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# LV49152V

**Bi-CMOS LSI** 

# Class-D Audio Power Amplifier BTL 15W × 2ch



http://onsemi.com

#### Overview

The LV49152V is a 15W per channel stereo digital power amplifier that takes analog inputs. The LV49152V uses unique Our developed feedback technology to achieve excellent audio quality despite being a class D amplifier and can be used to implement high quality flat display panel (FDP) based systems.

#### **Features**

- BTL output, class D amplifier system
- Unique Our developed feedback technology achieves superb audio quality
- High-efficiency class D amplifier
- Soft muting function reduces impulse noise at power on/off
- Full complement of built-in protection circuits : over current protection, thermal protection, and low power supply voltage protection circuits
- Built in Power limiter

#### **Functions**

• Power :  $15W \times 2ch$  output (VD = 15V, R<sub>L</sub> =  $8\Omega$ , fin = 1kHz, AES17, THD + N = 10%)

• Efficiency : 93% (VD = 15V,  $R_L = 8\Omega$ , fin = 1kHz,  $P_O = 15W$ )

• THD + N : 0.08% (VD = 15V, R<sub>L</sub> =  $8\Omega$ , fin = 1kHz, P<sub>O</sub> = 1W, Filter : AES17)

• Noise : 90µVrms (Filter : A-weight)

• Package SSOP44J (275mil)

#### **Specifications**

#### **Absolute Maximum Ratings** at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	VD	Supply voltage	20	V
Allowable power dissipation	Pd max	Our PCB, Soldered *	5.05	W
Package thermal resistance	θјс	Our PCB, Soldered *	2.1	°C/W
		Our PCB, Not soldered *	3.6	°C/W
Maximum junction temperature	Tj max		150	°C
Operating temperature	Topr		-25 to +75	°C
Storage temperature	Tstg		-50 to +150	°C

<sup>\*:</sup> Mounted on a specified board 110.0mm × 100.0mm × 1.5mm, glass epoxy (two-layer)

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

#### **Recommended Operating Range** at Ta = 25°C

Dans market	O. made ad	Symbol Conditions		Ratings		
Parameter	Symbol			typ	max	Unit
Supply voltage range	VD	Supply voltage	9	15	18	V
Load impedance range	RL	Speaker load	4	8		Ω

# Electrical Characteristics at Ta = 25 °C, VD = 15 V, RL = $8\Omega$ , L = $33\mu H$ (TOKO : A7502BY-330M), C = $0.1\mu F$ , CL = $0.47\mu F$

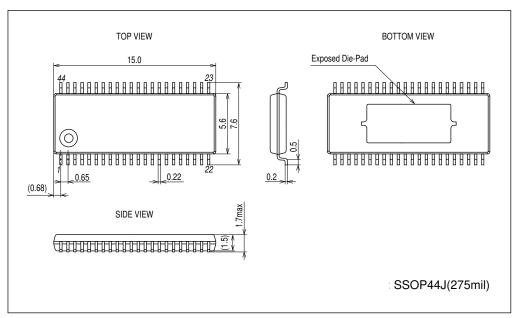
Davarantas	O. was la a l	Conditions	Ratings			Unit
Parameter	Symbol	Conditions	min	typ	max	Unit
Standby current	Ist	STBY = L, MUTE = L		1	10	μΑ
Mute current	Imute	STBY = H, MUTE = L	14	20	26	mA
Quiescent current	Icco	STBY = H, MUTE = H	35	45	55	mA
Voltage gain	VG	$fin = 1kHz, V_O = 0dBm$	28	30	32	dB
Offset voltage	Voffset	Rg = 0	-150		150	mV
Total harmonic distortion	THD+N	$P_O = 1W$ , fin = 1kHz, AES17		0.08	0.4	%
Output power	P <sub>O</sub> @10%	THD+N = 10%, AES17	13	15		W
Channel separation	CHsep.	Rg = 0, V <sub>O</sub> = 0dBm, DIN AUDIO	55	70		dB
Ripple rejection ratio	SVRR	fr = 100Hz, Vr = 0dBm, Rg = 0, DIN AUDIO	50	60		dB
Noise	V <sub>NO</sub>	Rg = 0, A-weight		90	300	μVrms
High-level input voltage	V <sub>IH</sub>	STBY and MUTE pin	3		VD	V
Low-level input voltage	V <sub>IL</sub>	STBY and MUTE pin	0		1	V
Under voltage protection UPPER	UV_UPPER	VD voltage measure		8.0		V
Under voltage protection LOWER	UV_LOWER	VD voltage measure		7.0		V

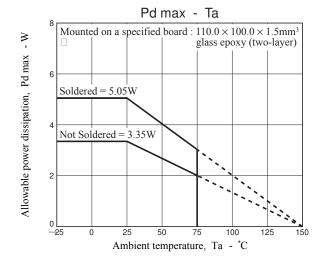
Note: The values of these characteristics were measured in the Our test environment. The actual values in an end system will vary depending on the printed circuit board pattern, the external components actually used, and other factors.

