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Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

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Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China



# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### General Description

The MAX44248 is an ultra-precision, low-noise, zero-drift dual operational amplifier featuring very low-power operation with a wide supply range. The device incorporates a patented auto-zero circuit that constantly measures and compensates the input offset to eliminate drift over time and temperature as well as the effect of 1/f noise. The device also features integrated EMI filters to reduce high-frequency signal demodulation on the output. The op amp operates from either a single 2.7V to 36V supply or dual  $\pm 1.35$ V to  $\pm 18$ V supply. The device is unity-gain stable with a 1MHz gain-bandwidth product and a low 90 $\mu$ A supply current per amplifier.

The low offset and noise specifications and high supply range make the device ideal for sensor interfaces and transmitters.

The device is available in 8-pin  $\mu$ MAX® and SO packages and is specified over the  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  automotive operating temperature range.

### Applications

Sensors Interfaces  
 4-20mA and 0 to 10V Transmitters  
 PLC Analog I/O Modules  
 Weight Scales  
 Portable Medical Devices

### Features

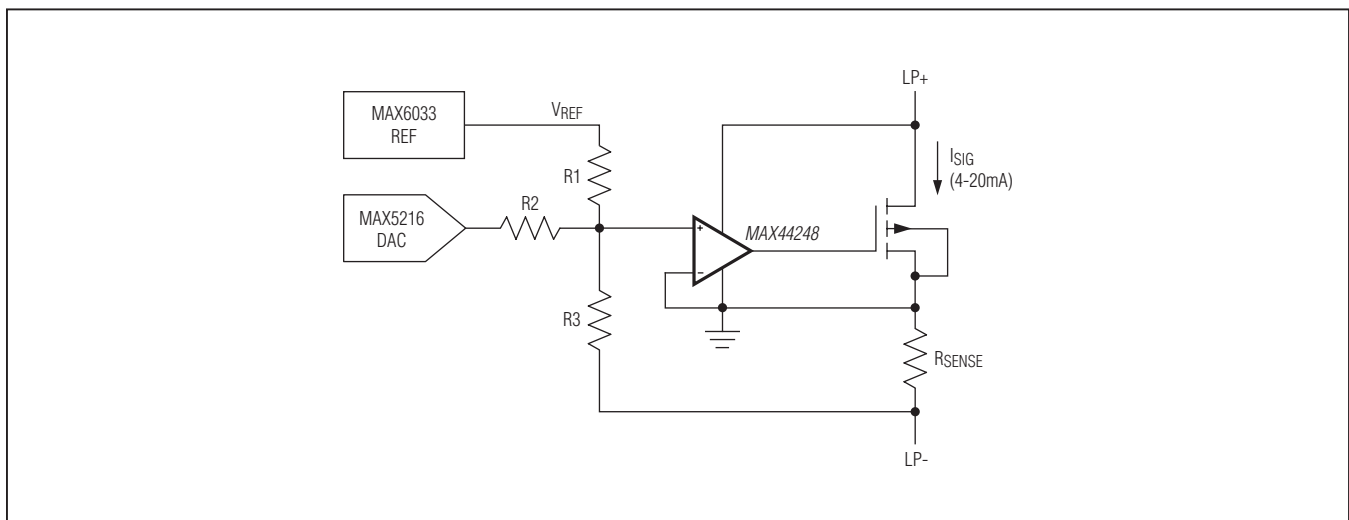
- ◆ Very Low Input Voltage Offset 7.5 $\mu$ V (max)
- ◆ Low 30nV/ $^{\circ}\text{C}$  Offset Drift (max)
- ◆ Low 90 $\mu$ A Quiescent Current per Amplifier
- ◆ Low Input Noise  
     50nV/ $\sqrt{\text{Hz}}$  at 1kHz  
     0.5 $\mu$ V<sub>p-p</sub> from 0.1Hz to 10Hz
- ◆ 1MHz Gain-Bandwidth Product
- ◆ EMI Suppression Circuitry
- ◆ Rail-to-Rail Output
- ◆ 2.7V to 36V Supply Range
- ◆ 8-Pin  $\mu$ MAX and SO package

[Ordering Information](#) appears at end of data sheet.

For related parts and recommended products to use with this part, refer to [www.maximintegrated.com/MAX44248.related](http://www.maximintegrated.com/MAX44248.related).

$\mu$ MAX is a registered trademark of Maxim Integrated Products, Inc.

### Typical Operating Circuit



For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at [www.maximintegrated.com](http://www.maximintegrated.com).

## 36V, Precision, Low-Power, 90µA, Dual Op Amp

### ABSOLUTE MAXIMUM RATINGS

$V_{DD}$ to $V_{SS}$ .....	-0.3V to +40V	Operating Temperature Range .....	-40°C to +125°C
Common-Mode Input Voltage .....	( $V_{SS} - 0.3V$ ) to ( $V_{DD} + 0.3V$ )	Storage Temperature .....	-65°C to +150°C
Differential Input Voltage $IN_+$ , $IN_-$ .....	6V	Junction Temperature .....	+150°C
Continuous Input Current Into Any Pin .....	±20mA	Lead Temperature (soldering, 10s) .....	+300°C
Output Voltage to $V_{SS}$ ( $OUT_-$ ) .....	- 0.3V to ( $V_{DD} + 0.3V$ )	Soldering Temperature (reflow) .....	+260°C
Output Short-Circuit Duration ( $OUT_-$ ) .....	1s		

### PACKAGE THERMAL CHARACTERISTICS (Note 1)

$\mu$ MAX	SO		
Junction-to-Ambient Thermal Resistance ( $\theta_{JA}$ ) .....	206.3°C/W	Junction-to-Ambient Thermal Resistance ( $\theta_{JA}$ ) .....	132°C/W
Junction-to-Case Thermal Resistance ( $\theta_{JC}$ ) .....	42°C/W	Junction-to-Case Thermal Resistance ( $\theta_{JC}$ ) .....	38°C/W

**Note 1:** Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to [www.maximintegrated.com/thermal-tutorial](http://www.maximintegrated.com/thermal-tutorial).

