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19-0526; Rev 0; 5/06



# **10-Bit, 7.5Msps, Full-Duplex** Analog Front-End

### **General Description**

The MAX19710 is an ultra-low-power, highly integrated mixed-signal analog front-end (AFE) ideal for wideband communication applications operating in full-duplex (FD) mode. Optimized for high dynamic performance and ultra-low power, the device integrates a dual 10-bit, 7.5Msps receive (Rx) ADC; dual 10-bit, 7.5Msps transmit (Tx) DAC; three fast-settling 12-bit aux-DAC channels for ancillary RF front-end control; and a 10-bit, 333ksps housekeeping aux-ADC. The typical operating power in FD mode is 30mW at a 7.5MHz clock frequency.

The Rx ADCs feature 54.8dB SINAD and 79.8dBc SFDR at 3.3MHz input frequency with a 7.5MHz clock frequency. The analog I/Q input amplifiers are fully differential and accept  $1.024V_{P-P}$  full-scale signals. Typical I/Q channel matching is  $\pm 0.01^{\circ}$  phase and  $\pm 0.01dB$  gain.

The Tx DACs feature 73.8dBc SFDR at  $f_{OUT}$  = 620kHz and  $f_{CLK}$  = 7.5MHz. The analog I-Q full-scale output voltage range is  $\pm 400 \text{mV}$  differential. The output DC common-mode voltage is from 0.89V to 1.36V. The I/Q channel offset is adjustable to optimize radio lineup sideband/carrier suppression. Typical I-Q channel matching is  $\pm 0.01dB$  gain and  $\pm 0.15^\circ$  phase.

Two independent 10-bit parallel, high-speed digital buses used by the Rx ADC and Tx DAC allow fullduplex operation for frequency-division duplex applications. The Rx ADC and Tx DAC can be disabled independently to optimize power management. A 3-wire serial interface controls power-management modes, the aux-DAC channels, and the aux-ADC channels.

The MAX19710 operates on a single 2.7V to 3.3V analog supply and 1.8V to 3.3V digital I/O supply. The MAX19710 is specified for the extended (-40°C to +85°C) temperature range and is available in a 56-pin, thin QFN package. The *Selector Guide* at the end of the data sheet lists other pin-compatible versions in this AFE family. For time-division duplex (TDD) applications, refer to the MAX19705–MAX19708 AFE family of products.

### **Applications**

Broadband Access Radio Private Mobile Radio

Portable Communication Equipment

### **Ordering Information**

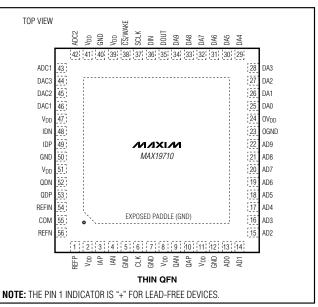
PART*	PIN-PACKAGE	PKG CODE
MAX19710ETN	56 Thin QFN-EP**	T5677-1
MAX19710ETN+	56 Thin QFN-EP**	T5677-1

\*All devices are specified over the -40°C to +85°C operating range. \*\*EP = Exposed paddle. +Denotes lead-free package.

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### Features

- Dual 10-Bit, 7.5Msps Rx ADC and Dual 10-Bit, 7.5Msps Tx DAC
- ♦ Ultra-Low Power 30mW at f<sub>CLK</sub> = 7.5MHz, FD Mode 21.3mW at f<sub>CLK</sub> = 7.5MHz, Slow Rx Mode 21.9mW at f<sub>CLK</sub> = 7.5MHz, Slow Tx Mode Low-Current Standby and Shutdown Modes
- Programmable Tx DAC Common-Mode DC Level and I/Q Offset Trim
- Excellent Dynamic Performance SNR = 54.9dB at f<sub>IN</sub> = 3.3MHz (Rx ADC) SFDR = 73.8dBc at f<sub>OUT</sub> = 620kHz (Tx DAC)
- ♦ Three 12-Bit, 1µs Aux-DACs
- 10-Bit, 333ksps Aux-ADC with 4:1 Input Mux and Data Averaging
- Excellent Gain/Phase Match ±0.01° Phase, ±0.01dB Gain (Rx ADC) at f<sub>IN</sub> = 1.8MHz
- Multiplexed Parallel Digital I/O
- Serial-Interface Control
- Versatile Power-Control Circuits Shutdown, Standby, Idle, Tx/Rx Disable
- Miniature 56-Pin Thin QFN Package (7mm x 7mm x 0.8mm)



Functional Diagram and Selector Guide appear at end of data sheet.

Maxim Integrated Products 1

Pin Configuration

nput Mux and

MAX19710

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

VDD to GND, OVDD to OGND GND to OGND	
IAP, IAN, QAP, QAN, IDP, IDN, QDP,	
QDN, DAC1, DAC2, DAC3 to GND	0.3V to VDD
ADC1, ADC2 to GND	0.3V to (VDD + 0.3V)
REFP, REFN, REFIN, COM to GND	
AD0-AD9, DA0-DA9, SCLK, DIN, CS/W	
CLK, DOUT to OGND	0.3V to (OV <sub>DD</sub> + 0.3V)

Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
56-Pin Thin QFN-EP (derate 27.8mW/°C abo	ve +70°C)2.22W
Thermal Resistance $\theta_{JA}$	
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	60°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **ELECTRICAL CHARACTERISTICS**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF$  on all digital outputs,  $f_{CLK} = 7.5MHz$  (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, CREFP = CREFN = C\_{COM} = 0.33\muF,  $C_L < 5pF$  on all aux-DAC outputs,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
POWER REQUIREMENTS	·	·				•
Analog Supply Voltage	V <sub>DD</sub>		2.7	3.0	3.3	V
Output Supply Voltage	OV <sub>DD</sub>		1.8		V <sub>DD</sub>	V
		FD mode: $f_{CLK} = 7.5$ MHz, $f_{OUT} = 620$ kHz on both DAC channels; $f_{IN} = 1.875$ MHz on both ADC channels; aux-DACs ON and at midscale, aux-ADC ON		10	12	
		SPI2-Tx mode: $f_{CLK} = 7.5MHz$ , $f_{OUT} = 620$ kHz on both DAC channels; Rx ADC OFF; aux-DACs ON and at midscale, aux-ADC ON		7.3	9	
V <sub>DD</sub> Supply Current		SPI1-Rx mode: $f_{CLK} = 7.5MHz$ , $f_{IN} = 1.875MHz$ on both ADC channels; Tx DAC OFF (Tx DAC outputs at 0V); aux-DACs ON and at midscale, aux-ADC ON		7.1	9	mA
		SPI4-Tx mode: $f_{CLK} = 7.5MHz$ , $f_{OUT} = 620kHz$ on both DAC channels; Rx ADC ON (output tri-stated); aux-DACs ON and at midscale, aux-ADC ON		9.7	12	

### ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF$  on all digital outputs,  $f_{CLK} = 7.5MHz$  (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, CREFP = C\_{REFN} = C\_{COM} = 0.33\muF, C<sub>L</sub> < 5pF on all aux-DAC outputs, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS	
		SPI3-Rx mode: $f_{CLK} = 7.5MHz$ , $f_{IN} = 1.875MHz$ on both channels; Tx DAC ON (Tx DAC outputs at midscale); aux-DACs ON and at midscale, aux-ADC ON		9.5	12		
V <sub>DD</sub> Supply Current		Standby mode: CLK = 0 or OV <sub>DD</sub> ; aux-DACs ON and at midscale, aux-ADC ON		2.7	3.5	mA	
		Idle mode: $f_{CLK} = 7.5MHz$ ; aux-DACs ON and at midscale, aux-ADC ON		4.6	6		
		Shutdown mode: CLK = 0 or OV <sub>DD</sub> , or aux-ADC OFF		0.5	5	μA	
OV <sub>DD</sub> Supply Current		FD mode: $f_{CLK} = 7.5MHz$ , $f_{OUT} = 620kHz$ on both DAC channels; $f_{IN} = 1.875MHz$ on both ADC channels; aux-DACs ON and at midscale, aux-ADC ON		0.94			
		SPI1-Rx and SPI3-Rx modes: $f_{CLK} =$ 7.5MHz, $f_{IN} =$ 1.875MHz on both ADC channels; DAC input bus tri-stated; aux-DACs ON and at midscale, aux-ADC ON		0.90		mA	
		SPI2-Tx and SPI4-Tx modes: $f_{CLK} =$ 7.5MHz, $f_{OUT} =$ 620kHz on both DAC channels; ADC output bus tri-stated; aux- DACs ON and at midscale, aux-ADC ON		52			
		Standby mode: CLK = 0 or OV <sub>DD</sub> ; aux- DACs ON and at midscale, aux-ADC ON		0.1		μA	
		Idle mode: f <sub>CLK</sub> = 7.5MHz; aux-DACs ON and at midscale, aux-ADC ON		12.8			
		Shutdown mode: CLK = 0 or OV <sub>DD</sub> , or aux-ADC OFF		0.1			
Rx ADC DC ACCURACY							
Resolution	N		10			Bits	
Integral Nonlinearity	INL			±0.5		LSB	
Differential Nonlinearity	DNL	No missing codes over temperature (Note 2)	-0.8	±0.4	+1.0	LSB	
Offset Error		Residual DC offset error	-5	±0.2	+5	%FS	
Gain Error		Includes reference error	-5	±0.9	+5	%FS	
DC Gain Matching			-0.15	±0.04	+0.15	dB	
Offset Matching				±11		LSB	



### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF$  on all digital outputs,  $f_{CLK} = 7.5MHz$  (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output,  $C_{REFP} = C_{REFN} = C_{COM} = 0.33\mu$ F,  $C_L < 5pF$  on all aux-DAC outputs,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}$ C.) (Note 1)

	±30 ±0.2 ±0.05 ±0.512 V <sub>DD</sub> /2 720 5		ppm/°C LSB V V
	±0.05 ±0.512 V <sub>DD</sub> / 2 720		V
	±0.512 V <sub>DD</sub> / 2 720		V
	V <sub>DD</sub> / 2 720		
	V <sub>DD</sub> / 2 720		
	720		V
			v
	5		kΩ
			pF
		7.5	MHz
	5		Clock
	5.5		Cycles
53.2	54.8		
	54.9		dB
53.1	54.7		dD
	54.8		dB
64.2	73.9		dDa
	79.8		dBc
	-71.7	-62.8	dBc
	-74.3		uвс
	-76.8		dDo
	-83.8		dBc
	-72		dBc
	-83		dBc
	3.5		ns
	2		psrms
	2		ns
=	-91		dB
	±0.01		dB
			Т
	53.1	53.1  54.9    53.1  54.7    54.8  54.8    64.2  73.9    79.8  -71.7    -74.3  -76.8    -83.8  -72    -83.3  -52    2  2	53.1  54.9    53.1  54.7    54.8  64.2    64.2  73.9    79.8  -71.7    -76.8  -76.8    -72  -83.8    3.5  2    2  2    -91  -91

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF$  on all digital outputs,  $f_{CLK} = 7.5MHz$  (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, CREFP = C\_{REFN} = C\_{COM} = 0.33\muF, C<sub>L</sub> < 5pF on all aux-DAC outputs, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Tx DAC DC ACCURACY						
Resolution	Ν		10			Bits
Integral Nonlinearity	INL			±0.3		LSB
Differential Nonlinearity	DNL	Guaranteed monotonic (Note 2)	-0.75	±0.2	+0.75	LSB
Residual DC Offset	Vos		-4	±1.2	+4	mV
Full-Scale Gain Error			-40	±1.6	+40	mV
Tx DAC DYNAMIC PERFORMANC	E					
DAC Conversion Rate	fCLK	(Note 3)			7.5	MHz
In-Band Noise Density	ND	$f_{OUT} = 620 \text{kHz}$		-120		dBFS/Hz
Third-Order Intermodulation Distortion	IM3	$f_{OUT1} = 620$ kHz, $f_{OUT2} = 640$ kHz		-79		dBc
Glitch Impulse				10		pV•s
Spurious-Free Dynamic Range to Nyquist	SFDR	f <sub>OUT</sub> = 620kHz	61	73.8		dBc
Total Harmonic Distortion to Nyquist	THD	f <sub>OUT</sub> = 620kHz		-72.2	-59.7	dBc
Signal-to-Noise Ratio to Nyquist	SNR	$f_{OUT} = 620 \text{kHz}$		55.1		dB
Tx DAC INTERCHANNEL CHARA	CTERISTICS	•				1
I-to-Q Output Isolation		f <sub>OUTX,Y</sub> = 2MHz, f <sub>OUTY,X</sub> = 2.2MHz		92		dB
Gain Mismatch Between I and Q Channels		Measured at DC	-0.4	±0.01	+0.4	dB
Phase Mismatch Between I and Q Channels		f <sub>OUT</sub> = 620kHz		±0.15		Degrees
Differential Output Impedance				800		Ω
Tx DAC ANALOG OUTPUT		·				
Full-Scale Output Voltage	V <sub>FS</sub>			±400		mV
		Bits CM1 = 0, CM0 = 0 (default)	1.29	1.36	1.42	
		Bits CM1 = 0, CM0 = 1	1.14	1.2	1.27	
Output Common-Mode Voltage	VCOMD	Bits CM1 = 1, CM0 = 0	0.96	1.05	1.15	V
		Bits CM1 = 1, CM0 = 1	0.78	0.89	1.03	_
Rx ADC-Tx DAC INTERCHANNEL	CHARACTE	RISTICS	1			1
Receive Transmit Isolation		ADC $f_{INI} = f_{INQ} = 1.8MHz$ , DAC $f_{OUTI} = f_{OUTQ} = 620$ kHz		92		dB
AUXILIARY ADCs (ADC1, ADC2)						
Resolution	N		10			Bits

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### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF$  on all digital outputs,  $f_{CLK} = 7.5MHz$  (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output,  $C_{REFP} = C_{REFN} = C_{COM} = 0.33 \mu$ F,  $C_L < 5$ pF on all aux-DAC outputs,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}$ C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Full-Scale Reference	Vocc	AD1 = 0 (default)		2.048		V
	VREF	AD1 = 1		V <sub>DD</sub>		v
Analog Input Range				0 to		V
				VREF		v
Analog Input Impedance		Measured at DC		500		kΩ
Input-Leakage Current		Measured at unselected input from 0 to VREF		±0.1		μΑ
Gain Error	GE	Includes reference error, $AD1 = 0$	-5		+5	%FS
Zero-Code Error	ZE			±2		mV
Differential Nonlinearity	DNL			±0.6		LSB
Integral Nonlinearity	INL			±0.6		LSB
Supply Current				210		μA
AUXILIARY DACs (DAC1, DAC2, D	AC3)					
Resolution	Ν		12			Bits
Integral Nonlinearity	INL	From code 100 to code 4000		±1.25		LSB
Differential Nonlinearity	DNL	Guaranteed monotonic over code 100 to code 4000 (Note 2)	-1.0	±0.65	+1.2	LSB
Output-Voltage Low	V <sub>OL</sub>	$R_L > 200 k\Omega$			0.2	V
Output-Voltage High	Voh	$R_L > 200 k\Omega$	2.57			V
DC Output Impedance		DC output at midscale		4		Ω
Settling Time		From code 1024 to code 3072, within ±10 LSB		1		μs
Glitch Impulse		From code 0 to code 4095		24		nV•s
Rx ADC-Tx DAC TIMING CHARAC	TERISTICS					
CLK Rise to Channel-I Output Data Valid	tdoi	Figure 3 (Note 2)	5.5	8.2	11.5	ns
CLK Fall to Channel-Q Output Data Valid	t <sub>DOQ</sub>	Figure 3 (Note 2)	6.5	9.3	13.0	ns
I-DAC DATA to CLK Fall Setup Time	tDSI	Figure 5 (Note 2)	10			ns
Q-DAC DATA to CLK Rise Setup Time	tdsq	Figure 5 (Note 2)	10			ns
CLK Fall to I-DAC Data Hold Time	tDHI	Figure 5 (Note 2)	0			ns
CLK Rise to Q-DAC Data Hold Time	t <sub>DHQ</sub>	Figure 5 (Note 2)	0			ns
CLK Duty Cycle				50		%
CLK Duty-Cycle Variation				±15		%
Digital Output Rise/Fall Time		20% to 80%		2.4		ns

