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### 5270 (H) x 3516 (V) Full Frame CCD Image Sensor

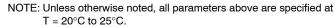
#### Description

The KAF-18500 is a dual output, high performance color CCD (charge coupled device) image sensor with 5270 (H) x 3516 (V) photoactive pixels designed for a wide range of color image sensing applications including digital imaging. Each pixel contains anti-blooming protection by means of a lateral overflow drain thereby preventing image corruption during high light level conditions. Each of the 6.8  $\mu$ m square pixels are selectively covered with red, green or blue pigmented filters for color separation. Microlenses are added for improved sensitivity.

The sensor utilizes the TRUESENSE Transparent Gate Electrode to improve sensitivity compared to the use of a standard front side illuminated polysilicon electrode.

| Parameter                                      | Typical Value                               |
|--|---|
| Architecture                                   | Full Frame CCD with Square<br>Pixels        |
| Total Number of Pixels                         | 5422 (H) x 3610 (V) = 19.6 M                |
| Number of Effective Pixels                     | 5310 (H) x 3556 (V) = 18.8 M                |
| Number of Active Pixels                        | 5270 (H) x 3516 (V) = 18.5 M                |
| Pixel Size                                     | 6.8 μm (H) x 6.8 μm (V)                     |
| Imager Size                                    | 43.1 mm (diagonal),<br>35 mm Optical format |
| Chip Size                                      | 37.8 mm (H) x 26.4 mm (V)                   |
| Aspect Ratio                                   | 3:2   |
| Saturation Signal                              | 42 ke⁻                                      |
| Charge to Voltage Conversion                   | 25 μV/e <sup>-</sup>                        |
| Quantum Efficiency (RGB)                       | 30%, 45%, 40%                               |
| Read Noise (f = 24 MHz)                        | 15.7 e⁻                                     |
| Dark Signal (T = 60°C)                         | 50 pA/cm <sup>2</sup>                       |
| Dark Current Doubling Temperature              | 5.3°C                                       |
| Linear Dynamic Range (f = 24 MHz,<br>T = 60°C) | 68.1 dB                                     |
| Charge Transfer Efficiency                     | 0.999995                                    |
| (HCTE/VCTE)                                    | 0.999998                                    |
| Blooming Protection<br>(4 ms exposure time)    | 5600 X saturation exposure                  |
| Maximum Data Rate                              | 24 MHz                                      |
| Readout Mode                                   | Dual Output Only                            |
| Package  | PGA   |
| Cover Glass                                    | AR coated (S8612)                           |

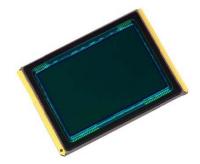
#### **Table 1. GENERAL SPECIFICATIONS**

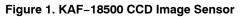




#### **ON Semiconductor®**

www.onsemi.com





#### Features

- TRUESENSE Transparent Gate Electrode for High Sensitivity
- High Resolution, 35 mm Format
- Broad Dynamic Range
- Low Noise
- Large Image Area

#### Applications

• Digital Still Cameras

#### ORDERING INFORMATION

See detailed ordering and shipping information on page 2 of this data sheet.

#### **ORDERING INFORMATION**

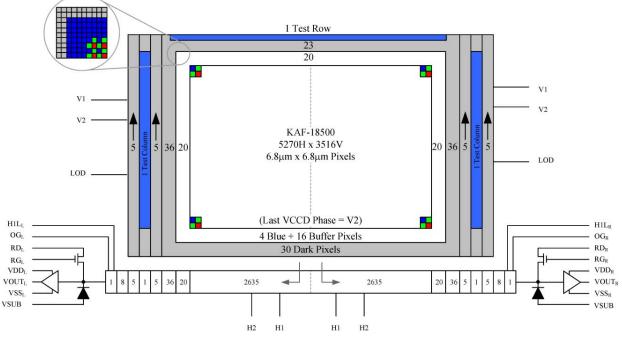
#### Table 2. ORDERING INFORMATION

| Part Number            | Description  | Marking Code     |
|------------------------|--|------------------|
| KAF-18500-NXA-JH-AA-08 | Special Color, Aperture, Enhanced, ESD, LOD, Microlens,<br>Sealed IR Cover Glass, 0.8 mm glass                     | KAF-18500-NXA-08 |
| KAF-18500-NXA-JH-AE-08 | Special Color, Aperture, Enhanced, ESD, LOD, Microlens,<br>Sealed IR Cover Glass, 0.8 mm glass [Engineering Grade] |                  |

See the ON Semiconductor *Device Nomenclature* document (TND310/D) for a full description of the naming convention used for image sensors. For reference documentation, including information on evaluation kits, please visit our web site at www.onsemi.com.

#### **DEVICE DESCRIPTION**

#### Architecture



2711 Pixels/Line/Output



#### **Dark Reference Pixels**

Surrounding the periphery of the device is a border of light shielded pixels creating a dark region. Within this dark region, exist light shielded pixels that include 36 leading dark pixels on every line. There are also 30 full dark lines at the start and 23 full dark lines at the end of every frame. Under normal circumstances, these pixels do not respond to light and may be used as a dark reference.

