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## Multicell Battery Monitors

## **FEATURES**

- Measures Up to 12 Battery Cells in Series
- Stackable Architecture Supports 100s of Cells
- Built-In isoSPI™ Interface: 1Mbps Isolated Serial Communications Uses a Single Twisted Pair, Up to 100 Meters Low EMI Susceptibility and Emissions
- 1.2mV Maximum Total Measurement Error
- 290µs to Measure All Cells in a System
- Synchronized Voltage and Current Measurement
- 16-Bit Delta-Sigma ADC with Frequency Programmable 3rd Order Noise Filter
- Engineered for ISO26262 Compliant Systems
- Passive Cell Balancing with Programmable Timer
- 5 General Purpose Digital I/O or Analog Inputs: Temperature or other Sensor Inputs Configurable as an I<sup>2</sup>C or SPI Master
- 4µA Sleep Mode Supply Current
- 48-Lead SSOP Package

## **APPLICATIONS**

- Electric and Hybrid Electric Vehicles
- Backup Battery Systems
- Grid Energy Storage
- High Power Portable Equipment

## DESCRIPTION

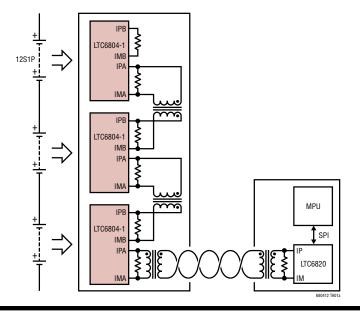
The LTC®6804 is a 3rd generation multicell battery stack monitor that measures up to 12 series connected battery cells with a total measurement error of less than 1.2mV. The cell measurement range of 0V to 5V makes the LTC6804 suitable for most battery chemistries. All 12 cell voltages can be captured in 290µs, and lower data acquisition rates can be selected for high noise reduction.

Multiple LTC6804 devices can be connected in series, permitting simultaneous cell monitoring of long, high voltage battery strings. Each LTC6804 has an isoSPI interface for high speed, RF-immune, local area communications. Using the LTC6804-1, multiple devices are connected in a daisy-chain with one host processor connection for all devices. Using the LTC6804-2, multiple devices are connected in parallel to the host processor, with each device individually addressed.

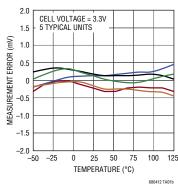
Additional features include passive balancing for each cell, an onboard 5V regulator, and 5 general purpose I/O lines. In sleep mode, current consumption is reduced to  $4\mu A$ . The LTC6804 can be powered directly from the battery, or from an isolated supply.

T, LT, LTC, LTM, Linear Technology and the Linear logo are registered and isoSPI is a trademark of Linear Technology Corporation. All other trademarks are the property of their respective owners. Protected by U.S. patents, including 8908799, 9182428, 9270133.

## TYPICAL APPLICATION



## Total Measurement Error vs Temperature of 5 Typical Units





## LTC6804-1/LTC6804-2

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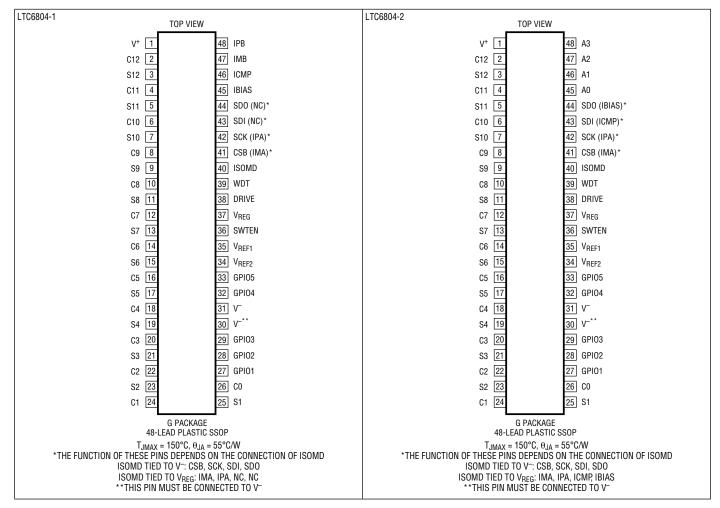
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## **ABSOLUTE MAXIMUM RATINGS** (Note 1)

Total Supply Voltage V <sup>+</sup> to V <sup>-</sup> . Input Voltage (Relative to V <sup>-</sup> )	75V
C0	0.3V to 0.3V
C12	0.3V to 75V
C(n)	0.3V to MIN (8 • n, 75V)
S(n)	0.3V to MIN (8 • n, 75V)
IPA, IMA, IPB, IMB	0.3V to $\dot{V}_{REG} + 0.3\dot{V}$
DRIVE Pin	
All Other Pins	0.3V to 6V
Voltage Between Inputs	
V <sup>+</sup> to C12	5.5V
C(n) to C(n – 1)	0.3V to 8V
S(n) to C(n – 1)	
C12 to C8	

C8 to C40.3V to 25V C4 to C00.3V to 25V
Current In/Out of Pins
All Pins Except V <sub>REG</sub> , IPA, IMA, IPB, IMB, S(n)10mA
IPA, IMA, IPB, IMB30mA
Operating Temperature Range
LTC6804I40°C to 85°C
LTC6804H
Specified Temperature Range
LTC6804I40°C to 85°C
LTC6804H
Junction Temperature
Storage Temperature65°C to 150°C
Lead Temperature (Soldering, 10sec)300°C

## PIN CONFIGURATION





## ORDER INFORMATION http://www.linear.com/product/LTC6804-1#orderinfo

TUBE	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE
LTC6804IG-1#PBF	LTC6804IG-1#TRPBF	LTC6804G-1	48-Lead Plastic SSOP	-40°C to 85°C
LTC6804HG-1#PBF	LTC6804HG-1#TRPBF	LTC6804G-1	48-Lead Plastic SSOP	-40°C to 125°C
LTC6804IG-2#PBF	LTC6804IG-2#TRPBF	LTC6804G-2	48-Lead Plastic SSOP	-40°C to 85°C
LTC6804HG-2#PBF	LTC6804HG-2#TRPBF	LTC6804G-2	48-Lead Plastic SSOP	-40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. \*The temperature grade is identified by a label on the shipping container. Parts ending with PBF are RoHS and WEEE compliant.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/. Some packages are available in 500 unit reels through designated sales channels with #TRMPBF suffix.

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . The test conditions are $V^+ = 39.6V$ , $V_{REG} = 5.0V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN TYP	MAX	UNITS
ADC DC Spec	ifications					
	Measurement Resolution		•	0.1		mV/bit
	ADC Offset Voltage	(Note 2)	•	0.1		mV
	ADC Gain Error	(Note 2)	•	0.01 0.02		% %
	Total Measurement Error (TME) in	$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 0$		±0.2		mV
	Normal Mode	C(n) to $C(n-1) = 2.0$		±0.1	±0.8	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 2.0$	•		±1.4	mV
		C(n) to $C(n-1) = 3.3$		±0.2	±1.2	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 3.3$	•		±2.2	mV
		C(n) to $C(n-1) = 4.2$		±0.3	±1.6	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 4.2$	•		±2.8	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 5.0$		±1		mV
		Sum of Cells, V(CO) = V <sup>-</sup>	•	±0.2	±0.75	%
		Internal Temperature, T = Maximum Specified Temperature		±5		°C
		V <sub>REG</sub> Pin	•	±0.1	±0.25	%
		V <sub>REF2</sub> Pin	•	±0.02	±0.1	%
		Digital Supply Voltage V <sub>REGD</sub>	•	±0.1	±1	%

