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Ordering Information

Part Number	Package	Operating Temperature
M08888G-11 *	28 pin, 4.5 mm x 4.5 mm QFN	-40 °C to +85 °C
M08888-11EVM	Evaluation board with M08888-11	-40 °C to +85 °C

* The letter "G" designator in the part number indicates that the device is RoHS-compliant.

Revision History

Revision	Level	Date	Description
V4	Release	July 2015	Updated register references.
E (V3)	Release	September 2011	Updated Ordering Information
D (V2)	Release	August 2011	Add operating specifications at Tc=120°C in Tables 1-5, 1-6 and 1-8. Change pin24 from DIS to GND.
C (V1)	Release	June 2011	Update Product Specifications, Functional Descriptions and Register Descriptions
B (V1P)	Preliminary	October 2010	Update Product Specifications, Functional Description and Register Descriptions. Swapped pinout of ADDR0 (pin 19) and ADDR1 (pin 18).
A (V1A)	Advance	June 2010	Initial

Conventions

Throughout this document, pins will be identified with italics (example *IOUT1*) while x or X means 0,1,2 to indicate the different channels.

1.0 Product Specification

1.1 Absolute Maximum Ratings

These are the absolute maximum ratings at or beyond which the IC can be expected to fail or be damaged. Reliable operation at these extremes for any length of time is not implied.

Table 1-1. Absolute Maximum Ratings

Symbol	Parameter	Rating	Units
DV _{DD} , ALV _{DD}	1.8 V Digital and Analog Voltage at pins DVDD and ALVDD	1.98	V
DHV _{DD} , AHV _{DD}	3.3 V Digital and Analog Voltage at pins DVDD and ALVDD	3.63	V
IOUT0, IOUT1, IOUT2	Output pins for driving LED/Laser - maximum voltage	5.5	V
T _{JCTN}	Junction Temperature	-40 to +125	°C
T _{STG}	Storage Temperature	-65 to +150	°C
SERSEL	Serial data format select input	-0.4 to 3.63	V
APC_IN0, APC_IN1, APC_IN2	RGB Photodiode Feedback Input Voltage	-0.4 to ALV _{DD} + 0.4	V
I _{APC_IN0} , I _{APC_IN1} , I _{APC_IN2}	RGB Photodiode Feedback Input Current	-0.5 to 4	mA
I_VREF	Current into Reference Voltage Pin	-0.12 to +0.12	mA
I0_ON, I1_ON, I2_ON	Enable LED/Laser output	-0.4 to 3.63	V
CLK_IN, \overline{CS} , SI, SCLK, SO	SPI inputs and output	-0.4 to 3.63	V
SCLK_S, SDA_S	I ² C interface	-0.4 to 3.63	V
REGREF	External DC-DC converter control signal	-0.4 to DV _{DD} + 0.4	V
I_REGREF	Current into or out of REGREF	-0.12 to +0.12	mA

1.2 DC Characteristics

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V}$, $ALV_{DD} = 1.8\text{ V}$, $DHV_{DD} = 3.3\text{ V}$, $AHV_{DD} = 3.3\text{ V}$ unless otherwise noted.

Table 1-2. DC Characteristics

Symbol	PARAMETER	CONDITIONS	Min	Typ	Max	Units
DV_{DD}	1.8 V supply for digital circuitry		1.71	1.8	1.89	V
ALV_{DD}	1.8 V supply for analog circuitry		1.71	1.8	1.89	V
DHV_{DD}	3.3 V supply for digital circuitry		3.13	3.3	3.47	V
AHV_{DD}	3.3 V supply for analog circuitry		3.13	3.3	3.47	V
I_{LVDD}	1.8 V Supply Current (DV_{DD} and ALV_{DD}) ^{1, 2}	Open Loop	—	18.5	21.5	mA
		Closed Loop (IPC)	—	8.5	11	
		Closed Loop (CPC)	—	22	30	
I_{HVDD}	3.3 V (DHV_{DD} and AHV_{DD}) ¹ Standby Current - 3.3 V	Open Loop	—	7.5	13	mA
		Closed Loop (IPC) additional to open loop current	—	—	0.05	
		Closed Loop (CPC)	—	2.5	3.2	
LV_{POR}	1.8 V SET Threshold	For positive going supply	—	1.5	—	V
	1.8 V RESET Threshold	For negative going supply	—	1.4	—	
HV_{POR}	3.3 V SET Threshold	For positive going supply	—	2.72	—	V
	3.3 V RESET Threshold	For negative going supply	—	2.62	—	
T_c	Case Temp.	Measured on top of M08888 case	-40	—	85	$^\circ\text{C}$

NOTES:

- Excludes serial interface (SPI/I²C) current and LED current
- I_{LVDD} will be increase by 0.5% of the I_{x_OUT} current when I_{x_OUT} is active. For the specified values inputs are toggling at 1 kHz at 50% of maximum I_{x_OUT} current.

1.3 APC Input Characteristics (register 4Ah[4]=1b)

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-3. APC Input Characteristics (photodiode cathode connected to pin APC_INX and photodiode anode connected to ground)

Parameter	Conditions	Minimum	Typical	Maximum	Units
Full scale input	APC_IN0,1,2	3.1	3.25	3.4	mA
PD capacitance ¹	APC_IN0,1,2	—	—	80	pF
Input bias	APC_IN0,1,2	—	1.8	—	V
Maximum input voltage		—	—	1.89	V

NOTES:

- Care should be taken in routing of each PD input so that total capacitance on the pin including routing does not exceed 80 pF

1.4 APC Input Characteristics (register 4Ah[4]=0b)

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-4. APC Input Characteristics (Current Sink Input)

Parameter	Conditions	Minimum	Typical	Maximum	Units
Full scale input	APC_IN0,1,2 ²	3.1	3.25	3.4	mA
PD capacitance ¹	APC_IN0,1,2 ²	—	—	300	pF
Min Input bias	APC_IN0,1,2 ²	—	0.6	—	V
Maximum input voltage		—	—	1.89	V

NOTES:

- Care should be taken in routing of each PD input so that total capacitance on the pin including routing does not exceed 300 pF
- Only a single channel can be used (broadband monitor photodetector). If multiple channel are used accuracy is not guaranteed.

1.5 CPC Target DAC

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-5. CPC Target DAC

Parameter	Conditions	Minimum	Typical	Maximum	Units
Resolution		—	13	—	bits
Conversion rate		—	30	—	Msp/s
Full scale monitor photodetector current		—	3.25	3.4	mA
Step size		—	350	515	nA
CPC Control Loop Accuracy ¹		-6	—	6	%

NOTES:

- At $T_c = 120\text{ }^\circ\text{C}$ control loop accuracy is $\pm 7\%$

1.6 Integrating Power Control

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-6. Integrating Power Control

Parameter	Conditions	Minimum	Typical	Maximum	Units
Resolution		—	10	—	bits
Count variation	Part to part	-35		+35	%
Stability ²	For targets $> 200\text{ }\mu\text{A}$ and register $0x0B=50h$	-11	—	+11	%

NOTES:

- This will correspond to a total power (LED/MPD current?) variation: monotonicity will still be guaranteed by the architecture
- Variation of integration target over supply and temperature. At $T_c = 120\text{ }^\circ\text{C}$ accuracy is $\pm 14\%$

1.7 LED Drivers

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-7. LED Drivers

Parameter	Conditions	Minimum	Typical	Maximum	Units
Headroom required ¹	At 2A full scale	—	—	0.3	V
Maximum allowable voltage headroom ²		—	—	5.5	V
Rise/fall time ³	20-80% Into 1 Ω electrical output, no snubber network	—	—	200	ns

NOTES:

- Required headroom scales with output current, maximum output current requires maximum headroom (see Section 2.3).
- To prevent damage at output pins do not exceed this voltage. Also verify power sequencing and power dissipation.
- Guaranteed by design

Table 1-8. Output Current DACs

Parameter	Conditions	Minimum	Typical	Maximum	Units
Resolution		—	10	—	bits
Conversion rate		—	12.5	—	Msp/s
Full scale <i>IOUTX</i>	Referred to the current output	—	2	—	A
<i>IOUTX</i> absolute accuracy	Referred to the current output ¹	-8	—	8	%

