# **E**hipsmall

Chipsmall Limited consists of a professional team with an average of over 10 year of expertise in the distribution of electronic components. Based in Hongkong, we have already established firm and mutual-benefit business relationships with customers from,Europe,America and south Asia,supplying obsolete and hard-to-find components to meet their specific needs.

With the principle of "Quality Parts,Customers Priority,Honest Operation,and Considerate Service",our business mainly focus on the distribution of electronic components. Line cards we deal with include Microchip,ALPS,ROHM,Xilinx,Pulse,ON,Everlight and Freescale. Main products comprise IC,Modules,Potentiometer,IC Socket,Relay,Connector.Our parts cover such applications as commercial,industrial, and automotives areas.

We are looking forward to setting up business relationship with you and hope to provide you with the best service and solution. Let us make a better world for our industry!



## Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832 Email & Skype: info@chipsmall.com Web: www.chipsmall.com Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China







**Dual-Output RGB / 6-Channel WLED Driver**

**with LED-SenseTM Temperature & Color Compensation**

## **FEATURES**

- **Six PowerLiteTM Linear LDO current drivers with 25 mV drop-out in a common cathode topology with up to 25 mA per channel**
- **LED current programmable from 0 to 25 mA in 200 linear steps**
- **Three separately controlled driver banks (2 LED each) supports RGB LED applications.**
- **Integrated digital temperature sensor with 10 bit ADC; 1<sup>0</sup>C resolution with 5<sup>0</sup>C accuracy**
- **LED-SenseTM\* temperature compensation algorithm continually monitors LED V-I parameters and adjusts brightness per user loaded PWM correction**
- **Three integrated PWM generators support RGB color correction and dimming with 12-bit resolution and 256 user programmable logarithmic steps (~ 0.17 dB per step)**
- **I <sup>2</sup>C serial programming interface; additional address pin allows 4 unique slave addresses.**
- **Power efficiency up to 98%; average efficiency > 80% in Li-ion battery applications**
- **Low current shutdown mode (< 1 µA); Low current software "standby mode" (< 5 µA)**
- **Soft start and current limiting**
- **LED Short circuit detection and protection, LED open detection**
- **Thermal shutdown protection**
- **Low EMI.**
- **Available in 3 x 3 x 0.8 mm<sup>3</sup> 16-pin TQFN or ultra small WCSP 3 x 4 ball grid (0.4mm pitch).**

### **APPLICATION**

- **Keypad and Display Backlight**
- **Cellular Phones**
- **Digital Still Cameras**
- **PDAs and Smartphones**

#### **DESCRIPTION**

The LDS8160 is a dual-output RGB or 6-channel white LED driver with three temperature compensation circuits for each bank of two LED drivers. It supports both RGB LED and WLED backlighting and keypad in portable applications.

Three 8-bit DACs set the current level for each LED bank (A, B, & C) from 0 to 25mA in 0.125mA steps.



Each channel contains a linear LDO current driver in a common cathode (i.e., current source) topology.

The LDO drivers have a typical dropout voltage of 25mV at maximum rated current. This provides a low power and low EMI solution in Li-ion battery applications without voltage boosting and associated external capacitors and components.

Three 12-bit PWM generators with "smooth" logarithmic control support Temperature vs. LED Luminosity adjustments as well, as RGB color correction and dimming. The PWM generators are programmable via an I<sup>2</sup>C serial interface. User programmed 8-bit codes are converted to 12-bit resolution logarithmic steps of  $\sim$  0.17 dB per step. The PWM frequency is ~280 Hz to minimize noise.

The LED-Sense $TM$  temperature compensation engine includes a multiplexed 10-bit ADC and digital processing circuits. The algorithm continually measures the V-I characteristics of the LEDs and an on-chip temperature diode to determine LED junction temperatures to within 5ºC accuracy.

Three user-programmable temperature correction tables (LUTs) store PWM adjustment codes for every 5ºC increment from -35ºC to 120ºC. These codes drive the PWM engine to adjust for luminosity variations and/or high temperature current de-rating. The three correction LUTs support independent correction for 3-color RGB applications.

The EN logic input functions as a chip enable. A logic HIGH applied at EN allows the LDS8160 to respond to  ${}^{\circ}C$  communication. A serial address pin, SADD, supports use in multi-target applications. The device operates from 2.3V to 5.5V.

The LDS8160 is available in a 0.4mm pitch 12-ball WCSP or a 3 x 3 x 0.8 mm 16-lead TQFN packages.



## **TYPICAL APPLICATION CIRCUIT**





## **ABSOLUTE MAXIMUM RATINGS**



### **RECOMMENDED OPERATING CONDITIONS**



Typical application circuit with external components is shown on page 1.

