



Chipsmall Limited consists of a professional team with an average of over 10 year of expertise in the distribution of electronic components. Based in Hongkong, we have already established firm and mutual-benefit business relationships with customers from,Europe,America and south Asia,supplying obsolete and hard-to-find components to meet their specific needs.

With the principle of “Quality Parts,Customers Priority,Honest Operation,and Considerate Service”,our business mainly focus on the distribution of electronic components. Line cards we deal with include Microchip,ALPS,ROHM,Xilinx,Pulse,ON,Everlight and Freescale. Main products comprise IC,Modules,Potentiometer,IC Socket,Relay,Connector.Our parts cover such applications as commercial,industrial, and automotives areas.

We are looking forward to setting up business relationship with you and hope to provide you with the best service and solution. Let us make a better world for our industry!



## Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

Email & Skype: info@chipsmall.com Web: www.chipsmall.com

Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China



# Digital Dual Output, 7-Phase Configurable PWM Controller with Adaptive Voltage Scaling (AVSBus) Bus

## ISL68137

The [ISL68137](#) is a digital dual output, flexible multiphase ( $X+Y \leq 7$ ) PWM controller with AVSBus (Adaptive Voltage Scaling interface). The ISL68137 can be configured to support any desired phase assignments up to a maximum of 7 phases across the 2 outputs ( $X+Y$ ). For example, 6+1, 5+2, 4+2, 3+3, 3+2, or even a single output operation as a 7+0 configuration are supported. The flexible phase arrangement, combined with PMBus and AVSBus interfaces, allows the device to support any demanding power supply requirement. The ISL68137 with AVSBus complements PMBus by providing a common interface that accelerates point-to-point communication between the controller and the load to statically and dynamically control the processor voltage, thus delivering a balanced and power efficient solution. AVSBus can be used exclusively once the device is configured via PMBus. The ISL68137 utilizes Intersil's proprietary linear synthetic digital current modulation scheme to achieve the industry's best combination of transient response and ease of tuning while addressing the challenges of modern multiphase designs.

Device configuration and telemetry monitoring is accomplished using Intersil's intuitive PowerNavigator™ GUI. The ISL68137 device supports on-chip nonvolatile memory to store various configuration settings that are user selectable via pin-strap, giving system designers increased power density to configure and deploy multiple configurations. The device supports an automatic phase add/drop feature to allow maximum efficiency across all load ranges. Thresholds for automatic phase add/drop are user programmable using the powerful PowerNavigator™ GUI.

The ISL68137 supports a comprehensive fault management system to enable the design of highly reliable systems. From a multitiered overcurrent protection scheme, to the configurable power-good and output overvoltage/undervoltage fault thresholds and temperature monitoring, virtually any need is accommodated.

With minimal external components, easy configuration, robust fault management and highly accurate regulation capability, implementing a high performance multiphase regulator has never been easier.

## Applications

- Networking equipment
- Telecom/datacom equipment
- Server/storage equipment
- Point-of-load power supply (Memory, DSP, ASIC, FPGA)

## Features

- PMBus 1.3 and AVSBus compliant
  - Telemetry -  $V_{IN}$ ,  $V_{OUT}$ ,  $I_{OUT}$ , power IN/OUT, temperature and various fault status registers
  - Individual AVSBus interface enables high speed voltage changes
- Advanced linear digital modulation scheme
  - Zero latency synthetic current control for excellent HF current balance
  - Dual edge modulation for fastest transient response
- Auto phase add/drop for excellent load vs efficiency profile
- Flexible phase configuration
  - 7+0, 6+1, 5+2, 4+3 phase operation
  - Operation using less than 7 phases between 2 outputs is also supported
- Diode braking for overshoot reduction
- Differential remote voltage sensing supports  $\pm 0.5\%$  closed loop system accuracy over load, line and temperature
- Highly accurate current sensing for excellent load line regulation and accurate OCP
  - Supports ISL99227 60A smart power stages
  - Supports DCR sense with integrated temperature compensation
- Supports phase doubler (ISL6617A) for up to 14-phase operation
- Comprehensive fault management enables high reliability systems
  - Pulse-by-pulse phase current limiting
  - Total output current protection
  - Output and input OV/UV
  - Open voltage sense detect
  - Black box recording capability for faults
- Intuitive configuration via [PowerNavigator™](#) GUI
  - NVM to store up to 8 configurations
- Pb-free (RoHS compliant)

## Related Literature

- For a full list of related documents, visit our website
  - [ISL68137](#) product page



## Table of Contents

<b>Ordering Information</b>	<b>3</b>
<b>Pin Configuration</b>	<b>3</b>
<b>Functional Pin Descriptions</b>	<b>4</b>
<b>Driver, DrMOS and Smart Power Stage Recommendation</b>	<b>5</b>
<b>Internal Block Diagram</b>	<b>5</b>
<b>Typical Application: 6+1 Configuration with ISL99227 SPS</b>	<b>6</b>
<b>Typical Application: 4+3 Configuration with ISL99227 SPS</b>	<b>7</b>
<b>Typical Application: 5+2 Configuration with DCR Sensing</b>	<b>8</b>
<b>Typical Application: Phase Doubler</b>	<b>9</b>
<b>Absolute Maximum Ratings</b>	<b>10</b>
<b>Thermal Information</b>	<b>10</b>
<b>Recommended Operating Conditions</b>	<b>10</b>
<b>Electrical Specifications</b>	<b>10</b>
<b>Typical Performance Curves</b>	<b>12</b>
<b>Functional Description</b>	<b>13</b>
Overview	13
PWM Modulation Scheme	13
PMBus Address Selection	13
Phase Configuration	13
Automatic Phase Add and Drop	13
Output Voltage Configuration	14
Switching Frequency	14
Current Sensing	14
Temperature Sensing	15
Temperature Compensation	16
Lossless Input Current and Power Sensing	16
Voltage Regulation	17
Current Feedback	17
Power-On Reset (POR)	17
Soft-Start Delay and Ramp Times	17
Stored Configuration Selection	17
<b>Fault Monitoring and Protection</b>	<b>17</b>
Power-Good Signals	18
Output Voltage Protection	18
Output Current Protection	18
Smart Power Stage OC Fault Detect	20
Thermal Monitoring and Protection	20
<b>Layout and Design Considerations</b>	<b>21</b>
<b>PMBus Operation</b>	<b>22</b>
<b>PMBus Protocol</b>	<b>23</b>
<b>PMBus Command Summary</b>	<b>24</b>
PMBus Use Guidelines	25
PMBus Data Formats	25
<b>PMBus Command Detail</b>	<b>26</b>
<b>Adaptive Voltage Scaling (AVSBus) Functionality and Operation</b>	<b>48</b>
AVSBus Master Send Subframe	48
AVSBus Slave Response Subframe	48
<b>AVSBus Command Detail</b>	<b>49</b>
<b>Revision History</b>	<b>52</b>
<b>About Intersil</b>	<b>52</b>
<b>Package Outline Drawing</b>	<b>53</b>

## Ordering Information

PART NUMBER (Notes 1, 2, 3)	PART MARKING	TEMP. RANGE (°C)	PACKAGE (RoHS COMPLIANT)	PKG. DWG. #
ISL68137IRAZ	ISL68137 IRZ	-40 to +85	48 Ld 6x6 QFN	L48.6x6B

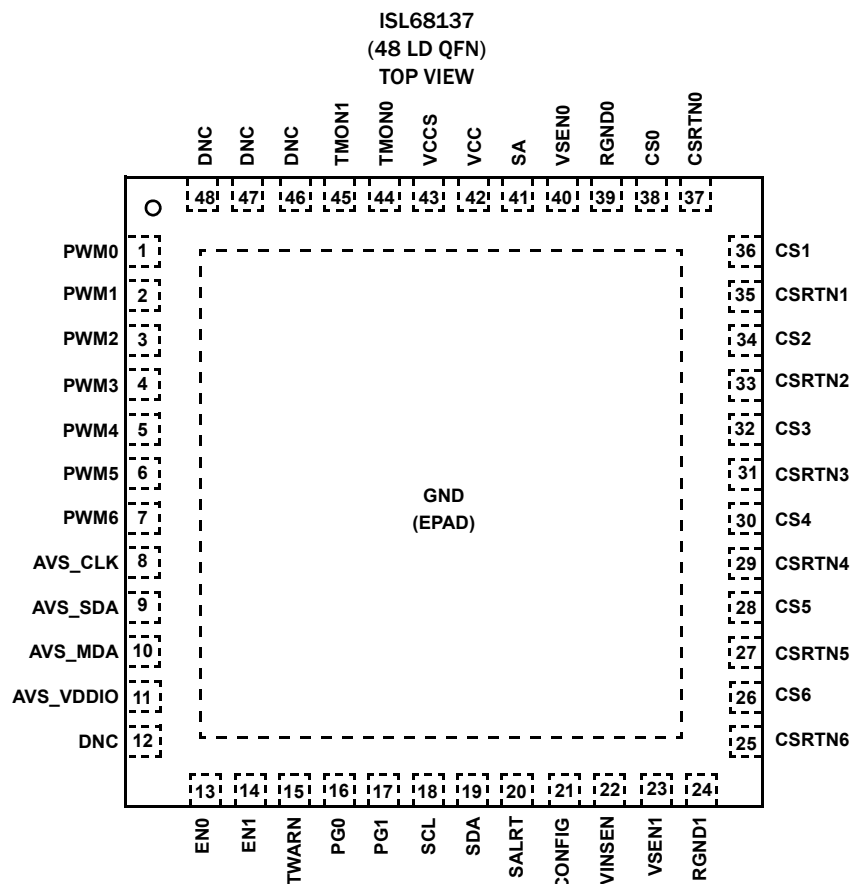
### NOTES:

1. Add "-T" suffix for 4k unit or "-T7A" suffix for 250 unit tape and reel options. Please refer to [TB347](#) for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. For Moisture Sensitivity Level (MSL), please see product information page for [ISL68137](#). For more information on MSL please see techbrief [TB363](#).

