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### LTC6363 Family

### Precision, Low Power Differential Amplifier/ADC Driver Family

- <sup>n</sup> **Available with User Set Gain or Fixed Gain of 0.5V/V, 1V/V, or 2V/V**
- <sup>n</sup> **2.9nV/√Hz Input-Referred Noise**
- 2mA Maximum Supply Current
- <sup>n</sup> **45ppm Max Gain Error**
- 0.5ppm/<sup>o</sup>C Max Gain Error Drift
- <sup>n</sup> **94dB Min CMRR**
- 100uV Max Offset Voltage
- 50nA Max Input Offset Current
- Fast Settling: 720ns to 18-Bit, 8V<sub>P-P</sub> Output
- 2.8V  $(\pm 1.4V)$  to 11V  $(\pm 5.5V)$  Supply Voltage Range
- Differential Rail-to-Rail Outputs
- Input Common Mode Range Includes Ground
- Low Distortion: 115dB SFDR at 2kHz, 18V<sub>P-P</sub><br>■ 500MHz Gain-Bandwidth Product
- <sup>n</sup> 500MHz Gain-Bandwidth Product
- $\Box$  35MHz –3dB Bandwidth
- **n** Low Power Shutdown:  $20\mu$ A (V<sub>S</sub> = 3V)
- $\blacksquare$  8-lead MSOP and 2mm  $\times$  3mm 8-Lead DFN Packages

### **APPLICATIONS**

- 20-Bit, 18-Bit and 16-Bit SAR ADC Drivers
- Single-Ended-to-Differential Conversion
- **Low Power ADC Drivers**
- Level Shifter
- Differential Line Drivers
- Battery-Powered Instrumentation

### TYPICAL APPLICATION



### FEATURES DESCRIPTION

The LTC®6363 family consists of four fully differential, low power, low noise amplifiers with rail-to-rail outputs optimized to drive SAR ADCs. The LTC6363 is a stand-alone differential amplifier, where the gain is typically set using four external resistors. The LTC6363-0.5, LTC6363-1, and LTC6363-2 each have internal matched resistors to create fixed gain blocks with gains of 0.5V/V, 1V/V, and 2V/V respectively. Each of the fixed-gain amplifiers features precision laser trimmed on-chip resistors for accurate, ultrastable gain and excellent CMRR.

#### **Family Selection Table**



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#### **LTC6363-1 Driving LTC2378-20 fIN = 2kHz, –1dBFS,131k-Point FFT**



1 6363fb 6363 TA01b

### ABSOLUTE MAXIMUM RATINGS **(Note 1)**





### PIN CONFIGURATION



### ORDER INFORMATION **http://www.linear.com/product/LTC6363#orderinfo**



### ORDER INFORMATION

#### **Lead Free Finish**



TRM = 500 pieces. \*Temperature grades are identified by a label on the shipping container.

Consult ADI Marketing for parts specified with wider operating temperature ranges.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/. Some packages are available in 500 unit reels through designated sales channels with #TRMPBF suffix.

#### **ELECTRICAL CHARACTERISTICS** Complete LTC6363 Family. The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications and typical values are at  $T_A = 25^{\circ}$ C. V<sup>+</sup> = 10V, V<sup>-</sup> = 0V, V<sub>CM</sub> =  $V_{\text{OCM}} = V_{\text{ICM}} = 5V$ ,  $V_{\overline{\text{SHDN}}} = \text{open}$ .  $V_{\text{S}}$  is defined as (V<sup>+</sup> – V<sup>-</sup>).  $V_{\text{OUTCM}}$  is defined as (V<sub>+OUT</sub> + V<sub>-OUT</sub>)/2.  $V_{\text{ICM}}$  is defined as (V<sub>+IN</sub> + V<sub>-IN</sub>)/2. **VOUTDIFF is defined as (V+OUT – V–OUT).**