### **ELECTRICAL OPERATING CHARACTERISTICS**

(Over recommended operating conditions unless specified otherwise) Vin = 3.6V, Cin = 1  $\mu$ F, EN = High, T<sub>AMB</sub> = 25°C







**Note:** 1.  $Vdx = Vin - V<sub>F</sub>$ ,

2. Vdx = Vin – VF, at which  $I_{\text{LED}}$  decreases by 10% from set value

3. Minimum LED forward voltage, which will be interpreted as "LED SHORT" condition

## **I <sup>2</sup>C CHARACTERISTICS**

Over recommended operating conditions unless otherwise specified for  $2.7 \leq$  VIN  $\leq$  5.5V, over full ambient temperature range -40 to +85°C.





**Figure 1: I<sup>2</sup>C Bus Timing Diagram**

#### **READ OPERATION:**

**Option 1**: Standard protocol sequential read:



where Reg. m is the last addressed in the write operation register

**Option 2:** Random access:



From reg. m, where Reg. m is the last addressed in the write operation register



**Option 3:** Random access with combined (extended) protocol:



#### **WRITE OPERATION:**

**Option 1:** Standard protocol sequencial write:



At  $k = 4$  data are send to register m and cycle repeats

**Option 2:** Combined (extended) protocol:



S: Start Condition Sr Start Repeat Condition R, W: Read bit (1), Write bit (0) A: Acknowledge (SDAT high) A\*: Not Acknowledge (SDAT low) P: Stop Condition Slave Address: Device address 7 bits (MSB first). Register Address: Device register address 8 bits Data: Data to read or write 8 bits - send by master - send by slave

## **I <sup>2</sup>C BUS PROTOCOL**

Standard protocol



Combined protocol:





## **WRITE INSTRUCTION SEQUENCE**

Standard protocol:



### **Write Instruction Example - Setting 20mA Current in LEDB1 and LEDB2**





### **REGISTER CONFIGURATION AND PROGRAMMING**









#### **Table 1**





#### **Table 2**



**Note:** \*) Value by default

#### **Table 3**



**Note:** \*) Value by default

#### **Table 4**



**Note:** \*) Value by default

\*\*) Trim code defined by customer

Bit 7 = 1 — Software reset: resets device, all registers reset/cleared.

Bit 6 = 1 — Standby (oscillator disabled, all registers retain programmed values.)

#### **Table 5: Ta-Tj Temperature Gradient Offset**

( set offset code to match reference De-rate point in LUT from LED Tj to Ta. Typically LED and Si are equal)



**Note:** \*) Value by default







#### **Table 7: T-code values vs. Temperature (for registers 4Ah and 4Bh)**



#### **Table 8: ΔPWM Code Allocation**



**Table 9: ΔPWM Codes vs. Number of Adjustment Steps**





### **PROGRAMMING EXAMPLES**



Note: XX – The LDS8160 I<sup>2</sup>C customer-selected slave address followed by binary 0 for write command, i.e. if I<sup>2</sup>C slave address is

001 0001 (see Table 10), XX = 0010 0010 (bin) = 22h<br>YY – The LDS8160 I<sup>2</sup>C customer-selected slave address followed by binary 1 for read command, i.e. if I<sup>2</sup>C slave address is 001 0001 (see Table 10), YY = 0010 0011 (bin) = 23h



### **PIN DESCRIPTION**





Top view: TQFN 16-lead 3 X 3 mm

### **PIN FUNCTION**

**VIN** is the supply pin. A small 1μF ceramic bypass capacitor is required between the  $V_{IN}$  pin and ground near the device. The operating input voltage range is from 2.3 V to 5.5 V.

**EN** is the enable input for the device. Guaranteed levels of logic high and logic low are set at 1.3 V and 0.4V respectively. When EN is initially taken high, the device becomes enabled and can communicate through  $I^2C$  interface after a 500  $\mu$ sec wakeup (initialization) period.

**SDAT** is the  ${}^{2}C$  serial data line. This is a bidirectional line allowing data to be written into and read from the registers of the LDS8160

**SCLK** is the I<sup>2</sup>C serial clock input.

**SADD** is I<sup>2</sup>C Serial interface Addresses Programming pin that allows choice of one of four  $I^2C$  addresses preprogrammed in device.

**GND** is the ground reference for internal circuitry. The pin must be connected to the ground plane on the PCB.

**LEDA1 – LEDC2** provide the internal regulated current sources for each of the LED anodes. These pins enter high-impedance zero current state whenever the device is in shutdown mode.