## **Package Dimensions**

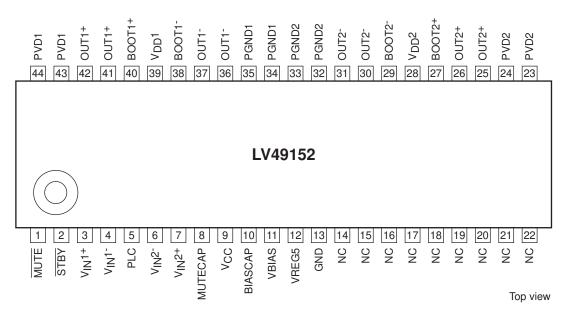
unit: mm (typ)

3285

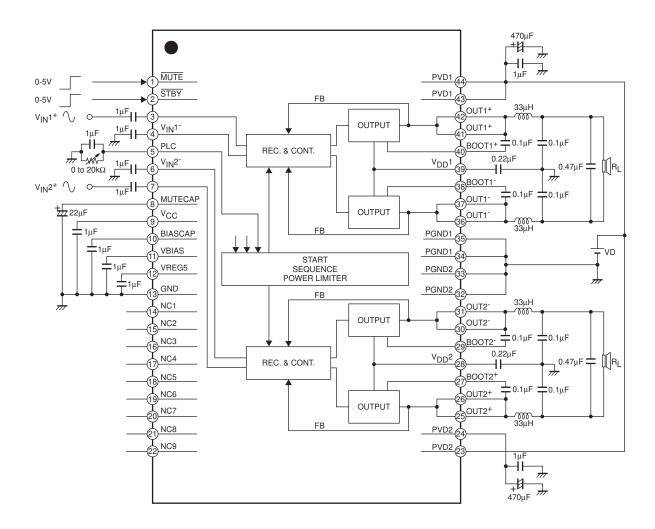




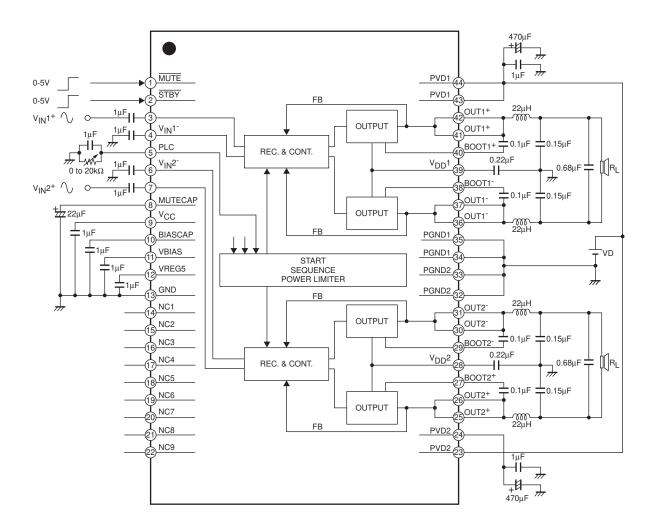
## **Pin Assignment**



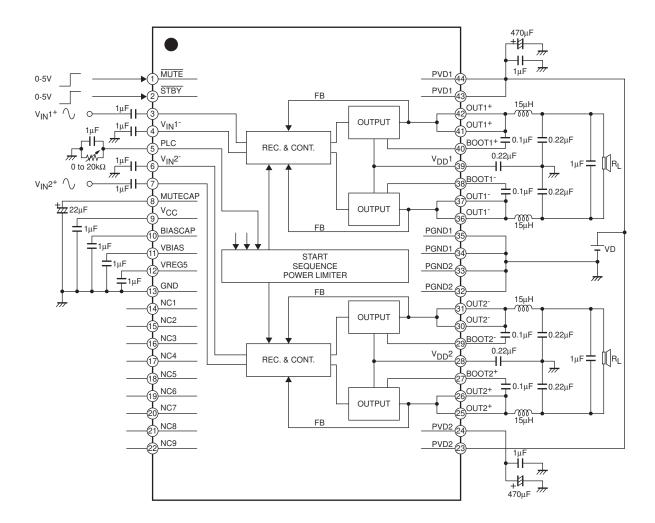
# Block Diagram and Application Circuit Example 1 $(R_L$ = $8\Omega)$