### **ELECTRICAL CHARACTERISTICS** Complete LTC6363 Family. The  $\bullet$  denotes the specifications which apply

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LTC6363 Only. The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications and typical values are at T<sub>A</sub> = 25°C. V<sup>+</sup> = 10V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>OCM</sub> = V<sub>ICM</sub> = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V<sup>+</sup> – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>–OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>–IN</sub>)/2. V<sub>OUTDIFF</sub> is defined as (V<sub>+OUT</sub> – V<sub>–OUT</sub>). Typical specifications apply to **the internal amplifier inside all versions.**



**LTC6363 Only. The**  $\bullet$  **denotes the specifications which apply over the full operating temperature range, otherwise specifications and typical values are at TA = 25°C. V<sup>+</sup> = 10V, V– = 0V, VCM = VOCM = VICM =**  5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V<sup>+</sup> – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>-OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>-IN</sub>)/2. V<sub>OUTDIFF</sub> is **defined as (V+OUT – V–OUT). Typical specifications apply to the internal amplifier inside all versions.**



**LTC6363-0.5 Only. The**  $\bullet$  **denotes the specifications which apply over the full** operating temperature range, otherwise specifications and typical values are at T<sub>A</sub> = 25°C. V<sup>+</sup> = 10V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>OCM</sub> = V<sub>ICM</sub> = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V<sup>+</sup> – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>-OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>-IN</sub>)/2. V<sub>OUTDIFF</sub> is defined **as (V+OUT – V–OUT).**



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**LTC6363-0.5 Only. The**  $\bullet$  **denotes the specifications which apply over the full** operating temperature range, otherwise specifications and typical values are at T<sub>A</sub> = 25°C. V<sup>+</sup> = 10V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>OCM</sub> = V<sub>ICM</sub> = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V<sup>+</sup> – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>–OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>–IN</sub>)/2. V<sub>OUTDIFF</sub> is defined  $\overline{as}$   $(\overline{V_{+}}_{01} - \overline{V_{-}}_{01})$ .

<b>SYMBOL</b>	<b>PARAMETER</b>	<b>CONDITIONS</b>		MIN	<b>TYP</b>	<b>MAX</b>	<b>UNITS</b>
<sub>sc</sub>	Output Short-Circuit Current, Either Output Pin, Sinking	$V_S = 3V$ , Output Shorted to 1.5V $V_S = 5V$ , Output Shorted to 2.5V $V_S = 10V$ , Output Shorted to 5V		12 13 14	25 35 40		mA mA mA
	Output Short-Circuit Current, Either Output Pin, Sourcing	$V_S = 3V$ , Output Shorted to 1.5V $V_S = 5V$ , Output Shorted to 2.5V $V_S = 10V$ , Output Shorted to 5V	$\bullet$	25 27 30	55 75 90		mA mA mA
$t_{-3dB}$	–3dB Bandwidth				35		<b>MHz</b>
SR	<b>Slew Rate</b>	Differential 18V <sub>P-P</sub> Output			44		$V/\mu s$
FPBW (Note 11)	<b>Full Power Bandwidth</b>	10V <sub>P-P</sub> Output 18V <sub>P-P</sub> Output			1.4 0.8		<b>MHz</b> <b>MHz</b>
HD2/HD3	2nd/3rd order Harmonic Distortion Single-Ended Input	$f = 1$ kHz, $V_{OUT} = 10V_{P-P}$ $f = 10kHz$ , $V_{OIII} = 10V_{P-P}$ $f = 100$ kHz, $V_{OIII} = 10V_{P-P}$			$-125/-122$ $-108/-111$ –87/–78		dBc dBc dBc
ts	Settling Time to a 8V <sub>P-P</sub> Output Step	0.1% 0.01% $0.0015\%$ (16-Bit) 4ppm (18-Bit)			420 440 550 740		ns ns ns ns

LTC6363-1 Only. The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications and typical values are at T<sub>A</sub> = 25°C. V\* = 10V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>OCM</sub> = V<sub>ICM</sub> = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V\* – V<sup>-</sup>). V<sub>OUTCM</sub> is defined **as (V+OUT + V–OUT)/2. VICM is defined as (V+IN + V–IN)/2. VOUTDIFF is defined as (V+OUT – V–OUT).**



