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Industrial Broadband Power-Line Modem

General Description

The MAX2982 power-line transceiver utilizes state-ofthe-art CMOS design techniques to deliver the highest level of performance, flexibility, and operational temperature range at reduced cost. This highly integrated design combines the Media Access Control (MAC) and the Physical (PHY) layers in a single device. The MAX2982 digital baseband and its companion device, the MAX2981 analog front-end (AFE) with integrated line driver, offer a complete high-speed power-line communication solution fully compliant with HomePlug \otimes 1.0 Powerline Alliance Specification.

The MAX2982 offers reliable broadband communication for industrial environments. The PHY layer is comprised of 84-carrier OFDM modulation engine and Forward Error Correcting (FEC) blocks. The OFDM engine can modulate the signals in one of four modes of operation, namely DBPSK, DQPSK (1/2 rate FEC), DQPSK (3/4 rate FEC) and the ROBO mode. The MAX2982 offers -1dB SNR performance in ROBO mode, a robust mode of operation, to maintain communication over harsh industrial line conditions. Additionally, advanced narrow-band interference rejection circuitry provides immunity from jammer signals.

The MAX2982 offers extensive flexibility by integrating an ARM946E-S™ microprocessor allowing feature enhancement, worldwide regulatory compliance, and improved testability. Optional spectral shaping and notching profiles provide an unparalleled level of flexibility in system design. Additionally, the automatic channel adaptation and interference rejection features of the MAX2982 guarantee outstanding performance. Privacy is provided by a hard-macro DES encryption with key management.

The MAX2982 supports an IEEE® 802.3 standard Media Independent Interface (MII), Reduced Media Independent Interface (RMII), synchronous FIFO supporting a glue-free interface to microcontrollers, and 10/100 Ethernet MAC. These interfaces and standards compliance simplify configuration of monitoring and control networks. Fast response time and an integrated temperature sensor make the MAX2982 an excellent solution for real-time control over power lines. The MAX2982 operates over the -40° C to $+105^{\circ}$ C temperature range and is available in a 128-pin, lead-free, LQFP package.

HomePlug is a registered trademark of HomePlug Powerline Alliance, Inc.

ARM946E-S is a trademark of ARM Limited.

IEEE is a registered service mark of the Institute of Electrical and Electronics Engineers, Inc.

Features

- ♦ Single-Chip Power-Line Networking Transceiver
- ♦ Integrated Temperature Sensor
- ◆ Up to 14Mbps Data Rate
- ◆ Low-Rate Adaptation (LORA) Operation Option Provides -2dB SNR Performance at 1Mbps
- ◆ 4.49MHz to 20.7MHz Frequency Band
- ◆ Flexible MAC/PHY Field Upgradable Firmware using TFTP Spectral Shaping Including Bandwidth and Notching Capability Programmable Preamble 128kB Internal SRAM
- ♦ Advanced Narrowband Interference Rejection **Circuitry**
- ◆ 84-Carrier, OFDM-Based PHY Automatic Channel Adaptation FEC (Forward Error Correction) DQPSK, DBPSK Modulation Enhanced ROBO Mode with -1dB SNR
- ♦ Large Bridge Table: Up to 512 Addresses
- ♦ 56-Bit DES Encryption with Key Management for Secure Communication
- ♦ On-Chip Communication Interfaces **UART** 10/100 Ethernet MII/RMII High-Speed Synchronous FIFO
- ◆ HomePlug 1.0 Compliant
- ◆ AEC-Q100-REV-G Automotive Grade Qualification

Applications

Industrial Automation Motor Control Remote Monitoring and Control Building Automation Broadband Over Shared Coax/Copper Line

Ordering Information appears at end of data sheet.

Typical Application Circuit appears at end of data sheet.

For related parts and recommended products to use with this part, refer to **www.maxim-ic.com/MAX2982.related**.

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For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

Industrial Broadband Power-Line Modem

Typical Application Circuit

Industrial Broadband Power-Line Modem

ABSOLUTE MAXIMUM RATINGS

PACKAGE THERMAL CHARACTERISTICS (Note 1)

Junction-to-Ambient Thermal Resistance (θ JA)30°C/W Junction-to-Case Thermal Resistance (θ JC)........................8°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{DD33} = +3.3V, V_{DD12} = V_{DVDD} = V_{AVDD} = +1.2V, V_{AVSS} = V_{DVSS} = V_{DGND} = 0V, T_A = -40$ to +105°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

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*The parametic values (min, typ, max limits) shown in the Electrical Characteristics table supersede values quoted elsewhere in this data sheet.