TABLE 1. KEY DIFFERENCES BETWEEN FAMILY OF PARTS

PART NUMBER	PHASE CONFIGURATION OUTPUT X/OUTPUT Y	SPECIFICATION SUPPORTED	PACKAGE
ISL68137	X+Y ≤ 7	PMBus/AVSBus	QFN 48 Ld, 6x6mm
ISL68134	X+Y ≤ 4	PMBus/AVSBus	TQFN 40 Ld, 5x5mm
ISL68127	X+Y ≤ 7	PMBus	QFN 48 Ld, 6x6mm
ISL68124	X+Y ≤ 4	PMBus	TQFN 40 Ld, 5x5mm

## Pin Configuration



# ISL68137

## Functional Pin Descriptions Refer to [Table 4 on page 21](#) for design layout considerations.

PIN NUMBER	PIN NAME	DESCRIPTION
7, 6, 5, 4, 3, 2, 1	PWM[6:0]	Pulse width modulation outputs. Connect these pins to the PWM input pins of 3.3V logic compatible Intersil smart power stages, driver IC(s) or power stages.
8	AVS_CLK	AVSBus clock input pin. Connect to ground if not used.
9	AVS_SDA	AVSBus data output pin. Leave open if not used.
10	AVS_MDA	AVSBus data input pin. Connect to ground if not used.
11	AVS_VDDIO	AVSBus reference voltage input pin. Leave open if not used.
12, 46, 47, 48	DNC	Do not connect any signals to these pins.
13	EN0	Input pin used for enable control of Output 0. Active high. Connect to ground if not used.
14	EN1	Input pin used for enable control of Output 1. Active high. Connect to ground if not used.
15	TWARN	Thermal warning flag. This open-drain output will be pulled low in the event of a sensed over-temperature at TMON pins without disabling the regulators. Maximum pull-up voltage is $V_{CC}$ .
16	PG0	Open-drain power-good indicators for Output 0. Maximum pull-up voltage is $V_{CC}$ .
17	PG1	Open-drain power-good indicators for Output 1. Maximum pull-up voltage is $V_{CC}$ .
18	SCL	Serial clock signal pin for SMBus interface. Maximum pull-up voltage is $V_{CC}$ .
19	SDA	Serial data signal pin for SMBus interface. Maximum pull-up voltage is $V_{CC}$ .
20	SALRT	Serial alert signal pin for SMBus interface. Maximum pull-up voltage is $V_{CC}$ .
21	CONFIG	Configuration ID selection pin. See <a href="#">Table 3</a> for more details.
22	VINSEN	Input voltage sense pin. Connect to VIN through a resistor divider (typically 40.2k/10k) with a 10nF decoupling capacitor.
23	VSEN1	Positive differential voltage sense input for Output 1. Connect to positive remote sensing point. Connect to ground if not used.
24	RGND1	Negative differential voltage sense input for Output 1. Connect to negative remote sensing point. Connect to ground if not used.
25, 27, 29, 31, 33, 35, 37	CSRTN[6:0]	The CS and CSRTN pins are current sense inputs to individual phase differential amplifiers. Unused phases should have their respective current sense inputs grounded. The ISL68137 supports smart power stage, DCR and resistor sensing. Connection details depend on current sense method chosen.
26, 28, 30, 32, 34, 36, 38	CS[6:0]	
39	RGND0	Negative differential voltage sense input for Output 0. Connect to negative remote sensing point. Connect to ground if not used.
40	VSEN0	Positive differential voltage sense input for Output 0. Connect to positive remote sensing point. Connect to ground if not used.
41	SA	PMBus Address selection pin. See <a href="#">Table 2 on page 13</a> for more details.
42	VCC	Chip primary bias input. Connect this pin directly to a +3.3V supply with a high quality MLCC bypass capacitor.
43	VCCS	Internally generated 1.2V LDO logic supply from VCC. Decouple with 4.7 $\mu$ F or greater MLCC (X5R or better).
44	TMON0	Input pin for external temperature measurement at Output 0. Supports diode based temperature sensing as well as smart power stage sensing. Refer to section <a href="#">"Temperature Compensation" on page 16</a> for more information.
45	TMON1	Input pin for external temperature measurement at Output 1. Supports diode based temperature sensing as well as smart power stage sensing. Refer to section <a href="#">"Temperature Compensation" on page 16</a> for more information.
EPAD	GND	Package pad serves as GND return for all chip functions. Connect directly to system GND plane with multiple thermal vias.

## Driver, DrMOS and Smart Power Stage Recommendation

INTERSIL PART NUMBER	QUIESCENT CURRENT (mA)	GATE DRIVE VOLTAGE (V)	NUMBER OF DRIVERS	COMMENTS
ISL99227	4.85	5	Single	60A, 5x5 smart power stage
ISL99140	0.19	5	Single	40A, 6x6 DrMOS
ISL6596	0.19	5	Single	Connect ISL6596 VCTRL to 3.3V
ISL6617A	5	N/A	N/A	Phase doubler with 5V PWM output to be compatible with a 60A DrMOS or with 60A smart power stage. Supports up to a 14-phase design.