NOTES:

- For driver headroom > specified in Table 1-7. Measured at 1000 mA. At $T_c = 120\text{ }^\circ\text{C}$ accuracy is -8% to +10%

Table 1-9. Scale DACs

Parameter	Conditions	Minimum	Typical	Maximum	Units
Resolution		—	6	—	bits
Minimum scale value	Referred to the current output, equivalent to code 000000b	180	200	220	mA
Maximum scale value	Referred to the current output, equivalent to code 111111b	1.775	2	2.2	A
Scale step	Referred to the current output	25	28.6	31	mA

1.8 DC-DC Converters Reference Generators

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-10. DC-DC Converters Reference Generators

Parameter	Conditions	Minimum	Typical	Maximum	Units
Voltage compliance		0.5	1.2	1.3 V	V
DAC resolution		—	9	—	bits
DAC DNL		-1.3	—	1.3	LSB
DAC full scale	(regref_setup[1]=0b)	—	100	110	μA
DAC full scale	(regref_setup[1]=1b)	—	200	220	μA
DC-DC Converter Headroom Error	Configured as per Table 2-4 using a Texas Instruments TPS63020 DC-DC converter and decimation set to 64.	-20	—	+20	mV

1.9 Internal Temperature Sensor

Typical values: $T_c = -40\text{ }^\circ\text{C}$ to $100\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-11. Internal Temperature Sensor

Parameter	Conditions	Minimum	Typical	Maximum	Units
Range		—	-40 to 125	—	$^\circ\text{C}$
Temperature step		—	0.65	—	$^\circ\text{C}$
Absolute accuracy ¹		-10	—	+10	$^\circ\text{C}$

NOTES:

1. After system calibration at room temperature (one point calibration).

1.10 Light Sources Alarm

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-12. Light Sources Alarm

Parameter	Conditions	Minimum	Typical	Maximum	Units
Light sensor alarm thresholds ¹		50	—	200	mV
Threshold accuracy		—	+/-15	—	mV
Alarm response time		—	5	—	μs

NOTES:

1. Threshold can be programmed through register alarm_setup0/1 to 50 mV, 100 mV, 150 mV, 200 mV.

1.11 CMOS Pins Characteristics

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_c = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = +3.3\text{ V}$

Table 1-13. CMOS Pins Characteristics

Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{IH} ¹		$0.65 DV_{DD}$	—	3.63	V
V_{IL}		0	—	$0.35 DV_{DD}$	V
V_{OH}		$DV_{DD} - 0.4$	—	DV_{DD}	V
V_{OL}		0	—	0.4	V
Rise/fall time ²	Maximum load of 5 pF. SPI mode	—	3	—	ns

NOTES:

1. Digital pins are 3.3 V (+/-10%) tolerant
2. In I²C mode, rise/fall time depends on load and pull up resistor.

1.12 Slave I²C Timing Specifications^{1,2}

Min and Max values: T_c= -40 °C to 85 °C, DV_{DD}=1.8 V+/-5%, ALV_{DD}=1.8 V+/-5%, DHV_{DD}=3.3 V+/-5%, AHV_{DD} =3.3 V+/-5% unless otherwise noted. **Typical values:** T_c=25 °C, DV_{DD}, ALV_{DD}=1.8 V, AHV_{DD} =3.3 V

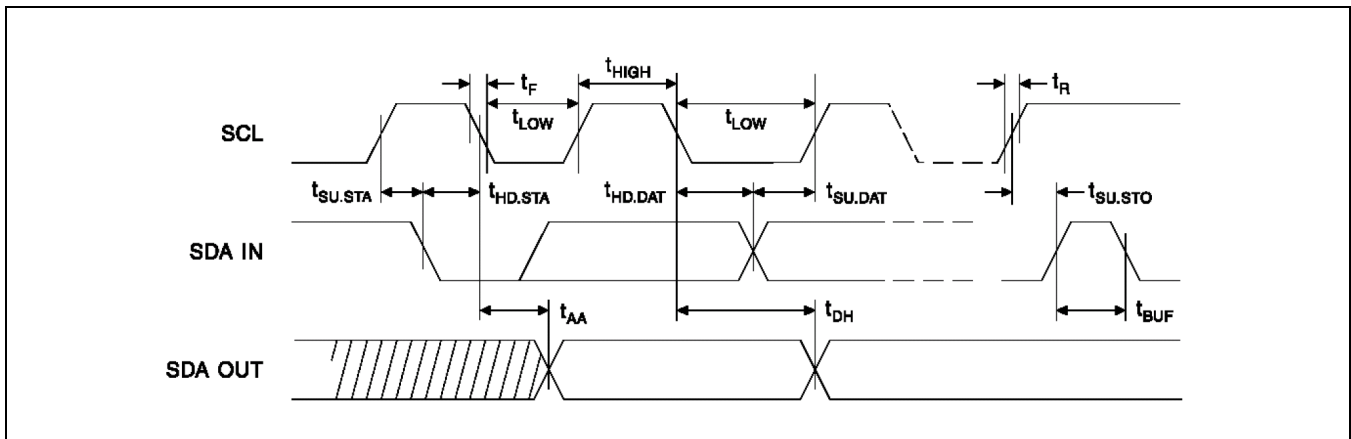
Table 1-14. Slave I²C Timing Specifications^{1,2}

Parameter	Symbol (refer to figure below)	Minimum	Typical	Maximum	Units
Clock Frequency, SCL_M	f _{SCL_MASTER}	—	—	3.4	MHz
Clock Pulse Width Low	t _{LOW}	160	—	—	ns
Clock Pulse Width High	t _{HIGH}	60	—	—	ns
Clock Low to Data Out Valid	t _{AA}	0	—	70	ns
Start Hold Time	t _{HDSTA}	160	—	—	ns
Start Set-up Time	t _{SUSTA}	160	—	—	ns
Data In Hold Time	t _{HDDAT}	0	—	—	ns
Data In Set-up Time	t _{SUDAT}	10	—	—	ns
Outputs (SDA_M, SCL_M, SDA_S and SCL_S) internal pull-up resistor value ³	R _{PULL-UP}	—	250	—	kΩ
Stop Set-up Time	t _{SUSTO}	160	—	—	ns
Data Out Hold Time	t _{DH}	5	—	—	ns

NOTES:

1. Guaranteed by design and characterization.
2. Specified at recommended operating conditions.
3. 4.7 kΩ should be added externally.

Figure 1-1. Slave I²C Timing



1.13 High Speed Serial Interface Timing Specifications

Min and Max values: $T_c = -40\text{ }^\circ\text{C}$ to $85\text{ }^\circ\text{C}$, $DV_{DD} = 1.8\text{ V} \pm 5\%$, $ALV_{DD} = 1.8\text{ V} \pm 5\%$, $DHV_{DD} = 3.3\text{ V} \pm 5\%$, $AHV_{DD} = 3.3\text{ V} \pm 5\%$ unless otherwise noted. **Typical values:** $T_a = 25\text{ }^\circ\text{C}$, DV_{DD} , $ALV_{DD} = 1.8\text{ V}$, $AHV_{DD} = 3.3\text{ V}$

Table 1-15. High speed serial interface timing specifications

Parameter	Symbol (refer to figure below)	Minimum	Typical	Maximum	Units
Clock Frequency	$f_{clk} = 1/T_{clk}$	—	—	3.4	MHz
Data in to clk hold time	T_{dh}	160	—	—	ns
Data in to clk set-up time	T_{ds}	60	—	—	ns
Enable to clk set up time	T_{cs}	0	—	70	ns
Enable to clk hold time	T_{ch}	160	—	—	ns
Read data output valid following rising edge of SCLK	T_{dd}	160	—	—	ns
SCLK duty cycle		45	—	55	%

NOTES:

- Maximum output capacitance of 30 pF.

