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Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

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Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China





Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

MAXQ7670

General Description

The MAXQ7670 is a highly integrated solution for measuring multiple analog signals and outputting the results on a control area network (CAN) bus. The device operates from a single 5V supply and incorporates a high-performance, 16-bit reduced instruction set computing (RISC) core, a SAR ADC, and a CAN 2.0B controller, supporting transfer rates up to 1Mbps. The 10-bit SAR ADC includes an amplifier with programmable gains of 1V/V or 16V/V, 8 input channels, and conversion rates up to 250ksps. The eight single-ended ADC inputs can be configured as four unipolar or bipolar, fully differential inputs. For single-supply operation, the external 5V supply powers the digital I/Os and two separate integrated linear regulators that supply the 2.5V digital core and the 3.3V analog circuitry. Each supply rail has a dedicated power-supply supervisor that provides brownout detection and power-on reset (POR) functions. The 16-bit RISC microcontroller (μ C) includes 64KB (32K x 16) of non-volatile program/data flash and 2KB (1K x 16) of data RAM. Other features of the MAXQ7670 include a 4-wire SPI™ interface, a JTAG interface for in-system programming and debugging, an integrated 15MHz RC oscillator, external crystal oscillator support, a timer/counter with pulse-width modulation (PWM) capability, and seven GPIO pins with interrupt and wake-up capability.

The system-on-a-chip (SoC) MAXQ7670 is a μ C-based, smart data acquisition system. As a member of the MAXQ® family of 16-bit, RISC μ Cs, the MAXQ7670 is ideal for low-cost, low-power, embedded-applications such as automotive, industrial controls, and building automation. The flexible, modular architecture used in the MAXQ μ Cs allows development of targeted products for specific applications with minimal effort.

The MAXQ7670 is available in a 40-pin, 5mm x 5mm TQFN package, and is specified to operate over the -40°C to +125°C automotive temperature range.

Applications

Automotive Steering Angle and Torque Sensors
CAN-Based Automotive Sensor Applications
Industrial Control
Building Automation

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Note: Some revisions of this device may incorporate deviations from published specifications known as errata. Multiple revisions of any device may be simultaneously available through various sales channels. For information about device errata, go to:

<http://www.maxim-ic.com/errata>



Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

Features

- ◆ **High-Performance, Low-Power, 16-Bit RISC Core**
0.166MHz to 16MHz Operation, Approaching 1MIPs/MHz
Low Power (< 1mA/MIPS, $V_{DD} = +2.5V$)
16-Bit Instruction Word, 16-Bit Data Bus
33 Instructions, Most Require Only One Clock Cycle
16-Level Hardware Stack
16 x 16-Bit, General-Purpose Working Registers
Three Independent Data Pointers with Auto-Increment/Decrement
Low-Power, Divide-by-256, Power-Management Modes (PMM) and Stop Mode
- ◆ **Program and Data Memory**
64KB Internal Nonvolatile Program/Data Flash
2KB Internal Data RAM
- ◆ **SAR ADC**
8 Single-Ended/4 Differential Channels,
10-Bit Resolution with No Missing Codes
PGA Gain = 1V/V or 16V/V
250ksps (150.9ksps with PGA Gain = 16V/V)
- ◆ **Timer/Digital I/O Peripherals**
CAN 2.0B Controller (15 Message Centers)
Serial Peripheral Interface (SPI)
JTAG Interface (Extensive Debug and Emulation Support)
Single 16-Bit/Dual 8-Bit Timer/PWM
Seven General-Purpose, Digital I/O Pins with External Interrupt/Wake-Up Features
- ◆ **Oscillator/Clock Module**
Internal Oscillator Supports External Crystal (8MHz or 16MHz)
Integrated 15MHz RC Oscillator
External Clock Source Operation
Programmable Watchdog Timer
- ◆ **Power-Management Module**
Power-On Reset
Power-Supply Supervisor/Brownout Detection
Integrated +2.5V and +3.3V Linear Regulators

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAXQ7670ATL/V+	-40°C to +125°C	40 TQFN-EP*

V denotes an automotive qualified part.

+Denotes a lead-free/RoHS-compliant package.

*EP = Exposed pad.

Typical Application Circuit and Pin Configuration appear at end of data sheet.

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ABSOLUTE MAXIMUM RATINGS

DVDD to DGND	-0.3V to +3V	Continuous Power Dissipation (T _A = +70°C)	
DVDDIO to GNDIO	-0.3V to +5.5V	40-Pin TQFN (derate 36mW/°C above +70°C)	2857mW
AVDD to AGND	-0.3V to +4V	Continuous Current into Any Pin.....	±50mA
DGND to GNDIO	-0.3V to +0.3V	Operating Temperature Range	-40°C to +125°C
GNDIO to AGND	-0.3V to +0.3V	Junction Temperature.....	+150°C
AGND to DGND.....	-0.3V to +0.3V	Storage Temperature Range	-65°C to +150°C
Analog Inputs to AGND	-0.3V to (V _{AVDD} + 0.3V)	Lead Temperature (soldering, 10s)	+300°C
RESET, Digital Inputs/Outputs to			
GNDIO	-0.3V to (V _{DVDDIO} + 0.3V)		
XIN, XOUT to DGND	-0.3V to (V _{DVDD} + 0.3V)		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{DVDDIO} = +5.0V, V_{AVDD} = +3.3V, V_{DVDD} = +2.5V, V_{REFADC} = +3.3V, system clock = 16MHz. T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER REQUIREMENTS						
Supply Voltage Ranges	DVDD	REGEN ² = DVDDIO, DV _{DD} ≤ AV _{DD} , DV _{DD} ≤ DV _{DDIO}	2.25	2.5	2.75	V
	AVDD	LRAPD = 1, AV _{DD} ≤ DV _{DDIO}	3.0	3.3	3.6	
	DVDDIO		4.5	5.0	5.25	
AVDD Supply Current	I _{AVDD}	Shutdown (Note 2)		3	10	μA
		All analog functions enabled		6	7	mA
Analog Module Incremental Subfunction Supply Current	ΔI _{AVDD}	ADC, 50ksps, 4MHz ADCCLK		5200		μA
		ADC, 250ksps, 4MHz ADCCLK		5600		
		AVDD brownout interrupt monitor		3		
		PGA enabled		5500		
DVDD Supply Current	I _{DVDD}	CPU in stop mode, all peripherals disabled		25	200	μA
		High speed/2MHz mode (Note 3)		2.0	2.5	
		High speed/16MHz mode (Note 4)		11.3		
		Low speed/625kHz mode (Note 5)		0.95		
		Program flash erase or write		14	23	
Digital Peripheral Incremental Subfunction Supply Current	ΔI _{DVDD}	DVDDIO brownout reset monitor		1		μA
		HF crystal oscillator		60		
		Internal fixed-frequency oscillator		50		
DVDDIO Supply Current	I _{DVDDIO}	All digital I/Os static at GNDIO or DV _{DDIO}		2	20	μA
		CAN transmitting, timer output switching (Note 6)		0.2	0.3	mA

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DDIO} = +5.0V$, $V_{AVDD} = +3.3V$, $V_{DVDD} = +2.5V$, $V_{REFADC} = +3.3V$, system clock = 16MHz. $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
MEMORY SECTION						
Flash Memory Size		Program or data storage		64		KB
Flash Page Size		16-bit word size		256		Words
Flash Erase/Write Endurance		Program or data (Note 7)	10,000			Cycles
Flash Data Retention (Note 7)		All flash, $T_A = +25^\circ C$	100			Years
		All flash, $T_A = +85^\circ C$	15			
Flash Erase Time		Flash page erase	20		50	ms
		Entire flash mass erase	200		500	
Flash Programming Time		Flash single word programming	20		40	μs
		Entire flash programming	0.66		1.31	s
RAM Memory Size				2		KB
Utility ROM Size		16-bit word size		4		KWords
ANALOG SENSE PATH (Includes PGA and ADC)						
Resolution	NADC	No missing codes	10			Bits
Integral Nonlinearity	INL _{ADC}	PGA gain = 16V/V, bipolar mode, $V_{IN} = \pm 100mV$, 150.9ksps		± 0.5	± 1	LSB ₁₀
		PGA gain = 1V/V, unipolar mode, $V_{IN} = +1.0V$, 250ksps		± 0.4	± 1	
Differential Nonlinearity	DNL _{ADC}	PGA gain = 1V/V or 16V/V		± 0.4	± 1	LSB ₁₀
Input-Referred Offset Error		Test at $T_A = +25^\circ C$, PGA gain = 1V/V or 16V/V		± 1	± 10	mV
Offset-Error Temperature Coefficient		PGA gain = 16V/V, bipolar mode		± 2		$\mu V/^\circ C$
Gain Error		PGA gain = 16V/V, bipolar mode, excludes offset and reference error, test at $T_A = +25^\circ C$	-2		+2	%
Gain-Error Temperature Coefficient		PGA gain = 16V/V, bipolar mode		± 5		ppm/ $^\circ C$
Conversion Clock Frequency	f _{ADCCLK}	f _{SYSCLK} = 8MHz or 16MHz	0.5		4.0	MHz
Sample Rate	f _{SAMPLE}	PGA gain = 16V/V, f _{ADCCLK} = 4MHz			150.9	ksps
		PGA gain = 1V/V, f _{ADCCLK} = 4MHz			250	
Channel Select, Track-and-Hold Acquisition Time	t _{ACQ}	PGA gain = 16V/V, 13.5 ADCCLK cycles at 4MHz		3.375		μs
		PGA gain = 1V/V, three ADCCLK cycles at 4MHz		0.75		
Conversion Time	t _{CONV}	13 ADCCLK cycles at 4MHz		3.25		μs

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DDIO} = +5.0V$, $V_{AVDD} = +3.3V$, $V_{DVDD} = +2.5V$, $V_{REFADC} = +3.3V$, system clock = 16MHz. $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Channel Select Plus Conversion Time	$t_{ACQ} + t_{CONV}$	PGA gain = 16V/V, 26.5 ADCCLK cycles at 4MHz		6.625		μs
		PGA gain = 1V/V, 16 ADCCLK cycles at 4MHz		4		
Turn-On Time	t_{RECOV}			10		μs
Aperture Delay				60		ns
Aperture Jitter				100		pSP-P
Differential Input Voltage Range		At AIN0–AIN7, unipolar mode, PGA gain = 1V/V	0		V_{REFADC}	V
		At AIN0–AIN7, unipolar mode, PGA gain = 16V/V	0		0.125	
		At AIN0–AIN7, bipolar mode, PGA gain = 1V/V	$-V_{REFADC}/2$		$+V_{REFADC}/2$	
		At AIN0–AIN7, bipolar mode, PGA gain = 16V/V	$-V_{REFADC}/32$		$+V_{REFADC}/32$	
Absolute Input Voltage Range		At AIN0–AIN7	0		V_{AVDD}	V
Input Leakage Current		At AIN0–AIN7		± 0.1		μA
Input-Referred Noise		At AIN0–AIN7, PGA gain = 16V/V		50		μV_{RMS}
		At AIN0–AIN7, PGA gain = 1V/V		400		
Small-Signal Bandwidth (-3dB)		$V_{IN} = 12mV_{P-P}$, PGA gain = 16V/V		33		MHz
		$V_{IN} = 200mV_{P-P}$, PGA gain = 1V/V		23		
Large-Signal Bandwidth (-3dB)		$V_{IN} = 150mV_{P-P}$, PGA gain = 16V/V		33		MHz
		$V_{IN} = 2.5V_{P-P}$, PGA gain = 1V/V		19		
Input Capacitance (Note 8)		Single-ended, any AIN0–AIN7, PGA gain = 16V/V		16		μF
		Single-ended, any AIN0–AIN7, PGA gain = 1V/V		13		
Input Common-Mode Rejection Ratio	CMRR	AIN0–AIN7, $V_{CM} =$ differential input range		75		dB
Power-Supply Rejection Ratio	PSRR	$AV_{DD} = 3.0V$ to $3.6V$		90		dB
EXTERNAL REFERENCE INPUTS						
REFADC Input Voltage Range			1.0	3.3	V_{AVDD}	V
REFADC Leakage Current		ADC disabled		1		μA
Input Capacitance		(Note 9)		20		μF
+3.3V (AVDD) LINEAR REGULATOR						
AVDD Output Voltage		LRAPD = 0	3.15	3.3	3.45	V
No-Load Quiescent Current		LRAPD = 0, all internal analog peripherals disabled		10		μA

