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LIS331DLF

MEMS digital output motion sensor
ultra low-power high performance 3-axes "nano" accelerometer

Features

- Wide supply voltage, 2.16 V to 3.6 V
- Low voltage compatible IOs, 1.8 V
- Ultra low-power mode consumption down to 10 µA
- ±2g/±4g/±8g dynamically selectable full-scale
- I^{2}C/SPI digital output interface
- 6 bit resolution
- 2 independent programmable interrupt generators for free-fall and motion detection
- Sleep to wake-up function
- 6D orientation detection
- Embedded self-test
- 10000 g high shock survivability
- ECOPACK® RoHS and "Green" compliant (see *Section 8*)
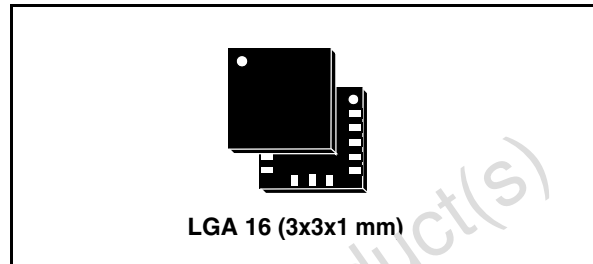
Applications

- Display orientation
- Gaming
- User interface
- Motion activated functions
- Intelligent power saving for handheld devices
- Impact recognition and logging

Description

The LIS331DLF is an ultra low-power high performance three axes linear accelerometer

LGA 16 (3x3x1 mm)

belonging to the "nano" family, with digital I^{2}C/SPI serial interface standard output.

The device features ultra low-power operational modes that allow advanced power saving and smart sleep to wake-up functions.

The LIS331DLF has dynamically user selectable full scales of ±2g/±4g/±8g and it is capable of measuring accelerations with output data rates from 0.5 Hz to 400 Hz.

The self-test capability allows the user to check the functioning of the sensor in the final application.

The device may be configured to generate interrupt signal by inertial wake-up/free-fall events as well as by the position of the device itself. Thresholds and timing of interrupt generators are programmable by the end user on the fly.

The LIS331DLF is available in small thin plastic land grid array package (LGA) and it is guaranteed to operate over an extended temperature range from -40 °C to +85 °C.

Table 1.   Device summary

Order codes	Temperature range [°C]	Package	Packaging
LIS331DLF	-40 to +85	LGA 16	Tray
LIS331DLFTR	-40 to +85	LGA 16	Tape and reel

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1 Block diagram and pin description

1.1 Block diagram

Figure 1. Block diagram
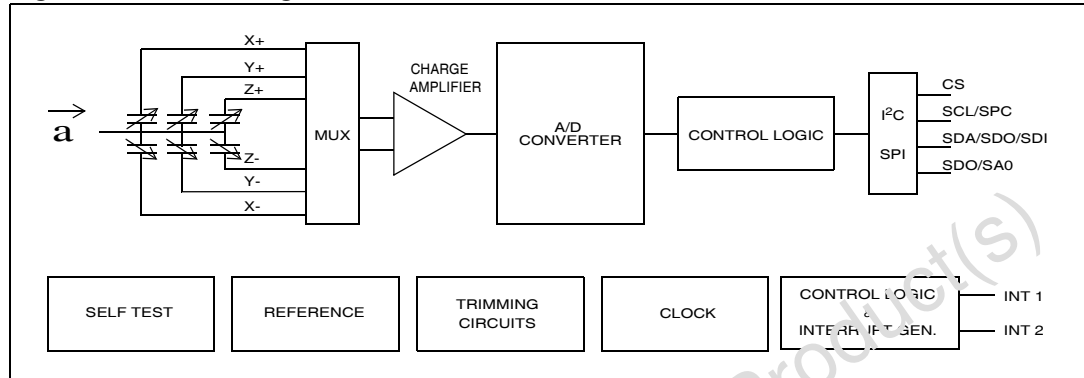
1.2 Pin description

Figure 2. Pin connection
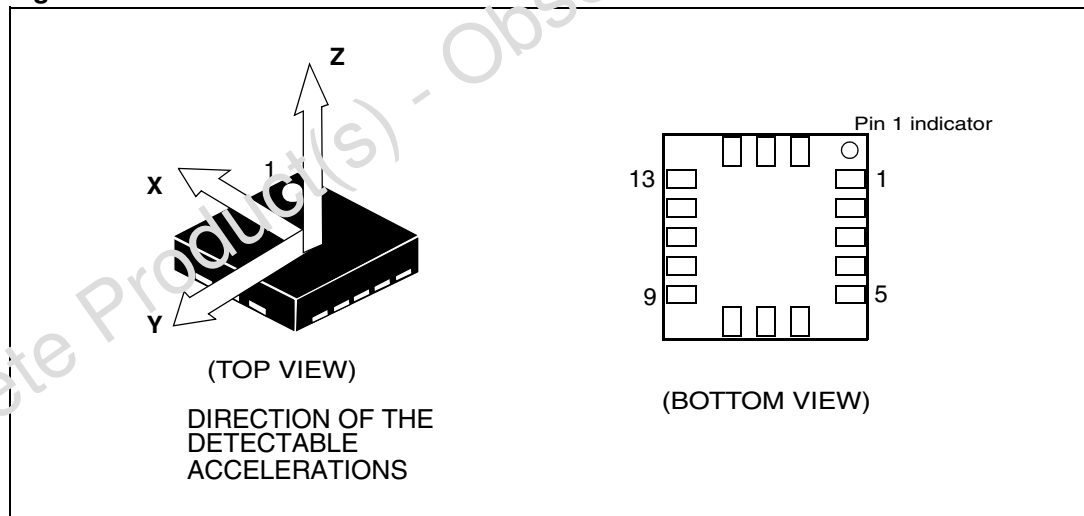
Table 2. **Pin description**

