

Chipsmall Limited consists of a professional team with an average of over 10 year of expertise in the distribution of electronic components. Based in Hongkong, we have already established firm and mutual-benefit business relationships with customers from, Europe, America and south Asia, supplying obsolete and hard-to-find components to meet their specific needs.

With the principle of "Quality Parts, Customers Priority, Honest Operation, and Considerate Service", our business mainly focus on the distribution of electronic components. Line cards we deal with include Microchip, ALPS, ROHM, Xilinx, Pulse, ON, Everlight and Freescale. Main products comprise IC, Modules, Potentiometer, IC Socket, Relay, Connector. Our parts cover such applications as commercial, industrial, and automotives areas.

We are looking forward to setting up business relationship with you and hope to provide you with the best service and solution. Let us make a better world for our industry!



# Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

Email & Skype: info@chipsmall.com Web: www.chipsmall.com

Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China







#### **Monolithic Linear IC**

# 2-Channel Power Amplifier



http://onsemi.com

#### Overview

The LA4814JA buili-in the power amplifier circuit capable of low-voltage (2.7V and up) operation and has additionally a standby function to reduce the current drain. It is a power amplifier IC optimal for speaker drive used in battery-driven portable equipment and other such products.

# **Applications**

Mini radio cassette players/recorders, portable radios, transceivers and other portable audio devices

#### **Features**

• On-chip 2-channel power amplifier

Output power 1 = 350mW typ. ( $V_{CC} = 5.0V$ ,  $R_L = 4\Omega$ , THD = 10%)

Output power 2 = 150mW typ. ( $V_{CC} = 3.6V$ ,  $R_{L} = 4\Omega$ , THD = 10%)

• Enables monaural BTL output system by changing externally connected components

Output power 3 = 700 mW typ. (V<sub>CC</sub> = 5.0 V, R<sub>L</sub> =  $8 \Omega$ , THD = 10 %)

Output power 4 = 320 mW typ. (V<sub>CC</sub> = 3.6V, R<sub>L</sub> =  $8\Omega$ , THD = 10%)

Low-voltage operation possible

 $V_{CC} = 2.7V$  and up

• Standby function

Current drain at standby =  $0.1\mu A$  typ. ( $V_{CC} = 5V$ )

• Voltage gain setting possible

Voltage gain = 3 to 20dB

• Second amplifier stop control function

Reducing the pop noise at startup (in BTL mode)

# **Specifications**

# **Maximum Ratings** at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V <sub>CC</sub> max		8	٧
Allowable power dissipation	Pd max	*	1.3	W
Maximum junction temperature	Tj max		150	°C
Operating temperature	Topr		-40 to +85	°C
Storage temperature	Tstg		-40 to +150	°C

 $<sup>^{*}</sup>$  Mounted on Our evaluation board : Double-sided board with dimensions of  $60 \text{mm} \times 60 \text{mm} \times 1.6 \text{mm}$ 

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

## **Operating Conditions** at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	V <sub>CC</sub>		5	٧
Recommended load resistance	RL	Single ended mode	4 to 32	Ω
		BTL mode	8 to 32	Ω
Operating supply voltage range	V <sub>CC</sub> op	Single ended mode, $R_L = 6$ to $32\Omega$	2.7 to 7	٧
		Single ended mode, $R_L = 4$ to $6\Omega$	2.7 to 5.5	V
		BTL mode, $R_L = 16$ to $32\Omega$	2.7 to 7	٧
		BTL mode, $R_L = 8$ to $16\Omega$	2.7 to 5.5	V

<sup>\*</sup> Determine the supply voltage to be used with due consideration of allowable power dissipation.

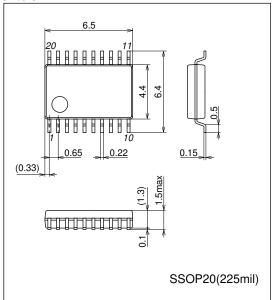
# **Electrical Characteristics** at $Ta = 25^{\circ}C$ , $V_{CC} = 5.0V$ , $R_L = 4\Omega$ , fin = 1kHz

