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# LT1813/LT1814

# Dual/Quad 3mA, 100MHz, 750V/µs Operational Amplifiers

#### DESCRIPTION

The LT<sup>®</sup>1813/LT1814 are dual and quad, low power, high speed, very high slew rate operational amplifiers with excellent DC performance. The LT1813/LT1814 feature reduced supply current, lower input offset voltage, lower input bias current and higher DC gain than other devices with comparable bandwidth. The circuit topology is a voltage feedback amplifier with the slewing characteristics of a current feedback amplifier.

The output drives a  $100\Omega$  load to  $\pm 3.5$ V with  $\pm 5$ V supplies. On a single 5V supply, the output swings from 1.1V to 3.9V with a  $100\Omega$  load connected to 2.5V. The amplifiers are stable with a 1000pF capacitive load making them useful in buffer and cable driver applications.

The LT1813/LT1814 are manufactured on Linear Technology's advanced low voltage complementary bipolar process. The LT1813 dual op amp is available in 8-pin MSOP, SO and 3mm x 3mm low profile (0.8mm) dual fine pitch leadless packages (DFN). The quad LT1814 is available in 14-pin SO and 16-pin SSOP packages. A single version, the LT1812, is also available (see separate data sheet).

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#### FEATURES

- 100MHz Gain Bandwidth Product
- 750V/µs Slew Rate
- 3.6mA Maximum Supply Current per Amplifier
- Tiny 3mm x 3mm x 0.8mm DFN Package
- 8nV/√Hz Input Noise Voltage
- Unity-Gain Stable
- 1.5mV Maximum Input Offset Voltage
- 4µA Maximum Input Bias Current
- 400nA Maximum Input Offset Current
- 40mA Minimum Output Current, V<sub>OUT</sub> = ±3V
- $\pm 3.5V$  Minimum Input CMR,  $V_S = \pm 5V$
- 30ns Settling Time to 0.1%, 5V Step
- Specified at ±5V, Single 5V Supplies
- Operating Temperature Range: –40°C to 85°C

# APPLICATIONS

- Active Filters
- Wideband Amplifiers
- Buffers
- Video Amplification
- Communication Receivers
- Cable Drivers
- Data Acquisition Systems

## TYPICAL APPLICATION

R1 **Filter Frequency Response**  $\sim$ R R = 499Ω  $V_{S} = \pm 5V$ RQ С R<sub>G</sub> 0 V<sub>IN</sub> = 5V<sub>P-P</sub> DISTORTION:  $R1 = 499\Omega$ VIN  $R_F = 475\Omega$ OUTPUT MAGNITUDE (6dB/DIV)  $R_Q = 49.9\Omega$ 2nd < -76dB1/4 LT1814  $R_{G} = 499\Omega$  $R_{F}$ 3rd < -90dB1/4 LT1814 C = 3.3nF ACROSS FREQ  $f_{\rm C} = 100 \text{kHz}$ Q = 10 RANGE BANDPASS 1/4 LT1814 OUT GAIN = 1 GAIN = С Q = R RQ R  $f_{\rm C} = \frac{1}{2\pi R_{\rm F} C}$ 1/4 LT1814 1k 100k 1M 10k 10M FREQUENCY (Hz) 1814 TA02 1814 TA01 18134fa

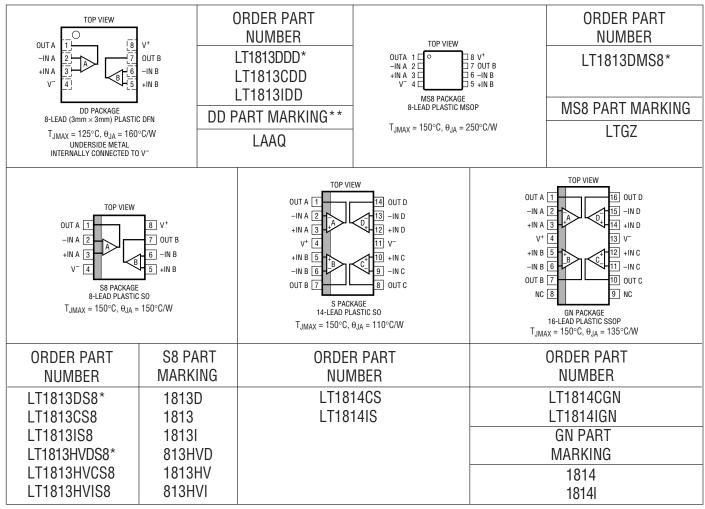
Bandpass Filter with Independently Settable Gain, Q and  $\mathbf{f}_{\mathbf{C}}$ 



### ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V <sup>+</sup> to V <sup>-</sup> )
LT1813/LT1814 12.6V
LT1813HV 13.5V
Differential Input Voltage (Transient Only, Note 2) ±6V
Input Voltage±V <sub>S</sub>
Output Short-Circuit Duration (Note 3) Indefinite
Operating Temperature Range40°C to 85°C

### PACKAGE/ORDER INFORMATION



Consult LTC marketing for parts specified with wider operating temperature ranges. \*See Note 9.

\*\*The temperature grades are identified by a label on the shipping container.