#### **Dummy Pixels**

Within each horizontal shift register there are 20 leading additional shift phases required before the dark reference pixels: (1 + 8 + 5 + 1 + 5) (See Figure 2). These pixels are designated as dummy pixels and should not be used to determine a dark reference level.

#### **Active Buffer Pixels**

Forming the outer boundary of the effective active pixel region, there are 20 unshielded active buffer pixels between the photoactive area and the dark reference. These pixels are light sensitive but they are not tested for defects and non–uniformities. For the leading 20 active column pixels, the first 4 pixels are covered with blue pigment while the remaining are arranged in a Bayer pattern (R, GR, GB, B).

#### Image Acquisition

An electronic representation of an image is formed when incident photons falling on the sensor plane create electron-hole within device. pairs the These photon-induced electrons are collected locally by the formation of potential wells at each photogate or pixel site. The number of electrons collected is linearly dependent on light level and exposure time and non-linearly dependent on wavelength. When the pixel's capacity is reached, excess electrons are discharged into the lateral overflow drain to prevent crosstalk or 'blooming'. During the integration period, the V1 and V2 register clocks are held at a constant (low) level.

#### **Charge Transport**

The integrated charge from each photogate (pixel) is transported to the output using a two-step process. Each line (row) of charge is first transported from the vertical CCD's to a horizontal CCD register using the V1 and V2 register clocks. The horizontal CCD is presented with a new line on the falling edge of V2 while H1 is held high. The horizontal CCD's then transport each line, pixel by pixel, to the output structure by alternately clocking the H1 and H2 pins in a complementary fashion. A separate connection to the last H1 phase (H1L) is provided to improve the transfer speed of charge to the floating diffusion output amplifier. On each falling edge of H1L a new charge packet is dumped onto a floating diffusion and sensed by the output amplifier.

#### HORIZONTAL REGISTER

#### **Output Structure**

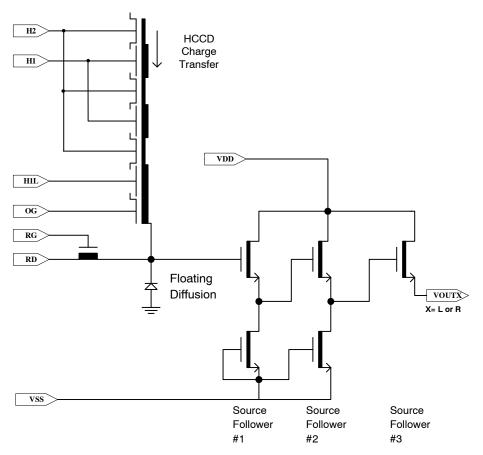
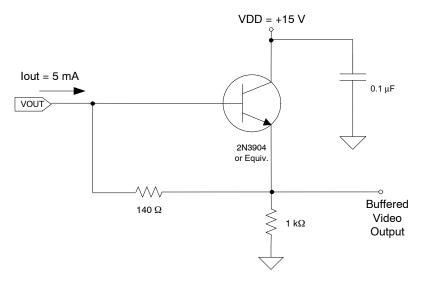


Figure 3. Output Architecture (Left or Right)

The output consists of a floating diffusion capacitance connected to a three-stage source follower. Charge presented to the floating diffusion (FD) is converted into a voltage and is current amplified in order to drive off-chip loads. The resulting voltage change seen at the output is linearly related to the amount of charge placed on the FD. Once the signal has been sampled by the system electronics, the reset gate (RG) is clocked to remove the signal and FD is reset to the potential applied by reset drain (RD). Increased signal at the floating diffusion reduces the voltage seen at the output pin. To activate the output structures, an off-chip current source must be added to the VOUT pins of the device. See Figure 4.

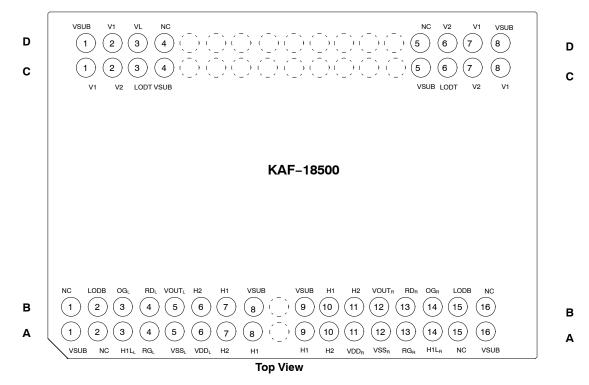


Note: Component values may be revised based on operating conditions and other design considerations.

