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STEREO HEADPHONE DRIVER



January 2016

GENERAL DESCRIPTION

The IS31AP4912 is stereo headphone drivers designed to allow the removal of the output DC-blocking capacitors for reduced component count and cost. The IS31AP4912 is ideal for small portable electronics where size and cost are critical design parameters.

The IS31AP4912 integrates click-and-pop suppression circuitry and thermal protect circuit. The gain of the amplifier is adjusted via external resistors.

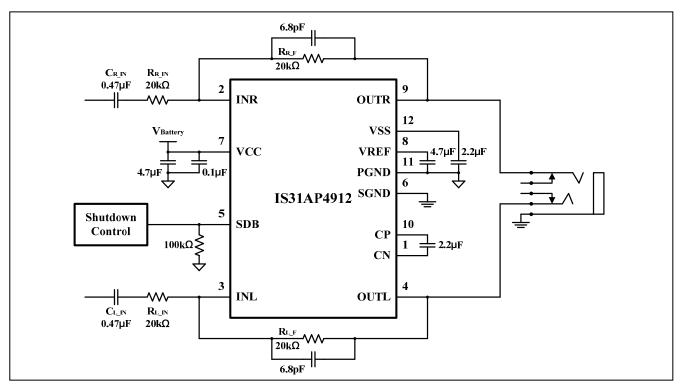
IS31AP4912 is available in UTQFN-12 (2mm × 2mm) packages. It operates from 2.7V to 5.5V over the temperature range of -40°C to +85°C.

FEATURES

- No output DC-blocking capacitors
- Supply voltage from 2.7V ~ 5.5V
- Low output noise (7µV)
- High SNR (103dB)
- -95dB PSRR
- Thermal protect circuit
- Integrated click-and-pop suppression circuitry
- UTQFN-12 (2mm × 2mm) package

APPLICATIONS

- Cellular handsets and PDAs
- Notebook PC
- MP3
- Portable gaming



TYPICAL APPLICATION CIRCUIT

Figure 1 Typical Application Circuit

Note: The SGND and PGND pins of the IS31AP4912 must be routed separately back to the decoupling capacitor in order to provide proper device operation. If the SGND and PGND pins are connected directly to each other, the part functions without risk of failure, but the noise and THD performance do not meet the specifications.



PIN CONFIGURATION

| Package | Pin Configuration (Top View) | |
|----------|--|---------------------------------|
| UTQFN-12 | $CN = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$ $INR = \begin{bmatrix} 2 \\ 2 \\ 1 \end{bmatrix}$ $INL = \begin{bmatrix} 3 \\ 1 \\ 1 \end{bmatrix}$ $INL = \begin{bmatrix} 3 \\ 1 \\ 1 \end{bmatrix}$ | OUTR 29 VREF 27 VCC |

PIN DESCRIPTION

| No. | Pin | Description |
|-----|------|---|
| 1 | CN | Charge pump flying capacitor negative terminal. |
| 2 | INR | Right channel audio input. |
| 3 | INL | Left channel audio input. |
| 4 | OUTL | Left channel audio output. |
| 5 | SDB | Shutdown control terminal, active low. |
| 6 | SGND | Signal Ground. |
| 7 | VCC | Supply voltage. |
| 8 | VREF | Internal produced supply voltage for charge pump and audio power amplifier. |
| 9 | OUTR | Right channel audio output. |
| 10 | CP | Charge pump flying capacitor positive terminal. |
| 11 | PGND | Power ground. |
| 12 | VSS | Output from charge pump. |



ORDERING INFORMATION Industrial Range: -40°C to +85°C

| Order Part No. | Package | QTY/Reel | |
|---------------------|---------------------|----------|--|
| IS31AP4912-UTLS2-TR | UTQFN-12, Lead-free | 3000 | |

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b.) the user assume all such risks; and

c.) potential liability of Integrated Silicon Solution, Inc is adequately protected under the circumstances