## Internal Block Diagram

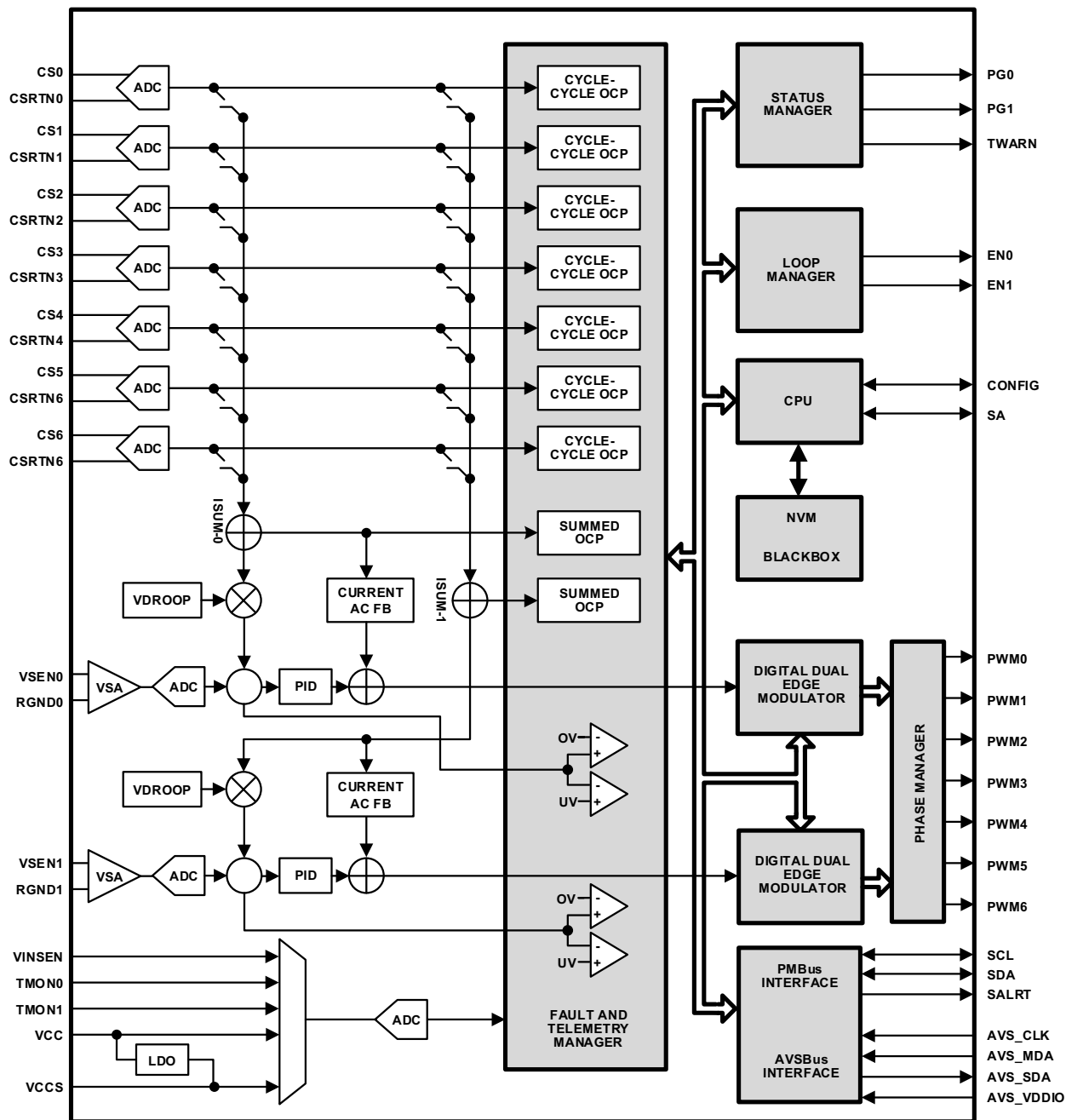


FIGURE 1. INTERNAL BLOCK DIAGRAM

## Typical Application: 6+1 Configuration with ISL99227 SPS

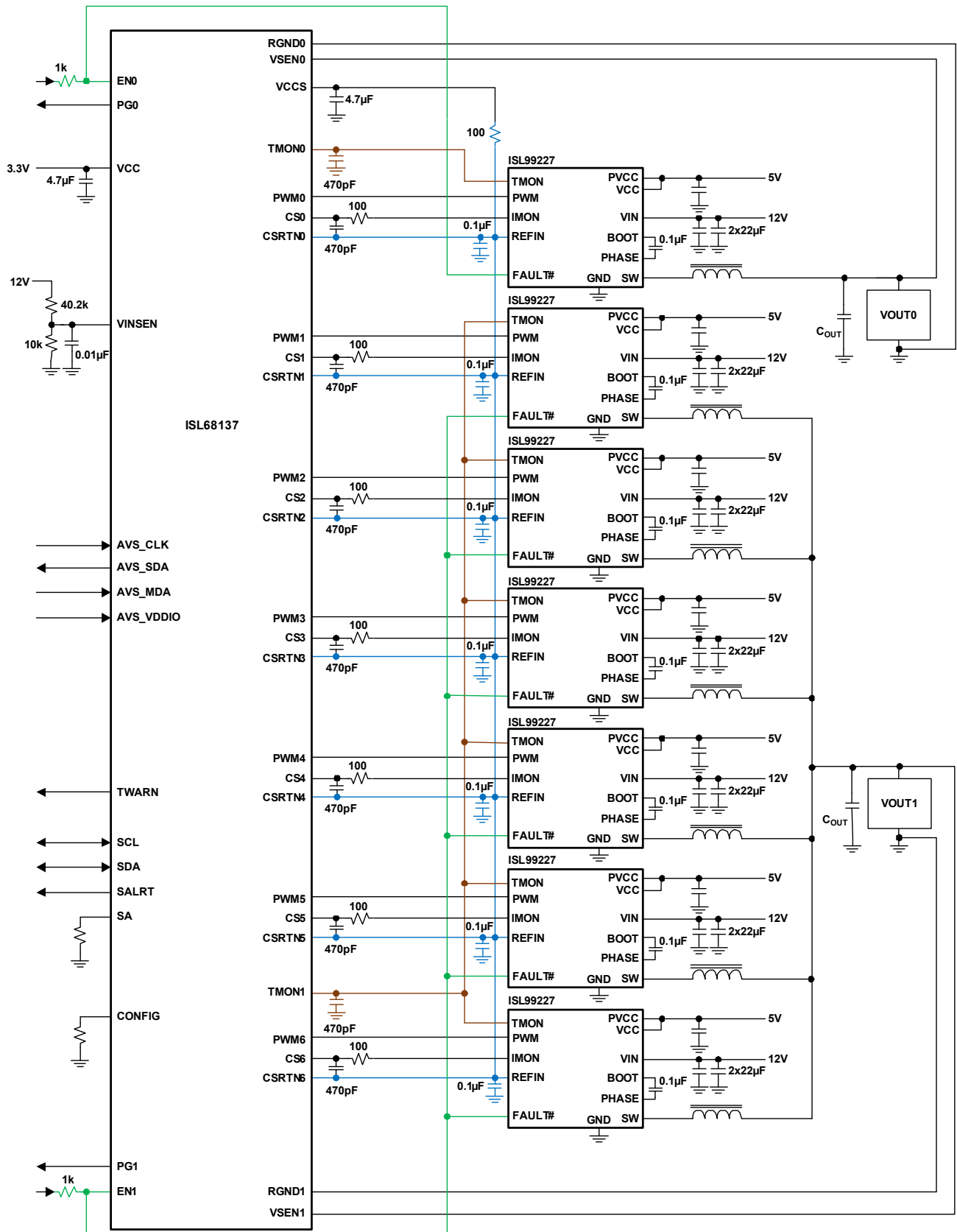


FIGURE 2. TYPICAL APPLICATION: 6+1 CONFIGURATION WITH ISL99227 SPS

## Typical Application: 4+3 Configuration with ISL99227 SPS

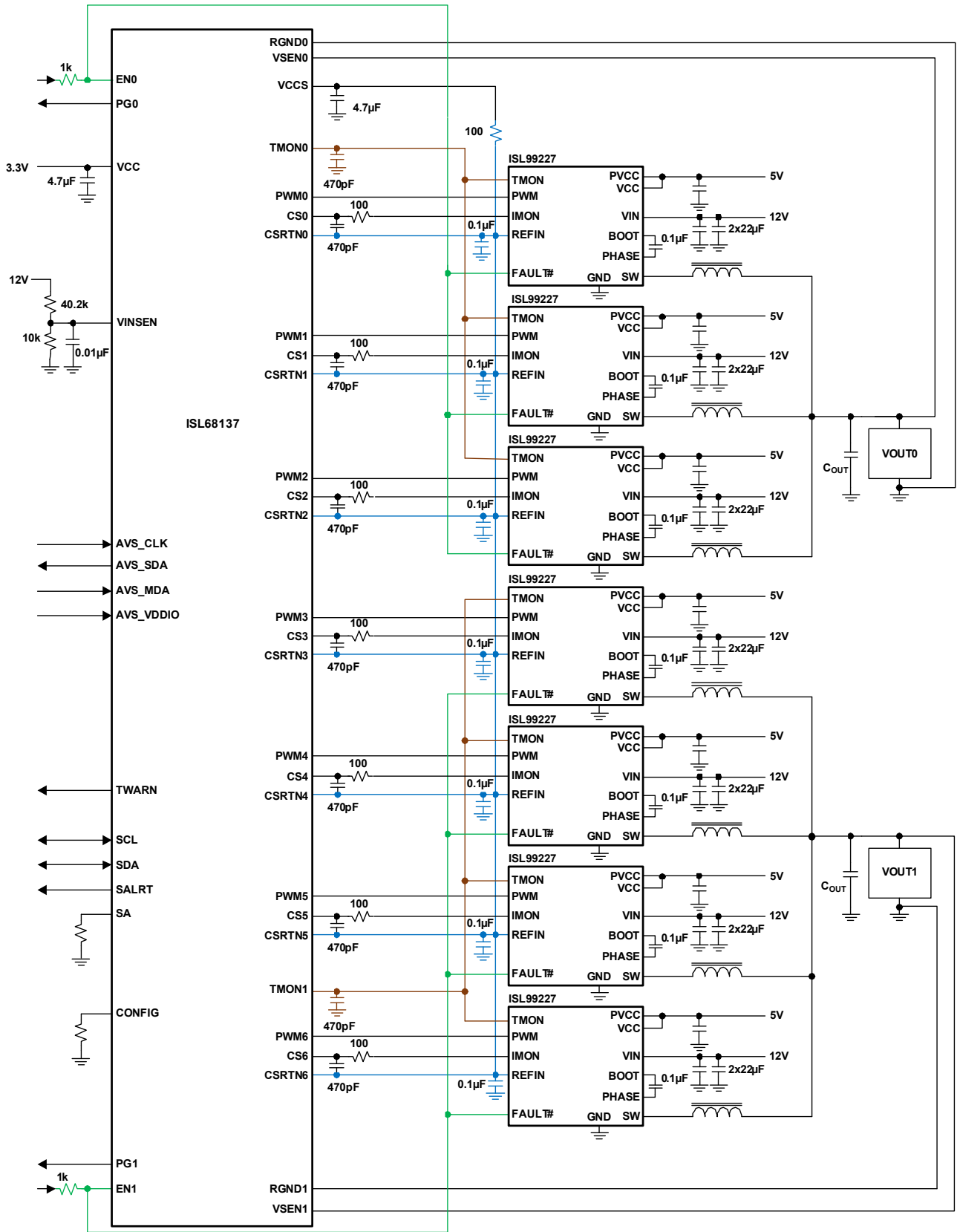


FIGURE 3. TYPICAL APPLICATION: 4+3 CONFIGURATION WITH ISL99227 SPS





## Typical Application: Phase Doubler

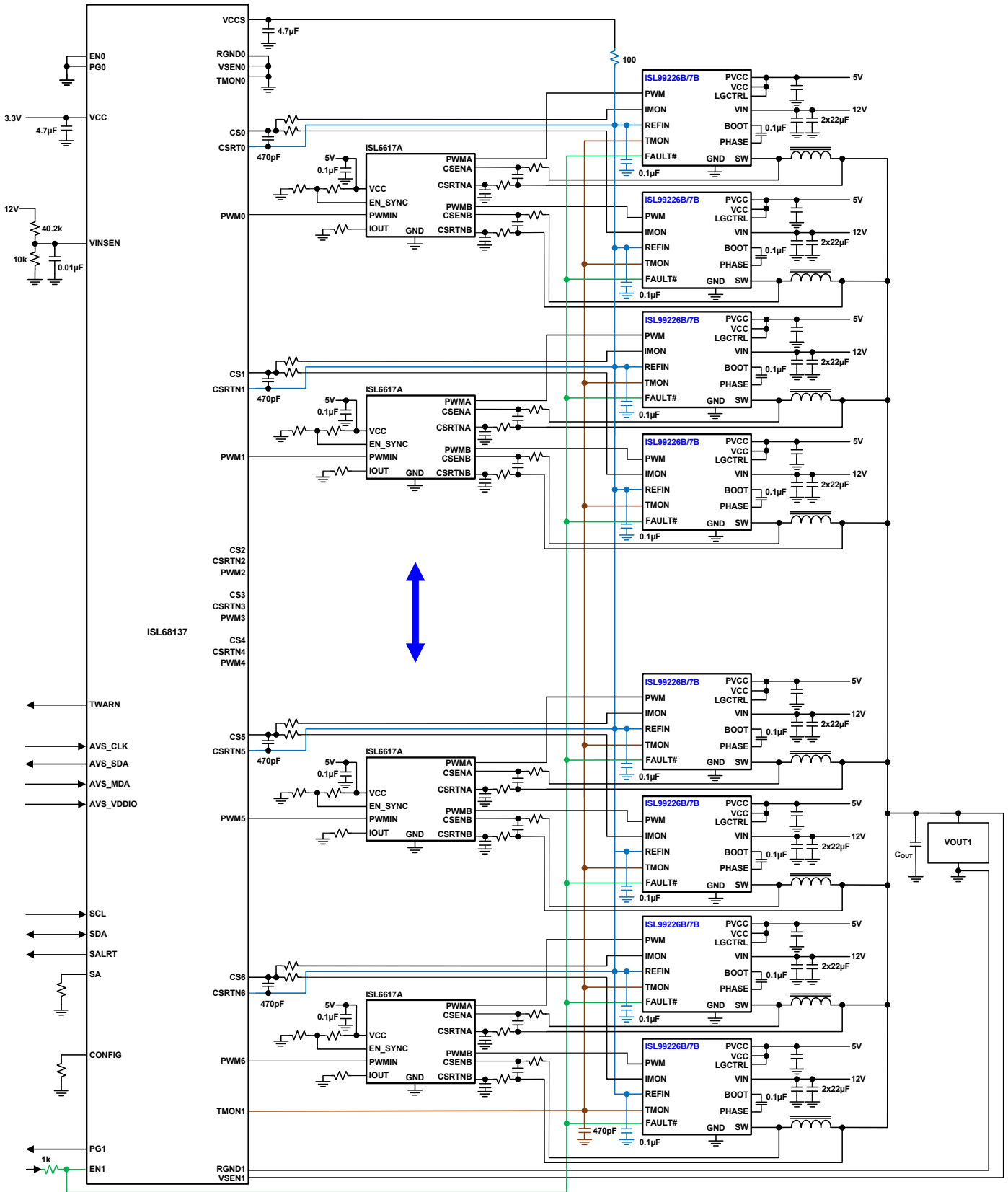


FIGURE 5. TYPICAL APPLICATION: PHASE DOUBLER

# ISL68137

## Absolute Maximum Ratings

VCC	+4.3V
VCCS	+1.6V
All Other Pins	(GND - 0.3V) to VCC + 0.3V
ESD Rating:	
Human Body Model (Tested per JS-001-2014)	2kV
Charged Device Model (Tested per JS-002-2014)	1kV
Latch-Up (Tested per JESD-78D; Class 2, Level A)	100mA

## Thermal Information

Thermal Resistance (Notes 4, 5)	$\theta_{JA}$ (°C/W)	$\theta_{JC}$ (°C/W)
48 Ld 6x6 QFN Package	27	1
Maximum Junction Temperature	+150°C	
Maximum Storage Temperature Range	-65°C to +150°C	
Pb-Free Reflow Profile	see <a href="#">TB493</a>	

## Recommended Operating Conditions

Supply Voltage, VCC	+3.3V ±5%
Ambient Temperature	-40°C to +85°C
Output Voltage	0V to 3.05V

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

### NOTES:

- $\theta_{JA}$  is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See [TB379](#).
- For  $\theta_{JC}$ , the "case temp" location is the center of the exposed metal pad on the package underside.

**Electrical Specifications** Recommended operating conditions, VCC = 3.3V, unless otherwise specified. **Boldface limits apply across the operating temperature range -40°C to +85°C.**

PARAMETER	TEST CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
<b>VCC SUPPLY CURRENT</b>					
Nominal Supply Current	VCC = 3.3VDC; EN1/2 = VIH, fSW = 400kHz		90.5		mA
Shutdown Supply Current	VCC = 3.3VDC; EN1/2 = 0V, no switching		11.4		mA
<b>VCCS LDO SUPPLY</b>					
Output Voltage		<b>1.20</b>	1.25	<b>1.30</b>	V
Maximum Current Capability	Excluding internal load	<b>50</b>			mA
<b>POWER-ON RESET AND INPUT VOLTAGE LOCKOUT</b>					
VCC Rising POR Threshold			2.7	<b>2.9</b>	V
VCC Falling POR Threshold		<b>1.0</b>			V
Enable (ENO and EN1) Input Threshold			2.3		V
Enable (ENO and EN1) LOW to HIGH Ramp Delay (TON_DELAY)		<b>200</b>			µs
POR to Initialization Complete Time			30	<b>40</b>	ms
<b>OUTPUT VOLTAGE CHARACTERISTICS (Note 6)</b>					
Output Voltage Adjustment Range		<b>0.25</b>		<b>3.05</b>	V
Output Voltage Set-Point Accuracy	Set-point 0.8V to 3.05V	<b>-0.5</b>		<b>0.5</b>	%
	Set-point 0.25V to <0.8V	<b>-5</b>		<b>5</b>	mV
<b>VOLTAGE SENSE AMPLIFIER</b>					
Open Sense Current	Only during open pin check of initialization		22		µA
Input Impedance (VSEN - RGND)			200		kΩ
Maximum Common-Mode Input			VCC - 0.2		V
Maximum Differential Input (VSEN - RGND)				<b>3.05</b>	V
<b>CURRENT SENSE AND OVERCURRENT PROTECTION</b>					
Maximum Common-Mode Input (SPS mode)	CSRTNx - GND		1.6		V
Maximum Common-Mode Input (DCR mode)	CSRTNx - GND		3.3		V
Current Sense Accuracy	ISEN to ADC accuracy	<b>-2</b>		<b>2</b>	%