Figure 4. Typical Output Structure Load Diagram

#### PHYSICAL DESCRIPTION

#### Pin Description and Device Orientation





| Pin | Name             | Description                                |  |  |  |
|-----|------------------|--|--|--|--|
| 1   | VSUB             | Substrate                                  |  |  |  |
| 2   | NC               | Physical pin with no connection on die     |  |  |  |
| 3   | H1LL             | Horizontal Phase 1, last phase, left side  |  |  |  |
| 4   | RGL              | Reset Drain, left side                     |  |  |  |
| 5   | VSSL             | Output Amplifier Return, left side         |  |  |  |
| 6   | VDDL             | Output Amplifier Supply, left side         |  |  |  |
| 7   | H2               | Horizontal Phase 2                         |  |  |  |
| 8   | H1               | Horizontal Phase 1                         |  |  |  |
| 9   | H1               | Horizontal Phase 1                         |  |  |  |
| 10  | H2               | Horizontal Phase 2                         |  |  |  |
| 11  | VDD <sub>R</sub> | Output Amplifier Supply, right side        |  |  |  |
| 12  | VSS <sub>R</sub> | Output Amplifier Return, right side        |  |  |  |
| 13  | RG <sub>R</sub>  | Reset Gate, right side                     |  |  |  |
| 14  | H1L <sub>R</sub> | Horizontal Phase 1, last phase, right side |  |  |  |
| 15  | NC               | Physical pin with no connection on die     |  |  |  |
| 16  | VSUB             | Substrate                                  |  |  |  |

#### Table 3. PGA GRID ROW A

| Table 4. | PGA GRID ROV | VΒ |
|----------|--------------|----|
|          |              |    |

| Pin | Name              | Description                            |  |  |  |
|-----|-------------------|--|--|--|--|
| 1   | NC                | Physical pin with no connection on die |  |  |  |
| 2   | LODB              | Lateral Overflow Drain, bottom         |  |  |  |
| 3   | OGL               | Output Gate, left side                 |  |  |  |
| 4   | $RD_L$            | Reset Gate, left side                  |  |  |  |
| 5   | VOUTL             | Video Output, left side                |  |  |  |
| 6   | H2                | Horizontal Phase 2                     |  |  |  |
| 7   | H1                | Horizontal Phase 1                     |  |  |  |
| 8   | VSUB              | Substrate                              |  |  |  |
| 9   | VSUB              | Substrate                              |  |  |  |
| 10  | H1                | Horizontal Phase 1                     |  |  |  |
| 11  | H2                | Horizontal Phase 2                     |  |  |  |
| 12  | VOUT <sub>R</sub> | Video Output, right side               |  |  |  |
| 13  | RD <sub>R</sub>   | Reset Drain, right side                |  |  |  |
| 14  | OG <sub>R</sub>   | Output Gate, right side                |  |  |  |
| 15  | LODB              | Lateral Overflow Drain, bottom         |  |  |  |
| 16  | NC                | Physical pin with no connection on die |  |  |  |

#### Table 5. PGA GRID ROW C

| Pin | Name | Description                 |  |
|-----|------|-----------------------------|--|
| 1   | V1   | Vertical Phase 1            |  |
| 2   | V2   | Vertical Phase 2            |  |
| 3   | LODT | Lateral Overflow Drain, top |  |
| 4   | VSUB | Substrate                   |  |
| 5   | VSUB | Substrate                   |  |
| 6   | LODT | Lateral Overflow Drain, top |  |
| 7   | V2   | Vertical Phase 2            |  |
| 8   | V1   | Vertical Phase 1            |  |

#### Table 6. PGA GRID ROW D

| Pin | Name | Description                            |  |  |  |
|-----|------|--|--|--|--|
| 1   | VSUB | Substrate                              |  |  |  |
| 2   | V1   | Vertical Phase 1                       |  |  |  |
| 3   | V2   | Vertical Phase 2                       |  |  |  |
| 4   | NC   | Physical pin with no connection on die |  |  |  |
| 5   | NC   | Physical pin with no connection on die |  |  |  |
| 6   | V2   | Vertical Phase 2                       |  |  |  |
| 7   | V1   | Vertical Phase 1                       |  |  |  |
| 8   | VSUB | Substrate                              |  |  |  |

#### **IMAGING PERFORMANCE**

#### Table 7. TYPICAL OPERATIONAL CONDITIONS

| Description   | Condition   | Notes                     |
|---|---|---------------------------|
| Frame time (t <sub>readout</sub> + t <sub>int</sub> ) | Varies, see below   | Includes overclock pixels |
| Readout time (t <sub>readout</sub> )                  | 527 ms  |                           |
| Integration time (t <sub>int</sub> )                  | Varies per test: Bright Field 250 ms, Dark Field 1 sec,<br>Saturation 250 ms, Low light 33 ms |                           |
| Horizontal clock frequency                            | 24 MHz  |                           |
| Temperature   | 20 – 25°C   | Room temperature          |
| Mode  | integrate – readout cycle   |                           |
| Operation   | Nominal operating voltages and timing with min. vertical pulse width $t_{Vw}$ = 11 $\mu s$    |                           |