**PAD** is the exposed pad underneath the package. For best thermal performance, the tab should be soldered to the PCB and connected to the ground plane

**TST** is a test pin used by factory only. Leave it floating (no external connection)



## **BLOCK DIAGRAM**



**Figure 2: LDS8160 Functional Block Diagram**

### **BASIC OPERATION**

The LDS8160 may operate in follow modes:

- a) Normal Operation Mode
- b) Custom Operation Modes
- c) Normal Standby Mode
- d) Low Power (LP) Standby Mode
- e) Programming Modes
- f) Shutdown Mode

#### **NORMAL OPERATION MODE**

At power-up,  $V_{IN}$  should be in the range from 2.3 V to 5.5 V (max). If  $V_{IN}$  is slow rising, EN pin should be logic LOW at least until  $V_{IN}$  reaches 2.3 V level. When EN is taken HIGH, a soft-start power-up sequence begins and performs internal circuits reset that requires less than 100 µs.

An initialization sequence then begins taking less than 10 ms. This sequence determines the user-<br>selected  $1<sup>2</sup>C$  slave address. loads factory selected  $I^2C$  slave address, loads factory programmed settings, and conducts initial diagnostics for open/shorted LEDs.

At this point, the  $I^2C$  interface is ready for communication and the LDS8160 may be userprogrammed. Upon programming completion for all required initial parameters and features' settings, a calibration command is given by setting bit 4 of the

Control Register (1Fh) HIGH. This starts the calibration sequence of the LDS8160 LED-Sense™ temperature compensation circuits. The calibration process takes approximately 16 ms.

The user can then additionally program the DC current and PWM duty cycles for the LEDs. A PWM ramp-up sequence occurs after the writing to the PWM registers. This ramp-up delay in less than 250 ms in the default soft-start ramp mode, or can be 64 ms using the optional fast (4x) ramp mode (bit 3 of Register 19h = HIGH). A further option is available to bypass the soft-start PWM ramp mode entirely and the initialization time will be reduced to just the calibration sequence time of  $\sim$  16ms. The initial softstart ramp mode can be bypassed by setting bit 4 of register 19h HIGH.

The calibration parameters for the temperature measurement engine and all customer-set parameters remain intact until the part is reset or powered-down. Additionally, the user can re-calibrate LDS8160 during times when LED currents are brought to zero and the system is thermally stabilized by programming the calibration command bit as discussed.

Factory preset values (upon completion of the powerup initialization but prior to user programming) are as follow (see Table3):

- a) All LEDs are disabled and  $I_{LEDA, B, C} = 0$ ;
- b) RGB mode with three independent Luminosity vs. Temperature correction tables (LUTs) selected and three PWM generators;
- c) PWM dimming control in Logarithmic Mode with PWM generators running by 120 $^{\rm 0}$  phase shift;
- d) LED temperature compensation enabled with LUTs in Logarithmic Mode Soft start/shutdown enabled;
- e) Internal Diode for temperature compensation is enabled
- f) LEDs are used as sensors for temperature compensation control.

#### **LED Current Setting**

Current setting registers 00h – 02h should be programmed using  $I^2C$  interface and desired LEDs should be enabled using register 03h before LEDs turn on.

The standard  $I^2C$  interface procedure is used to program  $I_{LED}$  current (see section " $I<sup>2</sup>C$  INTERFACE"). LDS8160 should be addressed with slave address chosen followed by register address (00h, 01h, or



02h) and data that represents the code for the desired LED current. (See Table 10 for accessible slave addresses.)

Code for LED current is determined as  $I_{LED}/0.125$  mA in hex format, i.e. 20 mA current code =  $20/0.125$  =  $160$  (dec) = A0h.

The LDS8160 maximum current should not exceed 25 mA per LED (i.e. current code should not exceed 200 (dec) = C8h) to meet all electrical specifications.

To turn LEDs ON/OFF register 03h should be addressed with data that represents the desired combination of LEDs turned ON/OFF (see Table 1); i.e. if LEDC1, LEDC2, LEDA1, LEDA2 should be ON, and LEDB1, LEDB2 should be OFF, binary code that should be written into register 03h is 110011 (bin)  $=$ 33h.

The LDS8160 allows two ways for LED current setting. One of them is using registers 00h – 02h (static mode) and other one by using the PWM signal to decrease average LED current value set by these registers (dynamic mode).