**ELECTRICAL CHARACTERISTICS** The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T<sub>A</sub> = 25°C. V<sub>S</sub> = ±5V, V<sub>CM</sub> = 0V, unless otherwise noted. (Note 8)

SYMBOL	PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS
V <sub>OS</sub>	Input Offset Voltage (Note 4)	$T_A = 0^{\circ}C \text{ to } 70^{\circ}C$	•	0.5	1.5 2	mV mV
$\frac{\Delta V_{0S}}{\Delta T}$	Input Offset Voltage Drift (Note 7)	$T_{A} = -40^{\circ}C \text{ to } 85^{\circ}C$ $T_{A} = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_{A} = -40^{\circ}C \text{ to } 85^{\circ}C$		10 10	3 15 30	mV µV/°C µV/°C
I <sub>OS</sub>	Input Offset Current	$T_{A} = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_{A} = -40^{\circ}C \text{ to } 85^{\circ}C$	•	50	400 500 600	nA nA nA
I <sub>B</sub>	Input Bias Current	$T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•	-0.9	±4 ±5 ±6	μΑ μΑ μΑ
e <sub>n</sub>	Input Noise Voltage Density	f = 10kHz		8		nV/√Hz
i <sub>n</sub>	Input Noise Current Density	f = 10kHz		1		pA/√Hz
R <sub>IN</sub>	Input Resistance	V <sub>CM</sub> = 3.5V Differential	3	10 1.5		ΜΩ ΜΩ
CIN	Input Capacitance			2		pF
V <sub>CM</sub>	Input Voltage Range	Guaranteed by CMRR $T_A = -40^{\circ}C$ to $85^{\circ}C$	±3.5 ±3.5	±4.2		V V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5V$ $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	75 • 73 • 72	85		dB dB dB
	Minimum Supply Voltage	Guaranteed by PSRR T <sub>A</sub> = -40°C to 85°C	•	±1.25	±2 ±2	V V
PSRR	Power Supply Rejection Ratio	$V_{S} = \pm 2V \text{ to } \pm 5.5V$ $T_{A} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_{A} = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	78 • 76 • 75	97		dB dB dB
		$V_{S} = \pm 2V \text{ to } \pm 6.5V \text{ (LT1813HV)}$ $T_{A} = 0^{\circ}\text{C} \text{ to } 70^{\circ}\text{C}$ $T_{A} = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}$	75 • 73 • 72	97		dB dB dB
A <sub>VOL</sub>	Large-Signal Voltage Gain	$V_{0UT} = \pm 3V, R_L = 500\Omega$ $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	1.5 1.0 0.8	3		V/mV V/mV V/mV
		$V_{OUT} = \pm 3V, R_L = 100\Omega$ $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	1.0 • 0.7 • 0.6	2.5		V/mV V/mV V/mV
V <sub>OUT</sub>	Maximum Output Swing (Positive/Negative)	$ \begin{array}{l} R_{L} = 500\Omega, \ 30 \text{mV} \ \text{Overdrive} \\ T_{A} = 0^\circ \text{C} \ \text{to} \ 70^\circ \text{C} \\ T_{A} = -40^\circ \text{C} \ \text{to} \ 85^\circ \text{C} \end{array} $		±4		V V V
		$ \begin{array}{l} R_{L} = 100\Omega, \ 30 \text{mV} \ \text{Overdrive} \\ T_{A} = 0^{\circ} \text{C} \ \text{to} \ 70^{\circ} \text{C} \\ T_{A} = -40^{\circ} \text{C} \ \text{to} \ 85^{\circ} \text{C} \end{array} $	±3.35 ±3.25 ±3.15			V V V



**ELECTRICAL CHARACTERISTICS** The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^{\circ}$ C.  $V_S = \pm 5$ V,  $V_{CM} = 0$ V, unless otherwise noted. (Note 8)

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
I <sub>OUT</sub>	Maximum Output Current	$V_{OUT} = \pm 3V, 30mV \text{ Overdrive}$ $T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	•	±40 ±35 ±30	±60		mA mA mA
I <sub>SC</sub>	Output Short-Circuit Current	$V_{OUT} = 0V$ , 1V Overdrive (Note 3) $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•	±75 ±60 ±55	±100		mA mA mA
SR	Slew Rate	$A_V = -1 \text{ (Note 5)}$ $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•	500 400 350	750		V/µs V/µs V/µs
FPBW	Full Power Bandwidth	6V <sub>P-P</sub> (Note 6)			40		MHz
GBW	Gain Bandwidth Product	$f = 200 \text{ kHz},  \text{R}_{\text{L}} = 500 \Omega$ $T_{\text{A}} = 0^{\circ} \text{C to } 70^{\circ} \text{C}$ $T_{\text{A}} = -40^{\circ} \text{C to } 85^{\circ} \text{C}$	•	75 65 60	100		MHz MHz MHz
-3dB BW	–3dB Bandwidth	$A_V = 1, R_L = 500\Omega$			200		MHz
t <sub>r</sub> , t <sub>f</sub>	Rise Time, Fall Time	$A_V = 1, 10\%$ to 90%, 0.1V, $R_L = 100\Omega$			2		ns
t <sub>PD</sub>	Propagation Delay (Note 10)	$A_V = 1,50\%$ to 50%, 0.1V, $R_L = 100\Omega$			2.8		ns
OS	Overshoot	$A_V = 1, 0.1V, R_L = 100\Omega$			25		%
ts	Settling Time	$A_V = -1, 0.1\%, 5V$			30		ns
THD	Total Harmonic Distortion	$A_V$ = 2, f = 1MHz, $V_{OUT}$ = 2 $V_{P-P}$ , $R_L$ = 500 $\Omega$			-76		dB
dG	Differential Gain	$A_V = 2, V_{OUT} = 2V_{P-P}, R_L = 150\Omega$			0.12		%
dP	Differential Phase	$A_V = 2, V_{OUT} = 2V_{P-P}, R_L = 150\Omega$			0.07		DEG
R <sub>OUT</sub>	Output Resistance	$A_V = 1$ , f = 1MHz			0.4		Ω
	Channel Separation	$ \begin{array}{l} V_{OUT}=\pm 3V, \ R_L=100\Omega \\ T_A=0^\circ C \ to \ 70^\circ C \\ T_A=-40^\circ C \ to \ 85^\circ C \end{array} $	•	82 81 80	100		dB dB dB
I <sub>S</sub>	Supply Current	Per Amplifier $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		3	3.6 4.5 5.0	mA mA mA
		Per Amplifier, $V_S = \pm 6.5V$ , (LT1813HV only) $T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•			4.0 5.0 5.5	mA mA mA