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF on all digital outputs, f_{CLK} = 7.5MHz (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, CREFP = CREFN = C_{COM} = 0.33\muF, C_L < 5pF on all aux-DAC outputs, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 1)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
SERIAL-INTERFACE TIMING CHAI	RACTERISTI	CS (Figures 6 and 8, Note 2)				•
Falling Edge of CS/WAKE to Rising Edge of First SCLK Time	tcss		10			ns
DIN to SCLK Setup Time	t <sub>DS</sub>		10			ns
DIN to SCLK Hold Time	tDН		0			ns
SCLK Pulse-Width High	tсн		25			ns
SCLK Pulse-Width Low	tCL		25			ns
SCLK Period	tCP		50			ns
SCLK to CS/WAKE Setup Time	tcs		10			ns
CS/WAKE High Pulse Width	tcsw		80			ns
CS/WAKE High to DOUT Active High	tCSD	Bit AD0 set		200		ns
CS/WAKE High to DOUT Low (Aux-ADC Conversion Time)	tCONV	Bit AD0 set, no averaging, $f_{CLK} = 7.5 MHz$ , CLK divider = 2		4.3		μs
DOUT Low to CS/WAKE Setup Time	tDCS	Bit AD0, AD10 set		200		ns
SCLK Low to DOUT Data Out	tCD	Bit AD0, AD10 set			14.5	ns
CS/WAKE High to DOUT High Impedance	tCHZ	Bit AD0, AD10 set		200		ns
MODE-RECOVERY TIMING CHARA	ACTERISTICS	S (Figure 7)				
		From shutdown to Rx mode, ADC settles to within 1dB SINAD		500		
		From shutdown to Tx mode, DAC settles to within 10 LSB error		26.2		
Shutdown Wake-Up Time (With CLK)	twake,sd	From aux-ADC enable to aux-ADC start conversion		10		μs
		From shutdown to aux-DAC output valid		28		
		From shutdown to FD mode, ADC settles to within 1dB SINAD, DAC settles to within 10 LSB error		500		
		From idle to Rx mode, ADC settles to within 1dB SINAD		7.3		
Idle Wake-Up Time	twake,sto	From idle to Tx mode, DAC settles to 10 LSB error		5.2		μs
(With CLK)		From idle to FD mode, ADC settles to within 1dB SINAD, DAC settles to within 10 LSB error		7.3		

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF on all digital outputs, f_{CLK} = 7.5MHz (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, C_{REFP} = C_{REFN} = C_{COM} = 0.33\mu$ F, C<sub>L</sub> < 5pF on all aux-DAC outputs, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
		From standby to Rx mode, ADC settles to within 1dB SINAD		7.5		
Standby Wake-Up Time (With CLK)	<sup>t</sup> WAKE,ST1	From standby to Tx mode, DAC settles to 10 LSB error		22.2		μs
		From standby to FD mode, ADC settles to within 1dB SINAD, DAC settles to within 10 LSB error		22.2		
Enable Time from Tx to Rx, Fast Mode	<sup>t</sup> ENABLE,RX	ADC settles to within 1dB SINAD		0.1		μs
Enable Time from Rx to Tx, Fast Mode	<sup>t</sup> ENABLE,TX	DAC settles to within 10 LSB error		0.1		μs
Enable Time from Tx to Rx, Slow Mode	t <sub>ENABLE,RX</sub>	ADC settles to within 1dB SINAD		7.3		μs
Enable Time from Rx to Tx, Slow Mode	tenable,tx	DAC settles to within 10 LSB error		5.2		μs
INTERNAL REFERENCE (V <sub>REFIN</sub> =	VDD; VREFP,	VREFN, VCOM levels are generated interna	lly)			
Positive Reference		VREFP - VCOM		0.256		V
Negative Reference		VREFN - VCOM		-0.256		V
Common-Mode Output Voltage	V <sub>COM</sub>		V <sub>DD</sub> / 2 - 0.15	V <sub>DD</sub> / 2	V <sub>DD</sub> / 2 + 0.15	V
Maximum REFP/REFN/COM Source Current	ISOURCE			2		mA
Maximum REFP/REFN/COM Sink Current	ISINK			2		mA
Differential Reference Output	VREF	VREFP - VREFN	+0.490	+0.512	+0.534	V
Differential Reference Temperature Coefficient	REFTC			±30		ppm/°C
<b>BUFFERED EXTERNAL REFEREN</b>	CE (external	V <sub>REFIN</sub> = 1.024V applied; V <sub>REFP</sub> , V <sub>REFN</sub> , V <sub>C</sub>	OM level	s are gen	erated in	ternally)
Reference Input Voltage	VREFIN			1.024		V
Differential Reference Output	VDIFF	VREFP - VREFN		0.512		V
Common-Mode Output Voltage	VCOM			V <sub>DD</sub> / 2		V
Maximum REFP/REFN/COM Source Current	ISOURCE			2		mA
Maximum REFP/REFN/COM Sink Current	ISINK			2		mA
REFIN Input Current				-0.7		μA
REFIN Input Resistance				500		kΩ

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF$  on all digital outputs,  $f_{CLK} = 7.5MHz$  (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, CREFP = CREFN = CCOM = 0.33\muF, C\_L < 5pF on all aux-DAC outputs,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
DIGITAL INPUTS (CLK, SCLK, D	IN, CS/WAKE,	DA9–DA0)				
Input High Threshold	VINH		0.7 x OV <sub>DI</sub>	D		V
Input Low Threshold	VINL			0.3 x	OVDD	V
	Dhu	CLK, SCLK, DIN, $\overline{CS}$ /WAKE = OGND or OV <sub>DD</sub>	-1		+1	
Input Leakage	DI <sub>IN</sub>	$DA9-DA0 = OV_{DD}$	-1		+1	μA
		DA9–DA0 = OGND	-5		+5	
Input Capacitance	DCIN			5		pF
DIGITAL OUTPUTS (AD9-AD0, D	OUT)					
Output-Voltage Low	Vol	I <sub>SINK</sub> = 200μA		0.2 x	OVDD	V
Output-Voltage High	VOH	ISOURCE = 200µA	0.8 x OV <sub>DI</sub>	D		V
Tri-State Leakage Current	ILEAK		-1		+1	μA
Tri-State Output Capacitance	Cout			5		pF

Note 1: Specifications from  $T_A = +25^{\circ}C$  to  $+85^{\circ}C$  guaranteed by production tests. Specifications at  $T_A < +25^{\circ}C$  guaranteed by design and characterization.

Note 2: Guaranteed by design and characterization.

**Note 3:** The minimum clock frequency ( $f_{CLK}$ ) for the MAX19710 is 1.5MHz (typ). The minimum aux-ADC sample rate clock frequency ( $A_{CLK}$ ) is determined by  $f_{CLK}$  and the chosen aux-ADC clock-divider value. The minimum aux-ADC  $A_{CLK} > 1.5MHz / 128 = 11.7kHz$ . The aux-ADC conversion time does not include the time to clock the serial data out of DOUT. The maximum conversion time (for no averaging, NAVG = 1) will be  $t_{CONV}$  (max) = ( $12 \times 1 \times 128$ ) /  $1.5MHz = 1024\mu$ s.

Note 4: SNR, SINAD, SFDR, HD3, and THD are based on a differential analog input voltage of -0.5dBFS referenced to the amplitude of the digital outputs. SINAD and THD are calculated using HD2 through HD6.

Note 5: Crosstalk rejection is measured by applying a high-frequency test tone to one channel and a low-frequency tone to the second channel. FFTs are performed on each channel. The parameter is specified as the power ratio of the first and second channel FFT test tones.

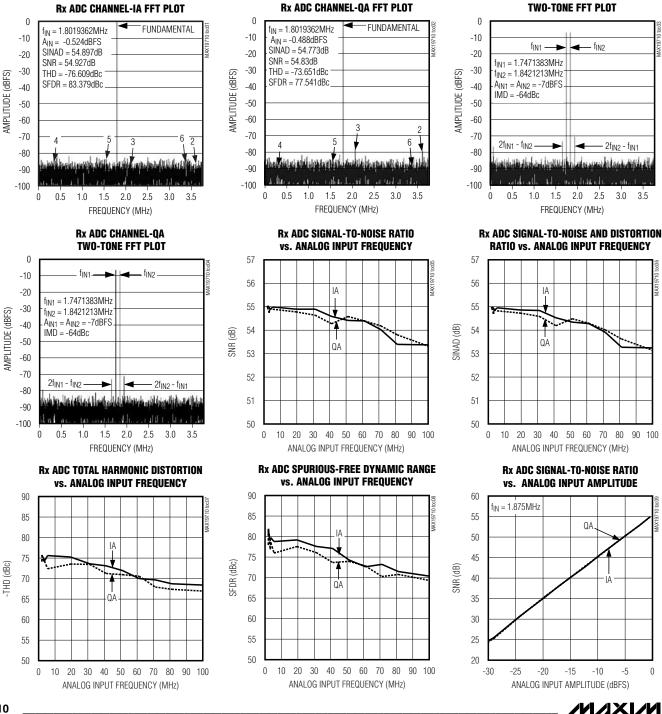
**Note 6:** Amplitude and phase matching are measured by applying the same signal to each channel, and comparing the two output signals using a sine-wave fit.