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25\,^{\circ}$ C. The test conditions are $V^+ = 39.6V$ , $V_{REG} = 5.0V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
	Total Measurement Error (TME) in	$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 0$			±0.1		mV
	Filtered Mode	C(n) to $C(n-1) = 2.0$			±0.1	±0.8	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 2.0$	•			±1.4	mV
		C(n) to $C(n-1) = 3.3$			±0.2	±1.2	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 3.3$	•			±2.2	mV
		C(n) to $C(n-1) = 4.2$			±0.3	±1.6	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 4.2$	•			±2.8	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 5.0$			±1		mV
		Sum of Cells, V(CO) = V <sup>-</sup>	•		±0.2	±0.75	%
		Internal Temperature, T = Maximum Specified Temperature			±5		°C
		V <sub>REG</sub> Pin	•		±0.1	±0.25	%
		V <sub>REF2</sub> Pin	•	:	±0.02	±0.1	%
		Digital Supply Voltage V <sub>REGD</sub>	•		±0.1	±1	%
	Total Measurement Error (TME) in	$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^{-}=0$			±2		mV
	Fast Mode	$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 2.0$	•			±4	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 3.3$	•			±4.7	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 4.2$	•			±8.3	mV
		$C(n)$ to $C(n-1)$ , $GPIO(n)$ to $V^- = 5.0$			±10		mV
		Sum of Cells, V(CO) = V <sup>-</sup>	•		±0.3	±1	%
		Internal Temperature, T = Maximum Specified Temperature			±5		°C
		V <sub>REG</sub> Pin	•		±0.3	±1	%
		V <sub>REF2</sub> Pin	•		±0.1	±0.25	%
		Digital Supply Voltage V <sub>REGD</sub>	•		±0.2	±2	%
	Input Range	C(n), n = 1 to 12	•	C(n - 1)		C(n-1) + 5	V
		CO	•		0		
		GPIO(n), n = 1 to 5	•	0		5	V
IL	Input Leakage Current When Inputs Are Not Being Measured	C(n), n = 0 to 12	•		10	±250	nA
	The first zoing inducator	GPIO(n), n = 1 to 5	•		10	±250	nA
	Input Current When Inputs Are	C(n), n = 0 to 12			±2		μА
	Being Measured	GPIO(n), n = 1 to 5			±2		μА
	Input Current During Open Wire Detection		•	70	100	130	μА



# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25\,^{\circ}$ C. The test conditions are $V^+ = 39.6V$ , $V_{REG} = 5.0V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS			MIN	TYP	MAX	UNITS
Voltage Refe	rence Specifications	1						
V <sub>REF1</sub>	1st Reference Voltage	V <sub>REF1</sub> Pin, No Load		•	3.1	3.2	3.3	V
	1st Reference Voltage TC	V <sub>REF1</sub> Pin, No Load				3		ppm/°C
	1st Reference Voltage Hysteresis	V <sub>REF1</sub> Pin, No Load				20		ppm
	1st Reference Long Term Drift	V <sub>REF1</sub> Pin, No Load				20		ppm/√kHr
$V_{REF2}$	2nd Reference Voltage	V <sub>REF2</sub> Pin, No Load		•	2.990	3	3.010	V
		V <sub>REF2</sub> Pin, 5k Load to V <sup>-</sup>		•	2.988	3	3.012	V
	2nd Reference Voltage TC	V <sub>REF2</sub> Pin, No Load				10		ppm/°C
	2nd Reference Voltage Hysteresis	V <sub>REF2</sub> Pin, No Load	,			100		ppm
	2nd Reference Long Term Drift	V <sub>REF2</sub> Pin, No Load				60		ppm/√kHr
General DC S	pecifications							
$I_{VP}$	V <sup>+</sup> Supply Current	State: Core = SLEEP, isoSPI = IDLE	$V_{REG} = 0V$			3.8	6	μA
	(See Figure 1: LTC6804 Operation State Diagram)		$V_{REG} = 0V$	•		3.8	10	μА
	Ctato Liag.am,		$V_{REG} = 5V$			1.6	3	μА
			$V_{REG} = 5V$	•		1.6	5	μA
		State: Core = STANDBY			18	32	50	μА
				•	10	32	60	μA
		State: Core = REFUP or MEASURE			0.4	0.55	0.7	mA
				•	0.375	0.55	0.725	mA
I <sub>REG(CORE)</sub>	V <sub>REG</sub> Supply Current	$V_{REG}$ Supply Current State: Core = SLEEP, isoSPI = IDLE $V_{REG}$ =	$V_{REG} = 5V$			2.2	4	μА
	(See Figure 1: LTC6804 Operation	V <sub>RE</sub>		•		2.2	6	μА
	State diagram)	State: Core = STANDBY  State: Core = REFUP  State: Core = MEASURE			10	35	60	μА
				•	6	35	65	μА
					0.2	0.45	0.7	mA
				•	0.15	0.45	0.75	mA
					10.8	11.5	12.2	mA
				•	10.7	11.5	12.3	mA
I <sub>REG(isoSPI)</sub>	Additional V <sub>REG</sub> Supply Current if	LTC6804-2: ISOMD = 1,	READY	•	3.9	4.8	5.8	mA
	isoSPI in READY/ACTIVE States	$R_{B1} + R_{B2} = 2k$	ACTIVE	•	5.1	6.1	7.3	mA
	Note: ACTIVE State Current Assumes t <sub>CLK</sub> = 1µs, (Note 3)	LTC6804-1: ISOMD = 0,	READY	•	3.7	4.6	5.6	mA
	OLK PS/(SSS)	$R_{B1} + R_{B2} = 2k$	ACTIVE	•	5.7	6.8	8.1	mA
		LTC6804-1: ISOMD = 1,	READY	•	6.5	7.8	9.5	mA
		$R_{B1} + R_{B2} = 2k$	ACTIVE	•	10.2	11.3	13.3	mA
		LTC6804-2: ISOMD = 1,	READY	•	1.3	2.1	3	mA
		$R_{B1} + R_{B2} = 20k$	ACTIVE	•	1.6	2.5	3.5	mA
		LTC6804-1: ISOMD = 0,	READY	•	1.1	1.9	2.8	mA
		$R_{B1} + R_{B2} = 20k$	ACTIVE	•	1.5	2.3	3.3	mA
		LTC6804-1: ISOMD = 1,	READY	•	2.1	3.3	4.9	mA
		$R_{B1} + R_{B2} = 20k$	ACTIVE	•	2.7	4.1	5.8	mA

LINEAD TECHNOLOGY

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25\,^{\circ}$ C. The test conditions are $V^+ = 39.6V$ , $V_{REG} = 5.0V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
	V <sup>+</sup> Supply Voltage	TME Specifications Met (Note 6)	•	11	40	55	٧
$V_{REG}$	V <sub>REG</sub> Supply Voltage	TME Supply Rejection < 1mV/V	•	4.5	5	5.5	V
	DRIVE output voltage	Sourcing 1µA		5.4	5.6	5.8	V
			•	5.2	5.6	6.0	V
		Sourcing 500µA	•	5.1	5.6	6.1	V
V <sub>REGD</sub>	Digital Supply Voltage		•	2.7	3.0	3.6	V
	Discharge Switch ON Resistance	V <sub>CELL</sub> = 3.6V	•		10	25	Ω
	Thermal Shutdown Temperature				150		°C
V <sub>OL(WDT)</sub>	Watchdog Timer Pin Low	WDT Pin Sinking 4mA	•			0.4	V
V <sub>OL(GPIO)</sub>	General Purpose I/O Pin Low	GPIO Pin Sinking 4mA (Used as Digital Output)	•			0.4	V
ADC Timing Spe	ecifications						
t <sub>C</sub> YCLE	Measurement + Calibration Cycle	Measure 12 Cells	•	2120	2335	2480	μs
(Figure 3)	Time When Starting from the REFUP State in Normal Mode	Measure 2 Cells	•	365	405	430	μs
	TELOT Otato III Normal Mode	Measure 12 Cells and 2 GPIO Inputs	•	2845	3133	3325	μs
	Measurement + Calibration Cycle	Measure 12 Cells	•	183	201.3	213.5	ms
	Time When Starting from the REFUP State in Filtered Mode	Measure 2 Cells	•	30.54	33.6	35.64	ms
	NEFUR State III FIITETEU WOUE	Measure 12 Cells and 2 GPIO Inputs	•	244	268.4	284.7	ms
	Measurement + Calibration Cycle Time When Starting from the REFUP State in Fast Mode	Measure 12 Cells	•	1010	1113	1185	μs
		Measure 2 Cells	•	180	201	215	μs
		Measure 12 Cells and 2 GPIO Inputs	•	1420	1564	1660	μs
t <sub>SKEW1</sub> (Figure 6)	Skew Time. The Time Difference between C12 and GPIO2 Measurements, Command = ADCVAX	Fast Mode	•	189	208	221	μs
		Normal Mode	•	493	543	576	μs
t <sub>SKEW2</sub> (Figure 3)	Skew Time. The Time Difference between C12 and C0	Fast Mode	•	211	233	248	μs
	Measurements, Command = ADCV	Normal Mode	•	609	670	711	μs
t <sub>WAKE</sub>	Regulator Start-Up Time	V <sub>REG</sub> Generated from Drive Pin (Figure 28)			100	300	μs
t <sub>SLEEP</sub>	Watchdog or Software Discharge	SWTEN Pin = 0 or DCT0[3:0] = 0000	•	1.8	2	2.2	sec
JULI	Timer	SWTEN Pin = 1 and DCTO[3:0] ≠ 0000		0.5		120	mir
t <sub>REFUP</sub>	Reference Wake-Up Time	State: Core = STANDBY		2.7	3.5	4.4	ms
(Figure 1, Figures 3 to 7)	The second of th	State: Core = REFUP	•			0	ms
$f_S$	ADC Clock Frequency		•	3.0	3.3	3.5	MHz
SPI Interface DO	C Specifications						
V <sub>IH(SPI)</sub>	SPI Pin Digital Input Voltage High	Pins CSB, SCK, SDI	•	2.3			V
V <sub>IL(SPI)</sub>	SPI Pin Digital Input Voltage Low	Pins CSB, SCK, SDI	•			0.8	V
V <sub>IH(CFG)</sub>	Configuration Pin Digital Input Voltage High	Pins ISOMD, SWTEN, GPI01 to GPI05, A0 to A3	•	2.7			V
V <sub>IL(CFG)</sub>	Configuration Pin Digital Input Voltage Low	Pins ISOMD, SWTEN, GPI01 to GPI05, A0 to A3	•			1.2	V



# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . The test conditions are $V^+ = 39.6V$ , $V_{REG} = 5.0V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I <sub>LEAK(DIG)</sub>	Digital Input Current	Pins CSB, SCK, SDI, ISOMD, SWTEN, A0 to A3	•			±1	μА
$V_{OL(SDO)}$	Digital Output Low	Pin SDO Sinking 1mA	•			0.3	V
isoSPI DC Spe	ecifications (See Figure 16)					,	
V <sub>BIAS</sub>	Voltage on IBIAS Pin	READY/ACTIVE State IDLE State	•	1.9	2.0	2.1	V
I <sub>B</sub>	Isolated Interface Bias Current	R <sub>BIAS</sub> = 2k to 20k	•	0.1		1.0	mA
A <sub>IB</sub>	Isolated Interface Current Gain	$\label{eq:lambda} \begin{array}{c} V_A \leq 1.6V & I_B = 1 mA \\ I_B = 0.1 mA \end{array}$	•	18 18	20 20	22 24.5	mA/mA mA/mA
$\overline{V_A}$	Transmitter Pulse Amplitude	$V_A =  V_{IP} - V_{IM} $	•			1.6	V
V <sub>ICMP</sub>	Threshold-Setting Voltage on ICMP Pin	V <sub>TCMP</sub> = A <sub>TCMP</sub> • V <sub>ICMP</sub>	•	0.2		1.5	V
I <sub>LEAK(ICMP)</sub>	Input Leakage Current on ICMP Pin	V <sub>ICMP</sub> = 0V to V <sub>REG</sub>	•			±1	μΑ
I <sub>LEAK(IP/IM)</sub>	Leakage Current on IP and IM Pins	IDLE State, V <sub>IP</sub> or V <sub>IM</sub> = 0V to V <sub>REG</sub>	•		,	±1	μΑ
A <sub>TCMP</sub>	Receiver Comparator Threshold Voltage Gain	$V_{CM} = V_{REG}/2$ to $V_{REG} - 0.2V$ , $V_{ICMP} = 0.2V$ to 1.5V	•	0.4	0.5	0.6	V/V
V <sub>CM</sub>	Receiver Common Mode Bias	IP/IM Not Driving		(V <sub>REG</sub>	- V <sub>ICMP</sub> /3	– 167mV)	V
R <sub>IN</sub>	Receiver Input Resistance	Single-Ended to IPA, IMA, IPB, IMB	•	27	35	43	kΩ
isoSPI Idle/Wa	akeup Specifications (See Figure 21)					,	
V <sub>WAKE</sub>	Differential Wake-Up Voltage	t <sub>DWELL</sub> = 240ns	•	200			m۷
t <sub>DWELL</sub>	Dwell Time at V <sub>WAKE</sub> Before Wake Detection	V <sub>WAKE</sub> = 200mV	•	240			ns
t <sub>READY</sub>	Startup Time After Wake Detection		•			10	μs
t <sub>IDLE</sub>	Idle Timeout Duration		•	4.3	5.5	6.7	ms
isoSPI Pulse 1	Fiming Specifications (See Figure 19)						
t <sub>1/2PW(CS)</sub>	Chip-Select Half-Pulse Width		•	120	150	180	ns
t <sub>INV(CS)</sub>	Chip-Select Pulse Inversion Delay		•			200	ns
t <sub>1/2PW(D)</sub>	Data Half-Pulse Width		•	40	50	60	ns
t <sub>INV(D)</sub>	Data Pulse Inversion Delay		•			70	ns
	equirements (See Figure 15 and Figure	20)					
t <sub>CLK</sub>	SCK Period	(Note 4)	•	1			μs
t <sub>1</sub>	SDI Setup Time before SCK Rising Edge		•	25			ns
t <sub>2</sub>	SDI Hold Time after SCK Rising Edge		•	25			ns
$\overline{t_3}$	SCK Low	$t_{CLK} = t_3 + t_4 \ge 1 \mu s$	•	200			ns
t <sub>4</sub>	SCK High	$t_{CLK} = t_3 + t_4 \ge 1 \mu s$	•	200	,		ns
t <sub>5</sub>	CSB Rising Edge to CSB Falling Edge		•	0.65			μs
t <sub>6</sub>	SCK Rising Edge to CSB Rising Edge	(Note 4)	•	0.8			μs
t <sub>7</sub>	CSB Falling Edge to SCK Rising Edge	(Note 4)	•	1			μs

LINEAD TECHNOLOGY

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . The test conditions are $V^+ = 39.6V$ , $V_{REG} = 5.0V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS		
isoSPI Timin	isoSPI Timing Specifications (See Figure 19)								
t <sub>8</sub>	SCK Falling Edge to SDO Valid	(Note 5)	•			60	ns		
t <sub>9</sub>	SCK Rising Edge to Short ±1 Transmit		•			50	ns		
t <sub>10</sub>	CSB Transition to Long ±1 Transmit		•			60	ns		
t <sub>11</sub>	CSB Rising Edge to SDO Rising	(Note 5)	•			200	ns		
t <sub>RTN</sub>	Data Return Delay		•		430	525	ns		
t <sub>DSY(CS)</sub>	Chip-Select Daisy-Chain Delay		•		150	200	ns		
t <sub>DSY(D)</sub>	Data Daisy-Chain Delay		•		300	360	ns		
t <sub>LAG</sub>	Data Daisy-Chain Lag (vs Chip- Select)		•	0	35	70	ns		
t <sub>6(GOV)</sub>	Data to Chip-Select Pulse Governor		•	0.8		1.05	μs		