#### **Table 8. SPECIFICATIONS**

| Description                                       | Symbol                            | Min            | Nom.                  | Max | Units             | Notes  | Verification Plan   |
|---|-----------------------------------|----------------|-----------------------|-----|-------------------|--------|---|
| Saturation Signal                                 | Vsat<br>Ne⁻ <sub>sat</sub><br>Q/V | 900<br>(35000) | 1086<br>42000<br>25.6 |     | mV<br>e⁻<br>μV/e⁻ | 1, 19  | die <sup>17</sup><br>design <sup>18</sup><br>design <sup>18</sup> |
| Peak Quantum Efficiency red<br>green<br>blue      | Rr<br>Rg<br>Rb                    |                | 30<br>45<br>40        |     | %                 |        | design <sup>18</sup>  |
| High Level Photoresponse<br>Non-Linearity         | Le_High                           |                | 2                     | 10  | %                 | 2      | die <sup>17</sup>   |
| Low Level Photoresponse<br>Non-Linearity          | Le_Low                            |                | 2                     | 10  | %                 | 2      | die <sup>17</sup>   |
| Photo Response Non-Uniformity<br>each color plane | PRNU                              |                | 4.5                   | 25  | %р–р              | 3      | die <sup>17</sup>   |
| Integration Dark Signal                           | Vdark, int                        |                | 4                     | 10  | mV/s              | 4, 16  | die <sup>17</sup>   |
| Readout Dark Signal                               | Vdark, read                       |                | 12                    | 20  | mV/s              | 15, 16 | die <sup>17</sup>   |
| Dark Signal Non–Uniformity                        | DSNU                              |                | 0.5                   | 4   | mV p–p            | 5      | die <sup>17</sup>   |
| Dark Signal Doubling Temperature                  | ΔΤ                                |                | 5.3                   |     | °C                |        | design <sup>18</sup>  |
| Read Noise  | N <sub>R</sub>                    |                | 15.7                  |     | e⁻ rms            |        | design <sup>18</sup>  |
| Total Noise                                       | N                                 |                | 18.9                  |     | e⁻ rms            | 6      | design <sup>18</sup>  |
| Linear Dynamic Range                              | DR                                |                | 68.1                  |     | dB                | 7      | design <sup>18</sup>  |
| Red–Green Hue Shift<br>Blue–Green Hue Shift       | RGHueUnif<br>BGHueUnif            |                | 1.8                   | 12  | %                 | 8      | die <sup>17</sup>   |
| Horizontal Charge Transfer Efficiency             | HCTE                              | 0.999995       | 0.999995              |     |                   | 9      | die <sup>17</sup>   |
| Vertical Charge Transfer Efficiency               | VCTE                              | 0.999999       | 0.999999              |     |                   |        | die <sup>17</sup>   |
| Blooming Protection                               | Xab                               |                | 5600                  |     | x Vsat            | 10     | design <sup>18</sup>  |
| DC Offset, output amplifier                       | Vodc                              | 6.0            | 8                     | 9.5 | V                 | 11     | die <sup>17</sup>   |
| Output Amplifier Bandwidth                        | f <sub>–3dB</sub>                 |                | 232                   |     | MHz               | 12     | design <sup>18</sup>  |
| Output Impedance, Amplifier                       | R <sub>OUT</sub>                  | 100            | 137                   | 300 | Ω                 |        | die <sup>17</sup>   |
| Hclk Feedthru                                     | V <sub>hft</sub>                  |                | 3.7                   | 20  | mV                | 13     | die <sup>17</sup>   |
| Reset Feedthru                                    | V <sub>rft</sub>                  |                | 0.5                   |     | V                 | 14     | design <sup>18</sup>  |

1. Increasing output load currents to improve bandwidth will decrease the conversion factor (Q/V).

2. Worst case deviation (from 10 mV to Vsat min), relative to a linear fit applied between 0 and 85% of Vsat min.

3. Difference between the maximum and minimum average signal levels of 148 x 148 blocks within the sensor on a per color basis as a % of average signal level.

4.  $T = 60^{\circ}$ C. Average non-illuminated signal with respect to over-clocked vertical register signal.

5. T = 60°C. Absolute difference between the maximum and minimum average signal levels of 148 x 148 blocks within the sensor.

6. rms deviation of a multi-sampled pixel measured in the dark including amplifier and system noise sources.

7. 20log (0.95 \* Vsat/VN). Specified at T = 60°C.

8. Gradual variations in hue (red with respect to green pixels and blue with respect to green pixels) in regions of interest (148 x 148 blocks) within the sensor.

9. Measured per transfer at Vsat min. Typically, no degradation in CTE is observed up to 24 MHz.

10. Xab is the number of times above the Vsat illumination level that the sensor will bloom by spot size doubling. The spot size is 10% of the imager height. Xab is measured at 4 ms.

11. Video level offset with respect to ground.

12. Last stage only. Assumes 5 pF off-chip load.

13. Amount of artificial signal due to H1 coupling.

14. Amplitude of feedthrough pulse in VOUT due to RG coupling.

15. T = 60°C. Average non-illuminated signal collected due to the read out time.

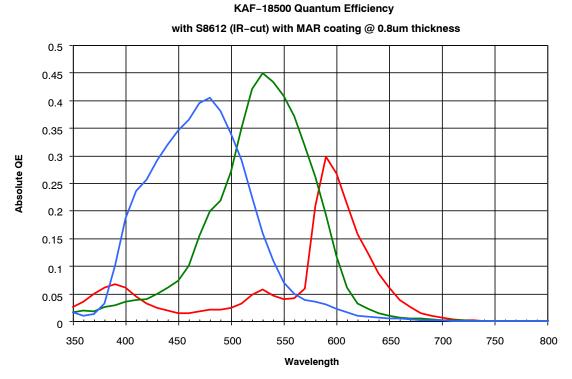
16. Total dark signal = (Vdark,int x t<sub>int</sub>) + (Vdark,read x t<sub>readout</sub>).