Industrial Broadband Power-Line Modem

ELECTRICAL CHARACTERISTICS* (continued)

 $(V_{DD33} = +3.3V, V_{DD12} = V_{DVDD} = V_{AVDD} = +1.2V, V_{AVSS} = V_{DVSS} = V_{DGND} = 0V, T_A = -40$ to $+105^{\circ}$ C, unless otherwise noted. Typical values are at T_A = +25°C.)

AC TIMING CHARACTERISTICS*

 $(V_{DD33} = +3.3V, V_{DD12} = V_{DVDD} = V_{AVDD} = +1.2V, V_{AVSS} = V_{DVSS} = V_{DGND} = 0V, T_A = -40°C$ to +105°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

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Industrial Broadband Power-Line Modem

AC TIMING CHARACTERISTICS* (continued)

(V_{DD33} = +3.3V, V_{DD12} = V_{DVDD} = V_{AVDD} = +1.2V, V_{AVSS} = V_{DVSS} = V_{DGND} = 0V, T_A = -40°C to +105°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

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*The parametic values (min, typ, max limits) shown in the Electrical Characteristics table supersede values quoted elsewhere in this data sheet.

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 $(T_A = +25^{\circ}C,$ unless otherwise noted.)

Typical Operating Characteristics

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Pin Configuration

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Pin Description

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PIN NAME TYPE FUNCTION 33 ETHRXD[1]/
MIIDAT[5] I/O Ethernet MII Receive Data Bit 1/MII/FIFO Mode MII/FIFO Transmit/ Receive Data [5] 34 **ETHRXD[2]/**
MIIDAT[6] I/O Ethernet MII Receive Data Bit 2/MII/FIFO Mode MII/FIFO Transmit/ Receive Data [6] 35 ETHRXD[3]/
MIIDAT[7] I/O Ethernet MII Receive Data Bit 3/MII/FIFO Mode MII/FIFO Transmit/ Receive Data [7] 36 ETHRXER/
BUFCS I Ethernet MII Receive Error/MII/FIFO Mode Active-Low FIFO Chip Select 39 GPIO[16] I/O General-Purpose Input/Output 16. GPIO[16] is in three-state during boot-up. Connect a 100k Ω pullup or pulldown resistor to GPIO[16] if not used. 40 GPIO[17] I/O General-Purpose Input/Output 17. GPIO[17] is in three-state during boot-up. Connect a 100k Ω pullup or pulldown resistor to GPIO[17] if not used. 43 GPIO[0] I/O General-Purpose Input/Output 0. GPIO[0] is in three-state during boot-up. Connect a 100k Ω pullup or pulldown resistor to GPIO[0] if not used. 44 GPIO[1] I/O General-Purpose Input/Output 1. GPIO[1] is in three-state during boot-up. Connect a 100k Ω pullup or pulldown resistor to GPIO[1] if not used. 45 GPIO[2] I/O Reserved 46 GPIO[3] I/O General-Purpose Input/Output 3. GPIO[3] is used for upper layer interface bit 2 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 2k Ω pulldown resistor according to Table 14. 47 GPIO[4] I/O General-Purpose Input/Output 4. GPIO[4] is used for AFE interface serial clock signal (output) and upper layer interface bit 0 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 2k Ω pulldown resistor according to Table 14. 48 GPIO[5] \vert \vert \vert GPIO[5] \vert GPIO[5] is used for AFE interface serial data signal (input/output). Connect a 100k Ω pullup resistor. 51 GPIO[23] I/O General-Purpose Input/Output 23. GPIO[23] is used for the boot pin bit 2 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 1k Ω pulldown resistor according to Table 11. 52 GPIO [6] I/O General-Purpose Input/Output 6. GPIO[6] is used for AFE interface serial write signal (output) and upper layer interface bit 1 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 2k Ω pulldown resistor according to Table 14. 53 GPIO[7] I/O General-Purpose Input/Output 7. GPIO[7] is used for AFE interface power-down signal. Connect a $2k\Omega$ pullup resistor. 56 GPIO[8] I/O General-Purpose Input/Output 8. GPIO[8] is used for nonvolatile memory serial clock signal (output). Connect a $10k\Omega$ pullup resistor to V_{DD33}. 57 GPIO[9] | I/O | General-Purpose Input/Output 9. GPIO[9] is used for serial data in nonvolatile memory interface.