ABSOLUTE MAXIMUM RATINGS

| -0.3V ~ +6.0V |
|-------------------------------|
| -0.3V ~ V _{CC} +0.3V |
| 150°C |
| -65°C ~ +150°C |
| -40°C ~ +85°C |
| 63.1°C/W |
| ±8kV |
| ±1kV |
| |

Note:

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other condition beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

| V_{CC} = 2.7V ~ 5.5V, T_A = 25°C, unless otherwise noted. T | Typical value is $T_A = 25^{\circ}C$, $V_{CC} = 3.6V$. |
|---|--|
|---|--|

| Symbol | Parameter | Condition | Min. | Тур. | Max. | Unit |
|------------------|--------------------------|-----------------------|------|------|------|------|
| V _{CC} | Supply voltage | | 2.7 | | 5.5 | V |
| I _{CC} | Quiescent current | No load | | 3 | 5.5 | mA |
| I _{SD} | Shutdown current | V _{SDB} = 0V | | | 1 | μA |
| f _{OSC} | Operating frequency | | | 250 | | kHz |
| V _{os} | Output offset voltage | V _{IN} = 0V | | 1 | | mV |
| V _{IH} | High-level input voltage | | 1.4 | | | V |
| VIL | Low-level input voltage | | | | 0.4 | V |

ELECTRICAL CHARACTERISTICS (NOTE 1)

 $T_A = 25^{\circ}C$, $V_{CC} = 3.6V$, unless otherwise noted.

| Symbol | Parameter | Condition | Min. | Тур. | Max. | Unit |
|-----------------|--------------------------------------|---|------|-------|------|------|
| Po | Output power | THD+N = 1%, R_L = 32Ω, f = 1kHz | | 30 | | mW |
| THD+N | Total harmonic distortion plus noise | P _O = 20mW, R _L = 32Ω, f = 1kHz | | 0.024 | | % |
| t _{wu} | Wake-up time from shutdown | | | 39 | | ms |
| | SRR Power supply rejection ratio | V _{P-P} = 200mV, R _L = 32Ω, f = 217Hz | | -95 | | dB |
| PSRR | | V _{P-P} = 200mV, R _L = 32Ω, f = 1kHz | | -93 | | dB |
| V _{NO} | Output voltage noise | | | 7 | | μV |
| SNR | Signal-to-noise ratio | P _O = 30mW, THD+N = 0.1% | | 103 | | dB |

Note 1: Guaranteed by design.



TYPICAL PERFORMANCE CHARACTERISTIC

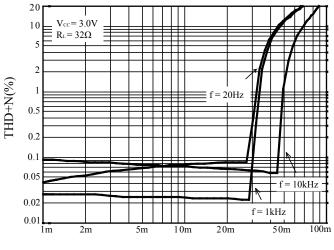
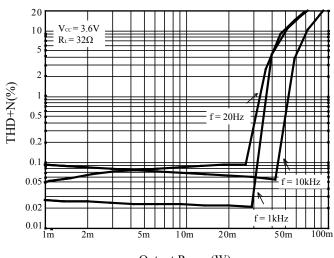
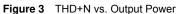


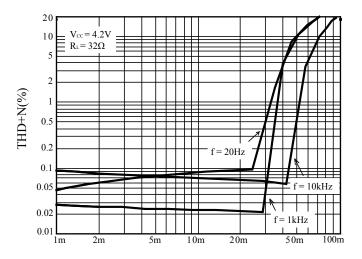


Figure 2 THD+N vs. Output Power

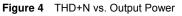


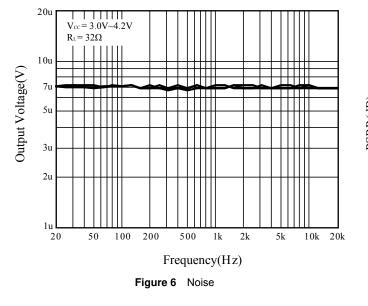


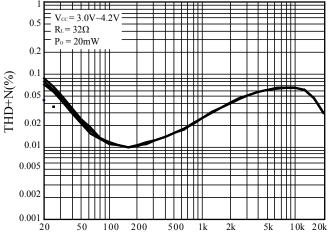




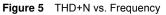
Output Power(W)