**ELECTRICAL CHARACTERISTICS** temperature range, otherwise specifications are at  $T_A = 25^{\circ}C$ .  $V_S = 5V$ ,  $V_{CM} = 2.5V$ ,  $R_L$  to 2.5V, unless otherwise noted. (Note 8)

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V <sub>OS</sub>	Input Offset Voltage (Note 4)	$T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•		0.7	2.0 2.5 3.5	mV mV mV
$\frac{\Delta V_{0S}}{\Delta T}$	Input Offset Voltage Drift (Note 7)	$T_A = 0^{\circ}C \text{ to } 70^{\circ}C$ $T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	•		10 10	15 30	μV/°C μV/°C
I <sub>OS</sub>	Input Offset Current	$T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		50	400 500 600	nA nA nA
I <sub>B</sub>	Input Bias Current	$T_A = 0^{\circ}C$ to $70^{\circ}C$ $T_A = -40^{\circ}C$ to $85^{\circ}C$	•		-1	±4 ±5 ±6	μΑ μΑ μΑ
e <sub>n</sub>	Input Noise Voltage Density	f = 10kHz			8		nV/√Hz
i <sub>n</sub>	Input Noise Current Density	f = 10kHz			1		pA/√Hz
R <sub>IN</sub>	Input Resistance	V <sub>CM</sub> = 3.5V Differential		3	10 1.5		ΜΩ ΜΩ
CIN	Input Capacitance				2		pF
V <sub>CM</sub>	Input Voltage Range (Positive)	Guaranteed by CMRR $T_A = -40^{\circ}C$ to $85^{\circ}C$	•	3.5 3.5	4.2		V V
	Input Voltage Range (Negative)	Guaranteed by CMRR T <sub>A</sub> = -40°C to 85°C	•		0.8	1.5 1.5	V V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1.5V \text{ to } 3.5V$ $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•	73 71 70	82		dB dB dB
	Minimum Supply Voltage	Guaranteed by PSRR T <sub>A</sub> = -40°C to 85°C	•		2.5	4 4	V V
A <sub>VOL</sub>	Large-Signal Voltage Gain	$V_{OUT} = 1.5V \text{ to } 3.5V, \text{ R}_{L} = 500\Omega$ $T_{A} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_{A} = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•	1.0 0.7 0.6	2		V/mV V/mV V/mV
		$ \begin{array}{l} V_{OUT} = 1.5V \text{ to } 3.5V, \ R_L = 100\Omega \\ T_A = 0^\circ \text{C to } 70^\circ \text{C} \\ T_A = -40^\circ \text{C to } 85^\circ \text{C} \end{array} $	• •	0.7 0.5 0.4	1.5		V/mV V/mV V/mV
V <sub>OUT</sub>	Maximum Output Swing (Positive)	$R_L = 500\Omega$ , 30mV Overdrive $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•	3.9 3.8 3.7	4.1		V V V
		$R_L = 100\Omega$ , 30mV Overdrive $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•	3.7 3.6 3.5	3.9		V V V
	Maximum Output Swing (Negative)	$ \begin{array}{l} R_{L} = 500\Omega, \ 30 \text{mV} \ \text{Overdrive} \\ T_{A} = 0^\circ \text{C} \ \text{to} \ 70^\circ \text{C} \\ T_{A} = -40^\circ \text{C} \ \text{to} \ 85^\circ \text{C} \end{array} $	•		0.9	1.1 1.2 1.3	V V V
		$ \begin{array}{l} R_{L} = 100\Omega, \ 30 \text{mV} \ \text{Overdrive} \\ T_{A} = 0^\circ \text{C} \ \text{to} \ 70^\circ \text{C} \\ T_{A} = -40^\circ \text{C} \ \text{to} \ 85^\circ \text{C} \end{array} $	•		1.1	1.3 1.4 1.5	V V V

# **ELECTRICAL CHARACTERISTICS** The • denotes the specifications which apply over the full operating

temperature range, otherwise specifications are at  $T_A = 25^{\circ}C$ .  $V_S = 5V$ ,  $V_{CM} = 2.5V$ ,  $R_L$  to 2.5V, unless otherwise noted. (Note 8)