Pin Description (continued)

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Pin Description (continued)

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Pin Description (continued)

Pin Description by Function

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Pin Description by Function (continued)

Industrial Broadband Power-Line Modem

CONTACT $\qquad \qquad | \quad$ NAME \qquad TYPE $\qquad \qquad$ FUNCTION 58 GPIO[10] I/O General-Purpose Input/Output 10. GPIO[10] is used for nonvolatile memory chip select signal (output). Connect a $10k\Omega$ pullup resistor. 59 GPIO[11] I/O General-Purpose Input/Output 11. GPIO[11] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[11]. 60 GPIO[12] | I/O General-Purpose Input/Output 12. GPIO[12] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[12]. 61 GPIO[13] | I/O General-Purpose Input/Output 13. GPIO[13] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[13]. 62 GPIO[14] | I/O General-Purpose Input/Output 14. GPIO[14] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[14]. 65 GPIO[15] | I/O General-Purpose Input/Output 15. GPIO[15] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[15]. 39 GPIO[16] I/O General-Purpose Input/Output 16. GPIO[16] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[16] if not used. 40 GPIO[17] I/O General-Purpose Input/Output 17. GPIO[17] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[17] if not used. 66 GPIO[18] | I/O General-Purpose Input/Output 18. GPIO[18] is in three-state during bootup. Connect a 100k Ω pullup or pulldown resistor to GPIO[18]. 71 GPIO[21] I/O General-Purpose Input/Output 21. GPIO[21] is used for AFE interface collision LED (output) and boot pin bit 0 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 1k Ω pulldown resistor according to Table 11. 72 GPIO[22] I/O General-Purpose Input/Output 22. GPIO[22] is used for AFE interface link status activity LED (output) and boot pin bit 1 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 1k Ω pulldown resistor according to Table 11. 51 GPIO[23] I/O General-Purpose Input/Output 23. GPIO[23] is used for the boot pin bit 2 (input). Connect a 10k Ω pullup resistor to V_{DD33} or a 1k Ω pulldown resistor according to Table 11. SHARED UPPER-LAYER INTERFACE 22 **ETHTXD[0]/**
MIIDAT[0] I/O Ethernet MII Transmit Data Bit 0/MII/FIFO Mode MII/FIFO Transmit/Receive Data [0] 25 **ETHTXD[1]/**
MIIDAT[1] I/O Ethernet MII Transmit Data Bit 1/MII/FIFO Mode MII/FIFO Transmit/Receive Data [1] 26 **ETHTXD**[2]/
MIIDAT[2] I/O | Ethernet MII Transmit Data Bit 2/MII/FIFO Mode MII/FIFO Transmit/Receive Data [2] 27 **ETHTXD[3]/**
MIIDAT[3] I/O | Ethernet MII Transmit Data Bit 3/MII/FIFO Mode MII/FIFO Transmit/Receive Data [3] 32 **ETHRXD**[0]/
MIIDAT[4] IO Ethernet MII Receive Data Bit 0/MII/FIFO Mode MII Transmit/Receive Data [4] 33 **ETHRXD[1]/**
MIIDAT[5] I/O Ethernet MII Receive Data Bit 1/MII/FIFO Mode MII/FIFO Transmit/Receive Data [5] 34 **ETHRXD[2]/**
MIIDAT[6] IO Ethernet MII Receive Data Bit 2/MII/FIFO Mode MII/FIFO Transmit/Receive Data [6]

Pin Description by Function (continued)

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Pin Description by Function (continued)

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Pin Description by Function (continued)

Functional Diagram

MAXIM

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Detailed Description

The MAX2982 power-line transceiver device is a state-ofthe-art CMOS device with high performance and extended operating temperature range to deliver reliable communications in industrial applications. This highly integrated design combines the MAC with the PHY layer in a single device. The MAX2982, with the MAX2981 analog frontend, forms a complete HomePlug 1.0-compliant solution with a substantially reduced system bill of materials.

