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With the principle of "Quality Parts, Customers Priority, Honest Operation, and Considerate Service", our business mainly focus on the distribution of electronic components. Line cards we deal with include Microchip, ALPS, ROHM, Xilinx, Pulse, ON, Everlight and Freescale. Main products comprise IC, Modules, Potentiometer, IC Socket, Relay, Connector. Our parts cover such applications as commercial, industrial, and automotives areas.

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PD - 95661 IRL3714ZPbF IRL3714ZSPbF IRL3714ZLPbF HEXFET® Power MOSFET

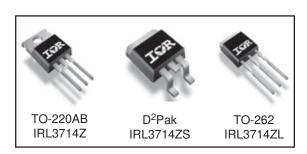
V_{DSS}	R _{DS(on)} max	Qg
20V	16mΩ	4.8nC

Applications

- High Frequency Synchronous Buck Converters for Computer Processor Power
- Lead-Free

Benefits

- Low R_{DS(on)} at 4.5V V_{GS}
- Ultra-Low Gate Impedance
- Fully Characterized Avalanche Voltage and Current



Absolute Maximum Ratings

	Parameter	Max.	Units
V _{DS}	Drain-to-Source Voltage	20	V
V _{GS}	Gate-to-Source Voltage	± 20	
I _D @ T _C = 25°C	Continuous Drain Current, V _{GS} @ 10V	36 ^⑤	Α
I _D @ T _C = 100°C	Continuous Drain Current, V _{GS} @ 10V	25 [©]	
I _{DM}	Pulsed Drain Current ①	140	
P _D @T _C = 25°C	Maximum Power Dissipation	35	W
P _D @T _C = 100°C	Maximum Power Dissipation	18	
	Linear Derating Factor	0.23	W/°C
T _J	Operating Junction and	-55 to + 175	°C
T _{STG}	Storage Temperature Range		
	Soldering Temperature, for 10 seconds	300 (1.6mm from case)	

Thermal Resistance

	Parameter	Тур.	Max.	Units
$R_{\theta JC}$	Junction-to-Case		4.3	°C/W
$R_{\theta CS}$	Case-to-Sink, Flat Greased Surface ③			
$R_{\theta JA}$	Junction-to-Ambient ③ —		62	
$R_{\theta JA}$	Junction-to-Ambient (PCB Mount) ®		40	

Notes ① through @ are on page 12

IRL3714Z/S/LPbF

International **TOR** Rectifier

Static @ T_J = 25°C (unless otherwise specified)

	Parameter	Min.	Тур.	Max.	Units	Conditions
BV _{DSS}	Drain-to-Source Breakdown Voltage	20			٧	$V_{GS} = 0V, I_D = 250\mu A$
$\Delta BV_{DSS}/\Delta T_{J}$	Breakdown Voltage Temp. Coefficient		0.015		mV/°C	Reference to 25°C, I _D = 1mA
R _{DS(on)}	Static Drain-to-Source On-Resistance		13	16	mΩ	V _{GS} = 10V, I _D = 15A ③
			21	26		V _{GS} = 4.5V, I _D = 12A ③
$V_{GS(th)}$	Gate Threshold Voltage	1.65	2.1	2.55	٧	$V_{DS} = V_{GS}$, $I_D = 250\mu A$
$\Delta V_{GS(th)}/\Delta T_J$	Gate Threshold Voltage Coefficient		-5.2		mV/°C	
I _{DSS}	Drain-to-Source Leakage Current			1.0	μΑ	$V_{DS} = 16V, V_{GS} = 0V$
				150		$V_{DS} = 16V, V_{GS} = 0V, T_{J} = 125^{\circ}C$
I _{GSS}	Gate-to-Source Forward Leakage			100	nA	V _{GS} = 20V
	Gate-to-Source Reverse Leakage			-100		V _{GS} = -20V
gfs	Forward Transconductance	21			S	$V_{DS} = 10V, I_{D} = 14A$
Q_g	Total Gate Charge		4.8	7.2		
Q _{gs1}	Pre-Vth Gate-to-Source Charge		1.7	_	1	$V_{DS} = 10V$
Q _{gs2}	Post-Vth Gate-to-Source Charge		0.80		nC	$V_{GS} = 4.5V$
Q_{gd}	Gate-to-Drain Charge		1.7			I _D = 14A
Q _{godr}	Gate Charge Overdrive		0.60			See Fig. 16
Q _{sw}	Switch Charge (Q _{gs2} + Q _{gd})		2.5		1	
Q _{oss}	Output Charge		2.7		nC	$V_{DS} = 10V$, $V_{GS} = 0V$
t _{d(on)}	Turn-On Delay Time		6.0			V _{DD} = 10V, V _{GS} = 4.5V ③
t _r	Rise Time		13			I _D = 14A
t _{d(off)}	Turn-Off Delay Time		10		ns	Clamped Inductive Load
t _f	Fall Time		5.0		1	
C _{iss}	Input Capacitance		550			$V_{GS} = 0V$
Coss	Output Capacitance		180		pF	$V_{DS} = 10V$
C _{rss}	Reverse Transfer Capacitance		99		1	f = 1.0 MHz