Figure 1-2. Serial Interface Sequential Write

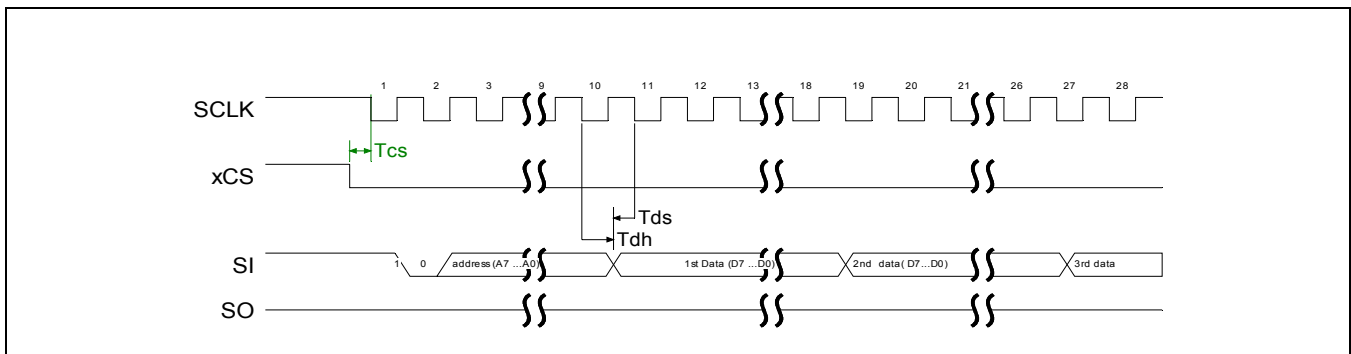


Figure 1-3. Serial Interface Random Write

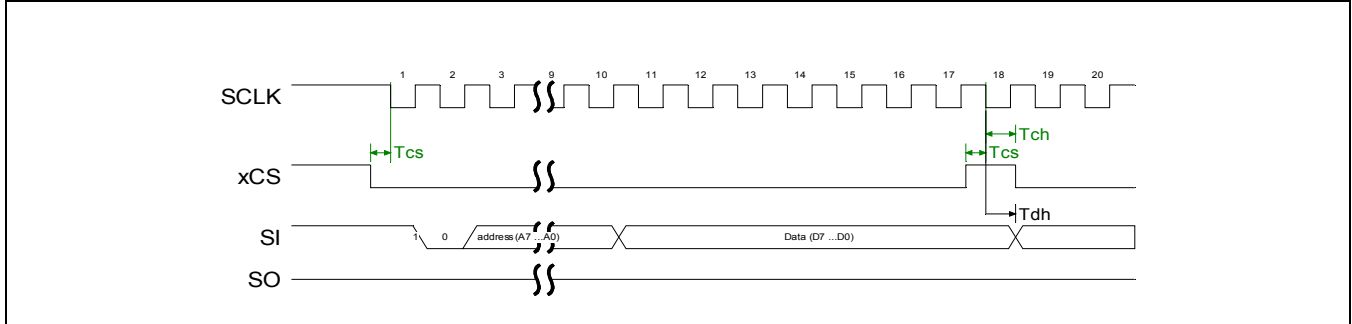


Figure 1-4. Serial Interface Sequential Read

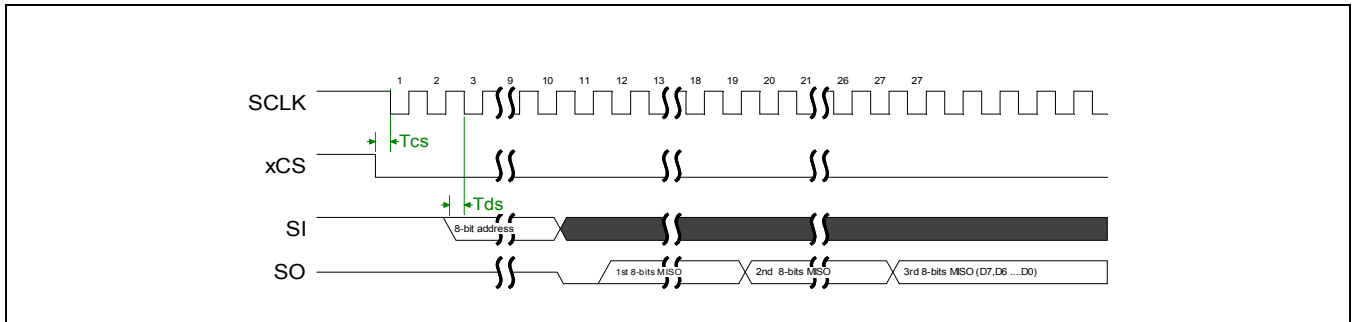
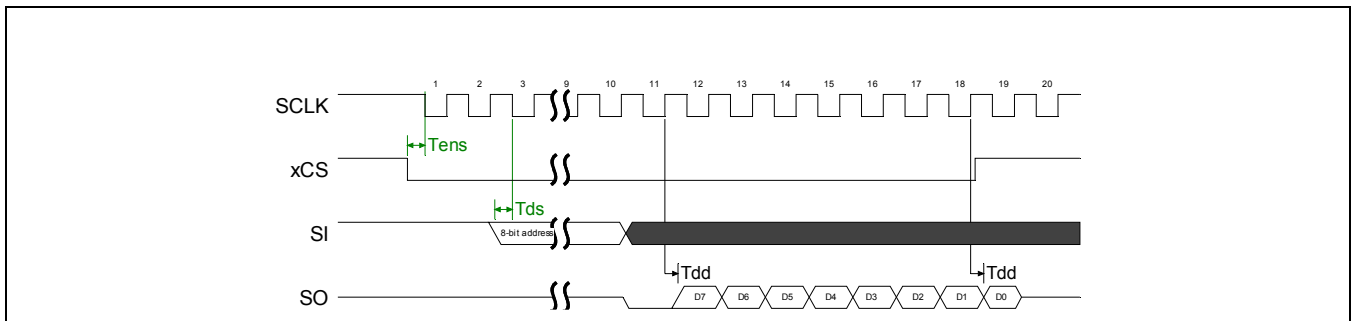


Figure 1-5. Serial Interface Random Read



1.14 M08888 Pinout

The M08888 is packaged in a 4.5 x 4.5 mm 28-pin QFN package with 0.5 mm pin pitch.

