# imall

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### **General Description**

The MAX9270 deserializer uses Maxim's gigabit multimedia serial link (GMSL) technology. The device functions the same as the MAX9260 deserializer without an output enable (ENABLE) pin. Outputs are enabled or disabled by a register bit. The deserializer pairs with any GMSL serializer to form a complete digital serial link for joint transmission of high-speed video, audio, and control data.

The deserializer accepts a maximum serial payload data rate of 2.5Gbps for a 15m shielded twisted-pair (STP) cable. The 24-bit or 32-bit width parallel interface operates up to a maximum bus clock of 104MHz or 78MHz, respectively. This serial link supports display panels from QVGA (320 x 240) up to XGA (1280 x 768), or dual-view WVGA (2 x 854 x 480).

The 24-bit or 32-bit mode handles 21 or 29 bits of data, along with an I<sup>2</sup>S input, supporting 4- to 32-bit audio word lengths and an 8kHz to 192kHz sample rate. The embedded control channel forms a full-duplex, differential 100kbps to 1Mbps UART link between the serializer and deserializer. The host electronic control unit (ECU) or microcontroller ( $\mu$ C) resides either on the serializer (for video display) or the deserializer (for image sensing). In addition, the control channel enables ECU/ $\mu$ C control of peripherals in the remote side of the serial link through I<sup>2</sup>C (base mode) or a user-defined full-duplex UART format (bypass mode).

The channel equalizer extends the link length and enhances the link reliability. Spread spectrum is available to reduce EMI on the parallel output data signals. The differential link complies with the ISO 10605 and IEC 61000-4-2 ESD-protection standards.

This device uses a 3.3V core supply and a 1.8V to 3.3V I/O supply. The device is available in a 56-pin TQFN package (8mm x 8mm x 0.75mm) with an exposed pad. Electrical performance is guaranteed over the -40°C to  $+105^{\circ}$ C automotive temperature range.

### **Applications**

High-Speed Serial-Data Transmission for Display High-Speed Serial-Data Transmission for Image Sensing

Automotive Navigation, Infotainment, and Image-Sensing Systems

### Features

- Pairs with Any GMSL Serializer
- 2.5Gbps Payload Rate, AC-Coupled Serial Link with 8b/10b Line Coding
- ◆ 24-Bit or 32-Bit Programmable Parallel Output Bus Supports Up to XGA (1280 x 768) or Dual-View WVGA (2 x 854 x 480) Panels with 18-Bit or 24-Bit Color
- 8.33MHz to 104MHz (24-Bit Bus) or 6.25MHz to 78MHz (32-Bit Bus) Parallel Data Rate
- Support Two/Three 10-Bit Camera Links at 104MHz/78MHz Maximum Pixel Clock
- 4-Bit to 32-Bit Word Length, 8kHz to 192kHz I<sup>2</sup>S Audio Channel Supports High-Definition Audio
- Embedded Half-/Full-Duplex Bidirectional Control Channel (100kbps to 1Mbps)
- Separate Interrupt Signal Supports Touch-Screen Functions for Display Panels
- Remote-End I<sup>2</sup>C Master for Peripherals
- Line Equalizer Extends Link Length
- Programmable Spread Spectrum on the Parallel
   Data Outputs Reduce EMI
- Does Not Require an External Clock
- Auto Data-Rate Detection Allows "On-The-Fly" Data-Rate Change
- Built-In PRBS Checker for BER Testing
- ISO 10605 and IEC 61000-4-2 ESD Protection
- ♦ -40°C to +105°C Operating Temperature Range
- Patent Pending

### Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX9270GTN/V+	-40°C to +105°C	56 TQFN-EP*

*N* denotes an automotive qualified part. +Denotes a lead(Pb)-free/RoHS-compliant package.

\*EP = Exposed pad.

T = Tape and reel.

Typical Applications Circuit appears at end of data sheet.

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Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

AVDD to EP0.5V to +3.9V	
DVDD to EP0.5V to +3.9V	
IOVDD to EP0.5V to +3.9V	
IN+, IN- to EP0.5V to +1.9V	
All Other Pins to EP0.5V to (IOVDD + 0.5V)	
IN+, IN- Short Circuit to Ground or	
SupplyContinuous	
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
56-Pin TQFN (derate 47.6mW/°C above +70°C)3809.5mW	
ESD Protection	
Human Body Model (R <sub>D</sub> = $1.5k\Omega$ , C <sub>S</sub> = $100pF$ )	
(IN+, IN-) to EP±8kV	
All Other Pins to EP±4kV	

IEC 61000-4-2 ( $R_D = 330\Omega$ , $C_S = 150pF$ )	
Contact Discharge	
(IN+, IN-) to EP	±8kV
Air Discharge	
(IN+, IN-) to EP	±10kV
ISO 10605 (R <sub>D</sub> = $2k\Omega$ , C <sub>S</sub> = $330pF$ )	
Contact Discharge	
(IN+, IN-) to EP	±8kV
Air Discharge	
(IN+, IN-) to EP	±20kV
Operating Temperature Range	40°C to +105°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

### PACKAGE THERMAL CHARACTERISTICS (Note 1)

56 TQFN

Junction-to-Ambient Thermal Resistance ( $\theta_{JA}$ )......21°C/W Junction-to-Case Thermal Resistance ( $\theta_{JC}$ )...........1°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to <u>www.maxim-ic.com/thermal-tutorial</u>.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### DC ELECTRICAL CHARACTERISTICS

 $(V_{DVDD} = V_{AVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential), EP connected to PCB ground (GND), T<sub>A</sub> = -40°C to +105°C, unless otherwise noted. Typical values are at V<sub>DVDD</sub> = V<sub>AVDD</sub> = V<sub>IOVDD</sub> = 3.3V, T<sub>A</sub> = +25°C.)

		51			,				
PARAMETER	SYMBOL	CONDITIONS		MIN	TYP MAX	UNITS			
SINGLE-ENDED INPUTS (ENABLE, INT, PWDN, SSEN, BWS, ES, DRS, MS, CDS, EQS, DCS)									
High-Level Input Voltage	VIH1			0.65 x Viovdd		V			
Low-Level Input Voltage	VIL1				0.35 x Viovdd	V			
Input Current	lin1	$V_{IN} = 0$ to $V_{I0}$	$V_{IN} = 0$ to $V_{IOVDD}$		+10	μA			
Input Clamp Voltage	VCL	I <sub>CL</sub> = -18mA			-1.5	V			
SINGLE-ENDED OUTPUTS (D	OUT_, SD, WS	, SCK, PCLK	OUT)						
Lligh Lougl Output Valtage			VDCS = VGND	VIOVDD - 0.3		V			
High-Level Output Voltage	VOH	$I_{OH} = -2mA$	VDCS = VIOVDD	VIOVDD - 0.2					
Low-Level Output Voltage VOL-	Vou	/	VDCS = VGND		0.3	- V			
	VOL1	$I_{OL} = 2mA$	VDCS = VIOVDD		0.2				

### DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DVDD} = V_{AVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential), EP connected to PCB ground (GND), T<sub>A</sub> = -40°C to +105°C, unless otherwise noted. Typical values are at V<sub>DVDD</sub> = V<sub>AVDD</sub> = V<sub>IOVDD</sub> = 3.3V, T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL		CON	IDITIONS		MIN	TYP	MAX	UNITS
			Vo =	0V,	VIOVDD = 3.0V to 3.6V	15	25	39	
		DOUT_, SD,	VDCS	= VGND	VIOVDD = 1.7V to 1.9V	3	7	13	
		WS, SCK	V <sub>O</sub> =	0V,	VIOVDD = 3.0V to 3.6V	20	35	63	
			VDCS	= VIOVDD	VIOVDD = 1.7V to 1.9V	5	10	21	0
Output Short-Circuit Current	IOS		V <sub>O</sub> =	0V,	VIOVDD = 3.0V to 3.6V	15	33	50	mA
			VDCS	= VGND	VIOVDD = 1.7V to 1.9V	5	10	17	
		PCLKOUT	V <sub>O</sub> =	0V,	VIOVDD = 3.0V to 3.6V	30	54	97	
			VDCS = VIOVE	= VIOVDD	VIOVDD = 1.7V to 1.9V	9	16	32	1
I <sup>2</sup> C AND UART I/O, OPEN-DRAII	N OUTPUTS	(RX/SDA, TX/	SCL, E	ERR, GPIO	, LOCK)				
High-Level Input Voltage	VIH2					0.7 x Viovdd			V
Low-Level Input Voltage	VIL2							0.3 x Viovdd	V
Input Current	I <sub>IN2</sub>	$V_{IN} = 0$ to $V_{I0}$ (Note 2)	DDVC	RX/SDA, GPIO, ER		-110 -80		+1 +1	μΑ
Low-Level Open-Drain Output Voltage	V <sub>OL2</sub>	I <sub>OL</sub> = 3mA		VIOVDD =	1.7V to 1.9V 3.0V to 3.6V		·	0.4	V V
DIFFERENTIAL OUTPUTS FOR	REVERSE C	ONTROL CH	ANNEL	(IN+, IN-)					
Differential High Output Peak Voltage, (V <sub>IN</sub> +) - (V <sub>IN</sub> -)	VROH	No high-spee (Figure 1)	ed data	transmiss	ion	30		60	mV
Differential Low Output Peak Voltage, (VIN+) - (VIN-)	V <sub>ROL</sub>	No high-spee (Figure 1)	ed data	transmiss	ion	-60		-30	mV
DIFFERENTIAL INPUTS (IN+, IN-	-)	1							
Differential High Input Threshold (Peak), (VIN+) - (VIN-)	VIDH(P)	(Figure 2)					40	90	mV
Differential Low Input Threshold (Peak), (VIN+) - (VIN-)	VIDL(P)	(Figure 2)				-90	-40		mV
Input Common-Mode Voltage, ((VIN+) + (VIN-))/2	VCMR					1	1.3	1.6	V
Differential Input Resistance (Internal)	Rı					80	100	130	Ω

### DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DVDD} = V_{AVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential), EP connected to PCB ground (GND),  $T_A = -40^{\circ}C$  to  $+105^{\circ}C$ , unless otherwise noted. Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 3.3V$ ,  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL	CONDIT	IONS	MIN	ТҮР	MAX	UNITS
POWER SUPPLY							
		V <sub>BWS</sub> = V <sub>GND</sub> ,	2% spread spectrum active		113	166	
		fpclkout = 16.6MHz	Spread spectrum disabled		105	155	
		VBWS = VGND,	2% spread spectrum active		122	181	- mA
Worst-Case Supply Current (Figure 3)	luco	fPCLKOUT = 33.3MHz	Spread spectrum disabled		110	165	
	Iwcs	VBWS = VGND, fPCLKOUT = 66.6MHz	2% spread spectrum active		137	211	
			Spread spectrum disabled		120	188	
		VBWS = VGND, fPCLKOUT = 104MHz	2% spread spectrum active		159	247	
			Spread spectrum disabled		135	214	
Sleep-Mode Supply Current	Iccs				80	130	μA
Power-Down Supply Current	Iccz	VPWDN = VGND			19	70	μA

### **AC ELECTRICAL CHARACTERISTICS**

 $(V_{DVDD} = V_{AVDD} = 3.0V$  to 3.6V,  $V_{IOVDD} = 1.7V$  to 3.6V,  $R_L = 100\Omega \pm 1\%$  (differential), EP connected to PCB ground (GND),  $T_A = -40^{\circ}C$  to  $+105^{\circ}C$ , unless otherwise noted. Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 3.3V$ ,  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS		
PARALLEL CLOCK OUTPUT (PCLKOUT)								
		VBWS = VGND, VDRS = VIOVDD	8.33		16.66			
Clock Fragueney	foot your	VBWS = VGND, VDRS = VGND	16.66		104	MHz		
Clock Frequency	fPCLKOUT	VBWS = VIOVDD, VDRS = VIOVDD	6.25		12.5			
		VBWS = VIOVDD, VDRS = VGND	12.5		78			
Clock Duty Cycle	DC	tHIGH/tT or tLOW/tT (Figure 4)	40	50	60	%		
Clock Jitter	tj	Period jitter, RMS, spread off, 3.125Gbps, PRBS pattern, UI = 1/fPCLKOUT		0.05		UI		
I <sup>2</sup> C/UART PORT TIMING								
Output Rise Time	t <sub>R</sub>	30% to 70%, CL = 10pF to 100pF, 1k $\Omega$ pullup to IOVDD	20		150	ns		
Output Fall Time	tF	70% to 30%, CL = 10pF to 100pF, 1k $\Omega$ pullup to IOVDD	20		150	ns		
Input Setup Time	tSET	I <sup>2</sup> C only (Figure 5)	100			ns		
Input Hold Time	thold	I <sup>2</sup> C only (Figure 5)	0			ns		

### AC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DVDD} = V_{AVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential), EP connected to PCB ground (GND),  $T_A = -40^{\circ}C$  to  $+105^{\circ}C$ , unless otherwise noted. Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 3.3V$ ,  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL	CON	DITIONS	MIN	ТҮР	MAX	UNITS
SWITCHING CHARACTERISTICS	\$						
		20% to 80%,	$V_{DCS} = V_{IOVDD},$ $C_{L} = 10pF$	0.4		2.2	
PCLKOUT Rise-and-Fall Time	t <sub>R</sub> , t <sub>F</sub>	$V_{IOVDD} = 1.7V$ to 1.9	$V_{DCS} = V_{GND},$ $C_{L} = 5pF$	0.5		2.8	ns
	ч <del>п</del> , ч <del>г</del>	20% to 80%,	$V_{DCS} = V_{IOVDD},$ $C_L = 10pF$	0.25		1.7	
		$V_{IOVDD} = 3.0V$ to 3.6	$\begin{array}{l} \text{SV}  \text{V}_{\text{DCS}} = \text{V}_{\text{GND}}, \\ \text{C}_{\text{L}} = 5\text{pF} \end{array}$	0.3		2.0	
		20% to 80%,	$V_{DCS} = V_{IOVDD},$ $C_L = 10pF$	0.5		3.1	
Parallel Data Rise-and-Fall Time	to to	$V_{IOVDD} = 1.7V$ to 1.9	$V_{DCS} = V_{GND},$ $C_L = 5pF$	0.6		3.8	
(Figure 6)	tR, tF	20% to 80%,	$V_{DCS} = V_{IOVDD},$ $C_{L} = 10pF$	0.3		2.2	- ns
		$V_{IOVDD} = 3.0V$ to 3.6	$\begin{array}{l} \text{SV}  \  \  V_{\text{DCS}} = V_{\text{GND}}, \\ \text{C}_{\text{L}} = 5 \text{pF} \end{array}$	0.4		2.4	
Deserializer Delay	top	Spread spectrum en	abled (Figure 7)			2880	Bits
Desenanzer Delay	tsD	Spread spectrum disabled (Figure 7)				750	
Lock Time	tlock	Spread spectrum en	abled (Figure 8)			1500	μs
LOCK TIME	LOCK	Spread spectrum off	(Figure 8)			1000	μο
Power-Up Time	tpu	(Figure 9)				2500	μs
Reverse Control-Channel Output Rise Time	tR	No high-speed trans	mission (Figure 1)	180		400	ns
Reverse Control-Channel Output Fall Time	tF	No high-speed trans	mission (Figure 1)	180		400	ns
I <sup>2</sup> S OUTPUT TIMING							
		$t_{WS} = 1/f_{WS}$ , rising	fws = 48kHz or 44.1kHz		0.4e - 3 x tws	0.5e - 3 x tws	
WS Jitter	taj-ws	(falling) edge to falling (rising) edge	f <sub>WS</sub> = 96kHz		0.8e - 3 x tws	1e - 3 x tws	ns
		(Note 3)	f <sub>WS</sub> = 192kHz		1.6e - 3 × tws	2e - 3 x tws	1
			nws = 16 bits, fws = 48kHz or 44.1kHz		13е - 3 x tscк	16e - 3 x tsck	
SCK Jitter	taj-sck	tAJ-SCK ing edge to rising fv edge	nws = 24 bits, fws = 96kHz		39е - 3 х tscк	48e - 3 x tsck	ns
			$n_{WS} = 32 \text{ bits},$ $f_{WS} = 192 \text{kHz}$		0.1 x tsck	0.13	