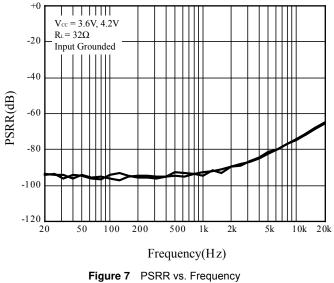






Frequency(Hz)

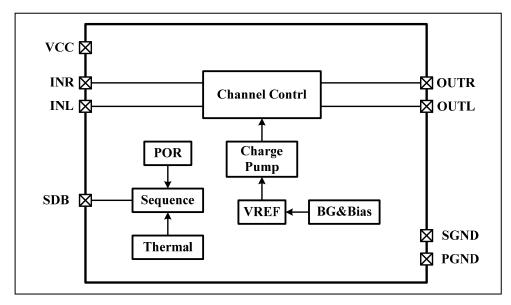




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FUNCTIONAL BLOCK DIAGRAM





APPLICATION INFORMATION

CHARGE PUMP CONVERTER

IS31AP4912 integrate a charge pump converter to change input supply voltage (V_{CC}) into a negative voltage providing a 0V reference voltage for output.

The charge pump converter only needs three external components: supply decoupling capacitor, output bypass capacitor and flying capacitor.

Choose low ESR capacitors to ensure the best operating performance and place the capacitors as close as possible to the IS31AP4912.

GAIN SETTING

The input resistors (R_{IN}) and feedback resistors (R_F) set the gain of the amplifier according to Equation (1).

$$Gain = \frac{R_F}{R_{IN}} \left(\frac{V}{V}\right) \qquad (1)$$

For example, in Figure 1:

 $\mathsf{R}_{\mathsf{F}} = 20\mathsf{k}\Omega, \, \mathsf{R}_{\mathsf{IN}} = 20\mathsf{k}\Omega, \\ 20$

so, $Gain = \frac{20}{20} = 1 \left(\frac{V}{V}\right)$ Resistor matching is very important in the amplifiers.

The balance of the output on the reference voltage depends on matched ratios of the resistors. CMRR, PSRR, and cancellation of the second harmonic distortion diminish if resistor mismatch occurs. Therefore, it is recommended to use 1% tolerance resistors or better to keep the performance optimized. Matching is more important than overall tolerance. Resistor arrays with 1% matching can be used with a tolerance greater than 1%.

Place the input resistors very close to the IS31AP4912 to limit noise injection on the high-impedance nodes.

INPUT CAPACITOR (CIN)

The input capacitors and input resistors form a high pass filter with the corner frequency, f_c , determined in Equation (2).

$$f_c = \frac{1}{2\pi R_{IN} C_{IN}}$$
 (2)

For example, in Figure 1: $R_{IN} = 20k\Omega$, $C_{IN} = 0.47\mu$ F,

so,
$$f_c = \frac{1}{2\pi \times 20k\Omega \times 0.47\,\mu F} \approx 17\,Hz$$

The value of the input capacitor is important to consider as it directly affects the bass (low frequency) performance of the circuit. The capacitors should have a tolerance of $\pm 10\%$ or better, because any mismatch in capacitance causes an impedance mismatch at the corner frequency and below.

DESIGN NOTE

COMPONENT SELECTION

The value and ESR of the output capacitor for charge pump will affect output ripple and transient performance. A X7R or X5R ceramic capacitor in 2.2μ F should be recommended. The flying capacitor should use a 2.2μ F X7R or X5R ceramic capacitor.

All the capacitors should support at least 10V.

