## imall

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### Monolithic Linear IC Multi-Power Supply System IC for Car Audio Systems



The LV56801P is a multi-power supply system IC that provides four regulator outputs and two high side switches as well as a number of protection functions including overcurrent protection, overvoltage protection and overheat protection. It is an optimal power supply IC for car audio and car entertainment systems and similar products.

#### Features

- Four regulator output systems
  - For microcontroller: 3.3V output voltage, 200mA maximum output current
  - For CD drive: 8.0V output voltage, 1300mA maximum output current

For illumination: 8 to 12V output voltage (output can be set with external resistors), 300mA maximum output current

For audio systems: 8 to 9V output voltage (output voltage can be set with external resistors), 300mA maximum output current

• Two V<sub>CC</sub>-linked high side switch systems

EXT: 350mA maximum output current, 0.5V voltage difference between input and output.

ANT: 300mA maximum output current, 0.5V voltage difference between input and output.

• Two V<sub>DD</sub> 3.3V-linked high side switch systems

SW5V: 200mA maximum output current, 0.25V voltage difference between input and output.

ACC (accessory voltage detection output): 100mA maximum output current, 0.25V voltage difference between input and output.

- Overcurrent protection function
- Overvoltage protection function, typ 21V (excluding VDD 3.3V output)
- Overheat protection function, typ 175°C
- On-chip accessory voltage detection circuit
- P-channel LDMOS used for power output block

CAUTION)

The protection functions are provided in order to improve the ability of the ICs to withstand breakdown, and they are not intended to guarantee safety when used under conditions outside the safe operating area or rated operating conditions.

Use of the ICs under any conditions exceeding the safe operating area or above the IOmax, and especially use in overcurrent protection areas or under conditions in which they are subject to thermal protection, may reduce their reliability and result in permanent breakdown.



#### Specifications

#### **Absolute Maximum Ratings** at $Ta = 25^{\circ}C$

Parameter	Conditions	Conditions		Ratings	Unit
Supply voltage	V <sub>CC</sub> max			36	V
Peak supply voltage	V <sub>CC</sub> peak	See below for the waveform a	applied.	50	V
Allowable Power dissipation	Pd max	Independent IC	Ta ≤ 25°C	1.5	W
		Al heat sink *		5.6	W
		With an infinity heat sink		32.5	W
Junction temperature	Tj max			150	°C
Operating ambient temperature	Topr			-40 to +85	°C
Storage temperature	Tstg			-55 to +150	°C

\* : When the Aluminum heat sink (50mm  $\times$  50mm  $\times$  1.5mm) is used

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

#### Allowable Operating range at Ta = 25°C

Parameter	Conditions	Ratings	Unit
Operating supply voltage 1	V <sub>DD</sub> output, SW output, ACC output	7.5 to 16	V
Operating supply voltage 2 ILM output at 10V		12 to 16	V
	ILM output at 8V	10 to 16	V
Operating supply voltage 3 Audio output at 9V		10 to 16	V
Operating supply voltage 4 CD output (CD output current = 1.3A)		10.5 to 16	V
	CD output (CD output current $\leq$ 1A)	10 to 16	V

#### Electrical Characteristics at $Ta = 25^{\circ}C$ , $V_{CC} = 14.4V$

\*: All the specifications are defined based on the tests that Tj is almost equal to Ta (=25°C). To suppress the rise of Tj in the junction temperature as much as possible, it tests by the pulse loading.