For dynamic mode, the LDS8160 integrates 3 digital PWM generators that operate at a frequency of  $\sim$  285 Hz. In Logarithmic Mode, the PWM generators are 12-bit resolution and can be programmed with an 8 bit code to provide 256 internally mapped 12-bit logarithmic duty cycle steps to adjust the dimming level. In Linear Mode, the PWM generates 256 linear duty cycle steps to adjust the dimming levels from the user programmed 8-bit code.

The advantage of PWM dimming is stable LED color temperature / wavelength that are determined by the maximum LED current value set by registers 00h – 02h.

To use the dynamic PWM mode for LED current setting, the maximum  $I_{LED}$  value should be first set by registers 00h – 02h as described above in static mode and the desired PWM dimming should be set by registers 05h – 07h. In Logarithmic Mode, set by default, dimming resolution is approximately -0.17 dB per step with 0dB dimming, or 100% duty cycle, at the  $256^{\text{th}}$  step.

#### **Global PWM Dimming**

The LDS8160 allows Global PWM Dimming control of all three banks in the RGB Logarithmic mode, set by default. It is convenient, because it allows the user to simultaneously change LED brightness equally across to all three channels independent of the maximum static current setting (registers 00h, 01h and 02h) in a particular channel.

For example, to decrease LED brightness by 50% (-6dB) at all three LED banks, Global PWM Dimming data code written in register 04h should be  $6/0.17 =$ 35 (decimal) = 23h (see Figure 6: Global Dimming in Logarithmic Mode in percent vs. register 04h data  $(0\%$  dimming = full LED brightness).

The LDS8160 integrates temperature measurement and compensation processing to maintain stable LED brightness across varying ambient temperature and de-rate power dissipated by LEDs, if the LED die temperature exceeds a preset value.

#### **Figure 3: Dynamic Mode Dimming in Logarithmic Mode in dB vs. registers 05h – 07h data (0dB dimming = full LED brightness)**



**Figure 4: Dynamic Mode Dimming in Logarithmic Mode in percent vs. registers 05h – 07h data (0% dimming = full LED brightness)**



Measured temperatures are encoded into 5-bit T-codes representing  $5^0C$  temperature intervals from  $-35$  to  $+120^{\circ}$ C. The measured T-code addresses stored ΔPWM adjustment codes to adjust the dimming level and therefore average current through the LEDs. The user loads specific ΔPWM codes into the LUTs to maintain constant average current and therefore luminosity over temperature.



LUT corrections codes are added/subtracted to/from the user-set duty cycle/dimming codes (dynamic and/or global) for the channel to correct LED brightness.

The LDS8160 integrates a 10-bit ADC and digital processing to determine LED temperatures approximately every 2.5 seconds. The proprietary  $LED-Sense^{TM}$  algorithm allows direct measurement of LED junction temperatures on the LEDA1, LEDB1, and LEDC1 driver channels. Additionally an on-chip silicon temperature sensing diode is also measured to enhance temperature estimation accuracy.

#### **Figure 5: Global Dimming in Logarithmic Mode in dB vs. register 04h data (0dB dimming = full LED brightness)**



**Figure 6: Global Dimming in Logarithmic Mode in percent vs. register 04h data (0% dimming = full LED brightness)**



In normal operation mode, the LDS8160 senses the LED temperatures from all 3 available channels when in the default RGB (3 channel) mode, or only from the LEDA1 channel when used in the WLED (single channel) mode.

#### **Temperature vs. PWM Duty Cycle Profiles**

The user must load the PWM correction look up tables (LUTs) prior to operation. For the LDS8160 all three tables, LUT-B, LUT-G and LUT-R require loading (even if using same data for a WLED application) with the user correction profiles prior to operation. For RGB applications, LUT-B which drives LEDA1 And LEDA2 respectively should be assigned as the Blue color channel. LUT-G which drives LEDB1 and LEDB2 should be assigned as the Green color channel, and LUT-R which drives LEDC1 and LEDC2 should be the RED channel.

The correction tables are based upon LED vendor characteristics for luminosity vs temperature and current, LED current de-rating specifications, and user system thermal design parameters. Figure 7 shows an actual Luminosity vs. Temperature curve of the NSSM038AT-E RGB LED available from Nichia Corp.





Figure 8 shows the typical LED characteristic of decreasing illumination over temperature, but each color changes differently. This results in white light color shifts over temperature if not accounted for. It is typical to see RED LED Luminosity vs Temperature to change by  $\pm$  50% relative to the 25<sup>o</sup>C level.

Figure 9 shows that luminosity is linearly dependent with LED forward currents ≤ 30 mA. Therefore loss of LED luminosity over temperature can be compensated for by associated increases in LED current.