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DVDDIO} = +5.0V$, $V_{AVDD} = +3.3V$, $V_{DVDD} = +2.5V$, $V_{REFADC} = +3.3V$, system clock = 16MHz. $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Current Capability		LRAPD = 0	50			mA
Output Short-Circuit Current		LRAPD = 0, AVDD shorted to AGND		100		mA
Maximum AVDD Bypass Capacitor to AGND		LRAPD = 0		0.47		μF
+2.5V (DVDD) LINEAR REGULATOR						
DVDD Output Voltage		$\overline{REGEN2} = GNDIO$	2.38	2.5	2.62	V
No-Load Quiescent Current		$\overline{REGEN2} = GNDIO$, all internal digital peripherals disabled		15		μA
Output Current Capability		$\overline{REGEN2} = GNDIO$	50			mA
Output Short-Circuit Current		$\overline{REGEN2} = GNDIO$, DVDD shorted to DGND		100		mA
Maximum DVDD Bypass Capacitor to DGND		$\overline{REGEN2} = GNDIO$		0.47		μF
SUPPLY-VOLTAGE SUPERVISORS AND BROWNOUT DETECTION						
DVDD Reset Threshold		Asserts \overline{RESET} if V_{DVDD} is below this threshold	2.1		2.25	V
DVDD Interrupt Threshold		Generates an interrupt if V_{DVDD} falls below this threshold	2.25		2.38	V
Minimum DVDD Interrupt and Reset Threshold Difference				0.14		V
AVDD Interrupt Threshold		Generates an interrupt if V_{AVDD} falls below this threshold	3.0		3.15	V
DVDDIO Interrupt Threshold		Generates an interrupt if V_{DVDDIO} falls below this threshold	4.5		4.75	V
Operational Range		DVDD	1		2.75	V
		AVDD	1		3.6	
		DVDDIO	1		5.25	
Supervisor Hysteresis				± 0.7		%
CAN INTERFACE						
CAN Baud Rate		$f_{CANCLK} = 8MHz$			1	Mbps
CANCLK Mean Frequency Error		8MHz or 16MHz, 50ppm external crystal		60		ppm
CANCLK Total Frequency Error		8MHz or 16MHz, 50ppm external crystal; measured over a 12ms interval; mean plus peak cycle jitter		< 0.5		%

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

ELECTRICAL CHARACTERISTICS (continued)

($V_{DVDDIO} = +5.0V$, $V_{AVDD} = +3.3V$, $V_{DVDD} = +2.5V$, $V_{REFADC} = +3.3V$, system clock = 16MHz. $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
HIGH-FREQUENCY CRYSTAL OSCILLATOR						
Clock Frequency		Using external crystal		8 or 16	16	MHz
		External input (Note 10)	0.166		16.67	
Stability		Excluding crystal drift		25		ppm
Startup Time		f _{SYCLK} cycles		65,535		Cycles
XIN Input Low Voltage		Driven with external clock source			0.3 x V_{DVDD}	V
XIN Input High Voltage		Driven with external clock source	0.7 x V_{DVDD}			V
INTERNAL FIXED-FREQUENCY OSCILLATOR						
Frequency	f _{IFFCLK}	$T_A = T_{MIN}$ to T_{MAX}	13.8	15	16.35	MHz
Tolerance		$T_A = +25^\circ C$		±0.4		%
Temperature Drift		$T_A = T_{MIN}$ to T_{MAX}		5		%
Power-Supply Rejection		$T_A = +25^\circ C$, $DVDD = 2.25V$ to $2.75V$		±1.5		%
RESET (\overline{RESET})						
\overline{RESET} Internal Pullup Resistance		Pulled up to $DVDDIO$		55		k Ω
\overline{RESET} Output Low Voltage		\overline{RESET} asserted, no external load			0.4	V
\overline{RESET} Output High Voltage		\overline{RESET} deasserted, no external load	0.9 x V_{DVDDIO}			V
\overline{RESET} Input Low Voltage		Driven with external clock source			0.3 x V_{DVDD}	V
\overline{RESET} Input High Voltage		Driven with external clock source	0.7 x V_{DVDDIO}			V
DIGITAL INPUTS ($P0_$, $CANRXD$, $MISO$, $MOSI$, \overline{SS}, $SCLK$, TCK, TDI, TMS)						
Input Low Voltage					0.8	V
Input High Voltage			2.1			V
Input Hysteresis				500		mV
Input Leakage Current		$V_{IN} = GNDIO$ or V_{DVDDIO} , pullup disabled	-10	±0.01	+10	μA
Input Pullup Resistance				55		k Ω
Input Pulldown Resistance				55		k Ω
Input Capacitance				15		pF
DIGITAL OUTPUTS ($P0_$, $CANTXD$, $MOSI$, $SCLK$, \overline{SS}, TDO)						
Output Low Voltage		$I_{SINK} = 0.5mA$			0.4	V
Output High Voltage		$I_{SOURCE} = 0.5mA$	$V_{DVDDIO} - 0.5$			V

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DDIO} = +5.0V$, $V_{AVDD} = +3.3V$, $V_{DVDD} = +2.5V$, $V_{REFADC} = +3.3V$, system clock = 16MHz. $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Capacitance		I/O pins three-state		15		pF
Maximum Output Impedance		PDO._ = 0		880		Ω
		PDO._ = 1		450		
SYSTEM CLOCK						
System Clock Frequency	f_{SYSCLK}	From any clock source	0		16.67	MHz
SPI INTERFACE TIMING						
SPI Master Operating Frequency	f_{MCLK}	$0.5 \times f_{SYSCLK}$			8	MHz
SPI Slave Mode Operating Frequency	f_{SCLK}				$f_{SYSCLK}/8$	MHz
SCLK Output Pulse-Width High/Low	t_{MCH} , t_{MCL}		t_{SYSCLK} - 25			ns
SCLK Input Pulse-Width High/Low	t_{SCH} , t_{SCL}			t_{SYSCLK}		ns
MOSI Output Hold Time After SCLK Sample Edge	t_{MOH}		t_{SYSCLK} - 25			ns
MOSI Output Setup Time to SCLK Sample Edge	t_{MOS}		t_{SYSCLK} - 25			ns
MISO Input Setup Time to SCLK Sample Edge	t_{MIS}		30			ns
MISO Input Hold Time After SCLK Sample Edge	t_{MIH}		0			ns
SCLK Inactive to MOSI Inactive	t_{MLH}		t_{SYSCLK} - 25			ns
MOSI Input Setup Time to SCLK Sample Edge	t_{SIS}		30			ns
MOSI Input Hold Time After SCLK Sample Edge	t_{SIH}		t_{SYSCLK} + 25			ns
MISO Output Valid After SCLK Shift Edge Transition	t_{SOV}				$3 t_{SYSCLK}$ + 25	ns
MISO Output Disabled After \overline{SS} Edge Rise	t_{SLH}				$2 t_{SYSCLK}$ + 50	ns
\overline{SS} Falling Edge to MISO Active	t_{SOE}		$2 t_{SYSCLK}$ + 2.5			ns

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DVDDIO} = +5.0V$, $V_{AVDD} = +3.3V$, $V_{DVDD} = +2.5V$, $V_{REFADC} = +3.3V$, system clock = 16MHz. $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
\overline{SS} Falling Edge to First SCLK Sample Edge	tSSE		2 tSYSCLK + 5			ns
SCLK Inactive to \overline{SS} Rising Edge	tSD		tSYSCLK + 10			ns
Minimum \overline{CS} Pulse Width	tscw		tSYSCLK + 10			ns

- Note 1:** All devices are 100% production tested at $T_A = +25^\circ C$ and $+125^\circ C$. Temperature limits to $T_A = -40^\circ C$ are guaranteed by design.
- Note 2:** All analog functions disabled and all digital inputs connected to supply or ground.
- Note 3:** High-speed/8 mode without CAN; $V_{DVDD} = +2.5V$, CPU and 16-bit timer running at 2MHz from an external, 16MHz crystal oscillator; all other peripherals disabled; all digital I/Os static at V_{DVDDIO} or GNDIO; $T_A = T_{MIN}$ to T_{MAX} .
- Note 4:** High-speed/1 mode with CAN; $V_{DVDD} = +2.5V$, CPU and 16-bit timer running at 16MHz from an external, 16MHz crystal oscillator; CAN enabled and communicating at 500kbps; all other peripherals disabled, all digital I/Os (except CANTXD and CANRXD) static at V_{DVDDIO} or GNDIO, $T_A = T_{MIN}$ to T_{MAX} .
- Note 5:** Low speed, PMM1 mode without CAN; $V_{DVDD} = +2.5V$, CPU and one timer running from an external, 16MHz crystal oscillator in PMM1 mode; all other peripherals disabled; all digital I/Os static at V_{DVDDIO} or GNDIO, $T_A = T_{MIN}$ to T_{MAX} .
- Note 6:** CAN transmitting at 500kbps; 16-bit timer output switching at 500kHz; all active I/Os are loaded with a 20pF capacitor; all remaining digital I/Os are static at V_{DVDDIO} or GNDIO, $T_A = T_{MIN}$ to T_{MAX} .
- Note 7:** Guaranteed by design and characterization.
- Note 8:** This is not a static capacitance. It is the capacitance presented to the analog input when the T/H amplifier is in sample mode.
- Note 9:** The switched capacitor on the REFADC input can disturb the reference voltage. To reduce this disturbance, place a 0.1 μF capacitor from REFADC to AGND as close as possible to REFADC.
- Note 10:** The digital design is fully static. However, the lower clock limit is set by a clock detect circuit. The MAXQ7670 switches to the internal RC clock if the external input goes below 166kHz. This clock detect circuit also acts to detect a crystal failure when a crystal is used.

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

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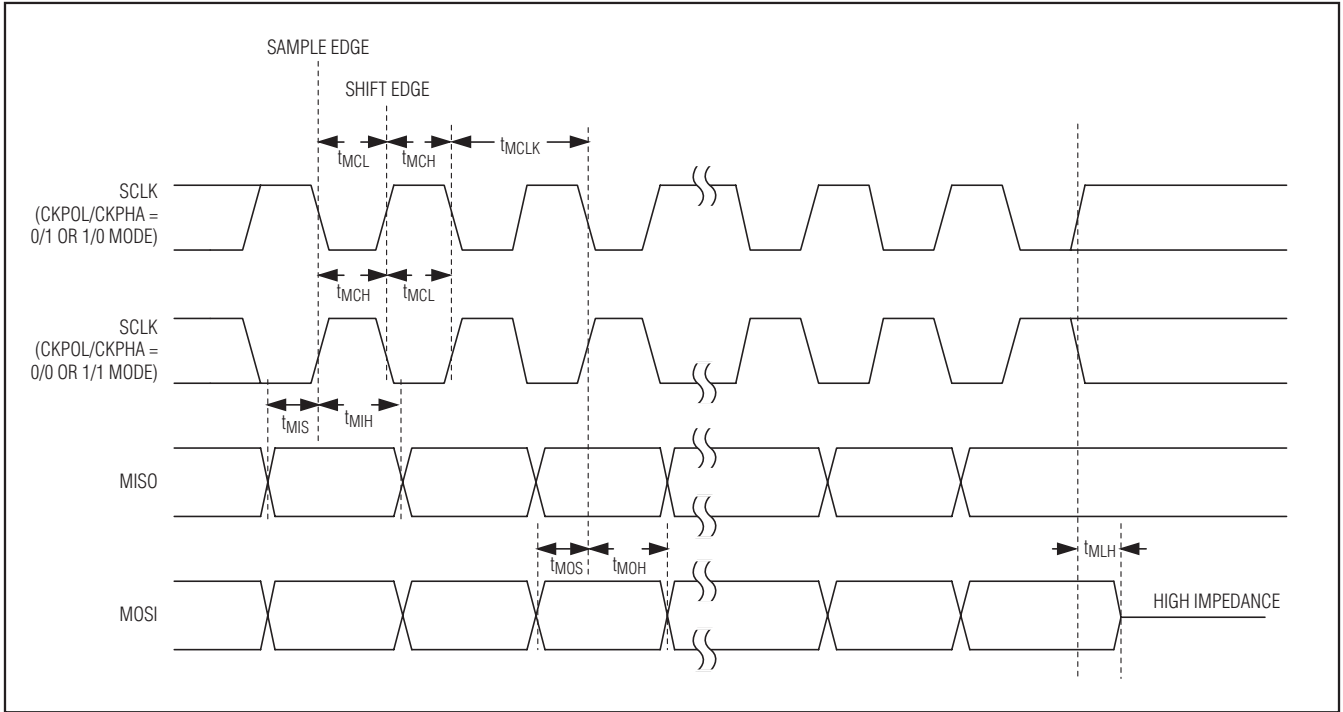