MAX19710

### Typical Operating Characteristics

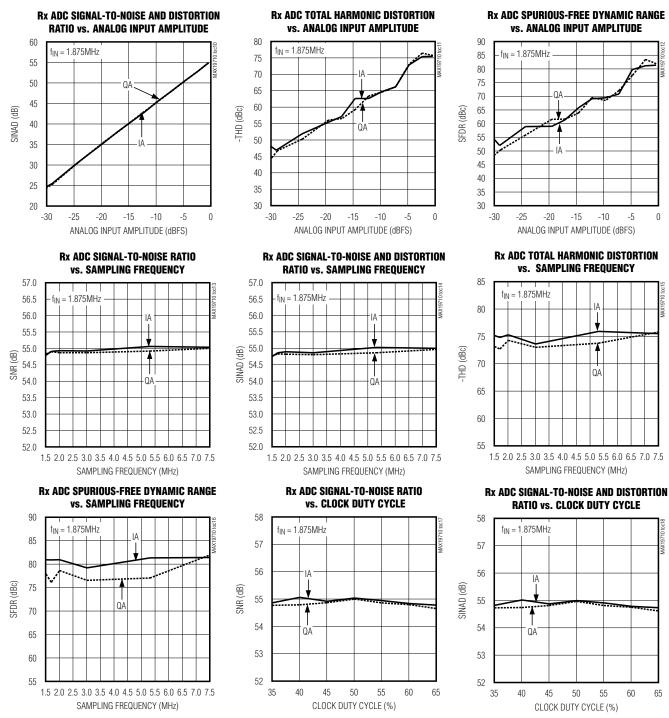
**Rx ADC CHANNEL-IA** 

(V<sub>DD</sub> = 3V, OV<sub>DD</sub> = 1.8V, internal reference (1.024V), C<sub>L</sub> ≈ 10pF on all digital outputs, f<sub>CLK</sub> = 7.5MHz (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output,  $C_{REFP} = C_{REFN} = C_{COM} = 0.33 \mu$ F,  $T_A = +25$ °C, unless otherwise noted.)



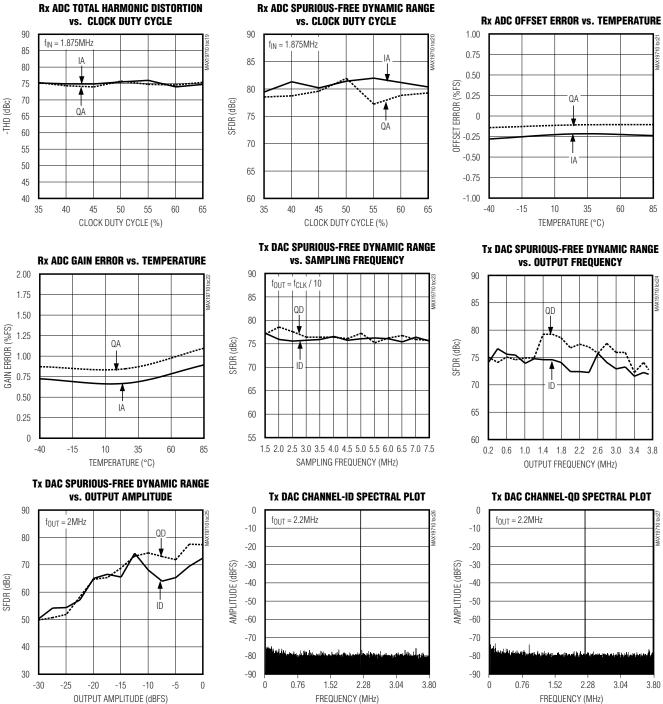
### **Typical Operating Characteristics (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF on all digital outputs, f_{CLK} = 7.5MHz (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, C_{REFP} = C_{REFN} = C_{COM} = 0.33\mu$ F, TA = +25°C, unless otherwise noted.)



### Typical Operating Characteristics (continued)

(V<sub>DD</sub> = 3V, OV<sub>DD</sub> = 1.8V, internal reference (1.024V), C<sub>L</sub> ≈ 10pF on all digital outputs, f<sub>CLK</sub> = 7.5MHz (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output,  $C_{REFP} = C_{REFN} = C_{COM} = 0.33 \mu$ F,  $T_A = +25^{\circ}$ C, unless otherwise noted.)

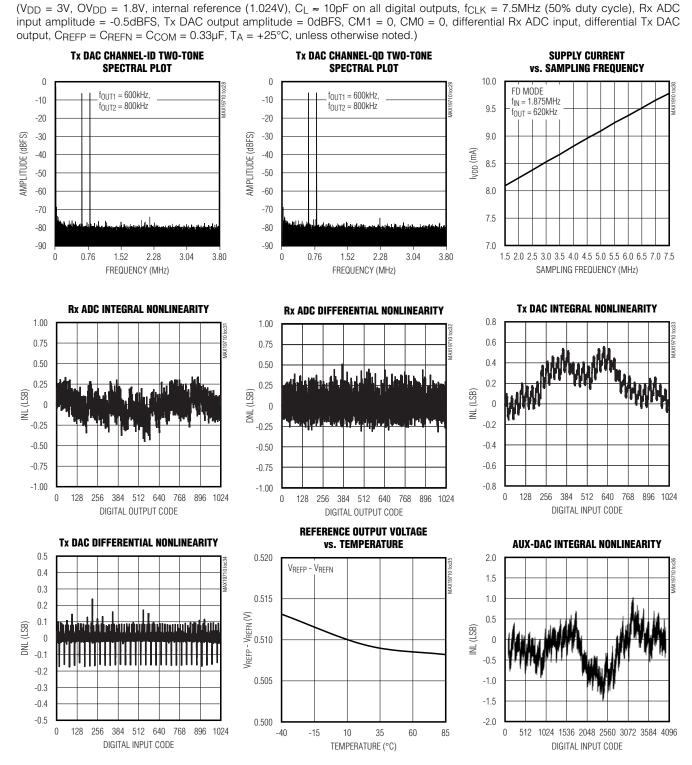


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Typical Operating Characteristics (continued)



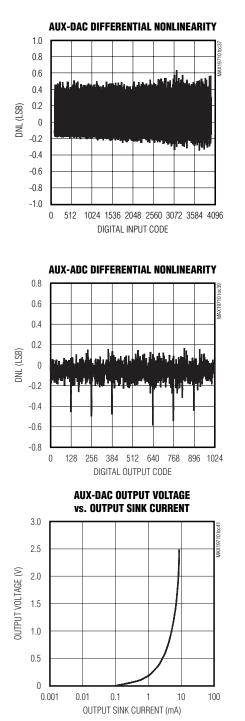
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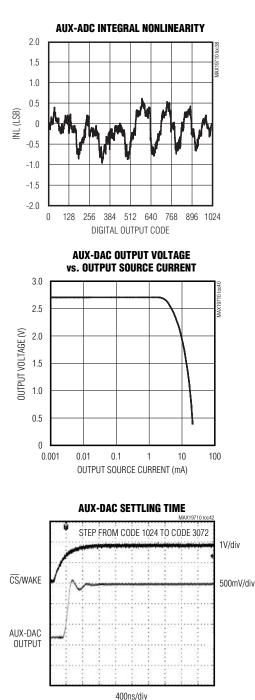
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### **Typical Operating Characteristics (continued)**

 $(V_{DD} = 3V, OV_{DD} = 1.8V, internal reference (1.024V), C_L \approx 10pF on all digital outputs, f_{CLK} = 7.5MHz (50% duty cycle), Rx ADC input amplitude = -0.5dBFS, Tx DAC output amplitude = 0dBFS, CM1 = 0, CM0 = 0, differential Rx ADC input, differential Tx DAC output, C_{REFP} = C_{REFN} = C_{COM} = 0.33\mu$ F, T<sub>A</sub> = +25°C, unless otherwise noted.)







