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LT3461/LT3461A

1.3MHz/3MHz Step-Up DC/DC Converters with Integrated Schottky in ThinSOT

The LT®3461/LT3461A are general purpose fixed frequency

current mode step-up DC/DC converters. Both devices

feature an integrated Schottky and a low V_{CESAT} switch al-

lowing a small converter footprint and lower parts cost. The

LT3461 switches at 1.3MHz while the LT3461A switches

at 3MHz. These high switching frequencies enable the use

of tiny, low cost and low height capacitors and inductors.

The constant switching frequency results in predictable

output noise that is easy to filter, and the inductor based

topology ensures an input free from switching noise typi-

cally present with charge pump solutions. The high voltage switch in the LT3461/LT3461A is rated at 40V making the

The LT3461/LT3461A are available in a low profile (1mm)

∠, LT, LTC, LTM, Linear Technology and the Linear logo are registered trademarks and

device ideal for boost converters up to 38V.

ThinSOT is a trademark of Linear Technology Corporation.

DESCRIPTION

SOT-23 package.

FEATURES

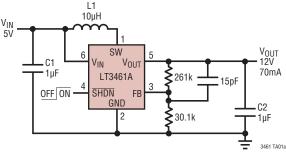
- Integrated Schottky Rectifier
- Fixed Frequency 1.3MHz/3MHz Operation
- High Output Voltage: Up to 38V
- Low V_{CESAT} Switch: 260mV at 250mA
- 12V at 70mA from 5V Input
- 5V at 115mA from 3.3V Input
- Wide Input Range: 2.5V to 16V
- Uses Small Surface Mount Components
- Low Shutdown Current: <1µA</p>
- Soft-Start
- Low Profile (1mm) SOT-23 (ThinSOT™) Package

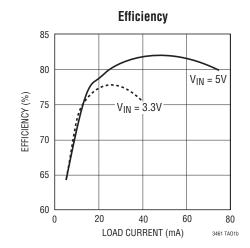
APPLICATIONS

- Digital Cameras
- CCD Bias Supply
- XDSL Power Supply
- TFT-LCD Bias Supply
- Local 5V or 12V Supply
- Medical Diagnostic Equipment
- Battery Backup

TYPICAL APPLICATION

5V to 12V, 70mA Step-Up DC/DC Converter

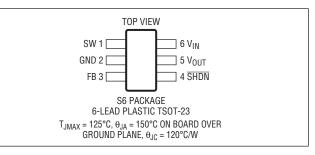




ABSOLUTE MAXIMUM RATINGS

(Note	1)
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PIN CONFIGURATION



ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3461AES6#PBF	LT3461AES6#TRPBF	LTAHG	8-Lead Plastic TSOT-23	-40°C to 85°C
LT3461ES6#PBF	LT3461ES6#TRPBF	LTAEB	8-Lead Plastic TSOT-23	-40°C to 85°C

Consult LTC Marketing for parts specified with wider operating temperature ranges.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/. Some packages are available in 500 unit reels through designated sales channels with #TRMPBF suffix.