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### ELECTRICAL CHARACTERISTICS

( $V_{DD} = 10V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ ,  $T_A = -40^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Typical values are at  $+25^\circ C$ .) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>POWER SUPPLY</b>						
Supply Voltage Range	$V_{DD}$	Guaranteed by PSRR	2.7		36	V
Power-Supply Rejection Ratio (Note 3)	PSRR	$T_A = +25^\circ C$ , $V_{IN+} = V_{IN-} = V_{DD}/2 - 1V$	140	148		dB
		$-40^\circ C < T_A < +125^\circ C$	133			
Quiescent Current per Amplifier	$I_{DD}$	$T_A = +25^\circ C$		90	120	$\mu A$
		$-40^\circ C < T_A < +125^\circ C$			130	
<b>DC SPECIFICATIONS</b>						
Input Common-Mode Range	$V_{CM}$	Guaranteed by CMRR test	$V_{SS} - 0.05$		$V_{DD} - 1.5$	V
Common-Mode Rejection Ratio (Note 3)	CMRR	$T_A = +25^\circ C$ , $V_{CM} = V_{SS} - 0.05V$ to $V_{DD} - 1.5V$	126	130		dB
		$-40^\circ C < T_A < +125^\circ C$ , $V_{CM} = V_{SS} - 0.05V$ to $V_{DD} - 1.5V$	120			
Input Offset Voltage (Note 3)	$V_{OS}$	$T_A = +25^\circ C$		2	7.5	$\mu V$
		$-40^\circ C < T_A < +125^\circ C$			10	
Input Offset Voltage Drift (Note 3)	$TC V_{OS}$			10	30	nV/°C
Input Bias Current (Note 3)	$I_B$	$T_A = +25^\circ C$		150	300	$\mu A$
		$-40^\circ C < T_A < +125^\circ C$			700	
Input Offset Current (Note 3)	$I_{OS}$	$T_A = +25^\circ C$		300	600	$\mu A$
		$-40^\circ C < T_A < +125^\circ C$			1400	



# MAX44248

## 36V, Precision, Low-Power, 90µA, Dual Op Amp

### ELECTRICAL CHARACTERISTICS (continued)

( $V_{DD} = 10V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ ,  $T_A = -40^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Typical values are at  $+25^\circ C$ .) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Open-Loop Gain (Note 3)	$A_{VOL}$	$V_{SS} + 0.5V \leq V_{OUT} \leq V_{DD} - 0.5V$	$T_A = +25^\circ C$	140	150		dB
			$-40^\circ C < T_A < +125^\circ C$	135			
Output Short-Circuit Current		To $V_{DD}$ or $V_{SS}$ , noncontinuous			40		mA
Output Voltage Swing	$V_{DD} - V_{OUT}$		$T_A = +25^\circ C$			80	mV
			$-40^\circ C < T_A < +125^\circ C$			110	
	$V_{OUT} - V_{SS}$		$T_A = +25^\circ C$			50	
			$-40^\circ C < T_A < +125^\circ C$			75	
<b>AC SPECIFICATIONS</b>							
Input Voltage-Noise Density	$e_N$	$f = 1kHz$			50		$nV/\sqrt{Hz}$
Input Voltage Noise		$0.1Hz < f < 10Hz$			500		$nV_{p-p}$
Input Current-Noise Density	$i_N$	$f = 1kHz$			0.1		$pA/\sqrt{Hz}$
Gain-Bandwidth Product	GBW				1		MHz
Slew Rate	SR	$A_V = 1V/V$ , $V_{OUT} = 2V_{p-p}$			0.7		$V/\mu s$
Capacitive Loading	$C_L$	No sustained oscillation, $A_V = 1V/V$			400		pF
Total Harmonic Distortion Plus Noise	THD+N	$V_{OUT} = 2V_{p-p}$ , $A_V = +1V/V$ , $f = 1kHz$			-100		dB
EMI Rejection Ratio	EMIRR	$V_{RF\_PEAK} = 100mV$	$f = 400MHz$		75		dB
			$f = 900MHz$		78		
			$f = 1800MHz$		80		
			$f = 2400MHz$		90		

### ELECTRICAL CHARACTERISTICS

( $V_{DD} = 30V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ ,  $T_A = -40^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Typical values are at  $+25^\circ C$ .) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
<b>POWER SUPPLY</b>							
Quiescent Current per Amplifier	$I_{DD}$	$T_A = +25^\circ C$			90	120	$\mu A$
		$-40^\circ C < T_A < +125^\circ C$				130	
<b>DC SPECIFICATIONS</b>							
Input Common-Mode Range	$V_{CM}$	Guaranteed by CMRR test		$V_{SS} - 0.05$		$V_{DD} - 1.5$	V
Common-Mode Rejection Ratio (Note 3)	CMRR	$T_A = +25^\circ C$ , $V_{CM} = V_{SS} - 0.05V$ to $V_{DD} - 1.5V$		130	140		dB
		$-40^\circ C < T_A < +125^\circ C$ , $V_{CM} = V_{SS} - 0.05V$ to $V_{DD} - 1.5V$		126			

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## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### ELECTRICAL CHARACTERISTICS (continued)