**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

**Note 2:** The ADC specifications are guaranteed by the Total Measurement Error specification.

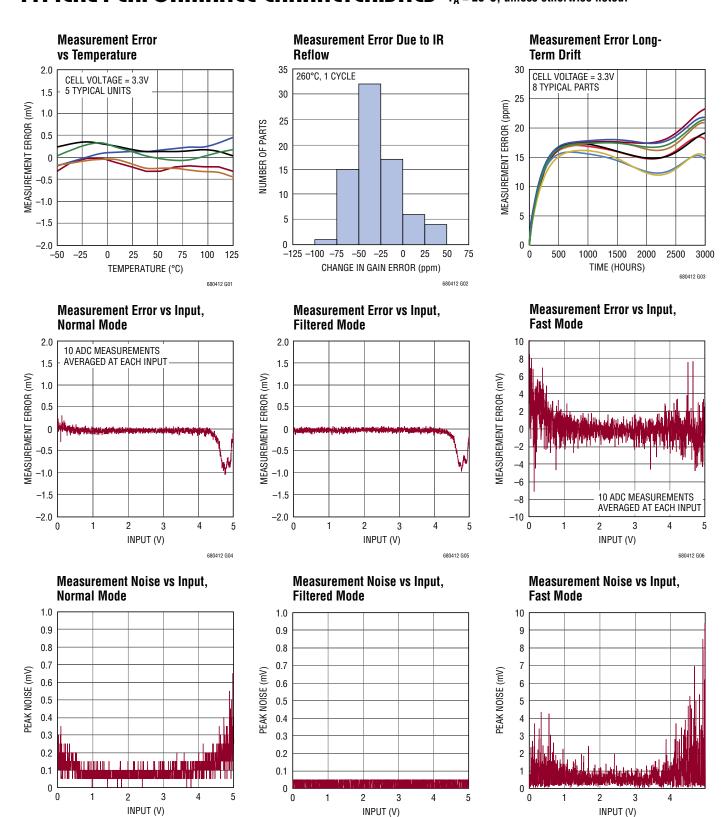
**Note 3:** The ACTIVE state current is calculated from DC measurements. The ACTIVE state current is the additional average supply current into  $V_{REG}$  when there is continuous 1MHz communications on the isoSPI ports with 50% data 1's and 50% data 0's. Slower clock rates reduce the supply current. See Applications Information section for additional details.

**Note 4:** These timing specifications are dependent on the delay through the cable, and include allowances for 50ns of delay each direction. 50ns corresponds to 10m of CAT-5 cable (which has a velocity of propagation of 66% the speed of light). Use of longer cables would require derating these specs by the amount of additional delay.

**Note 5:** These specifications do not include rise or fall time of SDO. While fall time (typically 5ns due to the internal pull-down transistor) is not a concern, rising-edge transition time  $t_{RISE}$  is dependent on the pull-up resistance and load capacitance on the SDO pin. The time constant must be chosen such that SDO meets the setup time requirements of the MCU.

**Note 6:**  $V^+$  needs to be greater than or equal to the highest C(n) voltage for accurate measurements. See the graph Top Cell Measurement Error vs  $V^+$ .

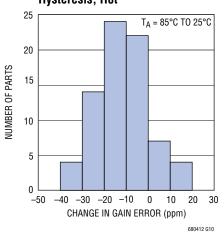




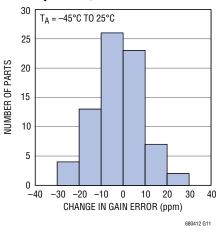
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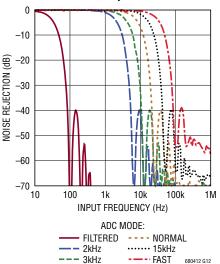




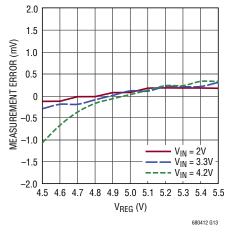
#### **Measurement Gain Error** Hysteresis, Cold



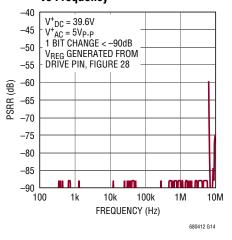




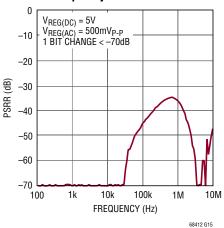
## Measurement Error vs V<sub>REG</sub>



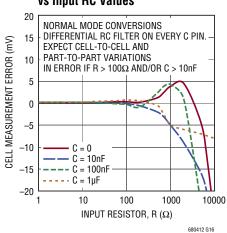
### Measurement Error V+ PSRR vs Frequency



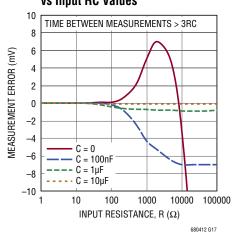
### Measurement Error V<sub>REG</sub> PSRR vs Frequency



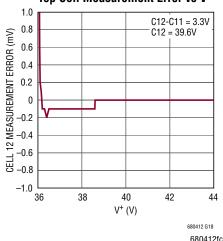
### **Cell Measurement Error** vs Input RC Values

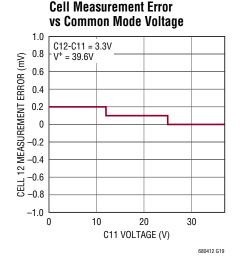


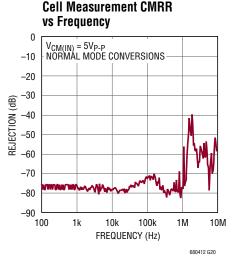
## **GPIO Measurement Error** vs Input RC Values