# Application Circuit Example 2 ( $R_L$ = $6\Omega$ )



# Application Circuit Example 3 $(R_L$ = $4\Omega)$



# LV49152V

# Pin Equivalent Circuit

Pin No.	Pin name	I/O	Description	Equivalent Circuit
1	MUTE	I	Mute control pin	VD
				250kΩ≨
				<b>↑</b>
				10kΩ
				100kΩŞ
				CND
2	STBY	I	Standby control pin	GND
2	3161	'	Standby Control pin	VD
				250kΩ ≹
				10kΩ   K
				100kΩ ≸
				GND
3	V <sub>IN</sub> 1+	I	Input pin, CH1 plus	VD
				300Ω
				GND VBIAS
4	V <sub>IN</sub> 1-	I	Input pin, CH1 minus	VD_
				<b>*</b>
				4 300Ω
				\$30kΩ +
				GND VBIAS
5	PLC	ı	Power level control pin	VD
				<u>φ</u> ά
				<b> </b>
				5 200Ω
				<b>A</b>
				GND T_

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Pin No.	Pin name	I/O	Description	Equivalent Circuit				
6	V <sub>IN</sub> 2 <sup>-</sup>	1	Input pin, CH2 minus	VD 300Ω \$30kΩ VBIAS GND				
7	V <sub>IN</sub> 2+	I	Input pin, CH2 plus	VD 300Ω ₹30kΩ VBIAS GND				
8	MUTECAP	0	Muteing sysytem capcitor connection	$V_{DD}$ $V$				
9	Vcc	0	Internal power supply decupling capacitor connection	9 GND				
10	BIASCAP	0	Internal regulator decupling capacitor connection	VD 1kΩ 1kΩ 1kΩ 1kΩ M \$100kΩ				

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Pin No.	Pin name	I/O	Description	Equivalent Circuit				
11	VBIAS	0	Internal regulator decupling capacitor connection  Internal regulator decupling capacitor connection	VD \$500Ω GND VD				
				(12)				
13	GND		Analog Ground					
14	NC		Non connection					
15	NC		Non connection					
16	NC		Non connection					
17	NC		Non connection					
18	NC		Non connection					
19	NC		Non connection					
20	NC		Non connection					
21	NC		Non connection					
22	NC		Non connection					
23	PVD2		CH2 power supply					
24	PVD2	_	CH2 power supply					
25	OUT2+	0	Output pin, CH2 plus	VD (25) (SND)				
26	OUT2+	0	Output pin, CH2 plus	VD (26) (GND)				

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Pin No.	Pin name	I/O	Description	Equivalent Circuit				
27	BOOT2+	I/O	Boot strap pin, CH2 plus					
28	V <sub>DD</sub> 2	0	CH2 internal regulator decupling capacitor connection					
29	BOOT2 <sup>-</sup>	I/O	Boot strap pin, CH2 minus					
30	OUT2-	0	Output pin, CH2 minus	VD 30 GND				
31	OUT2-	0	Output pin, CH2 minus	VD ————————————————————————————————————				
32	PGND2		CH2 Power Ground					
33	PGND2		CH2 Power Ground					
34	PGND1		CH1 Power Ground					
35	PGND1		CH1 Power Ground					
36	OUT1 <sup>-</sup>	0	Output pin, CH1 minus	VD 36				
37	OUT1-	0	Output pin, CH1 minus	VD  WE  GND				
38	BOOT1-	I/O	Boot strap pin, CH1 minus					
39	V <sub>DD</sub> 1	0	CH1 internal regulator decupling capacitor connection					
40	BOOT1+	I/O	Boot strap pin, CH1 plus					

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Pin No.	Pin name	I/O	Description	Equivalent Circuit
41	OUT1+	0	Output pin, CH1 plus	VD (41)
42	OUT1+	0	Output pin, CH1 plus	VD (42)
43	PVD1		CH1 power supply	
44	PVD1		CH1 power supply	

#### **Operation Mode Summary**

#### STBY mode ( $\overline{STBY} = L$ and $\overline{MUTE} = L$ )

Each bias becomes off state when the regulator in IC has been turned off.