Figure 9 gives the total RGB Power de-rating specification for the same Nichia NSSM038AT-E RGB LED. Total power is the combined power ( $V_F$  x  $|F|$  of each color LED. This curve specifies the maximum RGB LED power that insures not exceeding the maximum specified junction temperature with maximum ambient operating temperature of 85ºC.

**IXYS Display** 

#### **Figure 8: Luminosity vs. LED Forward Current for Nichia NSSM038AT-E RGB LED**



**Figure 9: Total power (combined R, G, and B diodes) power de-rating curve (NSSM038AT-E RGB LED from Nichia)**



Figure 10 shows the final plot of typical LDS8160 PWM LUT correction profiles that could be programmed by the user to adjust for this RGB LED. This accumulated correction takes into account both the Luminosity vs Temperature variations and the adjustments to meet the higher temperature power de-rating specification.

Given the 5ºC increments of the temperature adjustment intervals for the LDS8160, the currents are slowly ramped to equalize loss of light output before the de-rating profile begins. Once de-rating begins, the PWM duty cycle is reduced, lowering LED driver current, to insure meeting and regulating to the desired maximum operating temperature.

**IXYS Display** 

**Figure 10: Example LDS8160 Accumulated PWM Correction Curves for Nichia NSMM038AT-E RGB LED for ILED nominal (R, G, & B) = 15 mA @ 25ºC**



Appendix 1 describes how to generate PWM LUT correction profiles. Additionally software tools and support is available from the factory to assist customers to generate LUT tables for specific LEDs and applications. Please consult the factory or a sales representative.

#### **Global Dimming Limitations**

The final PWM dimming code value is the algebraic sum of three codes: Dynamic Dimming code, Global Dimming Code, and the Temperature Compensation Code. If this sum is equal to or below zero, the LED in that particular channel is disabled. It means that the Global Dimming dynamic range is limited by Dynamic Dimming and the Temperature Correction Table used.

#### As an example:

If the user set PWM Dynamic Dimming in a particular channel is set to -20 dB (registers  $05h - 07h$  data = code 143 (dec)) and the LED-Sense™ Temperature vs. PWM Correction requires 7 steps correction dimming (data code 7 (dec)), the resultant allowable additional Global Dimming range =  $143 - 7 = 136$ (dec) steps or  $\sim$  - 23.1 dB.

#### **I <sup>2</sup>C Interface**

The LDS8160 uses a 2-wire serial  $I^2C$ -bus interface. The SDAT and SCLK lines comply with the  $I^2C$ electrical specification and should be terminated with pull-up resistors to the logic voltage supply. When the bus is not used, both lines are high. The device supports a maximum bus speed of 400kbit/s. The serial bit sequence is shown at *REGISTER CONFIGURATION AND PROGRAMMING* section for read and write operations into the registers. Read and write instructions are initiated by the master controller/CPU and acknowledged by the slave LED driver.

The LDS8160 allows user to choose between one of four preprogrammed  $I^2C$  addresses by connecting SADD pin  $(#3)$  either to ground, SCLK, SDAT or  $V_{IN}$ pin (see Table 10). Consult factory about other addresses available.





For further details on the  $I^2C$  protocol, please refer to the  $I^2C$ -Bus Specification, document number 9398-393-40011, from Philips Semiconductors.

#### **Recommended User Register Initialization**

Table 11 is provided as a recommended user I2C register initialization and calibration sequence for the the LDS8160 for an RGB LED application. RED values in the table mean these registers are user/system dependent. Any values shown are for example only.

#### **Unused LED Channels**

For applications with less than six white or two RGB LEDs, unused LED banks can be disabled via the I<sup>-C</sup> interface by addressing register 03h with data that represent desired combination of LEDs turned ON/OFF (see Table 1).

The LDS8160 unused LED outputs can be left open.

#### **LED short/open protection**

The LDS8160 runs a LED short/open diagnostic routine upon the power up sequence. It detects both LED pins shorted to ground and LED pins that are open or shorted to  $V_{IN}$  (fault conditions).

The results for short to GND detection are stored in Diagnostics Register 1Ch. Bits from bit 5 to bit 0 indicate a short status as  $bit = 1$  for LEDC2 - LEDA1 respectively, if the corresponding bit in the LED Faults detection Diagnostics register, 1Dh, is also High=1. A short to GND is detected if the measured LED pin voltage is less than  $\sim 0.14$  V independent of the programmed LED current. Every channel detected as shorted, is disabled

#### **Table 11: Recommended Register Load Sequence for LDS8160**



Test results for open or short to  $V_{IN}$  LED pins are stored in the LED Faults Diagnostics Register 1Dh, Bits from bit 5 to bit 0 represent LEDC2 - LEDA1 respectively with bit  $= 1$  indicates a fault condition at this particular LED pin. If the corresponding bit in register 1Ch is also High  $= 1$ , than the LED is shorted to GND as prior discussed. However when the bit in 1Dh is High  $= 1$  and the corresponding bit in 1Ch is Low  $= 0$ , than the fault is either a short to Vin or open.