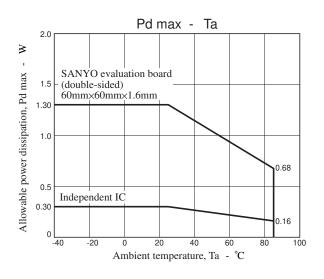
- ·	Symbol	0 1111		Ratings		
Parameter		Conditions	min	typ	max	Unit
Quiescent current drain	ICCOP	No signal		8.6	15	mA
Standby current drain	ISTBY	No signal, V8 = Low		0.1	10	μΑ
Maximum output power	POMAX	THD = 10%	220	350		mW
BTL maximum output power	POMXB	BTL mode, $R_L = 8\Omega$ , THD = 10%		700		mW
Voltage gain	VG	V <sub>IN</sub> = -30dBV	8.2	9.7	11.2	dB
Voltage gain use range	VGU		3		20	dB
Channel balance	CHB	$V_{IN} = -30 dBV$	-2	0	2	dB
Total harmonic distortion	THD	V <sub>IN</sub> = -30dBV		0.35	1	%
Output noise voltage	VNOUT	Rg = $620\Omega$ , 20 to $20kHz$		15	50	μVrms
Channel separation	CHSEP	VOUT = -10dBV, 20 to 20kHz -70 -81			dBV	
Ripple rejection ratio	SVRR	Rg = $620\Omega$ , fr = $100$ Hz, Vr = $-20$ dBV	53		dB	
Output DC offset voltage	VOF	Rg = $620\Omega$ , V3-V12, in BTL mode	-30	0	30	mV
Reference voltage	VREF			2.2		V
Pin 8 control HIGH voltage	V8H	(Power amplifier operation mode) 1.6		Vcc	V	
Pin 8 control LOW voltage	V8L	(Power amplifier standby mode) 0		0.3	V	
Pin 9 control HIGH voltage	V9H	(Second amplifier standby mode) 1.6 V		v <sub>CC</sub>	V	
Pin 9 control LOW voltage	V9L	(Second amplifier operation mode)	0 0.3		V	

# **Package Dimensions**

unit: mm (typ)

3179C





# **Pin Functions**

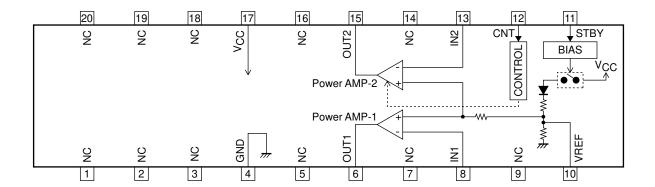
Pin No. Pin Name	Pin Voltage V <sub>CC</sub> = 5V	Description	Equivalent Circuit	
2	NC	-	No connect	
3	NC	-	No connect	
4	GND	0	Ground pin	
5	NC	-	No connect	
6 15	OUT1 OUT2	2.2	Power amplifier output pin	VCC VCC GND GND
7	NC	-	No connect	

Continued on next page.

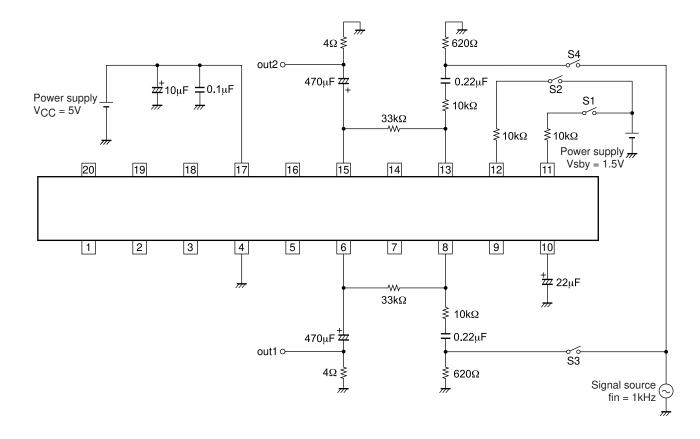
### Continued from preceding page.