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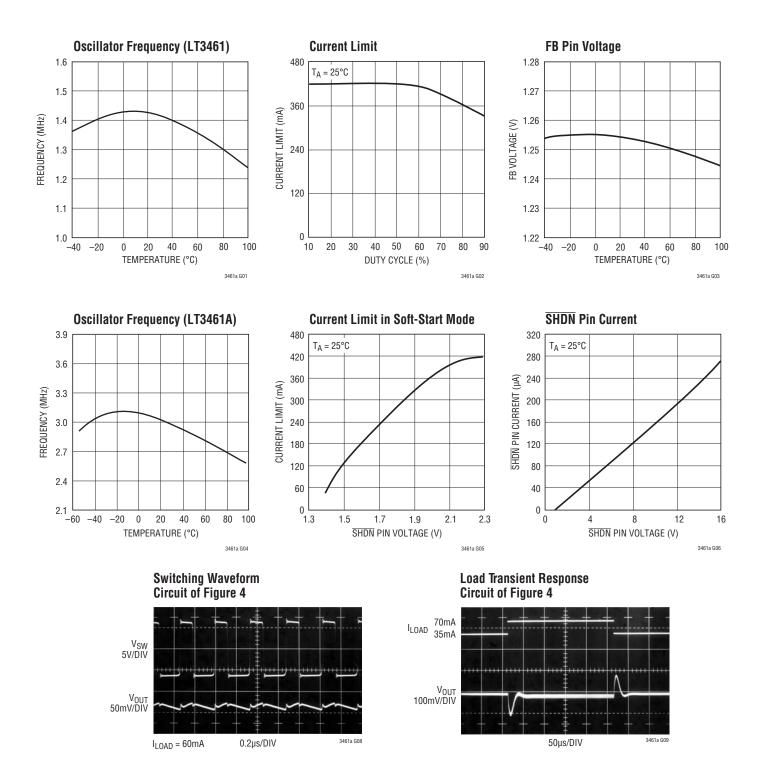
ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T_A = 25°C, V_{IN} = 3V, V_{SHDN} = 3V, unless otherwise noted.

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Minimum Operating Voltage			2.5			V
Maximum Operating Voltage					16	V
Feedback Voltage		•	1.235 1.225	1.255	1.275 1.280	V V
Feedback Line Regulation			0.005		%/V	
FB Pin Bias Current		•		40	100	nA
Supply Current	$\frac{FB = 1.3V, Not Switching}{SHDN} = 0V$			2.8 0.1	3.6 0.5	mA μA
Switching Frequency (LT3461A)		•	2.1	3.0	3.9	MHz
Switching Frequency (LT3461)		•	1.0	1.3	1.7	MHz
Maximum Duty Cycle (LT3461A)		•	82			%
Maximum Duty Cycle (LT3461)		•	92			%
Switch Current Limit			300	420	600	mA
Switch V _{CESAT}	I _{SW} = 250mA			260	350	mV
Switch Leakage Current	V _{SW} = 5V			0.01	1	μA
Schottky Forward Voltage	I _{SCHOTTKY} = 250mA			800	1100	mV
Schottky Reverse Leakage	$V_{OUT} - SW = 40V$			0.03	4	μA
SHDN Voltage High			1.5			V
SHDN Voltage Low					0.4	V
SHDN Pin Bias Current				35	50	μA

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The LT3461E/LT3461AE is guaranteed to meet specifications from 0°C to 70°C. Specifications over the -40°C to 85°C operating temperature range are assured by design, characterization and correlation with statistical process controls.

TYPICAL PERFORMANCE CHARACTERISTICS



LINEAR

PIN FUNCTIONS

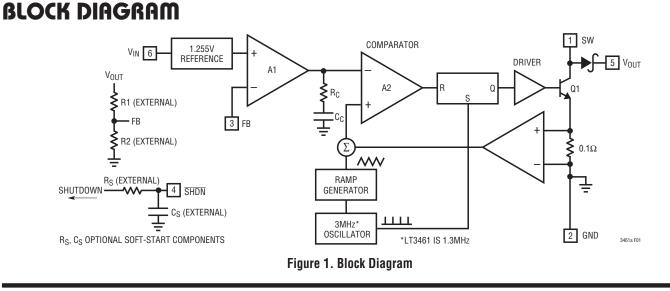
SW (Pin 1): Switch Pin. Connect inductor here. Minimize trace at this pin to reduce EMI.

GND (Pin 2): Ground Pin. Tie directly to local ground plane.

FB (Pin 3): Feedback Pin. Reference voltage is 1.255V. Connect resistor divider tap here. Minimize trace area at FB. Set V_{OUT} according to V_{OUT} = 1.255V (1 + R1/R2). **SHDN** (Pin 4): Shutdown Pin. Tie to 1.5V or higher to enable device; 0.4V or less to disable device. Also functions as soft-start. Use RC filter (47k, 47nF typ) as shown in Figure 1.