Parameter	Cumbel	Conditions		Ratings		Unit	
Parameter	Symbol	Conditions	min	typ	max	Unit	
Current drain	ICC	$V_{DD}$ no load, CTRL1/2 = $[L/L]$ , ACC = 0V		400	800	μA	
CTRL1 Input				·			
Low input voltage	V <sub>IL</sub> 1		0		0.5	V	
M1 input voltage	V <sub>IM1</sub> 1		0.8	1.1	1.4	V	
M2 input voltage	V <sub>IM2</sub> 1		1.9	2.2	2.5	V	
High input voltage	V <sub>IH</sub> 1		2.9	3.3	5.5	V	
Input impedance	R <sub>IH</sub> 1		350	500	650	kΩ	
CTRL2 Input	·	·	•				
Low input voltage	V <sub>IL</sub> 2		0		0.5	V	
M input voltage	V <sub>IM</sub> 2		1.1	1.65	2.1	V	
High input voltage	V <sub>IH</sub> 2		2.5	3.3	5.5	V	
Input impedance	R <sub>IH</sub> 2		350	500	650	kΩ	
V <sub>DD</sub> 3.3V Output *1		The V <sub>DD</sub> 3.3V output supplie	es the output c	urrents of SW	/ 3.3V and A	CC 3.3V	
Output voltage 1	V <sub>O</sub> 1	I <sub>O</sub> 1 = 200mA, I <sub>O</sub> 7, I <sub>O</sub> 8 = 0A	3.13	3.3	3.47	V	
Output voltage 2	V <sub>O</sub> 1'	I <sub>O</sub> 1 = 200mA, I <sub>O</sub> 7 = 200mA, I <sub>O</sub> 8 = 100mA	3.13	3.3	3.47	V	
Output total current	Ito1	$V_01 \ge 3.13V$ , $Ito1 = I_01 + I_07 + I_08$	500			mA	
Line regulation	ΔV <sub>OLN</sub> 1	$7.5V < V_{CC} < 16V, I_O1 = 200mA *2$		30	90	mV	
Load regulation	$\Delta V_{OLD}$ 1	1mA < I <sub>O</sub> 1 < 200mA *2		70	150	mV	
Dropout voltage 1	V <sub>DROP</sub> 1	I <sub>O</sub> 1 = 200mA *2		1.0	1.5	V	
Dropout voltage 2	V <sub>DROP</sub> 1'	I <sub>O</sub> 1 = 100mA *2		0.7	1.05	V	
Dropout voltage 3	V <sub>DROP</sub> 1"	$I_{O}1 + I_{O}7 + I_{O}8 = 500$ mA		2.5	3.75	V	
Ripple rejection	R <sub>REJ</sub> 1	f = 120Hz, I <sub>O</sub> 1 = 200mA *2	40	50		dB	
CD Output ; CTRL2 = H	•	•					
Output voltage	V <sub>O</sub> 2	I <sub>O</sub> 2 = 1000mA	7.6	8.0	8.4	V	

\*1 : The V<sub>DD</sub> 3.3V output also supplies the output currents of SW 3.3V and ACC 3.3V. Therefore, the current supply capability of the V<sub>DD</sub> 3.3V output and its other electrical characteristics are affected by the output statuses of SW 3.3V and ACC 3.3V. \*2 : SW 3.3V and ACC 3.3V are not subject to a load.