Pin#	Name	Function
1	Vdd_IO	Power supply for I/O pins
2	NC	Not connected
3	NC	Not connected
4	SCL SPC	I^2C serial clock (SCL) SPI serial port clock (SPC)
5	GND	0 V supply
6	SDA SDI SDO	I^2C serial data (SDA) SPI serial data input (SDI) 3-wire interface serial data output (SDO)
7	SDO SA0	SPI serial data output (SDO) I^2C less significant bit of the device address (SA0)
8	CS	SPI enable I^2C/SPI mode selection (1: I^2C mode; 0: SPI enabled)
9	INT 2	Inertial interrupt 2
10	Reserved	Connect to GND
11	INT 1	Inertial interrupt 1
12	GND	0 V supply
13	GND	0 V supply
14	Vdd	Power supply
15	Reserved	Connect to Vdd
16	GND	0 V supply

2   Mechanical and electrical specifications

2.1   Mechanical characteristics

Table 3.   Mechanical characteristics @ Vdd = 2.5 V, T = 25 °C unless otherwise noted [1]

Symbol	Parameter	Test conditions	Min.	Typ.[2]	Max.	Unit
FS	Measurement range[3]	FS bit set to 00		±2.0		g
		FS bit set to 01		±4.0		
		FS bit set to 11		±8.0		
So	Sensitivity	FS bit set to 00 6 bit representation		16		LSB/g
		FS bit set to 01 6 bit representation		8		
		FS bit set to 11 6 bit representation		4		
Dres	Device resolution	FS bit set to 00 ODR = 50 Hz		62.5		mg
TCSo	Sensitivity change vs temperature	FS bit set to 00		±0.01		%/°C
TyOff	Typical zero-g level offset accuracy[4],[5]	FS bit set to 00		±60		mg
TCOff	Zero-g level change vs temperature	Max delta from 25°C		±0.5		mg/°C
Vst	Self-test output change[6],[7],[8]	FS bit set to 00 X axis	1	5	8	LSb
		FS bit set to 00 Y axis	-1	-5	-8	LSb
		FS bit set to 00 Z axis	1	5	8	LSb
Top	Operating temperature range		-40		+85	°C
Wh	Product weight			20		mgram

1.   The product is factory calibrated at 2.5 V. The operational power supply range is from 2.16 V to 3.6 V.

2.   Typical specifications are not guaranteed

3.   Verified by wafer level test and measurement of initial offset and sensitivity

4.   Typical zero-g level offset value after MSL3 preconditioning

5.   Offset can be eliminated by enabling the built-in high pass filter

6.   The sign of "Self-test output change" is defined by CTRL_REG4 STsign bit (*Table 28*), for all axes.

7.   Self-test output changes with the power supply. "Self-test output change" is defined as
      OUTPUT[LSb]$_{(CTRL\_REG4\ ST\ bit=1)}$ - OUTPUT[LSb]$_{(CTRL\_REG4\ ST\ bit=0)}$. 1LSb=4$g$/64 at 6 bit representation, ±2 g full-scale

8.   Output data reach 99% of final value after 1/ODR+1 ms when enabling Self-test mode, due to device filtering

2.2 Electrical characteristics

Table 4. Electrical characteristics @ Vdd = 2.5 V, T = 25 °C unless otherwise noted [1]

Symbol	Parameter	Test conditions	Min.	Typ.[2]	Max.	Unit
Vdd	Supply voltage		2.16	2.5	3.6	V
Vdd_IO	I/O pins supply voltage[3]		1.71		Vdd+0.1	V
Idd	Current consumption in normal mode			250		µA
IddLP	Current consumption in low-power mode			10		µA
IddPdn	Current consumption in power-down mode			1		µA
VIH	Digital high level input voltage		0.8*Vdd_IO			V
VIL	Digital low level input voltage				0.2*Vdd_IO	V
VOH	High level output voltage		0.9*Vdd_IO			V
VOL	Low level output voltage				0.1*Vdd_IO	V
ODR	Output data rate in normal mode	DR bit set to 00		50		Hz
		DR bit set to 01		100		
		DR bit set to 10		400		
ODR$_{LP}$	Output data rate in low-power mode	PM bit set to 010		0.5		Hz
		PM bit set to 011		1		
		PM bit set to 100		2		
		PM bit set to 101		5		
		PM bit set to 110		10		
BW	System bandwidth[4]			ODR/2		Hz
Ton	Turn-on time[5]	ODR = 100 Hz		1/ODR+1ms		s
Top	Operating temperature range		-40		+85	°C

1. The product is factory calibrated at 2.5 V. The operational power supply range is from 2.16 V to 3.6 V.

2. Typical specification are not guaranteed

3. It is possible to remove Vdd maintaining Vdd_IO without blocking the communication busses, in this condition the measurement chain is powered off.

4. Refer to *Table 20* for filter cut-off frequency

5. Time to obtain valid data after exiting power-down mode

2.3 Communication interface characteristics

2.3.1 SPI - serial peripheral interface

Subject to general operating conditions for Vdd and Top.