**LTC6363-1 Only. The**  $\bullet$  **denotes the specifications which apply over the full** operating temperature range, otherwise specifications and typical values are at T<sub>A</sub> = 25°C. V<sup>+</sup> = 10V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>OCM</sub> = V<sub>ICM</sub> = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V<sup>+</sup> – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>–OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>–IN</sub>)/2. V<sub>OUTDIFF</sub> is defined **as (V+OUT – V–OUT).**



**LTC6363-2 Only. The**  $\bullet$  **denotes the specifications which apply over the full operating temperature range, otherwise specifications and typical values are at TA = 25°C. V<sup>+</sup> = 10V, V– = 0V, VCM = VOCM = VICM** = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V\* – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>-OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>-IN</sub>)/2. V<sub>OUTDIFF</sub> is **defined as (V+OUT – V–OUT).**



9

**LTC6363-2 Only. The**  $\bullet$  **denotes the specifications which apply over the full operating temperature range, otherwise specifications and typical values are at TA = 25°C. V<sup>+</sup> = 10V, V– = 0V, VCM = VOCM = VICM** = 5V, V<sub>SHDN</sub> = open. V<sub>S</sub> is defined as (V\* – V<sup>-</sup>). V<sub>OUTCM</sub> is defined as (V<sub>+OUT</sub> + V<sub>-OUT</sub>)/2. V<sub>ICM</sub> is defined as (V<sub>+IN</sub> + V<sub>-IN</sub>)/2. V<sub>OUTDIFF</sub> is  $\det(\mathbf{C} \cdot \mathbf{C}) = \det(\mathbf{C} \cdot \mathbf{C}) - \det(\mathbf{C} \cdot \mathbf{C})$ .



**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

**Note 2:** Absolute Maximum input voltage for the LTC6363-0.5/LTC6363-1/ LTC6363-2 is conservatively calculated assuming the worst case output voltage. For details on this calculation refer to the Input Pin Protection section.

**Note 3:** In the LTC6363, if input pins  $(+1N, -1N, V_{OCM}$  and  $\overline{SHDN}$ ) must exceed either supply voltage, the input current must be limited to less than 10mA. Additionally, if the differential input voltage exceeds 1.4V, the input current must be limited to less than 10mA. In the LTC6363-0.5/LTC6363-1/ LTC6363-2 versions, the same limits apply to  $V_{\text{OCM}}$ ,  $\overline{\text{SHDN}}$  and the internal amplifier's inputs. Please see the Input Common Mode Voltage Range and Input Pin Protection sections for additional details on calculating the input voltages on the internal amplifier's inputs while in a feedback configuration.

**Note 4:** A heat sink may be required to keep the junction temperature below the absolute maximum rating when the output is shorted indefinitely.

**Note 5:** The LTC6363I and LTC6363I-0.5/LTC6363I-1/LTC6363I-2 are guaranteed functional over the operating temperature range of –40°C to 85°C. The LTC6363H and LTC6363H-0.5/LTC6363H-1/LTC6363H-2 are guaranteed functional over the operating temperature range of –40°C to 125°C.

**Note 6:** The LTC6363I and LTC6363I-0.5/LTC6363I-1/LTC6363I-2 are guaranteed to meet specified performance from –40°C to 85°C. The LTC6363H and LTC6363H-0.5/LTC6363H-1/LTC6363H-2 are guaranteed to meet specified performance from –40°C to 125°C.

**Note 7:** Differential offset voltage, differential offset voltage drift, and PSRR are referred to the internal amplifier's input (summing junction) to allow for direct comparison of gain blocks with discrete amplifiers. CMRRI, CMRRIO and voltage noise are referenced to the LTC6363-0.5, LTC6363-1 and LTC6363-2's input pins. Refer to the Test Circuits section for more details.

**Note 8:** Maximum differential offset voltage drift, input offset current drift and differential gain drift are determined by sampling typical parts. Drift is not guaranteed by test or QA sampled at this value.