The most of circuits becomes off state.

The supply current :  $1\mu A$  (typical).

#### MUTE mode ( $\overline{STBY} = H$ and $\overline{MUTE} = L$ )

Each bias becomes on state when the regulator in IC has been turned on.

When more than half of the circuits are active, the amplifier in the output stages become off.

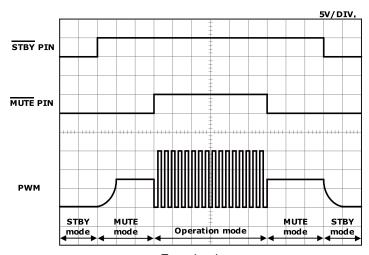
The supply current : 20mA (typical).

#### Operation mode ( $\overline{STBY} = H$ and $\overline{MUTE} = H$ )

The LV49152V operates as D-class amplifier.

The output signal is synchronized with the input signal.

The supply current : 45mA (typical)

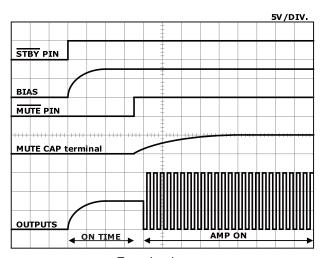


Function image

#### ON TIME/OFF TIME

#### **ON TIME**

Please secure ON TIME of 350msec or more for reducing Pop noise.

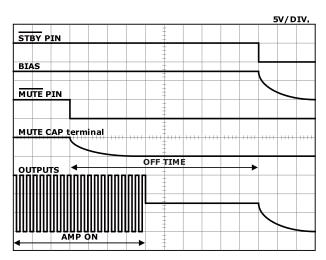


Function image

ON TIME • • • the time until the MUTE pin is set to high level after the STBY pin is set to high level

#### OFF TIME

Please secure OFF TIME of 1000msec or more for reducing Pop noise.



Function image

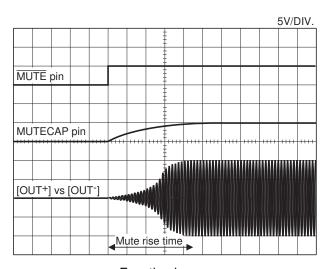
OFF TIME ••• the time until the  $\overline{STBY}$  pin is set to low level after the  $\overline{MUTE}$  pin is set to low level

#### **SOFT MUTE**

The soft mute circuit is able to use fade in/fade out function, and can set Rise time and fall time by the time constant of the MUTECAP capacitor.

#### FADE IN

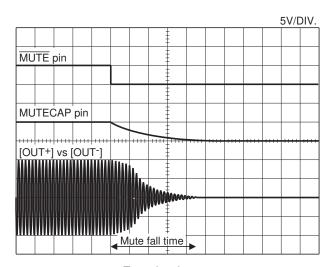
Mute rise time is Applpx.450msec in our recommended external components.



Function image

#### FADE OUT

Mute fall time is Applpx.450msec in our recommended external components.



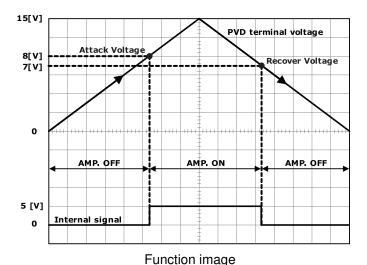
Function image

#### Power supply lowering protection circuit

Since the instable operation in the low voltage is prevented by using this circuit, after the voltage of the PVD pin is monitored and the voltage below the Attack voltage (PVD = 8V typ.), AMP is turned off.

Also, to prevent the instable operation when the voltage of the PVD pin is decreased by any cause during operations, the Attack voltage (PVD = 7V typ.) is set.

The voltage of Attack and Recover has hysteresis (About 1V) to prevent ON/OFF continuous action of the power supply lowering protection circuit.



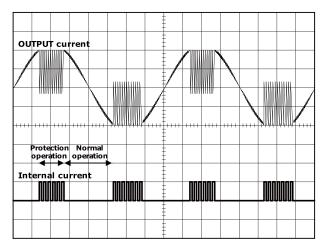
Also, this IC is designed to turn off AMP in the same sequence that the MUTE is on as a pop noise measures when the plug of products are put off.