Pin No. Pin Name		Pin Voltage	Description	Equivalent Circuit	
		V <sub>CC</sub> = 5V	Description	Equivalent Gircuit	
8 13	IN1 IN2	2.2	Input pin	V <sub>CC</sub> V <sub>CC</sub> V <sub>CC</sub> GND	
9	NC	-	No connect		
10	VREF	2.2	Ripple filter pin (For connection of capacitor for filter)	Vcc Vcc	
				V <sub>CC</sub> \$100kΩ 100kΩ \$107kΩ GND	
11	STBY	-	Standby pin Standby mode at 0V to 0.3V Operation mode at 1.6V to V <sub>CC</sub>	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	
12	CNT	-	Second amplifier stop control pin Second amplifier operation at 0V to 0.3V Second amplifier stop at 1.6V to V <sub>CC</sub>	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	
14	NC	-	No connect		
16	NC	-	No connect		
17	V <sub>CC</sub>	5	Power supply pin		
18	NC	-	No connect		
19	NC	-	No connect		
20	NC	-	No connect		

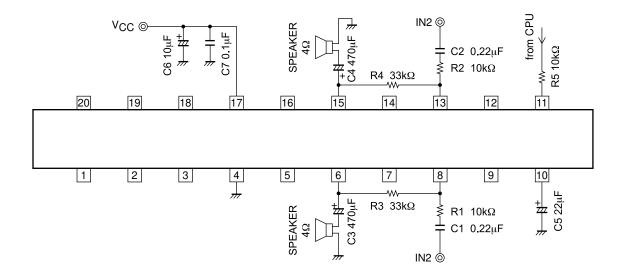
# **Block Diagram**



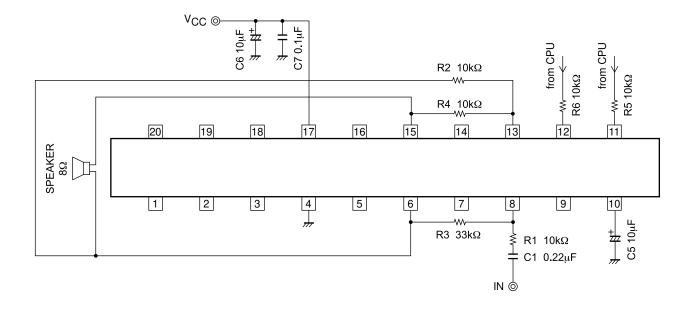
# **Test Circuit**



# **Application Circuit Example 1. (2-channel single ended mode)**



# **Application Circuit Example 2. (monaural BTL mode)**



## **Cautions for Use**

#### 1. Input coupling capacitors (C1, C2)

C1 and C2 are input coupling capacitors that are used to cut DC voltage. However, the input coupling capacitor C1 (C2) and input resistor R1 (R2) make up the high-pass filter, attenuating the bass frequency. Therefore, the capacitance value must be selected with due consideration of the cut-off frequency.

The cut-off frequency is expressed by the following formula:

fc = 
$$1/2 \pi \times R1 \times C1$$
 (=  $1/2 \pi \times R2 \times C2$ )

Note with care that this capacitance value affects the pop noise at startup. To increase this capacitance value, it is necessary to increase the capacitance value of pin 10 capacitor (C5) to soften the startup characteristics.

#### 2. Pin 10 capacitor (C5)

This capacitor C5 is designed for the ripple filter. Its purpose is to make up a low-pass filter with a  $100k\Omega$  internal resistor for reducing the ripple component of the power supply and improve the ripple rejection ratio.

Inside the IC, the startup characteristics of the pin 10 voltage are used to drive the automatic pop noise reduction circuit, and care must be taken with the pop noise when the C5 capacitance value is to be set lower.

However, when the IC is used in BTL mode, the automatic pop noise reduction function mentioned above has no effect. Instead, a pop noise reduction method that utilizes the second amplifier control function is used so that the capacitance value must be determined while factoring in the ripple rejection ratio or startup time.

Recommended capacitance value : Min.  $22\mu F$  (in 2-channel mode)

10μF (in mono BTL mode)

#### 3. Bypass capacitor (C7)

The purpose of the bypass capacitor C7 is to reject the high-frequency components that cannot be rejected by the power supply capacitor (chemical capacitor C6). Place the capacitor as near to the IC as possible, and use a ceramic capacitor with excellent high-frequency characteristics.

#### 4. Standby function

The standby function serves to place the IC in standby mode to minimize the current drain.

a) When using the standby function (when using microcomputer control)

By applying the following voltages to the standby pin (pin 11), the mode changeover can be performed between standby and operation.

Operation mode  $\cdots$  V11  $\geq$  1.6V Standby mode  $\cdots$  V11  $\leq$  0.3V

However, set the resistance of resistor R5 inserted in series in such a way that the condition in the following formula is met.

$$R5 \le 24.6 \times (Vstby - 1.6) k\Omega$$

The pin 11 inrush current is expressed by the following formula:

$$I11 = (40 \times \text{Vstby} - 26.3)/(1+0.04 \times \text{R5}) \mu\text{A}$$

#### b) When not using the standby function (microcomputer control is not possible)

By applying a voltage from the power supply (pin 17) to the standby pin (pin 11), the IC can be turned on without the control of the microcomputer when the power is turned on.