Table 5. SPI slave timing values

Symbol	Parameter	Value [1]		Unit
		Min.	Max.	
tc(SPC)	SPI clock cycle	100		ns
fc(SPC)	SPI clock frequency		10	MHz
tsu(CS)	CS setup time	6		ns
th(CS)	CS hold time	8		
tsu(SI)	SDI input setup time	5		
th(SI)	SDI input hold time	15		
tv(SO)	SDO valid output time		50	
th(SO)	SDO output hold time	9		
tdis(SO)	SDO output disable time		50	

Figure 3. SPI slave timing diagram [2]
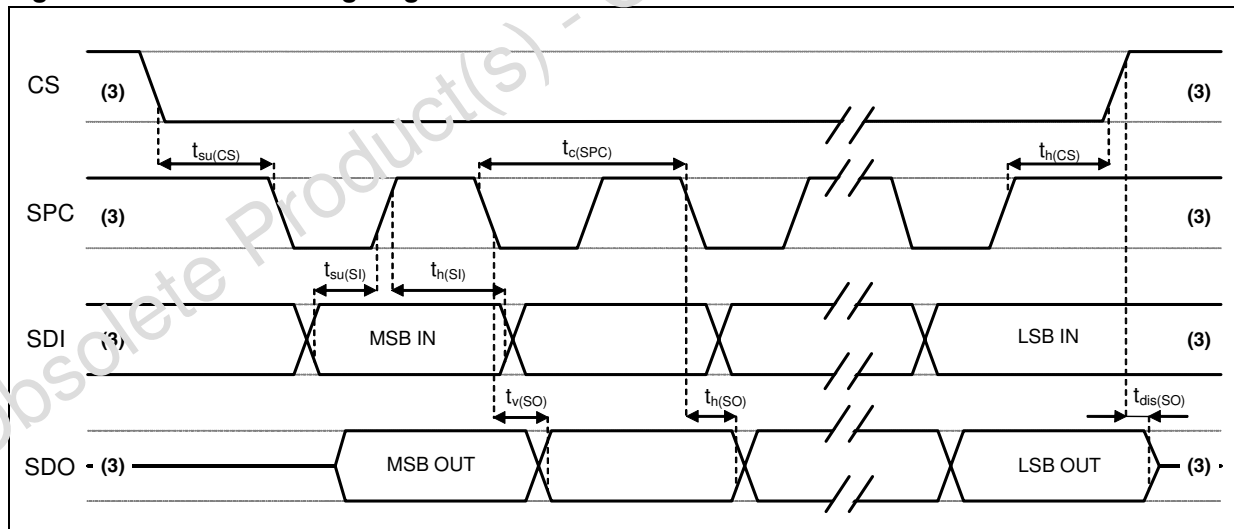
1. Values are guaranteed at 10 MHz clock frequency for SPI with both 4 and 3 wires, based on characterization results, not tested in production

2. Measurement points are done at 0.2·Vdd_IO and 0.8·Vdd_IO, for both Input and output port

3. When no communication is on-going, data on CS, SPC, SDI and SDO are driven by internal pull-up resistors

2.3.2 I^{2}C - Inter IC control interface

Subject to general operating conditions for Vdd and top.

Table 6. I^{2}C slave timing values

Symbol	Parameter	I^{2}C standard mode [1]		I^{2}C fast mode [1]		Unit
		Min	Max	Min	Max	
f$_{(SCL)}$	SCL clock frequency	0	100	0	400	KHz
t$_{w(SCLL)}$	SCL clock low time	4.7		1.3		μs
t$_{w(SCLH)}$	SCL clock high time	4.0		0.6		
t$_{su(SDA)}$	SDA setup time	250		100		ns
t$_{h(SDA)}$	SDA data hold time	0.01	3.45	0.01	0.9	μs
t$_{r(SDA)}$ t$_{r(SCL)}$	SDA and SCL rise time		1000	20 + 0.1C$_b$ [2]	300	ns
t$_{f(SDA)}$ t$_{f(SCL)}$	SDA and SCL fall time		300	20 + 0.1C$_b$ [2]	300	
t$_{h(ST)}$	START condition hold time	4		0.6		μs
t$_{su(SR)}$	Repeated START condition setup time	4.7		0.6		
t$_{su(SP)}$	STOP condition setup time	4		0.6		
t$_{w(SP:SR)}$	Bus free time between STOP and START condition	4.7		1.3		

1. Data based on standard I^{2}C protocol requirement, not tested in production

2. Cb = total capacitance of one bus line in pF

Figure 4. I^{2}C Slave timing diagram [a]
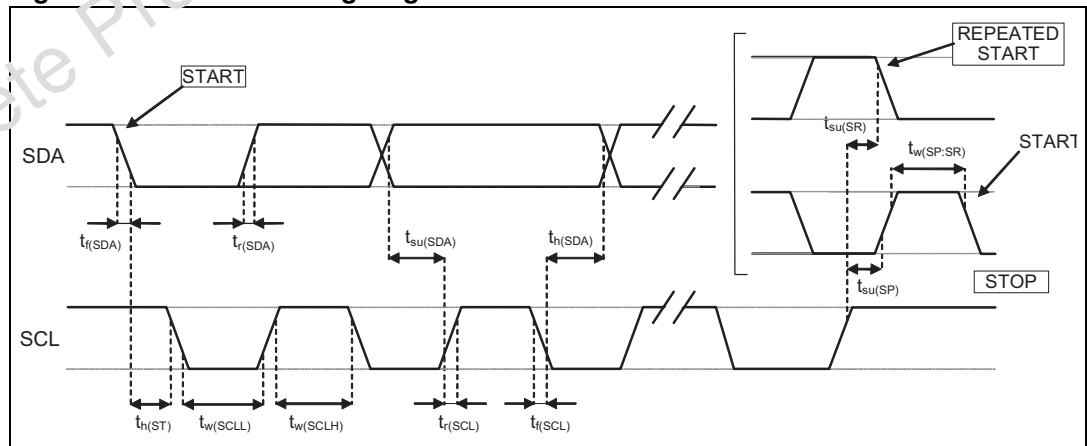
a. Measurement points are done at 0.2·Vdd_IO and 0.8·Vdd_IO, for both port

2.4 Absolute maximum ratings

Stresses above those listed as "absolute maximum ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device under these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 7. Absolute maximum ratings