MII/RMII/FIFO Interface

The MII/RMII/FIFO block is the data and control interface layer of the MAX2982 transceiver. This layer is designed to operate with IEEE 802.3 standard MII/RMII or other devices using the FIFO interface. Refer to the MAX2982 programming reference manual for information on initialization and control of the HomePlug 1.0 MAC through the MII/RMII/FIFO interface. The interface signals connecting to the external host are shown in Figure 1.

The interface is a data channel that transfers data in packets whose flow is controlled by the carrier-sense (MIICRS) signal. The MIICRS signal controls the halfduplex transmission between the external host and the HomePlug MAC. While a frame reception is in progress (MIICRS and MIIRXDV are high), the external host must wait until the completion of reception and the deassertion of MIICRS before starting a transmission. When sending two consecutive frames, the minimum time the external host needs to wait is the one-frame transfer time plus an interframe gap (IFG).

The MII signals MIICOL and MIITXER are not used, as the power-line networking device is able to detect and manage all transmission failures. The signals MIITXCLK and MIIRXCLK have the same source and are referred to as MIICLK in this data sheet.

In MII mode, the data is transferred synchronously with a 2.5MHz/25MHz clock. Data transmission in MII is in nibble format so the data transmission rate is 10Mbps/100Mbps.

In RMII mode, the data is transferred synchronously with a 5/50MHz clock. Data transmission in RMII is in di-bit (two-bit) format so the data transmission rate is 10Mbps/100Mbps.

In FIFO mode, data is read and written in byte format on each positive edge of BUFRDN and BUFWRN. The only limitation in this mode is that BUFRDN and BUFWRN must be low for at least three pulses of MIICLK to be considered a valid signal.

Figure 1. Ethernet MAC and MAX2982 Connection in MII Mode

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The upper layer interface can be selected according to the settings shown in Table 1.

MII Interface Signals

Table 2 describes the signals that provide data, status, and control to and from the MAX2982 in MII mode.

MII MAC and PHY Connections

Figure 1 illustrates the connections between Ethernet/ MAC and MAX2982 in MII mode. Although the TX and RX data paths are full duplex, the MII interface operates in half-duplex mode. MIIRXDV is never asserted at the same time as MIITXEN.

On transmit, the MAX2982 asserts MIICRS some time after MIITXEN is asserted, and drops MIICRS after MIITXEN is deasserted and the MAX2982 is ready to receive another packet. When MIICRS falls, the MAC times out an interframe gap (IFG) and asserts MIITXEN again when there is another packet to send. This differs from nominal behavior of MIICRS in that MIICRS can extend past the end of the packet by an arbitrary amount of time, while the MAX2982 is gaining access to the channel and transmitting the packet.

MACs in 10Mbps mode do not use a jabber timeout, so there is no timing restriction on how long MIICRS can assert other than timeouts (IFG) the MAX2982 implements.

Table 1. Upper Layer Interface Selection GPIO Settings

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Transmissions can "cut through" or begin to be modulated onto the wire as soon as the transfer begins when the MII fills the MAX2982 buffer faster than data needs to be made available to the modulator. When a packet arrives at MAX2982, the device attempts to gain access to the channel. This may not happen before the entire packet is transferred across the MII interface, so the MAX2982 buffers at least one Ethernet packet to perform this rate adaptation.

On receive, when the MAX2982 anticipates a packet to be demodulated, the device raises MIICRS to seize the half-duplex MII channel, waits one interframe gap time (IFG), then defers to MIITXEN when MIITXEN has been asserted plus an IFG. The device raises MIIRXDV to transfer the packet. At the end of the transfer, the MAX2982 drops MIICRS unless the transmit buffer is full or there is another receive packet ready to transfer. Figure 2 illustrates how one receive transfer is followed by a second, when the device defers to MIITXEN. Data reception maintains priority over transmission to ensure that the buffer empties faster than packets arrive off the wire. The longest that the receiver needs to wait is the time to transfer one TX frame plus an IFG or approximately 134µs. However, minimum size frames can arrive at a peak rate of one every 65us, so the receive side buffer must accommodate multiple frames (but only a little more than one Ethernet packet of data).