#### Over current protection circuit

The over current protection circuit is a protection circuit \* to protect the output DMOS from the over current and corresponds to any mode of the power supply, GND and a load short.

The protection operation is performed when the current reaches the detection current value set out in IC and the output DMOS is compulsorily turned off for about 20µsec.

After compulsorily tuning off the output DMOS, when the Amplifier is automatically reset in usual operation and the over current flows continuously, the protection operation is performed again.



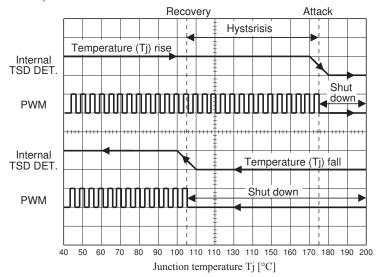
Function image

<sup>\*</sup> The over current protection circuit is a function to avoid the abnormal state like the output short-circuit temporarily. Unfortunately, we cannot guarantee that IC is not destroyed.

#### Thermal protection circuit

The LV49152V includes a thermal protection circuit to prevent damage to or destruction of the IC should abnormal internal heat generation occur.

This means that should the IC junction temperature (Tj) rise above about 175°C due to inadequate heat dissipation or other reason, the thermal protection circuit will operate to stop IC operation should the temperature rise further. If the temperature is reduced by lowering the input level or other means, the thermal protection circuit will recover automatically (about 105°C).



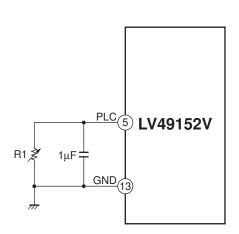
Function image

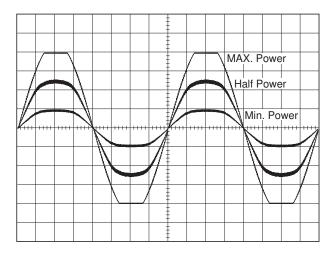
<sup>\*</sup> The thermal protection circuit is a function to avoid the abnormal state temporarily. Unfortunately, we cannot guarantee that IC is not destroyed.

#### **PLC**

The PLC (power level control) function is able to control the maximum index modulation by setting a value of external PLC resistance R1 voluntarily, and prevent a PWM signal from becoming the over modulation mode. In addition, this circuit can be use as output power limit circuit because the PLC function can set the maximum index modulation voluntarily, and variable from 2W to 15W with output power linearly in the state that made the power supply voltage and load resistance fixation. Because the PLC function can set the suitable rated output with the same power supply voltage/speaker regardless of screen size in flat screen televisions by this, set can plan the commonization of the board.

Furthermore, The PLC function can reduce abnormal noise in the hard clip so that output wave pattern becomes the soft clip when it limited output power.

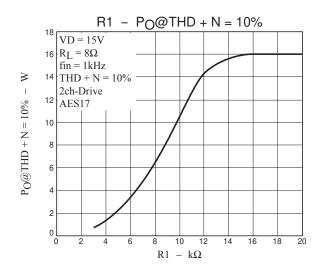




Function image

#### Measuring condition

VD = 15V,  $R_L = 8\Omega$ ,  $L = 33\mu H$  (TOKO : A7502BY-330M),  $C = 0.1 \mu F$ ,  $C_L = 0.47 \mu F$ ,  $Ta = 25^{\circ}C$ 

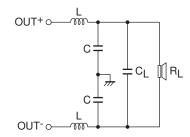


R1 [kΩ]	Po@10% [W]
3.0	0.694
3.6	1.073
4.7	1.982
6.2	3.642
7.5	5.562
8.2	6.855
9.1	8.591
10	10.64
13	15.32
15	15.94
20	16.01

Setting example of the output power limit value

- \* When it is used this function as output power limit, please use the high-precision resistance such as the metal film resistor when precision of the electricity value is necessary.
- \* The value of external PLC resistance R1 please connects more than  $3k\Omega$ .
- \* When it is changed a value of external PLC resistance R1, please turn off an amplifier.

#### Cut-off frequency calculation method and the output LC filter setting



The cut off frequency fc of the output LC filter is calculated by the following formula.

$$fc = \frac{1}{2\pi\sqrt{2LC_L}}$$

Also, by setting the cut off frequency fc, the value of CL and L is calculated by using the following formula.

$$C_L = \frac{1}{2\sqrt{2} \times \pi \ R_L fc}$$

$$L = \frac{\sqrt{2} \times R_L}{4\pi \text{ fc}}$$

In general, the value from 20% to 30% of CL is set to C.