Symbol	Ratings	Maximum value	Unit
Vdd	Supply voltage	-0.3 to 6	V
Vdd_IO	I/O pins supply voltage	-0.3 to 6	V
Vin	Input voltage on any control pin (CS, SCL/SPC, SDA/SDI/SDO, SDO/SA0)	-0.3 to Vdd_IO +0.3	V
A_{POW}	Acceleration (any axis, powered, Vdd = 2.5 V)	3000 g for 0.5 ms	
		10000 g for 0.1 ms	
A_{UNP}	Acceleration (any axis, unpowered)	3000 g for 0.5 ms	
		10000 g for 0.1 ms	
T_{OP}	Operating temperature range	-40 to +85	°C
T_{STG}	Storage temperature range	-40 to +125	°C
ESD	Electrostatic discharge protection	4 (HBM)	kV
		1.5 (CDM)	kV
		200 (MM)	V

Note: *Supply voltage on any pin should never exceed 6.0 V*

This is a mechanical shock sensitive device, improper handling can cause permanent damages to the part

This is an ESD sensitive device, improper handling can cause permanent damages to the part

2.5 Terminology

2.5.1 Sensitivity

Sensitivity describes the gain of the sensor and can be determined e.g. by applying 1 *g* acceleration to it. As the sensor can measure DC accelerations this can be done easily by pointing the axis of interest towards the center of the earth, noting the output value, rotating the sensor by 180 degrees (pointing to the sky) and noting the output value again. By doing so, ±1 *g* acceleration is applied to the sensor. Subtracting the larger output value from the smaller one, and dividing the result by 2, leads to the actual sensitivity of the sensor. This value changes very little over temperature and also time. The Sensitivity Tolerance describes the range of Sensitivities of a large population of sensors.

2.5.2 Zero-*g* level

Zero-*g* level offset (TyOff) describes the deviation of an actual output signal from the ideal output signal if no acceleration is present. A sensor in a steady state on a horizontal surface will measure 0 *g* in X axis and 0 *g* in Y axis whereas the Z axis will measure 1 *g*. The output is ideally in the middle of the dynamic range of the sensor (content of OUT registers 00h, data expressed as 2's complement number). A deviation from ideal value in this case is called Zero-*g* offset. Offset is to some extent a result of stress to MEMS sensor and therefore the offset can slightly change after mounting the sensor onto a printed circuit board or exposing it to extensive mechanical stress. Offset changes little over temperature, see "Zero-*g* level change vs. temperature". The Zero-*g* level tolerance (TyOff) describes the Standard deviation of the range of Zero-*g* levels of a population of sensors.

2.5.3 Self-test

Self-test allows to check the sensor functionality without moving it. The Self-test function is off when the self-test bit (ST) of CTRL_REG4 (control register 4) is programmed to '0'. When the self-test bit of CTRL_REG4 is programmed to '1' an actuation force is applied to the sensor, simulating a definite input acceleration. In this case the sensor outputs will exhibit a change in their DC levels which are related to the selected full scale through the device sensitivity. When self-test is activated, the device output level is given by the algebraic sum of the signals produced by the acceleration acting on the sensor and by the electrostatic test-force. If the output signals change within the amplitude specified inside *Table 3*, then the sensor is working properly and the parameters of the interface chip are within the defined specifications.

2.5.4 Sleep to wake-up

The "sleep to wake-up" function, in conjunction with low-power mode, allows to further reduce the system power consumption and develop new smart applications.
LIS331DLF may be set in a low-power operating mode, characterized by lower date rates refreshments. In this way the device, even if sleeping, keep on sensing acceleration and generating interrupt requests.
When the "sleep to wake-up" function is activated, LIS331DLF is able to automatically wake-up as soon as the interrupt event has been detected, increasing the output data rate and bandwidth.
With this feature the system may be efficiently switched from low-power mode to full-performance depending on user-selectable positioning and acceleration events, thus ensuring power saving and flexibility.

3   Functionality

The LIS331DLF is a "nano", low-power, digital output 3-axis linear accelerometer packaged in a LGA package. The complete device includes a sensing element and an IC interface able to take the information from the sensing element and to provide a signal to the external world through an I^2C/SPI serial interface.

3.1   Sensing element

A proprietary process is used to create a surface micro-machined accelerometer. The technology allows to carry out suspended silicon structures which are attached to the substrate in a few points called anchors and are free to move in the direction of the sensed acceleration. To be compatible with the traditional packaging techniques a cap is placed on top of the sensing element to avoid blocking the moving parts during the moulding phase of the plastic encapsulation.

When an acceleration is applied to the sensor the proof mass displaces from its nominal position, causing an imbalance in the capacitive half-bridge. This imbalance is measured using charge integration in response to a voltage pulse applied to the capacitor.

At steady state the nominal value of the capacitors are few pF and when an acceleration is applied the maximum variation of the capacitive load is in the fF range.

3.2   IC interface

The complete measurement chain is composed by a low-noise capacitive amplifier which converts the capacitive unbalancing of the MEMS sensor into an analog voltage that is finally available to the user by an analog-to-digital converter.

The acceleration data may be accessed through an I^2C/SPI interface thus making the device particularly suitable for direct interfacing with a microcontroller.

The LIS331DLF features a data-ready signal (RDY) which indicates when a new set of measured acceleration data is available thus simplifying data synchronization in the digital system that uses the device.

The LIS331DLF may also be configured to generate an inertial wake-up and free-fall interrupt signal accordingly to a programmed acceleration event along the enabled axes. Both Free-Fall and Wake-Up can be available simultaneously on two different pins.

3.3   Factory calibration

The IC interface is factory calibrated for sensitivity (So) and Zero-g level (TyOff).

The trimming values are stored inside the device in a non volatile memory. Any time the device is turned on, the trimming parameters are downloaded into the registers to be used during the active operation. This allows to use the device without further calibration.