**Note 9:** Input common mode range is tested by verifying that at the limits stated in the Electrical Characteristics table, the differential offset ( $V_{OSDIFF}$ ) and common mode offset ( $V_{\text{OSCM}}$ ) have not deviated by more than  $\pm 200 \mu$ V and  $\pm 10$ mV respectively compared to the V<sub>ICM</sub> = 5V (at V<sub>S</sub> = 10V),  $V_{IGM}$  = 2.5V (at  $V_S$  = 5V) and  $V_{IGM}$  = 1.5V (at  $V_S$  = 3V) cases.

Output common mode range is tested by verifying that at the limits stated in the Electrical Characteristics table, the common mode offset ( $V_{\text{OSCM}}$ ) has not deviated by more than  $\pm 15$ mV compared to the V<sub>OCM</sub> = 5V (at V<sub>S</sub> = 10V),  $V_{OCM}$  = 2.5V (at  $V_S$  = 5V) and  $V_{OCM}$  = 1.5V (at  $V_S$  = 3V) cases.

**Note 10:** Input bias current is defined as the average of the input currents flowing into the input pins (–IN and +IN). Input Offset current is defined as the difference between the input bias currents  $(I_{OS} = I_B^+ - I_B^-)$ .

**Note 11:** Full power bandwidth is calculated from the slew rate.  $FPBW = SR/(2 \cdot \pi \cdot V_P)$ 

#### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to all parts in the LTC6363 family.**



6363fb

11

### TYPICAL PERFORMANCE CHARACTERISTICS

**Applicable to the LTC6363 only.**



**Differential Input Offset Voltage vs Input Common Mode Voltage**





**Typical Distribution of Input** 



**Input Bias Current vs Input Common Mode Voltage**



**Output Overdrive Recovery** VINDIFF VOUTDIF  $V_S = \pm 5V$  $V_{\text{INDIFF}} = 24V_{\text{P-P}}$  $R_{LOAD} = 2k$  DIFF 1μs/DIV 4V/DIV 6363 G09

**Input Offset Current vs Temperature**



**Input Noise Density vs Frequency**



#### **Slew Rate vs Temperature**



### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363 only.**













**Open-Loop Gain and Phase vs Frequency**



 $V_S = \pm 5V$ V<sub>ICM</sub> = V<sub>OCM</sub> = 0V<br>R<sub>I</sub> = R<sub>F</sub> = 1k  $R_{LOAD} = 2k$  DIFF V<sub>INDIFF</sub> = 18V<sub>P-P</sub><br>SINGLE-ENDED INPUT V–OUT  $V_{+0UT}$ 500ns/DIV 1V/DIV 6363 G22

### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363 only.**



### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363-0.5 only.**





#### **Typical Distribution of Differential Gain Error**  $V_S = \pm 5V$ 72 UNITS

30



**Typical Distribution of Differential Gain Drift**



**Input Common Mode Rejection** 

FREQUENCY(Hz) 1k 10k 100k 1M 10M 100M

V<sub>S</sub> = ±5\<br>||| | ||||

6363 G37

**Ratio vs Frequency**

 $\frac{1}{16}$ 

60

70

80

COMMON MODE REJECTION RATIO (dB)

COMMON MODE REJECTION RATIO (dB)

90

100

110

 $V_S = \pm 5V$  $V<sub>ICM</sub> = V<sub>OCM</sub> = 0V$  $R_{\text{LOAD}} = 2k$  DIFF V<sub>OUTDIFF</sub> = 18V<sub>P-P</sub><br>SINGLE-ENDED INPUT SLEW MEASURED 10% to 90% TEMPERATURE (°C) –50 –25 0 25 50 75 100 125  $\frac{30}{-50}$ 35 40 45 50 55 60 SLEW RATE (V/µS)