V_{OUT} (Pin 5): Output Pin. Connect to resistor divider. Put capacitor close to pin and close to GND plane.

VIN (Pin 6): Input Supply Pin. Must be locally bypassed.



OPERATION

The LT3461/LT3461A uses a constant frequency, current mode control scheme to provide excellent line and load regulation. Operation can be best understood by referring to the block diagram in Figure 1. At the start of each oscillator cycle, the SR latch is set, which turns on the power switch Q1. A voltage proportional to the switch current is added to a stabilizing ramp and the resulting sum is fed into the positive terminal of the PWM comparator A2. When this voltage exceeds the level at the negative input of A2, the SR latch is reset turning off the power switch. The level at the negative input of A2 is set by the error amplifier A1, and is simply an amplified version of the difference between the feedback voltage and the reference voltage of 1.255V. In this manner, the error amplifier sets the correct peak current level to keep the output in regulation. If the error amplifier's output increases, more current is delivered to the output; if it decreases, less current is delivered.

Layout Hints

The high speed operation of the LT3461/LT3461A demands careful attention to board layout. You will not get advertised performance with careless layout. Figure 2 shows the recommended component placement.

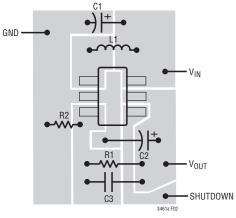


Figure 2. Suggested Layout

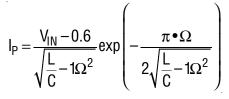




APPLICATIONS INFORMATION

Inrush Current

The LT3461 has a built-in Schottky diode. When supply voltage is applied to the $V_{\mbox{\scriptsize IN}}$ pin, the voltage difference between V_{IN} and V_{OUT} generates inrush current flowing from input through the inductor and the Schottky diode to charge the output capacitor. The maximum nonrepetitive surge current the Schottky diode in the LT3461 can sustain is 1.5A. The selection of inductor and capacitor value should ensure the peak of the inrush current to be below 1.5A. In addition, turn-on of the LT3461 should be delayed until the inrush current is less than the maximum current limit. The peak inrush current can be calculated as follows:



where L is the inductance. r is the resistance of the inductor and C is the output capacitance.

Table 3 gives inrush peak currents for some component selections.

Table 3. Inrush Peak Current				
V _{IN} (V)	L (µH)	C (µF)	I _P (A)	
5	4.7	1	1.1	
5	10	1	0.9	
5	10	1	0.9	

Thermal Considerations

Significant power dissipation can occur on the LT3461 and LT3461A, particularly at high input voltage. Device load, voltage drops in the power path components, and switching losses are the major contributors. It is important to measure device power dissipation in an application to ensure that the LT3461 does not exceed the absolute maximum operating junction temperature of 125°C over the operating ambient temperature range. Generally, for supply voltages below 5V the integrated current limit function provides adequate protection for nonfault conditions. For supply voltages above 5V, Figures 3a and 3b show the recommended operating region of the LT3461 and LT3461A, respectively. These graphs are based on 250mW on-chip dissipation. Improvement of these numbers can be expected if the LT3461 is supplied from a separate low voltage rail.

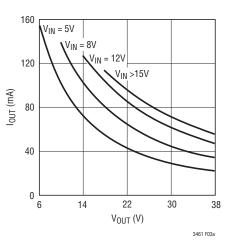


Figure 3a. LT3461 Operating Region

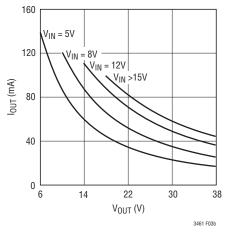


Figure 3b. LT3461A Operating Region

Switching Frequency

The key difference between the LT3461 and LT3461A is the faster switching frequency of the LT3461A. At 3MHz, the LT3461A switches at twice the rate of the LT3461. The higher switching frequency of the LT3461A allows physically smaller inductors and capacitors to be used in a given application, but with a slight decrease in efficiency and maximum output current when compared to the LT3461. Generally if efficiency and maximum output current are crucial, or a high output voltage is being generated, the LT3461 should be used. If application size and cost are more important, the LT3461A will be the better choice.