Transmitting

When a frame in the external host is ready to transmit and MIICRS is not high (the previous transmission has finished), the external host asserts MIITXEN, while data is ready on MIIDAT[3:0]. In response, the MAX2982 asserts MIICRS.While the external host keeps MIITXEN high, data is sampled synchronously with respect to MIICLK into the MAX2982 through MIIDAT. After transmission of the last byte of data and before the next positive edge of the MIICLK, MIITXEN is reset by the external host.

The transmission timing of the MII interface is illustrated in Figure 3, with details in Figure 4 and Table 3.

Figure 3. Transmission Behavior of the MII Interface

Figure 4. MII Interface Detailed Transmit Timing

Table 3. MII Interface Detailed Transmit Timing

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Receiving

When a frame is ready to send from the MAX2982 to the external host, the MAX2982 asserts MIIRXDV after IFG, while there is no transmission session in progress with respect to MIICRS.

Note: The receive process cannot start while a transmission is in progress.

While the MAX2982 keeps MIIRXDV high, data is sampled synchronously with respect to MIICLK from MAX2982 through MIIDAT. After the last byte of data is received, the MAX2982 resets MIIRXDV.

Receive timing of the MII interface is illustrated in Figure 5, with details in Figure 6 and Table 4.

Reduced Media Independent Interface (RMII)

Table 5 describes the signals that provide data, status, and control to the MAX2982 in RMII mode. In this mode, data is transmitted and received in bit pairs. The RMII mode connections are shown in Figure 7.

In case of an error in the received data, to eliminate the requirement for MIIRXER and still meet the requirement for undetected error rate, MIIDAT[5:4] replaces the decoded data in the receive stream with "10" until the end of carrier activity. By this replacement, the CRC check is guaranteed to reject the packet as being in error.

RMII Signal Timing

RMII transmit and receive timing are the same as for MII, except that the data are sent and received in di-bit format and MIICRS is removed.

Figure 5. Receive Behavior of the MII Interface

Figure 6. MII Interface Detailed Receive Timing

Table 5. RMII Signal Description

Table 4. MII Interface Detailed Receive Timing

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Table 5. RMII Signal Description (continued)

Figure 7. MAC-PHY Connection in RMII Mode

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packet to send.
Figure 8. External Host and MAX2982 Connection in FIFO mode packet to send.

Table 6. FIFO Signal Description

FIFO Interface Signals

The buffering FIFO interface supports synchronous operation and can be interfaced gluelessly to an external microprocessor memory bus. The interface is clocked by the external processor on the MIICLK pin.

The read and write pulse width is three MIICLK cycles.

The signals that provide data, status, and control to and from the MAX2982 are shown in Table 6. MIIRXDV should never be asserted at the same time as MIITXEN`, but the device is able to start transmission while receive is in progress. The MAX2982 gives higher priority to TX packets from the external host to avoid data loss.

On transmit, the MAX2982 asserts MIICRS after MIITXEN is asserted by the host. The host should not assert MIITXEN if MIICRS is already high. After MIITXEN is deasserted by the host, which means that the host has completed data transmission, MIICRS goes low when the MAX2982 is ready to receive another packet. When MIICRS falls, MIITXEN can be held low if there is another

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Transmissions can "cut through" or begin to be modulated onto the wire as soon as the transfer begins, as the interface fills the MAX2982 buffer faster than data needs to be made available to the modulator. When a packet arrives at the MAX2982, the device attempts to gain access to the channel. Since this may not happen before the entire packet is transferred across the interface, the MAX2982 FIFO features a 2Kbyte TX buffer to hold packets to perform this rate adaptation.

On receive, when the MAX2982 anticipates a packet to be demodulated, the device raises MIIRXDV to identify the upper layer that a packet is ready to transmit. MIIRXDV drops when the last byte is transmitted. Receive direction transfers maintain priority over the transmit direction to ensure that the buffer empties faster than packets arrive off the wire. The longest that the receiver needs to wait is the time to transfer one TX frame plus an IFG.

Transmitting

When the external host is ready to transmit a frame and MIICRS is not high (the previous transmission is finished), it asserts MIITXEN. The external host must assert MIITXEN if MIIRXDV is not high to avoid data loss. In response, the MAX2982 asserts MIICRS. While the external host keeps MIITXEN high, one byte of data is transmitted into the MAX2982 through MIIDAT_IN for each positive edge of BUFWR. After transmission of the last byte of data, the external host resets MIITXEN. Figure 9 shows the interactions between the external host and the MAX2982. There are two GPIOs indicating packet loss and completion of a packet transmission controlled by software. The host can use these signals to determine packet retransmission much faster than through TCP or a packet-based scheme. The BUFWR clock rate is 16MHz maximum at MIICLK of 66MHz.