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
I <sub>OUT</sub>	Maximum Output Current	$V_{OUT} = 1.5V \text{ or } 3.5V, 30\text{mV}$ Overdrive $T_A = 0^{\circ}\text{C} \text{ to } 70^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}$	•	±25 ±20 ±17	±35		mA mA mA
I <sub>SC</sub>	Output Short-Circuit Current	$V_{OUT} = 2.5V$ , 1V Overdrive (Note 3) $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•	±55 ±45 ±40	±75		mA mA mA
SR	Slew Rate	$A_V = -1 \text{ (Note 5)}$ $T_A = 0^\circ \text{C to } 70^\circ \text{C}$ $T_A = -40^\circ \text{C to } 85^\circ \text{C}$	•	200 150 125	350		V/µs V/µs V/µs
FPBW	Full Power Bandwidth	2V <sub>P-P</sub> (Note 6)			55		MHz
GBW	Gain Bandwidth Product	$f = 200 \text{kHz},  \text{R}_{\text{L}} = 500 \Omega$ $T_{\text{A}} = 0^{\circ} \text{C to } 70^{\circ} \text{C}$ $T_{\text{A}} = -40^{\circ} \text{C to } 85^{\circ} \text{C}$	•	65 55 50	94		MHz MHz MHz
-3dB BW	–3dB Bandwidth	$A_V = 1, R_L = 500\Omega$			180		MHz
t <sub>r</sub> , t <sub>f</sub>	Rise Time, Fall Time	$A_V$ = 1, 10% to 90%, 0.1V, $R_L$ = 100 $\Omega$			2.1		ns
t <sub>PD</sub>	Propagation Delay (Note 10)	$A_V$ = 1, 50% to 50%, 0.1V, $R_L$ = 100 $\Omega$			3		ns
OS	Overshoot	$A_V = 1, 0.1V, R_L = 100\Omega$			25		%
ts	Settling Time	$A_V = -1, 0.1\%, 2V$			30		ns
THD	Total Harmonic Distortion	$A_V = 2$ , f = 1MHz, $V_{OUT} = 2V_{P-P}$ , $R_L = 500\Omega$			-75		dB
dG	Differential Gain	$A_V = 2, V_{OUT} = 2V_{P-P}, R_L = 150\Omega$			0.22		%
dP	Differential Phase	$A_V = 2, V_{OUT} = 2V_{P-P}, R_L = 150\Omega$			0.21		DEG
R <sub>OUT</sub>	Output Resistance	$A_V = 1, f = 1MHz$			0.45		Ω
	Channel Separation	$V_{0UT} = 1.5V \text{ to } 3.5V, \text{ R}_{L} = 100\Omega$ $T_{A} = 0^{\circ}\text{C to } 70^{\circ}\text{C}$ $T_{A} = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	•	81 80 79	100		dB dB dB
I <sub>S</sub>	Supply Current	Per Amplifier $T_A = 0^{\circ}C$ to 70°C $T_A = -40^{\circ}C$ to 85°C	•		2.9	4.0 5.0 5.5	mA mA mA

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

**Note 2:** Differential inputs of ±6V are appropriate for transient operation only, such as during slewing. Large sustained differential inputs can cause excessive power dissipation and may damage the part.

**Note 3:** A heat sink may be required to keep the junction temperature below absolute maximum when the output is shorted indefinitely.

**Note 4:** Input offset voltage is pulse tested and is exclusive of warm-up drift.

**Note 5:** Slew rate is measured between  $\pm 2V$  at the output with  $\pm 3V$  input for  $\pm 5V$  supplies and  $2V_{P-P}$  at the output with a  $3V_{P-P}$  input for single 5V supplies.

Note 6: Full power bandwidth is calculated from the slew rate:

FPBW =  $SR/2\pi V_P$ 

Note 7: This parameter is not 100% tested

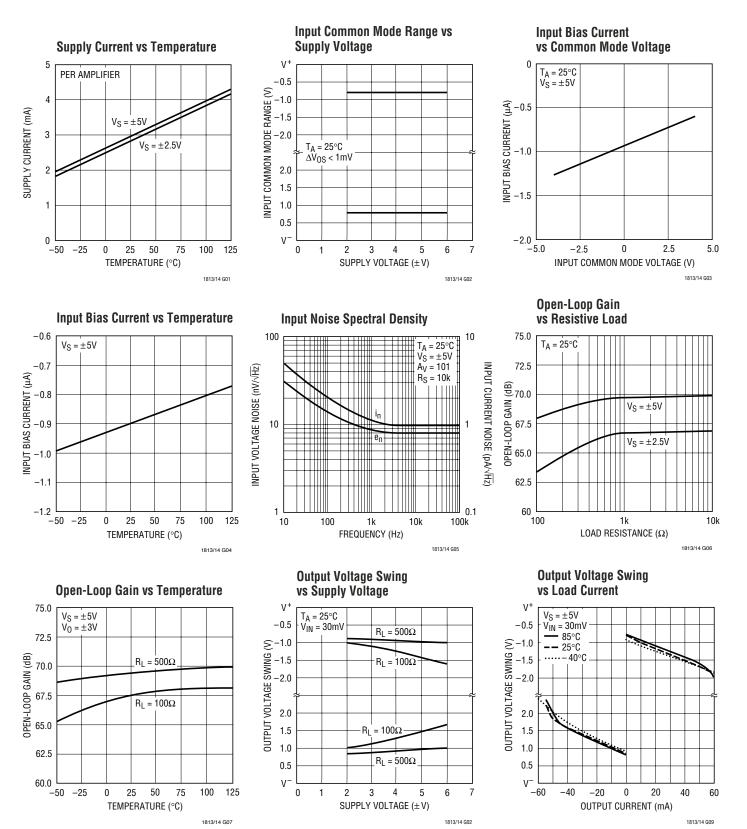
**Note 8:** The LT1813C/LT1814C are guaranteed to meet specified performance from 0°C to 70°C and is designed, characterized and expected to meet the extended temperature limits, but is not tested at -40°C and 85°C. The LT1813I/LT1814I are guaranteed to meet the extended temperature limits.