Figure 1-6. M08888 Pinout

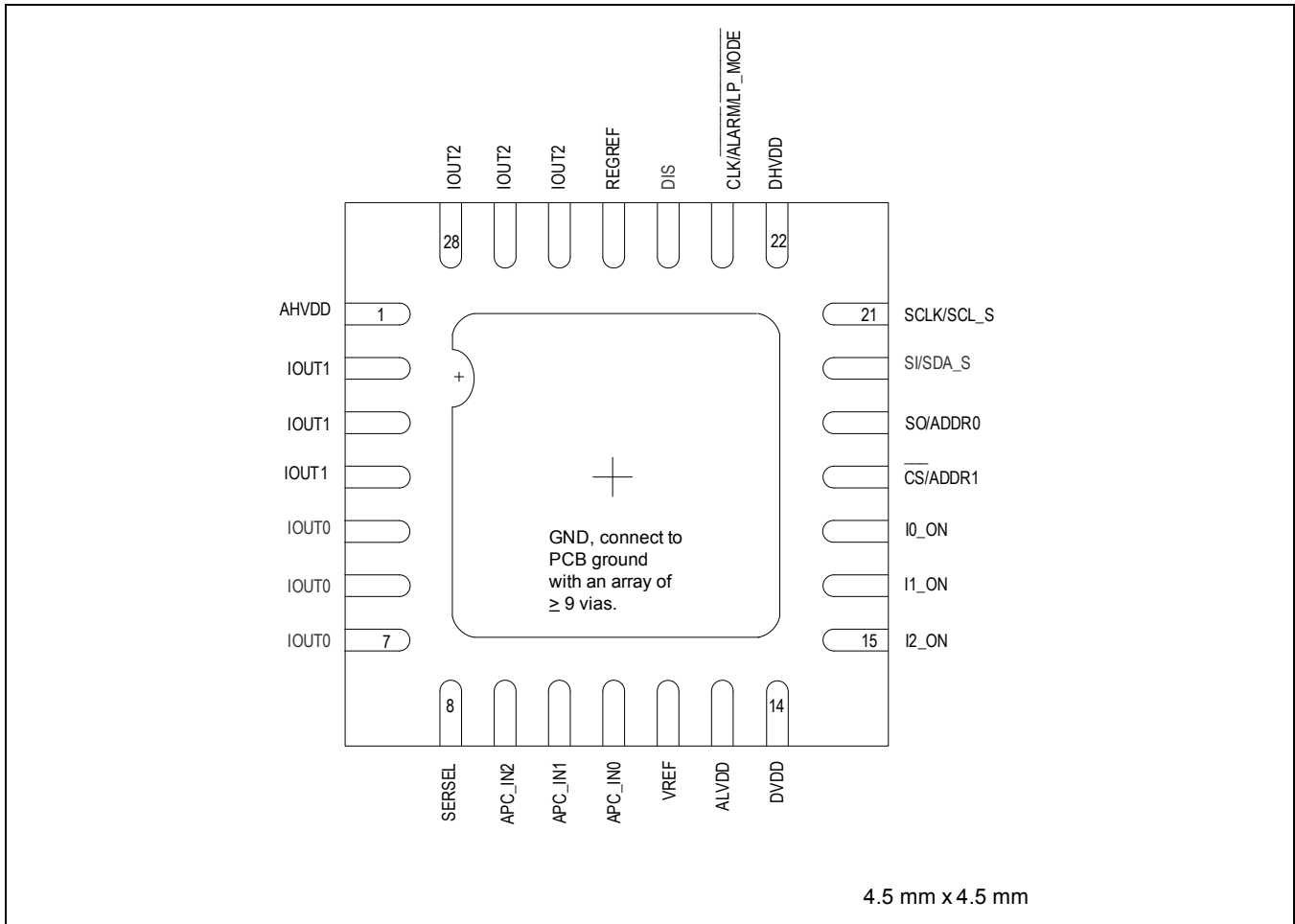


Table 1-16. Pin List and Descriptions

Pins	Name	Type	Function
1	AHVDD	Supply	3.3 V analog supply
2, 3, 4	IOUT1	Analog	Channel 1 output
5, 6, 7	IOUT0	Analog	Channel 0 output
8	SerSEL	CMOS (w/ pulldown)	Serial interface select (L=I ² C,H=SPI)
9, 10, 11	APC_IN	Analog	Monitor PD inputs (assignable)
12	VREF	Analog	Current reference generator
13	ALVDD	Supply	1.8 V analog supply
14	DVDD	Supply	1.8 V digital supply
15	I2_ON	CMOS (w/ pulldown)	Turns on driver 2
16	I1_ON	CMOS (w/ pulldown)	Turns on driver 1
17	I0_ON	CMOS (w/ pulldown)	Turns on driver 0
18	$\overline{CS}/ADDR0$	Open Drain (w/ pull-up)	Serial enable/I ² C address0
19	SO/ADDR1	Open Drain (w/ pull-up)	Serial data out/I ² C address1
20	SI/SDA_S	Open Drain (w/ pull-up)	Serial data in/I ² C data slave
21	SCLK/SCL_S	Open Drain (w/ pull-up)	Serial clock/I ² C clock slave
22	DHVDD	Supply	3.3 V digital supply
23	$\overline{CLK_IN}/\overline{ALARM}/\overline{LP_MODE}$	CMOS (w/ pulldown)	CLK_IN pin or ALARM or LP_MODE pin
24	DIS	CMOS	Disable Pin
25	REGREF	Analog	DC-DC converter control voltage
26, 27, 28	IOUT2	Analog	Channel 2 output
	GND	Supply	Ground

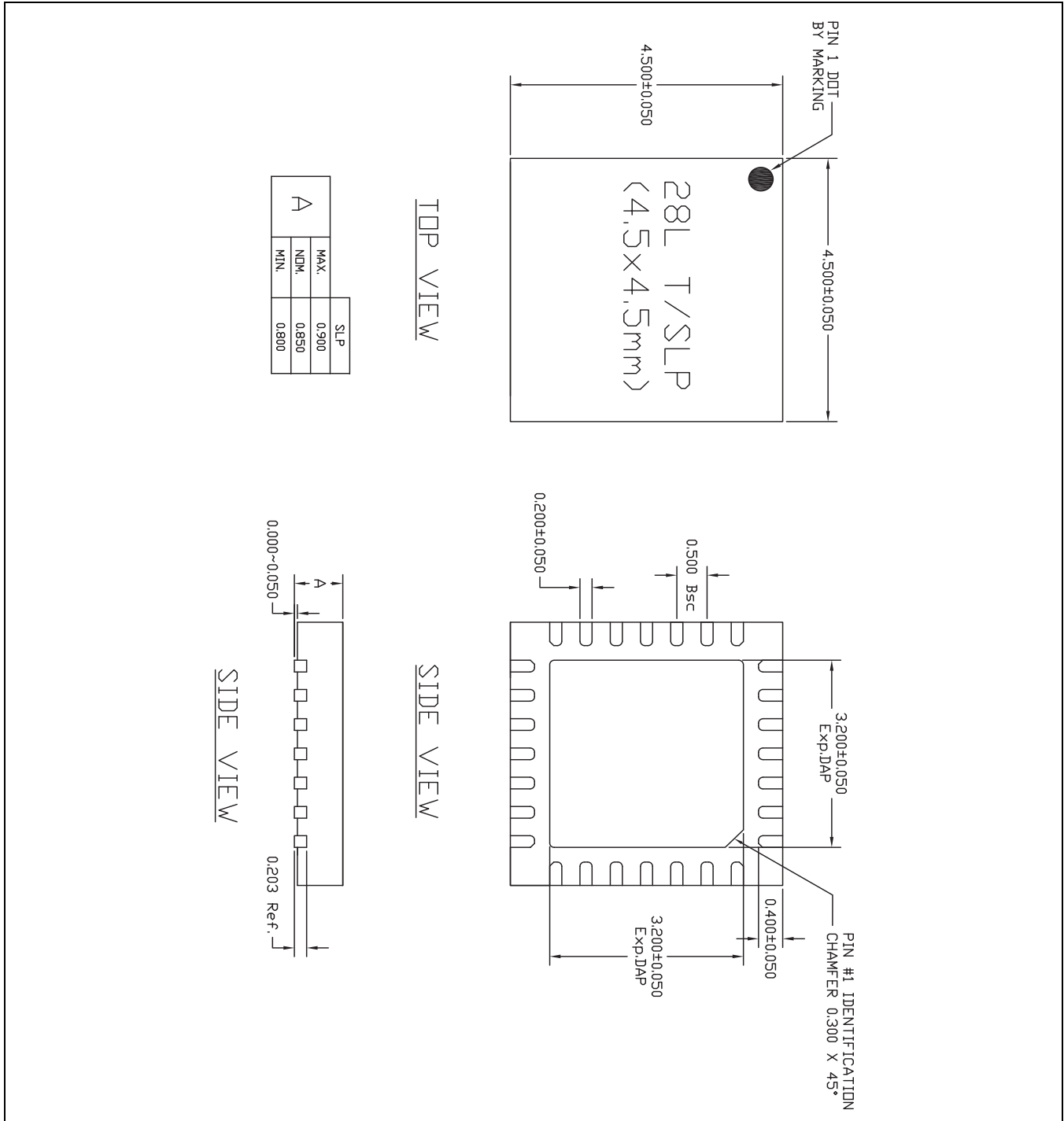
NOTES:

PD means pulled down, PU means pulled up.

4.7 μ F + 100nF should be used on each of the M08888 supply.

1.15 Package Information

Figure 1-7. Package Information



2.0 Functional Description

The M08888 is a highly integrated LED/laser driver for LCD/LCoS/DLP projection display applications. It provides control and monitoring of up to three LEDs/lasers, a temperature sensor and control of external DC-DC converters for optimal laser/led supply voltage.