Figure 1. SPI Timing Diagram in Master Mode

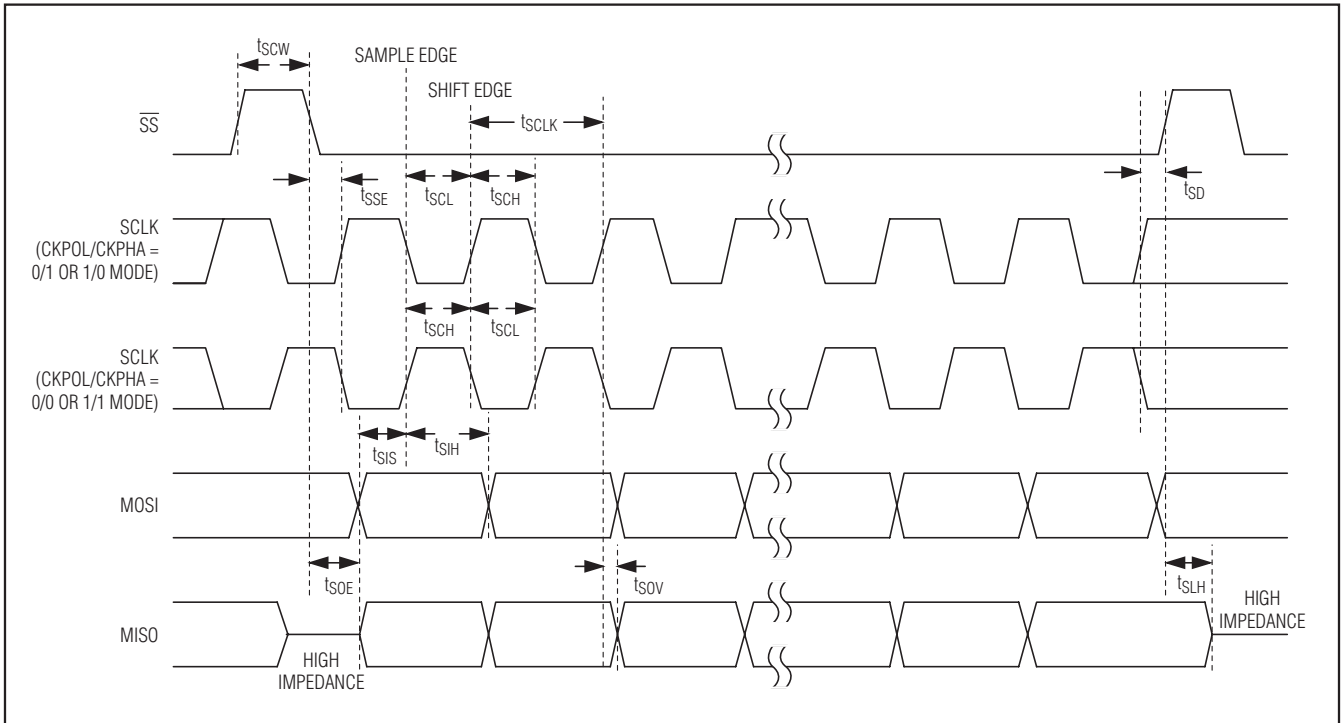
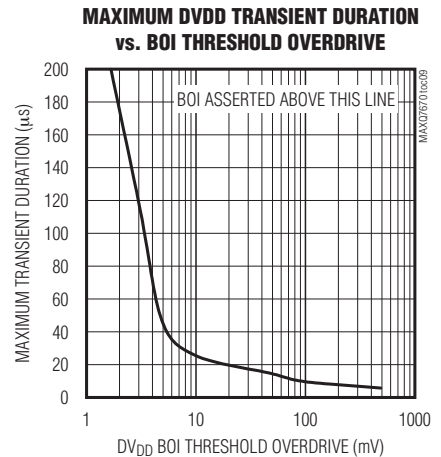
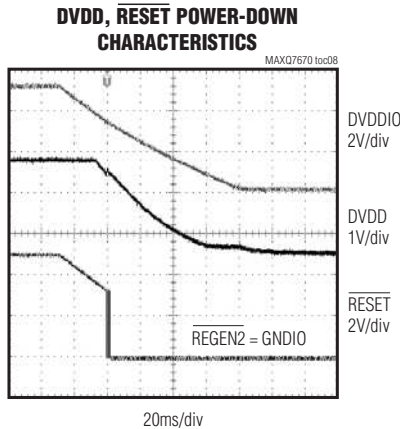
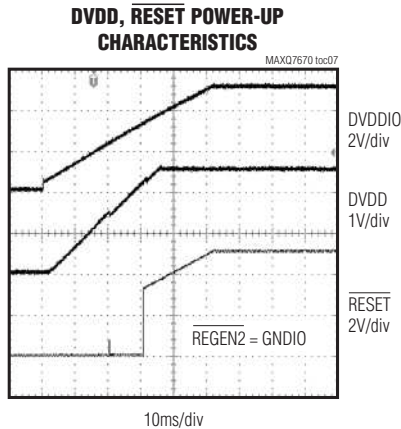
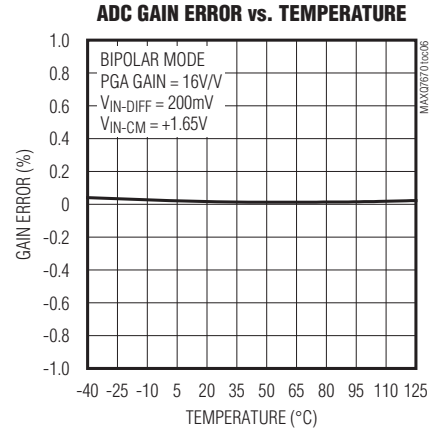
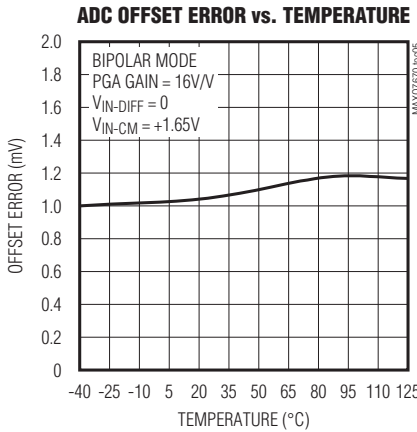
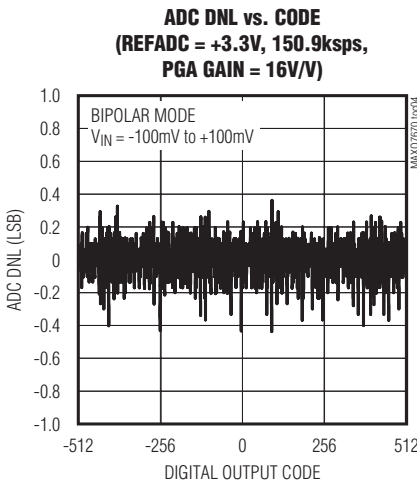
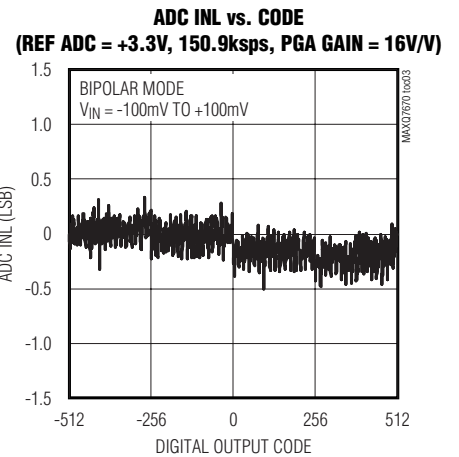
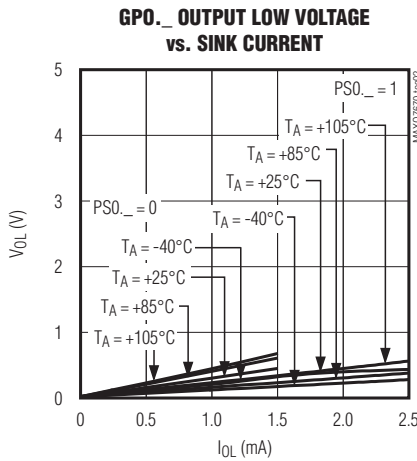
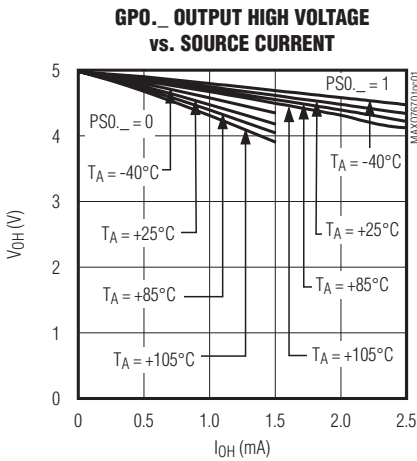


Figure 2. SPI Timing Diagram in Slave Mode

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

Typical Operating Characteristics

($V_{DVDDIO} = 5.0V$, $V_{AVDD} = 3.3V$, $V_{DVDD} = 2.5V$, $f_{SYSCLK} = 16MHz$, ADC resolution = 10 bits, $V_{REFDAC} = 3.3V$, $T_A = +25^\circ C$, unless otherwise noted.)



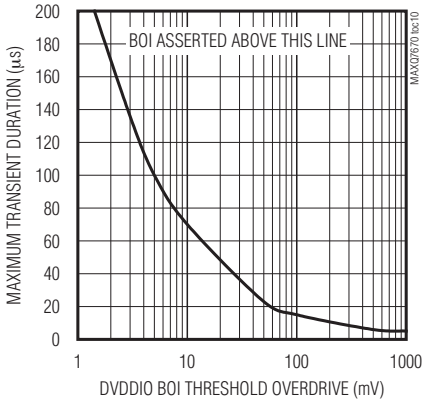
Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

MAXQ7670

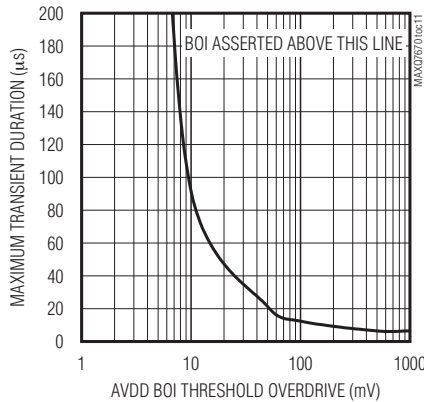
Typical Operating Characteristics (continued)

($V_{DVDDIO} = 5.0V$, $V_{AVDD} = 3.3V$, $V_{DVDD} = 2.5V$, $f_{SYSCLK} = 16MHz$, ADC resolution = 10 bits, $V_{REFDAC} = 3.3V$, $T_A = +25^\circ C$, unless otherwise noted.)