4 Application hints

Figure 5. LIS331DLF electrical connection

Digital signal from/to signal controller.Signal's levels are defined by proper selection of Vdd_IO
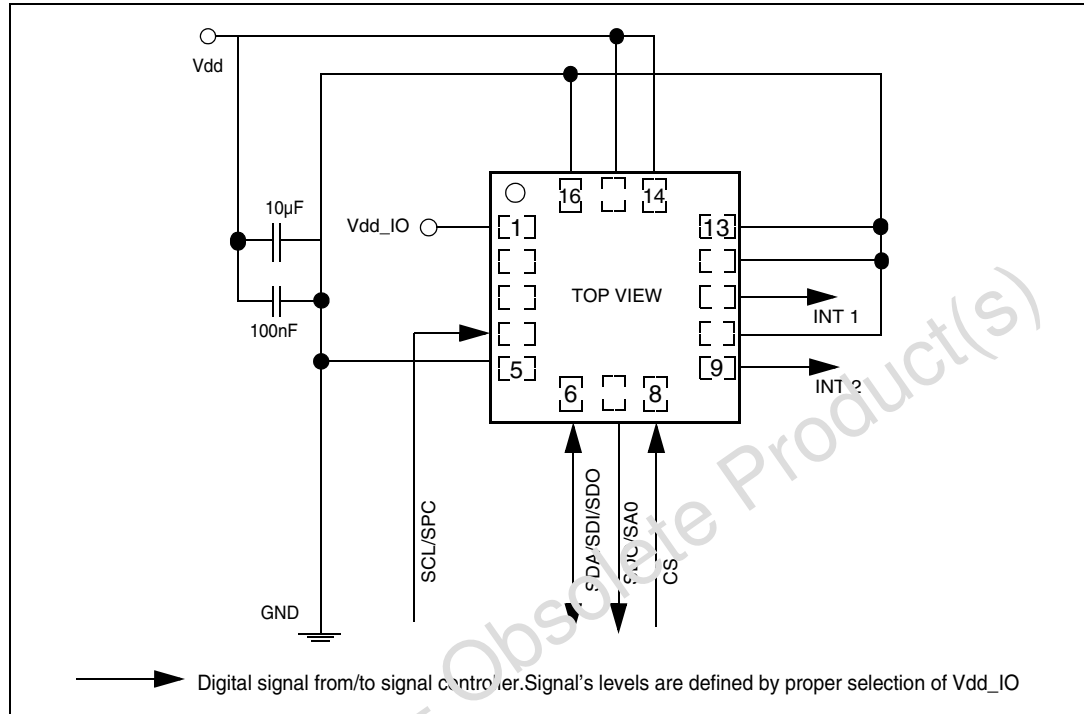
The device core is supplied through Vdd line while the I/O pads are supplied through Vdd_IO line. Power supply decoupling capacitors (100 nF ceramic, 10 µF Aluminum) should be placed as near as possible to the pin 14 of the device (common design practice).

All the voltage and ground supplies must be present at the same time to have proper behavior of the IC (refer to *Figure 5*). It is possible to remove Vdd maintaining Vdd_IO without blocking the communication bus, in this condition the measurement chain is powered off.

The functionality of the device and the measured acceleration data is selectable and accessible through the I²C or SPI interfaces.When using the I²C, CS must be tied high.

The functions, the threshold and the timing of the two interrupt pins (INT 1 and INT 2) can be completely programmed by the user through the I²C/SPI interface.

4.1 Soldering information

The LGA package is compliant with the ECOPACK®, RoHS and "Green" standard.
It is qualified for soldering heat resistance according to JEDEC J-STD-020C.

Leave "pin 1 indicator" unconnected during soldering.

Land pattern and soldering recommendations are available at www.st.com.

5   Digital interfaces

The registers embedded inside the LIS331DLF may be accessed through both the I²C and SPI serial interfaces. The latter may be SW configured to operate either in 3-wire or 4-wire interface mode.

The serial interfaces are mapped onto the same pads. To select/exploit the I²C interface, CS line must be tied high (i.e. connected to Vdd_IO).

Table 8.   Serial interface pin description

Pin name	Pin description
CS	SPI enable I²C/SPI mode selection (1: I²C mode; 0: SPI enabled)
SCL SPC	I²C serial clock (SCL) SPI serial port clock (SPC)
SDA SDI SDO	I²C serial data (SDA) SPI serial data input (SDI) 3-wire interface serial data output (SDO)
SA0 SDO	I²C less significant bit of the device address (SA0) SPI serial data output (SDO)

5.1   I²C serial interface

The LIS331DLF I²C is a bus slave. The I²C is employed to write data into registers whose content can also be read back.

The relevant I²C terminology is given in the table below.

Table 9.   Serial interface pin description

Term	Description
Transmitter	The device which sends data to the bus
Receiver	The device which receives data from the bus
Master	The device which initiates a transfer, generates clock signals and terminates a transfer
Slave	The device addressed by the master

There are two signals associated with the I²C bus: the serial clock line (SCL) and the serial data line (SDA). The latter is a bidirectional line used for sending and receiving the data to/from the interface. Both the lines are connected to Vdd_IO through a pull-up resistor embedded inside the LIS331DLF. When the bus is free both the lines are high.

The I²C interface is compliant with fast mode (400 kHz) I²C standards as well as with the normal mode.

5.1.1 I²C operation

The transaction on the bus is started through a START (ST) signal. A START condition is defined as a HIGH to LOW transition on the data line while the SCL line is held HIGH. After this has been transmitted by the Master, the bus is considered busy. The next byte of data transmitted after the start condition contains the address of the slave in the first 7 bits and the eighth bit tells whether the Master is receiving data from the slave or transmitting data to the slave. When an address is sent, each device in the system compares the first seven bits after a start condition with its address. If they match, the device considers itself addressed by the Master.

The slave address (SAD) associated to the LIS331DLF is 010100xb. **SDO**/**SA0** pad can be used to modify less significant bit of the device address. If SA0 pad is connected to voltage supply, LSb is '1' (address 0101001b) else if SA0 pad is connected to ground, LSb value is '0' (address 0101000b). This solution permits to connect and address two different accelerometers to the same I²C lines.