An open LED pin fault causes no harm in the LDS8160 or LED as the high side driver has no current path from  $V_{IN}$  or GND. Therefore, the fault detection status indicates only in the 1Dh diagnostic register, and no further action is required.

In the case of and LED directly shorted to  $V_{IN}$ , the full  $V_{IN}$  voltage will be connected to the LED and current can flow independent of the LDS8160 LED driver circuit directly to GND. The LDS8160 will detect the fault and indicate the status in Register 1Dh, however further action needs taken at the system level to shutdown  $V_{\text{IN}}$  power to prevent possible damage to the LED. The combined series resistance of the LED (typically  $\sim$  10 $\Omega$  or more) and additional board series resistance will result in current limiting but not sufficient to prevent damage to low power LEDs.

Besides the power-up diagnostic sequence, the user can re-initiate a diagnostic command at any time by setting bit 5 of the Digital Test Modes Register, 19h, to HIGH.

The LDS8160 restores LED current to programmed value at channels with detected shorts to GND after the fault condition is removed.

#### **Over-Temperature Protection**

If the die temperature exceeds +150°C the driver will enter shutdown mode. The LDS8160 requires restart after die temperature falls below 130°C.

#### **LED Selection**

If the power source is a Li-ion battery, LEDs with  $V_F =$ 1.9 V - 3.3 V are recommended to achieve highest efficiency performance and extended operation on a single battery charge.

#### **External Components**

The driver requires one external  $1 \mu F$  ceramic capacitors  $(C_{IN})$  X5R or X7R type.

#### **CUSTOM OPERATION MODES**

The LDS8160 allows the option to choose custom operating modes overwriting content of Configuration Register 1Eh (see Table 2).



Bit 0 of this register allows switching between standard and low power standby modes (see detailed description at "STANDBY MODE" section).

Bit 1 allows bypass soft start / ramp down if fast raising/falling LED current required.

Bit 2 allows disable LED temperature compensation if desired.

Bit 3 changes PWM generators start condition.

At normal operation mode, set by default, PWM pulse rising edge of each PWM generator is shifted by 120 $^0$  in respect to two others. It allows for a decrease in input current noise especially at high LED currents. However, it may be important for better color mix in RGB mode to start all three PWM pulses simultaneously. To do so, set register 1Eh bit  $3 = 1$ .

Bits 4, 5 are for factory use only.

The LDS8160 also provides the option for using an external remote temperature-sensing device such as a 2N3904. This option is available on channel LEDA1 In this case, channel LEDA1 should be disabled via register 03h and it cannot operate as a LED current source.

A further option is available to monitor temperatures and make adjustments only from sensing the onchip silicon diode temperature. This option is enabled by setting bit  $4 = 1$  in register 1Eh. In this mode, temperature correction is via LUTA only.

Bit 6 allows to change the PWM generators operation mode from linear to logarithmic.

In Linear Mode, Dynamic Dimming resolution is  $\sim$ 0.39% per LSB. Code 00h represents 100% Dimming, while code  $FFh = 0%$ 

Linear Dimming Mode recommended for WLED Mode operation only because it creates nonproportional Global Dimming in RGB Mode.

In Linear Dimming Mode, Dynamic Dimming resolution is ~0.39% per LSB. Code 00h represents 100% Dimming, while Code FFh = 0% (See Figure 11).

Bit 7 allows switch between RGB and WLED modes.

In RBG Mode, set by default, the LDS8160 uses three independent PWM generators for LED current dynamic dimming and three LUTs for independent luminosity vs temperature correction. In WLED Mode, the LDS8160 uses a single PWM generator to dim all six LEDs and one LUT for luminosity vs temperature correction. It is convenient if all six WLED should have identical brightness. However, if

two or three different brightness levels are required for LED banks A, B, and C using dynamic dimming, RGB Mode is recommended even with WLED.

**Figure 11: Global Dimming in Linear Mode in percent vs. register 04h data (0% dimming = full LED brightness)**



#### **STANDBY MODES**

The LDS8160 has two standby modes, which customers may set by  $\hat{f}^2C$  interface addressing register 1Fh with bit  $6 = 1$  (see Table 4).

In both standby modes,  $I^2C$  interface remains active and all registers store information.