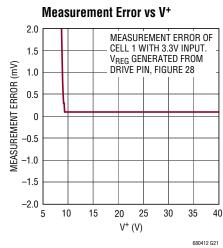


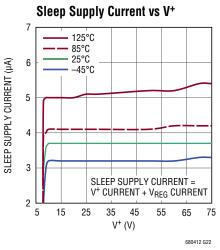
## Top Cell Measurement Error vs V+

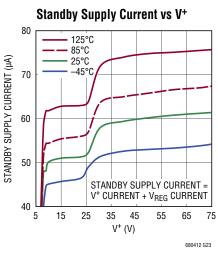


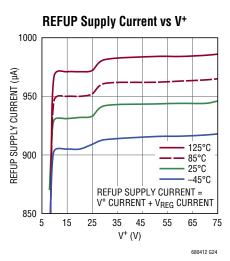


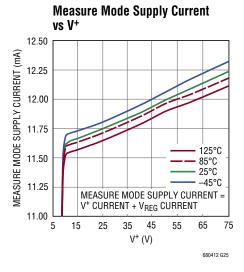


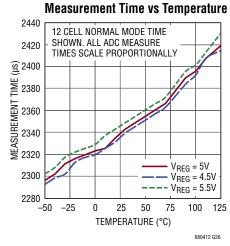


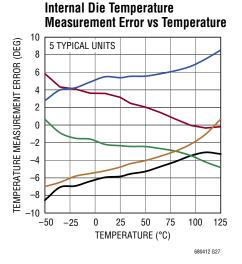




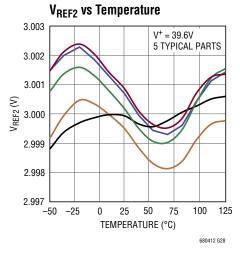


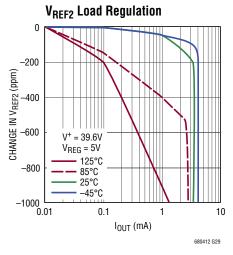


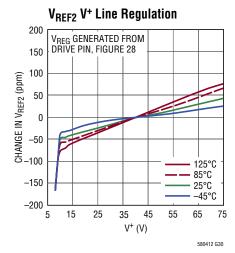


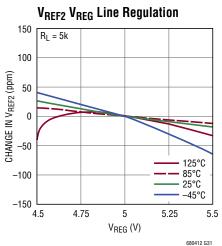


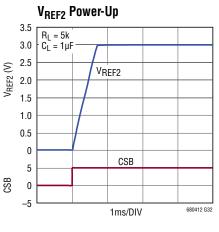


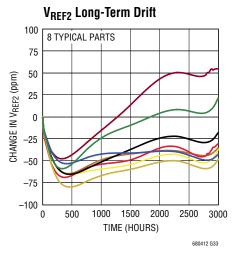


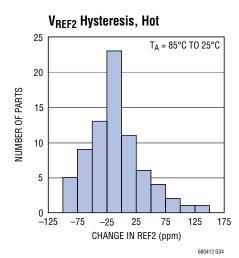


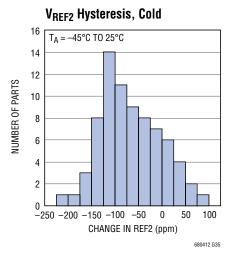


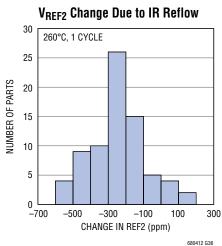




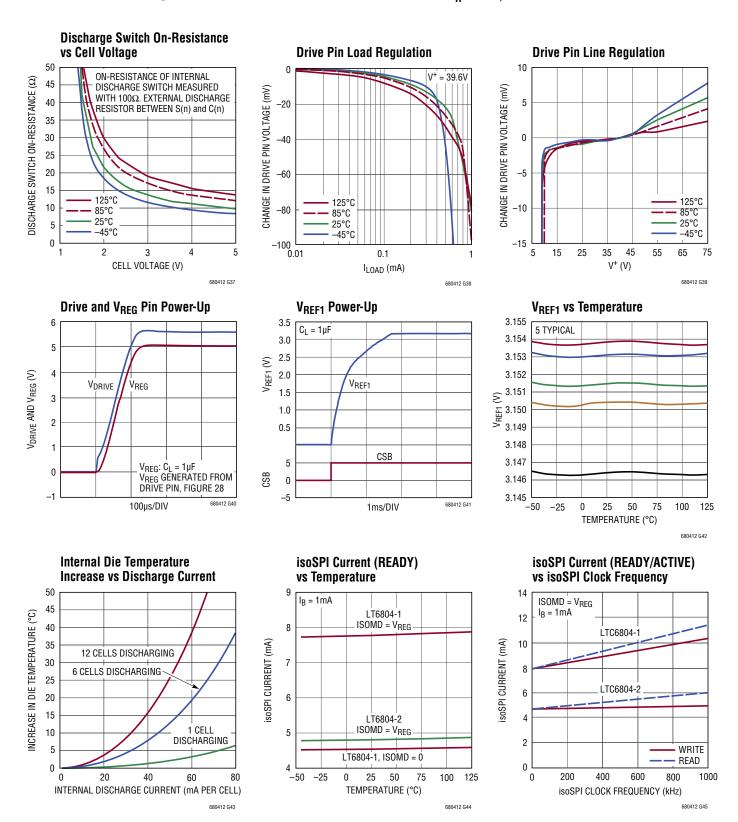


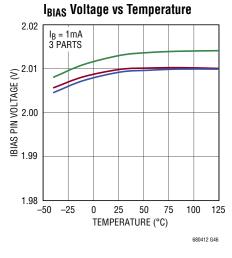


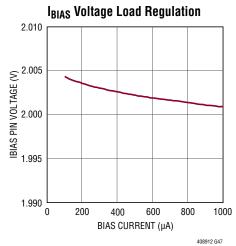


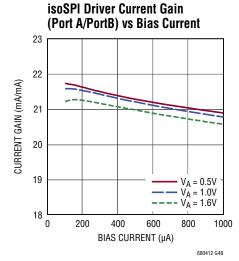


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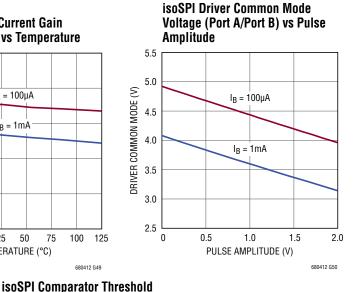


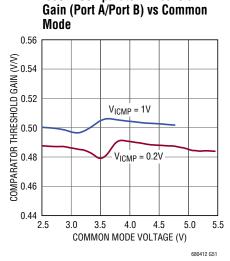


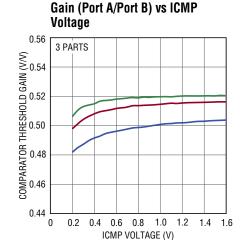


isoSPI Comparator Threshold

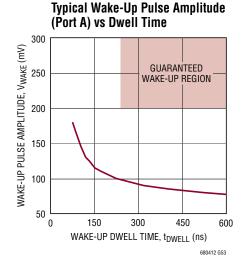
isoSPI Driver Current Gain (Port A/PortB) vs Temperature 23 22  $I_B = 100 \mu A$ CURRENT GAIN (mA/mA)  $I_B = 1 \text{mA}$ 20 19 18 -25 50 -50 0 25 75 100 125 TEMPERATURE (°C)







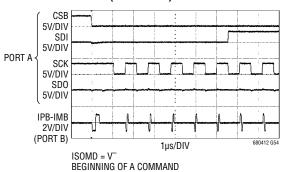
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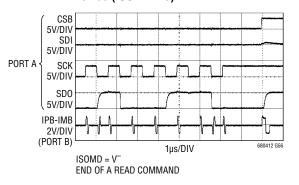
680412fc

680412 G52

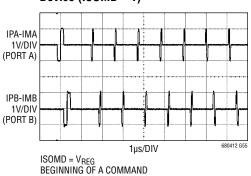
### Write Command to a Daisy-Chained Device (ISOMD = 0)



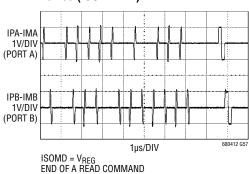
## Data Read-Back from a Daisy-Chained Device (ISOMD = 0)



### Write Command to a Daisy-Chained Device (ISOMD = 1)



## Data Read-Back from a Daisy-Chained Device (ISOMD = 1)





## PIN FUNCTIONS

CO to C12: Cell Inputs.

**S1 to S12**: Balance Inputs/Outputs. 12 N-MOSFETs are connected between S(n) and C(n-1) for discharging cells.

V+: Positive Supply Pin.

**V**<sup>-</sup>: Negative Supply Pins. The V<sup>-</sup> pins must be shorted together, external to the IC.

 $V_{REF2}$ : Buffered 2nd reference voltage for driving multiple 10k thermistors. Bypass with an external 1 $\mu$ F capacitor.

**V**<sub>REF1</sub>: ADC Reference Voltage. Bypass with an external 1µF capacitor. No DC loads allowed.

**GPIO[1:5]:** General Purpose I/O. Can be used as digital inputs or digital outputs, or as analog inputs with a measurement range from  $V^-$  to 5V. GPIO [3:5] can be used as an  $I^2$ C or SPI port.

**SWTEN:** Software Timer Enable. Connect this pin to V<sub>REG</sub> to enable the software timer.

**DRIVE:** Connect the base of an NPN to this pin. Connect the collector to  $V^+$  and the emitter to  $V_{REG}$ .

 $V_{REG}$ : 5V Regulator Input. Bypass with an external 1µF capacitor.

**ISOMD:** Serial Interface Mode. Connecting ISOMD to  $V_{REG}$  configures Pins 41 to 44 of the LTC6804 for 2-wire isolated interface (isoSPI) mode. Connecting ISOMD to  $V^-$  configures the LTC6804 for 4-wire SPI mode.

**WDT:** Watchdog Timer Output Pin. This is an open drain NMOS digital output. It can be left unconnected or connected with a 1M resistor to  $V_{REG}$ . If the LTC6804 does not receive a wake-up signal (see Figure 21) within 2 seconds, the watchdog timer circuit will reset the LTC6804 and the WDT pin will go high impedance.

#### **Serial Port Pins**

	LTC68 (DAISY-CH		LTC68 (ADDRES	
	ISOMD = V <sub>REG</sub>	ISOMD = V	ISOMD = V <sub>REG</sub>	ISOMD = V
PORT B	IPB	IPB	A3	A3
(Pins 45 to 48)	IMB	IMB	A2	A2
10 40)	ICMP	ICMP	A1	A1
	IBIAS	IBIAS	A0	A0
PORT A	(NC)	SD0	IBIAS	SD0
(Pins 41 to 44)	(NC)	SDI	ICMP	SDI
	IPA	SCK	IPA	SCK
	IMA	CSB	IMA	CSB

**CSB**, **SCK**, **SDI**, **SDO**: 4-Wire Serial Peripheral Interface (SPI). Active low chip select (CSB), serial clock (SCK), and serial data in (SDI) are digital inputs. Serial data out (SDO) is an open drain NMOS output pin. SDO requires a 5k pull-up resistor.