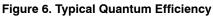
17. A parameter that is measured on every sensor during production testing.

18. A parameter that is quantified during the design verification activity.

19. Specified at  $T = 60^{\circ}$ C.

#### **TYPICAL PERFORMANCE CURVES**





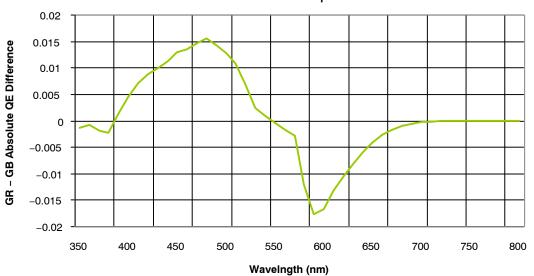


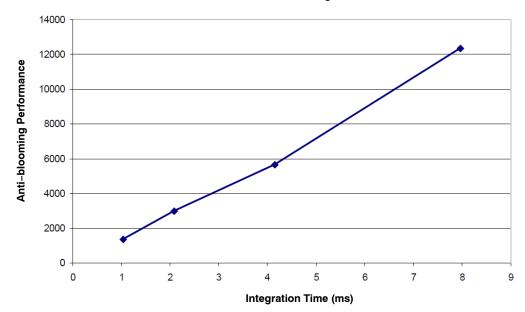


Figure 7. Typical GR–GB QE Difference

0.9 0.8 Normalized QE 0.7 0.6 0.5 0.4 <del>0.3</del> -LB 0.2 · CC 0 - RT -40 -35 -30 -25 -20 -15 -10 -5 0 5 10 15 20 25 30 35 40 Slash Angle

Angle QE about diagonal (BLUE light BLUE pixel)

Figure 8. Typical Normalized Angle QE



KAF-18500 Anti-blooming Performance

Figure 9. Typical Anti Blooming Performance

#### **DEFECT DEFINITIONS**

#### **Operating Conditions**

All defect tests performed at:

#### **Table 9. OPERATING CONDITIONS**

| Description                | Condition   | Notes            |
|----------------------------|---|------------------|
| Integration time (tint)    | Varies per test: Bright Field 250 ms, Dark Field 1 sec, Saturation 250 ms,<br>Low light 33 ms |                  |
| Horizontal clock frequency | 24 MHz  |                  |
| Temperature                | 20 – 25°C   | Room temperature |

#### **Table 10. SPECIFICATIONS**

| Classification | Points | Clusters, small<br>and large | Clusters, large | Columns | Includes Dead Columns |
|----------------|--------|------------------------------|-----------------|---------|-----------------------|
| Standard Grade | ≤ 4400 | ≤ 50                         | ≤5              | ≤ 15    | yes                   |

#### Point Defects

A pixel that deviates by more than 9 mV above neighboring pixels under non-illuminated conditions.

-or-

A pixel that deviates by more than 7% above or 11% below neighboring pixels under illuminated conditions.

#### Cluster Defect

Small clusters: A grouping of adjacent point defects that can number in size from 2 to 10 pixels.

Large clusters: A grouping of more than 10 pixels but not larger than 20 adjacent point defects. A single large cluster is not to exceed 5 adjacent pixels within the same color plane.

#### **Cluster Separation**

Cluster defects are separated by no less than 4 good pixels in any direction.

#### Column Defect

A grouping of more than 10 point defects along a single column

-or-

A column that deviates by more than 0.9 mV above or below neighboring columns under non-illuminated conditions.

-or-

A column that deviates by more than 1.5% above or below neighboring columns under illuminated conditions.

Column and cluster defects are separated by at least 4 good columns in the x direction. No multiple column defects (double or more) will be permitted.

#### Dead Columns

A column that deviates by more than 50% below neighboring columns under illuminated conditions.

#### Saturated Columns

A column that deviates by more than 100 mV above neighboring columns under non–illuminated conditions. No saturated columns are allowed.

#### OPERATION

#### Table 11. ABSOLUTE MAXIMUM RATINGS

| Description           | Symbol             | Minimum | Maximum | Units | Notes |
|-----------------------|--------------------|---------|---------|-------|-------|
| Diode Pin Voltages    | V <sub>diode</sub> | -0.5    | +17.5   | V     | 1, 2  |
| Gate Pin Voltages     | V <sub>gate1</sub> | -13.5   | +13.5   | V     | 1, 3  |
| Gate-Gate Voltages    | V <sub>1-2</sub>   | -13.5   | +13.5   | V     | 4, 5  |
| Output Bias Current   | l <sub>out</sub>   |         | -30     | mA    | 6     |
| LOD Diode Voltage     | V <sub>LODT</sub>  | -0.5    | +13.0   | V     | 7     |
| Operating Temperature | T <sub>OP</sub>    | 0       | 60      | °C    | 8     |

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Referenced to pin SUB.