Figure 10 shows the overall transmission timing of the FIFO interface.

Figure 9. Buffering FIFO Transmission Process from External Host

Figure 10. Transmission Timing of the Buffering (FIFO) Interface

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Figure 11. Buffering FIFO Interface Receive Process from the External Host View

Receiving

When the MAX2982 is ready to send a frame to the external host, the MAX2982 asserts MIIRXDV after an IFG when there is no transmission session in progress with respect to MIICRS. A receive process cannot start while a transmission is under progress. The FIFO features a 2Kb RX buffer to store received packets.

While the MAX2982 keeps MIIRXDV high, the device sends one byte of data on MIIDAT_OUT for each positive edge on BUFRD. The first two bytes represent the frame length in MSB first format. After the last byte of data is received, the MAX2982 resets MIIRXDV. The direction of bidirectional data I/Os is controlled through BUFCS and BUFRD. The MAX2982 enables data output drivers when $BUFCSN = 0$ and $BUFRDN = 0$. Figure 11 shows the interactions between the external host and the MAX2982.

Figure 12. Receive Timing of Buffering (FIFO) Interface

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FIFO Read/Write Timing

The FIFO interface is connected to an external data bus in half-duplex mode with independent buffers for TX and RX and MIICLK provided with external processor controls BUFRD and BUFWR timing as shown in Figures 13 and 14.

Figure 13. MAX2982 FIFO Read Timing Diagram

1) Minimum CLK frequency is 2.5MHz and maximum is 66MHz.

- 2) MIIDATA_OUT is valid maximum 10ns after the positive edge of T1. This means that worst case for t_{OV} = clock period - 10ns for pulse width of 2ns. $t_{\rm OV}$ = 2 x Clock Period -10ns for pulse width of 3ns.
- 3) MIIDATA OUT is three-stated maximum 12ns and minimum 5ns after the positive edge of BUFRD or BUFCS whichever is earlier, which is tOH.
- 4) MIIDATA_OUT is driven low-Z minimum 0ns after the negative edge of BUFRD.
- 5) CLK duty cycle is 40% to 60%.

Figure 14. MAX2982 FIFO Write Timing Diagram

- 1) MIIDATA_IN minimum setup time is 3ns at the positive edge of T2.
- 2) BUFWR and MIIDAT minimum hold time is 2ns at the positive edge of T2.
- 3) BUFWR pulse width is 3 clock cycles long.
- 4) Minimum CLK frequency is 2.5MHz and maximum is 66MHz.
- 5) CLK duty cycle is 40% to 60%.

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A typical interface between the MAX2982 and a microcontroller at a 66MHz clock rate is shown in Figure 15 with the following setting. \overline{WR} and \overline{RD} signals manage data transfer to/from the FIFO port through BUFWR and BUFRD. WR and RD are asserted low for three clock cycles and data is valid for at least 3ns.

Figure 15. Typical Interface Between a Microcontroller and MAX2982

Microcontroller settings:

CLKOUT Max 66MHz.

 $\overline{\text{RD}}$ and $\overline{\text{WR}}$ access phase set to 3 CLKOUT cycles.

GPIO[1]: When the MAX2982 is ready to send a frame to the microcontroller, the MAX2982 asserts MIIRXDV.

GPIO[2]: When the external host is ready to transmit a frame and MIICRS is not high (the previous transmission is finished), the microcontroller asserts MIITXEN. The external host must assert MIITXEN if MIIRXDV is not high to avoid data loss.

GPIO[3]: Upon assertion of MIITXEN, the MAX2982 asserts MIICRS.

GPIO[4]: When MIIRXER is asserted high, indicates to the microcontroller that an error has occurred during the frame reception.

Management Data Unit (MDU)

The MIIMDIO is a bidirectional data in/output for the Management Data Interface. The MIIMDC signal is a clock reference for the MIIMDIO signal. Figure 16 illustrates the write behavior of the MDU. Figure 17 illustrates the read behavior of the MDU.

Figure 16. Write Behavior of the Management Data Unit

Figure 17. Read Behavior of Management Data Unit