**A0 to A3:** Address Pins. These digital inputs are connected to  $V_{REG}$  or  $V^-$  to set the chip address for addressable serial commands.

**IPA**, **IMA**: Isolated 2-Wire Serial Interface Port A. IPA (plus) and IMA (minus) are a differential input/output pair.

**IPB**, **IMB**: Isolated 2-Wire Serial Interface Port B. IPB (plus) and IMB (minus) are a differential input/output pair.

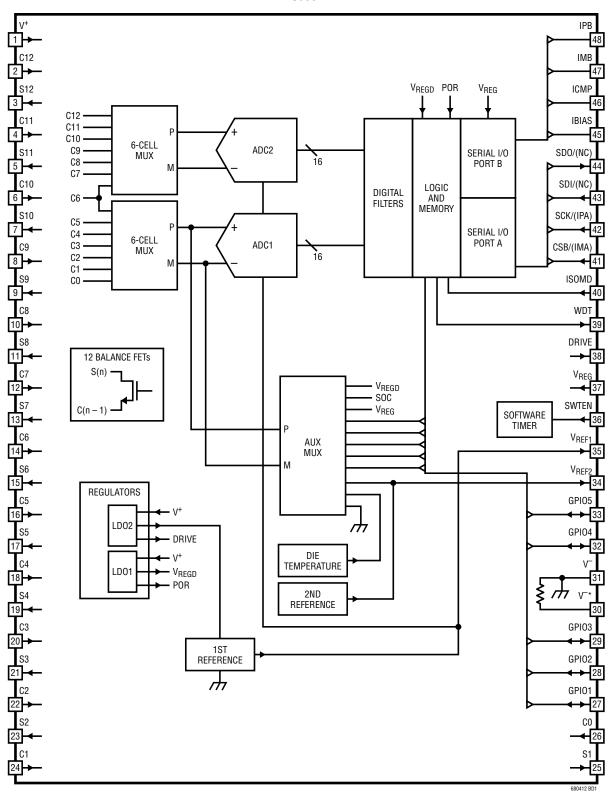
**IBIAS:** Isolated Interface Current Bias. Tie IBIAS to  $V^-$  through a resistor divider to set the interface output current level. When the isoSPI interface is enabled, the IBIAS pin voltage is 2V. The IPA/IMA or IPB/IMB output current drive is set to 20 times the current,  $I_B$ , sourced from the IBIAS pin.

**ICMP:** Isolated Interface Comparator Voltage Threshold Set. Tie this pin to the resistor divider between IBIAS and V<sup>-</sup> to set the voltage threshold of the isoSPI receiver comparators. The comparator thresholds are set to 1/2 the voltage on the ICMP pin.



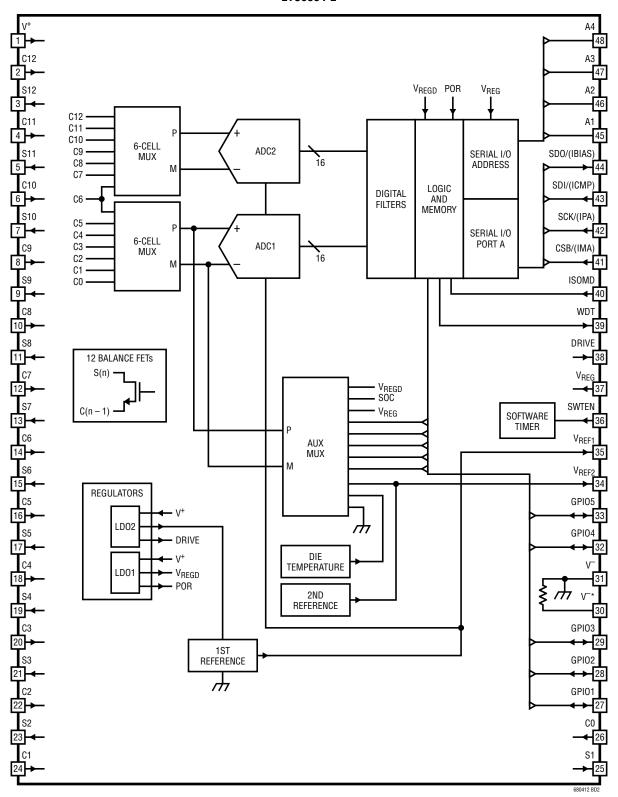
## **BLOCK DIAGRAM**

LTC6804-1



## **BLOCK DIAGRAM**

## LTC6804-2





### STATE DIAGRAM

The operation of the LTC6804 is divided into two separate sections: the core circuit and the isoSPI circuit. Both sections have an independent set of operating states, as well as a shutdown timeout.

## LTC6804 CORE STATE DESCRIPTIONS

#### **SLEEP State**

The reference and ADCs are powered down. The watchdog timer (see Watchdog and Software Discharge Timer) has timed out. The software discharge timer is either disabled or timed out. The supply currents are reduced to minimum levels. The isoSPI ports will be in the IDLE state.

If a WAKEUP signal is received (see Waking Up the Serial Interface), the LTC6804 will enter the STANDBY state.

### STANDBY State

The reference and the ADCs are off. The watchdog timer and/or the software discharge timer is running. The DRIVE pin powers the  $V_{REG}$  pin to 5V through an external transistor. (Alternatively,  $V_{REG}$  can be powered by an external supply).

When a valid ADC command is received or the REFON bit is set to 1 in the Configuration Register Group, the IC pauses for t<sub>REFUP</sub> to allow for the reference to power up and then enters either the REFUP or MEASURE state. If there is no WAKEUP signal for a duration t<sub>SLEEP</sub> (when both the watchdog and software discharge timer have expired) the LTC6804

returns to the SLEEP state. If the software discharge timer is disabled, only the watchdog timer is relevant.

#### **REFUP State**

To reach this state the REFON bit in the Configuration Register Group must be set to 1 (using the WRCFG command, see Table 36). The ADCs are off. The reference is powered up so that the LTC6804 can initiate ADC conversions more quickly than from the STANDBY state.

When a valid ADC command is received, the IC goes to the MEASURE state to begin the conversion. Otherwise, the LTC6804 will return to the STANDBY state when the REFON bit is set to 0, either manually (using WRCFG command) or automatically when the watchdog timer expires. (The LTC6804 will then move straight into the SLEEP state if both timers are expired).

#### **MEASURE State**

The LTC6804 performs ADC conversions in this state. The reference and ADCs are powered up.

After ADC conversions are complete the LTC6804 will transition to either the REFUP or STANDBY states, depending on the REFON bit. Additional ADC conversions can be initiated more quickly by setting REFON = 1 to take advantage of the REFUP state.

Note: Non-ADC commands do not cause a Core state transition. Only an ADC conversion or diagnostic commands will place the Core in the MEASURE state.

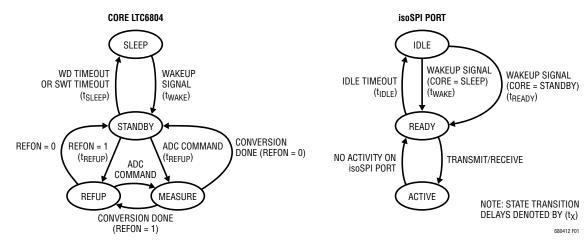


Figure 1. LTC6804 Operation State Diagram



#### isoSPI STATE DESCRIPTIONS

Note: The LTC6804-1 has two isoSPI ports (A and B), for daisy-chain communication. The LTC6804-2 has only one isoSPI port (A), for parallel-addressable communication.

#### **IDLE State**

The isoSPI ports are powered down.

When isoSPI port A receives a WAKEUP signal (see Waking Up the Serial Interface), the isoSPI enters the READY state. This transition happens quickly (within  $t_{READY}$ ) if the Core is in the STANDBY state because the DRIVE and  $v_{REG}$  pins are already biased up. If the Core is in the SLEEP state when the isoSPI receives a WAKEUP signal, then it transitions to the READY state within  $t_{WAKE}$ .