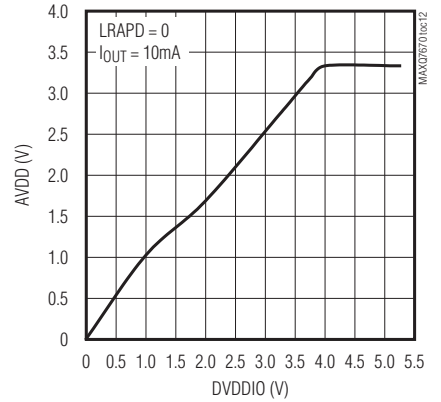
MAXIMUM DVDDIO TRANSIENT DURATION vs. BOI THRESHOLD OVERDRIVE



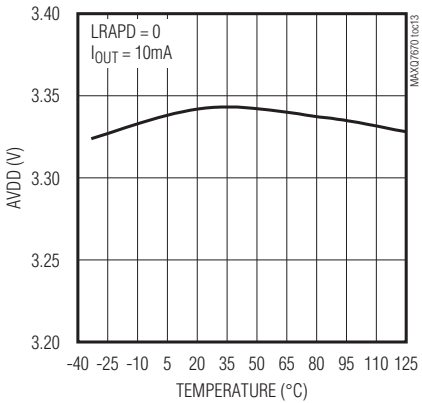
MAXIMUM AVDD TRANSIENT DURATION vs. BOI THRESHOLD OVERDRIVE



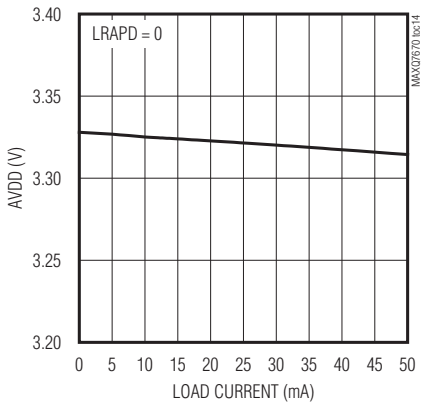
AVDD LINEAR REGULATOR OUTPUT VOLTAGE vs. DVDDIO SUPPLY VOLTAGE



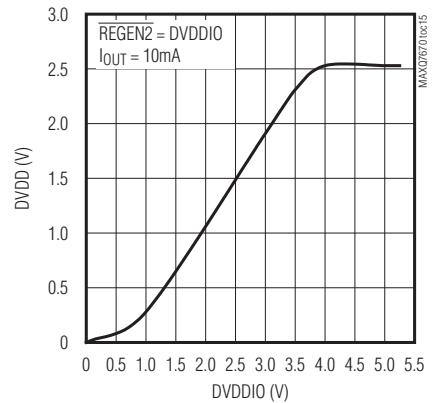
AVDD LINEAR REGULATOR OUTPUT VOLTAGE vs. TEMPERATURE



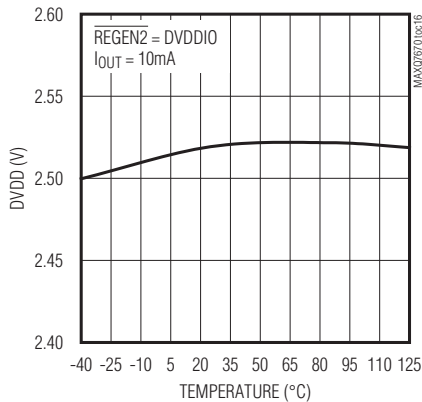
AVDD LINEAR REGULATOR OUTPUT VOLTAGE vs. LOAD CURRENT



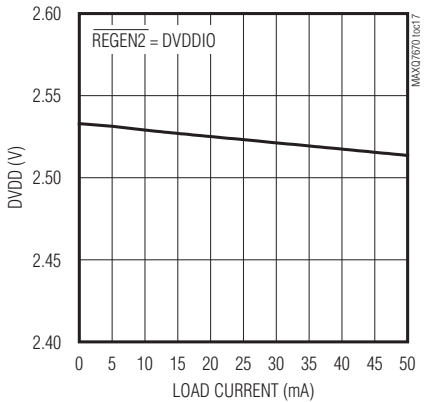
DVDD LINEAR REGULATOR OUTPUT VOLTAGE vs. DVDDIO SUPPLY VOLTAGE



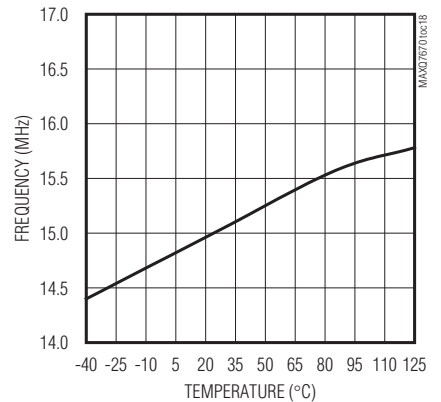
DVDD LINEAR REGULATOR OUTPUT VOLTAGE vs. TEMPERATURE



DVDD LINEAR REGULATOR OUTPUT VOLTAGE vs. LOAD CURRENT



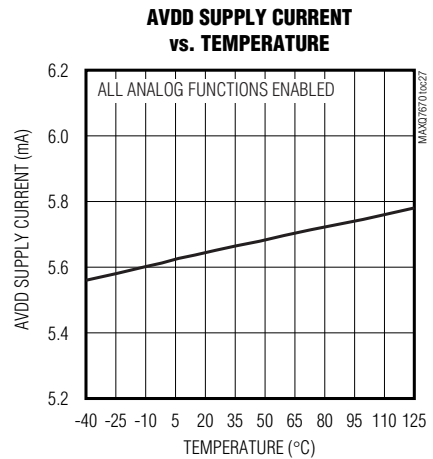
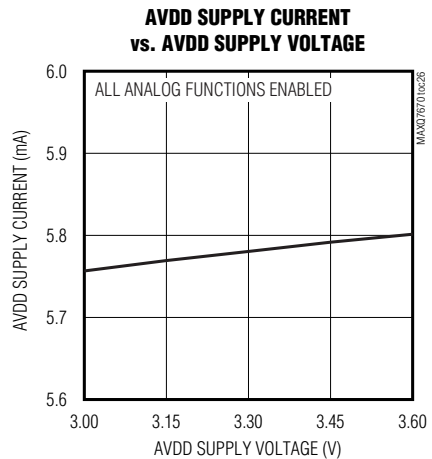
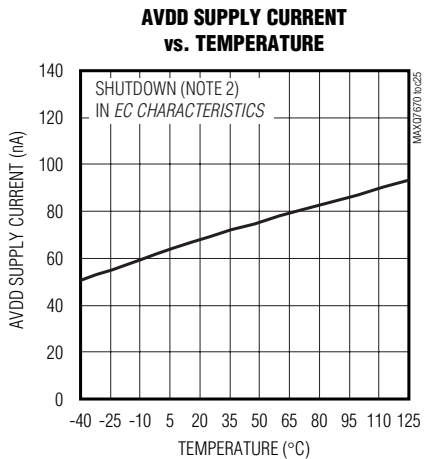
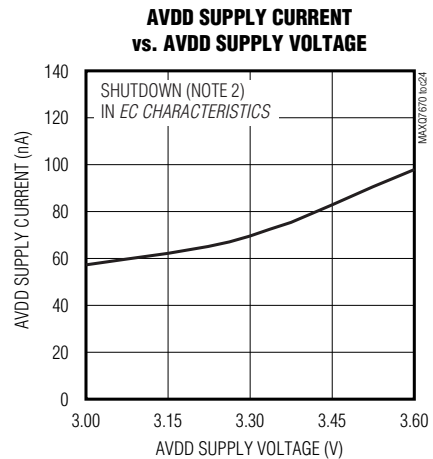
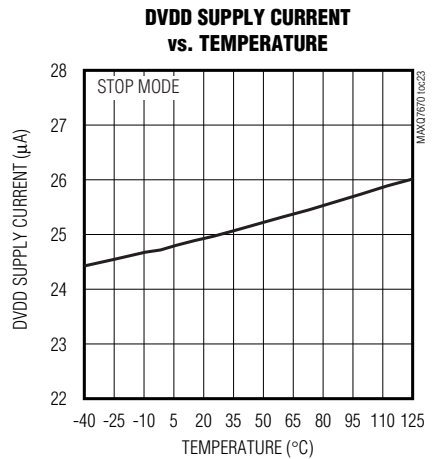
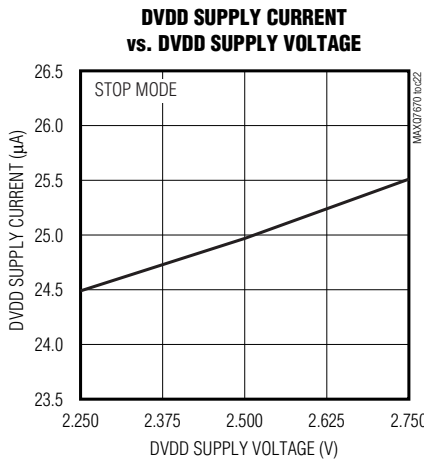
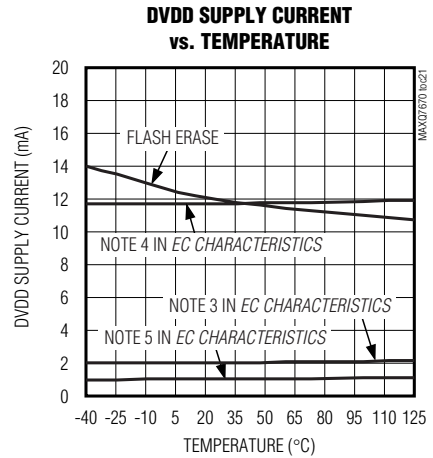
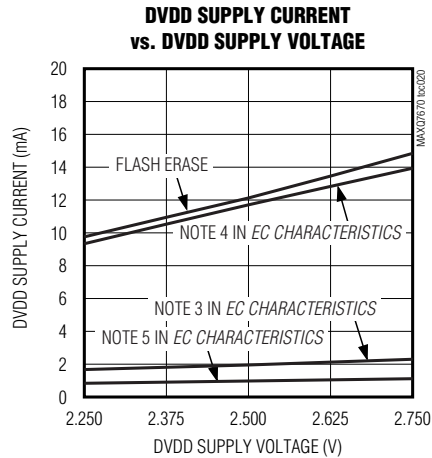
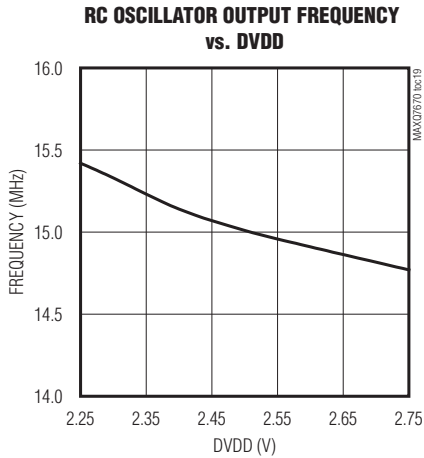
RC OSCILLATOR OUTPUT FREQUENCY vs. TEMPERATURE



Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

Typical Operating Characteristics (continued)

($V_{DDIO} = 5.0V$, $V_{AVDD} = 3.3V$, $V_{DVDD} = 2.5V$, $f_{SYSCLK} = 16MHz$, ADC resolution = 10 bits, $V_{REFDAC} = 3.3V$, $T_A = +25^\circ C$, unless otherwise noted.)