### AC ELECTRICAL CHARACTERISTICS (continued)

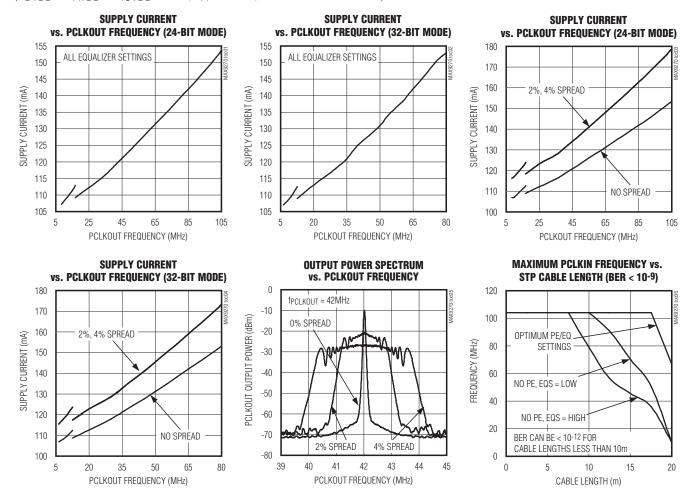
 $(V_{DVDD} = V_{AVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential), EP connected to PCB ground (GND), T<sub>A</sub> = -40°C to +105°C, unless otherwise noted. Typical values are at V<sub>DVDD</sub> = V<sub>AVDD</sub> = V<sub>IOVDD</sub> = 3.3V, T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Audio Skew Relative to Video	ASK	Video and auc	Video and audio synchronized		3 x tws	4 x tws	μs
SCK SD WS Bigg and Fall Time	to to	20% to 80%	VDCS = VIOVDD, CL = 10pF	0.3		3.1	ns
SCK, SD, WS Rise-and-Fall Time	t <sub>R</sub> , t <sub>F</sub>		$V_{DCS} = V_{GND}, C_L = 5pF$	0.4		3.8	ns
SD, WS Valid Time Before SCK	to) (D	toox - 1/foox	(Figure 11)	0.35	0.5		20
SD, WS Valid Time Belore SCK	tD/B	ISCK = MSCK	$t_{SCK} = 1/f_{SCK}$ (Figure 11)		x tsck		ns
SD, WS Valid Time After SCK	tova		(Figure 11)	0.35	0.5		20
SD, WS Valid Time Alter SCK	tdva	tsck = 1/fsck (Figure 11)		x tsck	x tsck		ns

**Note 2:** Minimum I<sub>IN</sub> due to voltage drop across the internal pullup resistor.

Note 3: Rising to rising edge jitter can be twice as large.

 $(V_{DVDD} = V_{AVDD} = V_{IOVDD} = 3.3V, T_A = +25^{\circ}C, unless otherwise noted.)$ 



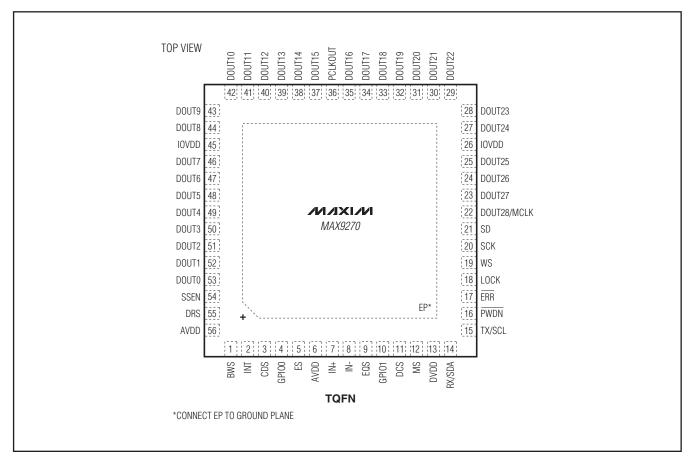
Typical Operating Characteristics

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6

### \_Pin Configuration

**MAX9270** 



### **Pin Description**

PIN	NAME	FUNCTION
1	BWS	Bus-Width Select. Parallel output bus-width selection input requires external pulldown or pullup resistors. Set BWS = low for 24-bit bus mode. Set BWS = high for 32-bit bus mode.
2	INT	Interrupt. Interrupt input requires external pulldown or pullup resistors. A transition on the INT input of the deserializer toggles the serializer's INT output.
3	CDS	Control-Direction Selection. Control-link-direction selection input requires external pull- down or pullup resistors. Set CDS = low for $\mu$ C use on the serializer side of the serial link. Set CDS = high for $\mu$ C use on the deserializer side of the serial link.
4	GPIO0	GPIO0. Open-drain general-purpose input/output with internal $60k\Omega$ pullup resistors to IOVDD. GPIO0 is high impedance during power-up and when $\overline{PWDN} = low$ .
5	ES	Edge Select. PCLKOUT edge-selection input requires external pulldown or pullup resistors. Set ES = low for a rising-edge trigger. Set ES = high for a falling-edge trigger.

### 7

### Pin Description (continued)

PIN	NAME	FUNCTION
6, 56	AVDD	3.3V Analog Power Supply. Bypass AVDD to EP with 0.1µF and 0.001µF capacitors as close as possible to the device with the smallest value capacitor closest to AVDD.
7, 8	IN+, IN-	Differential CML Input +/ Differential inputs of the serial link.
9	EQS	Equalizer Select. Deserializer equalizer-selection input requires external pulldown or pullup resistors. The state of EQS latches upon power-up or rising edge of $\overline{PWDN}$ . Set EQS = low for 10.7dB equalizer boost (EQTUNE = 1001). Set EQS = high for 5.2dB equalizer boost (EQTUNE = 0100).
10	GPIO1	GPIO1. Open-drain general-purpose input/output with internal $60k\Omega$ pullup resistors to IOVDD. GPIO1 is high impedance during power-up and when $\overline{PWDN} = low$ .
11	DCS	Drive Current Select. Driver current-selection input requires external pulldown or pullup resistors. Set DCS = high for stronger parallel data and clock output drivers. Set DCS = low for normal parallel data and clock drivers (see the <i>DC Electrical Characteristics</i> table).
12	MS	Mode Select. Control-link mode-selection/autostart mode selection input requires external pulldown or pullup resistors. MS sets the control-link mode when CDS = high (see the <i>Control-Channel and Register Programming</i> section). Set MS = low to select base mode. Set MS = high to select the bypass mode. MS sets autostart mode when CDS = low (see Tables 13 and 14).
13	DVDD	3.3V Digital Power Supply. Bypass DVDD to EP with 0.1µF and 0.001µF capacitors as close as possible to the device with the smaller value capacitor closest to DVDD.
14	RX/SDA	Receive/Serial Data. UART receive or I <sup>2</sup> C serial-data input/output with internal 30k $\Omega$ pullup to IOVDD. In UART mode, RX/SDA is the Rx input of the deserializer's UART. In I <sup>2</sup> C mode, RX/SDA is the SDA input/output of the serializer's I <sup>2</sup> C master.
15	TX/SCL	Transmit/Serial Clock. UART transmit or I <sup>2</sup> C serial-clock output with internal 30k $\Omega$ pullup to IOVDD. In UART mode, TX/SCL is the Tx output of the serializer's UART. In I <sup>2</sup> C mode, TX/SCL is the SCL output of the deserializer's I <sup>2</sup> C master.
16	PWDN	Power-Down. Active-low power-down input requires external pulldown or pullup resis- tors.
17	ERR	Error. Active-low open-drain video data error output with internal pullup to IOVDD. $\overline{\text{ERR}}$ goes low when the number of decoding errors during normal operation exceed a programmed error threshold or when at least one PRBS error is detected during PRBS test. $\overline{\text{ERR}}$ is high impendence when $\overline{\text{PWDN}}$ = low.
18	LOCK	Open-Drain Lock Output with Internal Pullup to IOVDD. LOCK = high indicates PLLs are locked with correct serial-word-boundary alignment. LOCK = low indicates PLLs are not locked or incorrect serial-word-boundary alignment. LOCK remains low when the configuration link is active. LOCK is high impedance when PWDN = low.
19	WS	Word Select. I <sup>2</sup> S word-select output.
20	SCK	Serial Clock. I <sup>2</sup> S serial-clock output
21	SD	Serial Data. I <sup>2</sup> S serial-data output. Disable I <sup>2</sup> S to use SD as an additional data output latched on the selected edge of PCLKOUT.

### \_Pin Description (continued)

PIN	NAME	FUNCTION
22–25, 27–35, 37–44, 46–53	DOUT28/MCLK, DOUT27, DOUT26, DOUT25, DOUT24–DOUT16, DOUT15–DOUT8, DOUT7–DOUT0	Data Output[0:28]. Parallel data outputs. Output data can be strobed on the selected edge of PCLKOUT. Set BWS = low (24-bit mode) to use DOUT0–DOUT20 (RGB and SYNC). DOUT21–DOUT28 are not used in 24-bit mode and are set to low. Set BWS = high (32-bit mode) to use DOUT0–DOUT28 (RGB, SYNC, and two extra outputs). DOUT28 can be used to output MCLK (see the <i>Additional MCLK Output for Audio Applications</i> section).
26, 45	IOVDD	1.8V to 3.3V Logic I/O Power Supply. Bypass IOVDD to EP with $0.1\mu$ F and $0.001\mu$ F capacitors as close as possible to the device with the smaller value capacitor closest to IOVDD.
36	PCLKOUT	Parallel Clock Output. Used for DOUT0-DOUT28.
54	SSEN	Spread-Spectrum Enable. Parallel output spread-spectrum enable input requires external pulldown or pullup resistors. The state of SSEN latches upon power-up or when resuming from power-down mode ( $\overline{PWDN} = low$ ). Set SSEN = high for ±2% spread spectrum on the parallel outputs. Set SSEN = low to use the parallel outputs without spread spectrum.
55	DRS	Data-Rate Select. Data-rate range-selection input requires external pulldown or pullup resistors. Set DRS = high for parallel input data rates of 8.33MHz to 16.66MHz (24-bit mode) or 6.25MHz to 12.5MHz (32-bit mode). Set DRS = low for parallel input data rates of 16.66MHz to 104MHz (24-bit mode) or 12.5MHz to 78MHz (32-bit mode).
_	EP	Exposed Pad. EP functions as the IC's ground connection. <b>MUST</b> connect EP to the ground plane to maximize thermal and electrical performance.

