequency ceramic capacitor mounted close to the module pins will further improve performance.
2. In order to provide good decoupling between $\mathrm{V}_{\mathrm{cc}}-\mathrm{V}_{\mathrm{SS}}$ and $\mathrm{V}_{\mathrm{B}}-\mathrm{V}_{\mathrm{O}}$ terminals, a capacitor connected between these terminals is recommended and should be located very close to the module pins. Additional high frequency capacitors, typically $0.1 \mu \mathrm{~F}$, are strongly recommended.
3. Low inductance shunt resistor should be used for phase leg current sensing. Similarly, the length of the traces from the pin to the corresponding shunt resistor should be kept as small as possible.
4. Value of the bootstrap capacitors depends upon the switching frequency. Their selection should be made based on IR design tip DN 98-2a or Figure 8.
5. Application conditions should guarantee minimum dead-time of 200 ns


Figure 3: Maximum phase current as function of switching frequency
Trapezoidal modulation, $120^{\circ}$ switching, $\mathrm{V}_{\mathrm{BUS}}=300 \mathrm{~V}$, Duty Cycle $=0.8$, without heatsink:
$\mathrm{T}_{\mathrm{a}}=55^{\circ} \mathrm{C}, \mathrm{T}_{\mathrm{J}}=150^{\circ} \mathrm{C}$


Figure 4: Maximum phase current as function of switching frequency
Trapezoidal modulation, $120^{\circ}$ switching, $V_{B u s}=300 \mathrm{~V}$, Duty Cycle $=0.8$, with heatsink:
$\mathrm{T}_{\mathrm{C}}=100^{\circ} \mathrm{C}, \mathrm{T}_{\mathrm{J}}=125^{\circ} \mathrm{C}$


Figure 5. FredFET Turn-on. Typical turn-on waveform $@ T_{j}=125^{\circ} \mathrm{C}, \mathrm{V}_{\text {BUS }}=300 \mathrm{~V}$


Figure 6. FredFET Turn-off. Typical turn-on waveform $@ T_{j}=125^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{BUS}}=300 \mathrm{~V}$


Figure 7: Normalized On-Resistance vs temperature
$V_{C C}=10 \mathrm{~V}, I_{D}=1.5 \mathrm{~A}$


Figure 8: Recommended minimum bootstrap capacitor value vs switching frequency

## Package Outline



Note 1: Marking for pin 1 identification
Note 2: Product Part Number
Note 3: Lot and Date code marking
Dimensioning and Tolerancing per ANSY Y14.5M-1992
Controlling Dimensions: INCH
Dimensions are shown in millimeters [inches]

Data and Specifications are subject to change without notice


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