2. Includes pins: RD, VDD, VSS, VOUT.

3. Includes pins: V1, V2, H1, H1L, H2, RG, OG.

Voltage difference between overlapping gates. Includes: V1 to V2; H1, H1L to H2; H1L to OG; V1 to H2. These inputs contain an ESD
protection circuit. Exceeding the maximum voltages will cause an uncontrolled current to flow in these circuits and may damage the input
pin.

5. Voltage difference between non-overlapping gates. Includes: V1 to H1, H1L; V2, OG to H2. These inputs contain an ESD protection circuit. Exceeding the maximum voltages will cause an uncontrolled current to flow in these circuits and may damage the input pin.

 Avoid shorting output pins to ground or any low impedance source during operation. Amplifier bandwidth increases at higher currents and lower load capacitance at the expense of reduced gain (sensitivity). Operation at the maximum values will reduce Mean Time to Failure (MTTF).

7. V1, H1, V2, H2, H1L, OG, and RD are tied to 0 V.

8. Noise performance will degrade at higher temperatures.

9. Absolute maximum rating is defined as a level or condition that should not be exceeded at any time per the description. If the level or condition is exceeded, the device will be degraded and may be damaged.

#### Power-up Sequence

The sequence chosen to perform an initial power-up is not critical for device reliability. A coordinated sequence may minimize noise and the following sequence is recommended:

- 1. Connect the ground pins (SUB).
- 2. Supply the appropriate biases and clocks to the remaining pins.

| Description             | Symbol           | Minimum | Nominal | Maximum | Units | Maximum DC<br>Current (mA)         | Notes |
|-------------------------|------------------|---------|---------|---------|-------|------------------------------------|-------|
| Reset Drain             | RD               | 11.3    | 11.5    | 11.7    | V     | I <sub>RD</sub> = 0.01             |       |
| Output Amplifier Return | VSS              | 0.5     | 0.7     | 1.0     | V     | I <sub>SS</sub> = 3.0              |       |
| Output Amplifier Supply | VDD              | 14.5    | 15.0    | 15.5    | V     | I <sub>OUT</sub> + I <sub>SS</sub> |       |
| Substrate               | SUB              |         | 0       |         | V     | 0.01                               |       |
| Output Gate             | OG               | -2.2    | -2.0    | -1.8    | V     | 0.01                               |       |
| Lateral Overflow Drain  | LOD              | 9.8     | 10.0    | 10.2    | V     | 0.01                               |       |
| Video Output Current    | I <sub>OUT</sub> |         | -5      | -10     | mA    |                                    | 1     |

#### Table 12. DC BIAS OPERATING CONDITIONS

1. An output load sink must be applied to VOUT to activate output amplifier – see Figure 4.

#### **AC Operating Conditions**

#### Table 13. CLOCK LEVELS

| Description        | Symbol         | Level     | Minimum | Nominal | Maximum | Units | Effective<br>Capacitance | Notes |
|--------------------|----------------|-----------|---------|---------|---------|-------|--------------------------|-------|
| V1 Low Level       | V1L            | Low       | -9.2    | -9.0    | -8.8    | V     | 245 nF                   | 1, 2  |
| V1 High Level      | V1H            | High      | 2.3     | 2.5     | 2.7     | V     | 245 nF                   | 1, 2  |
| V2 Low Level       | V2L            | Low       | -9.2    | -9.0    | -8.8    | V     | 303 nF                   | 1, 2  |
| V2 High Level      | V2H            | High      | 2.3     | 2.5     | 2.7     | V     | 303 nF                   | 1, 2  |
| H1, H2 (amplitude) | H1amp<br>H2amp | Amplitude | 6.5     | 6.75    | 7.0     | V     | See below                |       |
| H1 Low Level       | H1Low          | Low       | -4.7    | -4.5    | -4.3    | V     | 460 pF                   | 1     |
| H2 Low Level       | H2Low          | Low       | -5.2    | -5.0    | -4.8    | V     | 302 pF                   | 1     |
| H1L Low Level      | H1Llow         | Low       | -6.7    | -6.5    | -6.3    | V     | 15 pF                    | 1     |
| H1L High Level     | H1Lhigh        | High      | 1.3     | 1.5     | 1.7     | V     | 15 pF                    | 1     |
| RG Low Level       | RGL            | Low       | 0.3     | 0.5     | 0.7     | V     | 21 pF                    | 1     |
| RG High Level      | RGH            | High      | 7.8     | 8.0     | 8.2     | V     | 21 pF                    | 1     |

All pins draw less than 10 μA DC current. Capacitance values relative to SUB (substrate).
 Clock capacitance is the effective capacitance extrapolated from the rise and fall time measured while operating the sensor.