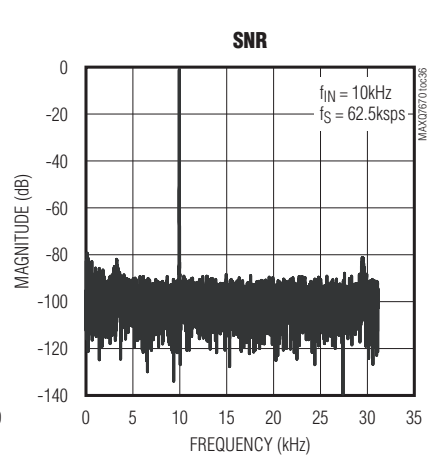
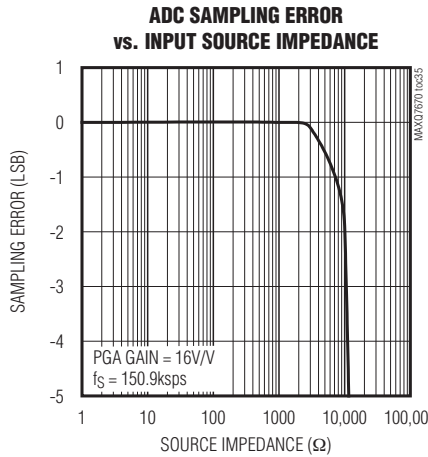
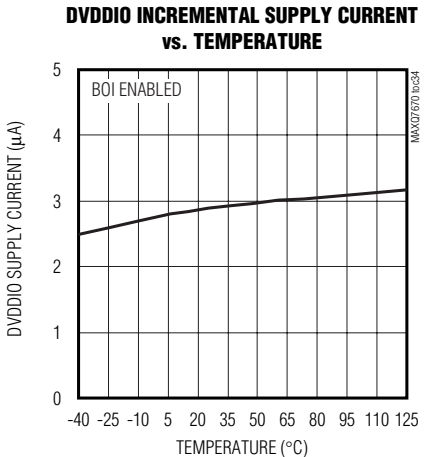
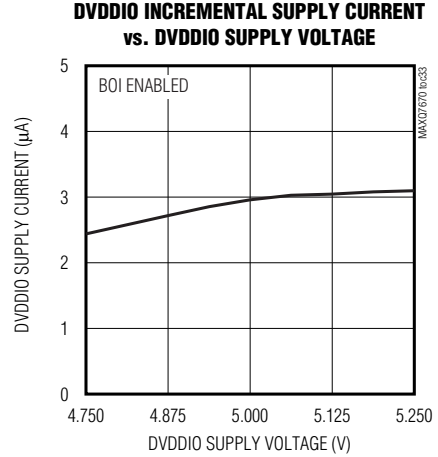
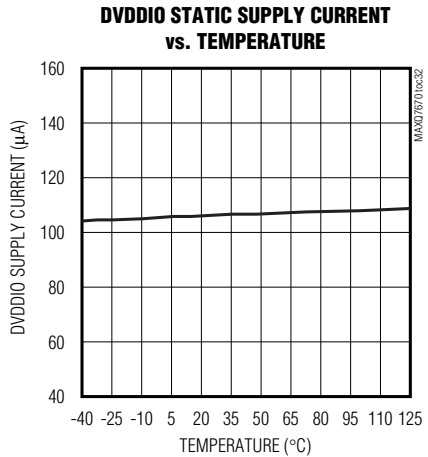
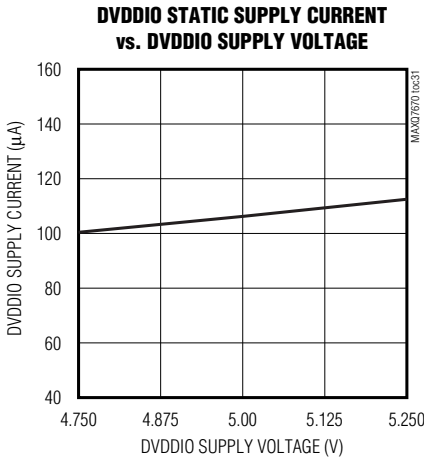
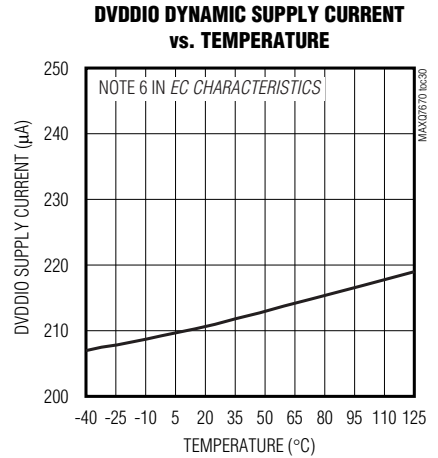
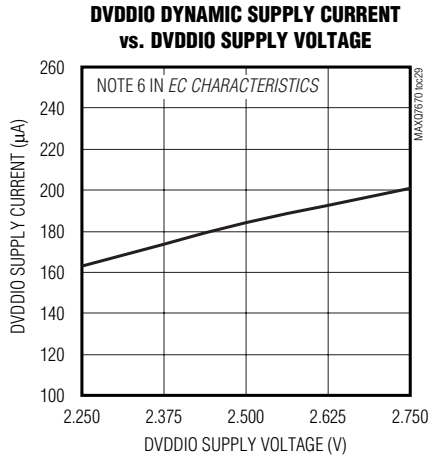
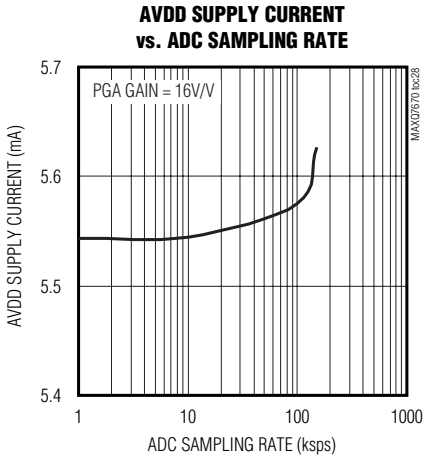


Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

MAXQ7670

Typical Operating Characteristics (continued)

($V_{DDIO} = 5.0V$, $V_{AVDD} = 3.3V$, $V_{D VDD} = 2.5V$, $f_{SYSCLK} = 16MHz$, ADC resolution = 10 bits, $V_{REFDAC} = 3.3V$, $T_A = +25^\circ C$, unless otherwise noted.)



Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

Pin Description

PIN	NAME	FUNCTION
1	AIN7	Analog Input Channel 7. AIN7 is multiplexed to the PGA or ADC as single-ended analog input 7 or as a differential input with AIN6. As a differential input, the polarity of AIN7 is negative.
2	AIN6	Analog Input Channel 6. AIN6 is multiplexed to the PGA or ADC as a single-ended analog input 6 or as a differential input with AIN7. As a differential input, the polarity of AIN6 is positive.
3	AIN5	Analog Input Channel 5. AIN5 is multiplexed to the PGA or ADC as single-ended analog input 5 or as a differential input with AIN4. As a differential input, the polarity of AIN5 is negative.
4	AIN4	Analog Input Channel 4. AIN4 is multiplexed to the PGA or ADC as single-ended analog input 4 or as a differential input with AIN5. As a differential input, the polarity of AIN4 is positive.
5	REFADC	ADC External Reference Input. Connect an external reference between 1V and V _{AVDD} .
6	AGND	Analog Ground
7	AIN3	Analog Input Channel 3. AIN3 is multiplexed to the PGA or ADC as single-ended analog input 3 or as a differential input with AIN2. As a differential input, the polarity of AIN3 is negative.
8	AIN2	Analog Input Channel 2. AIN2 is multiplexed to the PGA or ADC as single-ended analog input 2 or as a differential input with AIN3. As a differential input, the polarity of AIN2 is positive.
9	AIN1	Analog Input Channel 1. AIN1 is multiplexed to the PGA or ADC as single-ended analog input 1 or as a differential input with AIN0. As a differential input, the polarity of AIN1 is negative.
10	AIN0	Analog Input Channel 0. AIN0 is multiplexed to the PGA or ADC as single-ended analog input 0 or as a differential input with AIN1. As a differential input, the polarity of AIN0 is positive.
11	I.C.	Internally Connected. Connect to GNDIO for proper operation.
12	P0.0	Port 0 Bit 0. P0.0 is a general-purpose digital I/O with interrupt/wake-up capability.
13	P0.1	Port 0 Bit 1. P0.1 is a general-purpose digital I/O with interrupt/wake-up capability.
14	P0.2	Port 0 Bit 2. P0.2 is a general-purpose digital I/O with interrupt/wake-up capability.
15, 22, 38	GNDIO	Digital I/O Ground and Regulator Ground
16	CANRXD	CAN Bus Receiver Input. CAN receiver input.
17	CANTXD	CAN Bus Transmitter Output. CAN transmitter output.
18	\overline{SS}	Active-Low, SPI Port Slave Select Input. In SPI slave mode, this is the slave select input. In SPI master mode, this is an input and connection is optional (connect if mode fault enable is required, refer to the <i>MAXQ7670 User's Guide</i> for a description of the SPICN register). In master mode, use an available GPIO as a slave selector and pull \overline{SS} high to DVDDIO through a pullup resistor.
19	P0.6/T0	Port 0 Bit 6/Timer 0 I/O. P0.6 is a general-purpose digital I/O with interrupt/wake-up input capability. T0 is a primary timer/PWM input or output. The alternative function, T0, is selected using the T2CNA0 register. When this function is selected, it overrides the GPIO functionality.
20	P0.7/T0B	Port 0 Bit 7/Timer 0 Output. P0.7 is a general-purpose digital I/O with interrupt/wake-up input capability. T0B is a secondary timer/PWM output. The alternative function, T0B, is selected using the T2CNB0 register. When this function is selected, it overrides the GPIO functionality.
21, 39	DVDDIO	Digital I/O Supply Voltage and Regulator Supply Input. DVDDIO supplies all digital I/O except for XIN and XOUT, and supplies power to the two internal linear regulators, AVDD and DVDD. Bypass DVDDIO to GNDIO with a 0.1 μ F capacitor as close as possible to the device.

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

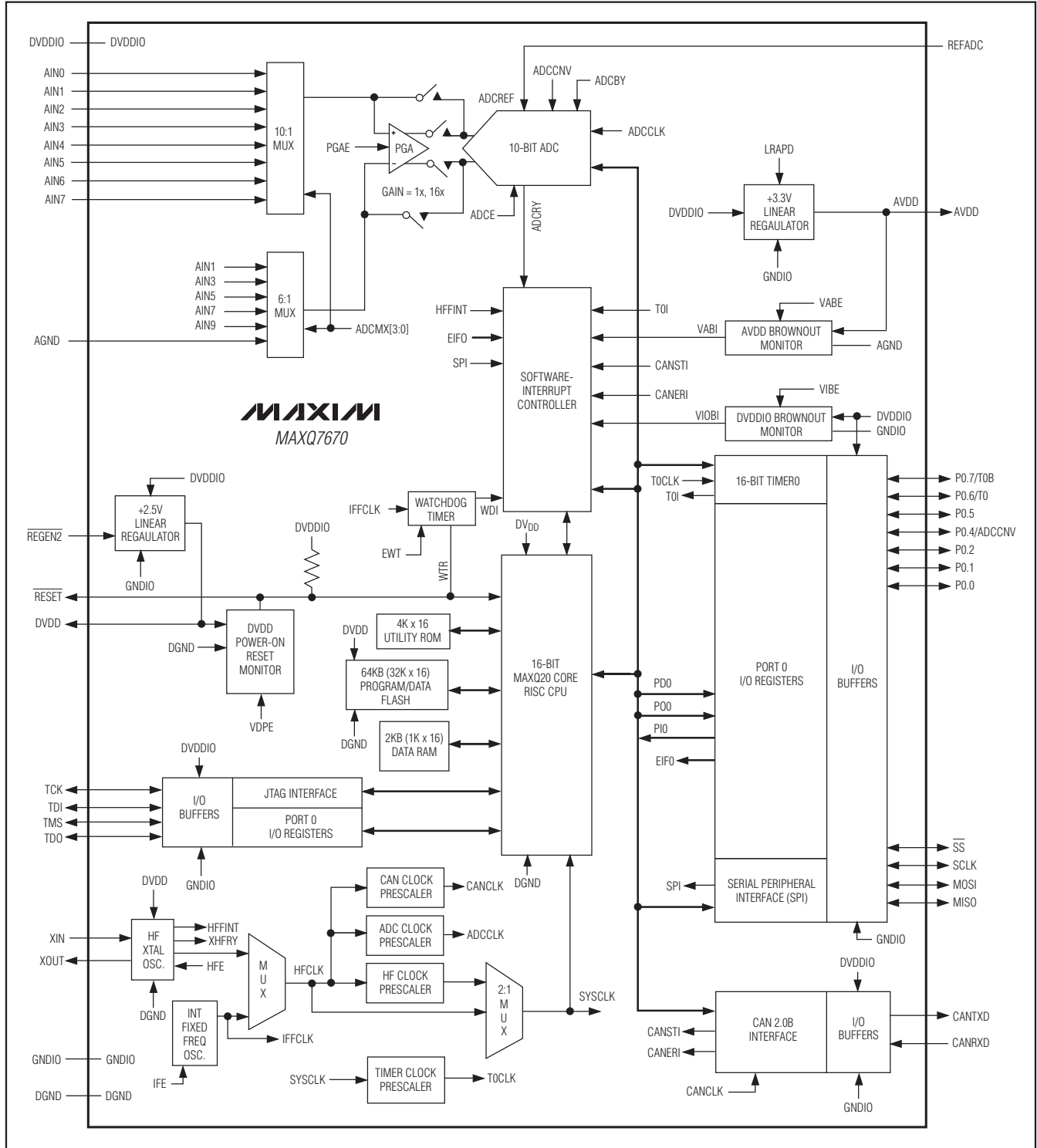
Pin Description (continued)