In order to reduce the pop noise when the IC is turned off, it is recommended that resistor R5 be inserted as shown in Fig.2. The resistance value indicated below is recommended for the inserted resistor R5.

$$V_{CC} = 5.0V : R5 = 82k\Omega$$
  
 $V_{CC} = 3.6V : R5 = 47k\Omega$ 

$$V_{CC} = 3.0V : R5 = 33k\Omega$$

#### 5. Second amplifier control function (only when BTL mode is used)

The second amplifier control function is a function to reduce the startup pop-noise in BTL mode. The pop noise can be reduced by first turning on the IC while the second amplifier is stopped, then after the potential inside the IC gets stabilized, turning on the second amplifier.

The values shown below are recommended for the control time.

C5 [μF]	2.2	3.3	4.7	10
Twu [ms]	200	250	300	500

<sup>\*</sup> Twu: Time after releasing standby to second amplifier turn-on

#### a) When using microcomputer control

The second amplifier can be controlled by applying the following voltages to pin 12.

Second amplifier operation mode  $\cdots$  V12  $\leq$  0.3V

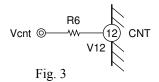
Second amplifier stop mode  $\cdots$  V12  $\geq$  1.6V

However, set the resistance value of the resistor R6 inserted in series in such a way that the condition in the following formula is met.

$$R6 \le 16.2 \times (Vent - 1.6) k\Omega$$

The pin 12 injected current is expressed by the following formula:

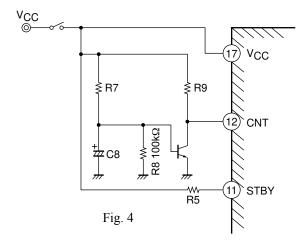
$$I12 = (57.6 \times Vent - 31.7)/(1+0.058 \times R6) \mu A$$



#### b) When microcomputer control is not possible

When the microcomputer cannot be used, the second amplifier can be controlled by adding the external components as shown in Fig. 4.

	V <sub>CC</sub> (V)				
	5 3.6 3				
R7 (kΩ)	10	6.8	6.8		
R9 (kΩ)	120	68	56		
C8 (μF)	100	100	100		



#### 6. Shorting between pins

When power is applied with pins left short-circuited, electrical deterioration or damage may result.

Therefore, check before power application if pins are short-circuited with solder, etc. during mounting of IC.

### 7.Load shorting

If the load is left short-circuited for a long period of time, electrical deterioration or damage may occur.

Never allow the load to short-circuit.

#### 8. Maximum rating

When IC is used near the maximum rating, there is a possibility that the maximum rating may be exceeded even under the smallest change of conditions, resulting in failure. Take sufficient margin for variation of supply voltage and use IC within a range where the maximum rating will never be exceeded.

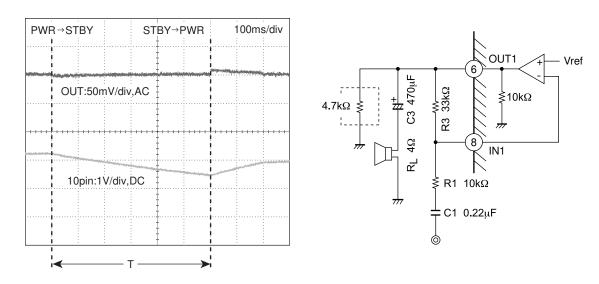
## 9. Turn-off transient response characteristics

If the IC is turned off and then turned back on while there is a potential difference between the pin 10 (reference voltage, plus input pin) and pins 8 and 15 (minus input pins), a louder pop noise than the one normally generated when power is switched on will be emitted. Therefore, in order to minimize the turn-on pop noise, smoothen the discharge of the input and output capacitors, and bring the potential of pin 10 and pins 8 and 15 to approximately the same level, then turn on the IC.

#### a) Single ended mode

When the continuous changeover of mode between standby and operation is necessary, it is recommended to insert a resistor between the output pins (pins 6 and 15) and ground to accelerate the turn-off transient response characteristic. The value shown below is recommended for the resistor used for discharge. In order to reduce pop noise, it is recommended that time necessary for turning the IC back on is greater than the following value.