**Note 9:** The LT1813D is 100% production tested at 25°C. It is designed, characterized and expected to meet the 0°C to 70°C specifications although it is not tested or QA sampled at these temperatures. The LT1813D is guaranteed functional from -40°C to 85°C but may not meet those specifications.

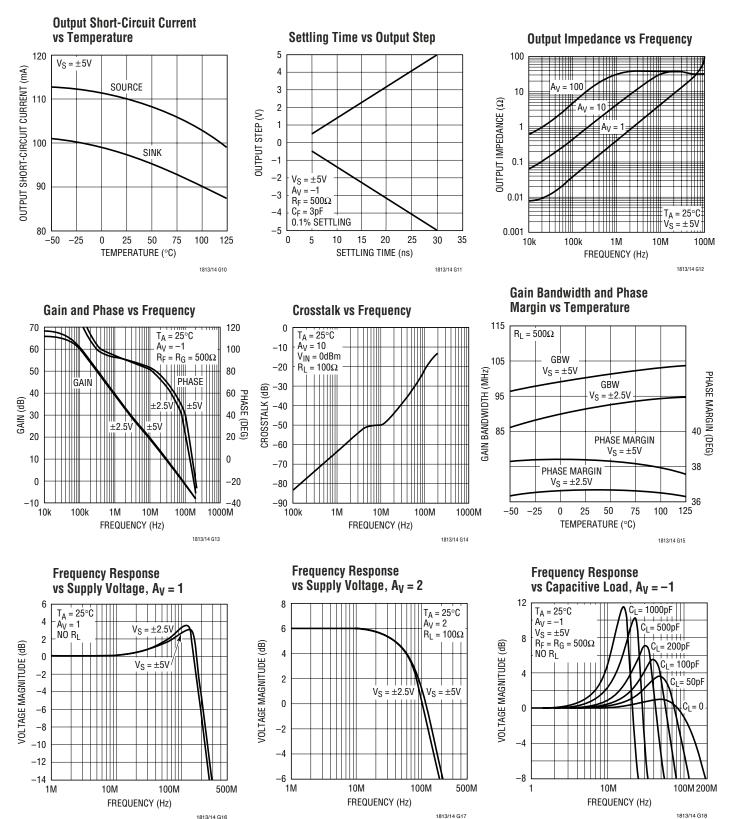
**Note 10:** Propagation delay is measured from the 50% point on the input waveform to the 50% point on the output waveform.



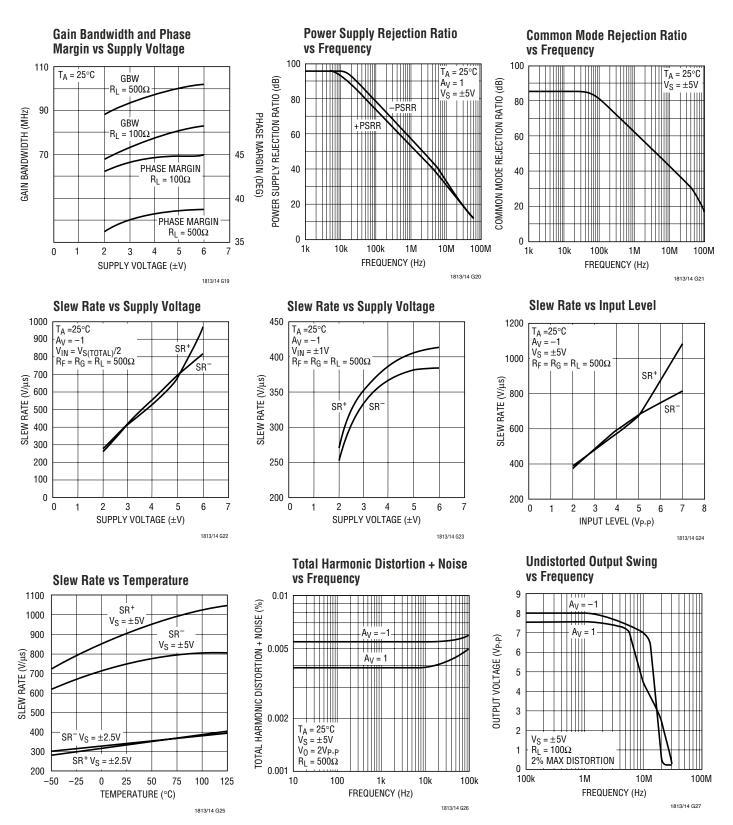




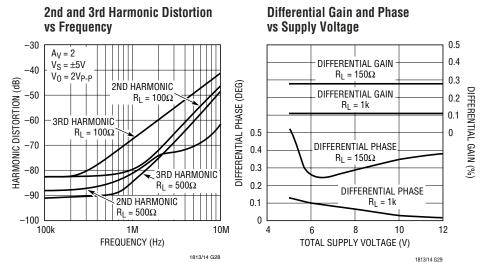
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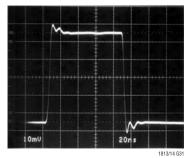