PIN	NAME	FUNCTION
23	SCLK	SPI Serial Clock. SCLK is the SPI interface serial clock I/O. In SPI master mode, SCLK is an output. While in SPI slave mode, SCLK is an input.
24	MOSI	SPI Serial Data I/O. MOSI is the SPI interface serial data output in master mode or serial data input in slave mode.
25	MISO	SPI Serial Data I/O. MISO is the SPI interface serial data input in master mode or serial data output in slave mode.
26	$\overline{\text{REGEN2}}$	Active-Low +2.5V Linear Regulator Enable Input. Connect $\overline{\text{REGEN2}}$ to GNDIO to enable the +2.5V linear regulator. Connect to DVDDIO to disable the +2.5V linear regulator.
27	TDO	JTAG Serial Test Data Output. TDO is the JTAG serial test, data output.
28	TMS	JTAG Test Mode Select. TMS is the JTAG test mode, select input.
29	TDI	JTAG Serial Test Data Input. TDI is the JTAG serial test, data input.
30	TCK	JTAG Serial Test Clock Input. TCK is the JTAG serial test, clock input.
31	P0.4/ ADCCNV	Port 0 Bit 4/ADC Start Conversion Control. P0.4 is a general-purpose digital I/O with interrupt/wake-up capability. ADCCNV is a firmware-configurable, rising or falling edge, start/convert signal used to trigger ADC conversions. The alternative function, ADCCNV, is selected using the register bits ACNT[2:0]. When using ADCCNV as a trigger for ADC conversion, set P0.4/ADCCNV as an input using the PD0 register. This action prevents any unintentional interference in the SARADC operation.
32	P0.5	Port 0 Bit 5. P0.5 is a general-purpose digital I/O with interrupt/wake-up capability.
33	$\overline{\text{RESET}}$	Reset Input/Output. Active-low input/output with internal 55k Ω pullup to DVDDIO. Drive low to reset the MAXQ7670. The MAXQ20 μC core holds $\overline{\text{RESET}}$ low during POR and during DVDD brownout conditions.
34	DGND	Digital Ground
35	XOUT	High-Frequency Crystal Output. Connect an external crystal to XIN and XOUT for normal operation, or leave unconnected if XIN is driven with an external clock source. Leave unconnected if an external clock source is not used.
36	XIN	High-Frequency Crystal Input. Connect an external crystal or resonator to XIN and XOUT for normal operation, or drive XIN with an external clock source. Leave unconnected if an external clock source is not used.
37	DVDD	Digital Supply Voltage. DVDD supplies internal digital core and flash memory. DVDD is directly connected to the output of the internal +2.5V linear regulator. Disable the internal regulator (through $\overline{\text{REGEN2}}$) to connect an external supply. Bypass DVDD to DGND with a 0.1 μF capacitor as close as possible to the device.
40	AVDD	Analog Supply Voltage. AVDD supplies PGA and ADC. AVDD is directly connected to the output of the internal +3.3V linear regulator. Disable the internal regulator (via software) to connect an external supply. Bypass AVDD to AGND with a 0.1 μF capacitor as close as possible to the device.
—	EP	Exposed Pad. Connect EP to the ground plane.

MAXQ7670

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

Block Diagram



Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

MAXQ7670

Detailed Description

The MAXQ7670 incorporates a 16-bit RISC arithmetic logic unit (ALU) with a Harvard memory architecture that addresses 64KB (32K x 16) of flash and 2048 bytes (1024 x 16) of RAM memory. This core combined with digital and analog peripherals provide versatile data-acquisition functions. The peripherals include up to seven digital I/Os, a 4-wire SPI interface, a CAN 2.0B bus, a JTAG interface, a timer, an integrated RC oscillator, two linear regulators, a watchdog timer, three power-supply supervisors, a 10-bit 250ksps SAR ADC with programmable-gain amplifier (PGA) and eight single-ended or four differential multiplexed inputs. The

power-efficient MAXQ20 μ C core consumes less than 1mA/MIPS. Refer to the *MAXQ7670 User's Guide* for more detailed information on configuring and programming the MAXQ7670.

Analog Input Peripheral

The integrated 10-bit ADC employs an ultra-low-power SAR-based conversion method and operates up to 250ksps with PGA = 1V/V (150.9ksps with PGA = 16V/V). The integrated 8-channel multiplexer (mux) and PGA allow the ADC to measure eight single-ended (relative to AGND) or four fully differential analog inputs with software-selectable input ranges through the PGA. See Figures 3 and 4.

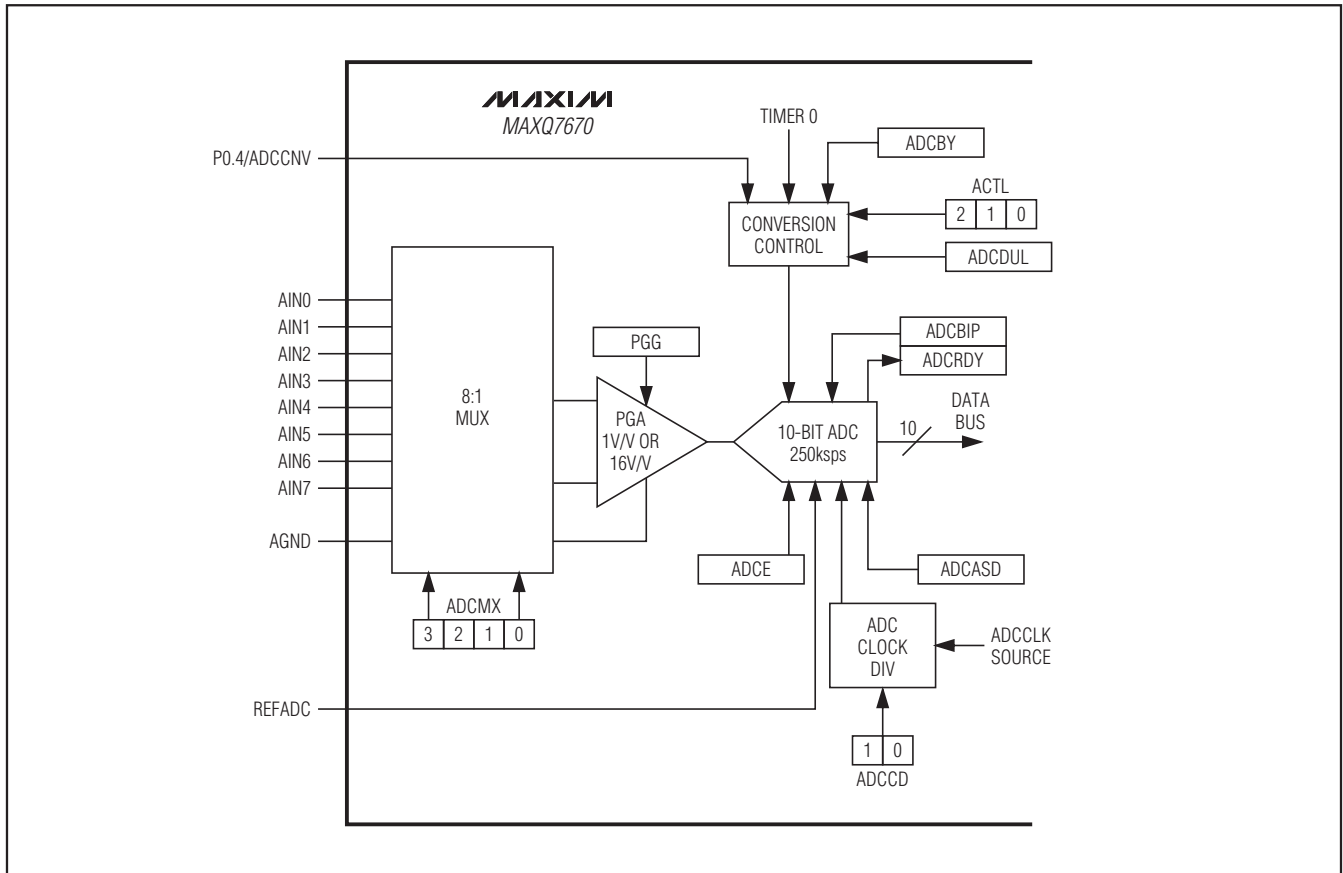


Figure 3. Simplified Analog Input Diagram (Eight Single-Ended Inputs)

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

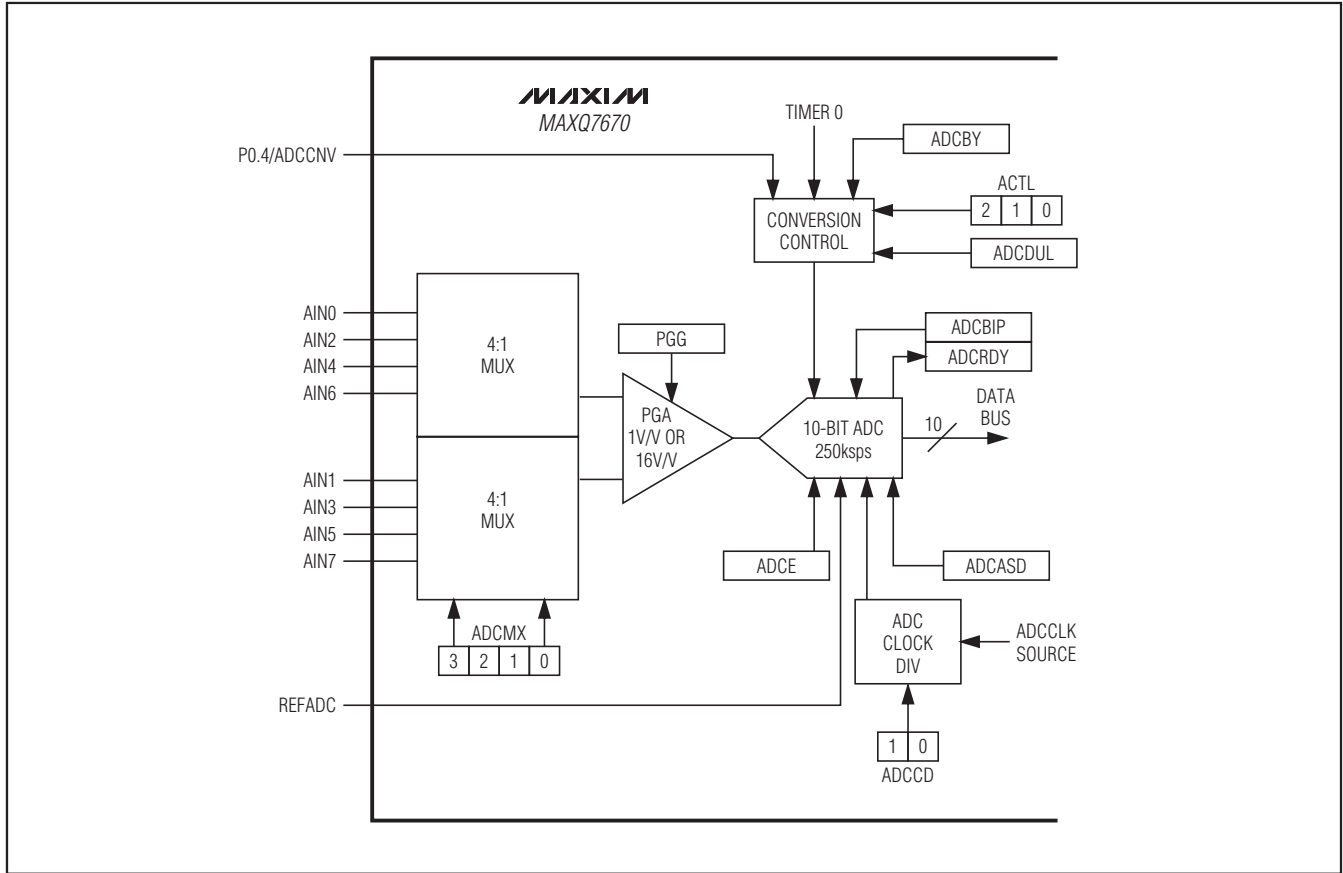


Figure 4. Simplified Analog Input Diagram (Four Fully Differential Inputs)

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

The MAXQ7670 ADC uses a fully differential SAR conversion technique and an integrated T/H (track and hold) block to convert voltage signals into a 10-bit digital result. Both single-ended and differential configurations are implemented using an analog input channel multiplexer that supports 8 single-ended or 4 differential channels.

In single-ended mode, the mux selects from either of the ground-referenced analog inputs AIN0–AIN7. In differential input configuration, analog inputs are selected from the following pairs: AIN0/AIN1, AIN2/AIN3, AIN4/AIN5, and AIN6/AIN7. Table 1 shows the single-ended and differential input configurations possible for the ADC mux.

Analog Input Track and Hold

A SAR conversion in the MAXQ7670 has different T/H cycles depending on whether a gain of 1 (bypass) or a gain of 16 (PGA enabled) is selected.