**Slew Rate vs Temperature**

#### **Differential Input Offset Voltage vs Input Common Mode Voltage**





6363 G35

#### **Large Signal Step Response**



### 15 6363fb

### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363-0.5 only.**



**Harmonic Distortion vs Output** 







**Settling Time vs Output Step** 





### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363-1 only.**





### **Typical Distribution of Differential Gain Error** V<sub>S</sub> = ±5V<br>72 UNITS

25



**Typical Distribution of Differential Gain Drift**









**Slew Rate vs Temperature**



#### **Differential Input Offset Voltage vs Input Common Mode Voltage**





**Large Signal Step Response**



### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363-1 only.**



**Harmonic Distortion vs Output** 







**Settling Time vs Output Step** 





### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363-2 only.**





#### **Typical Distribution of Differential Gain Error**  $V_S = \pm 5V$ 72 UNITS

25





**Input Common Mode Rejection** 

FREQUENCY(Hz)

 $V_S = \pm 5V$ 

6363 G67

**Ratio vs Frequency**

 $60$ <sup>L</sup><br>1k

70

80

COMMON MODE REJECTION RATIO (dB)

COMMON MODE REJECTION RATIO (dB)

90

100

110

**Slew Rate vs Temperature**  $V_S = \pm 5V$ VICM = VOCM = 0V RLOAD = 2k DIFF V<sub>OUTDIFF</sub> = 18V<sub>P-P</sub><br>SINGLE-ENDED INPUT SLEW MEASURED 10% to 90% TEMPERATURE (°C) –50 –25 0 25 50 75 100 125  $30 - 50$ 35 40 45 50 55 60 SLEW RATE (V/µS)

**Differential Input Offset Voltage vs Input Common Mode Voltage**





6363 G65

#### **Large Signal Step Response**



### TYPICAL PERFORMANCE CHARACTERISTICS **Applicable to the LTC6363-2 only.**



**Harmonic Distortion vs Output** 







**Settling Time vs Output Step** 





### PIN FUNCTIONS

**–IN (Pin 1):** Inverting Input of Amplifier. In the fixed-gain LTC6363-0.5/LTC6363-1/LTC6363-2 versions, this pin connects to a precision, on-chip resistor  $R_1$ .

**V<sub>OCM</sub>** (Pin 2): Output Common Mode Reference Voltage. Voltage applied to this pin sets the output common mode voltage level. If left floating, an internal resistor divider creates a default voltage approximately halfway between V<sup>+</sup> and V<sup>–</sup>. The V<sub>OCM</sub> pin should be decoupled to ground with a minimum of 0.1µF.

**V + (Pin 3):** Positive Power Supply. Operational supply range is 2.8V to 11V when  $V^- = 0V$ .

**+OUT (Pin 4):** Positive Output Pin. Output capable of swinging rail-to-rail.

**–OUT (Pin 5):** Negative Output Pin. Output capable of swinging rail-to-rail.

**V – (Pin 6/Exposed Pad Pin 9):** Negative Power Supply. Negative supply can be 0V, or taken negative as long as  $2.8V \le (V^+ - V^-) \le 11V$ .

**SHDN** (Pin 7): When the SHDN pin is floating or driven high, the LTC6363 family is in the normal (active) operating mode. When SHDN pin is connected to V– or driven low, the part is disabled and draws approximately 20µA of supply current  $(V_S = 3V)$ .

**+IN (Pin 8):** Noninverting Input of Amplifier. In the fixed LTC6363-0.5/LTC6363-1/LTC6363-2 versions, this pin connects to a precision,on-chip resistor  $\mathsf{R}_{\mathsf{I}}$ .

### BLOCK DIAGRAMS



**LTC6363-0.5/ LTC6363-1/ LTC6363-2**





### APPLICATIONS INFORMATION

#### **Functional Description**

The LTC6363 family consists of four fully differential, low power, low noise, precision amplifiers. The LTC6363 is an unconstrained, fully differential amplifier, typically used with four external resistors. The LTC6363-0.5, LTC6363-1, and LTC6363-2 (gains of 0.5, 1, and 2 respectively) are fully-differential fixed gain blocks featuring precision, laser trimmed, matched internal resistors for accurate, stable gain and excellent CMRR. The entire LT6363 family is optimized to convert a fully differential or single-ended signal to a low impedance, balanced differential output suitable for driving high performance, low power differential sigma-delta or SAR ADCs. The balanced differential nature of the amplifier also provides even-order harmonic distortion cancellation, and low susceptibility to common mode noise (e.g. power supply noise).