# ISL68137

**Electrical Specifications** Recommended operating conditions,  $V_{CC} = 3.3V$ , unless otherwise specified. **Boldface limits apply across the operating temperature range -40°C to +85°C. (Continued)**

PARAMETER	TEST CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
Average Overcurrent Threshold Resolution			0.1		A
Cycle-by-Cycle Current Limiting Threshold Accuracy			0.1		A
<b>DIGITAL DROOP</b>					
Droop Resolution			0.01		mV/A
<b>OSCILLATORS</b>					
Accuracy of Switching Frequency Setting	When set to 500kHz	<b>480</b>	500	<b>520</b>	kHz
Accuracy of Switching Frequency Setting		<b>-4</b>		<b>4</b>	%
Switching Frequency Range		<b>0.2</b>		<b>1.0</b>	MHz
<b>SOFT-START RATE AND VOLTAGE TRANSITION RATE</b>					
Minimum Soft-Start Ramp Rate	Programmable minimum rate		20		μs
Maximum Soft-Start Ramp Rate	Programmable maximum rate		10		ms
Soft-Start Ramp Rate Accuracy		<b>-4</b>		<b>4</b>	%
Minimum Transition Rate	Programmable minimum rate		0.1		mV/μs
Maximum Transition Rate	Programmable maximum rate		100		mV/μs
Transition Rate Accuracy		<b>-4</b>		<b>4</b>	%
<b>PWM OUTPUT</b>					
PWMx Output High Level	$I_{OUT} = 4mA$	<b><math>V_{CC} - 0.4</math></b>			V
PWMx Output Low Level	$I_{OUT} = 4mA$			<b>0.4</b>	V
PWMx Output Tri-State $I_{OL}$	$V_{OH} = V_{CC}$			<b>1</b>	μA
PWMx Output Tri-State $I_{OH}$	$V_{OL} = 0V$	<b>-1</b>			μA
<b>THERMAL MONITORING AND PROTECTION</b>					
Temperature Sensor Range		<b>-50</b>		<b>150</b>	°C
Temperature Sensor Accuracy	TMON to ADC accuracy	<b>-4.5</b>		<b>4.5</b>	%
TWARN Output Low Impedance		<b>4</b>	9	<b>13</b>	?
TWARN Hysteresis			3		°C
<b>POWER-GOOD AND PROTECTION MONITORS</b>					
PG Output Low Voltage	$I_{OUT} = 8mA$ load			<b>0.4</b>	V
PG Leakage Current	With pull-up resistor externally connected to VCC		0.5	<b>1.0</b>	μA
Overvoltage Protection Threshold Resolution			1		mV
Undervoltage Protection Threshold Resolution			1		mV
Overvoltage Protection Threshold when Disabled			$V_{CC} - 0.2$		V
<b>INPUT VOLTAGE SENSE</b>					
Input Voltage Accuracy	VINSEN to ADC accuracy	<b>-2.5</b>		<b>2.5</b>	%
Input Voltage Protection Threshold Resolution			1		mV
<b>AVSBus</b>					
AVSBus VDDIO Input Voltage Range		<b>0.90</b>		<b>3.63</b>	V
AVSBus CLK, MDA, Input High Level		<b><math>0.6 * V_{DDIO}</math></b>			V
AVSBus CLK, MDA, Input Low Level				<b><math>0.4 * V_{DDIO}</math></b>	V

# ISL68137

**Electrical Specifications** Recommended operating conditions,  $V_{CC} = 3.3V$ , unless otherwise specified. **Boldface limits apply across the operating temperature range  $-40^{\circ}C$  to  $+85^{\circ}C$ .** (Continued)

PARAMETER	TEST CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
AVSBus SDA, Output High Level		<b><math>0.8 * V_{DDIO}</math></b>			V
AVSBus SDA, Output Low Level				<b><math>0.2 * V_{DDIO}</math></b>	V
AVSBus CLK Maximum Frequency		<b>50</b>			MHz
AVSBus CLK Minimum Frequency				<b>5</b>	MHz
<b>SMBus/PMBus</b>					
SALERT, SDA Output Low Level	$I_{OUT} = 4mA$			<b>0.4</b>	V
SCL, SDA Input High/Low Threshold			1.25		V
SCL, SDA Input Hysteresis			2		mV
SCL Maximum Frequency		<b>0.05</b>		<b>2.00</b>	MHz

**NOTES:**

- These parts are designed and adjusted for accuracy with all errors in the voltage loop included.
- Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.

## Typical Performance Curves

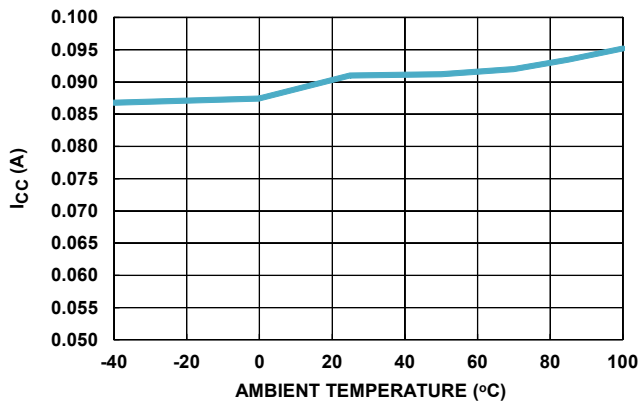


FIGURE 6. NOMINAL SUPPLY CURRENT vs TEMPERATURE

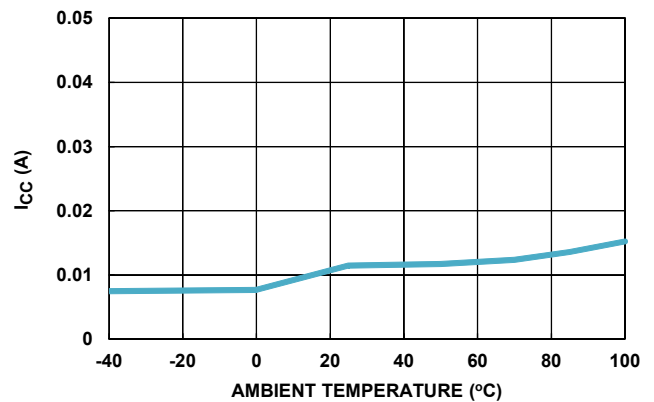


FIGURE 7. SHUTDOWN SUPPLY CURRENT vs TEMPERATURE



## Functional Description

### Overview

The ISL68137 is a digital dual output 7-phase PWM controller that can be programmed for single output 7+0, dual output 6+1, 5+2, or 4+3 phase operation. Operation using less than 7 phases between 2 outputs is also supported. Existing digital multiphase solutions utilize analog comparator based schemes (nonlinear) to bolster the inadequate transient response common to many digital multiphase solutions. The ISL68137 uses a linear voltage regulation scheme to address transient loads. As a result, it is much easier for users to configure and validate their designs when compared with nonlinear schemes. By combining a proprietary low noise and zero latency digital current sense scheme with cutting edge digital design techniques, Intersil is able to meet transient demands without resorting to nonlinear schemes. In addition, the ISL68137 can store up to 8 user configurations in NVM and allows the user to select the desired configuration via pin-strap (CONFIG). The result is a system that is easy to configure and deploy.

A number of performance enhancing features are supported in the ISL68137. These include AVSBus control, diode braking, automatic phase dropping, DCR/resistor/smart power stage current sense support, load line regulation and multiple temperature sensing options.

To facilitate configuration development, the PowerNavigator™ GUI provides a step-by-step arrangement for setup and parametric adjustment. Once a configuration has been set, the user may employ PowerNavigator™ to monitor telemetry or use a direct PMBus interface based on the supported command set.

### PWM Modulation Scheme

The ISL68137 uses Intersil's proprietary linear synthetic current modulation scheme to improve transient performance. This is a unique constant frequency, dual edge PWM modulation scheme with both PWM leading and trailing edges being independently moved to give the best response to transient loads. Current balance is an inherent part of the regulation scheme. The modulation scheme is capable of overlapping pulses should the load profile demand such operation. In addition, the modulator is capable of adding or removing pulses from a given cycle in response to regulation demands while still managing maximum average frequency to safe levels. For DC load conditions the operating frequency is constant.

### PMBus Address Selection

When communicating with multiple PMBus devices on a single bus, each device must have its own unique address so the host can distinguish between the devices. The device address can be set using a 1% resistor on the SA pin according to the pin-strap options listed in [Table 2](#).

TABLE 2. RESISTOR VALUES TO ADDRESS MAPPING

R SA (Ω)	PMBus ADDRESS	R SA (Ω)	PMBus ADDRESS
0	60h	1500	50h
180	61h	1800	51h
330	64h	2200	54h
470	65h	2700	55h
680	40h	3300	58h
820	41h	3900	59h
1000	44h	4700	5Ch
1200	45h	5600	5Dh

### Phase Configuration

The ISL68137 supports up to two regulated outputs through seven configurable phases. Either output is capable of controlling up to seven phases in any arbitrary mix. Phase assignments are accomplished via the PowerNavigator™ GUI.

While the device supports arbitrary phase assignment, it is good practice to assign phases to Output 1 in descending sequential numerical order starting from Phase 6. For example, a 4-phase rail could consist of phases 6, 5, 4 and 3. For Output 0, phases would be assigned starting from Phase 0 in ascending sequential numerical order.

### Automatic Phase Add and Drop

In order to produce the most optimal efficiency across a wide range of output loading, the modulator supports automatic dropping or adding of phases. Use of automatic phase dropping is optional. If automatic phase dropping is enabled, the number of active phases at any time is determined solely by load current. During operation, phases of Output 1 will drop beginning with the lowest phase number assigned. Phase dropping begins with the highest assigned phase number. [Figure 8](#) illustrates the typical characteristic of efficiency vs load current vs phase count.

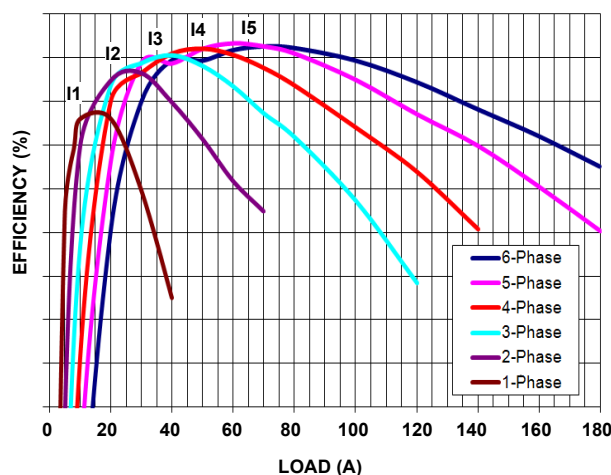


FIGURE 8. EFFICIENCY vs PHASE NUMBER

Phases are dropped one at a time with a user programmed drop delay between drop events. As an example, suppose the delay is set to 1ms and 3 phases are active. Should the load suddenly drop to a level needing only 1 phase, the ISL68137 will begin by dropping a phase after 1ms. An additional phase will be dropped each 1ms thereafter until only 1 phase remains.