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Parameter	Symbol	Conditions	Ratings			Unit
Falametei	Symbol	Conditions	min	typ	max	Unin
Output current	I <sub>O</sub> 2	$V_{O}2 \ge 7.6V$	1300			mA
Line regulation	$\Delta V_{OLN}^2$	$10.5V < V_{CC} < 16V, I_O2 = 1000mA$		50	100	mV
Load regulation	$\Delta V_{OLD}^2$	10mA < I <sub>O</sub> 2 < 1000mA		100	200	mV
Dropout voltage 1	V <sub>DROP</sub> 2	I <sub>O</sub> 2 = 1000mA		1.0	1.5	V
Dropout voltage 2	V <sub>DROP</sub> 2'	I <sub>O</sub> 2 = 500mA		0.5	0.75	V
Ripple rejection	R <sub>REJ</sub> 2	f = 120Hz, I <sub>O</sub> 2 = 1000mA	40	50		dB
AUDIO (8-9V) Output ; CTRL2 =	ГмЈ		•			
AUDIO_F pin voltage	V <sub>I</sub> 3		1.222	1.260	1.298	V
AUDIO_F pin inflow current	I <sub>IN</sub> 3		-1		1	μA
AUDIO output voltage 1	V <sub>O</sub> 3	I <sub>O</sub> 3 = 200mA, R2 = 30kΩ, R3 = 5.6kΩ *3	7.65	8.0	8.35	V
AUDIO output voltage 2	V <sub>O</sub> 3'	I <sub>O</sub> 3 = 200mA, R2 = 27kΩ, R3 = 4.7kΩ *3	8.13	8.5	8.87	V
AUDIO output voltage 3	V <sub>O</sub> 3"	$I_O^3 = 200$ mA, R2 = 24k $\Omega$ , R3 = 3.9k $\Omega$ *3	8.6	9.0	9.4	V
AUDIO output current	I <sub>O</sub> 3		300			mA
Line regulation	∆V <sub>OLN</sub> 3	$10V < V_{CC} < 16V, I_{O}3 = 200mA$		30	90	mV
Load regulation	∆V <sub>OLD</sub> 3	1mA < I <sub>O</sub> 3 < 200mA		70	150	mV
Dropout voltage 1	V <sub>DROP</sub> 3	$I_{O3} = 200$ mA		0.3	0.45	V
Dropout voltage 2	V <sub>DROP</sub> 3'	$I_{O3} = 100 \text{mA}$		0.15	0.23	V
Ripple rejection	R <sub>REJ</sub> 3	f = 120Hz, I <sub>O</sub> 3 = 200mA	40	50		dB
ILM (8-12V) Output ; CTRL1 =	- -		-			
ILM_F pin voltage	V <sub>1</sub> 4		1.222	1.260	1.298	V
ILM output voltage 1	V <sub>O</sub> 4	I <sub>O</sub> 4 = 200mA	11.4	12.0	12.6	V
ILM output voltage 2	V <sub>O</sub> 4'	$I_{\Omega}4 = 200$ mA, R1 = 270 k $\Omega$ *4	8.5	10.0	11.5	v
ILM output voltage 3	V <sub>O</sub> 4"	$I_{O}4 = 200$ mA, R1 = 100 k $\Omega$ *4	6.8	8.0	9.2	v
ILM output current	I <sub>0</sub> 4	R1 = 270kΩ	300	0.0	0.2	mA
Line regulation	ΔV <sub>OLN</sub> 4	$12V < V_{CC} < 16V, I_{O}4 = 200 \text{mA}, \text{R1} = 270 \text{k}\Omega$	000	30	90	mV
Load regulation	ΔV <sub>OLD</sub> 4	1mA < I <sub>O</sub> 4 < 200mA		70	150	mV
Dropout voltage 1		$I_0 4 = 200 \text{mA}$		0.7	1.05	V
Dropout voltage 2	VDROP4	$I_0 4 = 100 \text{mA}$		0.35	0.53	V
Ripple rejection	V <sub>DROP</sub> 4'	$f = 120Hz, I_0 4 = 200mA$	40	50	0.55	dB
Remoto (EXT) ; CTRL1 = M2	R <sub>REJ</sub> 4	1 - 120112, 104 - 20011A	40	50		UD
Output voltage	V <sub>O</sub> 5	I <sub>O</sub> 5 = 350mA	V <sub>CC</sub> -1.0	Voc 0.5		v
	-		350	V <sub>CC</sub> -0.5		
	I <sub>O</sub> 5	$V_{O}5 \ge V_{CC}-1.0$	350			mA
ANT remoto ; CTRL1 = [H]	N/ C	1.0.000	V 10	N/ 05		v
Output voltage	V <sub>O</sub> 6	I_06 = 300mA	V <sub>CC</sub> -1.0	V <sub>CC</sub> -0.5		
Output current	I <sub>O</sub> 6	$V_{O}6 \ge V_{CC}-1.0$	300			mA
SW 3.3V Output ; CTRL2 = M	N/ 7					
Output voltage 1	V <sub>0</sub> 7	$I_07 = 1$ mA, $I_01$ , $I_08 = 0$ A *5	V <sub>O</sub> 1-0.1	V <sub>O</sub> 1		V
Output voltage 2	V <sub>0</sub> 7'	$I_07 = 80\text{mA}, I_01, I_08 = 0\text{A} + 5$	V <sub>O</sub> 1-0.22	V <sub>O</sub> 1-0.1		V
Output voltage 3	V <sub>0</sub> 7"	$I_07 = 200$ mA, $I_01$ , $I_08 = 0$ A *5	V <sub>O</sub> 1-0.55	V <sub>O</sub> 1-0.25		V
Output current	I <sub>O</sub> 7	$V_{O}^{7} \ge 2.88$	200			mA
ACC Detection ; ACC Integration			1			
ACC detection voltage	V <sub>TH</sub> 8		2.75	3.0	3.25	V
Hysteresis width	V <sub>HIS</sub> 8		0.2	0.3	0.4	V
Input impedance	ZI8	(Pull-down resistance internal)	42	60	78	kΩ
ACC output voltage 1	V <sub>O</sub> 8	I <sub>O</sub> 8 = 0.5mA, I <sub>O</sub> 1, I <sub>O</sub> 7 = 0A *5	V <sub>O</sub> 1-0.1	V <sub>O</sub> 1		V
ACC output voltage 2	V <sub>O</sub> 8'	I <sub>O</sub> 8 = 40mA, I <sub>O</sub> 1, I <sub>O</sub> 7 = 0A *5	V <sub>O</sub> 1-0.22	V <sub>O</sub> 1-0.1		V
ACC output voltage 3	V <sub>O</sub> 8''	I <sub>O</sub> 8 = 100mA, I <sub>O</sub> 1, I <sub>O</sub> 7 = 0A *5	V <sub>O</sub> 1-0.55	V <sub>O</sub> 1-0.25		V
ACC output voltage	IO8	$V_{O}8 \ge 2.88$	100			mA