Recommended discharge resistor :  $R = 4.7k\Omega$ (Recommended turn-on time : T = 600ms)

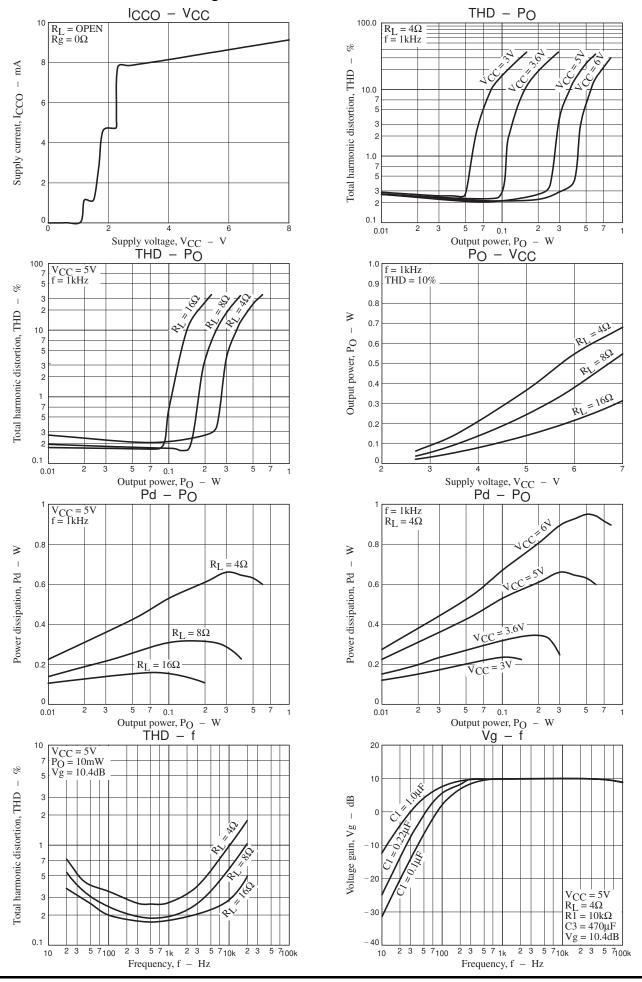


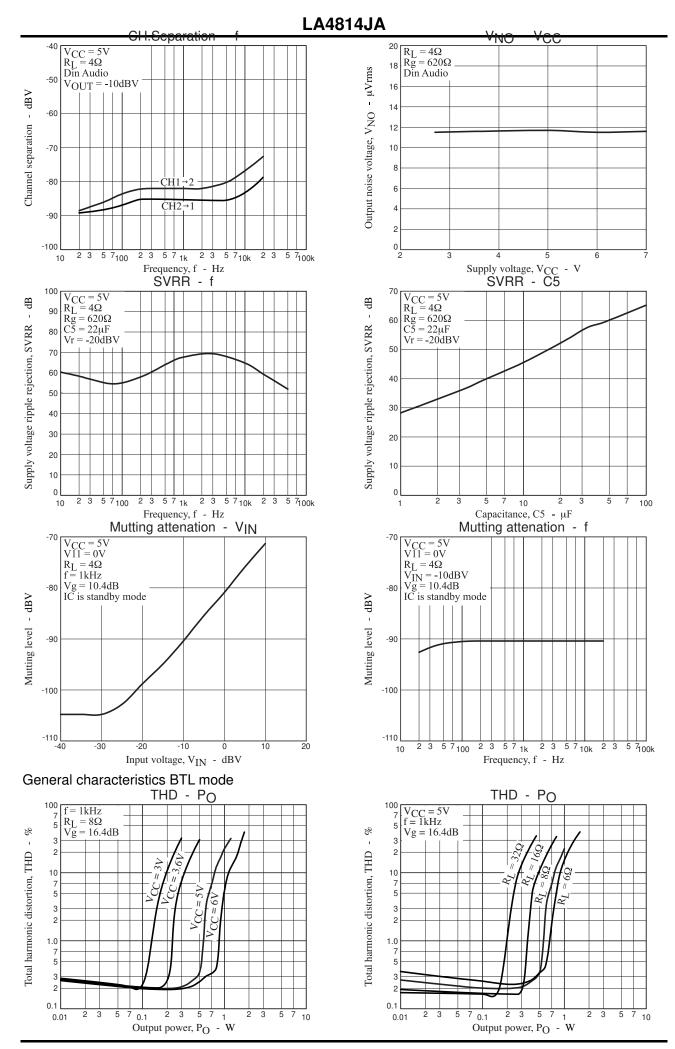
#### b) BTL mode

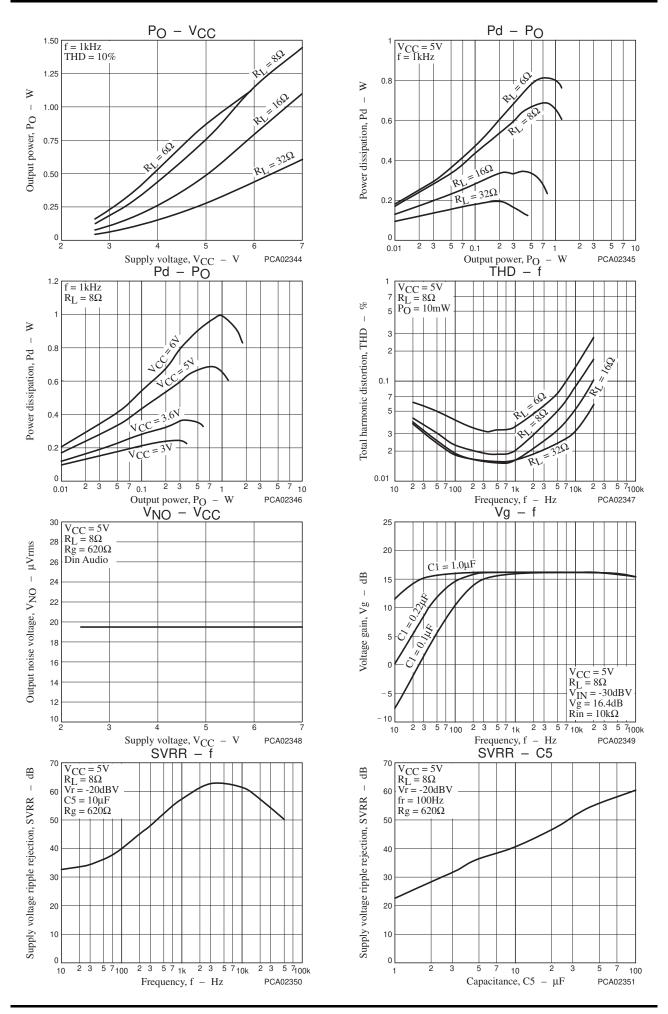
When the continuous changeover of mode between standby and operation is performed, it is recommended that the second amplifier control