Avalanche Characteristics

	Parameter	Тур.	Max.	Units
E _{AS}	Single Pulse Avalanche Energy ^②		23	mJ
I _{AR}	Avalanche Current ①		14	Α
E _{AR}	Repetitive Avalanche Energy ①		3.5	mJ

Diode Characteristics

	Parameter	Min.	Тур.	Max.	Units	Conditions
Is	Continuous Source Current			36®		MOSFET symbol
	(Body Diode)				Α	showing the
I _{SM}	Pulsed Source Current			140		integral reverse
	(Body Diode) ①					p-n junction diode.
V_{SD}	Diode Forward Voltage			1.0	V	$T_J = 25$ °C, $I_S = 14A$, $V_{GS} = 0V$ ③
t _{rr}	Reverse Recovery Time	_	8.3	12	ns	$T_J = 25^{\circ}C$, $I_F = 14A$, $V_{DD} = 10V$
Q _{rr}	Reverse Recovery Charge		1.5	2.3	nC	di/dt = 100A/µs ③

IRL3714Z/S/LPbF

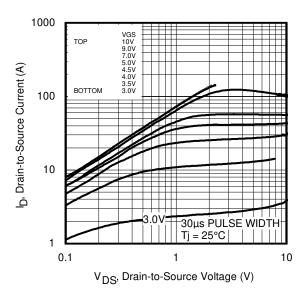


Fig 1. Typical Output Characteristics

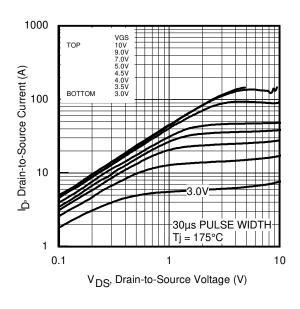


Fig 2. Typical Output Characteristics

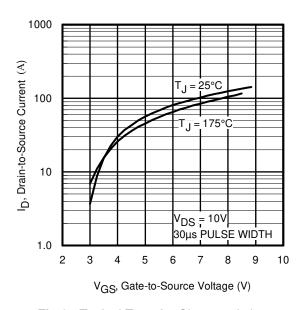


Fig 3. Typical Transfer Characteristics

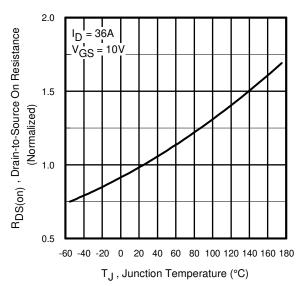


Fig 4. Normalized On-Resistance vs. Temperature

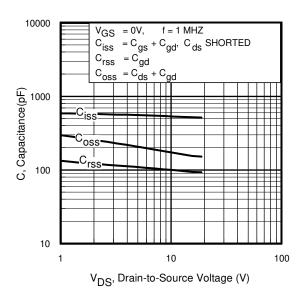


Fig 5. Typical Capacitance vs. Drain-to-Source Voltage

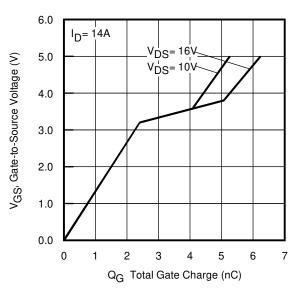


Fig 6. Typical Gate Charge vs. Gate-to-Source Voltage

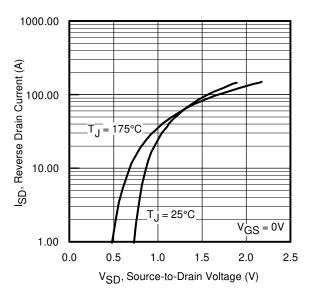


Fig 7. Typical Source-Drain Diode Forward Voltage

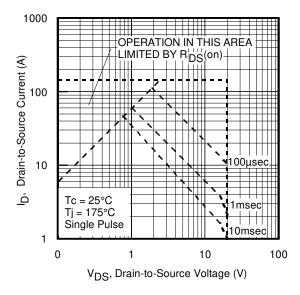
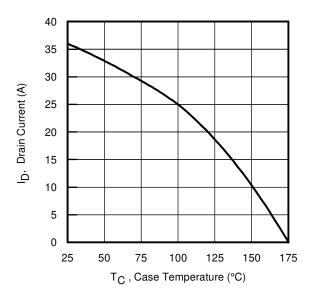


Fig 8. Maximum Safe Operating Area

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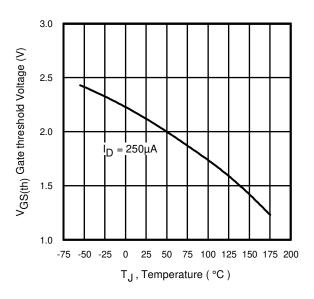