In addition to the described load current add/drop thresholds, the fast phase add function provides a very rapid response to transient load conditions. This feature continuously monitors the system regulation error and should it exceed the user set threshold, all dropped phases will be readied for use. In this way, there is no delay should all phases be needed to support a load transient. The fast phase add threshold is set in the PowerNavigator™ GUI. Output current threshold for adding and dropping phases can also be configured.

To ensure dropped phases have sufficient boot capacitor charge to turn on the high-side MOSFET after a long period of disable, a boot refresh circuit turns on the low-side MOSFET of each dropped phase to refresh the boot capacitor. Frequency of the boot refresh is programmable via PowerNavigator™.

## Output Voltage Configuration

Output voltage set points and thresholds for each output can be configured with PowerNavigator™ GUI. Parameters such as output voltage,  $V_{OUT}$  margin high/low and  $V_{OUT}$  OV/UV fault thresholds can be configured with GUI. Additionally, output voltage and margin high/low can be adjusted during regulation via PMBus command  $V_{OUT\_COMMAND}$ ,  $V_{OUT\_MARGIN\_HIGH}$  and  $V_{OUT\_MARGIN\_LOW}$  for further tuning. The following  $V_{OUT}$  relationships must be maintained for correct operation:  
 $V_{OUT\_OV\_FAULT\_LIMIT} > V_{OUT\_COMMAND}$   
 $(V_{OUT\_MARGIN\_HIGH}$  and  $V_{OUT\_MARGIN\_LOW}$ , if used)  $> V_{OUT\_UV\_FAULT\_LIMIT}$ . Additionally, the  $V_{OUT}$  commands are bounded by  $V_{OUT\_MAX}$  and  $V_{OUT\_MIN}$  to provide protection against incorrect set points being sent to the device. The ISL68137 also incorporates AVSBus functionality for high speed changes to the  $V_{OUT}$  target.

## Switching Frequency

Switching frequency is user configurable over a range of 200kHz to 1MHz.

## Current Sensing

The ISL68137 supports DCR, resistor and smart power stage current sensing. Connection to the various sense elements is accomplished via the CS and CSRTN pins. Current sensing inputs are high impedance differential inputs to reject noise and ground related inaccuracies.

To accommodate a wide range of effective sense resistance, information about the effective sense resistance and required per phase current capability is utilized by the GUI to properly configure the current sense circuitry.

## INDUCTOR DCR SENSING

DCR sensing takes advantage of the fact that an inductor winding has a resistive component (DCR) that will drop a voltage proportional to the inductor current. Figure 9 shows that the DCR is treated as a lumped element with one terminal inaccessible for measurement. Fortunately, a simple R-C network as shown in Figure 10 is capable of reproducing the hidden DCR voltage. By simply matching the R-C time constant to the L/DCR time constant, it is possible to precisely recreate the DCR voltage across the capacitor. This means that  $V_{DCR}(t) = V_C(t)$ , thus preserving even the high frequency characteristic of the DCR voltage.

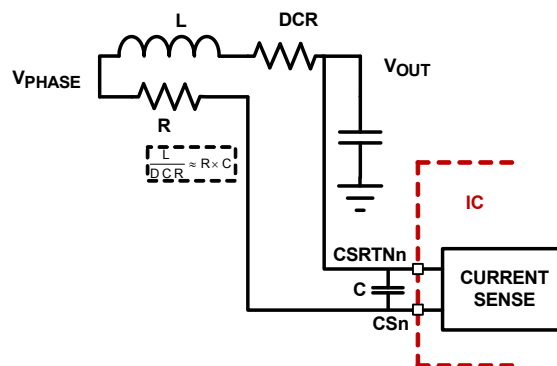


FIGURE 9. DCR SENSING CONFIGURATION

Modern inductors often have such low DCR values that the resulting signal is  $<10\text{mV}$ . To avoid noise problems, care must be taken in the PCB layout to properly place the R-C components and route the differential lines between controller and inductor. Figure 9 graphically shows one PCB design method that places the R component near the inductor VPHASE and the C component very close to the IC pins. This minimizes routing of the noisy VPHASE and maximizes filtering near the IC. Route the lines between the inductor and IC as a pair on a single layer directly to the controller. Care must be taken to avoid routing the pair near any switching signals including Phase, PWM etc. This is the method used by Intersil on evaluation board designs.

This method is sensing the resistance of a metal winding where the DCR value will increase with temperature. This must be compensated or the sensed (and reported) current will increase with temperature. In order to compensate the temperature effect, the ISL68137 provides temperature sensing options and an internal methodology to apply the correction.

## RESISTIVE SENSING

For more accurate current sensing, a dedicated current sense resistor  $R_{SENSE}$  in series with each output inductor can serve as the current sense element. This technique, however, reduces the overall converter efficiency due to the additional power loss on the current sense element  $R_{SENSE}$ .



Sensed temperature is utilized in the system for faults, telemetry and temperature compensation of sensed current.

## Temperature Compensation

The ISL68137 supports inductor DCR sensing, which generally requires temperature compensation due to the copper wire used to form inductors. Copper has a positive temperature coefficient of approximately 0.39%/°C. Since the voltage across the inductor is sensed for the output current information, the sensed current has the same positive temperature coefficient as the inductor DCR.

Compensating current sense for temperature variation generally requires that the current sensing element temperature and its temperature coefficient is known. While temperature coefficient is generally obtained easily, actual current sense element temperature is essentially impossible to measure directly. Instead, a temperature sensor (a BJT for the ISL68137) placed near the inductors is measured and the current sense element (DCR) temperature is calculated from that measurement. Calculating current sense element temperature is equivalent to applying gain and offset corrections to the temperature sensor measurement and the ISL68137 supports both corrections.

Figure 15 depicts the block diagram of temperature compensation. A BJT placed near the inductors used for DCR sensing is monitored by the IC utilizing the well known delta Vbe method of temperature sensing.  $T_{SENSE}$  is the direct measured temperature of the BJT. Because the BJT is not directly sensing DCR, corrections must be made such that  $T_{DCR}$  reflects the true DCR temperature. Corrections are applied according to the relationship shown in Equation 1, where  $k_{SLOPE}$  represents a gain scaling and  $T_{OFFSET}$  represents an offset correction. These parameters are provided by the designer via the PowerNavigator™ GUI:

$$T_{DCR} = k_{SLOPE} \cdot T_{SENSE} + T_{OFFSET} \quad (EQ. 1)$$

Once  $T_{DCR}$  has been determined, the compensated DCR value may be determined according to Equation 2, where  $DCR_{25}$  is the DCR at +25°C and  $T_C$  is the temperature coefficient of copper (3900 ppm/°C).  $T_{DCR} = T_{ACTUAL}$  here:

$$DCR_{CORR} = DCR_{25} \cdot (1 + T_C \cdot (T_{ACTUAL} - 25)) \quad (EQ. 2)$$

Thus, the temperature compensated DCR is now used to determine the actual value of current in the DCR sense element.

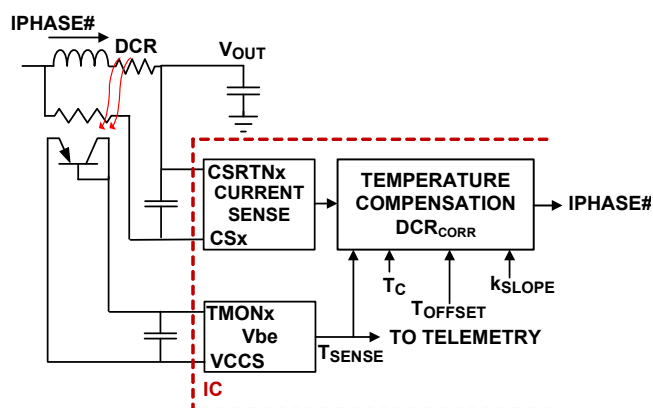


FIGURE 15. BLOCK DIAGRAM OF TEMPERATURE COMPENSATION

In the physical PCB design, the temperature sense diode (BJT) is placed close to the inductor of the phase that is never dropped during automatic phase drop operation. Additionally, a filter capacitor no larger than 500pF should be added near the IC between each TMON pin and VCCS. This is shown in Figure 16.

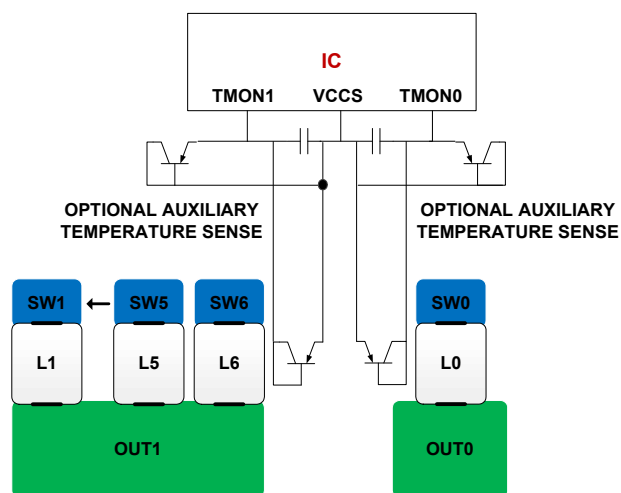


FIGURE 16. RECOMMENDED PLACEMENT OF BJT

## Lossless Input Current and Power Sensing

Input current telemetry is provided via an input current synthesizer. By utilizing the IC's ability to precisely determine its operational conditions, input current can be synthesized to a high degree of accuracy without the need for a lossy sense resistor. Fine-tuning of offset and gain are provided for in the GUI. Note that input current sense fine-tuning must be done after output current sense setup is finalized. With a precise knowledge of input current and voltage, input power may be computed.

Input current and power telemetry is accessed via a PMBus and easily monitored in the PowerNavigator™ GUI.  $V_{IN}$  is monitored directly by the VINSN pin through a 1:5 resistor divider as shown in Figure 17.

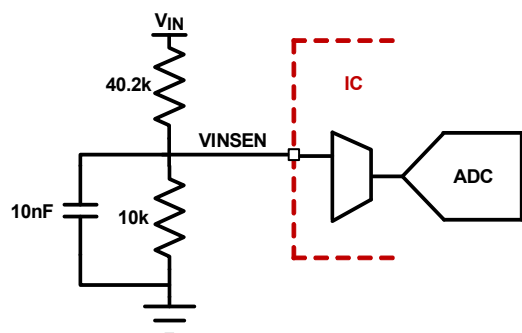


FIGURE 17. INPUT VOLTAGE SENSE CONFIGURATION

## Voltage Regulation

Output voltage is sensed through the remote sense differential amplifier and digitized. From this point, the regulation loop is entirely digital. Traditional PID controls are utilized in conjunction with several enhanced methods to compensate the voltage regulation loop and tune the transient response.

## Current Feedback

Current feedback in a voltage regulator is often utilized to ease the stability design of the voltage feedback path. Additionally, many microprocessors require the voltage regulator to have a controlled output resistance (known as load-line or droop regulation) and this is accomplished utilizing current feedback.