Each Laser/LED output consists of a 10 bit DAC which controls a high efficiency driver. If the desired maximum current is less than 2A the output resolution can be improved by a 6 bit independent scaling DAC.

The output stages require only 300 mV of headroom between the M08888 output and the Laser/LED cathode when driving 2 amperes. The headroom requirement can be scaled proportionally lower for lower currents.

The M08888 also incorporates safety and alarm features and a temperature monitor with 8 bit resolution.

The M08888 internal registers are loaded from an external micro controller through a slave I²C interface or a 4-wire high speed interface. The host micro controller can monitor the temperature sensor and read back the analog to digital converter outputs and status registers using either serial interface.

2.1 Operating Modes

The M08888 can operate in 3 different optical power control modes. The power control modes are open loop (OL) and 2 automatic power control (APC) modes: continuous power control (CPC) and integrating power control (IPC).

Different output channels can have different operating modes. Some channels may be configured as Open Loop and some channels may be configured to use APC (either CPC or IPC).

Open Loop mode is the simplest mode of operation and the Laser/LED current is set by writing the desired current to the output DAC.

The 2 automatic power control modes (APC) use monitor photodiode feedback to accurately adjust the Laser/LED output power to make the current from a photodiode match a target current. With APC control temperature compensation of the Laser/LED is automatic and color balance is simplified.

The user can select the polarity of the monitor photodetector if an APC mode is selected. By default the M08888 accepts a current source monitor photodetector (MPD) tied to the positive supply. It is possible to accept current sink MPD by selecting `ipc_setup[4]=1b`.

Configuration of the M08888 timer settings and pin 23 configuration will also affect optical power control.

2.1.1 Open Loop Output Control

In open loop mode the current for each Laser/LED is stored in register (`ioutx[9:0]`). The M08888 will shift the contents of `ioutx[9:0]` and `ioutx_scale[5:0]` registers to the output DAC when pin `Ix_ON` goes high and the output will then sink the programmed current through the Laser/LED.

By changing the scale setting `ioutx_scale` the full 10 bits of settability can be retained even for low Laser/LED currents. The scale can be changed on a frame by frame basis but when `ioutx_scale` is changed the rise time of the output stage will be significantly slower during the first `Ix_ON` period in which the change in the scale is made. The rise time will return to its normal value in subsequent `Ix_ON` periods if the scale is not changed.

The M08888 includes timers that can be used to delay, pulse width modulate or clock the Open Loop output signal. See [Section 2.4](#) for a description of how to use these features.

When pin 23 operates in `LP_MODE` (`input_ctrl[6:5]=10b`) the scale value of each channel can be different for the different pin 23 states. These scales are set in `ioutx_scale[5:0]` (pin 23=H) and `ioutx_scale_LP[5:0]` (pin 23=L).

Table 2-1. Basic Register Configuration for Open Loop Control (when REGREF is not controlling a DC-DC Converter)

Name	Address	Recommend Setting	Description
<code>out_ctrlx</code>	<code>0x05[7],0x06[7],0x07[7]</code>	1b	Set outputs to Open Loop Control if Regref is not used to control a DC-DC converter at this output.
<code>pd_fe</code>	<code>0x0A[0]</code>	1b	Power down photodiode amplifier (if all channels are under open loop control)
<code>ioutx_msb</code>	<code>0x10[1:0], 0x14[1:0], 0x18[1:0]</code>	xxb	Two most significant bits of output current setting
<code>ioutx_lsb</code>	<code>0x11, 0x15,0x19</code>	xxh	Eight least significant bits of output current setting
<code>ioutx_scale</code>	<code>0x12, 0x16, 0x1A</code>	11xxxxxb	Set the scaling of the output currents
<code>ioutx_scale_lp</code>	<code>0x13, 0x17, 0x1B</code>	11xxxxxb	Set the scaling of the output currents in Low Power mode
<code>apc0_ch</code>	<code>0x34[1:0]</code>	11b	Disable APC input for selected channel
<code>apc1_ch</code>	<code>0x34[3:2]</code>		
<code>apc2_ch</code>	<code>0x34[5:4]</code>		

2.1.2 Automatic Power Control

Automatic power control (APC) can keep the laser/led power constant and the color balanced by comparing the monitor photodetector (RGB color sensor) currents to target values programmed into the 13-bit target DACs.

At power-up the APC can be enabled/disabled independently for each channel by setting `apc_ctrl0[5:0]=11`. For channels with disabled APC the laser/led currents are controlled through the serial interface using bits `ioutx[9:0]`.

It is possible to freeze the APC loop for each channel by using `apcx_freeze[0]`. In this case the M08888 will stop updating the `IOUTx` currents independently of the state of `Ix_ON` pins. The photodiode (RGB sensor) target values are programmed via the serial interface in registers `target2[12:0]`, `target1[12:0]` and `target0[12:0]` respectively for `IOUT2`, `IOUT1` and `IOUT0`. When the corresponding color is turned on as signaled by the transition of `I0_ON`, `I1_ON` or `I2_ON` from low to high the LED/Laser drive currents are automatically adjusted up or down to always make the photodetector current (RGB sensor current) match the target current.

If desired, these target currents can be adjusted on a frame by frame basis to optimize contrast and save battery power depending on the brightness required for a particular frame.

The target values are stored in registers target2[12:0], target1[12:0] and target0[12:0] respectively for IOUT2, IOUT1 and IOUT0 before the corresponding color is turned on by the transition of I2_ON, I1_ON or I0_ON from low to high. This allows for the control of the LED/laser power in real time on a frame by frame basis.

When one of the I_x_ON signals is enabled (low to high transition), the target DAC value for that input is activated and the incoming monitor photodetector current is selected by the analog multiplexer from APC_IN0, APC_IN1 or APC_IN2. The monitor photodetector input is assigned to an output using apc_ctrl[5:0].

Each Laser/LED can have its own photodetector or all the Lasers/LEDs can share a single broadband photodetector. For example, if a single broadband photodetector is used and connected to APC_IN0 all the apc_ctrl[5:0] should be set to 00b. It should be noted that the index in the APC registers does not refer to the input channel but to the Laser/LED output channel. If the monitor photodetector feedback for IOUT2 is tied to APC_IN0 then apc_ctrl[5:4]=00 and the settings for the monitor photodetector should be programmed in the APC registers with index 2.

If more than one laser is turned on (2 or 3 of I0_ON, I1_ON or I2_ON are high at the same time) the M08888 freezes the update of the APC loop for as long as more than one of the I_x_ON signals are high. This prevents the APC loop from using incorrect photodetector information in case a single photodetector is used. In the case of multiple I_x_ON simultaneously the optical power tracked and adjusted is that of the channel corresponding to the last I_x_ON transitioning high. It should be noted that 2 or more low to high transitions of any of the I0_ON, I1_ON or I2_ON within 500 ns of each other would violate internal timing and will result in unpredictable operation. Care should be taken to prevent damage to the part when multiple lasers are enabled. The power dissipation of the M08888 should be kept below the level that, when multiplied by the thermal resistance of the package in the system and added to the maximum ambient temperature, does not exceed 125 °C.

The maximum photodiode current supported by the M08888 is 3.25 mA. The full scale value of the target DAC is 3.25 mA with a resolution of 13 bits (390 nA step size). As an example, if the maximum current from the photodiode is 200 µA then the possible target values are up to 512 decimal (200 µA/390 nA). In this example the maximum target value for targetx[12:0] would be 200h.

Readback of the output current DAC of each channel is possible by strobing the DACs using strbalarm_ctrl[1] and reading registers rb_ioutx[9:0]

When pin 23 is set to $\overline{LP_MODE}$ (input_ctrl[6:5]=10b) the target power value will be controlled by the state of pin 23. When pin 23 is high the Laser/LED current will be adjusted until the monitor photodiode current matches the target in register targetx[12:0] and when pin 23 is low the Laser/LED current will be adjusted until the monitor photodiode current matches the target in register targetx_LP[12:0].