In case of fc = 30kHz

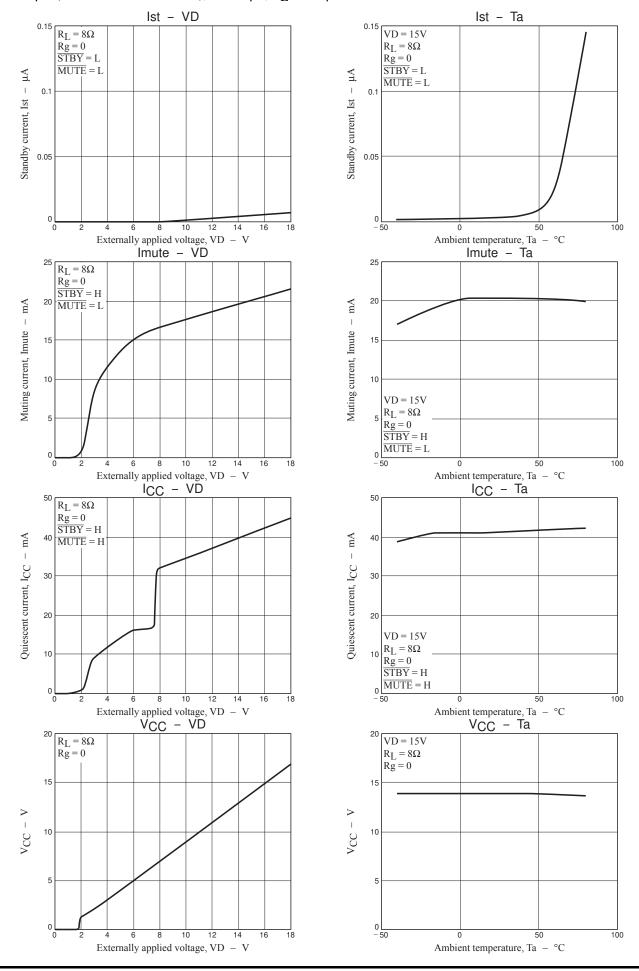
R <sub>L</sub> [Ω]	L [μH]	C <sub>L</sub> [μF]	C [μF]	Q
4	15	1	0.22	0.650
6	22	0.68	0.15	0.636
8	33	0.47	0.1	0.704
16	68	0.22	0.047	0.739

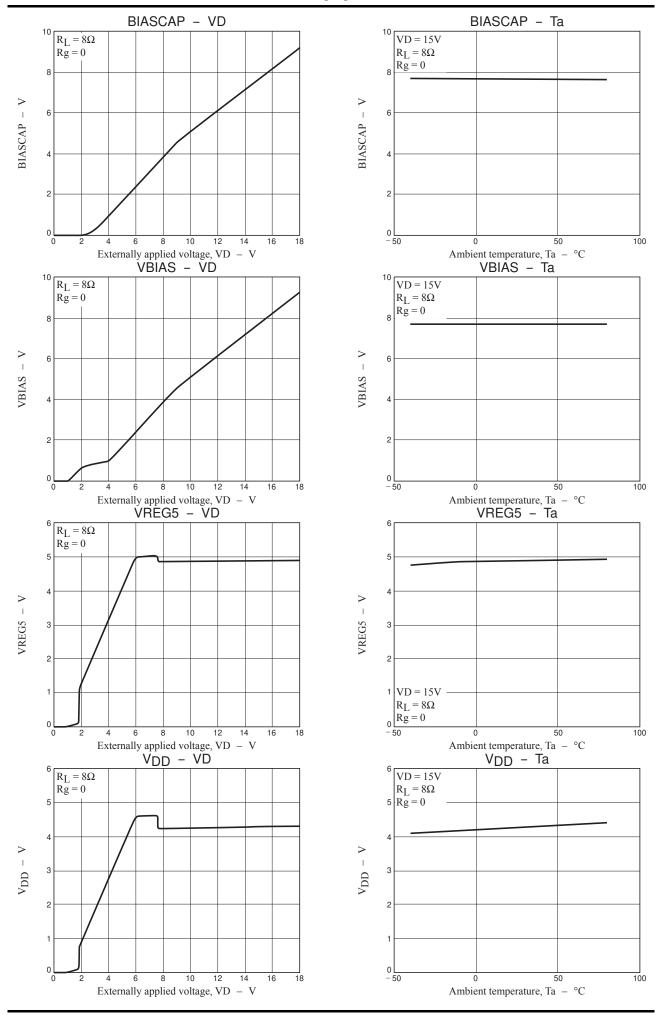
Above formula is common calculation method and is a measure of constant setting. In fact, it is necessary to set with each set that considers the speaker characteristics.