For applications requiring droop regulation, the designer simply specifies the output resistance desired using the PowerNavigator™ GUI.

Current feedback stability benefits are available for rails that do not specify droop regulation such as system agent. For these applications, the designer may enable AC current feedback in the GUI. With this configuration, the DC output voltage will be steady regardless of load current.

## Power-On Reset (POR)

Initialization of the ISL68137 begins after  $V_{CC}$  crosses its rising POR threshold. When POR conditions are met, the internal 1.2V LDO is enabled and basic digital subsystem integrity checks begin. During this process, the controller will load the selected user configuration from NVM as indicated by the CONFIG pin resistor value, read VIN UVLO thresholds from memory and start the telemetry subsystem. With telemetry enabled,  $V_{IN}$  may be monitored to determine when it exceeds its user programmable rising UVLO threshold. Once  $V_{CC}$  and  $V_{IN}$  satisfy their respective voltage conditions, the controller is in its shutdown state. It will transition to its active state and begin soft-start when the state of EN0/EN1 command a start-up. While in shutdown mode, the PWM outputs are held in a high-impedance state to assure the drivers remain off.

## Soft-Start Delay and Ramp Times

It may be necessary to set a delay from when an enable signal is received until the output voltage starts to ramp to its target value. In addition, the designer may wish to precisely set the time required for an output to ramp to its target value after the delay period has expired. These features may be used as part of an

overall inrush current management strategy or to precisely control how fast a load IC is turned on. The ISL68137 gives the system designer several options for precisely and independently controlling both the delay and ramp time periods. The soft-start delay period begins when the EN pin is asserted and ends when the delay time expires.

The soft-start delay and ramp-up/down times can be configured via PowerNavigator™ GUI. The device needs approximately 200 $\mu$ s after enable to initialize before starting to ramp up. When the soft-start ramp period is set to 0ms, the output ramps up as quickly as the output load capacitance and loop settings allow. It is recommended to set the ramps to a non-zero value to prevent inadvertent fault conditions due to excessive inrush current.

## Stored Configuration Selection

As many as eight configurations may be stored and used at any time using the on-board nonvolatile memory. Configurations are assigned an identifier number between 0 and 7 at power-up. The device will load the configuration indicated by the 1% resistor value detected on the CONFIG pin. Resistor values are used to indicate use of one of the eight possible configurations. [Table 3](#) provides the resistor value corresponding to each configuration identifier.

TABLE 3. RESISTOR VALUES TO CONFIGURATION MAPPING

R CONFIG ( $\Omega$ )	CONFIG ID
6800	0
1800	1
2200	2
2700	3
3300	4
3900	5
4700	6
5600	7

Only the most recent configuration with a given number can be loaded. The device supports a total of 8 stored operations. As an example, a configuration with the identifier 0 could be saved 8 times or configurations with all 8 identifiers could be stored one time each for a total of 8 save operations.

PowerNavigator™ provides a simple interface to save and load configurations.

## Fault Monitoring and Protection

The ISL68137 actively monitors temperature, input voltage, output voltage and output current to detect and report fault conditions. Fault monitors trigger configurable protective measures to prevent damage to a load. The power-good indicators, PG0/PG1, are provided for linking to external system monitors.

A high level of flexibility is provided in the ISL68137 fault logic. Faults may be enabled or disabled individually. Each fault type can also be configured to either latch off or retry indefinitely.



## Power-Good Signals

The PG0/PG1 pins are open-drain power-good outputs that indicate completion of the soft-start sequence and output voltage of the associated rail within the expected regulation range.

The PG pins may be associated or disassociated with a number of the available fault types. This allows a system design to be tailored for virtually any condition. In addition, these power-good indicators will be pulled low when a fault (OCP or OVP) condition or UV condition is detected on the associated rail.

## Output Voltage Protection

Output voltage is measured at the load sensing points differentially for regulation and the same measurement is used for OVP and UVP. The fault thresholds are set using PMBus commands. [Figure 18](#) shows a simplified OVP/UVP block diagram. The output voltage comparisons are done in the digital domain.

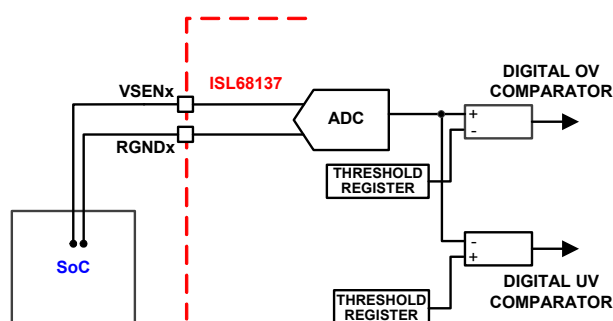


FIGURE 18. OVP, UVP COMPARATORS

The device responds to an output overvoltage condition by disabling the output, declaring a fault, setting the SALRT pin, setting the PG pin and then pulsing the LFET until the output voltage has dropped below the threshold. Similarly, the device responds to an output undervoltage condition by disabling the output, declaring a fault, setting the SALRT pin and setting the PG pin. The output will not restart until the EN pin is cycled (unless the device is configured to retry).

In addition, the ISL68137 features open pin sensing protection to detect an open of the output voltage sensing circuit. When this condition is detected, controller operation is suspended.

## Output Current Protection

The ISL68137 offers a comprehensive overcurrent protection scheme. Each phase is protected from both excessive peak current and sustained current. In addition, the system is protected from sustained total output overcurrent.

[Figure 19](#) depicts a block diagram of the system total output current protection scheme. In this scheme, the phase currents are summed to form ISUM. ISUM is then fed to dual response paths allowing the user to program separate LPF, threshold and response time. One path is intended to allow response more quickly than the other path. With this system, the user can allow high peak total current for a short time and a lower level of current for a sustained time. Note that neither of these paths affect PWM activity on a cycle-by-cycle basis. The characteristics of each path are easily set in PowerNavigator™.

In addition to total output current, the ISL68137 provides an individual phase peak current limit that will act on PWM in a cycle-by-cycle manner. This means that if a phase current is detected to exceed the OC threshold, the phase PWM signal will be inverted to move current away from the threshold. In addition to limiting positive or negative peak current on a cycle-by-cycle basis, individual phase OC can be configured to limit current indefinitely or to declare a fault after a programmable number of consecutive OC cycles. This feature is useful for applications where a fault shutdown of the system would not be acceptable but some ability to limit phase currents is desired. [Figures 22](#) and [23](#) on [page 19](#) depict this operation. If configured for indefinite current limit, the converter will act as a current source and  $V_{OUT}$  will not remain at its regulation point. It should be noted that in this case,  $V_{OUT}$  OV or UV protection action may occur, which could shut the regulator down.

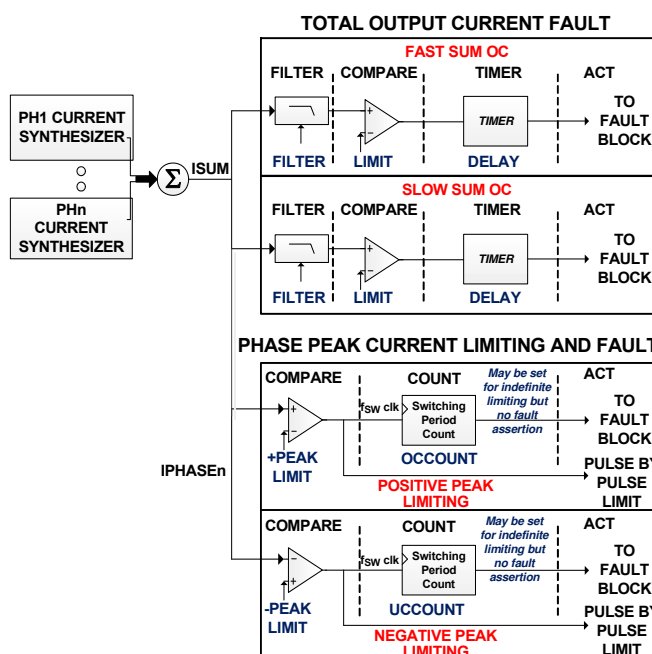


FIGURE 19. OCP FUNCTIONAL DIAGRAM

An example of the OCP\_Fast and OCP\_Slow waveforms are shown in [Figures 20](#) and [21](#).

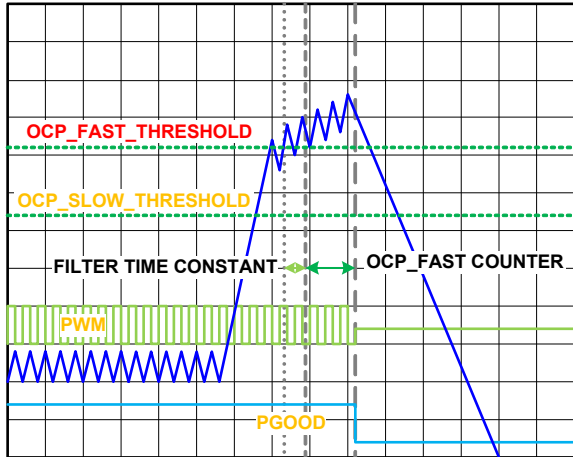


FIGURE 20. OCP\_FAST

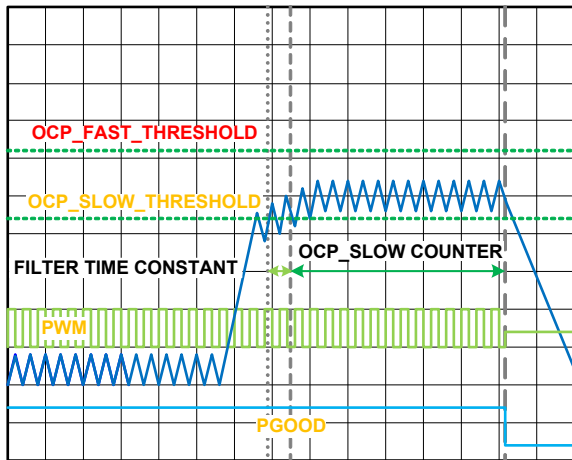


FIGURE 21. OCP\_SLOW

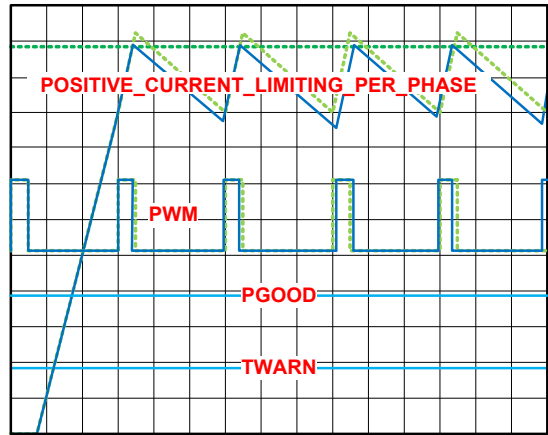


FIGURE 22. POSITIVE PEAK PHASE CURRENT LIMITING

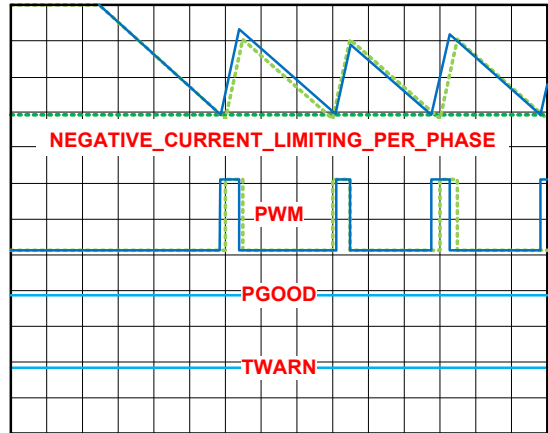


FIGURE 23. NEGATIVE PEAK PHASE CURRENT LIMITING

## Smart Power Stage OC Fault Detect

Intersil Smart Power Stage (SPS) devices will output a large signal on their IMON lines if peak current exceeds their preprogrammed threshold. (For more detail about this functionality, please refer to the relevant SPS datasheet.) The ISL68137 is equipped to detect this fault flag and immediately shut down. This detector is enabled on the GUI Overcurrent Fault setup screen.