LFLT PCLKIN SPREAD FILTER PLL PLL LMN0 LINE-FAULT LMN1 CLKDIV DET  $\geq$ 8b/10b DIN[N:0] CML FIFO ENCODE S Тχ PARITY OUT+| WS, SD, SCK AUDIO PRBS TERM OUT-FIFO GEN TX/SCL RX/SDA REV CH UART/I<sup>2</sup>C Rx GMSL SERIALIZER SPREAD CDR PLL PLL STP CABLE PCLKOUT  $(Z_0 = 50)$ EQ CLKDIV 8b/10b DOUT[N:0] CML FIFO DECODE РĆ S Rx PARITY IN-WS, SD, SCK AUDIO PRBS ///XI/// TERM IN+ **FIFO** CHECK MAX9270 TX/SCL RX/SDA REV CH UART/I<sup>2</sup>C Тχ DESERIALIZER

**Functional Diagram** 

**MAX9270** 

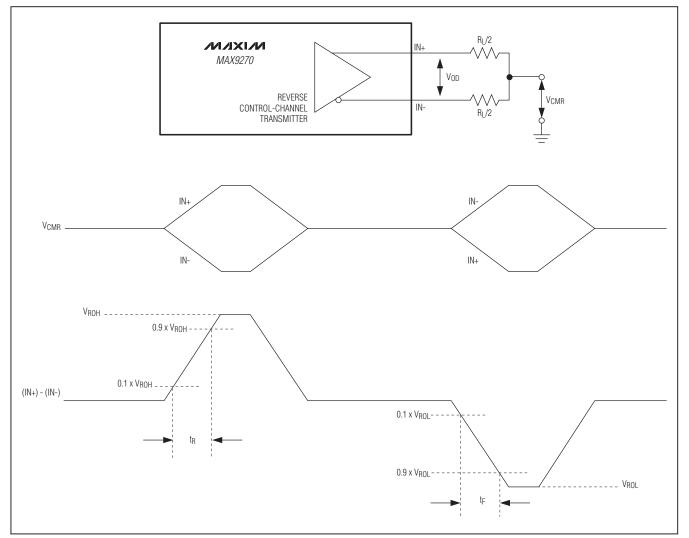
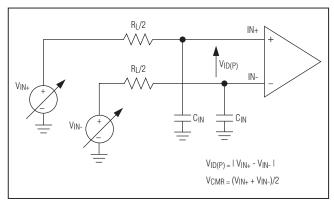
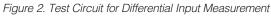


Figure 1. Reverse Control-Channel Output Parameters





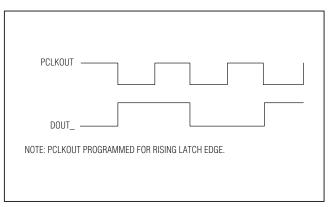


Figure 3. Worst-Case Pattern Output

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**MAX9270** 



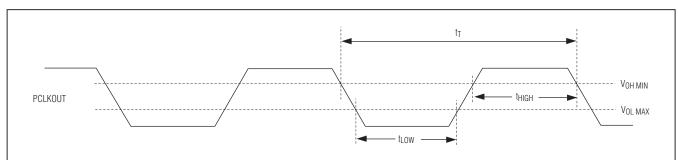


Figure 4. Clock Output High-and-Low Times

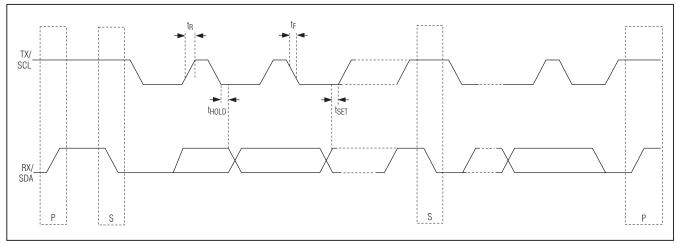


Figure 5. I<sup>2</sup>C Timing Parameters

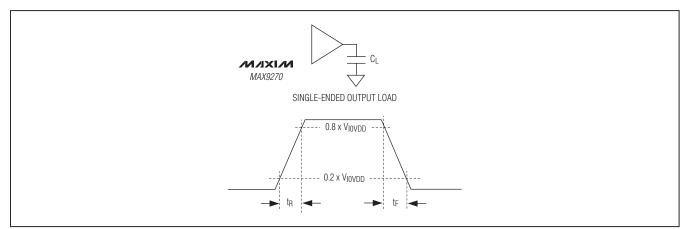


Figure 6. Output Rise-and-Fall Times

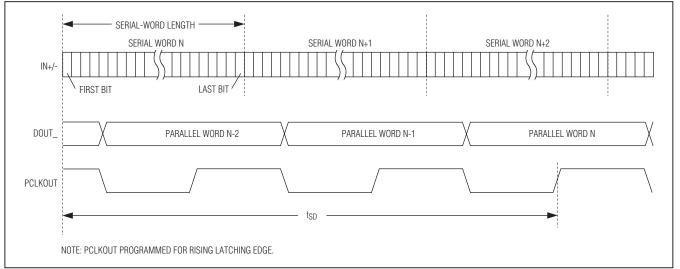


Figure 7. Deserializer Delay

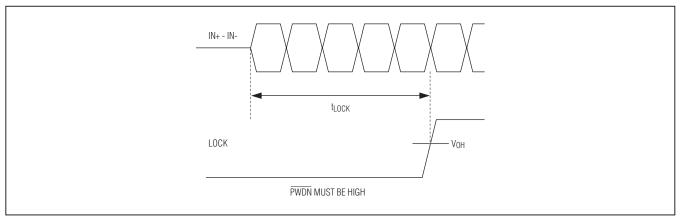


Figure 8. Lock Time

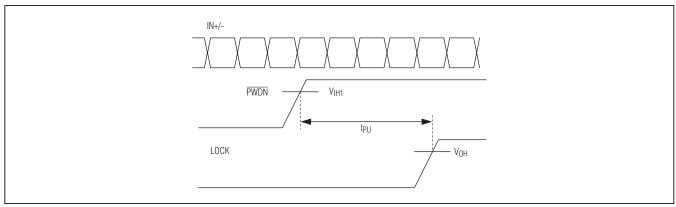


Figure 9. Power-Up Delay



**MAX9270** 

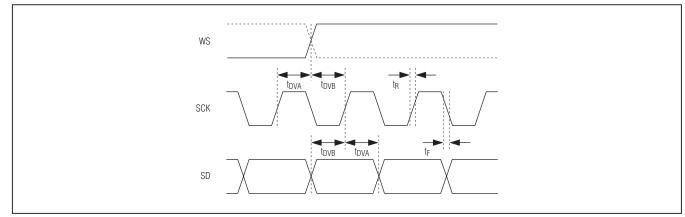


Figure 10. Output I<sup>2</sup>S Timing Parameters

**MAX9270** 

#### **Detailed Description**

The MAX9270 deserializer utilizes Maxim's GMSL technology. This device pairs with any GMSL serializer to form a complete digital serial link for joint transmission of high-speed video, audio, and control data for videodisplay or image-sensing applications. The MAX9270 functions the same as the MAX9260 without an output enable pin (ENABLE). Outputs are enabled by default and programmable by a register bit. The serial-payload data rate can reach up to 2.5Gbps for a 15m STP cable. The parallel interface is programmable for 24-bit or 32-bit width modes at the maximum bus clock of 104MHz or 78MHz, respectively. The minimum bus clock is 6.25MHz for the 32-bit mode and 8.33MHz for the 24-bit mode. With such a flexible data configuration, the GMSL is able to support XGA (1280 x 768) or dual-view WVGA (2 x 854 x 480) display panels. For image sensing, it supports three 10-bit camera links simultaneously with a pixel clock up to 78MHz. The 24-bit mode handles 21-bit data and control signals plus an I<sup>2</sup>S audio signal. The 32-bit mode handles 29-bit data and control signals plus an I<sup>2</sup>S audio signal. Any combination and sequence of color video data, video sync, and control signals make up the 21-bit or 29-bit parallel data on DOUT\_. The I2S port supports the sampled audio data at a rate from 8kHz to 192kHz and the audio word length of anywhere between 4 to 32 bits. The embedded control channel forms a UART link between the serializer and deserializer. The UART link can be set to half-duplex mode or full-duplex mode depending on the application. The GMSL supports UART rates from 100kbps to 1Mbps. Using this control link, a host ECU or  $\mu$ C communicates with the serializer and deserializer, as well as the peripherals in the remote side, such as backlight control, grayscale gamma correction, camera module, and touch screen. All serial communication (forward and reverse) uses differential signaling. The peripheral programming uses I<sup>2</sup>C format or the default GMSL UART format. A separate bypass mode enables communication using a full-duplex, user-defined UART format. The control link between the serializer/deserializer allows  $\mu$ C connectivity to either device or peripherals to support video-display or image-sensing applications.