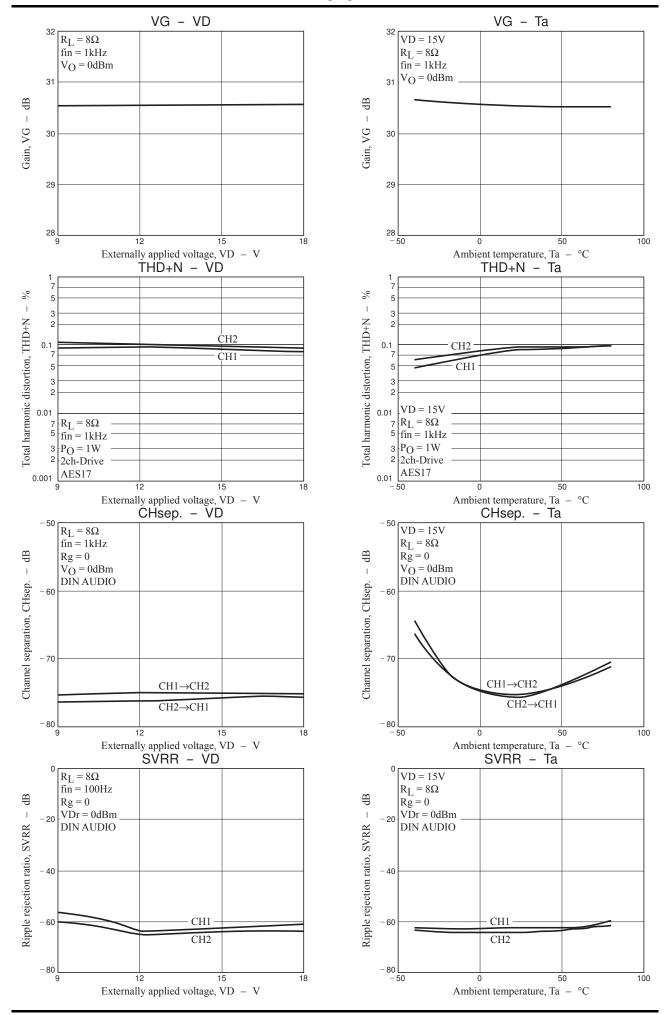
In addition, please set the fixed number to become  $Q \le 1$  in currents in the fc neighborhood increasing if Q value of the LC filter is big.