Gain = 1V/V

In gain = 1V/V, the conversion has a two-stage T/H cycle. In track mode, a positive input capacitor connects to the signal channel. A negative input capacitor connects to the reference channel. After the T/H enters hold mode, the difference between the signal and the reference channel is converted to a 10-bit value. This two-stage cycle takes 16 SARCLKs to complete.

Gain = 16V/V

In gain = 16V/V, the conversion has a three-stage T/H cycle: amplification, ADC track, and ADC hold. First, the PGA tracks the selected input and reference signals. The PGA amplifies the difference between the two signals and holds the result for the next stage, ADC track. The ADC tracks and converts the PGA result into a 10-bit value. The SAR operation itself does not change irrespective of the chosen gain. This three-stage cycle takes 26.5 SARCLKs to complete. Figure 5 shows the conversion timing differences between gain = 1V/V and gain = 16V/V.

Table 1. ADC Mux Input Configurations

SAR CHANNEL SELECT (REGISTER ACNT[14:11])	SIGNAL CHANNEL INTO ADC	REFERENCE CHANNEL INTO ADC	MEASUREMENT TYPE
0000	AIN0	AGND	Single-ended measurement on AIN0
0001	AIN1	AGND	Single-ended measurement on AIN1
0010	AIN2	AGND	Single-ended measurement on AIN2
0011	AIN3	AGND	Single-ended measurement on AIN3
0100	AIN4	AGND	Single-ended measurement on AIN4
0101	AIN5	AGND	Single-ended measurement on AIN5
0110	AIN6	AGND	Single-ended measurement on AIN6
0111	AIN7	AGND	Single-ended measurement on AIN7
1000	—	—	Reserved
1001	—	—	Reserved
1010	AIN0	AIN1	AIN0/AIN1
1011	AIN2	AIN3	AIN2/AIN3
1100	AIN4	AIN5	AIN4/AIN5
1101	AIN6	AIN7	AIN6/AIN7
1110	—	—	Reserved
1111	—	—	VCIM differential zero offset trim

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

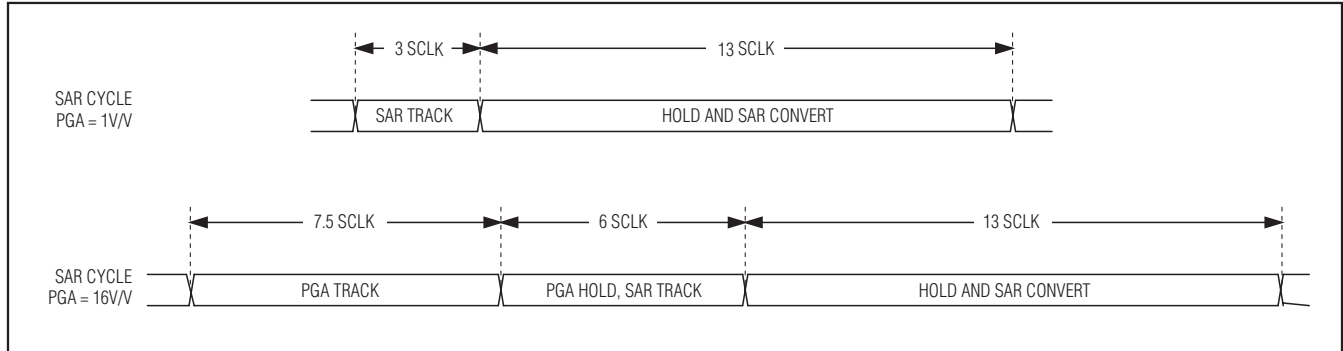


Figure 5. Conversion Timing Differences Between Gain = 1V/V and Gain = 16V/V

Input Impedance

The input-capacitance charging rate determines the time required for the T/H to acquire an input signal. The required acquisition time lengthens with the increase of the input signals source resistance. Any source below 5k Ω does not significantly affect the ADC's performance. A high-impedance source can be accommodated by placing a 1 μ F capacitor between the input channel and AGND. The combination of analog-input source impedance and the capacitance at the analog input creates an RC filter that limits the analog-input bandwidth.

Controlling ADC Conversions

Use the following methods to control the ADC conversion timing:

- 1) Software register bit control
- 2) Continuous conversion
- 3) Internal timer (T0)
- 4) External input through ADCCNV

Refer to the *MAXQ7670 User's Guide* for more detailed information on the ADC and mux.

POR and Brownout

The MAXQ7670 operates from a single, external +5V supply connected to the DVDDIO. DVDDIO is the supply rail for the digital I/O and the supply input for both integrated linear regulators. The +3.3V linear regulator powers AVDD, while the +2.5V linear regulator powers DVDD. Alternatively, connect $\overline{\text{REGEN2}}$ to DVDDIO and apply external power supplies to AVDD and DVDD.

Power supplies DVDDIO, DVDD, and AVDD each include a brownout monitor that alerts the μ C through an interrupt when the corresponding supply voltages drop below a defined threshold. This condition is generally referred to as brownout interrupt (BOI). Enable BOI by setting the VABE, VDBE, and VIBE bits in the

APE register. By continually checking for low supply voltages, appropriate action can be taken for brownout conditions.

Startup Using Internal Regulators

Once the +5V DVDDIO supply reaches approximately 1.25V, the +2.5V linear regulator turns on and DVDD begins ramping. Between the DVDD levels of 1V and the reset threshold, the DVDD monitor holds $\overline{\text{RESET}}$ low. DVDD releases $\overline{\text{RESET}}$ after reaching the reset threshold. The MAXQ7670 jumps to the reset vector location (8000h in the utility ROM). During this time, DVDD finishes ramping to its nominal voltage of +2.5V.

During this POR time, the software-enabled +3.3V linear regulator remains off. Turn on the +3.3V linear regulator after the MAXQ7670 has completed its bootup routines and is running application code. To turn on the +3.3V regulator, set the LRAPD bit in the APE register to 0. The AVDD supply begins ramping to its nominal voltage of +3.3V.

Brownout Detectors

The MAXQ7670 features brownout monitors for the +5V DVDDIO, +3.3V AVDD, and +2.5V DVDD power supplies. When enabled, these monitors generate interrupts when DVDDIO, AVDD, or DVDD fall below their respective brownout thresholds. Monitoring the supply rails alerts the μ C to brownout conditions so appropriate action can be taken. Under normal conditions the DVDDIO brownout monitor signals a falling +5V supply before the DVDD or AVDD brownout monitors indicate that the +2.5V or +3.3V are falling. The exceptions to this condition are:

- If either DVDD or AVDD are externally powered and the source of power is removed
- If there is some type of device failure that pulls the regulator outputs low without affecting the +5V DVDDIO supply

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

The DVDD reset supervisor resets the MAXQ7670 when the +2.5V DVDD falls below the reset threshold. The processor remains in reset until DVDD returns above the reset threshold. The μC does not execute commands in reset mode. See Figure 6 for the μC response to DVDD brownout and reset.

Refer to the *MAXQ7670 User's Guide* for detailed programming information, and a more thorough description of POR and brownout behavior.

Internal 3.3V Linear Regulator

The integrated 3.3V 50mA linear regulator or an external 3.3V supply powers AVDD. The integrated 3.3V regulator is inactive upon power-up. Enable the integrated regulator with software programming after power-up. When using an external supply, connect a regulated 3.3V supply to AVDD after applying DVDDIO.

Internal 2.5V Linear Regulator

The integrated 2.5V 50mA linear regulator or an external 2.5V supply applied at DVDD powers DVDD. Connect $\overline{\text{REGEN2}}$ to GNDIO to enable the integrated regulator. Connect $\overline{\text{REGEN2}}$ to DVDDIO to use an external supply. When using an external supply, connect a regulated 2.5V supply to DVDD after applying DVDDIO.

DVDDIO Current Requirements

Both internal linear regulators are capable of supplying up to 50mA each. When using the regulators to power AVDD and DVDD and to provide power to external devices, make sure DVDDIO's power input can source a current greater than the sum of the MAXQ7670 supply current and the load currents of the two regulators.

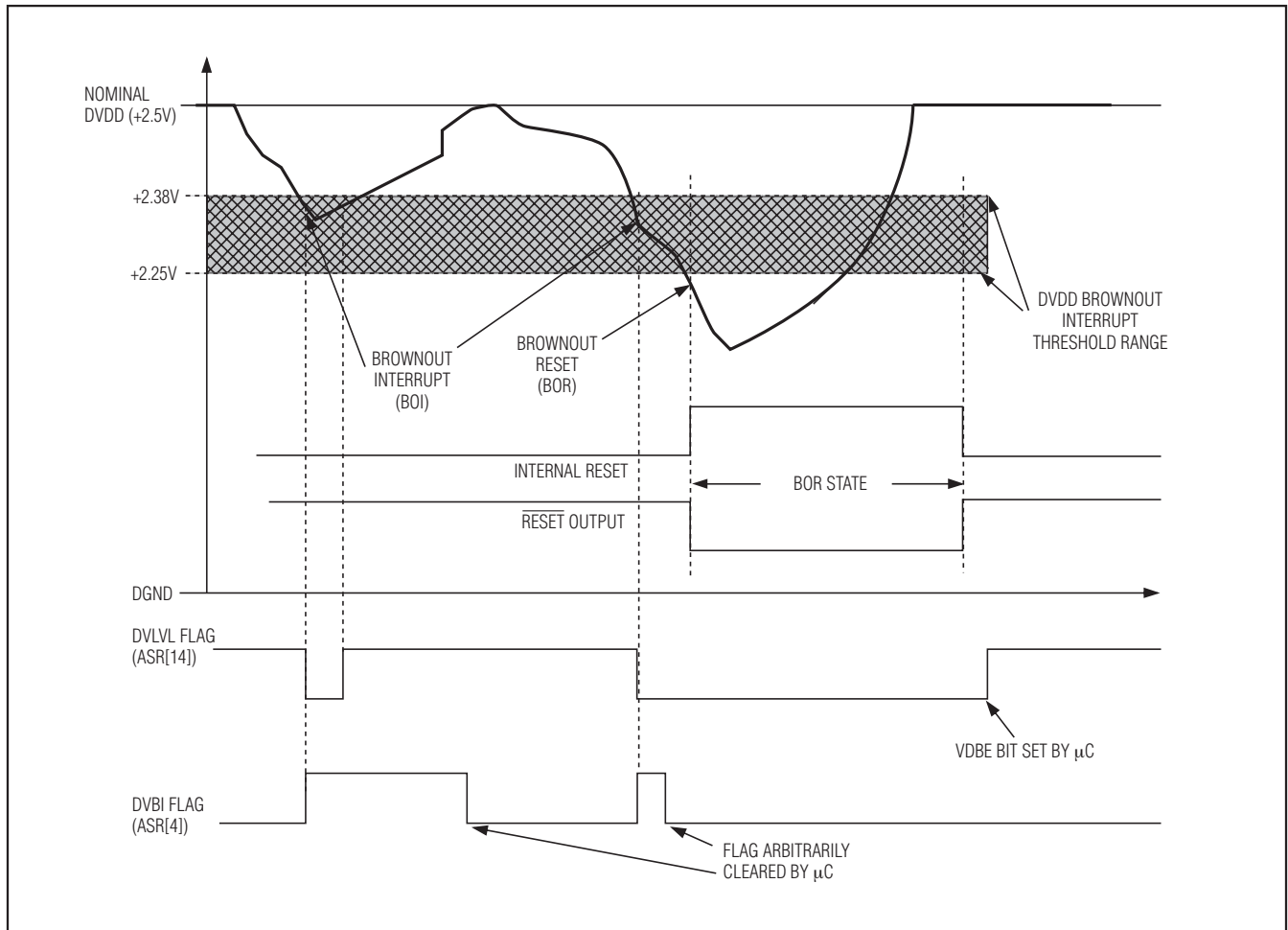


Figure 6. DVDD Brownout and Reset Behavior

Microcontroller with 10-Bit ADC, PGA, 64KB Flash, and CAN Interface

System Clock Generator

The MAXQ7670 oscillator module provides the master clock generator that supplies the system clock for the μ C core and all of the peripheral modules. The high-frequency oscillator operates with an 8MHz or 16MHz crystal. Alternatively, use the integrated RC oscillator in applications that do not require precise timing. The MAXQ7670 executes most instructions in a single SYSCLK period. The oscillator module contains all of the primary clock generation circuitry. Figure 7 shows a block diagram of the system clock module.