### **READY State**

The isoSPI port(s) are ready for communication. Port B is enabled only for LTC6804-1, and is not present on the LTC6804-2. The serial interface current in this state depends on if the part is LTC6804-1 or LTC6804-2, the status of the ISOMD pin, and  $R_{BIAS} = R_{B1} + R_{B2}$  (the external resistors tied to the IBIAS pin).

If there is no activity (i.e., no WAKEUP signal) on port A for greater than  $t_{IDLE}$  = 5.5ms, the LTC6804 goes to the IDLE state. When the serial interface is transmitting or receiving data the LTC6804 goes to the ACTIVE state.

### **ACTIVE State**

The LTC6804 is transmitting/receiving data using one or both of the isoSPI ports. The serial interface consumes maximum power in this state. The supply current increases with clock frequency as the density of isoSPI pulses increases.

## **POWER CONSUMPTION**

The LTC6804 is powered via two pins: V<sup>+</sup> and V<sub>REG</sub>. The V<sup>+</sup> input requires voltage greater than or equal to the top cell voltage, and it provides power to the high voltage elements of the core circuitry. The V<sub>REG</sub> input requires 5V and provides power to the remaining core circuitry and the isoSPI circuitry. The V<sub>REG</sub> input can be powered through an external transistor, driven by the regulated

DRIVE output pin. Alternatively, V<sub>REG</sub> can be powered by an external supply.

The power consumption varies according to the operational states. Table 1 and Table 2 provide equations to approximate the supply pin currents in each state. The V<sup>+</sup> pin current depends only on the Core state and not on the isoSPI state. However, the  $V_{REG}$  pin current depends on both the Core state and isoSPI state, and can therefore be divided into two components. The isoSPI interface draws current only from the  $V_{REG}$  pin.

 $I_{REG} = I_{REG(CORE)} + I_{REG(isoSPI)}$ 

Table 1. Core Supply Current

STATE		l <sub>V</sub> +	I <sub>REG(CORE)</sub>	
SLEEP	V <sub>REG</sub> = 0V	3.8µA	0μΑ	
SLEEP	V <sub>REG</sub> = 5V	1.6µA	2.2μΑ	
STANDBY		32μΑ	35μΑ	
REFUP		550µA	450μA	
MEASURE		550µA	11.5mA	

In the SLEEP state the  $V_{REG}$  pin will draw approximately 2.2 $\mu$ A if powered by a external supply. Otherwise, the V<sup>+</sup> pin will supply the necessary current.

## **ADC OPERATION**

There are two ADCs inside the LTC6804. The two ADCs operate simultaneously when measuring twelve cells. Only one ADC is used to measure the general purpose inputs. The following discussion uses the term ADC to refer to one or both ADCs, depending on the operation being performed. The following discussion will refer to ADC1 and ADC2 when it is necessary to distinguish between the two circuits, in timing diagrams, for example.

### **ADC Modes**

The ADCOPT bit (CFGR0[0]) in the configuration register group and the mode selection bits MD[1:0] in the conversion command together provide 6 modes of operation for the ADC which correspond to different over sampling ratios (OSR). The accuracy of these modes are summarized in Table 3. In each mode, the ADC first measures the inputs, and then performs a calibration of each channel. The names of the modes are based on the –3dB bandwidth of the ADC measurement.



680412f

Table 2. isoSPI Supply Current Equations

isoSPI STATE	DEVICE	ISOMD CONNECTION	I <sub>REG(isoSPI)</sub>
IDLE	LTC6804-1/LTC6804-2	N/A	0mA
READY	LTC6804-1	$V_{REG}$	$2.8\text{mA} + 5 \cdot I_B \text{ Note: } I_B = V_{BIAS}/(R_{B1} + R_{B2})$
		V-	1.6mA + 3 • I <sub>B</sub>
	LTC6804-2	$V_{REG}$	1.8mA + 3 • I <sub>B</sub>
		V-	0mA
ACTIVE	LTC6804-1	$V_{REG}$	Write: $2.8\text{mA} + 5 \cdot l_B + (2 \cdot l_B + 0.4\text{mA}) \cdot \frac{1\mu s}{t_{CLK}}$ Read: $2.8\text{mA} + 5 \cdot l_B + (3 \cdot l_B + 0.5\text{mA}) \cdot \frac{1\mu s}{t_{CLK}}$
		V-	$1.6\text{mA} + 3 \bullet l_{B} + (2 \bullet l_{B} + 0.2\text{mA}) \bullet \frac{1\mu s}{t_{CLK}}$
	LTC6804-2	$V_{REG}$	Write: $1.8\text{mA} + 3 \cdot l_B + (0.3\text{mA}) \cdot \frac{1\mu s}{t_{CLK}}$ Read: $1.8\text{mA} + 3 \cdot l_B + (l_B + 0.3\text{mA}) \cdot \frac{1\mu s}{t_{CLK}}$
		V-	0mA

Table 3. ADC Filter Bandwidth and Accuracy

MODE	-3dB FILTER BW	-40dB FILTER BW	TME SPEC AT 3.3V, 25°C	TME SPEC AT 3.3V,-40°C, 125°C
27kHz (Fast Mode)	27kHz	84kHz	±4.7mV	±4.7mV
14kHz	13.5kHz	42kHz	±4.7mV	±4.7mV
7kHz (Normal Mode)	6.8kHz	21kHz	±1.2mV	±2.2mV
3kHz	3.4kHz	10.5kHz	±1.2mV	±2.2mV
2kHz	1.7kHz	5.3kHz	±1.2mV	±2.2mV
26Hz (Filtered Mode)	26Hz	82Hz	±1.2mV	±2.2mV

Note: TME is the total measurement error.

## Mode 7kHz (Normal):

In this mode, the ADC has high resolution and low TME (total measurement error). This is considered the normal operating mode because of the optimum combination of speed and accuracy.

## Mode 27kHz (Fast):

In this mode, the ADC has maximum throughput but has some increase in TME (total measurement error). So this mode is also referred to as the fast mode. The increase in speed comes from a reduction in the oversampling ratio. This results in an increase in noise and average measurement error.

## Mode 26Hz (Filtered):

In this mode, the ADC digital filter –3dB frequency is lowered to 26Hz by increasing the OSR. This mode is also referred to as the filtered mode due to its low –3dB frequency. The accuracy is similar to the 7kHz (Normal) mode with lower noise.

## Modes 14kHz, 3kHz and 2kHz:

Modes 14kHz, 3kHz and 2kHz provide additional options to set the ADC digital filter—3dB frequency at 13.5kHz, 3.4kHz and 1.7kHz respectively. The accuracy of the 14kHz mode is similar to the 27kHz (fast) mode. The accuracy of 3kHz and 2kHz modes is similar to the 7kHz (normal) mode.



The conversion times for these modes are provided in Table 5. If the core is in STANDBY state, an additional  $t_{REFUP}$  time is required to power up the reference before beginning the ADC conversions. The reference can remain powered up between ADC conversions if the REFON bit in Configuration Register Group is set to 1 so the core is in REFUP state after a delay  $t_{REFUP}$ . Then, the subsequent ADC commands will not have the  $t_{REFUP}$  delay before beginning ADC conversions.

## **ADC Range and Resolution**

The C inputs and GPIO inputs have the same range and resolution. The ADC inside the LTC6804 has an approximate range from -0.82V to 5.73V. Negative readings are rounded to 0V. The format of the data is a 16-bit unsigned integer where the LSB represents  $100\mu V$ . Therefore, a reading of 0x80E8 (33,000 decimal) indicates a measurement of 3.3V.

Delta-Sigma ADCs have quantization noise which depends on the input voltage, especially at low over sampling ratios (OSR), such as in FAST mode. In some of the ADC modes, the quantization noise increases as the input voltage approaches the upper and lower limits of the ADC range. For example, the total measurement noise versus input voltage in normal and filtered modes is shown in Figure 2.

The specified range of the ADC is 0V to 5V. In Table 4, the precision range of the ADC is arbitrarily defined as 0.5V to 4.5V. This is the range where the quantization noise is relatively constant even in the lower OSR modes (see Figure 2). Table 4 summarizes the total noise in this range for all six ADC operating modes. Also shown is the noise

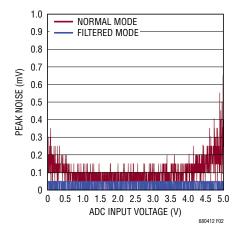


Figure 2. Measurement Noise vs Input Voltage

free resolution. For example, 14-bit noise free resolution in normal mode implies that the top 14 bits will be noise free with a DC input, but that the 15th and 16th least significant bits (LSB) will flicker.