The AC-coupled serial link uses 8b/10b coding. The deserializer features a programmable channel equalizer to extend the link length and enhance the link reliability. A programmable spread-spectrum feature reduces EMI on the parallel data outputs. The differential serial link input pins comply with the ISO 10605 and IEC 61000-4-2 ESD-protection standards. This device uses a 3.3V core supply and a 1.8V to 3.3V I/O supply.

#### **Register Mapping**

The  $\mu$ C configures various operating conditions of the GMSL through registers in the serializer/deserializer. The default device addresses stored in the R0 and R1 registers of the serializer/deserializer are 0x80. Write to the R0/R1 registers to change the device address of the serializer or deserializer.

 Table 1. Power-Up Default Register Map (see Table 12)

REGISTER ADDRESS (hex)	POWER-UP DEFAULT (hex)	POWER-UP DEFAULT SETTINGS (MSB FIRST)
0x00	0x80	SERID =1000000, serializer device identifier is 1000 000 RESERVED = 0
0x01	0x90	DESID =1001000, deserializer device identifier is 1001 000 RESERVED = 0
0x02	0x1F or 0x5F	SS = 00 (SSEN = low), SS = 01 (SSEN = high), spread-spectrum settings depend on SSEN pin state at power-up RESERVED = 0 AUDIOEN = 1, I <sup>2</sup> S channel enabled PRNG = 11, automatically detect the pixel clock range SRNG = 11, automatically detect serial-data rate
0x03	0x00	AUTOFM = 00, calibrate spread-modulation rate only once after locking RESERVED = 0 SDIV = 00000, autocalibrate sawtooth divider
0x04	0x03 or 0x83	LOCKED = 0, LOCK output = low (read only) OUTENB = 0, outputs enabled PRBSEN = 0, PRBS test disabled SLEEP = 0 or 1, SLEEP setting default depends on CDS and MS pin state at pow- er-up (see the <i>Link Startup Procedure</i> section) INTTYPE = 00, base mode uses I <sup>2</sup> C REVCCEN = 1, reverse control channel active (sending) FWDCCEN = 1, forward control channel active (receiving)
0x05	0x28 or 0x29	RESERVED = 0 HPFTUNE = 01, 3.75MHz equalizer highpass cutoff frequency PDHF = 0, high-frequency boosting disabled EQTUNE = 1000 (EQS = high, 10.7dB), EQTUNE = 1001 (EQS = low, 5.2dB), EQTUNE default setting depends on EQS pin state at power-up
0x06	0x0F	DISSTAG = 0, staggered outputs enabled AUTORST = 0, error registers/output auto reset disabled DISINT = 0, INT transmission enabled INT = 0, INT output = low (read only) GPI010UT = 1, GPI01 output set to high GPI01 = 1, GPI01 input = high (read only) GPI000UT = 1, GPI00 output set to high GPI00 = 1, GPI00 input = high (read only)
0x07	0x54	RESERVED = 01010100
0x08	0x30	RESERVED = 00110000
0x09	0xC8	RESERVED = 11001000
0x0A	0x12	RESERVED = 00010010
0x0B	0x20	RESERVED = 00100000

### Table 1. Power-Up Default Register Map (see Table 12) (continued)

REGISTER ADDRESS (hex)	POWER-UP DEFAULT (hex)	POWER-UP DEFAULT SETTINGS (MSB FIRST)
0x0C	0x00	ERRTHR = 00000000, error threshold set to zero for decoding errors
0x0D	0x00 (read only)	DECERR = 00000000, zero decoding errors detected
0x0E	0x00 (read only)	PRBSERR = 00000000, zero PRBS errors detected
0x12	0x00	MCLKSRC = 0, MCLK is derived from PCLKOUT (see Table 4) MCLKDIV = 0000000, MCLK output is disabled
0x1E	0x02 (read only)	ID = 00000010, device ID is 0x02
0x1F	0x0X (read only)	RESERVED = 0000 REVISION = XXXX

### Table 2. Bus-Width Selection Using BWS

BWS INPUT STATE	BUS WIDTH	PARALLEL BUS SIGNALS USED
Low	24	DOUT[0:20], WS, SCK, SD
High	32	DOUT[0:28], WS, SCK, SD

#### **Parallel Outputs**

The parallel bus uses two selectable bus widths, 24 bits and 32 bits. BWS selects the bus width according to Table 2. In 24-bit mode, DIN21–DIN28 are not used and are internally pulled down. For both modes, SD, SCK, and WS pins are dedicated for I<sup>2</sup>S audio data. The assignments of the first 21 or 29 signals are interchangeable and appear in the same order at both sides of the serial link. In image-sensing applications, disabling the I<sup>2</sup>S audio channel (through the internal registers) allows the serialization of three 10-bit camera data streams through DIN[0:28] plus SD inputs. The parallel bus accepts data clock rates from 8.33MHz to 104MHz for the 24-bit mode and 6.25MHz to 78MHz for the 32-bit mode.

Serial Link Signaling and Data Format

The serializer's high-speed data serial output uses CML signaling with programmable preemphasis and AC-coupling. The deserializer's high-speed receiver uses AC-coupling and programmable channel equalization. Together, the GMSL operates at up to 3.125Gbps over STP cable lengths up to 15m.

The serializer scrambles and encodes the parallel input bits, and sends the 8b/10b coded signal through the serial link. The deserializer recovers the embedded serial clock and then samples, decodes, and descrambles the data onto the parallel output bus. Figures 11 and 12 show the serial-data packet format prior to scrambling and 8b/10b coding. For the 24-bit or 32-bit mode, the first 21 or 29 serial bits map to DOUT[20:0] or DOUT[28:0], respectively. The audio channel bit (ACB) contains an encoded audio signal derived from the three I<sup>2</sup>S inputs (SD, SCK, and WS). The forward control channel (FCC) bit carries the forward control data. The last bit (PCB) is the parity bit of the previous 23 or 31 bits.

#### **Reverse Control Channel**

The GMSL uses the reverse control channel to send I<sup>2</sup>C/UART in the opposite direction of the video stream from the deserializer to the serializer. The reverse control channel and forward video data coexist on the same twisted pair forming a bidirectional link. The reverse control channel operates independently from the forward control channel. The reverse control channel is available 500µs after power-up. The serializer temporarily disables the reverse control channel for 350µs after starting/stopping the forward serial link.

#### **Parallel Data-Rate Selection**

The deserializer uses the DRS input to set the parallel data rate. Set DRS high to use a low-speed parallel data rate in the range of 6.25MHz to 12.5MHz (32-bit mode) or 8.33MHz to 16.66MHz (24-bit mode). Set DRS low



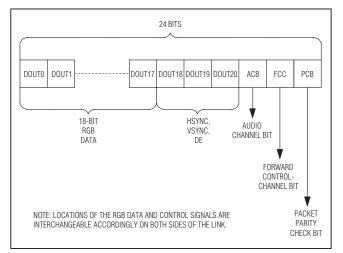


Figure 11. 24-Bit Mode Serial Link Data Format

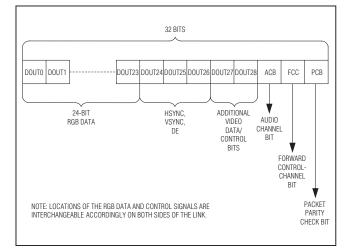


Figure 12. 32-Bit Mode Serial Link Data Format

### Table 3. Maximum Audio Sampling Rates for Various PCLK\_ Frequencies

WORD LENGTH (Bits)		PCLK_ FREQUENCY (DRS = LOW) (MHz)			PCLK_ FREQUENCY (DRS = HIGH) (MHz)			
	12.5	15	16.6	> 20	6.25	7.5	8.33	> 10
8	> 192	> 192	> 192	> 192	> 192	> 192	> 192	> 192
16	> 192	> 192	> 192	> 192	> 192	> 192	> 192	> 192
18	185.5	> 192	> 192	> 192	185.5	> 192	> 192	> 192
20	174.6	> 192	> 192	> 192	174.6	> 192	> 192	> 192
24	152.2	182.7	> 192	> 192	152.2	182.7	> 192	> 192
32	123.7	148.4	164.3	> 192	123.7	148.4	164.3	> 192

for normal operation with parallel data rates higher than 12.5MHz (32-bit mode) or 16.66MHz (24-bit mode).