Fig 9. Maximum Drain Current vs.
Case Temperature

Fig 10. Threshold Voltage vs. Temperature

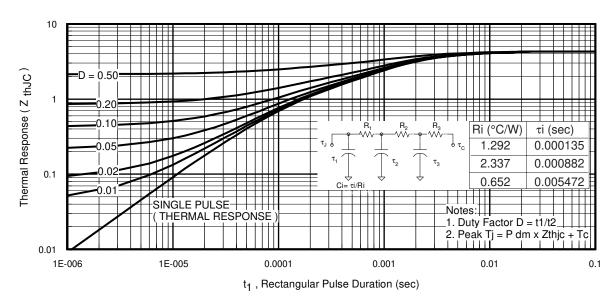


Fig 11. Maximum Effective Transient Thermal Impedance, Junction-to-Case

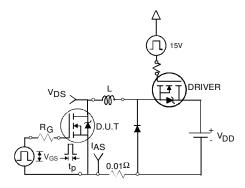


Fig 12a. Unclamped Inductive Test Circuit

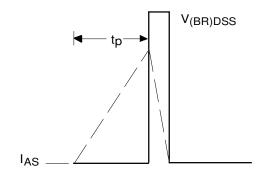


Fig 12b. Unclamped Inductive Waveforms

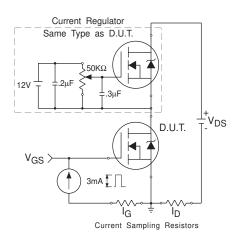


Fig 13. Gate Charge Test Circuit

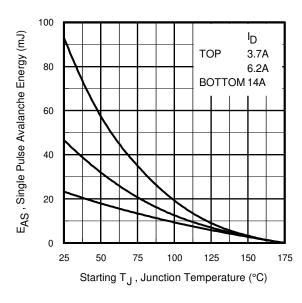


Fig 12c. Maximum Avalanche Energy vs. Drain Current

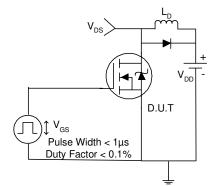


Fig 14a. Switching Time Test Circuit

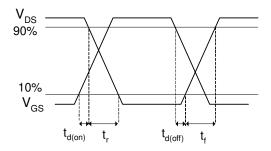


Fig 14b. Switching Time Waveforms www.irf.com

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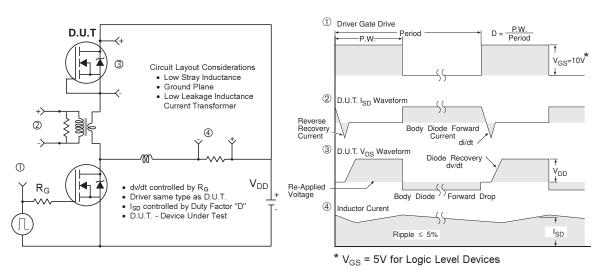


Fig 15. Peak Diode Recovery dv/dt Test Circuit for N-Channel HEXFET® Power MOSFETs

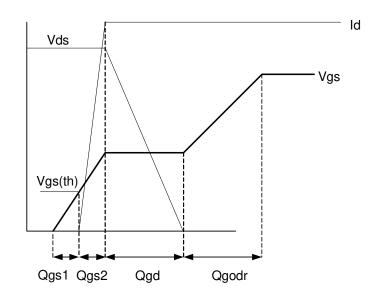


Fig 16. Gate Charge Waveform

Power MOSFET Selection for Non-Isolated DC/DC Converters

Control FET

Special attention has been given to the power losses in the switching elements of the circuit - Q1 and Q2. Power losses in the high side switch Q1, also called the Control FET, are impacted by the $R_{\rm ds(on)}$ of the MOSFET, but these conduction losses are only about one half of the total losses.

Power losses in the control switch Q1 are given by;

$$P_{loss} = P_{conduction} + P_{switching} + P_{drive} + P_{output}$$

This can be expanded and approximated by;