In Normal Standby Mode the LED drivers and internal clock are off; however, some internal circuits remain active resulting in a standby current from the  $V_{\text{IN}}$  power source of 125 µA typical. In this mode, the EN pin should be logic HIGH with signal level from 1.3 to  $V_{\text{IN}}$ voltage.

In Low Power (LP) Standby Mode most of the device is disabled and results in very low standby current from  $V_{\text{IN}}$  power source (5 µA typical). In LP Mode, the EN pin should be connected to a 1.8V voltage source capable to provide up to  $~100$   $~\mu$ A maximum dynamic current to LDS8160 digital core in case of any  ${}^1C$ interface activity.. If this voltage source is unavailable, Normal Standby Mode should be used.

To set LP Standby Mode, bit 0 in register 1Eh should be set to 1 (see Table 2) before addressing to register 1Fh.

#### **SHUTDOWN MODE**

To set LDS8160 in shutdown mode, EN pin should be logic low more than 10 ms. The LDS8160 shutdown current is less than 1 µA. The LDS8160 wakes up from shutdown mode with factory-preset data. To preserve customer-programmed data, use either Normal or LP standby modes.



#### **PROGRAMMING MODES**

The LDS8160 is factory preprogrammed with specific defaults for the Nichia NSSM038AT\_E RGB LEDs; however, application specific LEDs and other user system conditions may require user programming of the temperature compensation LUTs and other LED specific parameters.

After initialization and user programming the user should conduct an  $I^2C$  calibration sequence command by writing Bit  $4 = 1$  in the Control register 1Fh. This conducts a real time calibration of the initial starting temperature and the actual LED parameters. Upon completion, Bit 4 will be internally reset to 0, and the LDS8160 is ready for use.



### **TYPICAL CHARACTERISTICS**

(Over recommended operating conditions unless specified otherwise) Vin = 3.6V, Cin = 1  $\mu$ F, EN = High, T<sub>AMB</sub> = 25°C





**Output Driver Current vs. VDrop-Out Voltage**





## **PACKAGE DRAWING AND DIMENSIONS**

#### **16-PIN TQFN (HV3), 3mm x 3mm, 0.5mm PITCH**





Note:

- 1. All dimensions are in millimeters
- 2. Complies with JEDEC Standard MO-220



#### **ORDERING INFORMATION**



#### **Notes:**

- 1. Matte-Tin Plated Finish (RoHS-compliant)
- 2. Quantity per reel is 2000

### **EXAMPLE OF ORDERING INFORMATION**



#### **Notes:**

- 1) All packages are RoHS-compliant (Lead-free, Halogen-free).
- 2) The standard lead finish is Matte-Tin.
- 3) The device used in the above example is a LDS8160A 002–T2 (3x3 TQFN, Tape & Reel).
- 4) For additional package and temperature options, please contact your nearest IXYS Corp. Sales office.



#### **Appendix 1**

#### **CREATING LUT CORRECTION TABLES FOR LDS8160**

LED luminosity (or brightness) is proportional to forward current through the device and is dependent on temperature. To maintain a constant level of luminosity, the forward current should be adjusted vs. temperature. However, changing the static forward current also shifts the chromaticity of the LED, where each white or color LED has a different dependency with temperature.

The LDS8160 uses Dynamic Dimming control to change average LED current while maintaining the peak current thereby causing no color shift. The LED-Sense<sup>™</sup> temperature and color correction algorithm implements this current compensation feature by adjustment of the PWM duty cycles vs. the LEDs temperature. The LEDs' and an internal chip diode's I-V characteristics are routinely measured, digitized, and mapped to ΔPWM code adjustments stored in three integrated Luminosity vs. Temperature (LUT) lookup tables. Each LUT is assigned to one LED bank with two LED current drivers each. By default, banks A, B, and C are assigned to Blue, Green, and Red LEDs respectively. Additionally, the same LUTs can be used to insure current or power de-rating curve vs. temperature.

Figure A1.1 shows an actual Luminosity vs. Temperature curve of the NSSM038AT-E RGB LED available from Nichia Corp.





Figure A1.2 shows the total power (combined R, G, and B diodes) specification and de-rating for this RGB LED.



#### Figure A1.2: Total power (combined R, G, and B diodes) de-rating curve (NSSM038AT-E RGB LED from Nichia

Assuming that compensation should maintain Relative Luminosity  $= 1$  in full range of temperatures, the Compensation curve should be an inversion of the Luminosity vs. Temperature curve shown at Figure 1 (see Figure A1-3).