The outputs of the LTC6363 family are capable of swinging rail-to-rail and can source up to 90mA or sink up to 40mA of current. The LTC6363 family is optimized for high bandwidth and low power applications. Load capacitances above 50pF to ground or 25pF differentially should be decoupled with 10 $\Omega$  to 50 $\Omega$  of series resistance from each output to prevent oscillation or ringing.

### **SHDN Pin**

The LTC6363 family has a  $\overline{\text{SHDN}}$  pin which, when tied to V $^-$  or driven to below  $V_{II}$ , will shut down amplifier operation such that only 20 $\mu$ A (at V<sub>S</sub> = 3V) to 70 $\mu$ A (at V<sub>S</sub> = 10V) is drawn from the supplies. Pull-down circuitry should be capable of sinking at least 12µA to guarantee complete shutdown over all conditions. For normal amplifier operation, the SHDN pin should be left floating or tied to  $\mathsf{V}^*$  or driven to above  $\mathsf{V}_{\mathsf{IH}}$ .

#### **General Amplifier Applications**

In Figure 1, the gain to  $V_{\text{OUTDIFF}}$  from  $V_{\text{IND}}$  and  $V_{\text{INM}}$  is given by:

$$
V_{\text{OUTDIFF}} = V_{+OUT} - V_{-OUT} \approx \left(\frac{R_F}{R_I}\right) \bullet (V_{INP} - V_{INM})
$$

Note from the previous equation, the differential output voltage  $(V_{+OUT} - V_{-OUT})$  is independent of input and output common mode voltages, or the voltage at the common mode pin. This makes the LTC6363 family ideally suited



**Figure 1. Definitions and Terminology**

for pre-amplification, pre-attenuation, level shifting and conversion of single-ended signals to differential output signals for driving differential input ADCs or other devices.

#### **Output Common Mode and V<sub>OCM</sub> Pin**

The output common mode voltage is defined as the average of the two outputs:

$$
V_{OUTCM} = \left(\frac{V_{+OUT} + V_{-OUT}}{2}\right) = V_{OCM}
$$

As the equation shows, the output common mode voltage is independent of the input common mode voltage, and is instead determined by the voltage on the  $V_{\text{OCM}}$  pin, by means of an internal common mode feedback loop.

The  $V_{\text{OCM}}$  input connects to the base of a PNP transistor and an internal resistor divider network. If the  $V_{\text{OCM}}$  pin is left open, the resistor divider creates a default voltage approximately halfway between V<sup>+</sup> and V<sup>–</sup>. The V<sub>OCM</sub> pin can be overdriven to another voltage if desired for greater accuracy or flexibility. For example, when driving an ADC, if the ADC makes a reference available for setting the common mode voltage, it can be directly tied to the  $V_{\Omega CM}$ pin, as long as the ADC is capable of driving the 1.8M input resistance presented by the  $V_{OCM}$  pin. The Electrical Characteristics table specifies the valid range that can be applied to the  $V_{\text{OCM}}$  pin ( $V_{\text{OUTCMB}}$ ).

#### **Input Common Mode Voltage Range**

6363fb For all versions of the LTC6363, the input common mode voltage range,  $V_{ICMR}$ , specification refers to the voltage at the input pins of the part. The input common mode voltage range of the LTC6363-0.5, LTC6363-1 and LTC6363-2 are extended beyond that of the LTC6363 due to the resistor divider action of the on-chip resistors.