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APPLICATIONS INFORMATION

Inductor Selection

The inductors used with the LT3461/LT3461A should havea saturation current rating of 0.3A or greater. If the device is used in an application where the input supply will be hot-plugged, then the saturation current rating should be equal to or greater than the peak inrush current. For the LT3461, an inductor value between 10 μ H and 47 μ H, depending upon output voltage, will usually be the best choice for most designs. For the LT3461A, inductor values between 4.7 μ H and 15 μ H inductor will suffice for most applications. For best loop stability results, the inductor value selected should provide a ripple current of 70mA or more. For a given V_{IN} and V_{OUT} the inductor value to use with LT3461A is estimated by the formula:

L (in microhenries) =
$$\frac{D \bullet V_{IN} \bullet \sqrt{V_{OUT}} \bullet 1 \text{sec}}{1A \bullet \sqrt{1V}}$$

where D = $\frac{V_{OUT} + 1V - V_{IN}}{V_{OUT} + 1V}$

Use twice this value for the LT3461.

TYPICAL APPLICATIONS

Capacitor Selection

Low ESR capacitors should be used at the output to minimize the output voltage ripple. Multilayer ceramic capacitors using X5R/X7R dielectrics are preferred as

they have a low ESR and maintain capacitance over wide voltage and temperature range. A 2.2μ F output capacitor is sufficient for most applications using the LT3461, while a 1 μ F capacitor is sufficient for most applications using the LT3461A. High output voltages typically require less capacitance for loop stability. Always use a capacitor with sufficient voltage rating.

Either ceramic or solid tantalum capacitors may be used for the input decoupling capacitor, which should be placed as close as possible to the LT3461/LT3461A. A 1 μ F capacitor is sufficient for most applications.

Phase Lead Capacitor

A small value capacitor can be added across resistor R1 between the output and the FB pin to reduce output perturbation due to a load step and to improve transient response. This phase lead capacitor introduces a pole-zero pair to the feedback that boosts phase margin near the cross-over frequency. The following formula is useful to estimate the capacitor value needed:

$$C_{PL} = \frac{500k\Omega}{R2} \bullet 1pF$$

For an application running $50\mu A$ in the feedback divider, capacitor values from 10pF to 22pF work well.

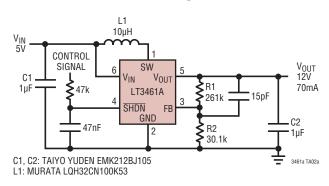
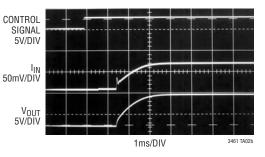


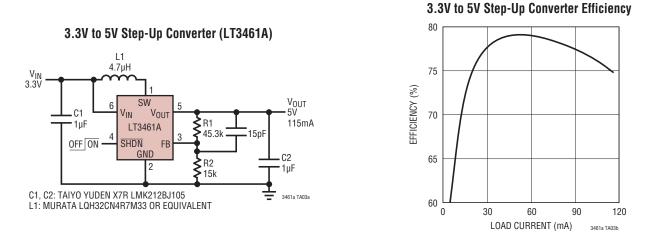
Figure 4. 5V to 12V with Soft-Start Circuit (LT3461A)