#### **Audio Channel**

The I<sup>2</sup>S audio channel supports audio sampling rates from 8kHz to 192kHz and audio word lengths from 4 bits to 32 bits. The audio bit clock (SCK) does not need to be synchronized with PCLKIN. The serializer automatically encodes audio data into a single bit stream synchronous with PCLKIN. The deserializer decodes the audio stream and stores audio words in a FIFO. Audio rate detection uses an internal oscillator to continuously determine the audio data rate and output the audio in I<sup>2</sup>S format. The audio channel is enabled by default. When the audio channel is disabled, the SD pins on both sides are treated as a regular parallel data pin.

PCLK\_ frequencies can limit the maximum supported audio sampling rate. Table 3 lists the maximum audio sampling rate for various PCLK\_ frequencies. Spread-

spectrum settings do not affect the I<sup>2</sup>S data rate or WS clock frequency.

#### Additional MCLK Output for Audio Applications

Some audio DACs such as the MAX9850 do not require a synchronous main clock (MCLK), while other DACs require MCLK to be a specific multiple of WS. If an audio DAC chip needs the MCLK to be a multiple of WS, synchronize the I<sup>2</sup>S audio data with PCLK\_ of the GMSL, which is typical for most applications. Select the PCLK\_ to be the multiple of WS, or use a clock synthesis chip, such as the MAX9491, to regenerate the required MCLK from PCLK\_ or SCK.

For audio applications that cannot directly use the PCLKOUT output, the deserializer provides a divided MCLK output on DOUT28 at the expense of one less parallel line in 32-bit mode (24-bit mode is not affected). By default, DOUT28 operates as a parallel data output and



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### Table 4. fSRC Settings

MCLKSRC SETTING (REGISTER 0x12, D7)	DATA-RATE SETTING	BIT-WIDTH SETTING	MCLK SOURCE FREQUENCY (fSRC)
	High speed	24-bit mode	3 x fpclkout
0	riigh speed	32-bit mode	4 x fpclkout
0	Low apod	24-bit mode	6 x fPCLKOUT
	Low speed	32-bit mode	8 x fpclkout
1	_	_	Internal oscillator (120MHz typ)

MCLK is turned off. Set MCLKDIV (deserializer register 0x12, D[6:0]) to a non-zero value to enable the MCLK output. Set MCLKDIV to 0x00 to disable MCLK and set DOUT28 as a parallel data output.

The output MCLK frequency is:

$$f_{MCLK} = \frac{f_{SRC}}{MCLKDIV}$$

where f<sub>SRC</sub> is the MCLK source frequency (Table 4) and MCLKDIV is the divider ratio from 1 to 127.

Choose MCLKDIV values so that  $f_{MCLK}$  is not greater than 60MHz. MCLK frequencies derived from PCLK\_ (MCLKSRC = 0) are not affected by spread-spectrum settings in the deserializer. Enabling spread spectrum in the serializer, however, introduces spread spectrum into MCLK. Spread-spectrum settings of either device do not affect MCLK frequencies derived from the internal oscillator. The internal oscillator frequency ranges from 100MHz to 150MHz over all process corners and operating conditions.

#### **Control-Channel and Register Programming**

The  $\mu$ C uses the control link to send and receive control data over the STP link simultaneously with the high-speed data. Configuring the CDS pin allows the  $\mu$ C to control the link from either the serializer or the deserializer side to support video-display or image-sensing applications.

The control link between the  $\mu$ C and the serializer/ deserializer runs in base mode or bypass mode according to the mode selection (MS) input of the device connected to the  $\mu$ C. Base mode is a half-duplex control link and the bypass mode is a full-duplex control link. In base mode, the  $\mu$ C is the host and accesses the registers of both the serializer/deserializer by using the GMSL UART protocol. The  $\mu$ C can also program the peripherals on the remote side by sending the UART packets converted to  $I^{2}C$  by the device on the remote side of the link (deserializer for LCD or serializer for image-sensing applications). The  $\mu$ C communicates with a UART peripheral in base mode (through INTTYPE register settings) using the half-duplex default GMSL UART protocol. The device addresses of the serializer and deserializer in the base mode are programmable. The default values are 0x80 and 0x90, respectively.

In base mode, when the peripheral interface uses  $I^{2}C$  (default), the serializer/deserializer only convert packets that have device addresses different from themselves to  $I^{2}C$ . The converted  $I^{2}C$  bit rate is the same as the original UART bit rate.

In bypass mode, the  $\mu$ C bypasses the GMSL and communicates with the peripherals directly using its own defined UART protocol. The  $\mu$ C cannot access the GMSL registers in this mode. Peripherals accessed through the forward control channel using the UART interface need to handle at least one PCLK\_ period of jitter due to the asynchronous sampling of the UART signal by PCLK\_.

The serializer embeds control signals going to the deserializer in the high-speed forward link. Do not send a low value longer than 100µs in either base or bypass mode. The deserializer uses a proprietary differential line coding to send signals back towards the serializer. The speed of the control link ranges from 100kbps to 1Mbps in both directions. The serializer/deserializer automatically detects the control-channel bit rate in base mode. Packet bit rates can vary up to 3.5x from the previous bit rate (see the *Changing the Data Frequency* section). Figure 13 shows the UART protocol for writing and reading in base mode between the µC and the serializer/deserializer.

Figure 14 shows the UART data format. Even parity is used. Figures 15 and 16 detail the formats of the SYNC byte (0x79) and ACK byte (0xC3). The  $\mu$ C and the connected slave chip generate the SYNC byte and ACK



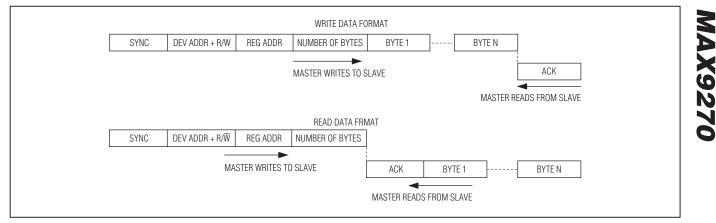


Figure 13. UART Protocol for Base Mode

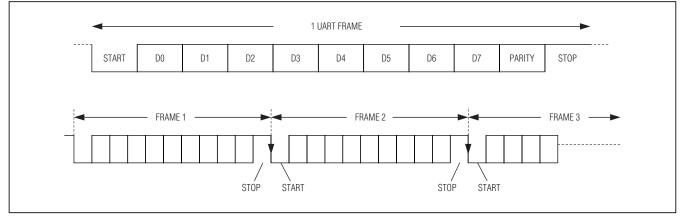


Figure 14. UART Data Format for Base Mode



Figure 15. SYNC Byte (0x79)

byte, respectively. Certain events such as device wakeup and interrupt generate signals on the control path and should be ignored by the  $\mu$ C. All data written to the internal registers do not take affect until after the acknowledge byte is sent. This allows the  $\mu$ C to verify that write commands are processed without error, even if the result of the write command directly affects the serial link. The slave uses the SYNC byte to synchronize with the host UART data rate automatically. If the INT or MS inputs of the device toggle while there is control-channel communication, the control-channel communication can be corrupted. In the event of a missed acknowledge, the

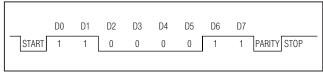


Figure 16. ACK Byte (0xC3)

 $\mu$ C should assume there was an error in the packet when the slave device receives it, or that an error occurred during the response from the slave device. In base mode, the  $\mu$ C must keep the UART Tx/Rx lines high for 16 bit times before starting to send a new packet.

As shown in Figure 17, the remote-side device converts the packets going to or coming from the peripherals from the UART format to the I<sup>2</sup>C format and vice versa. The remote device removes the byte number count and adds or receives the ACK between the data bytes of I<sup>2</sup>C. The I<sup>2</sup>C's data rate is the same as the UART data rate.