Data transfer with acknowledge is mandatory. The transmitter must release the SDA line during the acknowledge pulse. The receiver must then pull the data line LOW so that it remains stable low during the HIGH period of the acknowledge clock pulse. A receiver which has been addressed is obliged to generate an acknowledge after each byte of data received.

The I²C embedded inside the LIS331DLF behaves like a slave device and the following protocol must be adhered to. After the start condition (ST) a slave address is sent, once a slave acknowledge (SAK) has been returned, a 8-bit sub-address (SUB) is transmitted: the 7 LSb represent the actual register address while the MSB enables address auto increment. If the MSb of the SUB field is '1', the SUB (register address) is automatically increased to allow multiple data read/write.

The slave address is completed with a Read/Write bit. If the bit was '1' (Read), a repeated START (SR) condition must be issued after the two sub-address bytes; if the bit is '0' (Write) the Master will transmit to the slave with direction unchanged. *Table 10* explains how the SAD+Read/Write bit pattern is composed, listing all the possible configurations.

Table 10. SAD+Read/Write patterns

Command	SAD[6:1]	SAD[0] = SA0	R/W	SAD+R/W
Read	010100	0	1	01010001 (51h)
Write	010100	0	0	01010000 (50h)
Read	010100	1	1	01010011 (53h)
Write	010100	1	0	01010010 (52h)

Table 11. Transfer when master is writing one byte to slave

Master	ST	SAD + W		SUB		DATA		SP
Slave			SAK		SAK		SAK	

Table 12. Transfer when master is writing multiple bytes to slave:

Master	ST	SAD + W			SUB		DATA		DATA		SP
Slave			SAK			SAK		SAK		SAK	

Table 13. Transfer when master is receiving (reading) one byte of data from slave:

Master	ST	SAD + W		SUB		SR	SAD + R			NMAK	SP
Slave			SAK		SAK			SAK	DATA		

Table 14. Transfer when Master is receiving (reading) multiple bytes of data from slave

Master	ST	SAD+W		SUB		SR	SAD+R			MAK		MAK		NMAK	SP
Slave			SAK		SAK			SAK	DATA		DATA		DATA		

Data are transmitted in byte format (DATA). Each data transfer contains 8 bits. The number of bytes transferred per transfer is unlimited. Data is transferred with the Most Significant bit (MSb) first. If a receiver can't receive another complete byte of data until it has performed some other function, it can hold the clock line, SCL LOW to force the transmitter into a wait state. Data transfer only continues when the receiver is ready for another byte and releases the data line. If a slave receiver doesn't acknowledge the slave address (i.e. it is not able to receive because it is performing some real time function) the data line must be left HIGH by the slave. The Master can then abort the transfer. A LOW to HIGH transition on the SDA line while the SCL line is HIGH is defined as a STOP condition. Each data transfer must be terminated by the generation of a STOP (SP) condition.

In order to read multiple bytes, it is necessary to assert the most significant bit of the sub-address field. In other words, SUB(7) must be equal to 1 while SUB(6-0) represents the address of first register to be read.
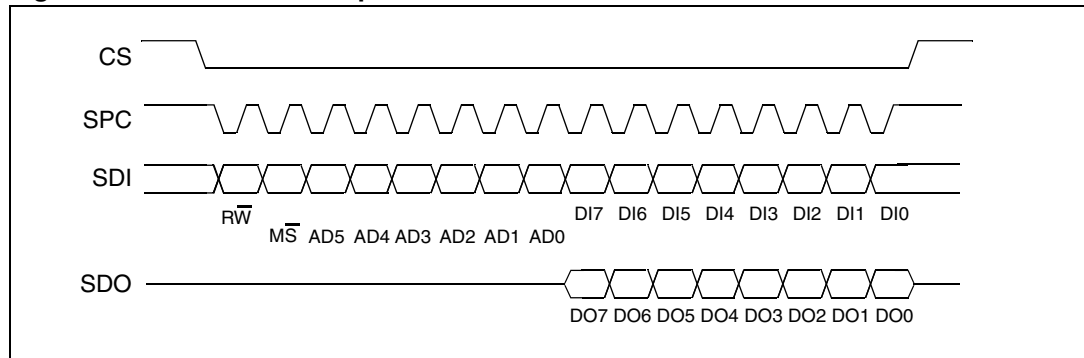
In the presented communication format MAK is master acknowledge and NMAK is no master acknowledge.

5.2 SPI bus interface

The LIS331DLF SPI is a bus slave. The SPI allows to write and read the registers of the device.

The Serial Interface interacts with the outside world with 4 wires: **CS**, **SPC**, **SDI** and **SDO**.

Figure 6.   Read and write protocol

CS is the serial port enable and it is controlled by the SPI master. It goes low at the start of the transmission and goes back high at the end. **SPC** is the Serial Port Clock and it is controlled by the SPI master. It is stopped high when **CS** is high (no transmission). **SDI** and **SDO** are respectively the serial port data input and output. Those lines are driven at the falling edge of **SPC** and should be captured at the rising edge of **SPC**.

Both the Read Register and Write Register commands are completed in 16 clock pulses or in multiple of 8 in case of multiple bytes read/write. Bit duration is the time between two falling edges of **SPC**. The first bit (bit 0) starts at the first falling edge of **SPC** after the falling edge of **CS** while the last bit (bit 15, bit 23, ...) starts at the last falling edge of SPC just before the rising edge of **CS**.

bit 0: R$\overline{W}$ bit. When 0, the data DI(7:0) is written into the device. When 1, the data DO(7:0) from the device is read. In latter case, the chip will drive **SDO** at the start of bit 8.

bit 1: M$\overline{S}$ bit. When 0, the address will remain unchanged in multiple read/write commands. When 1, the address is auto incremented in multiple read/write commands.

bit 2-7: address AD(5:0). This is the address field of the indexed register.

bit 8-15: data DI(7:0) (write mode). This is the data that is written into the device (MSb first).

bit 8-15: data DO(7:0) (read mode). This is the data that is read from the device (MSb first).