( $V_{DD} = 30V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ ,  $T_A = -40^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Typical values are at  $+25^\circ C$ .) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Input Offset Voltage (Note 3)	$V_{OS}$	$T_A = +25^\circ C$			2	7.5	$\mu V$
		$-40^\circ C < T_A < +125^\circ C$				10	
Input Offset Voltage Drift (Note 3)	TC $V_{OS}$				10	30	nV/ $^\circ C$
Input Bias Current (Note 3)	$I_B$	$T_A = +25^\circ C$			150	300	$\mu A$
		$-40^\circ C < T_A < +125^\circ C$				700	
Input Offset Current (Note 3)	$I_{OS}$	$T_A = +25^\circ C$			300	600	$\mu A$
		$-40^\circ C < T_A < +125^\circ C$				1400	
Open-Loop Gain (Note 3)	$A_{VOL}$	$V_{SS} + 0.5V \leq V_{OUT} \leq V_{DD} - 0.5V$	$T_A = +25^\circ C$	146	150		dB
			$-40^\circ C < T_A < +125^\circ C$	140			
Output Short-Circuit Current		To $V_{DD}$ or $V_{SS}$ , noncontinuous			40		mA
Output Voltage Swing	$V_{DD} - V_{OUT}$	$T_A = +25^\circ C$				200	mV
		$-40^\circ C < T_A < +125^\circ C$				270	
	$V_{OUT} - V_{SS}$	$T_A = +25^\circ C$				140	
		$-40^\circ C < T_A < +125^\circ C$				220	
<b>AC SPECIFICATIONS</b>							
Input Voltage-Noise Density	$e_N$	f = 1kHz			50		nV/ $\sqrt{Hz}$
Input Voltage Noise		0.1Hz < f < 10Hz			500		nV <sub>p-p</sub>
Input Current-Noise Density	$i_N$	f = 1kHz			0.1		pA/ $\sqrt{Hz}$
Gain-Bandwidth Product	GBW				1		MHz
Slew Rate	SR	$A_V = 1V/V$ , $V_{OUT} = 2V_{P-P}$			0.7		V/ $\mu s$
Capacitive Loading	$C_L$	No sustained oscillation, $A_V = 1V/V$			400		pF
Total Harmonic Distortion Plus Noise	THD+N	$V_{OUT} = 2V_{P-P}$ , $A_V = +1V/V$ , f = 1kHz			-100		dB
EMI Rejection Ratio	EMIRR	$V_{RF\_PEAK} = 100mV$	f = 400MHz		75		dB
			f = 900MHz		78		
			f = 1800MHz		80		
			f = 2400MHz		90		

**Note 2:** All devices are 100% production tested at  $T_A = +25^\circ C$ . Temperature limits are guaranteed by design.

**Note 3:** Guaranteed by design.

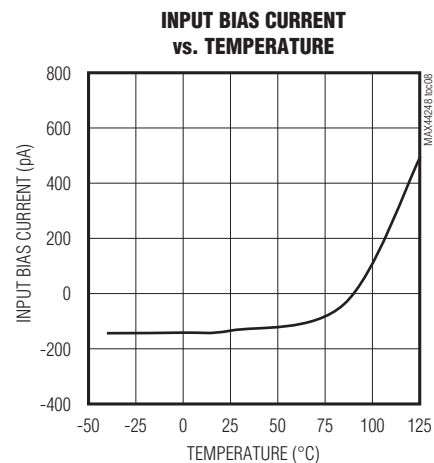
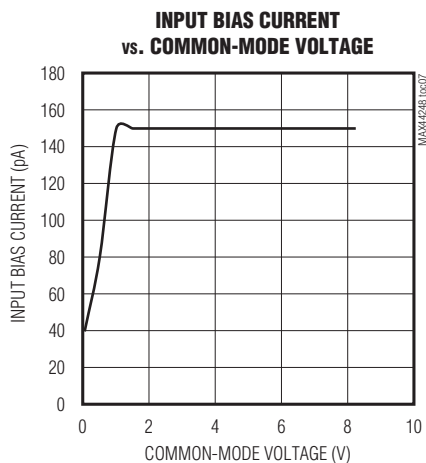
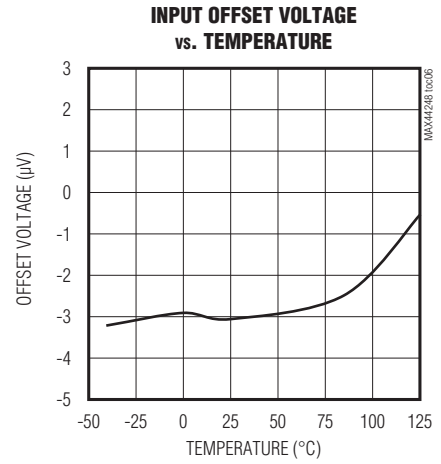
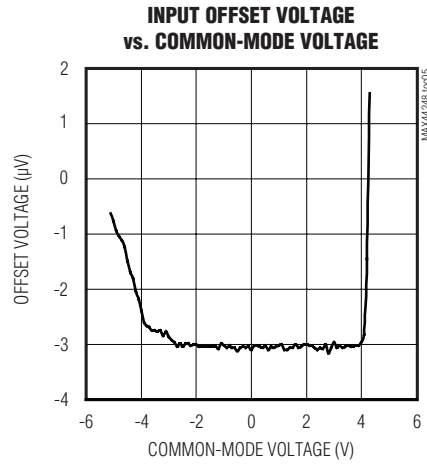
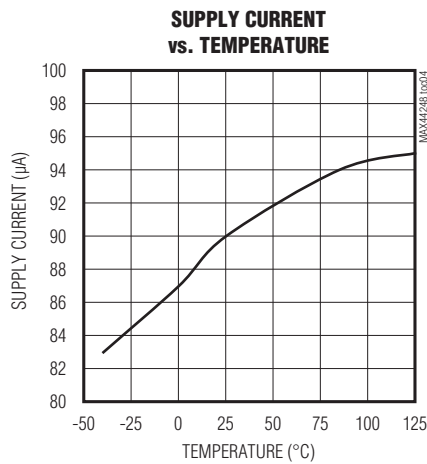
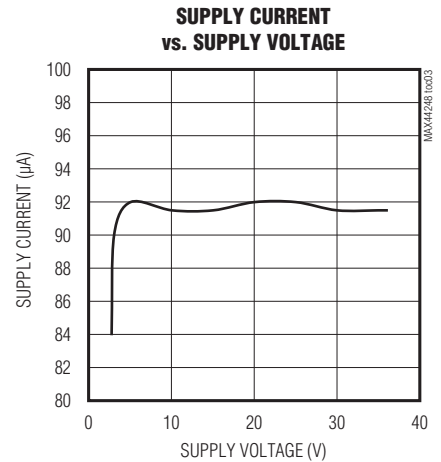
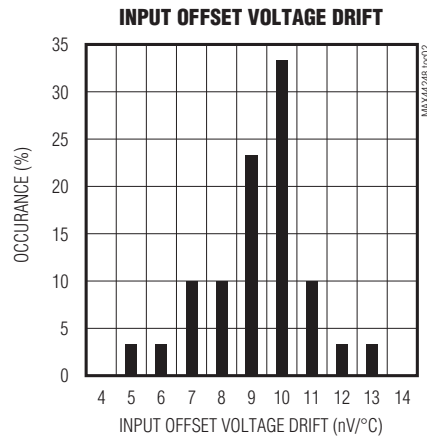
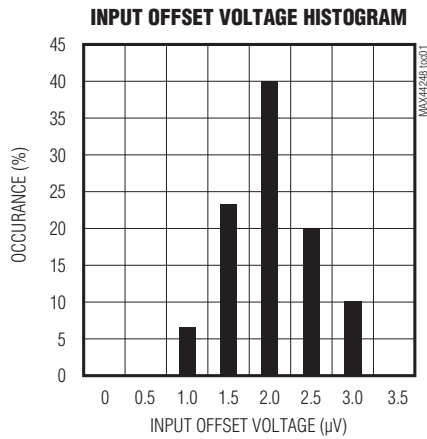
**Note 4:** At  $IN+$  and  $IN-$ . Defined as  $20\log(V_{RF\_PEAK}/\Delta V_{OS})$ .

# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### Typical Operating Characteristics

( $V_{DD} = 10V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ . Typical values are at  $T_A = +25^\circ C$ .)



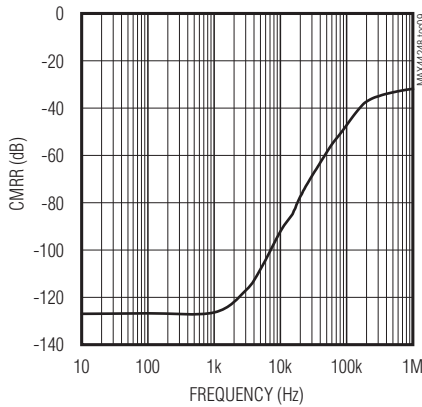
# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

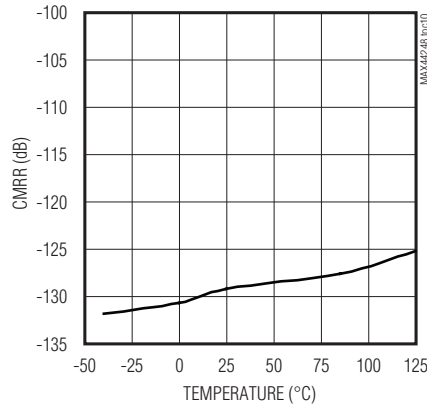
### Typical Operating Characteristics (continued)

( $V_{DD} = 10V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ . Typical values are at  $T_A = +25^\circ C$ .)

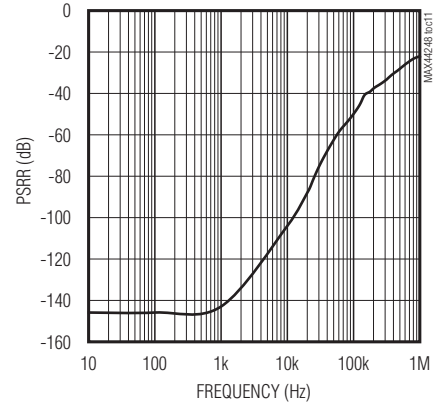
**COMMON-MODE REJECTION RATIO vs. FREQUENCY**



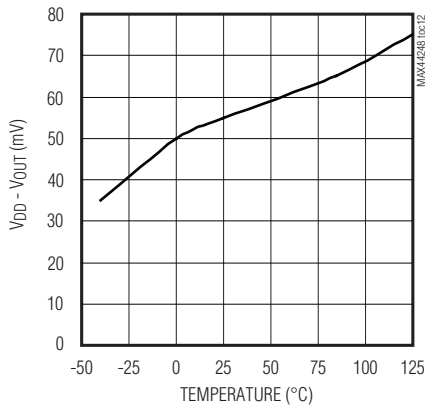
**COMMON-MODE REJECTION RATIO vs. TEMPERATURE**