function be used to reduce the turn-on pop noise. If this function is used, the pop noise level can be reduced regardless of the time taken for the IC to turn on after it is turned off.

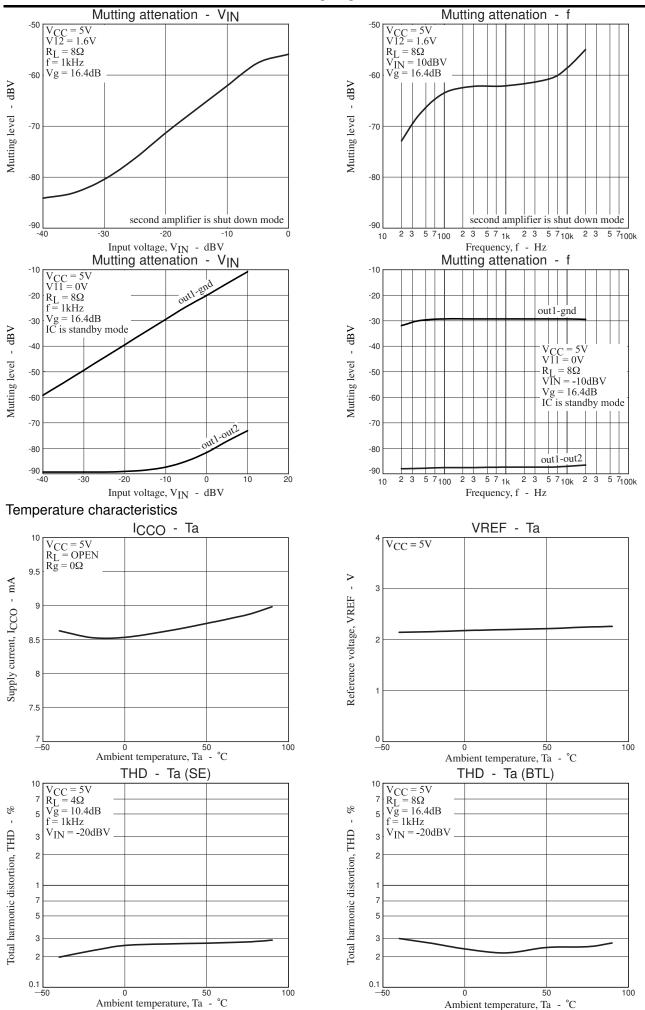
For details on the time taken for the second amplifier to turn on after the IC is turned on, refer to Section 5 "Second amplifier control function."