Input Current and Output Voltage





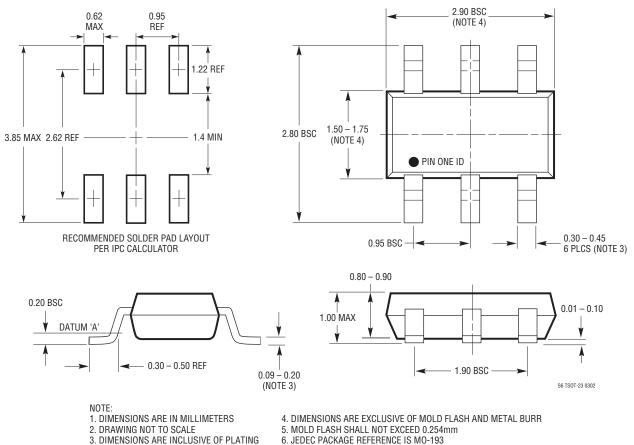
TYPICAL APPLICATIONS



PACKAGE DESCRIPTION

Please refer to http://www.linear.com/product/LT3461#packaging for the most recent package drawings.

S6 Package 6-Lead Plastic TSOT-23 (Reference LTC DWG # 05-08-1636)



3. DIMENSIONS ARE INCLUSIVE OF PLATING

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REVISION HISTORY (Revision history begins at Rev B)

REV	DATE	DESCRIPTION	PAGE NUMBER
А	01/16	Modified inrush current I _P equation.	6

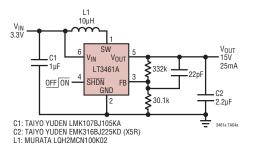




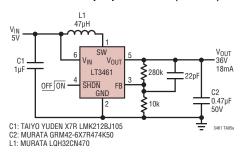


TYPICAL APPLICATION

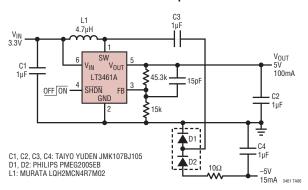
Low Profile (1mm) 3.3V to 15V Step-Up Converter



5V to 36V Step-Up Converter (LT3461)



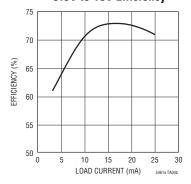
3.3V to ±5V Dual Output Converter



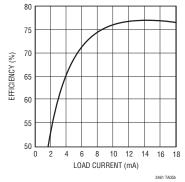
RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1615/LT1615-1	300mA/80mA (I _{SW}) Constant Off-Time, High Efficiency Step-Up DC/DC Converter	V_{IN} : 1.2V to 15V, $V_{OUT(MAX)}$ = 34V, I_Q = 20µA, I_{SD} <1µA, ThinSOT Package
LT1944/LT1944-1	Dual Output 350mA/100mA (I _{SW}), Constant Off-Time, High Efficiency Step-Up DC/DC Converter	V_{IN} : 1.2V to 15V, $V_{OUT(MAX)}$ = 34V, I_Q = 20µA, I_{SD} <1µA, MS Package
LTC3400/LTC3400B	600mA (I _{SW}), 1.2MHz, Synchronous Step-Up DC/DC Converter	V_{IN} : 0.85V to 5V, $V_{OUT(MAX)}$ = 5V, I_Q = 19µA/300µA, I_{SD} <1µA, ThinSOT Package
LT3460	T3460 0.32A (I _{SW}), 1.3MHz, High Efficiency Step-Up DC/DC Converter V _{IN} : 2.5V to 16V, V _{OUT(MAX)} = 36V, I _Q = 2mA, I _{SD} < SC70, ThinSOT Packages	
LT3465/LT3465A	Constant Current, 1.2MHz/2.7MHz, High Efficiency White LED Boost Regulator with Integrated Schottky Diode	V_{IN} : 2.7V to 16V, $V_{OUT(MAX)}$ = 34V, I_Q = 1.9mA, I_{SD} <1 $\mu A,$ ThinSOT Package

3.3V to 15V Efficiency



5V to 36V Efficiency



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