The M08888 is capable of accepting also current sink type monitor photodetector typical of LCOS panels. This can be done by setting register ipc_setup[4]=1b. When sink MPD is selected the mirror ratio can be selected between 1:1 and 4:1 by setting bit ipc_setup[5].

The MPD can have a maximum capacitance of up to 300 pF however in this case the M08888 will operate only with a single broadband photodetector.

2.1.2.1 Continuous Power Control

In continuous power control (CPC) mode, the M08888 continuously compares the monitor photodetector current to a target value and makes monitor photodiode current match the target value by adjusting the current in the laser/LED. For example, if the monitor photodetector current is below the target then the Laser/LED current is increased. The sign of this operation can be inverted using input_ctrl[4] (but it should not be changed unless it is certain that there is an inversion in the monitor photodiode signal).

The CPC loop is designed to settle to the desired output power in less than 50 μ s. To achieve optimal settling time, the CPC loop must be adjusted to the laser/monitor photodetector characteristics. This is done using the settings in `apcx_ctrl0`, `apcx_ctrl1` and `apcx_ctrl2`.

To further reduce the settling time the user can program the initial current from which the CPC loop will start. This can be 0, the value programmed in `ioutx[9:0]` register or the value to which the CPC had converged during the previous frame. The selection of the initial current is done using register `iturnonx[1:0]` (`apcx_ctrl1[3:2]`). Starting from the previously determined value will substantially accelerate settling time since it is likely that it will be starting at the proper level.

It is possible to “freeze” the APC loop for each channel by using `apcx_ctrl0[0]`. When `apcx_ctrl0[0]=1b` the M08888 will stop updating the `IOUTx` currents irregardless of the state of `Ix_ON` pins. It is possible also to delay operation the CPC loop immediately following the light source turn-on. This will allow the DC-DC converter to settle to the proper voltage before the APC starts adjusting the current. The CPC delay time can be programmed using register `apcx_ctrl0[2:1]`.

Table 2-2. Basic Register Configuration for CPC Control

Name	Address	Recommend Setting	Description
<code>out_ctrlx</code>	0x05[7], 0x06[7], 0x07[7]	1b	Set output to Closed Loop Control
<code>loop select</code>	0x34[6]	1b	Select CPC loop control
<code>apc0_ch</code>	0x34[1:0]	xxb	Select which photodiode (RGB sensor) input will control which output. Setting depends on hardware connection to photodiode(s) and which channels will use APC control. See Register description for details.
<code>apc1_ch</code>	0x34[3:2]		
<code>apc2_ch</code>	0x34[5:4]		
<code>targetx_msb</code>	0x35, 0x37, 0x39	000x xxxxb	Set the target at desired level. The LED drive current will be adjusted up or down until the photodiode current is equal to the target current for the channels that are controlled by CPC.
<code>targetx_lsb</code>	0x36, 0x38, 0x3A	xxh	
<code>apc0_ctrl2</code>	0x47	xxh	Set the length in clock counts of the initial, mid and min step intervals. Set the delay for the clock divider. The decimation factor and clock divider will also affect interval lengths.
<code>apc1_ctrl2</code>	0x44		
<code>apc2_ctrl2</code>	0x41		
<code>apc0_ctrl1</code>	0x48	xxh	Set the step size of the change in LED drive current for the initial and mid intervals. Set the initial LED current to be 0 mA, the ending value of the previous <code>I_on</code> period or the value set in <code>ioutx</code> register. Set the clock divider.
<code>apc1_ctrl1</code>	0x45		
<code>apc2_ctrl1</code>	0x42		
<code>apc0_ctrl0</code>	0x49	xxx00xx0b	Set the decimation factor (mid and min count changes will occur at rate divided by the decimation factor). Set the WAIT states (APC will not change LED current during WAIT states).
<code>apc1_ctrl0</code>	0x46		
<code>apc2_ctrl0</code>	0x43		

2.1.2.2 Integral Power Control

Integral power control (IPC) can be enabled by setting registers `apc_ctrl[6]=0b`.

The working principle of integral power control is to mimic the behavior of the human eye which integrates the optical power over the frame period. This is achieved in the M08888 by charging a capacitor with the monitor photodetector current.

When I_{x_ON} goes high the M08888 will drive the Laser/LED with a current defined by register $ioutx[9:0]$. The monitor photodiode current will charge an internal capacitor until its voltage matches an internal reference voltage (~ 0.8 V) at which time a counter is incremented and the capacitor is discharged. The monitor photodiode current continues to charge the capacitor and increment the counter until the count matches the target set by the user in registers $targetx[9:0]$. Once the counter reaches the value set by the user the M08888 stops driving the Laser/LED. By adjusting the count higher or lower the brightness of a frame can be adjusted. The target counter has 10bits and the M08888 can distinguish between 1024 different power levels. The target registers are the same of the CPC target registers but the resolution in IPC mode is limited to 10 bits instead of 13 bits.

In order to maximize the dynamic range and be able to utilize the entire 10bits the user must select the proper mirroring ratios for the monitor photodetector current ($ipdx_sel[4:0]$), the charging current ($ichx_sel[4:0]$) and the value of the capacitance ($capx_sel[3:0]$). The settings of these bits will depend on the frame (or subframe) duration and on the monitor photodetector current.

If it is desired to typically have the LED on for 1/2 the frame time then the registers should be configured such that:

$$1/2 \times FR/512 = 0.8 \text{ V} \times Cch / Ich$$

where FR is the frame or subframe rate, Cch is the charging capacitors selected through $capx_sel[3:0]$, and Ich is the resulting charging current obtained by multiplying the monitor photodetector current by the $ipdx_sel[4:0]$ and by the $ichx_sel[4:0]$ mirroring ratios. The factor of 512 is length of the 10 bit range.

$$Cch = 25 \text{ pF} + N \times 2.5 \text{ pF} \text{ where } N \text{ is the setting of } capx_sel[3:0].$$

$$Ich = (I_{MPD} / ipdx_sel[4:0]) \times ichx_sel[4:0] \text{ where } I_{MPD} \text{ is the monitor photodiode current}$$

Example:

Assume that the monitor photodiode current is 150 μ A when the LED is on at the desired amplitude, the I_{x_ON} time is 1 ms and it is desired that the LED be on approximately 50% of this time. The mirroring ratio should be set to the 200 μ A range $ichx_sel[4:0] = 00001b$, choose a Cch = 25 pF with $ichx_sel[4:0] = 00000b$ (any other value is also OK but using the default value means this register never needs to be written), choose a mirroring ratio such that the Cch can be charged to 0.8 V several hundred times in 0.5 ms. (if it is desired that it be charged ~ 250 times in 0.5 ms then:

$$0.5 \text{ ms}/250 = 0.8 \text{ V} \times 25 \text{ pF} / ((150 \mu\text{A} / 200 \mu\text{A}) \times ichx_sel[4:0])$$

$$0.002 \text{ ms} = 26.67e-12 / ichx_sel[4:0]$$

$$ichx_sel[4:0] = 2.667e-5 / 2 \text{ which is between } 10 \mu\text{A} \text{ and } 20 \mu\text{A} \text{ so set } ichx_sel[4:0] = 00001b$$

As in CPC mode, when pin23 operates in $\overline{LP_MODE}$ ($input_ctrl[6:5]=10b$) the target power value can be changed by toggling pin23. When pin 23 is high the Laser/LED current will be on until the monitor photodiode count matches the target in register $targetx[9:0]$ and when pin 23 is low the Laser/LED current will be on until the monitor photodiode count matches the $targetx_LP[9:0]$.

The monitor photodiode input can be changed from sinking to sourcing at $ipc_setup[4]$ and an additional scaling factor of 4:1 is available at $ipc_setup[5]$.