#### TIMING

#### **Table 14. REQUIREMENTS AND CHARACTERISTICS**

| Description                  | Symbol                              | Minimum | Nominal | Maximum | Units | Notes |
|------------------------------|-------------------------------------|---------|---------|---------|-------|-------|
| H1, H2 Clock Frequency       | f <sub>H</sub>                      |         |         | 24      | MHz   | 1, 2  |
| V1, V2 Clock Frequency       | f <sub>V</sub>                      |         |         | 45.5    | kHz   | 1, 2  |
| H1, H2 Rise, Fall Times      | t <sub>H1r</sub> , t <sub>H1f</sub> | 5       |         | 10      | %     | 3, 7  |
| V1, V2 Rise, Fall Times      | t <sub>V1r</sub> , t <sub>V1f</sub> | 5       |         | 10      | %     | 3     |
| V1 – V2 Cross–over           | V <sub>VCR</sub>                    | 1       |         |         | V     |       |
| H1 – H2 Cross–over           | V <sub>HCR</sub>                    | -3.0    | -1.5    | 0       | V     |       |
| H1L Rise – H2 Fall Crossover | V <sub>H1LCR</sub>                  | -2.0    |         | 1.0     | V     | 9     |
| H1, H2 Setup Time            | t <sub>HS</sub>                     | 1       | 5       |         | μs    |       |
| RG Clock Pulse Width         | t <sub>RGw</sub>                    | 5       |         |         | ns    | 4     |
| RG Rise, Fall Times          | t <sub>RGr</sub> , t <sub>RGf</sub> | 5       |         | 10      | %     | 3     |
| V1, V2 Clock Pulse Width     | t <sub>Vw</sub>                     | 11      |         |         | μs    | 2, 6  |
| Flush Clock Off Time         | t <sub>off</sub>                    | 4       |         |         | μs    | 2, 6  |
| Pixel Period (1 Count)       | t <sub>e</sub>                      | 42      |         |         | ns    | 2     |
| H1L - VOUT Delay             | t <sub>HV</sub>                     |         | 5       |         | ns    |       |
| RG - VOUT Delay              | t <sub>RV</sub>                     |         | 5       |         | ns    |       |
| Readout Time                 | t <sub>readout</sub>                | 505     |         |         | ms    | 6, 8  |
| Integration Time             | t <sub>int</sub>                    |         |         |         |       | 5, 6  |
| Line Time                    | t <sub>line</sub>                   | 140     |         |         | μs    | 6     |
| Fast Flush Time              | t <sub>flush</sub>                  | 88      |         |         | ms    |       |

1. 50% duty cycle values.

CTE will degrade above the nominal frequency.
 Relative to the pulse width (based on 50% of high/low levels).

RG should be clocked continuously.
 Integration time is user specified.

6. Longer times will degrade noise performance.

7. The maximum specification or 10 nsec whichever is greater based on the frequency of the horizontal clocks.
8. t<sub>readout</sub> = t<sub>line</sub> \* 3610 lines
9. The charge capacity near the output could be degraded if the voltage at the clock cross over point is outside this range.

#### **Edge Alignment**

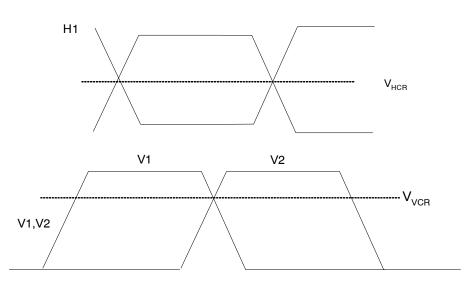
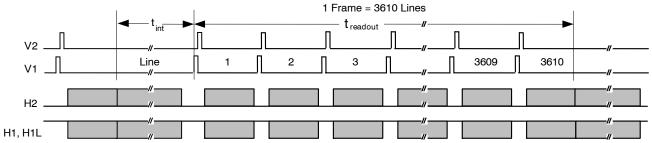