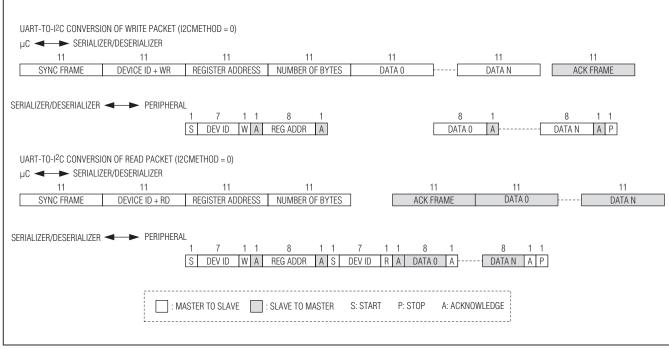


Figure 17. Format Conversion between UART and  $I^2C$  with Register Address (I2CMETHOD = 0)

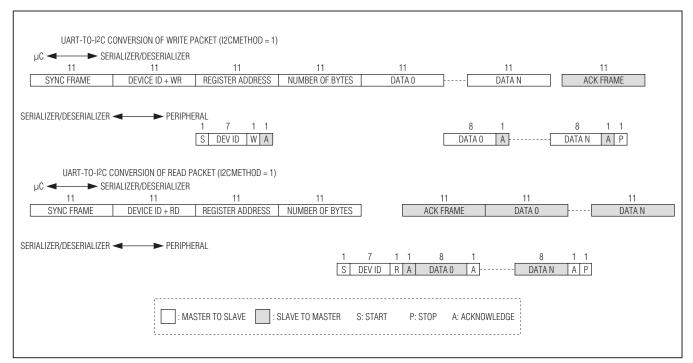


Figure 18. Format Conversion between UART and  $I^2C$  in Command-Byte-Only Mode (I2CMETHOD = 1)

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#### Interfacing Command-Byte-Only I<sup>2</sup>C Devices

The GMSL UART-to- $l^2C$  conversion interfaces with devices that do not require register addresses, such as the MAX7324 GPIO expander. Change the communication method of the  $l^2C$  master using the I2CMETHOD bit. I2CMETHOD = 1 sets command-byte-only mode, while I2CMETHOD = 0 sets normal mode where the first byte in the data stream is the register address. In this mode, the  $l^2C$  master ignores the register address byte and directly reads/writes the subsequent data bytes (Figure 18).

#### Interrupt Control

The INT of the serializer is the interrupt output and the INT of the deserializer is the interrupt input. The interrupt output on the serializer follows the transitions at the interrupt input of the deserializer. This interrupt function supports remote-side functions such as touch-screen peripherals,

BOOST SETTING (0x05 D[3:0])	TYPICAL BOOST GAIN (dB)		
0000	2.1		
0001	2.8		
0010	3.4		
0011	4.2		
	5.2		
0100	Power-up default (EQS = high)		
0101	6.2		
0110	7		
0111	8.2		
1000	9.4		
1001	10.7 Power-up default (EQS = low)		
1010	11.7		
1011	13		

#### Table 5. Cable Equalizer Boost Levels

#### Table 6. Parallel Output Spread

remote power-up, or remote monitoring. Interrupts that occur during periods where the reverse control channel is disabled, such as link startup/shutdown, are automatically resent once the reverse control channel becomes available again. Bit D4 of register 0x06 in the deserializer also stores the interrupt input state. Writing to the SETINT register bit also sets the INT output of the serializer. In addition, the  $\mu$ C sets the INT output of the serializer by writing to the SETINT register bit. In normal operation, the state of the interrupt output changes when the interrupt input on the deserializer toggles.

#### Line Equalizer

The deserializer includes an adjustable line equalizer to further compensate cable attenuation at high frequencies. The cable equalizer has 11 selectable levels of compensation from 2.1dB to 13dB (Table 5). The EQS input selects the default equalization level at power-up. The state of EQS is latched upon power-up or when resuming from power-down mode. To select other equalization levels, set the corresponding register bits in the deserializer (0x05 D[3:0]). Use equalization in the deserializer, together with preemphasis in the serializer to create the most reliable link for a given cable.

#### Spread Spectrum

To reduce the EMI generated by the transitions on the serial link and parallel outputs, the deserializer supports spread spectrum. Turning on spread spectrum on the deserializer spreads the parallel video outputs. Do not enable spread spectrum for both the serializer and deserializer. The two selectable spread-spectrum rates at the parallel outputs are  $\pm 2\%$  and  $\pm 4\%$  (Table 6).

Set the SSEN input high to select 2% spread at power-up and SSEN input low to select no spread at power-up. The state of SSEN is latched upon power-up or when resuming from power-down mode.

Turning on spread spectrum does not affect the audio data stream. Changes in the serializer spread settings only affect MCLK output if it is derived from PCLK\_(MCLKSRC = 0).

SS	SPREAD (%)			
00	No spread spectrum. Power-up default when SSEN = low.			
01	±2% spread spectrum. Power-up default when SSEN = high.			
10	No spread spectrum.			
11	±4% spread spectrum.			

The device includes a sawtooth divider to control the spread-modulation rate. Autodetection or manual programming of the PCLK\_ operation range guarantees a spread-spectrum modulation frequency within 20kHz to 40kHz. Additionally, manual configuration of the sawtooth divider (SDIV, 0x03 D[5:0]) allows the user to set a specific modulation frequency for a specific PCLK\_ rate. Always keep the modulation frequency between 20kHz to 40kHz to ensure proper operation.

#### Manual Programming of the Spread-Spectrum Divider

The modulation rate relates to the PCLK\_ frequency as follows:

$$f_M = (1 + DRS) \frac{f_{PCLK}}{MOD \times SDIV}$$

where:

 $f_M$  = Modulation frequency.

DRS = DRS pin input value (0 or 1).

 $f_{PCLK}$  = Parallel clock frequency (12.5MHz to 104MHz).

MOD = Modulation coefficient given in Table 7.

SDIV = 5-bit SDIV setting, manually programmed by the  $\mu$ C.

To program the SDIV setting, first look up the modulation coefficient according to the part number and desired bit-width and spread-spectrum settings. Solve the above equation for SDIV using the desired parallel clock and modulation frequencies. If the calculated SDIV value is larger than the maximum allowed SDIV value in Table 7, set SDIV to the maximum value.

**Sleep Mode** The deserializer includes a low-power sleep mode to reduce power consumption when it is not attached to the  $\mu$ C LCD applications. Set the SLEEP bit to 1 to initiate sleep mode. The deserializer sleeps after serial link inactivity or 8ms (whichever arrives first) after setting its SLEEP = 1. See the *Link Startup Procedure* section for details on waking up the device for different  $\mu$ C and starting conditions. The  $\mu$ C side device cannot enter into sleep mode, and its SLEEP bit remains at 0. Use the  $\overline{PWDN}$  input pin to bring the  $\mu$ C side device into a low-power state.

#### **Configuration Link Mode**

The GMSL includes a low-speed configuration link to allow control-data connection between the two devices in the absence of a valid parallel clock input. In either display or camera applications, the configuration link can be used to program equalizer/preemphasis or other registers before establishing the video link. An internal oscillator provides PCLK\_ for establishing the serial configuration link between the serializer and deserializer. The parallel output clock and data lines are disabled in the deserializer. The LOCK output remains low even after a successful configuration link lock. Set CLINKEN = 1 on the serializer to turn on the configuration link. The configuration link remains active as long as the video link has not been enabled. The video link overrides the configuration link and attempts to lock when SEREN = 1.

### Link Startup Procedure

Table 8 lists four startup cases for video-display applications. Table 9 lists two startup cases for image-sensing applications. In either display or image-sensing applications, the control link is always available after the highspeed data link or the configuration link is established and the GMSL registers or the peripherals are ready for programming.

#### **Video-Display Applications**

For the video-display application, with a remote display unit, connect the  $\mu$ C to the serializer and set CDS = low for both the serializer and deserializer. Table 8 summarizes the four startup cases based on the settings of  $\overline{\text{AUTOS}}$  and MS.

#### Case 1: Autostart Mode

After power-up or when PWDN transitions from low to high for both the serializer and deserializer, the serial link establishes if a stable PCLK\_ is present. The serializer locks to PCLK\_ and sends the serial data to the deserializer. The deserializer then detects activity on the serial link and locks to the input serial data.