Table 2-3. Basic Register Configuration for IPC Control

Name	Address	Recommend Setting	Description
ipc_setup[4]	0x4Ah[4]	0b or 1b	M08888 sinks MPD current (1b) or sources MPD current (0b)
out_ctrlx	0x05[7],0x06[7], 0x07[7]	1b	Set output to Closed Loop Control
loop select	0x34[6]	0b	Select IPC loop control
apc0_ch	0x34[1:0]	xxb	Select which photodiode (RGB sensor) input will control which output. Setting depends on hardware connection to photodiode(s) and which channels will use APC control. See Register description for details.
apc1_ch	0x34[3:2]		
apc2_ch	0x34[5:4]		
targetx_msb	0x35, 0x37, 0x39	000xxxxb	Set the target to the desired count. The LED drive current set in the ioutx register will be on until the current from the ichx current mirror charges the capacitor set at ipcx_ctrl2[3:0] to 0.8 V the number of times set in this target register (the capacitor at pcx_ctrl2[3:0] is discharged every time the 0.8 V comparator is tripped and recharging begins again until the target count is reached).
targetx_lsb	0x36, 0x38, 0x3A	xxh	
ioutx_msb	0x10[1:0], 0x14[1:0], 0x18[1:0]	xxb	Two most significant bits of output current setting
ioutx_lsb	0x11, 0x15, 0x19	xxh	Eight least significant bits of output current setting
ioutx_scale	0x12, 0x16, 0x1A	11xxxxb	Set the scaling of the output currents
ipdx_sel	0x4C[4:0] 0x4F[4:0] 0x52[4:0]	x xxxxb	Select a setting that is greater than the peak current expected from the monitor photodiode. Choosing a lower amplitude setting will give more resolution/accuracy in setting the IPC target.
ichx_sel	0x4D[4:0] 0x50[4:0] 0x53[4:0]	x xxxxb	Select a current range that can charge the IPC capacitor several hundred (but <511) times during the Ix_ON time.

2.2 Outputs

2.2.1 LASER/LED Current DACs

The M08888 includes three monotonic DACs which generate the currents for the three LED output drivers.

The three DACs have a maximum range from 0 to 2A, a resolution of 10 bits and a maximum update rate of 12.5 MSPS. The output rise time will be limited by the M08888 output current driver unless a single DC-DC converter is used for more than one Laser/LED, in which case the DC-DC converter settling time and overall supply loop behavior may determine the output rise time.

The full scale of each output DAC can be programmed through the ioutx_scale[5:0] bits independently for each channel. The scale DAC changes the full scale of each output current DAC from a minimum of 200 mA to a maximum of 2A in steps of 28.57 mA. At power-up the scale DACs are set at maximum scale (2A).

2.2.2 Output Current Drivers

The integrated output current drivers deliver the DAC currents to the Lasers/LEDs.

Each Laser/LED driver output ($IOUTx$) is controlled by the corresponding ON signal (Ix_ON) and the PWM and MPG setting as described later. Rise/fall time of the driver is typically 200 ns into a resistive electrical load connected to a stable supply voltage. If a DC-DC converter is employed the response time of the current output may also depend on the response time of the DC-DC converter and the series resistance of the light sources. The rise and fall time is specified for any transition of the $ioutx[9:0]$ registers for a constant $ioutx_scale[5:0]$ code.

The drivers require a worst case headroom of 300 mV. The headroom is proportionally lower at lower drive currents.

The M08888 typical driver headroom follows the following equation:

$$VLDD = 150 \text{ m}\Omega \times IOUT$$

The voltage at the laser driver output should never exceed 5.5 V. An external resistor should be used between the laser/LED cathode and ground to provide a small leakage current into the light source allowing the voltage at the $IOUTx$ to be reduced from the anode voltage by the laser/LED voltage drop. The value of the resistor should be chosen such that the current flowing is enough to create a voltage drop on the laser while keeping the laser current far below threshold or, in the case of LEDs, low enough so as to not cause light pollution in the system.

2.2.3 Recommended Snubbing Network at $IOUTx$ Pins

A snubbing network of 1 μF in series with 1.5 Ω should be placed in parallel with the LED/Laser at each $IOUTx$ pin. The inductance in series with the LED/Laser should be less than 500 nH.

2.3 Controlling the Output Voltage to Optimize Power Consumption

The system power dissipation will be dominated by the Laser/LED current and the bias voltage of the Laser/LED. For each ampere of Laser/LED current, each 100 mV of excess Laser/LED bias voltage results in 100 mW of wasted power.

The M08888 minimizes the voltage drop on the output stage and optimizes overall power dissipation by adjusting the anode voltage of the light sources through the external DC-DC converter. Given the expected maximum current for a particular LED/laser on a channel, the user can program the headroom required for each channel based on the equations above (Section 2.2.2) to optimize system power. Whether or not the M08888 controls the DC-DC converter output voltage the required headroom of the M08888 outputs must be maintained when the corresponding Ix_ON is high (see Section 2.2.2).

2.3.1 Control of External DC-DC Converters with the M08888

An external DC-DC converter is controlled by the M08888 by connecting pin $REGREF$ to the DC-DC converter feedback pin.

Typical register settings to allow control of an external DC-DC converter are shown below.

Table 2-4. Basic Register Configuration for Controlling an External DC-DC Control

Name	Address	Recommended Setting	Description
selfcal_hr	0x00[4]	1b	Power-up REGREF circuitry
opmode_ctrl1	0x01[7]	1b	Set DC-DC output to minimum when all I_x_ON are low
out_ctrlx	0x05, 0x06, 0x07	8Ch	Recommended Configuration
regref_setup	0x24	00h	Set REGREF operating mode
regref2_ctrl1	0x27	FFh	Set the headroom at $IOUT2$ to 310 mV and the decimation factor to 2048. Lower headroom values can be used depending on the maximum $IOUT2$ output current. See section 2.2.2
regref2_ctrl0	0x28	44h	Set the DAC update rate and the starting value at $IOUT2$.
regref1_ctrl1	0x2B	FFh	Set the headroom at $IOUT1$ to 310 mV and the decimation factor to 2048. Lower headroom values can be used depending on the maximum $IOUT1$ output current. See section 2.2.2
regref1_ctrl0	0x2C	44h	Set the DAC update rate and the starting value at $IOUT1$.
regref0_ctrl1	0x2F	FFh	Set the headroom at $IOUT0$ to 310 mV and the decimation factor to 2048. Lower headroom values can be used depending on the maximum $IOUT0$ output current. See section 2.2.2
regref0_ctrl0	0x30	44h	Set the DAC update rate and the starting value at $IOUT0$.

The M08888 DC-DC converter control circuitry uses a 9 bit DAC to set a feedback factor for the external DC-DC converter and adjust the anode voltage of the Laser/LEDs. The DAC is controlled by a digital filter with programmable update rate and decimation factor. The digital filter is fed by a comparator which increments or decrements the counter code depending on whether the headroom of the driver is higher or lower than the programmed headroom. The sign of this operation can be inverted through `input_ctrl[3]` (this bit should not be set unless it is known that the DC-DC converter feedback path is inverted from the usual polarity). The above operations are performed automatically by the M08888 and no interaction with the M08888 is required beyond initializing the register settings as described at the beginning of this section.

Once the LED is turned off the value of the DAC inputs are stored by the M08888. The next time this LED is active, the loop will automatically start from the stored DAC value. The initial value of the DAC code can be selected through `regrefx_ctrl0[3:2]` to be either the previously determined value, 0 or the value written in register `regrefx_dac`. The DAC codes for each channel can be read back by using the strobe bit (`strbalarm_ctrl[1]`) and registers `rb_regrefdacx[8:0]`.

The headroom, decimation filter and update rate can be programmed independently for each channel using `regrefx_ctrl1[7:3]`, `regrefx_ctrl1[2:0]` and `regrefx_ctrl0[7:5]` respectively.

If more than one output is being turned on at the same time, the `regref` will control the DC-DC converter for the headroom of the last output turned on.

If CPC is used, the user should program the headroom for the highest expected output current.

If CPC control is used there may be interactions between the settling of the DC-DC converter and the digital control method. If the settling of the DC-DC converter is such that the initial headroom is less than what is required by the Laser/LED then this may cause the CPC loop to rail (adjust the Laser/LED current to 2A) and potentially damage

the Laser/LED when the voltage rises to a high enough voltage to allow unrestricted forward current conduction. The interaction of the DC-DC converter digital control method with the CPC should therefore be carefully evaluated to avoid damage to the LEDs. One way to minimize interaction between the voltage supply and the CPC control is to let the voltage supply settle before enabling CPC control of the Laser/LED current. The update of the output current in CPC mode can also be delayed using register `apcx_ctrl0[2:1]` to allow proper settling of the DC-DC converter before the operation of control loop starts.