This characteristic must be fitted to the chosen nominal current at 25 $\mathrm{^{0}C}$ . Than the maximum current operating point is established and it must comply with the specified temperature de-rating curves for the LEDs.

Figure A1.5 represents LED Current vs. Temperature curve created for NSSM038AT-E RGB LED with 15ma chosen as the nominal current at  $25^{\circ}$ C, and a maximum power for the RGB LED of  $\sim$  133mW as depicted in Figure A1.4 showing the user-selected de-

rating curve. The user operating point must comply within the specification in Figure 2.



Figure A1.4: User Chosen Power and De-rating Curve starting at 55 $\mathrm{^0C}$  and shutdown at 85 $\mathrm{^0C}$ 

The maximum current of  $\sim$  18mA for the Red LED is limited by power dissipation at  $50^{\circ}$ C and decreases at higher temperatures in respect to the de-rating specification of Figure A1.2.





The LED Current vs. Temperature curves are then mapped to LDS8160 ΔPWM duty cycle codes that are loaded into each of the three LUTs as 32 4-bit words. Each word can represent from +7 to -7  $\Delta$ PWM steps for every 5<sup>o</sup>C temperature increment. The ΔPWM codes are loaded into registers 50h – 7Fh as 4-bit two's complement values (see Table 7 of main LDS8160 datasheet for code allocation).

To maintain correlation to typical LED vendor data, the tables establish  $25^{\circ}$ C as the zero-reference point. Therefore, "0" is the required  $\triangle$ PWM code value for 25<sup>o</sup>C. For temperatures above  $25^{\circ}$ C, the ΔPWM codes is the delta step change from the 5ºC temperature point lower than the current step, while for temperatures below  $25^0C$  the  $\triangle$ PWM code is the delta step change from the 5ºC temperature higher then the current step (i.e. closer to 25ºC). The compensation temperature range is from -35 to 120 $\mathrm{^0C}$ .

**IXYS Display** 

Example:

If  $\triangle$ PWM codes for the Red LED at 35<sup>°</sup>C are 0001 (1 step) and at  $40^{\circ}$ C 0010 (2 steps), register 77h should be loaded with code  $0010 0001$  (bin) = 21h.

The LDS8160 has three integrated PWM generators that allow programming of 256 logarithmic steps with 12-bit resolution in the LOG mode. Each PWM step is  $\sim$  0.17 dB from 300uA to 25mA in the 1-x scale mode and therefore  $\sim$  0.34 dB in the 2-x scale mode.

1-x scale is typically used in the temperature correction/compensation part of curve (as shown in Figure 5) A 2-x scale mode is also available to support the higher de-rating slope requirements

The LOG mode is required for RGB correction.

Linear mode operation and linear mode LUT correction codes are an option in WLED applications. If Linear WLED mode is chosen, all PWM related data for Dynamic Dimming and Temperature Compensation are entered as linear step codes, where each ΔPWM step is 1/256 of full brightness (100% Duty Cycle)

In WLED applications where Linear PWM option mode is chosen, only one PWM generator is active (i.e. the A or Blue channel). In Linear mode the PWM is 8-bit linear resolution where each bit represents is 1/256 of 100% duty cycle.

#### **Example: RGB LUT Table Generation**

Assume that the desired nominal forward current at  $25^{\circ}$ C is 15 mA at all three LEDs and the forward voltages for the R, G, B LEDs are  $\sim$  2.1 V, 3.2 V, and 3.2 V, respectively (per NSSM038AT-E datasheet).

If selected de-rating starts at  $50^{\circ}$ C, LED current values at this temperature would be (per the Luminosity Compensation Curve at Figure 3):

 $\sim$  1.2x the nominal value at 25<sup>o</sup>C, i.e. 15 x 1.2 = 18 mA for Red LED;

~ 1.04X the nominal value at  $25^{\circ}$ C, i.e. 15 x 1.04 = 15.6 mA for Green LED;

 $\sim$  1x the nominal value for Blue LED to maintain constant luminosity over temperature.

Users must also determine the typical forward voltage vs. Temperature coefficients, or "k" factors, of the LEDs used @ 1mA of forward current.

For the Nichia NSSM038AT-E these have been determined as;

 $-2.0$  mV/ $\mathrm{^0C}$  for RED LED,

**-** 1.5 mV/<sup>0</sup>C for Green LED, and

 $-1.3$  mV/ $^0$ C for Blue LED.

Therefore, at  $50^{\circ}$ C, forward voltages are  $V_F = 2.1V + [-2.0 \text{ mV}/^0\text{C} \times (50^0\text{C} - 25^0\text{C})] = 2.05V$  for Red