$$\begin{split} P_{loss} &= \left(I_{rms}^{2} \times R_{ds(on)}\right) \\ &+ \left(I \times \frac{Q_{gd}}{i_{g}} \times V_{in} \times f\right) + \left(I \times \frac{Q_{gs2}}{i_{g}} \times V_{in} \times f\right) \\ &+ \left(Q_{g} \times V_{g} \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) \end{split}$$

This simplified loss equation includes the terms ${\rm Q_{gs2}}$ and ${\rm Q_{oss}}$ which are new to Power MOSFET data sheets.

 Q_{gs2} is a sub element of traditional gate-source charge that is included in all MOSFET data sheets. The importance of splitting this gate-source charge into two sub elements, Q_{gs1} and Q_{gs2} , can be seen from Fig 16.

 Q_{gs2} indicates the charge that must be supplied by the gate driver between the time that the threshold voltage has been reached and the time the drain current rises to I_{dmax} at which time the drain voltage begins to change. Minimizing Q_{gs2} is a critical factor in reducing switching losses in Q1.

 $\rm Q_{oss}$ is the charge that must be supplied to the output capacitance of the MOSFET during every switching cycle. Figure A shows how $\rm Q_{oss}$ is formed by the parallel combination of the voltage dependant (nonlinear) capacitance's $\rm C_{ds}$ and $\rm C_{dg}$ when multiplied by the power supply input buss voltage.

Synchronous FET

The power loss equation for Q2 is approximated by;

$$\begin{split} P_{loss} &= P_{conduction} + P_{drive} + P_{output}^* \\ P_{loss} &= \left(I_{rms}^2 \times R_{ds(on)}\right) \\ &+ \left(Q_g \times V_g \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) + \left(Q_{rr} \times V_{in} \times f\right) \end{split}$$

*dissipated primarily in Q1.