Pin Description

PIN	NAME	FUNCTION			
1	REFP	Positive Reference Voltage Input Terminal. Bypass with a $0.33\mu$ F capacitor to GND as close to REFP as possible.			
2, 8, 11, 39, 41, 47, 51	V <sub>DD</sub>	Analog Supply Voltage. Bypass $V_{DD}$ to GND with a combination of a 2.2 $\mu F$ capacitor in parallel with a 0.1 $\mu F$ capacitor.			
3	IAP	Channel-IA Positive Analog Input. For single-ended operation, connect signal source to IAP.			
4	IAN	Channel-IA Negative Analog Input. For single-ended operation, connect IAN to COM.			
5, 7, 12, 40, 50	GND	Analog Ground. Connect all GND pins to ground plane.			
6	CLK	Conversion Clock Input. Clock signal for both receive ADCs and transmit DACs.			
9	QAN	Channel-QA Negative Analog Input. For single-ended operation, connect QAN to COM.			
10	QAP	Channel-QA Positive Analog Input. For single-ended operation, connect signal source to QAP.			
13–22	Beceive ADC Digital Outputs AD9 is the most significant bit (MSB) and AD0 is the lease				
23	OGND	Output-Driver Ground			
24	OV <sub>DD</sub>	Output-Driver Power Supply. Supply range from $+1.8V$ to V <sub>DD</sub> . Bypass OV <sub>DD</sub> to OGND with a combination of a 2.2µF capacitor in parallel with a 0.1µF capacitor.			
25–34	DA0-DA9	Transmit DAC Digital Inputs. DA9 is the most significant bit (MSB) and DA0 is the least significant bit (LSB). DA0–DA9 are internally pulled up to $OV_{DD}$ .			
35	DOUT	Aux-ADC Digital Output			
36	DIN	3-Wire Serial-Interface Data Input. Data is latched on the rising edge of SCLK.			
37	SCLK	3-Wire Serial-Interface Clock Input			
38	CS/WAKE	3-Wire Serial-Interface Chip-Select/WAKE Input. When the MAX19710 is in shutdown, CS/WAKE controls the wake-up function. See the <i>Wake-Up Function</i> section.			
42	ADC2	Selectable Auxiliary ADC Analog Input 2			
43	ADC1	Selectable Auxiliary ADC Analog Input 1			
44	DAC3	Auxiliary DAC3 Analog Output (V <sub>OUT</sub> = 0 at Power-Up)			
45	DAC2	Auxiliary DAC2 Analog Output (V <sub>OUT</sub> = 0 at Power-Up)			
46	DAC1	Auxiliary DAC1 Analog Output (AFC DAC, V <sub>OUT</sub> = 1.1V at Power-Up)			
48	IDN	Tx Path Channel-ID Differential Negative Output			
49	IDP	Tx Path Channel-ID Differential Positive Output			
52	QDN	Tx Path Channel-QD Differential Negative Output			
53	QDP	Tx Path Channel-QD Differential Positive Output			
54	REFIN	Reference Input. Connect to V <sub>DD</sub> for internal reference.			
55	COM	Common-Mode Voltage I/O. Bypass COM to GND with a 0.33µF capacitor.			
56	REFN	Negative Reference Voltage Input Terminal. Rx ADC conversion range is $\pm$ (V <sub>REFP</sub> - V <sub>REFN</sub> ). Bypass REFN to GND with a 0.33µF capacitor.			
	EP	Exposed Paddle. Exposed paddle is internally connected to GND. Connect EP to the GND plane.			

### **Detailed Description**

The MAX19710 integrates a dual, 10-bit Rx ADC and a dual, 10-bit Tx DAC while providing ultra-low power and high dynamic performance at 7.5Msps conversion rate. The Rx ADC analog input amplifiers are fully differ-

ential and accept 1.024VP-P full-scale signals. The Tx DAC analog outputs are fully differential with  $\pm$ 400mV full-scale output, selectable common-mode DC level, and adjustable channel ID–QD offset trim.

The MAX19710 integrates three 12-bit auxiliary DACs (aux-DACs) and a 10-bit, 333ksps auxiliary ADC (aux-ADC) with 4:1 input multiplexer. The aux-DAC channels feature 1µs settling time for fast AGC, VGA, and AFC level setting. The aux-ADC features data averaging to reduce processor overhead and a selectable clock-divider to program the conversion rate.

The MAX19710 includes a 3-wire serial interface to control operating modes and power management. The serial interface is SPI<sup>TM</sup> and MICROWIRE<sup>TM</sup> compatible. The MAX19710 serial interface selects shutdown, idle, standby, FD, transmit (Tx), and receive (Rx) modes, as well as controls aux-DAC and aux-ADC channels.

The MAX19710 features two independent, high-speed, 10-bit buses for the Rx ADC and Tx DAC, which allow full-duplex (FD) operation for frequency-division duplex applications. Each bus can be disabled to optimize power management through the 3-wire interface. The

MICROWIRE is a trademark of National Semiconductor Corp. SPI is a trademark of Motorola, Inc. MAX19710 operates from a single 2.7V to 3.3V analog supply and a 1.8V to 3.3V digital supply.

#### **Dual 10-Bit Rx ADC**

The ADC uses a seven-stage, fully differential, pipelined architecture that allows for high-speed conversion while minimizing power consumption. Samples taken at the inputs move progressively through the pipeline stages every half clock cycle. Including the delay through the output latch, the total clock-cycle latency is 5 clock cycles for channel IA and 5.5 clock cycles for channel QA. The ADC full-scale analog input range is  $\pm$ VREF with a V<sub>DD</sub> / 2 ( $\pm$ 0.8V) common-mode input range. VREF is the difference between VREFP and VREFN. See the *Reference Configurations* section for details.

#### Input Track-and-Hold (T/H) Circuits

Figure 1 displays a simplified diagram of the Rx ADC input track-and-hold (T/H) circuitry. Both ADC inputs (IAP, QAP, IAN, and QAN) can be driven either differen-

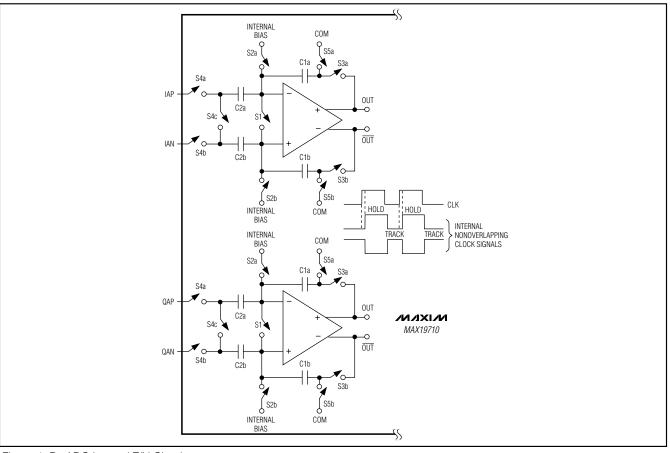


Figure 1. Rx ADC Internal T/H Circuits

DIFFERENTIAL INPUT VOLTAGE	DIFFERENTIAL INPUT (LSB)	OFFSET BINARY (AD0-AD9)	OUTPUT DECIMAL CODE	
V <sub>REF</sub> x 512/512	511 (+Full Scale - 1 LSB)	11 1111 1111	1023	
V <sub>REF</sub> x 511/512	510 (+Full Scale - 2 LSB)	11 1111 1110	1022	
V <sub>REF</sub> x 1/512	+1	10 0000 0001	513	
V <sub>REF</sub> x 0/512	0 (Bipolar Zero)	10 0000 0000	512	
-V <sub>REF</sub> x 1/512	-1	01 1111 1111	511	
-V <sub>REF</sub> x 511/512	-511 (-Full Scale + 1 LSB)	00 0000 0001	1	
-V <sub>REF</sub> x 512/512	-512 (-Full Scale)	00 0000 0000	0	

#### Table 1. Rx ADC Output Codes vs. Input Voltage

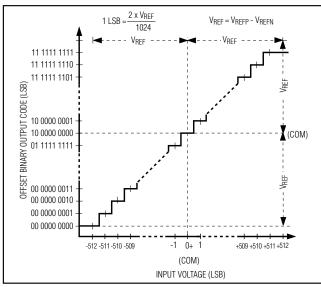


Figure 2. Rx ADC Transfer Function

tially or single-ended. Match the impedance of IAP and IAN, as well as QAP and QAN, and set the input signal common-mode voltage within the V<sub>DD</sub> / 2 ( $\pm$ 0.8V) Rx ADC range for optimum performance.