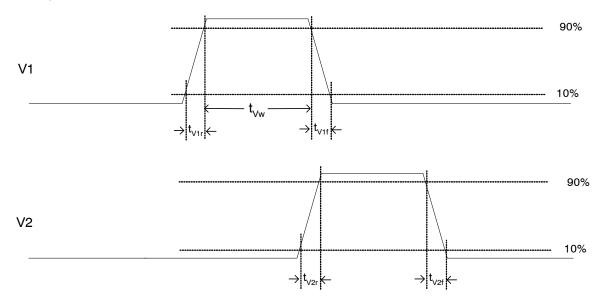
Figure 10. Timing Edge Alignment



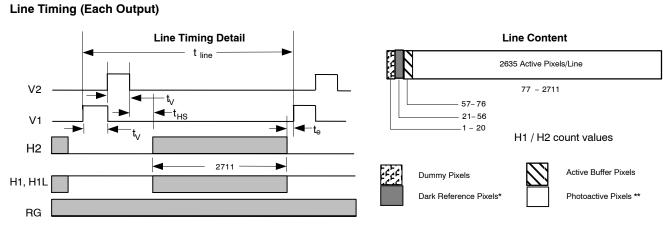
Frame Timing Detail













#### **Pixel Timing**

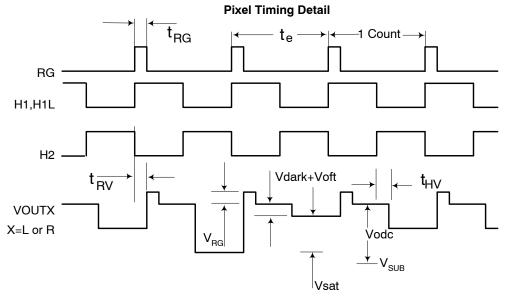
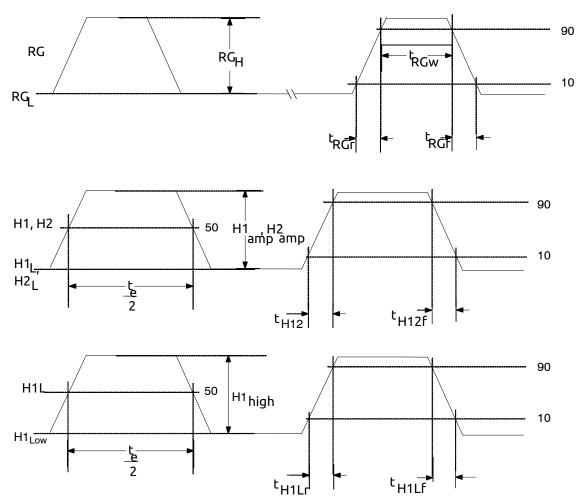


Figure 14. Pixel Timing

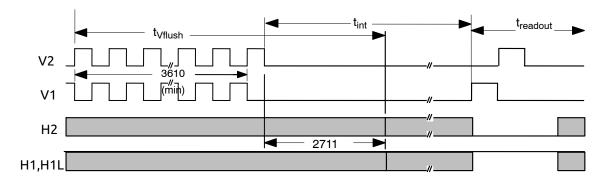
**Pixel Timing Detail** 

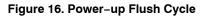




#### MODE OF OPERATION

Power-up Flush Cycle





#### STORAGE AND HANDLING

For information on ESD and cover glass care and cleanliness, please download the *Image Sensor Handling and Best Practices* Application Note (AN52561/D) from www.onsemi.com.

For information on soldering recommendations, please download the Soldering and Mounting Techniques Reference Manual (SOLDERRM/D) from www.onsemi.com. For quality and reliability information, please download the *Quality & Reliability* Handbook (HBD851/D) from www.onsemi.com.

For information on device numbering and ordering codes, please download the *Device Nomenclature* technical note (TND310/D) from <u>www.onsemi.com</u>.

For information on Standard terms and Conditions of Sale, please download <u>Terms and Conditions</u> from <u>www.onsemi.com</u>.

#### **MECHANICAL INFORMATION**

#### **Completed Assembly**

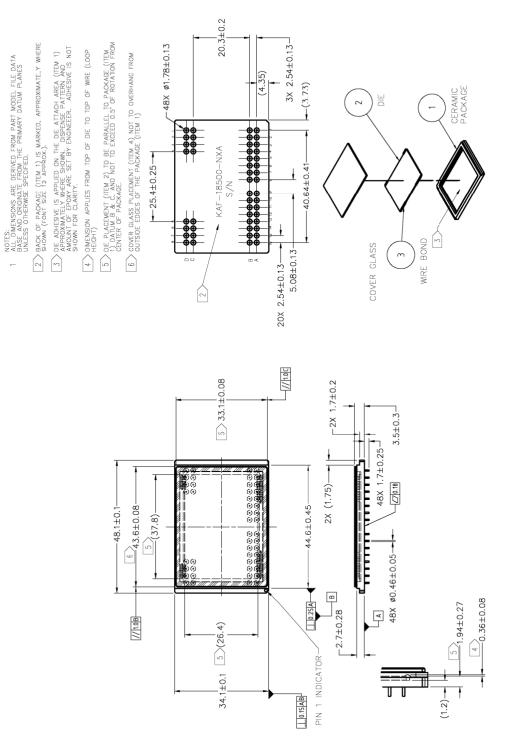
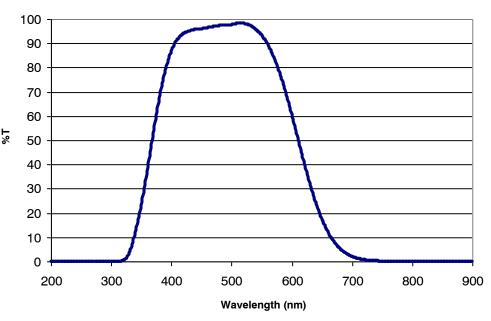


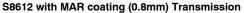
Figure 17. Completed Assembly Drawing

#### **Cover Glass Specification**

MAR Coated-IR Absorbing Cover Glass

- 1. Dust/Scratch/Digs/Defects: 20 micron max
- 2. Substrate material: Schott S8612 whose performance is as published and specifications controlled by Schott, North America. The cover glass supplied for this device is as shown in Figure 17. Data supplied in the graph below is a typical transmission of the AR coated material at 0.8 mm thickness.







3. Multilayer anti-reflective coating on two sides: Two-sided reflectance:

#### Table 15.

| Wavelength   | Transmission |  |  |
|--------------|--------------|--|--|
| 420 – 450 nm | < 2%         |  |  |
| 450 – 630 nm | < 1%         |  |  |
| 630 – 680 nm | < 2%         |  |  |

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