This feature functions by detecting signals which exceed the current sense ADC full-scale range. If this detector is disabled while using an Intersil SPS, the SPS Fault# signal must be connected to the controller Enable pin of the associated rail. This will ensure that an SPS OC event will be detected and the converter will shut down.

## Thermal Monitoring and Protection

The TWARN pin indicates the temperature status of the voltage regulator. The TWARN pin is an open-drain output and an external pull-up resistor is required. This signal is valid only after the controller is enabled.

The TWARN signal can be used to inform the system that the temperature of the voltage regulator is too high and the load should reduce its power consumption. TWARN only indicates a thermal warning, not a fault.

The thermal monitoring function block diagram is shown in [Figure 24](#). The ISL68137 has 2 over-temperature thresholds, which allow both warning and fault indications. Each temperature sensor threshold can be independently programmed in the PowerNavigator™ GUI. [Figure 25 on page 20](#) shows the thermal warning to TWARN and [Figure 26 on page 20](#) shows the over-temperature fault to shutdown. PGOOD and TWARN can be configured to indicate these warning and fault thresholds via the PowerNavigator™ GUI.

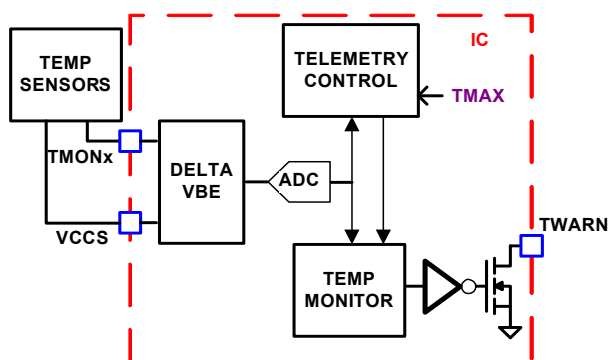


FIGURE 24. BLOCK DIAGRAM OF THERMAL MONITORING FUNCTION

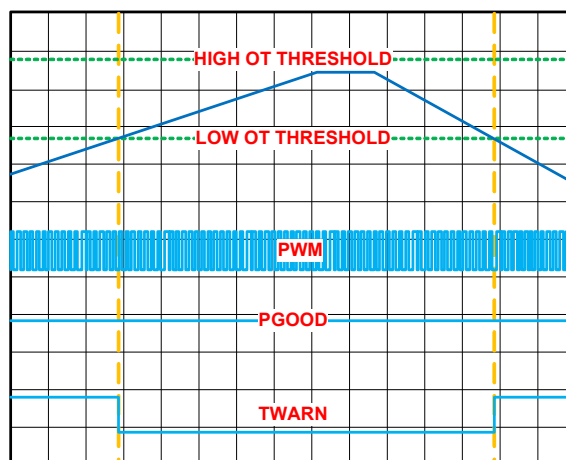


FIGURE 25. THERMAL WARNING TO TWARN

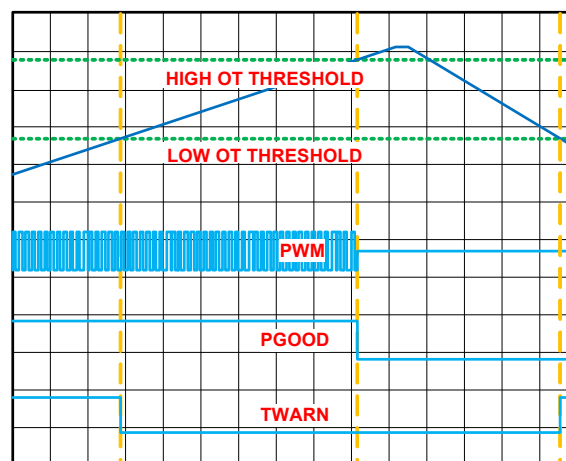


FIGURE 26. OVER-TEMPERATURE FAULT

## Layout and Design Considerations

In addition to [TB379](#), the following PCB layout and design strategies are intended to minimize the noise coupling, the impact of board parasitic impedances on converter performance and to optimize the heat dissipating capabilities of the printed circuit board. This section highlights some important practices, which should be followed during the layout process.

[Table 4](#) provides general guidance on best practices related to pin noise sensitivity. Use of good engineering judgment is required to implement designs based on criteria specific to the situation.

**TABLE 4. PIN DESIGN AND/OR LAYOUT CONSIDERATION**

PIN NAME	NOISE SENSITIVE	DESCRIPTION
VINSEN	Yes	Connects to the resistor divider between VIN and GND (see <a href="#">Figure 17 on page 17</a> ). Filter VINSEN with 10nF to GND.
RGNDx VSENx	Yes	Treat each of the remote voltage sense pairs as differential signals in the PCB layout. They should be routed side by side on the same layer. They should not be routed in proximity to noisy signals like PWM or Phase. Tie to ground when not used.
PGx	No	Open-drain. 3.3V maximum pull-up voltage. Tie to ground when not used.
SCL, SDA, SALRT	Yes	50kHz to 2MHz signal during communication, pair up with SALRT and route carefully. 20 mils spacing within SDA, SALRT and SCL; and more than 30 mils to all other signals. Refer to the SMBus design guidelines and place proper termination resistance for impedance matching. Tie to ground when not used.
AVS_CLK, AVS_SDA, AVS_MDA	Yes	Up to 50MHz signals during communication, route carefully. 20 mils spacing within CLK, SDA, MDA; and more than 30 mils to all other signals. Tie CLK and MDA to ground when not used.
TMONx	Yes	When diode sensing is utilized, VCCS is the return path for the delta Vbe currents. Utilize a separate VCCS route specifically for diode temp sense. A filter capacitor no greater than 500pF should be placed between each TMON pin and the VCCS pin near the IC. Tie to ground when not used.
TWARN	No	Open-drain. 3.3V maximum pull-up voltage.
VCC	Yes	Place at least 2.2µF MLCC decoupling capacitor directly at the pin.
VCCS	Yes	Place 4.7µF MLCC decoupling capacitor directly at the pin.
PWM	No	Avoid routing near noise sensitive analog lines such as current sense or voltage sense.

**TABLE 4. PIN DESIGN AND/OR LAYOUT CONSIDERATION (Continued)**

PIN NAME	NOISE SENSITIVE	DESCRIPTION
CSx CSRTNx	Yes	Treat each of the current sense pairs as differential signals in the PCB layout. They should be routed side by side on the same layer. They should not be routed in proximity to noisy signals like PWM or Phase. Proper routing of current sense is perhaps the most critical of all the layout tasks. Tie to ground when not used.
GND	Yes	This EPAD is the return of PWM output drivers. Use 4 or more vias to directly connect the EPAD to the power ground plane.
General Comments		The layer next to the top or bottom layer is preferred to be ground layers, while the signal layers can be sandwiched in the ground layers if possible.

## PMBus Operation

The ISL68137 PMBus slave address is pin selectable utilizing the SA pin and resistor value described in [Table 2 on page 13](#). For proper operation, users should follow the PMBus protocol, as shown in “[PMBus Protocol](#)” on page 23. The supported PMBus addresses are in 8-bit format (including write and read bit): 80-8E, A0-AE, B0-BE and C0-CE. The least significant bit of the 8-bit address is for write (0h) and read (1h). PMBus commands are in the range from 0x00h to 0xFFh. For the ISL68137, Page 0 corresponds to Output 0 and Page 1 to Output 1. For reference purposes, the 7-bit format addresses are also summarized in [Table 5](#).

TABLE 5. PMBus 8-BIT AND 7-BIT FORMAT ADDRESS (HEX)

8-BIT	7-BIT	8-BIT	7-BIT	8-BIT	7-BIT	8-BIT	7-BIT
80/81	40	A0/A1	50	B0/B1	58	C0/C1	60
82/83	41	A2/A3	51	B2/B3	59	C2/C3	61
88/89	44	A8/A9	54	B8/B9	5C	C8/C9	64
8A/8B	45	AA/AB	55	BA/BB	5D	CA/CB	65

The PMBus data formats follow PMBus Specification version 1.3 and SMBus version 2.0.

Basic PMBus telemetry commands are summarized in “[PMBus Command Summary](#)” on page 24.

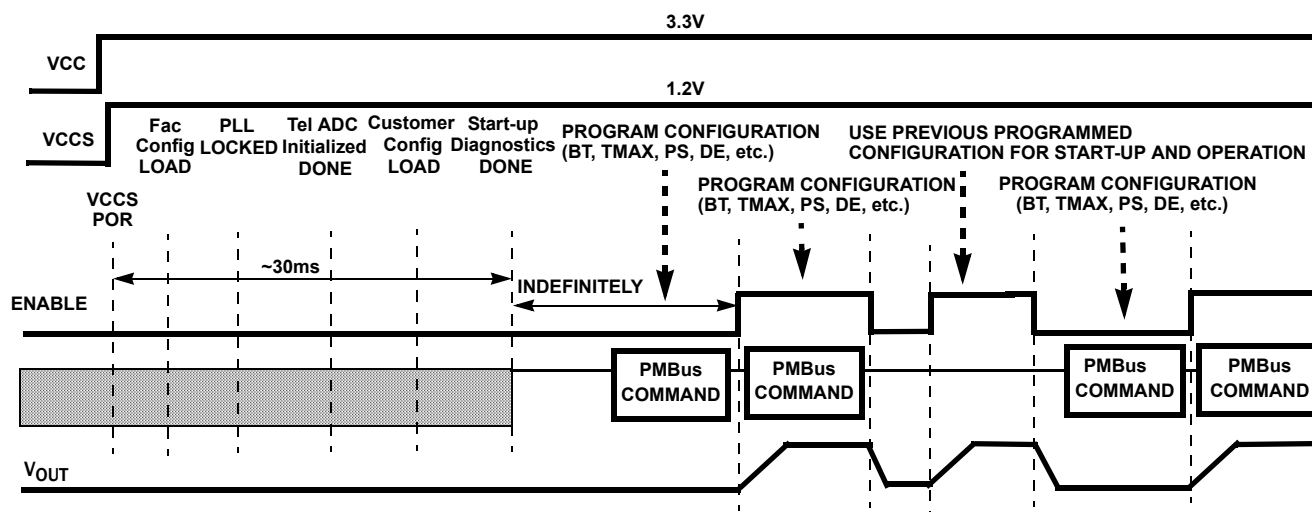
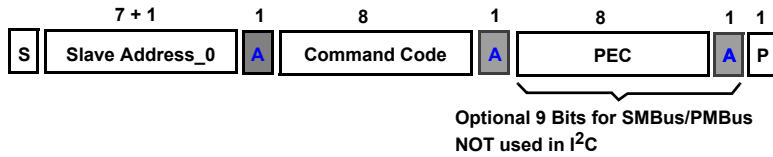