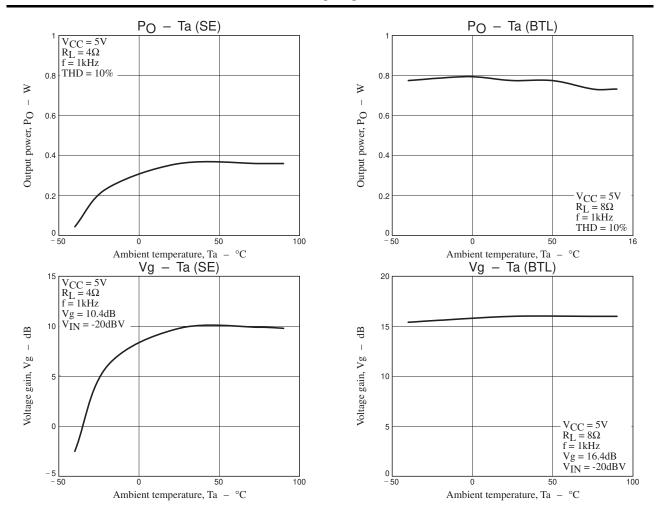
# General characteristics Single ended mode



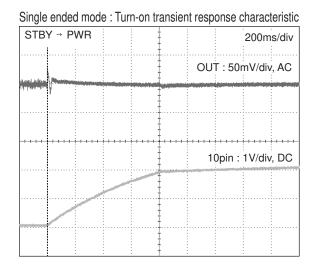


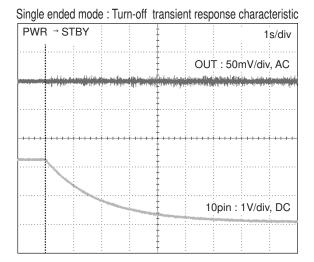


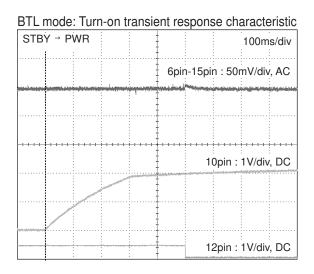


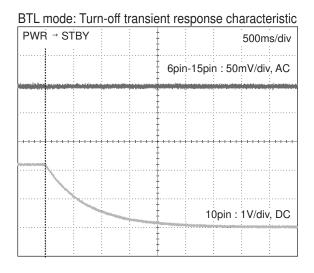


## Pop noise





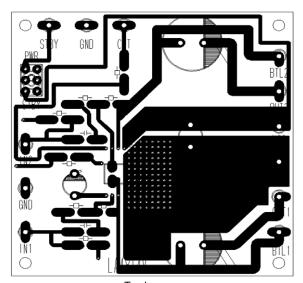


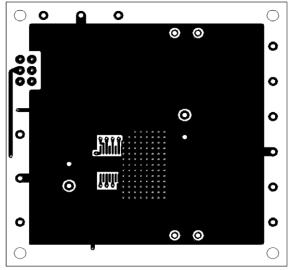


#### **Evaluation** board

1. Double-sided board

Size: 60mm×60mm×1.6mm





Top Layer

**Bottom Layer** 

ON Semiconductor and the ON logo are registered trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC owns the rights to a number of patents, trademarks, copyrights, trade secrets, and other intellectual property. A listing of SCILLC's product/patent coverage may be accessed at www.onsemi.com/site/pdf/Patent-Marking.pdf. SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. "Typical" parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the SCILLC product could create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that SCILLC was negligent regarding the design or manufacture of the part. SCILLC is an Equa