\*3 : When a component with a resistance accuracy of  $\pm 1\%$  is used

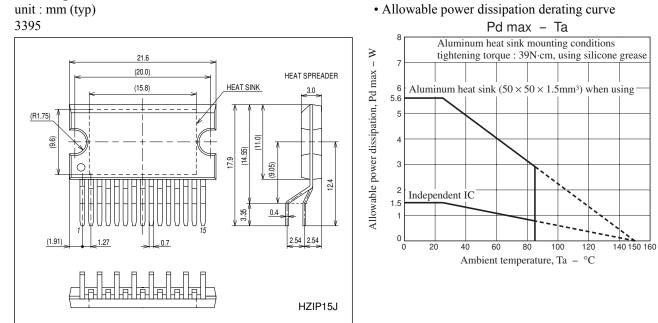
<Reference> When a component with a resistance accuracy of  $\pm 0.5\%$  is used, V\_O3" is 8.67V  $\leq 9.0V \leq 9.33V.$ 

\*4 : When a component with a resistance accuracy of  $\pm 1\%$  is used

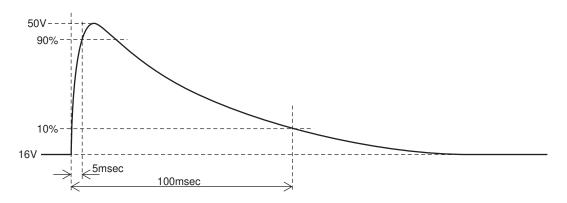
The absolute accuracy of the internal resistance is  $\pm 15\%$ .

\*5: Since the SW 3.3V and ACC 3.3V are output from V<sub>DD</sub> 3.3V through the SW, the voltage drops by an amount equivalent to the ON resistance of the SW.