In multiple read/write commands further blocks of 8 clock periods will be added. When M$\overline{S}$ bit is '0' the address used to read/write data remains the same for every block. When M$\overline{S}$ bit is '1' the address used to read/write data is increased at every block.

The function and the behavior of **SDI** and **SDO** remain unchanged.

5.2.1   SPI read

Figure 7.   SPI read protocol
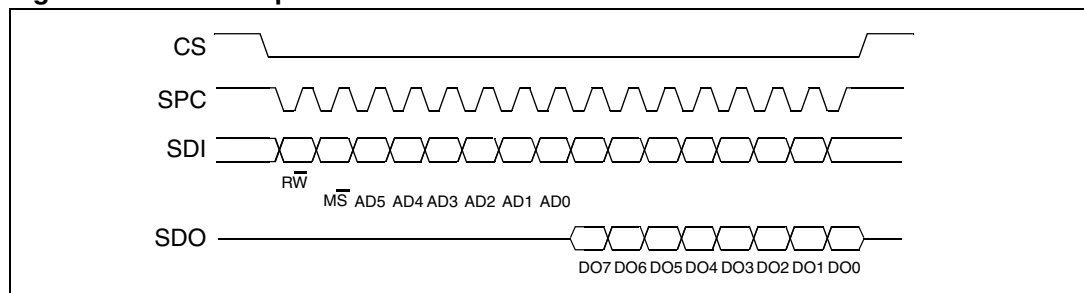
The SPI read command is performed with 16 clock pulses. Multiple byte read command is performed adding blocks of 8 clock pulses at the previous one.
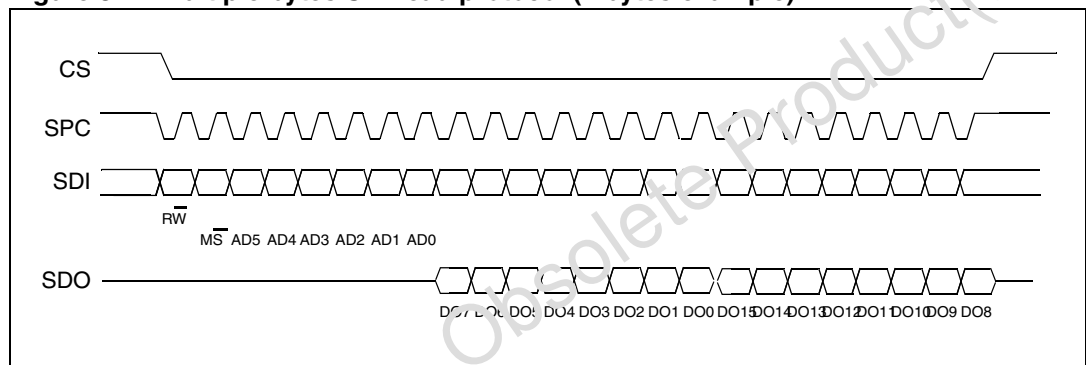
bit 0: READ bit. The value is 1.

bit 1: M$\overline{S}$ bit. When 0 do not increment address, when 1 increment address in multiple reading.

bit 2-7: address AD(5:0). This is the address field of the indexed register.

bit 8-15: data DO(7:0) (read mode). This is the data that will be read from the device (MSb first).
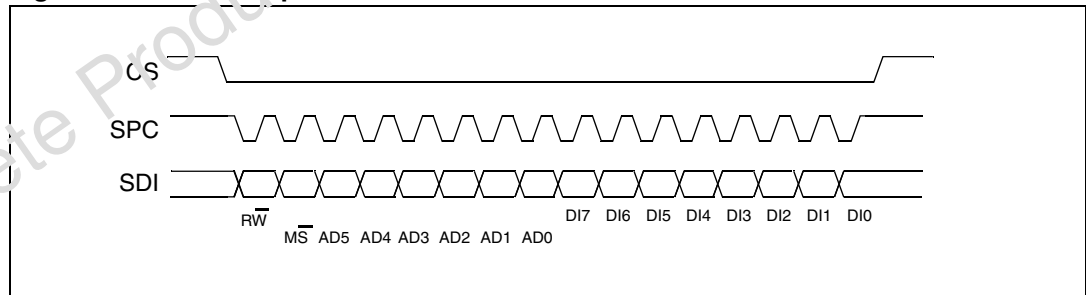
bit 16-... : data DO(...-8). Further data in multiple byte reading.

Figure 8.   Multiple bytes SPI read protocol (2 bytes example)

5.2.2   SPI write

Figure 9.   SPI write protocol

The SPI Write command is performed with 16 clock pulses. Multiple byte write command is performed adding blocks of 8 clock pulses at the previous one.

bit 0: WRITE bit. The value is 0.
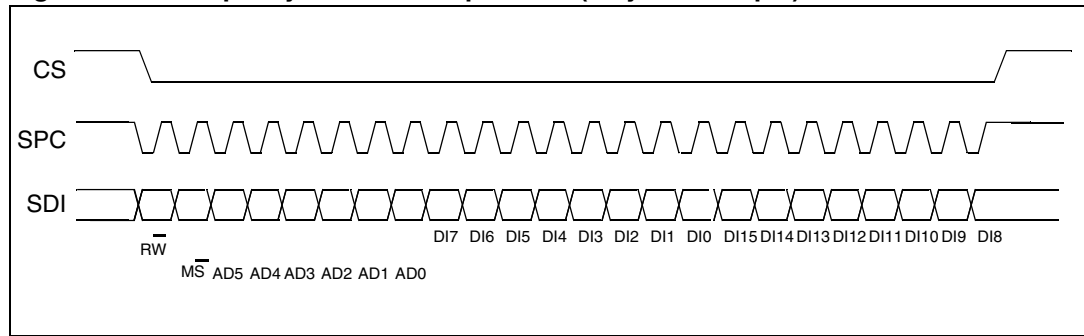
bit 1: M$\overline{S}$ bit. When 0 do not increment address, when 1 increment address in multiple writing.

bit 2 -7: address AD(5:0). This is the address field of the indexed register.

bit 8-15: data DI(7:0) (write mode). This is the data that will be written inside the device (MSb first).

bit 16-... : data DI(...-8). Further data in multiple byte writing.