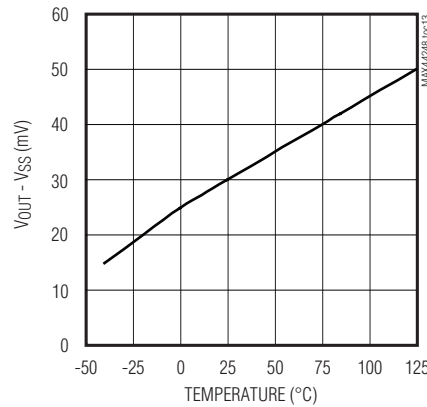
**POWER-SUPPLY REJECTION RATIO vs. FREQUENCY**



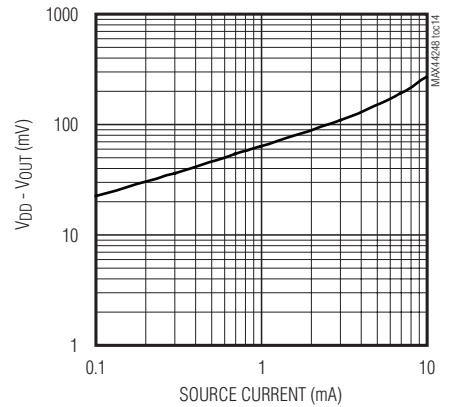
**OUTPUT VOLTAGE HIGH vs. TEMPERATURE**



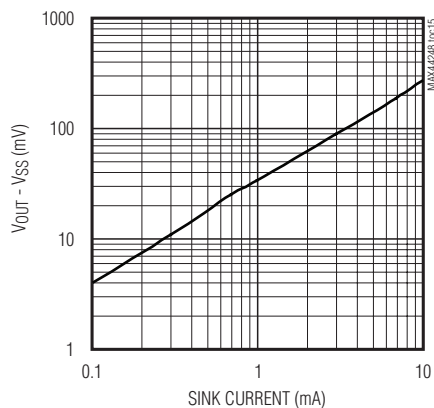
**OUTPUT VOLTAGE LOW vs. TEMPERATURE**



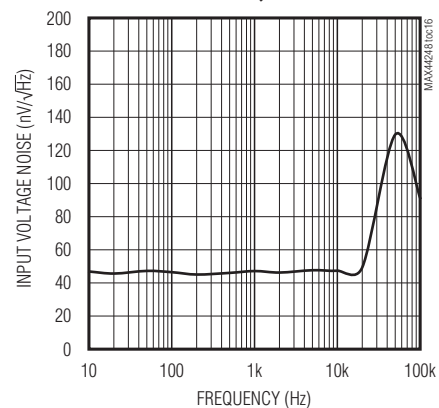
**OUTPUT VOLTAGE HIGH vs. SOURCE CURRENT**



**OUTPUT VOLTAGE LOW vs. SINK CURRENT**



**INPUT VOLTAGE NOISE vs. FREQUENCY**



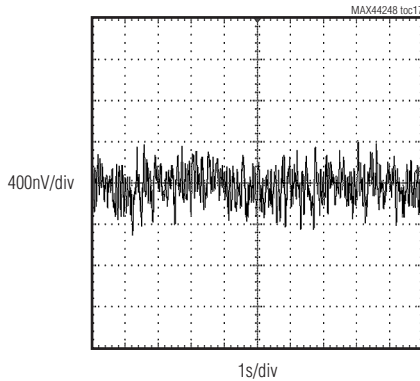
# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

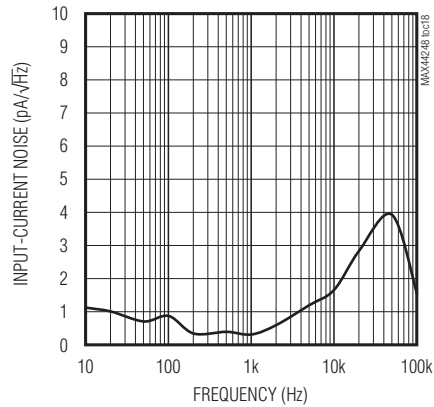
### Typical Operating Characteristics (continued)

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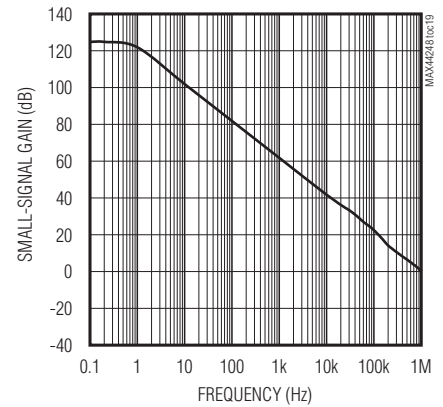
INPUT VOLTAGE 0.1Hz TO 10Hz NOISE