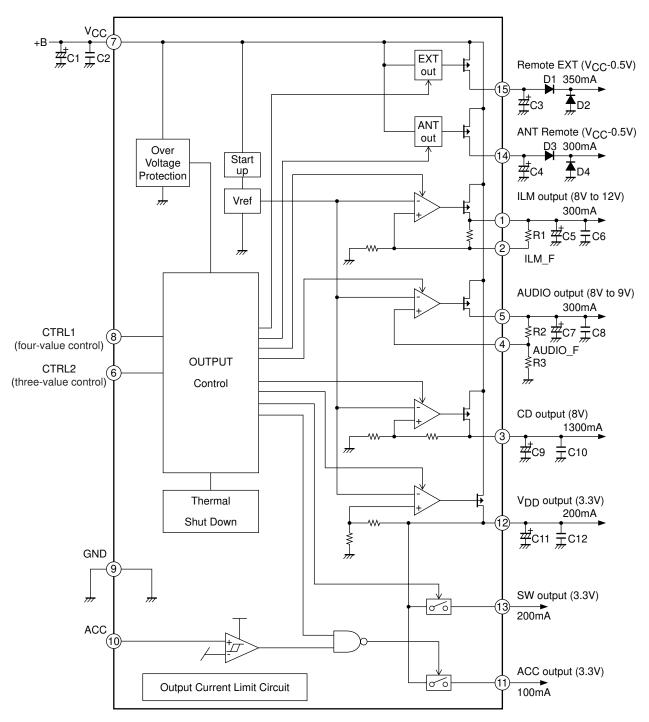
#### Package Dimensions



• Waveform applied during surge test



#### **Block Diagram**



#### **Pin Function**

Pin No.	Pin name	Description	Equivalent Circuit
1	ILM	ILM output pin ON when CTRL1 = M1, M2, H 12.0V/300mA	
2	ILM_F	ILM output voltage adjustment pin	(2) + (1)

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	om preceding pag		
Pin No.	Pin name	Description	Equivalent Circuit
3	CD	CD output pin ON when CTRL2 = M, H 8.0V/1.3A	$\begin{array}{c} \hline 7 \\ \hline \\ 3 \\ \hline \\ \\ \hline \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\$
4	AUDIO_F	AUIDO output voltage adjustment pin	
5	AUDIO	AUDIO output pin ON when CTRL2 = M, H	
6	CTRL2	CTRL2 input pin three-value input	
7	V <sub>CC</sub>	Supply terminal	
8	CTRL1	CTRL1 input pin four-value input	
	GND	GND pin	

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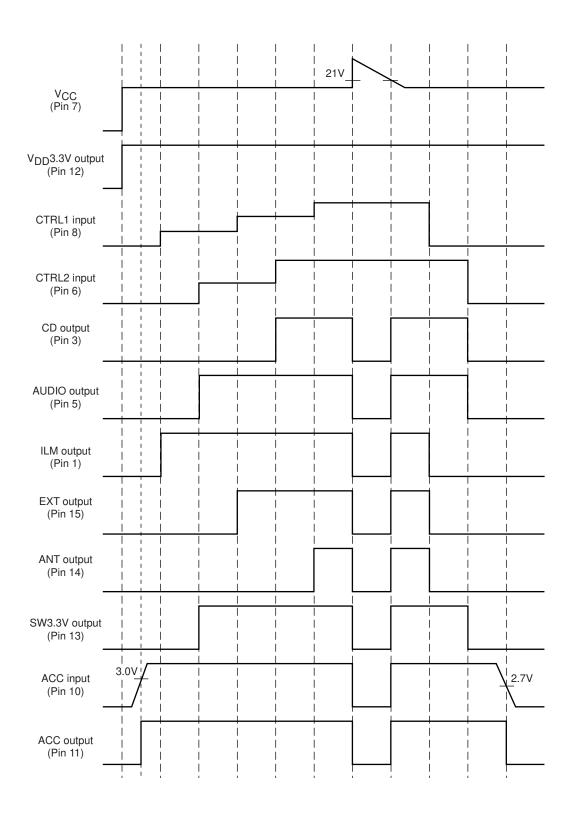
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Pin No.	Pin name	Description	Equivalent Circuit		
10	ACC	Accessory input	7 Vcc $45k\Omega$ $F$ $F$ $K\Omega$ 9 $GND$		
11	ACC3.3V	Accessory detection output ON when ACC > 3V			
12	V <sub>DD</sub> 3.3V	V <sub>DD</sub> 3.3V output pin 3.3V/200mA			
13	SW3.3V	SW3.3V output pin ON when CTRL2 = M, H	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		
14	ANT	ANT output pin ON when CTRL1 = H V <sub>CC</sub> -0.5V/300mA			
15	EXT	EXT output pin ON when CTRL1 = M2, H V <sub>CC</sub> -0.5V/350mA			