### APPLICATIONS INFORMATION

For LTC6363-0.5, LTC6363-1 and LTC6363-2 applications where the input is fully differential, the common mode voltage at the amplifier summing junction can be calculated using the following equation:

$$
V_{ICM\_AMP} = V_{ICM} \cdot \left(\frac{G}{G+1}\right) + V_{OCM} \cdot \left(\frac{1}{G+1}\right)
$$

Where G is the gain,  $V_{ICM}$  AMP is the common mode voltage at the amplifier's summing junction,  $V_{OCM}$  is the voltage applied to the  $V_{\text{OCM}}$  pin and  $V_{\text{ICM}}$  is the common mode voltage applied to the input pins of the LTC6363-0.5, LTC6363-1 or LTC6363-2. This equation is more useful when solved for  $V_{ICM}$ :

**Table 1. Valid Input Common Mode Voltage Range for Fixed-Gain Versions (Differential Inputs)**

<b>PART VERSION</b>	<b>GAIN</b>	SUPPLY (V)	$V_{OCM} (V)$	$V_{IGM} (V)$
LTC6363-0.5	0.5	3	0.5	$-1$ to 4.4
LTC6363-0.5	0.5	3	1.5	$-3$ to 2.4
LTC6363-0.5	0.5	3	2.5	$-5$ to 0.4
LTC6363-0.5	0.5	5	0.5	$-1$ to 10.4
LTC6363-0.5	0.5	5	2.5	$-5$ to 6.4
LTC6363-0.5	0.5	5	4.5	$-9$ to 2.4
LTC6363-0.5	0.5	10	0.5	$-1$ to 25.4
LTC6363-0.5	0.5	10	5	$-10$ to 16.4
LTC6363-0.5	0.5	10	9.5	$-19$ to $7.4$
LTC6363-1	1	3	0.5	$-0.5$ to 3.1
LTC6363-1	$\mathbf{1}$	3	1.5	$-1.5$ to 2.1
LTC6363-1	1	3	2.5	$-2.5$ to 1.1
LTC6363-1	1	5	0.5	$-0.5$ to $7.1$
LTC6363-1	1	5	2.5	$-2.5$ to 5.1
LTC6363-1	1	5	4.5	$-4.5$ to 3.1
LTC6363-1	1	10	0.5	$-0.5$ to 17.1
LTC6363-1	1	10	5	$-5$ to 12.6
LTC6363-1	$\mathbf{1}$	10	9.5	$-9.5$ to 8.1
LTC6363-2	$\overline{c}$	3	0.5	$-0.25$ to 2.45
LTC6363-2	$\overline{2}$	3	1.5	$-0.75$ to 1.95
LTC6363-2	$\overline{2}$	3	2.5	$-1.25$ to 1.45
LTC6363-2	2	5	0.5	$-0.25$ to 5.45
LTC6363-2	2	5	2.5	$-1.25$ to 4.45
LTC6363-2	2	5	4.5	$-2.25$ to 3.45
LTC6363-2	$\overline{c}$	10	0.5	$-0.25$ to 12.95
LTC6363-2	$\overline{2}$	10	5	$-2.5$ to 10.7
LTC6363-2	$\overline{2}$	10	9.5	$-4.75$ to 8.45

$$
V_{ICM} = \frac{V_{ICM\_AMP} \bullet (G+1) - V_{OCM}}{G}
$$

The minimum and maximum valid input common mode voltage can be computed using this equation by substituting for  $V_{ICM}$  AMP the minimum and maximum  $V_{ICMR}$ specification of the LTC6363: V<sup>-</sup> and V<sup>+</sup>-1.2V respectively. Table 1 lists various solutions to this equation.

The equation changes slightly if the LTC6363-0.5, LTC6363-1 or LTC6363-2 input is single ended since now the input common mode voltage at the amplifiers's summing junction is also a function of the input signal  $V_{INP}$  (where  $V_{INM} = 0$ ):

$$
V_{ICM}=\frac{V_{ICM\_AMP} \bullet (G+1)-V_{OCM}}{G}-\frac{V_{INP}}{2}
$$

In summary, the common mode voltage at the input pins of the LTC6363-0.5/LTC6363-1/LTC6363-2 ( $V_{ICM}$ ) is valid if it lies within the following range:

$$
\frac{V^-(G+1) - V_{OCM}}{G} \leq V_{ICM} \leq \frac{(V^+ - 1.2)(G+1) - V_{OCM}}{G}
$$