## **ADC Range vs Voltage Reference Value:**

Typical Delta-Sigma ADC's have a range which is exactly twice the value of the voltage reference, and the ADC measurement error is directly proportional to the error in the voltage reference. The LTC6804 ADC is not typical. The absolute value of  $V_{REF1}$  is trimmed up or down to compensate for gain errors in the ADC. Therefore, the ADC total measurement error (TME) specifications are superior to the  $V_{REF1}$  specifications. For example, the 25°C specification of the total measurement error when measuring 3.300V in 7kHz (normal) mode is  $\pm 1.2$ mV and the 25°C specification for  $V_{REF1}$  is 3.200V  $\pm 100$ mV.

Table 4. ADC Range and Resolution

MODE	FULL RANGE <sup>1</sup>	SPECIFIED Range	PRECISION Range <sup>2</sup>	LSB	FORMAT	MAX NOISE	NOISE FREE RESOLUTION <sup>3</sup>
27kHz (Fast)						$\pm 4mV_{P-P}$	10 Bits
14kHz		-0.8192V to 5.7344V 0V to 5V 0.5V to 4.5V				±1mV <sub>P-P</sub>	12 Bits
7kHz (Normal)			0.5)/+0.4.5)/	0.5V to 4.5V 100uV Unsigned 16 Bits	±250μV <sub>P-P</sub>	14 Bits	
3kHz			0.57 (0.4.57	Ιούμν	±150μV <sub>F</sub>	±150μV <sub>P-P</sub>	14 Bits
2kHz						±100μV <sub>P-P</sub>	15 Bits
26Hz (Filtered)						±50μV <sub>P-P</sub>	16 Bits

<sup>1.</sup> Negative readings are rounded to 0V.

<sup>3.</sup> NOISE FREE RESOLUTION is a measure of the noise level within the PRECISION RANGE.



<sup>2.</sup> PRECISION RANGE is the range over which the noise is less than MAX NOISE.

## **Measuring Cell Voltages (ADCV Command)**

The ADCV command initiates the measurement of the battery cell inputs, pins C0 through C12. This command has options to select the number of channels to measure and the ADC mode. See the section on Commands for the ADCV command format.

Figure 3 illustrates the timing of ADCV command which measures all twelve cells. After the receipt of the ADCV command to measure all 12 cells, ADC1 sequentially measures the bottom 6 cells. ADC2 sequentially measures the top 6 cells. After the cell measurements are complete, each channel is calibrated to remove any offset errors.

Table 5 shows the conversion times for the ADCV command measuring all 12 cells. The total conversion time is given by  $t_{6C}$  which indicates the end of the calibration step.

Figure 4 illustrates the timing of the ADCV command that measures only two cells.

Table 6 shows the conversion time for ADCV command measuring only 2 cells.  $t_{1C}$  indicates the total conversion time for this command.

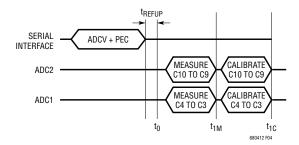


Figure 4. Timing for ADCV Command Measuring 2 Cells

Table 6. Conversion Times for ADCV Command Measuring Only 2 Cells in Different Modes

	CONVERSION TIMES (in µs)					
MODE	t <sub>0</sub>	t <sub>1M</sub>	t <sub>1C</sub>			
27kHz	0	57	201			
14kHz	0	86	230			
7kHz	0	144	405			
3kHz	0	260	521			
2kHz	0	493	754			
26Hz	0	29,817	33,568			

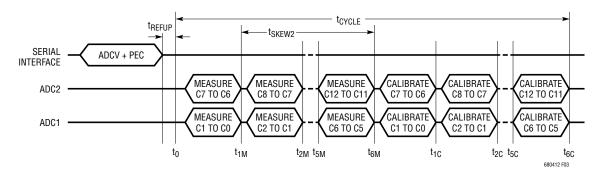


Figure 3. Timing for ADCV Command Measuring All 12 Cells

Table 5. Conversion Times for ADCV Command Measuring All 12 Cells in Different Modes

	CONVERSION TIMES (in µs)								
MODE	t <sub>0</sub>	t <sub>1M</sub>	t <sub>2M</sub>	t <sub>5M</sub>	t <sub>6M</sub>	t <sub>1C</sub>	t <sub>2C</sub>	t <sub>5C</sub>	t <sub>6C</sub>
27kHz	0	57	103	243	290	432	568	975	1,113
14kHz	0	86	162	389	465	606	742	1,149	1,288
7kHz	0	144	278	680	814	1,072	1,324	2,080	2,335
3kHz	0	260	511	1,262	1,512	1,770	2,022	2,778	3,033
2kHz	0	493	976	2,425	2,908	3,166	3,418	4,175	4,430
26Hz	0	29,817	59,623	149,043	178,850	182,599	186,342	197,571	201,317

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## **Under/Overvoltage Monitoring**

Whenever the C inputs are measured, the results are compared to undervoltage and overvoltage thresholds stored in memory. If the reading of a cell is above the overvoltage limit, a bit in memory is set as a flag. Similarly, measurement results below the undervoltage limit cause a flag to be set. The overvoltage and undervoltage thresholds are stored in the configuration register group. The flags are stored in the status register group B.

## Auxiliary (GPIO) Measurements (ADAX Command)

The ADAX command initiates the measurement of the GPIO inputs. This command has options to select which GPIO input to measure (GPIO1-5) and which ADC mode. The ADAX command also measures the 2nd reference. There are options in the ADAX command to measure each GPIO and the 2nd reference separately or to measure all 5 GPIOs and the 2nd reference in a single command. See the section on commands for the ADAX command format. All auxiliary measurements are relative to the V<sup>-</sup> pin voltage. This command can be used to read external temperature

by connecting the temperature sensors to the GPIOs. These sensors can be powered from the 2nd reference which is also measured by the ADAX command, resulting in precise ratiometric measurements.

Figure 5 illustrates the timing of the ADAX command measuring all GPIOs and the 2nd reference. Since all the 6 measurements are carried out on ADC1 alone, the conversion time for the ADAX command is similar to the ADCV command.

## Measuring Cell Voltages and GPIOs (ADCVAX Command)

The ADCVAX command combines twelve cell measurements with two GPIO measurements (GPIO1 and GPIO2). This command simplifies the synchronization of battery cell voltage and current measurements when current sensors are connected to GPIO1 or GPIO2 inputs. Figure 6 illustrates the timing of the ADCVAX command. See the section on commands for the ADCVAX command format. The synchronization of the current and voltage measurements, t<sub>SKEW1</sub>, in FAST MODE is within 208µs.

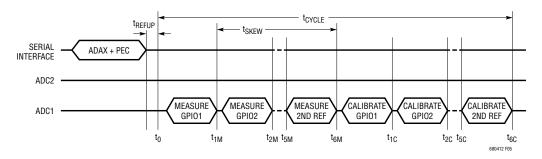


Figure 5. Timing for ADAX Command Measuring All GPIOs and 2nd Reference

Table 7. Conversion Times for ADAX Command Measuring All GPIOs and 2nd Reference in Different Modes

	CONVERSION TIMES (in μs)								
MODE	to	t <sub>1M</sub>	t <sub>2M</sub>	t <sub>5M</sub>	t <sub>6M</sub>	t <sub>1C</sub>	t <sub>2C</sub>	t <sub>5C</sub>	t <sub>6C</sub>
27kHz	0	57	103	243	290	432	568	975	1,113
14kHz	0	86	162	389	465	606	742	1,149	1,288
7kHz	0	144	278	680	814	1,072	1,324	2,080	2,335
3kHz	0	260	511	1,262	1,512	1,770	2,022	2,778	3,033
2kHz	0	493	976	2,425	2,908	3,166	3,418	4,175	4,430
26Hz	0	29,817	59,623	149,043	178,850	182,599	186,342	197,571	201,317