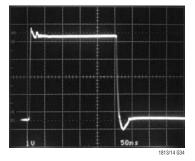




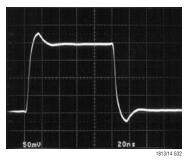
Small-Signal Transient (A<sub>V</sub> = 1)

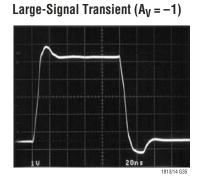


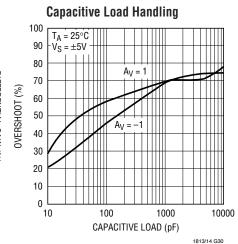
Large-Signal Transient ( $A_V = 1$ )



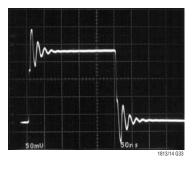




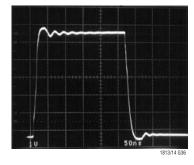




Small-Signal Transient  $(A_V = 1, C_L = 100 pF)$ 



Large-Signal Transient  $(A_V = -1, C_L = 200 pF)$ 





#### **APPLICATIONS INFORMATION**

#### Layout and Passive Components

The LT1813/LT1814 amplifiers are more tolerant of less than ideal board layouts than other high speed amplifiers. For optimum performance, a ground plane is recommended and trace lengths should be minimized, especially on the negative input lead.

Low ESL/ESR bypass capacitors should be placed directly at the positive and negative supply pins ( $0.01\mu$ F ceramics are recommended). For high drive current applications, additional  $1\mu$ F to  $10\mu$ F tantalums should be added.

The parallel combination of the feedback resistor and gain setting resistor on the inverting input combine with the input capacitance to form a pole that can cause peaking or even oscillations. If feedback resistors greater than 1k are used, a parallel capacitor of value:

 $C_F > R_G \bullet C_{IN}/R_F$ 

should be used to cancel the input pole and optimize dynamic performance. For applications where the DC noise gain is 1 and a large feedback resistor is used,  $C_F$  should be greater than or equal to  $C_{IN}$ . An example would be an I-to-V converter.

#### **Input Considerations**

The inputs of the LT1813/LT1814 amplifiers are connected to the base of an NPN and PNP bipolar transistor in parallel. The base currents are of opposite polarity and provide first order bias current cancellation. Due to variation in the matching of NPN and PNP beta, the polarity of the input bias current can be positive or negative. The offset current, however, does not depend on beta matching and is tightly controlled. Therefore, the use of balanced source resistance at each input is recommended for applications where DC accuracy must be maximized. For example, with a 100 $\Omega$  source resistance at each input, the 400nA maximum offset current results in only 40 $\mu$ V of extra offset, while without balance the 4 $\mu$ A maximum input bias current could result in a 0.4mV offset contribution.

The inputs can withstand differential input voltages of up to 6V without damage and without needing clamping or

series resistance for protection. This differential input voltage generates a large internal current (up to 40mA), which results in the high slew rate. In normal transient closed-loop operation, this does not increase power dissipation significantly because of the low duty cycle of the transient inputs. Sustained differential inputs, however, will result in excessive power dissipation and therefore **this device should not be used as a comparator.** 

#### **Capacitive Loading**

The LT1813/LT1814 are stable with capacitive loads from 0pF to 1000pF, which is outstanding for a 100MHz amplifier. The internal compensation circuitry accomplishes this by sensing the load induced output pole and adding compensation at the amplifier gain node as needed. As the capacitive load increases, both the bandwidth and phase margin decrease so there will be peaking in the frequency domain and ringing in the transient response. Coaxial cable can be driven directly, but for best pulse fidelity a resistor of value equal to the characteristic impedance of the cable (e.g.,  $75\Omega$ ) should be placed in series with the output. The receiving end of the cable should be terminated with the same value resistance to ground.

#### **Slew Rate**

The slew rate of the LT1813/LT1814 is proportional to the differential input voltage. Highest slew rates are therefore seen in the lowest gain configurations. For example, a 5V output step in a gain of 10 has a 0.5V input step, whereas in unity gain there is a 5V input step. The LT1813/LT1814 is tested for a slew rate in a gain of -1. Lower slew rates occur in higher gain configurations.