#### CTRL Pin Output Truth Table

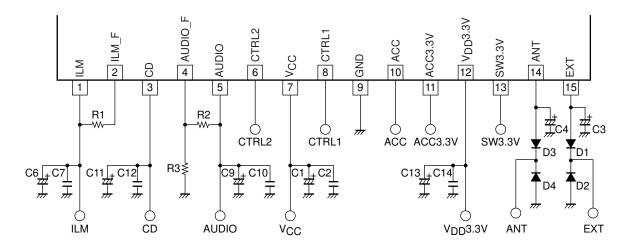
CTRL1	ANT	EXT	ILM
L	OFF	OFF	OFF
M1	OFF	OFF	ON
M2	OFF	ON	ON
Н	ON	ON	ON

CTRL2	CD	AUDIO	SW5
UTRLZ	CD	AUDIO	3005
L	OFF	OFF	OFF
М	OFF	ON	ON
Н	ON	ON	ON

#### **Timing Chart**



#### **Recommended Operation Circuit**



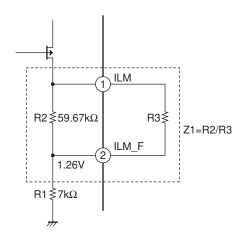
#### Peripheral parts list

Name of part	Description	Recommended value	Remarks
C1	Power supply bypass capacitor	100µF or more	These capacitors must be placed near
C2	Oscillation prevention capacitor	0.22µF or more	the $V_{CC}$ and GND pins.
C3	EXT output stabilization capacitor	2.2µF or more	
C4	ANT output stabilization capacitor	2.2µF or more	
C5, C7, C9, C11	Output stabilization capacitor	4.7μF or more	Electrolytic capacitor *
C6, C8, C10, C12	Output stabilization capacitor	0.22µF or more	Ceramic capacitor *
R1	Resistor for ILM voltage adjustment	ILM output voltage R1:without = 12.0V :270k $\Omega$ = 10.0V :100k $\Omega$ = 8.0V	A resistor with resistance accuracy as low as less than $\pm1\%$ must be used.
R2, R3	Resistor for AUDIO voltage setting	AUDIO output voltage R2/R3:30kΩ/5.6kΩ = 8.0V :27kΩ/4.7kΩ = 8.5V :24kΩ/3.9kΩ = 9.0V	A resistor with resistance accuracy as low as less than $\pm 1\%$ must be used.
D1, D2, D3, D4	Diode for internal device breakdown protection		

\* : In order to stabilize the regulator outputs, it is recommended that the electrolytic capacitor and ceramic capacitor be connected in parallel.

Furthermore, the values listed above do not guarantee stabilization during the overcurrent protection operations of the regulator, so oscillation may occur during an overcurrent protection operation.

• ILM output voltage setting method



The ILM\_F voltage is determined by the internal band gap voltage of the IC (typ = 1.26V).