For the synchronous MOSFET Q2, $R_{ds(on)}$ is an important characteristic; however, once again the importance of gate charge must not be overlooked since it impacts three critical areas. Under light load the MOSFET must still be turned on and off by the control IC so the gate drive losses become much more significant. Secondly, the output charge Q_{oss} and reverse recovery charge Q_{rr} both generate losses that are transfered to Q1 and increase the dissipation in that device. Thirdly, gate charge will impact the MOSFETs' susceptibility to Cdv/dt turn on.

The drain of Q2 is connected to the switching node of the converter and therefore sees transitions between ground and $V_{\rm in}.$ As Q1 turns on and off there is a rate of change of drain voltage dV/dt which is capacitively coupled to the gate of Q2 and can induce a voltage spike on the gate that is sufficient to turn the MOSFET on, resulting in shoot-through current . The ratio of $Q_{\rm gd}/Q_{\rm gs1}$ must be minimized to reduce the potential for Cdv/dt turn on.

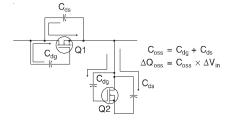
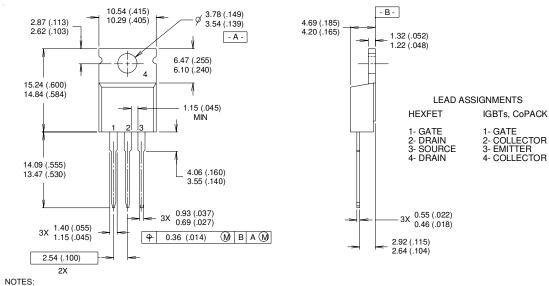


Figure A: Qoss Characteristic

IRL3714Z/S/LPbF

TO-220AB Package Outline

Dimensions are shown in millimeters (inches)



- 1 DIMENSIONING & TOLERANCING PER ANSI Y14.5M, 1982.
- 2 CONTROLLING DIMENSION: INCH

- $\,\,$ OUTLINE CONFORMS TO JEDEC OUTLINE TO-220AB.
- 4 HEATSINK & LEAD MEASUREMENTS DO NOT INCLUDE BURRS.

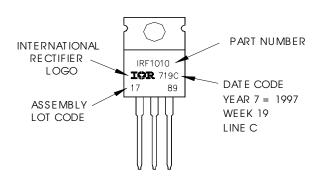
TO-220AB Part Marking Information

EXAMPLE: THIS IS AN IRF1010

LOT CODE 1789

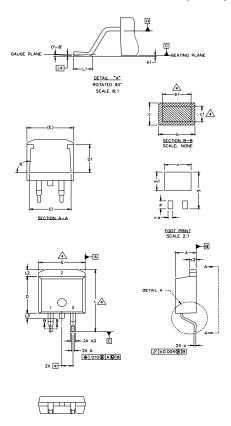
ASSEMBLED ON WW 19, 1997 IN THE ASSEMBLY LINE "C"

Note: "P" in assembly line position indicates "Lead-Free"



IRL3714Z/S/LPbF D²Pak Package Outline

Dimensions are shown in millimeters (inches)



S		DIMEN	SIONS		N.
0 B 0 L		ETERS			N O T E S
0			_	INCHES	
Ľ	MIN.	MAX.	MIN.	MAX.	š
Α	4.06	4.83	.160	.190	
A1		0.127		.005	
ь	0.51	0.99	.020	.039	
b1	0.51	0.89	.020	.035	4
b2	1,14	1,40	.045	.055	
С	0.43	0.63	.017	.025	
c1	0.38	0.74	.015	.029	4
c2	1,14	1,40	.045	.055	
D	8.51	9.65	.335	.380	3
D1	5.33		.210		
E	9.65	10.67	.380	.420	3
E1	6.22		.245		
е	2.54	BSC	.100	BSC	
L	14.61	15,88	.575	.625	
L1	1.78	2.79	.070	.110	
L2		1.65		.065	
L3	1.27	1.78	.050	.070	
L4	0,25	BSC	.010 BSC		
m	17,78		.700		
m1	8.89		.350		
n	11.43		.450		
0	2.08		.082		
р	3.81		.150		
Θ	90*	93'	90*	93*	

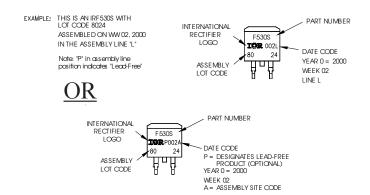
LEAD ASSIGNMENTS

HEXFET	IGBTs, CoPACK	DIODES
1 GATE 2 DRAIN 3 SOURCE	1 GATE 2 COLLECTOR 3 EMITTER	1,- ANODE * 2 CATHODE 3 ANODE

· PART DEPENDENT.