### Table 7. Modulation Coefficients and Maximum SDIV Settings

SPREAD-SPECTRUM SETTING (%)	MODULATION COEFFICIENT (decimal)	SDIV UPPER LIMIT (decimal)
4	208	15
2	208	30

CASE	AUTOS (SERIALIZER)	SERIALIZER POWER-UP STATE	MS (DESERIALIZER)	DESERIALIZER POWER-UP STATE	LINK STARTUP MODE
1	Low	Serialization enabled	Low	Normal (SLEEP = 0)	Both devices power up with the serial link active (autostart).
2	High	Serialization disabled	High	Sleep mode (SLEEP = 1)	Serial link is disabled and the deserializer powers up in sleep mode. Set SEREN = 1 or CLINKEN = 1 in the serializer to start the serial link and wake up the deserializer.
3	High	Serialization disabled	Low	Normal (SLEEP = 0)	Both devices power up in nor- mal mode with the serial link disabled. Set SEREN = 1 or CLINKEN = 1 in the serializer to start the serial link.
4	Low	Serialization enabled	High	Sleep mode (SLEEP = 1)	The deserializer starts in sleep mode. Link autostarts upon the serializer power-up. Use this case when the deserializer pow- ers up before the serializer.

### Table 8. Startup Selection for Video-Display Applications (CDS = Low)

#### Case 2: Standby Start Mode

After power-up, or when PWDN transitions from low to high for both the serializer and deserializer, the deserializer starts up in sleep mode, and the serializer stays in standby mode (does not send serial data). Use the  $\mu$ C and program the serializer to set SEREN = 1 to establish a video link or CLINKEN = 1 to establish the configuration link. After locking to a stable PCLK\_ (for SEREN = 1) or the internal oscillator (for CLINKEN = 1), the serializer sends a wake-up signal to the deserializer. The deserializer exits sleep mode after locking to the serial data and sets SLEEP = 0. If after 8ms the deserializer does not lock to the input serial data, the deserializer goes back to sleep, and the internal sleep bit remains uncleared (SLEEP = 1).

#### Case 3: Remote Side Autostart Mode

After power-up, or when  $\overline{PWDN}$  transitions from low to high, the remote device (deserializer) starts up and tries to lock to an incoming serial signal with sufficient power. The host side (serializer) is in standby mode and does not try to establish a link. Use the  $\mu C$  and program the serializer to set SEREN = 1 (and apply a stable PCLK\_) to establish a video link, or CLINKEN = 1 to establish the configuration link. In this case, the deserializer ignores the short wake-up signal sent from the serializer.

#### Case 4: Remote Side in Sleep Mode

After power-up or when PWDN transitions from low to high, the remote device (deserializer) starts up in sleep mode. The high-speed link establishes automatically after the serializer powers up with a stable PCLK\_ and sends a wake-up signal to the deserializer. Use this mode in applications where the deserializer powers up before the serializer.

#### **Image-Sensing Applications**

For image-sensing applications, with remote camera unit(s), connect the  $\mu$ C to the deserializer and set CDS = high for both the serializer and deserializer. The deserializer powers up normally (SLEEP = 0) and continuously tries to lock to a valid serial input. Table 9 summarizes the two startup cases, based on the state of the serializer AUTOS pin.

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### Table 9. Startup Selection for Image-Sensing Applications (CDS = High)

CASE	AUTOS (SERIALIZER)	SERIALIZER POWER-UP STATE	DESERIALIZER POWER-UP STATE	LINK STARTUP MODE
1	Low	Serialization enabled	Normal (SLEEP = 0)	Autostart.
2	High	Sleep mode (SLEEP = 1)	Normal (SLEEP = 0)	Serializer is in sleep mode. Wake up the serializer through the control channel ( $\mu$ C attached to the deserializer).

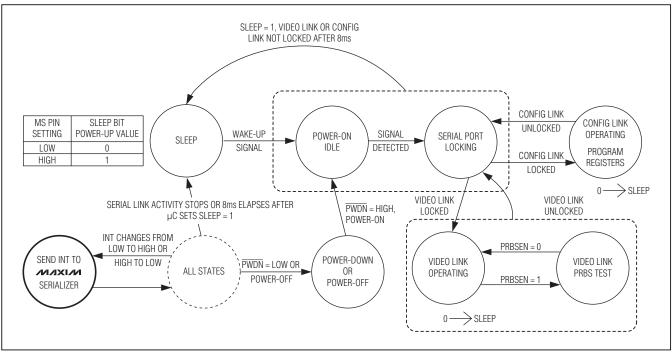


Figure 19. State Diagram, CDS = Low (LCD Application)

#### Case 1: Autostart Mode

After power-up, or when PWDN transitions from low to high, the serializer locks to a stable PCLKIN and sends the high-speed data to the deserializer. The deserializer locks to the serial data and outputs the parallel video data and PCLKOUT.

#### Case 2: Sleep Mode

After power-up, or when  $\overline{\text{PWDN}}$  transitions from low to high, the serializer starts up in sleep mode. To wake up the serializer, use the  $\mu$ C to send a regular UART frame

containing at least three rising edges (e.g., 0x66), at a bit rate no greater than 1Mbps. The low-power wake-up receiver of the serializer detects the wake-up frame over the reverse control channel and powers up. Reset the sleep bit (SLEEP = 0) of the serializer using a regular control-channel write packet to power up the device fully. Send the sleep bit write packet at least 500 $\mu$ s after the wake-up frame. The serializer goes back to sleep mode if its sleep bit is not cleared within 8ms (typ) after detecting a wake-up frame.



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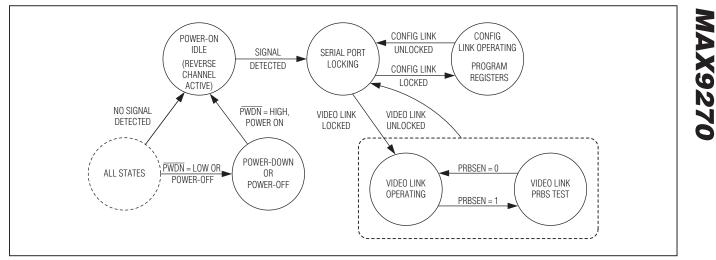


Figure 20. State Diagram, CDS = High (Camera Application)

#### Applications Information

#### **Error Checking**

The deserializer checks the serial link for errors and stores the number of detected decoding errors in the 8-bit register (DECERR, 0x0D). If a large number of decoding errors are detected within a short duration, the deserializer loses lock and stops the error counter. The deserializer then attempts to relock to the serial data. DECERR resets upon successful video link lock, successful readout of DECERR (through UART), or whenever auto-error reset is enabled. The deserializer does not check for decoding errors during the internal PRBS test and DECERR is reset to 0x00.

**ERR** Output The deserializer has an open-drain ERR output. This output asserts low whenever the number of decoding errors exceed the error threshold (ERRTHR, 0x0C) during normal operation, or when at least one PRBS error is detected during PRBS test. ERR reasserts high whenever DECERR (0x0D) resets, due to DECERR readout, video link lock, or autoerror reset.

#### Autoerror Reset

The default method to reset errors is to read the respective error registers in the deserializer (0x0D, 0x0E). Autoerror reset clears the decoding-error counter (DECERR) and the ERR output ~1 $\mu$ s after ERR goes low. Autoerror reset is disabled on power-up. Enable autoerror reset through AUTORST (0x06 D6). Autoerror reset does not run when the device is in PRBS test mode.

#### Self PRBS Test

The GMSL link includes a PRBS pattern generator and bit-error verification function. Set PRBSEN = 1 (0x04 D5) first in the serializer and then the deserializer to start the PRBS test. Set PRBSEN = 0 (0x04 D5) first in the deserializer and then the serializer to exit the PRBS self test. The deserializer uses an 8-bit register (0x0E) to count the number of detected errors. The control link also controls the start and stop of the error counting. During PRBS mode, the device does not count decoding errors and the ERR output reflects PRBS errors only. Autoerror reset does not run when the device is in PRBS mode.

### Microcontrollers on Both Sides of the GMSL Link (Dual $\mu$ C Control)

Usually the  $\mu$ C is either on the serializer side for videodisplay applications, or on the deserializer side for image-sensing applications. For the former case, both the CDS pins are set to low, and for the latter case, the CDS pins are set to high. However, if the CDS pin of the serializer is low and the CDS pin of the deserializer is high, then the serializer/deserializer can both connect to  $\mu$ Cs simultaneously. In such a case, the  $\mu$ Cs on either side can communicate with the GMSL UART protocol.

Contentions of the control link may happen if the  $\mu$ Cs on both sides are using the link at the same time. The GMSL does not provide the solution for contention avoidance. The serializer/deserealizer do not send an acknowledge frame when communication fails due to contention. Users can always implement a higher-layer protocol to