#### **Rx ADC System Timing Requirements**

Figure 3 shows the relationship between the clock, analog inputs, and the resulting output data. Channels IA and QA are sampled on the rising edge of the clock sig-

nal (CLK) and the resulting data is multiplexed at the AD0–AD9 outputs. Channel IA data is updated on the rising edge and channel QA data is updated on the falling edge of CLK. Including the delay through the output latch, the total clock-cycle latency is 5 clock cycles for channel IA and 5.5 clock cycles for channel QA.

#### Digital Output Data (AD0-AD9)

AD0–AD9 are the Rx ADC digital logic outputs of the MAX19710. The logic level is set by OV<sub>DD</sub> from 1.8V to V<sub>DD</sub>. The digital output coding is offset binary (Table 1). Keep the capacitive load on the digital outputs AD0–AD9 as low as possible (< 15pF) to avoid large digital currents feeding back into the analog portion of the MAX19710 and degrading its dynamic performance. Buffers on the digital outputs isolate the outputs from heavy capacitive loads. Adding 100 $\Omega$  resistors in series with the digital outputs close to the MAX19710 will help improve ADC performance. Refer to the MAX19710EVKIT schematic for an example of the digital outputs driving a digital buffer through 100 $\Omega$  series resistors.

During SHDN, IDLE, STBY, SPI2, and SPI4 states, digital outputs AD0–AD9 are tri-stated.

#### **Dual 10-Bit Tx DACs**

The dual 10-bit digital-to-analog converters (Tx DACs) operate with clock speeds up to 7.5MHz. The Tx DAC digital inputs, DA0–DA9, are multiplexed on a single 10-bit transmit bus. The voltage reference determines the Tx DAC full-scale voltage at IDP, IDN and QDP, QDN analog outputs. See the *Reference Configurations* section for setting the reference voltage.



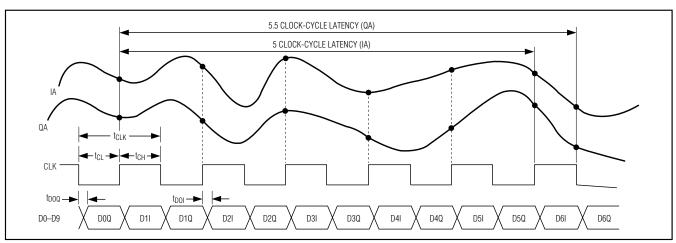


Figure 3. Rx ADC System Timing Diagram

### Table 2. Tx DAC Output Voltage vs. Input Codes

(Internal Reference Mode  $V_{REFDAC} = 1.024V$ , External Reference Mode  $V_{REFDAC} = V_{REFIN}$ ,  $V_{FS} = 400$  for 800mV<sub>P-P</sub> Full Scale)

DIFFERENTIAL OUTPUT VOLTAGE (V)	OFFSET BINARY (DA0–DA9)	INPUT DECIMAL CODE
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	11 1111 1111	1023
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	11 1111 1110	1022
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	10 0000 0001	513
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	10 0000 0000	512
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	01 1111 1111	511
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	00 0000 0001	1
$(V_{FS}) \frac{V_{REFDAC}}{1024} \times \frac{1023}{1023}$	00 0000 0000	0

The Tx DAC outputs (IDN, IDP, QDN, QDP) are biased at an adjustable common-mode DC level and designed to drive a differential input stage with  $\geq$  70k $\Omega$  input impedance. This simplifies the analog interface between RF quadrature upconverters and the MAX19710. Many RF upconverters require a 0.89V to 1.36V common-mode bias. The MAX19710 common-mode DC bias eliminates discrete level-setting resistors and code-generated level shifting while preserving the full dynamic range of each Tx DAC. **The Tx DAC differential analog outputs cannot be used in single-ended mode because of the**  **internally generated common-mode DC level.** Table 2 shows the Tx DAC output voltage vs. input codes. Table 10 shows the selection of DC common-mode levels. See Figure 4 for an illustration of the Tx DAC analog output levels.

The Tx DAC also features independent DC offset trim on each ID–QD channel. This feature is configured through the SPI interface. The DC offset correction is used to optimize sideband and carrier suppression in the Tx signal path (see Table 9).



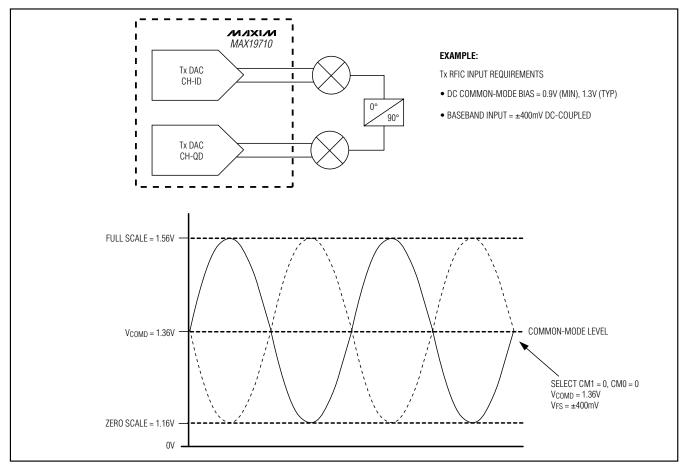


Figure 4. Tx DAC Common-Mode DC Level at IDN, IDP or QDN, QDP Differential Outputs

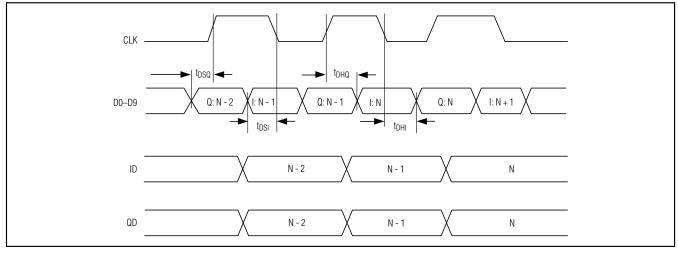


Figure 5. Tx DAC System Timing Diagram

**MAX19710** 

#### Tx DAC Timing

Figure 5 shows the relationship among the clock, input data, and analog outputs. Channel ID data is latched on the falling edge of the clock signal, and channel QD data is latched on the rising edge of the clock signal, at which point both ID and QD outputs are simultaneously updated.

#### 3-Wire Serial Interface and Operation Modes

The 3-wire serial interface controls the MAX19710 operation modes as well as the three 12-bit aux-DACs and the 10-bit aux-ADC. Upon power-up, program the MAX19710 to operate in the desired mode. Use the 3wire serial interface to program the device for shutdown, idle, standby, FD, Rx, Tx, aux-DAC controls, or aux-ADC conversion. A 16-bit data register sets the mode control as shown in Table 3. The 16-bit word is composed of four control bits (A3–A0) and 12 data bits (D11–D0). Data is shifted in MSB first (D11) and LSB last (A0) format. Table 4 shows the MAX19710 power-management modes. Table 5 shows the SPI-controlled Tx, Rx, and FD modes. The serial interface remains active in all modes.

#### SPI Register Description

Program the control bits, A3–A0, in the register as shown in Table 3 to select the operating mode. Modify A3–A0 bits to select from ENABLE-16, Aux-DAC1, Aux-DAC2, Aux-DAC3, IOFFSET, QOFFSET, COMSEL, Aux-ADC, ENABLE-8, and WAKEUP-SEL modes. ENABLE-16 is the default operating mode (see Table 6). This mode allows for shutdown, idle, and standby states as well as switching between FAST, SLOW, Rx and Tx modes. Tables 4 and 5 show the required SPI settings for each mode.

In ENABLE-16 mode, the aux-DACs have independent control bits E4, E5, and E6, and bit E9 enables the aux-ADC. Table 7 shows the auxiliary DAC enable codes. Table 8 shows the auxiliary ADC enable code. Bits E11 and E10 are reserved. Program bits E11 and E10 to logic-low. Bits E3, E7, and E8 are not used.