Formula for ILM voltage calculation

$$Z_{1} = R_{2} / / R_{3} = \frac{R_{2} \cdot R_{3}}{R_{2} + R_{3}}$$
$$ILM = \frac{1.26[V]}{R_{1}} \times Z_{1} + 1.26[V]$$
$$Z_{1} = \frac{(ILM - 1.26) \cdot R_{1}}{1.26} \qquad R_{3} = \frac{R_{2} \cdot Z_{1}}{R_{2} - Z_{1}}$$

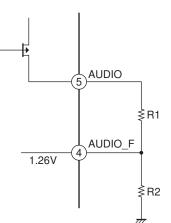
Example : ILM = 9V setting method

$$Z_1 = \frac{(9V - 1.26V) \cdot 7k\Omega}{1.26V} \cong 43k\Omega$$

When R3 = 150k, the ILM output voltage will be as follows:

$$Z_{1}' = \frac{59.67k\Omega \cdot 150k\Omega}{59.67k\Omega + 150k\Omega} \cong 42.69k\Omega$$
$$ILM = \frac{1.26V}{7k\Omega} \times 42.69k\Omega + 1.26V \cong \boxed{8.94V}$$

#### • AUDIO output voltage setting method



The AUDIO\_F voltage is determined by the internal band gap voltage of the IC (typ = 1.26V).

Formula for AUDIO voltage calculation

$$AUDIO = \frac{1.26[V]}{R_2} \times R_1 + 1.26[V]$$
$$\frac{R_1}{R_2} = \frac{(AUDIO - 1.26)}{1.26}$$

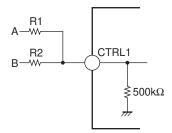
The circuit must be designed in such a way that the R1:R2 ratio satisfies the formula given above for the AUDIO voltage that has been set.

Example : AUDIO = 8.5V setting method

$$\frac{R_1}{R_2} = \frac{(8.5 - 1.26)}{1.26} \cong 5.75$$
$$\frac{R_1}{R_2} = \frac{27k\Omega}{4.7k\Omega} \cong 5.74$$
$$AUDIO = 1.26V \times 5.74 + 1.26V \cong 8.49V$$

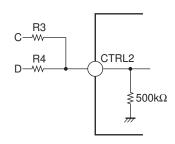
Note : In the above, the typical values are given in all instances for the values used and, as such, they will vary due to the effects of production-related variations of the IC and external resistors.

#### CTRL1 Application Circuit Example



(1)	1) <u>3.3V input: <math>R1 = 4.7k\Omega</math>, <math>R2 = 10k\Omega</math></u>				
	А	В	CTRL1		
	0V	0V	0V		
	0V	3.3V	1.05V		
	3.3V	0V	2.23V		
	3.3V	3.3V	3.20V		

• CTRL2 Application Circuit Example

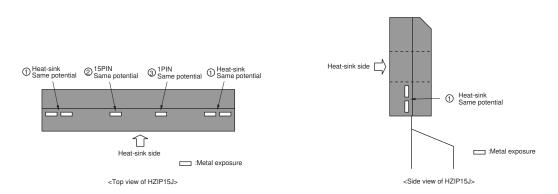


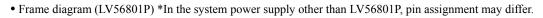
(1)	(1) 3.3V input: $R3 = R4 = 4.7k\Omega$				
	А	В	CTRL2		
	0V	0V	0V		
	0V	3.3V	1.61V		
	3.3V	0V	1.61V		
	3.3V	3.3V	3.29V		

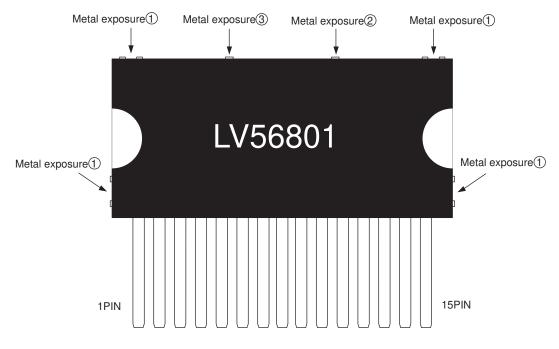
#### Caution for implementing LV56801P to a system board