Figure 10. Multiple bytes SPI write protocol (2 bytes example)

5.2.3 SPI read in 3-wires mode

3-wires mode is entered by setting to '1' bit SIM (SPI serial interface mode selection) in CTRL_REG4.

Figure 11. SPI read protocol in 3-wires mode
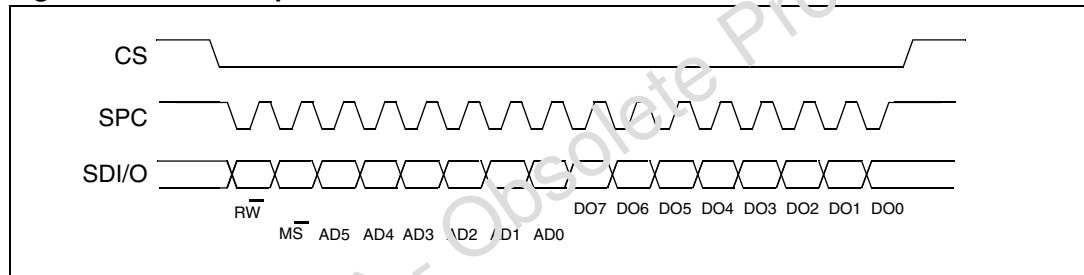
The SPI Read command is performed with 16 clock pulses:

bit 0: READ bit. The value is 1.

bit 1: MS bit. When 0 do not increment address, when 1 increment address in multiple reading.

bit 2-7: address AD(5:0). This is the address field of the indexed register.

bit 8-15: data DO(7:0) (read mode). This is the data that is read from the device (MSb first).

Multiple read command is also available in 3-wires mode.

6 Register mapping

The table given below provides a listing of the 8 bit registers embedded in the device and the related addresses:

Table 15. Register address map

Name	Type	Register address		Default	Comment
		Hex	**Binary**		
Reserved (do not modify)		00 - 0E			Reserved
WHO_AM_I	r	0F	000 1111	01010010	Dummy register
Reserved (do not modify)		10 - 1F			Reserved
CTRL_REG1	rw	20	010 0000	00000111	
CTRL_REG2	rw	21	010 0001	00000000	
CTRL_REG3	rw	22	010 0010	00000000	
CTRL_REG4	rw	23	010 0011	00000000	
CTRL_REG5	rw	24	010 0100	00000000	
HP_FILTER_RESET	r	25	010 0101		Dummy register
REFERENCE	rw	26	010 0110	00000000	
STATUS_REG	r	27	010 0111	00000000	
--	r	28	010 1000	00000000	Not used
OUT_X	r	29	010 1001	output	
--	r	2A	010 1010	00000000	Not used
OUT_Y	r	2B	010 1011	output	
--	r	2C	010 1100	00000000	Not used
OUT_Z	r	2D	010 1101	output	
Reserved (do not modify)		2E - 2F			Reserved
INT1_CFG	rw	30	011 0000	00000000	
INT1_SOURCE	r	31	011 0001	00000000	
INT1_THS	rw	32	011 0010	00000000	
INT1_DURATION	rw	33	011 0011	00000000	
INT2_CFG	rw	34	011 0100	00000000	
INT2_SOURCE	r	35	011 0101	00000000	
INT2_THS	rw	36	011 0110	00000000	
INT2_DURATION	rw	37	011 0111	00000000	
Reserved (do not modify)		38 - 3F			Reserved

Registers marked as *Reserved* must not be changed. The writing to those registers may cause permanent damages to the device.

The content of the registers that are loaded at boot should not be changed. They contain the factory calibration values. Their content is automatically restored when the device is powered-up.

7 Register description

The device contains a set of registers which are used to control its behavior and to retrieve acceleration data. The registers address, made of 7 bits, is used to identify them and to write the data through serial interface.

7.1 WHO_AM_I (0Fh)

Table 16.   WHO_AM_I register

0	1	0	1	0	0	1	0

Device identification register.

This register contains the device identifier that for LIS331DLF is set to 52h.

7.2 CTRL_REG1 (20h)

Table 17.   CTRL_REG1 register

PM2	PM1	PM0	DR1	DR0	Zen	Yen	Xen

Table 18.   CTRL_REG1 description

PM2 - PM0	Power mode selection. Default value: 000 (000: Power-down; Others: refer to *Table 19*)
DR1, DR0	Data rate selection. Default value: 00 (00:50 Hz; Others: refer to *Table 20*)
Zen	Z axis enable. Default value: 1 (0: Z axis disabled; 1: Z axis enabled)
Yen	Y axis enable. Default value: 1 (0: Y axis disabled; 1: Y axis enabled)
Xen	X axis enable. Default value: 1 (0: X axis disabled; 1: X axis enabled)

PM bits allow to select between power-down and two operating active modes. The device is in power-down mode when PD bits are set to "000" (default value after boot). *Table 19* shows all the possible power mode configurations and respective output data rates. Output data in the low-power modes are computed with low-pass filter cut-off frequency defined by DR1, DR0 bits.

DR bits, in the normal-mode operation, select the data rate at which acceleration samples are produced. In low-power mode they define the output data resolution. *Table 20* shows all the possible configuration for DR1 and DR0 bits.