Modes aux-DAC1, aux-DAC2, and aux-DAC3 select the aux-DAC channels named DAC1, DAC2, and DAC3 and hold the data inputs for each DAC. Bits \_D11-\_D0 are the data inputs for each aux-DAC and can be programmed through SPI. The MAX19710 also includes two 6-bit registers that can be programmed to adjust the offsets for the Tx DAC ID and QD channels independently (see Table 9). Use the COMSEL mode to select the output common-mode voltage with bits CM1 and CM0 (see Table 10). Use the aux-ADC mode to start the auxiliary ADC conversion (see the *10-Bit, 333ksps*)

*Auxiliary ADC* section for details). Use ENABLE-8 mode for faster enable and switching between shutdown, idle, and standby states as well as switching between FAST, SLOW, Rx and Tx modes and the FD mode.

The WAKEUP-SEL register selects the operating mode that the MAX19710 is to enter immediately after coming out of shutdown (Table 11). See the *Wake-Up Function* section for more information.

Shutdown mode offers the most dramatic power savings by shutting down all the analog sections (including the reference) of the MAX19710. In shutdown mode, the Rx ADC digital outputs are in tri-state mode, the Tx DAC digital inputs are internally pulled to OV<sub>DD</sub>, and the Tx DAC outputs are at OV. When the Rx ADC outputs transition from tri-state to active mode, the last converted word is placed on the digital output bus. The Tx DAC previously stored data is lost when coming out of shutdown mode. The wake-up time from shutdown mode is dominated by the time required to charge the capacitors at REFP, REFN, and COM. In internal reference mode and buffered external reference mode, the wake-up time is typically 500µs to enter Rx mode. 26.2µs to enter Tx mode, and 500µs to enter FD mode.

In all operating modes the Tx DAC inputs DA0–DA9 are internally pulled to  $OV_{DD}$ . To reduce the supply current of the MAX19710 in shutdown mode do not pull DA0–DA9 low. This consideration is especially important in shutdown mode to achieve the lowest quiescent current.

In idle mode, the reference and clock distribution circuits are powered, but all other functions are off. The Rx ADC outputs AD0–AD9 are forced to tri-state. The Tx DAC DA0–DA9 inputs are internally pulled to  $OV_{DD}$ , while the Tx DAC outputs are at 0V. The wake-up time is 7.3µs to enter Rx mode, 5.2µs to enter Tx mode, and 7.3µs to enter FD mode. When the Rx ADC outputs transition from tri-state to active, the last converted word is placed on the digital output bus.

In standby mode, the reference is powered but all other device functions are off. The wake-up time from standby mode is 7.5µs to enter Rx mode, 22.2µs to enter Tx mode, and 22.2µs to enter FD mode. When the Rx ADC outputs transition from tri-state to active, the last converted word is placed on the digital output bus.

#### FAST and SLOW Rx and Tx Modes

The MAX19710 features FAST and SLOW modes for switching between Rx and Tx operation. In FAST Tx mode, the Rx ADC core is powered on but the ADC digital outputs AD0–AD9 are tri-stated. The Tx DAC digital bus is active and the DAC core is fully operational.



REGISTER	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	A3	A2	<b>A</b> 1	A0
NAME	(MSB)	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1 (LSB)
ENABLE-16	E11 = 0 Reserved	E10 = 0 Reserved	E9	_	_	E6	E5	E4	_	E2	E1	E0	0	0	0	0
Aux-DAC1	1D11	1D10	1D9	1D8	1D7	1D6	1D5	1D4	1D3	1D2	1D1	1D0	0	0	0	1
Aux-DAC2	2D11	2D10	2D9	2D8	2D7	2D6	2D5	2D4	2D3	2D2	2D1	2D0	0	0	1	0
Aux-DAC3	3D11	3D10	3D9	3D8	3D7	3D6	3D5	3D4	3D3	3D2	3D1	3D0	0	0	1	1
IOFFSET	_	_		_			105	104	103	102	IO1	100	0	1	0	0
QOFFSET	_					_	Q05	Q04	QO3	Q02	Q01	QOO	0	1	0	1
COMSEL	—	_	_	_	_	_	_	_	_	_	CM1	CM0	0	1	1	0
Aux-ADC	AD11 = 0 Reserved	AD10	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	AD1	AD0	0	1	1	1
ENABLE-8	_	_	_	_	_	_	_	_		E2	E1	E0	1	0	0	0
WAKEUP-SEL	_	_		—		—	_		—	W2	W1	W0	1	0	0	1

### Table 3. MAX19710 Mode Control

— = Not used.

### Table 4. Power-Management Modes

	ADD	RESS			DATA	BITS	i		FUNCTION (POWER		
A3	A2	A1	<b>A</b> 0	E9*	E2	E1	E0	MODE	MANAGEMENT)	DESCRIPTION	COMMENT
				1	0	0	0	SHDN	SHUTDOWN	Rx ADC = OFF Tx DAC = OFF (TX DAC outputs at 0V) Aux-DAC = OFF Aux-ADC = OFF CLK = OFF REF = OFF	Device is in complete shutdown.
	0000 (16-Bit Mode) or 1000 (8-Bit Mode)			X**	0	0	1	IDLE	IDLE	Rx ADC = OFF Tx DAC = OFF (TX DAC outputs at 0V) Aux-DAC = Last State CLK = ON REF = ON	Fast turn-on time. Moderate idle power.
			X**	0	1	0	STBY	STANDBY	Rx ADC = OFF Tx DAC = OFF (TX DAC outputs at 0V) Aux-DAC = Last State CLK = OFF REF = ON	Slow turn-on time. Low standby power.	

X = Don't care.

\*Bit E9 is not available in 8-bit mode.

\*\* In IDLE and STBY modes, the aux-ADC can be turned on or off.



### Table 5. MAX19710 Tx, Rx, and FD Control Using SPI Commands

ADDRESS DATA BITS FUNCTION MODE DESCRIPTION COMMENT (Tx-Rx SWITCHING SPEED) A3 A2 A1 A0 E2 E1 E0 Bx Mode Rx ADC = ON Slow transition to Tx Rx Bus = Enabled mode from this SLOW 0 1 1 SPI1-Rx Tx DAC = OFF mode. (Tx DAC outputs at 0V) Low power. Tx Bus = OFF (all inputs are pulled high) Tx Mode: Slow transition to Rx Bx ADC = OFFmode from this SPI2-Tx 0 0 SLOW Rx Bus = Tri-state 1 mode. Tx DAC = ONLow power. Tx Bus = ON 0000 **Rx Mode:** Rx ADC = ON (16-Bit Mode) Rx Bus = Enabled and Fast transition to Tx 1000 Tx DAC = ONmode from this 1 0 SPI3-Rx FAST 1 (8-Bit Mode) (Tx DAC outputs at mode. Moderate midscale) power. Tx Bus = OFF (all inputs are pulled high) Tx Mode: Fast transition to Rx Rx ADC = ON mode from this SPI4-Tx FAST Rx Bus = Tri-state 1 1 0 mode. Moderate Tx DAC = ON power. Tx Bus = ON FD Mode: **Default Mode** Rx ADC = ONFast transition to any 1 1 1 FD FAST Rx Bus = ONmode. Moderate Tx DAC = ONpower. Tx Bus = ON

In FAST Rx mode, the Tx DAC core is powered on. The Tx DAC outputs are set to midscale. In this mode, the Tx DAC input bus is disconnected from the DAC core and DA0–DA9 are internally pulled to OV<sub>DD</sub>. The Rx ADC digital bus is active and the ADC core is fully operational.

In FAST mode, the switching time from Tx to Rx, or Rx to Tx is minimized because the converters are on and do not have to recover from a power-down state. In FAST mode, the switching time from Rx to Tx and Tx to Rx is  $0.1\mu$ s. Power consumption is higher in FAST mode because both Tx and Rx cores are always on.

In SLOW Tx mode, the Rx ADC core is powered off and the ADC digital outputs AD0–AD9 are tri-stated. The Tx DAC digital bus is active and the DAC core is fully operational. In SLOW Rx mode, the Tx DAC core is powered off. The Tx DAC outputs are set to 0. In SLOW Rx mode, the Tx DAC input bus is disconnected from the DAC core and DA0–DA9 are internally pulled to OV<sub>DD</sub>. The Rx ADC digital bus is active and the ADC core is fully operational. The switching times for SLOW modes are 5.2µs for Rx to Tx and 7.3µs for Tx to Rx.