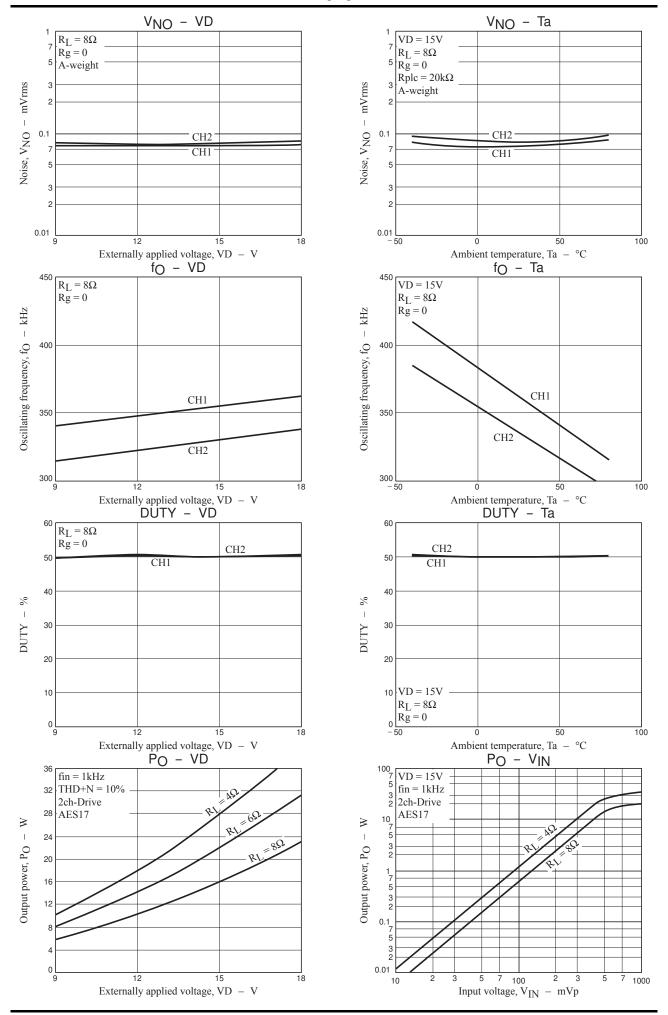
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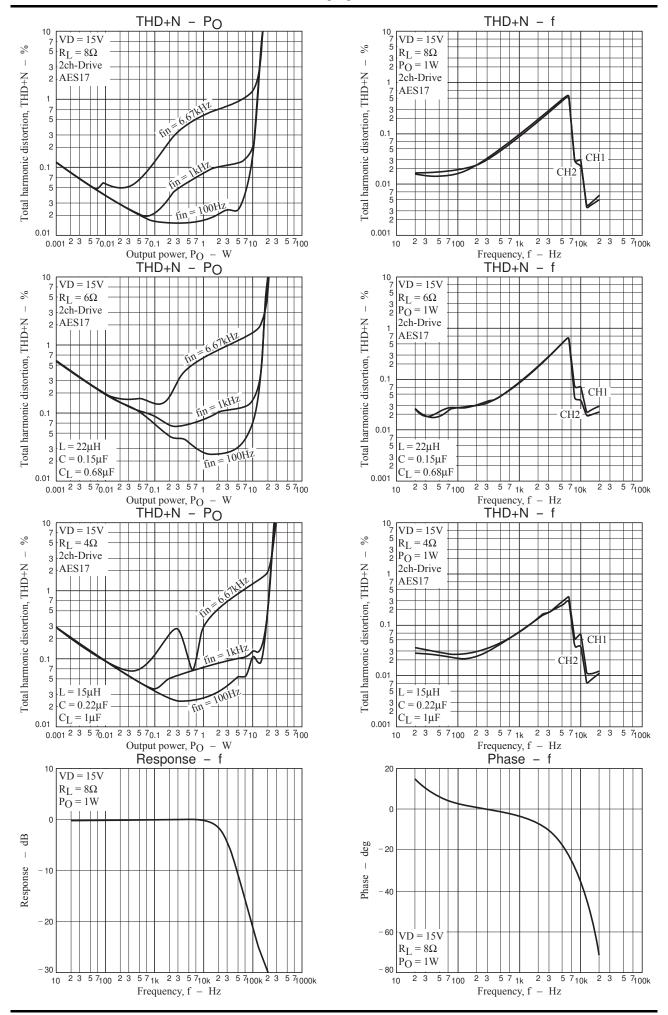
 $L = 33\mu H$  (TOKO : A7502BY-330M),  $C = 0.1\mu F$ ,  $C_L = 0.47\mu F$ 

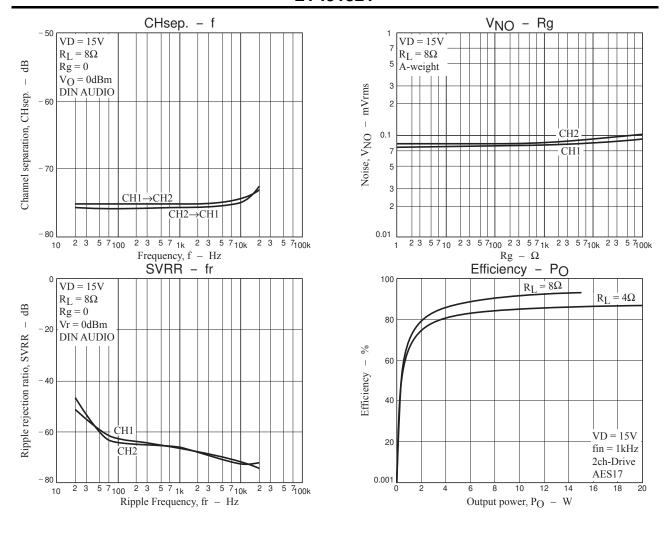












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