PCB LAYOUT

The decoupling capacitors should be placed close to the VCC pin and the output capacitors should be placed close to the VSS pin. The flying capacitor should be placed close to the CN and CP pins. The input capacitors and input resistors should be placed close to the INR and INL pins and the traces must be parallel to prevent noise. The traces of OUTR and OUTL pins connected to the headphone should be as possible as short and wide. The recommended width is 0.5mm.

Trace width should be at least 0.75mm for the power supply and the ground plane. The SGND and PGND pins of the IS31AP4912 must be routed separately back to the decoupling capacitor in order to provide proper device operation. If the SGND and PGND pins are connected directly to each other, the part functions without risk of failure, but the noise and THD performance do not meet the specifications.



CLASSIFICATION REFLOW PROFILES

| Profile Feature | Pb-Free Assembly | | |
|---|----------------------------------|--|--|
| Preheat & Soak Temperature min (Tsmin) Temperature max (Tsmax) Time (Tsmin to Tsmax) (ts) | 150°C 200°C 60-120 seconds | | |
| Average ramp-up rate (Tsmax to Tp) | 3°C/second max. | | |
| Liquidous temperature (TL) Time at liquidous (tL) | 217°C 60-150 seconds | | |
| Peak package body temperature (Tp)* | Max 260°C | | |
| Time (tp)** within 5°C of the specified classification temperature (Tc) | Max 30 seconds | | |
| Average ramp-down rate (Tp to Tsmax) | 6°C/second max. | | |
| Time 25°C to peak temperature | 8 minutes max. | | |

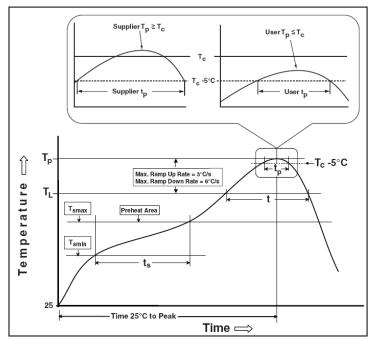
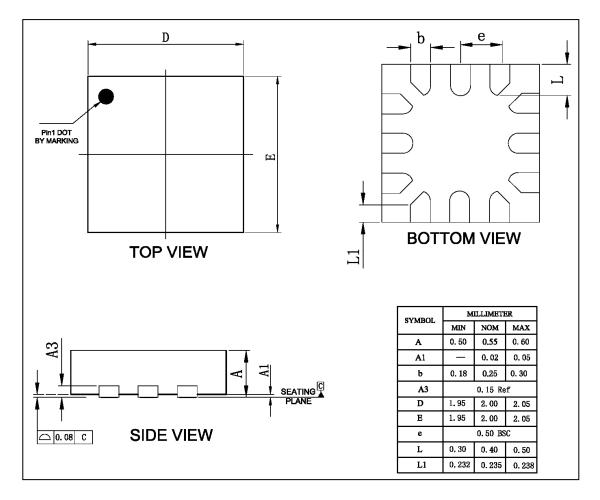


Figure 8 Classification Profile



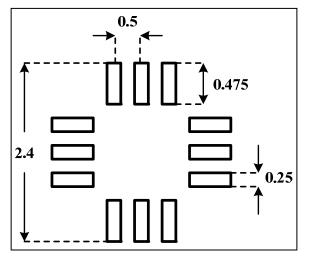
PACKAGING INFORMATION

UTQFN-12





RECOMMENDED LAND PATTERN



Note:

1. Land pattern complies to IPC-7351.

2. All dimensions in MM.

3. This document (including dimensions, notes & specs) is a recommendation based on typical circuit board manufacturing parameters. Since land pattern design depends on many factors unknown (eg. user's board manufacturing specs), user must determine suitability for use.



REVISION HISTORY

| Revision | Detail Information | Date |
|----------|---|------------|
| А | Initial release | 2011.11.17 |
| В | Update POD | 2013.06.06 |
| С | Add ESD value and θ_{JA} Add land pattern and update POD | 2015.12.23 |