#### **Power Dissipation**

The LT1813/LT1814 combine two or four amplifiers with high speed and large output drive in a small package. It is possible to exceed the maximum junction temperature specification under certain conditions. Maximum junction temperature ( $T_J$ ) is calculated from the ambient temperature ( $T_A$ ) and power dissipation ( $P_D$ ) as follows:

$$\mathsf{T}_\mathsf{J} = \mathsf{T}_\mathsf{A} + (\mathsf{P}_\mathsf{D} \bullet \Theta_\mathsf{J}_\mathsf{A})$$

### APPLICATIONS INFORMATION

Power dissipation is composed of two parts. The first is due to the quiescent supply current and the second is due to on-chip dissipation caused by the load current. The worst-case load induced power occurs when the output voltage is at 1/2 of either supply voltage (or the maximum swing if less than 1/2 the supply voltage). Therefore P<sub>DMAX</sub> is:

$$\begin{split} \mathsf{P}_{\mathsf{DMAX}} &= (\mathsf{V}^+ - \mathsf{V}^-) \bullet (\mathsf{I}_{\mathsf{SMAX}}) + (\mathsf{V}^+/2)^2 / \mathsf{R}_{\mathsf{L}} \text{ or} \\ \mathsf{P}_{\mathsf{DMAX}} &= (\mathsf{V}^+ - \mathsf{V}^-) \bullet (\mathsf{I}_{\mathsf{SMAX}}) + (\mathsf{V}^+ - \mathsf{V}_{\mathsf{OMAX}}) \bullet (\mathsf{V}_{\mathsf{OMAX}} / \mathsf{R}_{\mathsf{L}}) \end{split}$$

Example: LT1814S at 70°C, V<sub>S</sub> =  $\pm$ 5V, R<sub>L</sub>=100 $\Omega$ 

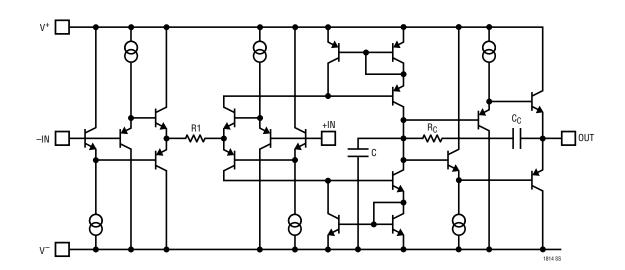
 $P_{DMAX} = (10V) \bullet (4.5mA) + (2.5V)^2/100\Omega = 108mW$  $T_{IMAX} = 70^{\circ}C + (4 \bullet 108mW) \bullet (100^{\circ}C/W) = 113^{\circ}C$ 

#### **Circuit Operation**

The LT1813/LT1814 circuit topology is a true voltage feedback amplifier that has the slewing behavior of a current feedback amplifier. The operation of the circuit can be understood by referring to the Simplified Schematic. Complementary NPN and PNP emitter followers buffer the inputs and drive an internal resistor. The input voltage appears across the resistor, generating current that is mirrored into the high impedance node. Complementary followers form an output stage that buffers the gain node from the load. The input resistor, input stage transconductance, and the capacitor on the high impedance node determine the bandwidth. The slew rate is determined by the current available to charge the gain node capacitance. This current is the differential input voltage divided by R1, so the slew rate is proportional to the input step. Highest slew rates are therefore seen in the lowest gain configurations.

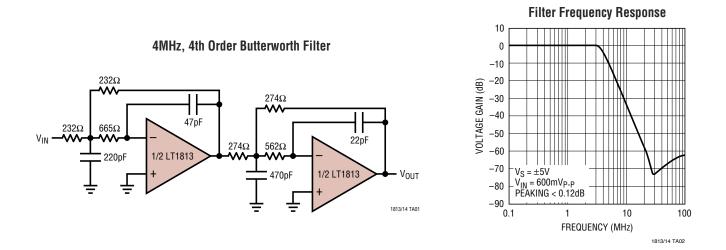
The RC network across the output stage is bootstrapped when the amplifier is driving a light or moderate load and has no effect under normal operation. When a heavy load (capacitive or resistive) is driven, the network is incompletely bootstrapped and adds to the compensation at the high impedance node. The added capacitance moves the unity-gain frequency away from the pole formed by the output impedance and the capacitive load. The zero created by the RC combination adds phase to ensure that the total phase lag does not exceed 180° (zero phase margin), and the amplifier remains stable. In this way, the LT1813/ LT1814 are stable with up to 1000pF capacitive loads in unity gain, and even higher capacitive loads in higher closed-loop gain configurations.