	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	
REGISTER NAME	16 (MSB)	15	14	13	12	11	10	9	8	7	6	5	
			0			0	0	0		1	1	1	
ENABLE-16	0	0	Aux-ADC = ON	_	—		Aux-DAC1 t Ix-DAC3 = (		—	FD mode			
Aux-DAC1	0	1	1	0	1	0	0	0	1	1	0	0	
					DA	AC1 outpu	ut set to 1.	1V					
Aux-DAC2	0	0	0	0	0	0	0	0	0	0	0	0	
					D		ut set to 0		1		1		
Aux-DAC3	0	0	0	0	0	0	0	0	0	0	0	0	
	DAC3 output set to 0V												
IOFFSET	_	_	_		_		0	0	0	0	0	0	
									o offset or	1			
QOFFSET		_	_		_	_	0	0	0	0	0	0	
								No	No offset on channel QD				
COMSEL								_			0	0	
											VCOMD	= 1.36V	
		0	0	0	0	0	0	0	0	0	0	0	
Aux-ADC	0		Aux				LE, Aux-Al Divider =				C1,		
ENABLE-8										1	1	1	
CNADLE-8		_	—	—			—				FD mode		
WAKEUP-SEL										1	1	1	
WAREUP-SEL	_	— — —	_	_	_		—	—		Wake-u	p state = I	-D mode	

### Table 6. MAX19710 Default (Power-On) Register Settings

## Table 7. Aux-DAC Enable Table(ENABLE-16 Mode)

E6	E5	E4	Aux-DAC3	Aux-DAC2	Aux-DAC1				
0	0	0	ON	ON	ON				
0	0	1	ON	ON	OFF				
0	1	0	ON	OFF	ON				
0	1	1	ON	OFF	OFF				
1	0	0	OFF	ON	ON				
1	0	1	OFF	ON	OFF				
1	1	0	OFF	OFF	ON				
1	1	1	OFF	OFF	OFF				
0	0	0	Default mode						

## Table 8. Aux-ADC Enable Table(ENABLE-16 Mode)

E9	SELECTION
0 (Default)	Aux-ADC is Powered ON
1	Aux-ADC is Powered OFF

Power consumption in SLOW Tx mode is 21.9mW, and 21.3mW in SLOW Rx mode. Power consumption in FAST Tx mode is 29.1mW, and 28.5mW in FAST Rx mode.

#### FD Mode

The MAX19710 features an FD mode, which is ideal for applications supporting frequency-division duplex. In FD mode, both Rx ADC and Tx DAC, as well as their

### Table 9. Offset Control Bits for ID and QD Channels (IOFFSET or QOFFSET Mode)

BITS I	BITS 105-100 WHEN IN IOFFSET MODE, BITS Q05-Q00 WHEN IN QOFFSET MODE									
IO5/QO5	IO4/QO4	IO3/QO3	IO2/QO2	I01/Q01	100/Q00	(VFS <sub>P-P</sub> / 1023)				
1	1	1	1	1	1	-31 LSB				
1	1	1	1	1	0	-30 LSB				
1	1	1	1	0	1	-29 LSB				
•	•	•	•	•	•	•				
•	•	•	•	•	•	•				
•	•	•	•	•	•	•				
1	0	0	0	1	0	-2 LSB				
1	0	0	0	0	1	-1 LSB				
1	0	0	0	0	0	0mV				
0	0	0	0	0	0	0mV (Default)				
0	0	0	0	0	1	1 LSB				
0	0	0	0	1	0	2 LSB				
•	•	•	•	•	•	•				
•	•	•	•	•	•	•				
•	•	•	•	•	•	•				
0	1	1	1	0	1	29 LSB				
0	1	1	1	1	0	30 LSB				
0	1	1	1	1	1	31 LSB				

Note: 1 LSB = (800mV<sub>P-P</sub> / 1023) = 0.782mV.

## Table 10. Common-Mode Select (COMSEL Mode)

CM1	CM0	Tx PATH OUTPUT COMMON MODE (V)
0	0	1.36 (Default)
0	1	1.20
1	0	1.05
1	1	0.89

respective digital buses, are active and the device can receive and transmit simultaneously. Switching from FD mode to other Rx or Tx modes is fast (0.1 $\mu$ s) since the on-board converters are already powered. Consequently, power consumption in this mode is the maximum of all operating modes. In FD mode the MAX19710 consumes 30mW.

#### Wake-Up Function

The MAX19710 uses the SPI interface to control the operating modes of the device including the shutdown and wake-up functions. Once the device has been placed in shutdown through the appropriate SPI command, the first pulse on CS/WAKE performs a wake-up

### Table 11. WAKEUP-SEL Register

W2	W1	WO	POWER MODE AFTER WAKE-UP (WAKE-UP STATE)
0	0	0	<b>Invalid Value</b> . This value is ignored when inadvertently written to the WAKEUP-SEL register.
0	0	1	IDLE
0	1	0	STBY
0	1	1	SPI1-SLOW Rx
1	0	0	SPI2-SLOW Tx
1	0	1	SPI3-FAST Rx
1	1	0	SPI4-FAST Tx
1	1	1	FD (Default)

function. At the first rising edge of  $\overline{CS}$ /WAKE, the MAX19710 is forced to a preset operating mode determined by the WAKEUP-SEL register. This mode is termed the wake-up state. If the WAKEUP-SEL register has not been programmed, the wake-up state for the MAX19710 is FD mode by default (Tables 6, 11). The WAKEUP-SEL register cannot be programmed with W2



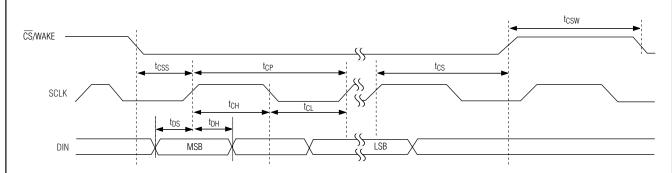


Figure 6. Serial-Interface Timing Diagram

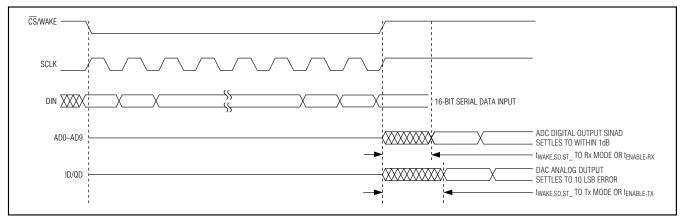


Figure 7. Mode-Recovery Timing Diagram

= 0, W1 = 0, and W0 = 0. If this value is inadvertently written to the device, it is ignored and the register continues to store its previous value. Upon wake-up, the MAX19710 enters the power mode determined by the WAKEUP-SEL register, however, all other settings (Tx DAC offset, Tx DAC common-mode voltage, aux-DAC settings, aux-ADC state) are restored to their values prior to shutdown.

The only SPI line that is monitored by the MAX19710 during shutdown is  $\overline{CS}$ /WAKE. Any information transmitted to the MAX19710 concurrent with the  $\overline{CS}$ /WAKE wake-up pulse is ignored.

#### SPI Timing

The serial digital interface is a standard 3-wire connection CS/WAKE, SCLK, DIN) compatible with SPI/QSPI<sup>TM</sup>/ MICROWIRE/DSP interfaces. Set CS/WAKE low to enable the serial data loading at DIN or output at DOUT. Following a CS/WAKE high-to-low transition, data is shifted synchronously, most significant bit first, on the rising edge of the serial clock (SCLK). After 16 bits are loaded into the serial input register, data is transferred to the latch when CS/WAKE transitions high. CS/WAKE must transition high for a minimum of 80ns before the next write sequence. SCLK can idle either high or low between transitions. Figure 6 shows the detailed timing diagram of the 3-wire serial interface.

#### **Mode-Recovery Timing**

Figure 7 shows the mode-recovery timing diagram. twake is the wake-up time when exiting shutdown, idle, or standby mode and entering Rx, Tx, or FD mode. tENABLE is the recovery time when switching between either Rx or Tx mode. twake or tENABLE is the time for the Rx ADC to settle within 1dB of specified SINAD performance and Tx DAC settling to 10 LSB error. twake and tENABLE times are measured after the 16-bit serial command is latched into the MAX19710 by a CS/WAKE transition high. In FAST mode, the recovery time is 0.1µs to switch between Tx or Rx modes.

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