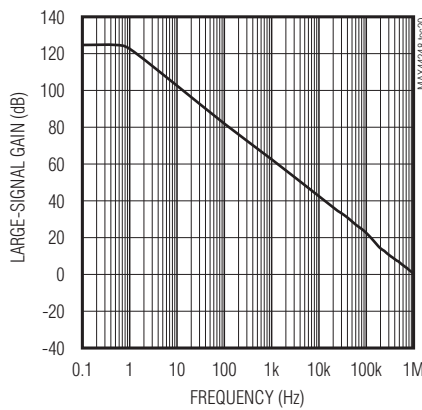
INPUT CURRENT NOISE vs. FREQUENCY



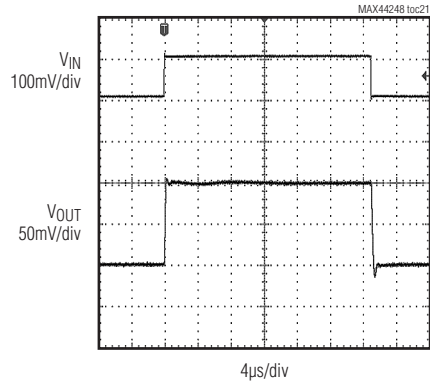
SMALL-SIGNAL GAIN vs. FREQUENCY



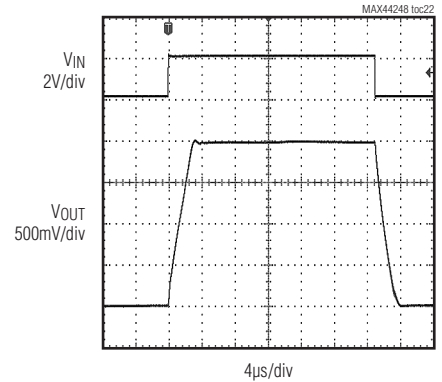
LARGE-SIGNAL GAIN vs. FREQUENCY



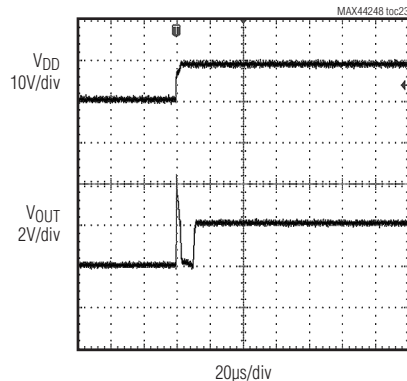
SMALL-SIGNAL STEP RESPONSE



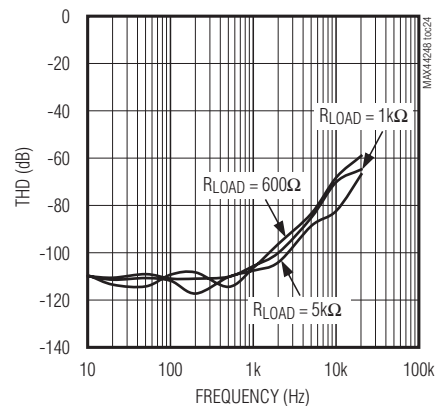
LARGE-SIGNAL STEP RESPONSE



POWER-UP TIME



TOTAL HARMONIC DISTORTION vs. FREQUENCY



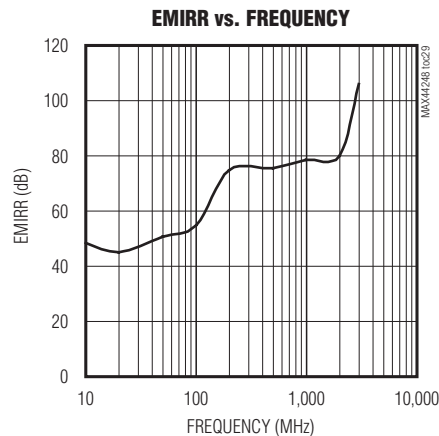
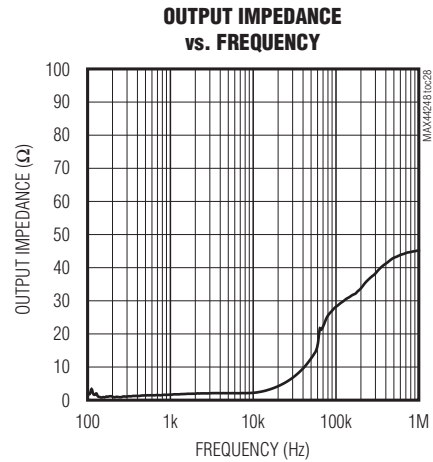
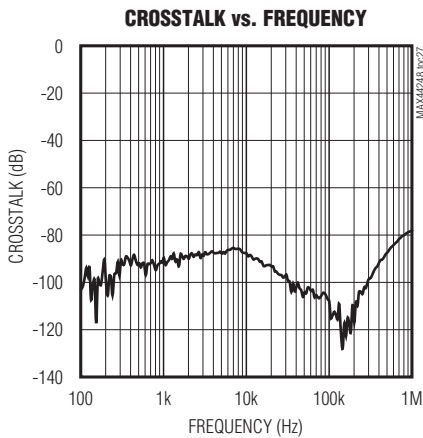
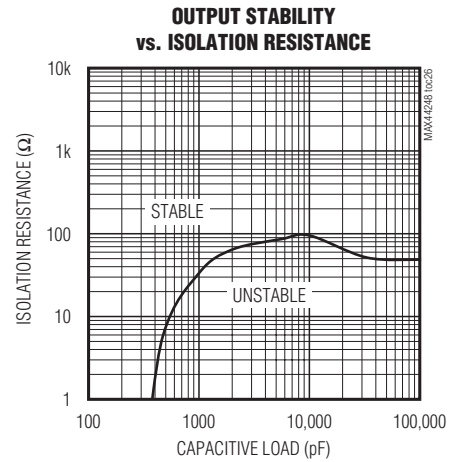
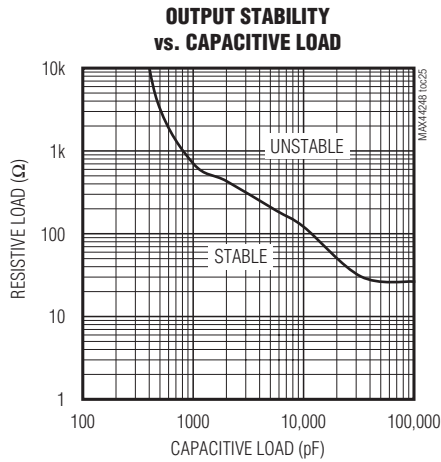


# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### Typical Operating Characteristics (continued)

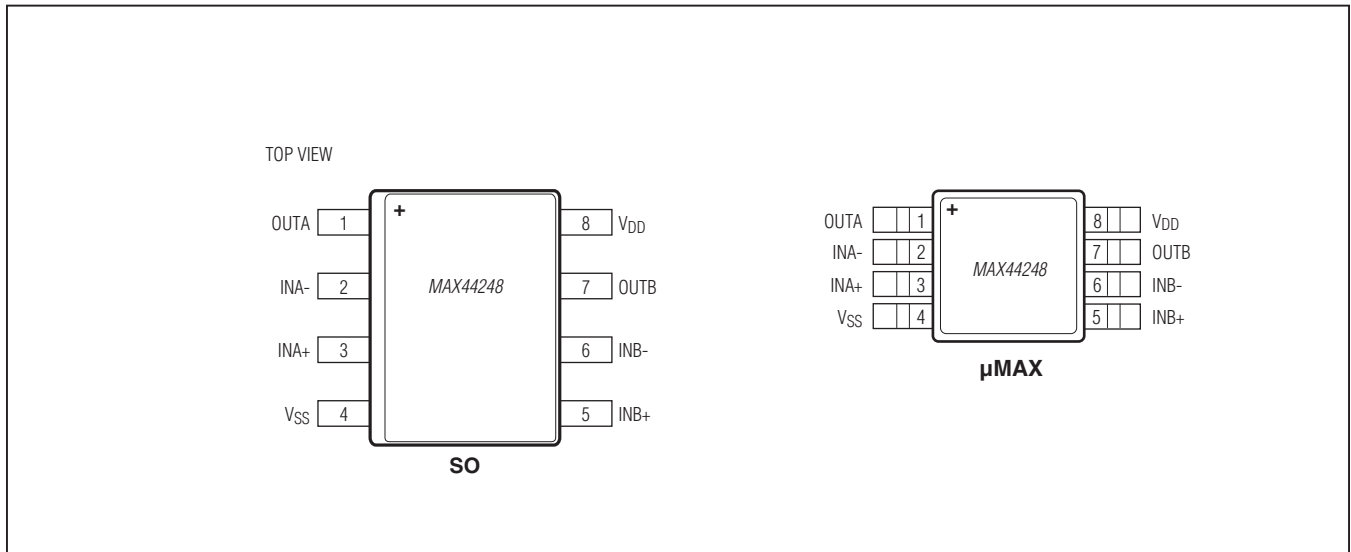
( $V_{DD} = 10V$ ,  $V_{SS} = 0V$ ,  $V_{IN+} = V_{IN-} = V_{DD}/2$ ,  $R_L = 5k\Omega$  to  $V_{DD}/2$ . Typical values are at  $T_A = +25^\circ C$ .)



# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### Pin Configurations



### Pin Description

PIN		NAME	FUNCTION
SO	μMAX		
1	1	OUTA	Channel A Output
2	2	INA-	Channel A Negative Input
3	3	INA+	Channel A Positive Input
4	4	VSS	Negative Supply Voltage
5	5	INB+	Channel B Positive Input
6	6	INB-	Channel B Negative Input
7	7	OUTB	Channel B Output
8	8	VDD	Positive Supply Voltage

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### ***Detailed Description***

The MAX44248 is a high-precision amplifier that has a less than 2 $\mu$ V (typ) input-referred offset and low input voltage-noise density at 10Hz. 1/f noise, in fact, is eliminated to improve the performance in low-frequency applications. These characteristics are achieved through an auto-zeroing technique that cancels the input offset voltage and 1/f noise of the amplifier.

#### ***External Noise Suppression in EMI Form***

The device has input EMI filters to prevent effects of radio frequency interference on the output. The EMI filters comprise passive devices that present significant higher impedance to higher frequency signals. See the EMIRR vs. Frequency graph in the *Typical Operating Characteristics* section for details.

#### ***High Supply Voltage Range***

The device features 90 $\mu$ A current consumption per channel and a voltage supply range from either 2.7V to 36V single supply or  $\pm$ 1.35V to  $\pm$ 18V split supply.

### ***Applications Information***

The device is an ultra-high precision operational amplifier with a high supply voltage range designed for load cell, medical instrumentation, and precision instrument applications.

#### ***4–20mA Current-Loop Communication***

Industrial environments typically have a large amount of broadcast electromagnetic interference (EMI) from high-voltage transients and switching motors. This combined with long cables for sensor communication leads to high-voltage noise on communication lines. Current-Loop communication is resistant to this noise because the EMI induced current is low. This configuration also allows for

low-power sensor applications to be powered from the communication lines.

The *Typical Operating Circuit* shows how the device can be used to make a current loop driver.

The circuit uses low-power components such as the MAX44248 op amp, the 16-bit MAX5216 DAC, and the high-precision 60 $\mu$ A-only MAX6033 reference. In this circuit, both the DAC and the reference are referred to the local ground. The MAX44248 op-amp inputs are capable of swinging to the negative supply (which is the local ground in this case). R3 acts as a current mirror with R<sub>SENSE</sub>. Therefore, if R<sub>SENSE</sub> = 50 $\Omega$  (i.e. 20mA will drop 1V) and if the current through R3 is 10 $\mu$ A when I<sub>OUT</sub> is 20mA (0.05% error) then R3 = 100k $\Omega$ . R1 is chosen along with the reference voltage to provide the 4mA offset. R2 = 512k $\Omega$  for 20mA full scale or R2 = 614k $\Omega$  for 20% over-range. R<sub>SENSE</sub> is ratiometric with R3, R1 independently sets the offset current and R2 independently sets the DAC scaling.