In HZIP15J, the package used in this IC, there are several metal exposure other than the connection pins and heat-sinks as shown in the following diagrams. In the diagrams, the electric potential of 2 and 3 are the same as Pin15 and Pin1, respectively. 2 (=Pin15) is EXT pin and 3 (=Pin1) is ILM output (regulator). When the IC is implemented to the system, make sure that no attachment clamp touches the exposed Pin1/ Pin15. When the exposed Pin1/ Pin15 touch the attachment clamp (same electrical potential as GND), ILM output or VCC enter the same state as time when GND was shorted. The electric potential of the exposed metal connected to heat-sinks 1 is the same as that of sub board of the IC (GND). Therefore, even if the exposed metal and GND of the system board are adjacent to each other, there should be no problem.

• HZIP15J external view



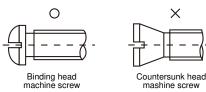


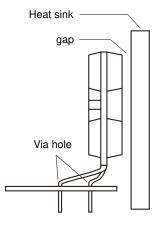


#### HZIP15J Heat sink attachment

Heat sinks are used to lower the semiconductor device junction temperature by leading the head generated by the device to the outer environment and dissipating that heat.

- a. Unless otherwise specified, for power ICs with tabs and power ICs with attached heat sinks, solder must not be applied to the heat sink or tabs.
- b. Heat sink attachment
  - $\cdot$  Use flat-head screws to attach heat sinks.
  - $\cdot$  Use also washer to protect the package.
  - · Use tightening torques in the ranges 39-59Ncm(4-6kgcm).
  - If tapping screws are used, do not use screws with a diameter larger than the holes in the semiconductor device itself.
  - Do not make gap, dust, or other contaminants to get between the semiconductor device and the tab or heat sink.
  - Take care a position of via hole .
  - $\cdot$  Do not allow dirt, dust, or other contaminants to get between the semiconductor device and the tab or heat sink.
  - · Verify that there are no press burrs or screw-hole burrs on the heat sink.
  - · Warping in heat sinks and printed circuit boards must be no more than
  - 0.05 mm between screw holes, for either concave or convex warping.
  - · Twisting must be limited to under 0.05 mm.
  - $\cdot$  Heat sink and semiconductor device are mounted in parallel.
  - Take care of electric or compressed air drivers
  - The speed of these torque wrenches should never exceed 700 rpm, and should typically be about 400 rpm.
- c. Silicone grease
  - $\cdot$  Spread the silicone grease evenly when mounting heat sinks.
  - · Our company recommends YG-6260 (Momentive Performance Materials Japan LLC)
- d. Mount
  - · First mount the heat sink on the semiconductor device, and then mount that assembly on the printed circuit board.
  - $\cdot$  When attaching a heat sink after mounting a semiconductor device into the printed circuit board, when tightening up a heat sink with the screw, the mechanical stress which is impossible to the semiconductor device and the pin doesn't hang.
- e. When mounting the semiconductor device to the heat sink using jigs, etc.,
  - $\cdot$  Take care not to allow the device to ride onto the jig or positioning dowel.
  - · Design the jig so that no unreasonable mechanical stress is not applied to the semiconductor device.
- f. Heat sink screw holes
  - · Be sure that chamfering and shear drop of heat sinks must not be larger than the diameter of screw head used.
  - $\cdot$  When using nuts, do not make the heat sink hole diameters larger than the diameter of the head of the screws used. A hole diameter about 15% larger than the diameter of the screw is desirable.
  - $\cdot$  When tap screws are used, be sure that the diameter of the holes in the heat sink are not too small. A diameter about 15% smaller than the diameter of the screw is desirable.
- g. There is a method to mount the semiconductor device to the heat sink by using a spring band. But this method is not recommended because of possible displacement due to fluctuation of the spring force with time or vibration.





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