- NOTES:
 1. DIMENSIONING AND TOLERANCING PER ASME Y14.6M-1994
- 2. DIMENSIONS ARE SHOWN IN MILLIMETERS [INCHES].
- 3. DIMENSION D & E DO NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.127 [.005"] PER SIDE, THESE DIMENSIONS ARE WEASURED AT THE OUTWOST EXTREMES OF THE PLASTIC BODY.
- DIMENSION 61 AND c1 APPLY TO BASE METAL ONLY. 5. CONTROLLING DIMENSION: INCH,

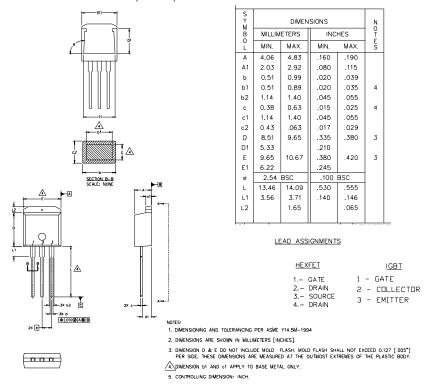
D²Pak Part Marking Information



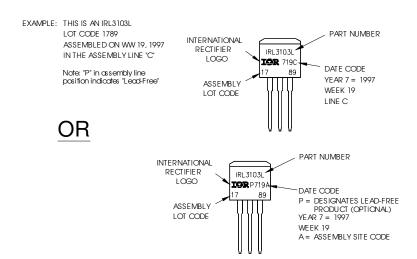
IRL3714Z/S/LPbF

TO-262 Package Outline

Dimensions are shown in millimeters (inches)



TO-262 Part Marking Information

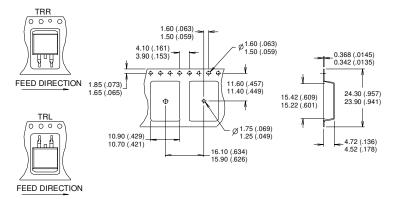


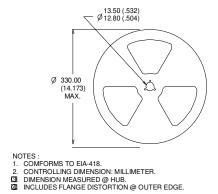
IRL3714Z/S/LPbF

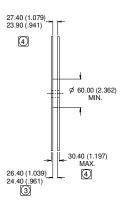
International TOR Rectifier

D²Pak Tape & Reel Information

Dimensions are shown in millimeters (inches)







Notes:

- ① Repetitive rating; pulse width limited by max. junction temperature.
- $\label{eq:tau_starting} \begin{tabular}{l} \begi$
- 3 Pulse width $\leq 400 \mu s$; duty cycle $\leq 2\%$.
- \oplus C_{oss} eff. is a fixed capacitance that gives the same charging time as C_{oss} while V_{DS} is rising from 0 to 80% V_{DSS}.
- ⑤ This is only applied to TO-220AB pakcage.
- ® This is applied to D²Pak, when mounted on 1" square PCB (FR-4 or G-10 Material). For recommended footprint and soldering techniques refer to application note #AN-994.

TO-220AB package is not recommended for Surface Mount Application.

Data and specifications subject to change without notice.

This product has been designed and qualified for the Industrial market.

Qualification Standards can be found on IR's Web site.



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TAC Fax: (310) 252-7903

Visit us at www.irf.com for sales contact information. 07/04

Note: For the most current drawings please refer to the IR website at: http://www.irf.com/package/