#### ***Driving High-Performance ADCs***

The MAX44248's low input offset voltage and low noise make this amplifier ideal for ADC buffering. Weight scale applications require a low-noise, precision amplifier in front of an ADC. [Figure 1](#) details an example of a load cell and amplifier driven from the same 5V supply, along with a 16-bit delta sigma ADC such as the MAX11205.

The MAX11205 is an ultra-low-power (< 300 $\mu$ A, max active current), high-resolution, serial output ADC. It provides the highest resolution per unit power in the industry and is optimized for applications that require very high dynamic range with low power such as sensors on a 4–20mA industrial control loop. The device provides a high-accuracy internal oscillator that requires no external components.

# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### Layout Guidelines

The MAX44248 features ultra-low voltage and noise. Therefore, to get optimum performance follow the layout guidelines.

Avoid temperature gradients at the junction of two dissimilar metals. The most common dissimilar metals used on a PCB are solder-to-component lead and solder-to-board trace. Dissimilar metals create a local thermocouple. A variation in temperature across the board can cause an additional offset due to Seebeck effect at the solder junctions. To minimize the Seebeck effect, place the amplifier away from potential heat

sources on the board, if possible. Orient the resistors such that both the ends are heated equally. It is a good practice to match the input signal path to ensure that the type and number of thermoelectric junctions remain the same. For example, consider using dummy  $0\Omega$  resistors oriented in such a way that the thermoelectric source, due to the real resistors in the signal path, are cancelled. It is recommended to flood the PCB with ground plane. The ground plane ensures that heat is distributed uniformly reducing the potential offset voltage degradation due to Seebeck effect.

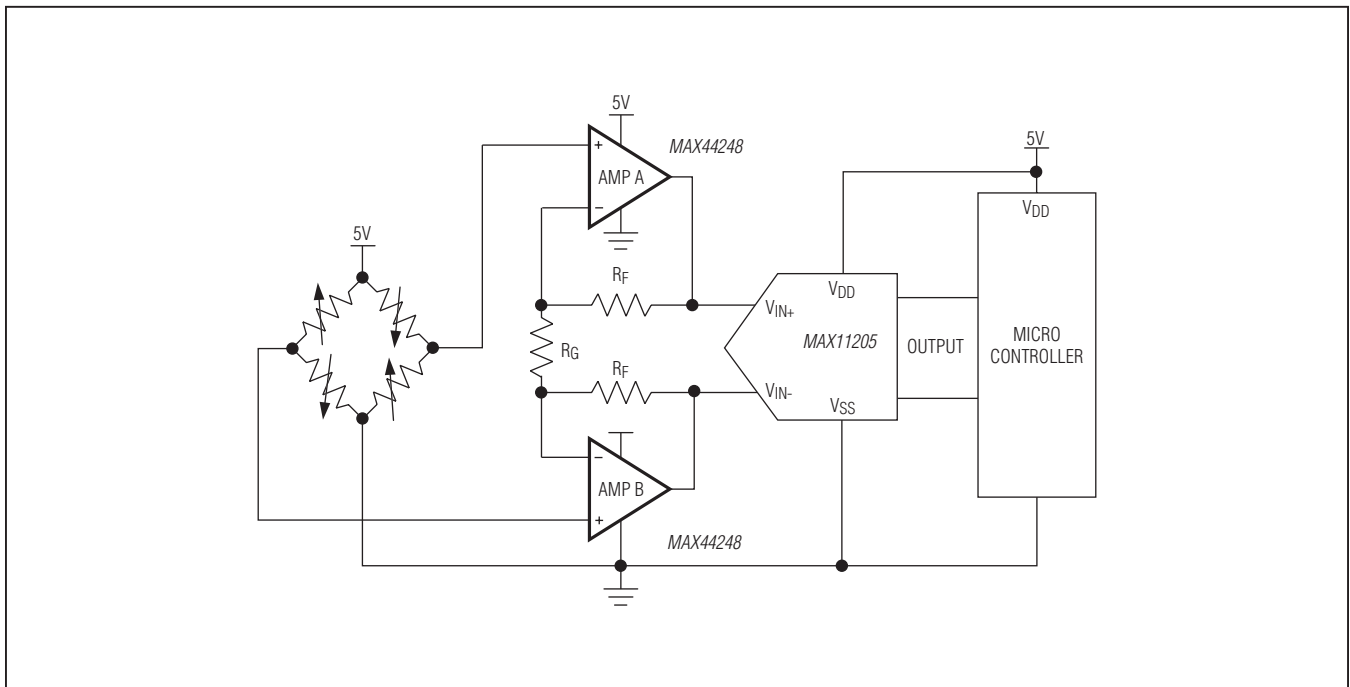


Figure 1. Weight Application

# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### Chip Information

PROCESS: BiCMOS

### Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX44248AUA+	-40°C to +125°C	8 $\mu$ MAX
MAX44248ASA+	-40°C to +125°C	8 SO

+Denotes a lead(Pb)-free/RoHS-compliant package.

### Package Information

For the latest package outline information and land patterns (footprints), go to [www.maximintegrated.com/packages](http://www.maximintegrated.com/packages). Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 SO	S8+4	<a href="#">21-0041</a>	<a href="#">90-0096</a>
8 $\mu$ MAX	U8+1	<a href="#">21-0036</a>	<a href="#">90-0092</a>



# MAX44248

## 36V, Precision, Low-Power, 90 $\mu$ A, Dual Op Amp

### Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	7/12	Initial release	—



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