#### SIMPLIFIED SCHEMATIC (one amplifier)

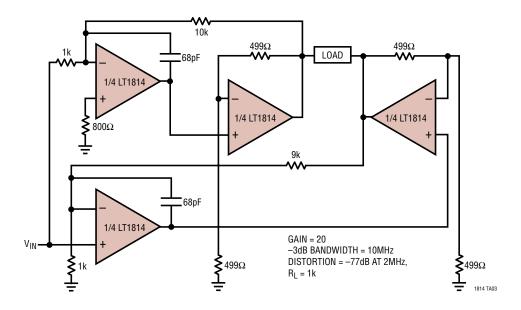




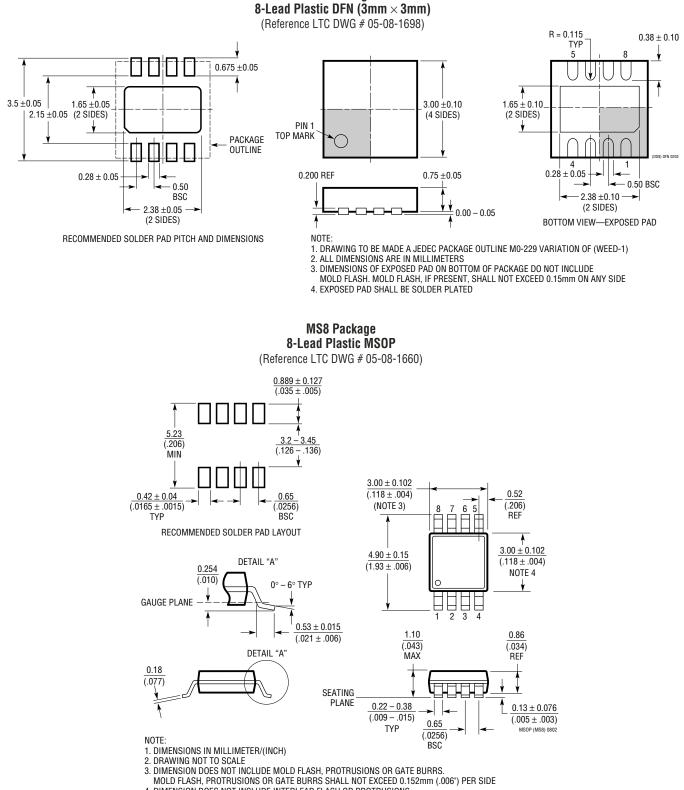
#### **TYPICAL APPLICATION**



Gain of 20 Composite Amplifier Drives Differential Load with Low Distortion



#### PACKAGE DESCRIPTION



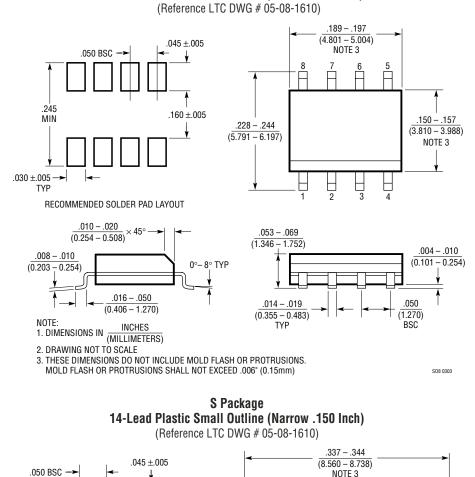
**DD** Package

4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.

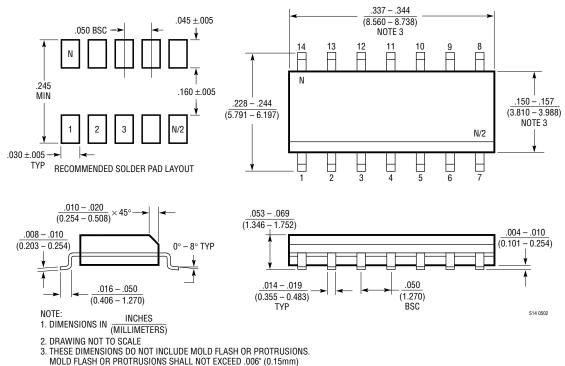
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006") PER SIDE 5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004") MAX



#### PACKAGE DESCRIPTION

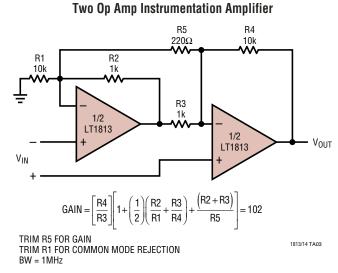


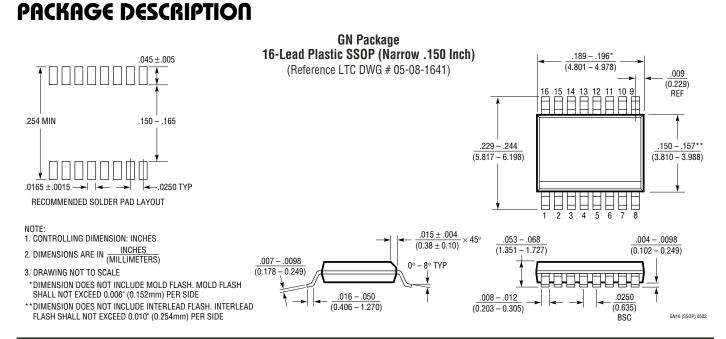
S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)



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#### TYPICAL APPLICATION





## **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LT1363/LT1364/LT1365	Single/Dual/Quad 70MHz, 1000V/µs, C-Load™ Op Amps	$\pm 2.5V$ to $\pm 15V$ Operation
LT1395/LT1396/LT1397	Single/Dual/Quad 400MHz Current Feedback Amplifiers	4.6mA Supply Current, 800V/µs, 80mA Output Current
LT1806/LT1807	Single/Dual 325MHz, 140V/µs Rail-to-Rail I/O Op Amps	Low Noise 3.5nV/√Hz
LT1809/LT1810	Single/Dual 180MHz, 350V/µs Rail-to-Rail I/O Op Amps	Low Distortion –90dBc at 5MHz
LT1812	Single 3mA, 100MHz, 750V/µs Op Amp	Single Version of LT1813/LT1814; 50µA Shutdown Option
LT1815/LT1816/LT1817	Single/Dual/Quad 220MHz, 1500V/µs Op Amps	6.5mA Supply Current, 6nV/√Hz Input Noise
C-Load is a tradomark of L	inear Technology Corporation	

C-Load is a trademark of Linear Technology Corporation.

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