The initial update of the IDAC which controls the LED/Laser supply headroom can be delayed through registers `regrefx_ctrl0[1:0]` to allow the DC-DC converter to settle.

In the case of integrating power control mode the speed of the DC-DC converter settling is not important: the light source current will be equal to the programmed current if the headroom is higher or equal to the required and will be smaller if the headroom is less than the required headroom however this will not matter as long as the frame/subframe time is long enough to guarantee that the integrated power over the time meets the target.

The monotonic DAC used for controlling external DC-DC converters has a full scale current of 100 μA and 9bits of resolution. The full scale can be increase by a factor of 2 to 200 μA by setting `regref_setup[1]=1b`.

If the `Ix_ON` signals do not overlap and there is a significant amount of time ($>10\ \mu\text{s}$) when `Ix_ON=L` (blanking period), `opmode_ctrl1[7]` should be set to 1. In this case during the blanking period the DAC current is set to 0 and all the `IOUTx` signals are disconnected so that an external resistor divider network can be used to set the light sources anode voltage. The impedance level at `REGREF` pin, defined by $R1//R2//\text{input impedance of the DC-DC converter}$, should always be greater than 500 $\text{k}\Omega$.

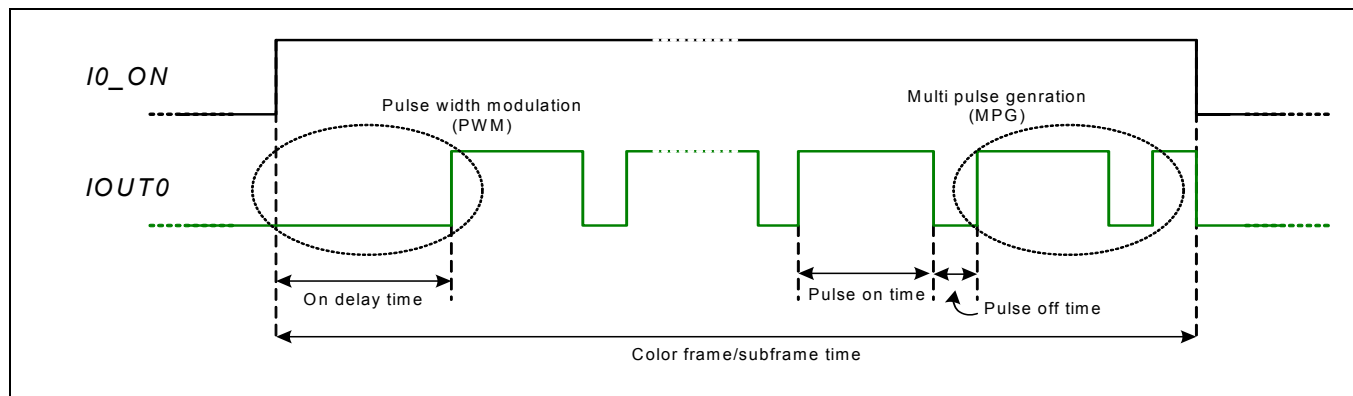
2.4 Timers

The M08888 features internal timers which allow an extra layer of control of the current by means of pulse width modulation (PWM) and multi pulse generation (MPG).

The clock source for the internal timer circuitry can be either the internal 25 MHz oscillator or an external clock fed through `CLK_IN` (pin 23).

With reference to the following diagram for channel 0 (`I0_ON`, `IOUT0`), PWM controls the “On delay time” while MPG adjusts independently both the “Pulse on time” and “Pulse off time”.

Figure 2-1. Example of PWM and MPG Timers



2.4.1 Pulse Width Modulation (PWM)

Activating pulse width modulation will allow the output current to be delayed with respect to the I_x_ON signal. This may be useful if a blanking period is needed to allow the LCD or DLP to settle or it may be useful to save power if a laser is used and the same optical output can be achieved with fewer coulombs if higher laser driver currents and shorter durations are used (the coulombs will be less with higher current and shorter duration if the laser threshold current is approximately the same with PWM activated at higher current as it is at lower current without PWM activated).

PWM works in all optical power control modes: Open Loop, CPC and IPC.

The PWM delay is implemented with a 10 bit counter that counts the clock cycles of an internal 25 MHz oscillator. At the rising edge of each I_x_ON the counter is decremented and the Laser/LED is turned on when the counter reaches 0. The laser is turned off as usual on the falling edge of the corresponding I_x_ON signal. The maximum delay achievable is $\sim 41 \mu\text{s}$ (1023 times the 40 ns period of the internal oscillator). Additional delay can be achieved by using the M08888 programmable divider of the internal clock. The internal clock can be divided down by a factor of 1, 2, 4, 8, 16, 32, 64, and 128. This can be obtained by writing register `clk_div_pwm[3:0]`. The maximum delay is therefore equal to $\sim 5.24\text{msec}$.

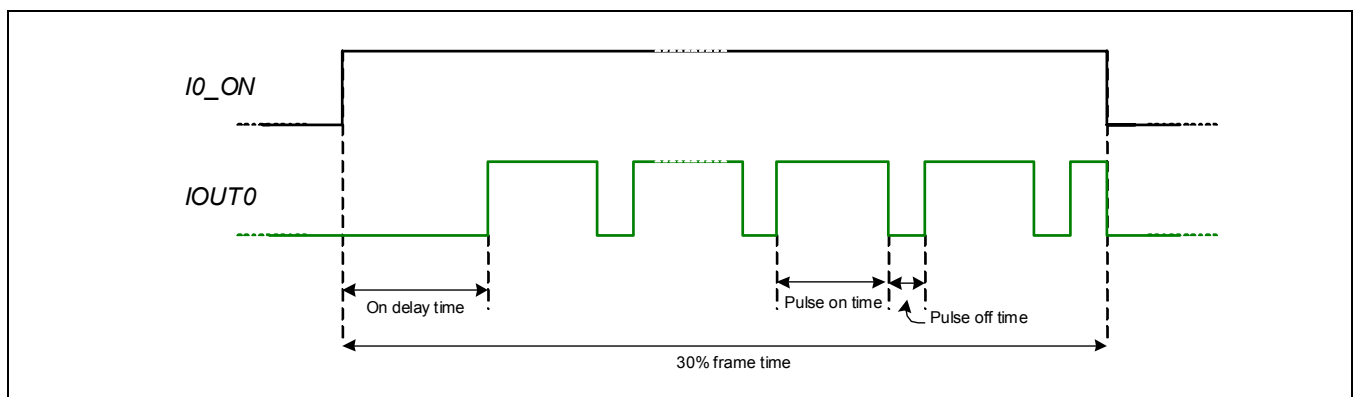
The values of the programmable counter are stored in register `on_countx[9:0]`. The rising edge of the I_x_ON signal strobes the corresponding `on_countx[9:0]` value into the M08888 in the same fashion as other output current settings such as registers `targetx[12:0]` and `ioutx[9:0]` registers. If the `on_count` register value is changed during the on time for that color the effect of the register change will be available during the next I_x_ON cycle.

If the `on_countx` is programmed to 000h the PWM feature is disabled for $IOUTx$.

The internal ring oscillator clock will vary by as much as $\pm 15\%$ over process, temperature and supply. If this accuracy is not acceptable then pin 23 can be defined to be a clock input and a more accurate external clock signal can be used. The PWM block is designed to operate with a maximum frequency of 25 MHz. The PWM generator will work at the speed of the signal at CLK_IN (pin 23) when the external clock is selected with register `clk_ctrl[2:1]`.

2.4.2 Multi Pulse Generator (MPG)

Figure 2-2. Multi Pulse Generator Timing



A less than 100% duty cycle pulsed waveform can be programmed into the M08888 using the Multi-Pulse Generator (MPG).