For Differential Inputs

$$
\frac{V^-(G+1) - V_{OCM}}{G} - \frac{V_{INP}}{2} \le V_{ICM}
$$

$$
\le \frac{(V^+ - 1.2)(G+1) - V_{OCM}}{G} - \frac{V_{INP}}{2}
$$

For Single-Ended Inputs ( $V_{INM} = 0$ )

#### **Input Pin Protection**

The absolute maximum input current of the LTC6363 amplifier input pins is  $\pm 10$ mA, as specified in the Absolute Maximum Ratings. The amplifier inside the LTC6363-0.5/ LTC6363-1/LTC6363-2 also has this same limitation but cannot be directly observed. Absolute maximum input voltage is specified for the LTC6363-0.5/LTC6363-1/ LTC6363-2 using the following equations:

$$
V^- - 10mA \cdot R_1 - \frac{(V_{OUT} - V^- + 0.3)}{G} - 0.3 \text{ to}
$$
  

$$
V^+ + 10mA \cdot R_1 + \frac{(V^+ + 0.3 - V_{OUT})}{G} + 0.3
$$

### APPLICATIONS INFORMATION

The output voltage is a variable in these equations because it affects how much current is flowing in  $R_F$ . This current also flows in  $\mathsf{R}_{\mathsf{I}}$  and increases the voltage which can be applied to the input without exceeding the 10mA limit on the amplifier's inputs. The absolute maximum input voltage is specified conservatively, assuming the output voltage is at  $V^+$  for the positive limit and  $V^-$  for the negative limit. This simplifies the equations:

$$
V^- - 10mA \cdot R_1 - 0.3 \cdot \left(1 + \frac{1}{G}\right) \text{ to}
$$
  

$$
V^+ + 10mA \cdot R_1 + 0.3 \cdot \left(1 + \frac{1}{G}\right)
$$

#### **Input Impedance and Loading Effects**

The low frequency input impedance looking into the  $V_{\text{IMP}}$ or  $V_{INM}$  input of Figure 1 depends on how the inputs are driven. For fully differential input sources ( $V_{\text{INP}} = -V_{\text{INM}}$ ), the input impedance seen at either input is simply:

$$
R_{\mathsf{INP}} = R_{\mathsf{INM}} = R_{\mathsf{I}}
$$

For single-ended inputs, due to the signal imbalance at the input, the input impedance increases over the balanced differential case. The input impedance looking into either input is:

$$
R_{INP} = R_{INM} = \frac{R_I}{1 - \left(\frac{1}{2}\right) \cdot \left(\frac{R_F}{R_I + R_F}\right)}
$$

Input signal sources with non-zero impedances can also cause feedback imbalance between the pair of feedback networks. For the best performance, it is recommended that the input source impedance be compensated. If impedance matching is required at the source, a termination resistor R1 should be chosen (see Figure 2) such that:

$$
R1 = \frac{R_{INM} \cdot R_S}{R_{INM} - R_S}
$$



**Figure 2. Optimal Compensation for Signal Source Impedance**

According to Figure 2, the input impedance looking into the differential amp  $(R_{INM})$  reflects the single-ended source case, given above. Also, R2 is chosen as:

$$
R2 = R1 || R_S = \frac{R1 \cdot R_S}{R1 + R_S}
$$

#### **Effects of Resistor Pair Mismatch**

Figure 3 shows a circuit diagram which takes into consideration resistor mismatch. Often, resistor mismatch limits CMRR well below amplifier specifications. Assuming infinite open-loop gain, the differential output relationship is given by the equation:

$$
V_{OUT(DIFF)} = V_{+OUT} - V_{-OUT}
$$
  
\n
$$
\approx V_{INDIFF} \cdot \frac{R_F}{R_I} + V_{CM} \cdot \frac{\Delta \beta}{\beta_{AVG}} - V_{OCM} \cdot \frac{\Delta \beta}{\beta_{AVG}}
$$

where  $R_F$  is the average of  $R_{F1}$  and  $R_{F2}$ , and  $R_I$  is the average of  $R_{11}$  and  $R_{12}$ .



**Figure 3. Real-World Application with Feedback Resistor Pair Mismatch**

25

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