FIGURE 27. SIMPLIFIED PMBus INITIALIZATION TIMING DIAGRAM



## PMBus Protocol

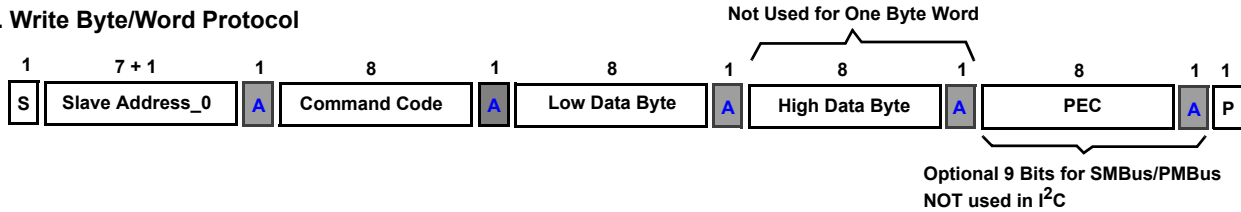
### 1. Send Byte Protocol



Example command: 03h Clear Faults  
(This will clear all of the bits in Status Byte for the selected Rail)

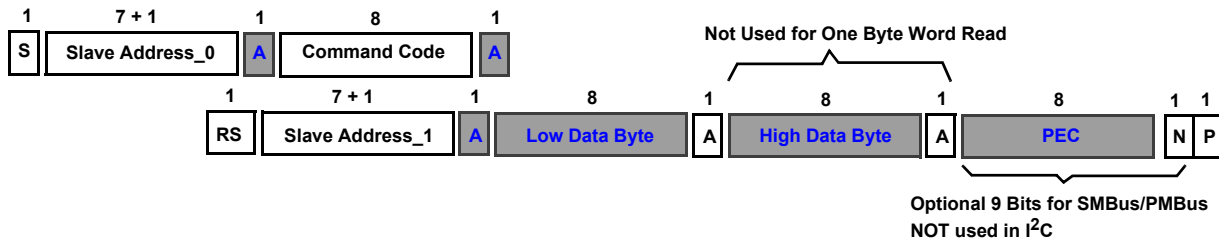
- S: Start Condition
  - A: Acknowledge ("0")
  - N: Not Acknowledge ("1")
  - W: Write ("0")
  - RS: Repeated Start Condition
  - R: Read ("1")
  - PEC: Packet Error Checking
  - P: Stop Condition
- Acknowledge or DATA from Slave, ISL68137 Controller

### 2. Write Byte/Word Protocol



Example command: 21h VOUT\_COMMAND

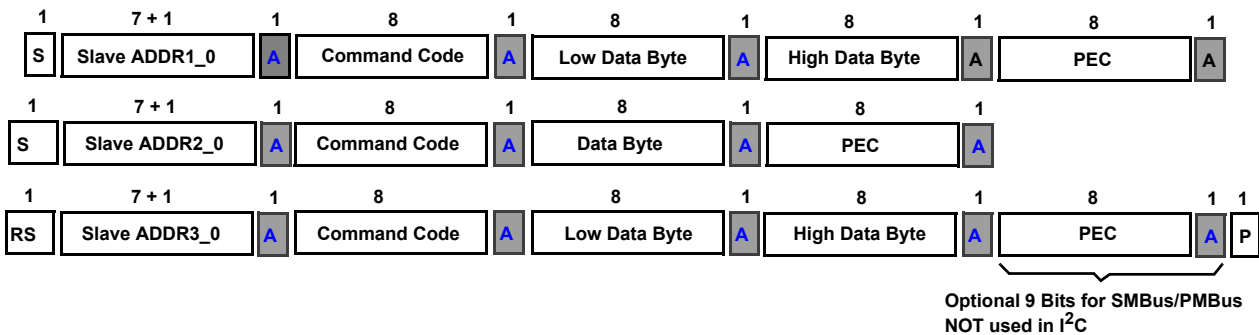
### 3. Read Byte/Word Protocol



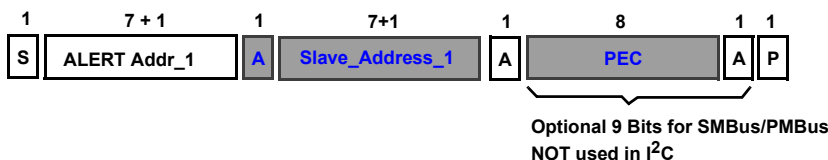
Example command: 8B READ\_VOUT (Two words, read voltage of the selected rail).

STOP (P) bit is NOT allowed before the repeated START condition when "reading" contents of a register.

### 4. Group Command Protocol - No more than one command can be sent to the same Address



### 5. Alert Response Address (ARA, 0001\_1001, 25h) for SMBus and PMBus, not used for I2C



## PMBus Command Summary

CODE	COMMAND NAME	DESCRIPTION	TYPE	DATA FORMAT	DEFAULT VALUE	DEFAULT SETTING
00h	PAGE	Selects Output 0, 1, or both	R/W	Bit	00h	Page 0
01h	OPERATION	Enable/disable, margin settings	R/W	Bit	08h	Off
02h	ON_OFF_CONFIG	On/off configuration settings	R/W	Bit	16h	ENABLE pin control
03h	CLEAR_FAULTS	Clears all fault bits in all registers and releases the SALRT pin	Write	N/A	N/A	
10h	WRITE_PROTECT	Write protection to sets of commands	R/W	Bit	00h	No write protection
20h	VOUT_MODE	Defines format for output voltage related commands	Read	Bit	40h	Direct format
21h	VOUT_COMMAND	Sets the nominal V <sub>OUT</sub> target	R/W	Direct	0384h	900mV
22h	VOUT_TRIM	Applies trim voltage to V <sub>OUT</sub> set-point	R/W	Direct	0000h	0mV
24h	VOUT_MAX	Absolute maximum voltage setting	R/W	Direct	08FCh	2300mV
25h	VOUT_MARGIN_HIGH	Sets V <sub>OUT</sub> target during margin high	R/W	Direct	0640h	1600mV
26h	VOUT_MARGIN_LOW	Sets V <sub>OUT</sub> target during margin low	R/W	Direct	00FAh	250mV
27h	VOUT_TRANSITION_RATE	Slew rate setting for V <sub>OUT</sub> changes	R/W	Direct	0064h	10mV/μs
28h	VOUT_DROOP	Sets the loadline (V/I slope) resistance for the output	R/W	Direct	0000h	0μV/A
2Bh	VOUT_MIN	Absolute minimum target voltage setting	R/W	Direct	0000h	0mV
40h	VOUT_OV_FAULT_LIMIT	Sets the V <sub>OUT</sub> overvoltage fault threshold	R/W	Direct	076Ch	1900mV
44h	VOUT_UV_FAULT_LIMIT	Sets the V <sub>OUT</sub> undervoltage fault threshold	R/W	Direct	0000h	0mV
4Fh	OT_FAULT_LIMIT	Sets the over-temperature fault threshold	R/W	Direct	007Dh	+125 °C
51h	OT_WARN_LIMIT	Sets the over-temperature warning threshold	R/W	Direct	07D0h	+2000 °C
55h	VIN_OV_FAULT_LIMIT	Sets the V <sub>IN</sub> overvoltage fault threshold	R/W	Direct	36B0h	14,000mV
59h	VIN_UV_FAULT_LIMIT	Sets the V <sub>IN</sub> undervoltage fault threshold	R/W	Direct	1F40h	8,000mV
5Bh	IIN_OC_FAULT_LIMIT	Sets the I <sub>IN</sub> overcurrent fault threshold	R/W	Direct	0032h	50A
60h	TON_DELAY	Sets the delay time from enable to V <sub>OUT</sub> rise	R/W	Direct	0014h	200μs
61h	TON_RISE	Turn-on rise time	R/W	Direct	01F4h	500μs
64h	TOFF_DELAY	Turn-off delay time	R/W	Direct	0000h	0μs
65h	TOFF_FALL	Turn-off fall time	R/W	Direct	01F4h	500μs
78h	STATUS_BYTE	First byte of STATUS_WORD	Read	Bit	N/A	N/A
79h	STATUS_WORD	Summary of critical faults	Read	Bit	N/A	N/A
7Ah	STATUS_VOUT	Reports V <sub>OUT</sub> faults	Read	Bit	N/A	N/A
7Bh	STATUS_IOUT	Reports I <sub>OUT</sub> faults	Read	Bit	N/A	N/A
7Ch	STATUS_INPUT	Reports input faults	Read	Bit	N/A	N/A
7Dh	STATUS_TEMPERATURE	Reports temperature warnings/faults	Read	Bit	N/A	N/A
7Eh	STATUS_CML	Reports communication, memory, logic errors	Read	Bit	N/A	N/A
80h	STATUS_MFR_SPECIFIC	Reports specific events	Read	Bit	N/A	N/A
88h	READ_VIN	Reports input voltage measurement	Read	Direct	N/A	N/A
89h	READ_IIN	Reports input current measurement	Read	Direct	N/A	N/A
8Bh	READ_VOUT	Reports output voltage measurement	Read	Direct	N/A	N/A
8Ch	READ_IOUT	Reports output current measurement	Read	Direct	N/A	N/A
8Dh	READ_TEMPERATURE_1	Reports power stage temperature measurement	Read	Direct	N/A	N/A

**PMBus Command Summary** (Continued)

CODE	COMMAND NAME	DESCRIPTION	TYPE	DATA FORMAT	DEFAULT VALUE	DEFAULT SETTING
8Eh	READ_TEMPERATURE_2	Reports TMON0 temperature measurement	Read	Direct	N/A	N/A
8Fh	READ_TEMPERATURE_3	Reports TMON1 temperature measurement	Read	Direct	N/A	N/A
96h	READ_POUT	Reports output power	Read	Direct	N/A	N/A
97h	READ_PIN	Reports input power	Read	Direct	N/A	N/A
98h	PMBUS_REVISION	Reports specific events	Read	Bit	33h	Revision 1.3
ADh	IC_DEVICE_ID	Reports device identification information	Read	Bit	49D22700h	ISL68137
A Eh	IC_DEVICE_REV	Reports device revision information	Read	Bit	N/A	N/A
E7h	APPLY_SETTINGS	Instructs device to apply PMBus setting changes	Write	BIT	01h	N/A
F2h	RESTORE_CONFIG	Allows selection of configurations from NVM	Write	BIT	N/A	N/A

**PMBus Use Guidelines**

All commands can be read at any time

Always disable the outputs when writing commands that change device settings. Exceptions to this rule are commands intended to be written while the device is enabled, for example, OPERATION.

**PMBus Data Formats****Direct (D)**

The Direct data format is a two byte two's complement binary integer.

**Bit Field (BIT)**

Break down of Bit Field is provided in [“PMBus Command Detail” on page 26](#).