The MAXQ7670 contains the following features for generating its master clock signal timing source:

- Internal, fast-starting, 15MHz RC oscillator eliminates external crystal
- Internal high-frequency oscillator that can drive an external 8MHz or 16MHz crystal
- External high-frequency 0.166MHz to 16MHz clock input
- Power-up timer
- Power-saving management modes
- Fail-safe modes

Watchdog Timer

The primary function of the watchdog timer is to supervise software execution, watching for stalled or stuck software. The watchdog timer performs a controlled system restart when the μ C fails to write to the watchdog timer register before a selectable timeout interval expires. A watchdog timer typically has four objectives:

- 1) To detect if a system is operating normally

- 2) To detect an infinite loop in any of the tasks
- 3) To detect an arbitration deadlock involving two or more tasks
- 4) To detect if some lower priority tasks are not getting to run because of higher priority tasks

As illustrated in Figure 8, the internal RC oscillator (CLK_RC) drives the watchdog timer through a series of dividers. The programmable divider output determines the timeout interval. When enabled, the interrupt flag WDIF sets. A system reset occurs after a time delay (based on the divider ratio) unless an interrupt service routine clears the watchdog interrupt.

The watchdog timer functions as the source of both the watchdog interrupt and the watchdog reset. The interrupt timeout has a default divide ratio of 2^{12} of the CLK_RC, with the watchdog reset set to timeout 2^9 clock cycles later. With the nominal RC oscillator value of 15MHz, an interrupt timeout occurs every 0.273ms, followed by a watchdog reset 34 μ s later. The watchdog timer resets to the default divide ratio following any reset event. Use the WD0 and WD1 bits in the WDCN register to increase the watchdog interrupt period. Changing the WD[1:0] bits before a watchdog interrupt timeout occurs (i.e. before the watchdog reset counter begins) resets the watchdog timer count. The watchdog reset timeout occurs 512 RC oscillator cycles after the watchdog interrupt timeout. For more information on the MAXQ7670 watchdog timer, refer to the *MAXQ7670 User's Guide*.

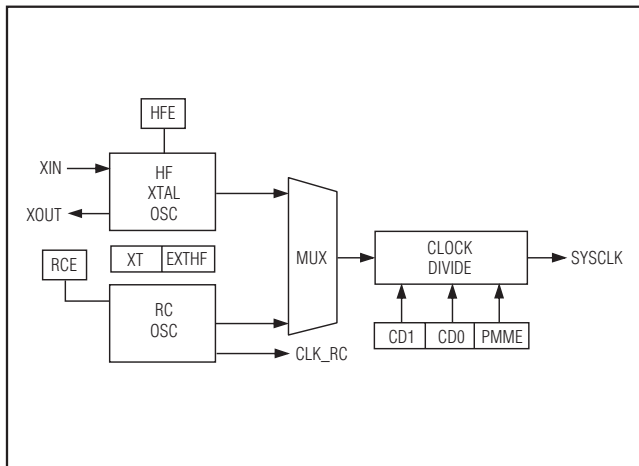


Figure 7. High-Frequency and RC Oscillator Functional Diagram

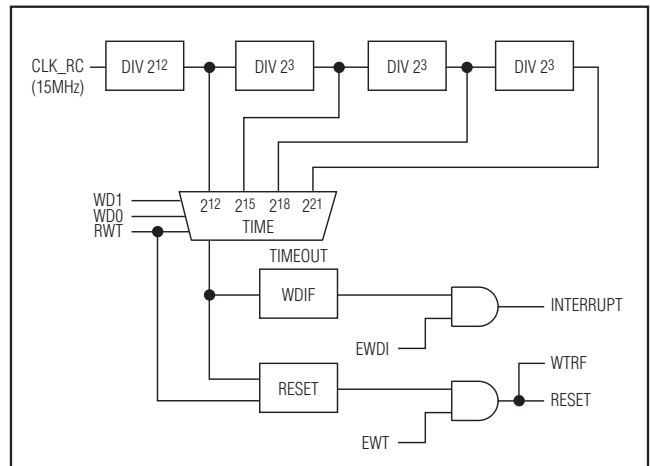


Figure 8. Watchdog Functional Diagram

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Timer and PWM

The MAXQ7670 includes a 16-bit timer channel. The timer offers two ports, T0 and T0B, to facilitate PWM outputs, and capture timing events. The autoreload 16-bit timer/counter offers the following functions:

- 8-/16-bit timer/counter
- Up/down autoreload
- Counter function of external pulse
- Capture
- Compare
- PWM output
- Event timer
- System supervisor

Refer to the *MAXQ7670 User's Guide* and Application Note 3205: *Using Timers in the MAXQ Family of Microcontrollers* for more information about the timer module.

CAN Interface Bus

The MAXQ7670 incorporates a fully compliant CAN 2.0B controller.

Two groups of registers provide the μ C interface to the CAN controller. To simplify the software associated with the operation of the CAN controllers, most of the global CAN status and controls as well as the individual message center control/status registers are located in the peripheral register map. The remaining registers associated with the data identification, identification masks, format, and data are located in a dual port memory to allow the CAN controller and the processor access to the required functions. The CAN controller can directly access the dual port memory. The processor accesses the dual port memory through a dedicated interface that consists of the CAN 0 data pointer (C0DP) and the CAN 0 data buffer (C0DB) special function registers. See Figure 9 for CAN controller details.

CAN Functional Description

The CAN module stores up to 15 messages. Each message consists of an acceptance identifier and 8 bytes of data. The MAXQ7670 supports both the standard 11-bit and extended 29-bit identification modes.

Configure each of the first 14 message centers either to transmit or receive. Message center 15 is a receive-only center, storing any message that centers 1–14 do not accept.

A message center only accepts an incoming message if the following conditions are satisfied:

- The incoming message's arbitration value matches the message center's acceptance identifier
- The first 2 data bytes of the incoming message match the bytes in the media arbitration registers (C0MA0 and C0MA1)

Use the global mask registers to mask out bits in the incoming message that do not require a comparison.

A message center, configured to transmit, meets these conditions: T/R = 1, TIH = 0, DTUP = 1, MSRDY = 1, and MTRQ = 1. The message center transmits its contents when it receives an incoming request message containing the same identifier (i.e., a remote frame).

Global control and status registers in the CAN unit enable the μ C to evaluate error messages, validate and locate new data, establish the bus timing for the CAN bus, establish the identification mask bits, and verify the source of individual messages. In addition, each message center is individually equipped with the necessary status and controls to establish directions, interrupt generation, identification mode (standard or extended), data field size, data status, automatic remote frame request and acknowledgment, and masked or nonmasked identification acceptance testing.

JTAG Interface Bus

The joint test action group (JTAG) IEEE® 1149.1 standard defines a unique method for in-circuit testing and programming. The MAXQ7670 conforms to this standard, implementing an external test access port (TAP) and internal TAP controller for communication with a JTAG bus master, such as an automatic test equipment (ATE). For detailed information on the TAP and TAP controller, refer to IEEE Standard 1149.1 on the IEEE website at www.standards.ieee.org. The JTAG on the MAXQ7670 does not support boundary scan test capability.

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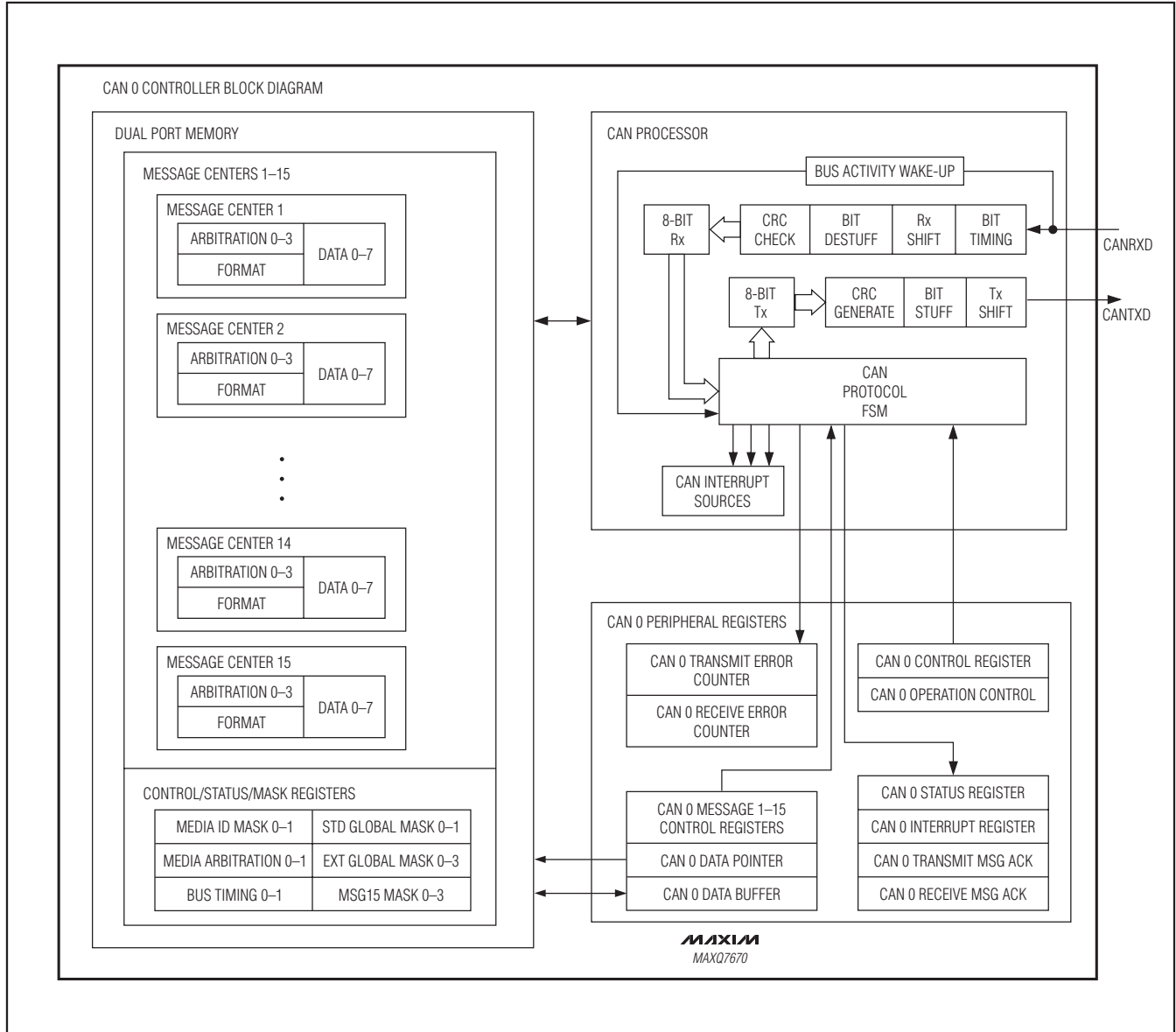


Figure 9. CAN 0 Controller Block Diagram

The TAP controller communicates synchronously with the host system (bus master) through four digital I/Os: test mode select (TMS), test clock (TCK), test data input (TDI), and test data output (TDO). The internal TAP module consists of several shift registers and a TAP controller (see Figure 11). The shift registers serve as transmit-and-receive data buffers for a debugger.

4-Wire SPI Bus

The MAXQ7670 includes a powerful hardware SPI module, providing serial communication with a wide variety of external devices. The SPI port on the MAXQ7670 is a fully independent module that is accessed through software. This full 4-wire, full-duplex serial bus module supports master and slave modes. The SPI clock

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frequency is limited to $SYSCLK/2$ in master mode and $SYSCLK/8$ in slave mode. Figure 10 shows the functional diagram of the SPI port. Figures 1 and 2 illustrate the timing parameters listed in the *Electrical Characteristics* table.

General-Purpose Digital I/Os

The MAXQ7670 provides seven general-purpose digital I/Os (GPIOs). Some of the GPIOs include an additional special function (SF), such as a timer input/output. For example, the state of P0.6/T0 is programmable to depend on timer channel 0 logic. When used as a port, each I/O is configurable for high-impedance, weak pullup to DVDDIO or pulldown to GNDIO. At power-up,

each GPIO is configured as an input with a pullup to DVDDIO. In addition, each GPIO can be programmed to cause an interrupt (on falling or rising edges). In stop mode, use any interrupt to wake-up the device.

The port direction (PD) register determines the input/output direction of each I/O. The port output (PO) register contains the current state of the logic output buffers. When an I/O is configured as an output, writing to the PO register controls the output logic state. Reading the PO register shows the current state of the output buffers, independent of the data direction. The port input (PI) register is a read-only register that always reflects the logic state of the I/Os.

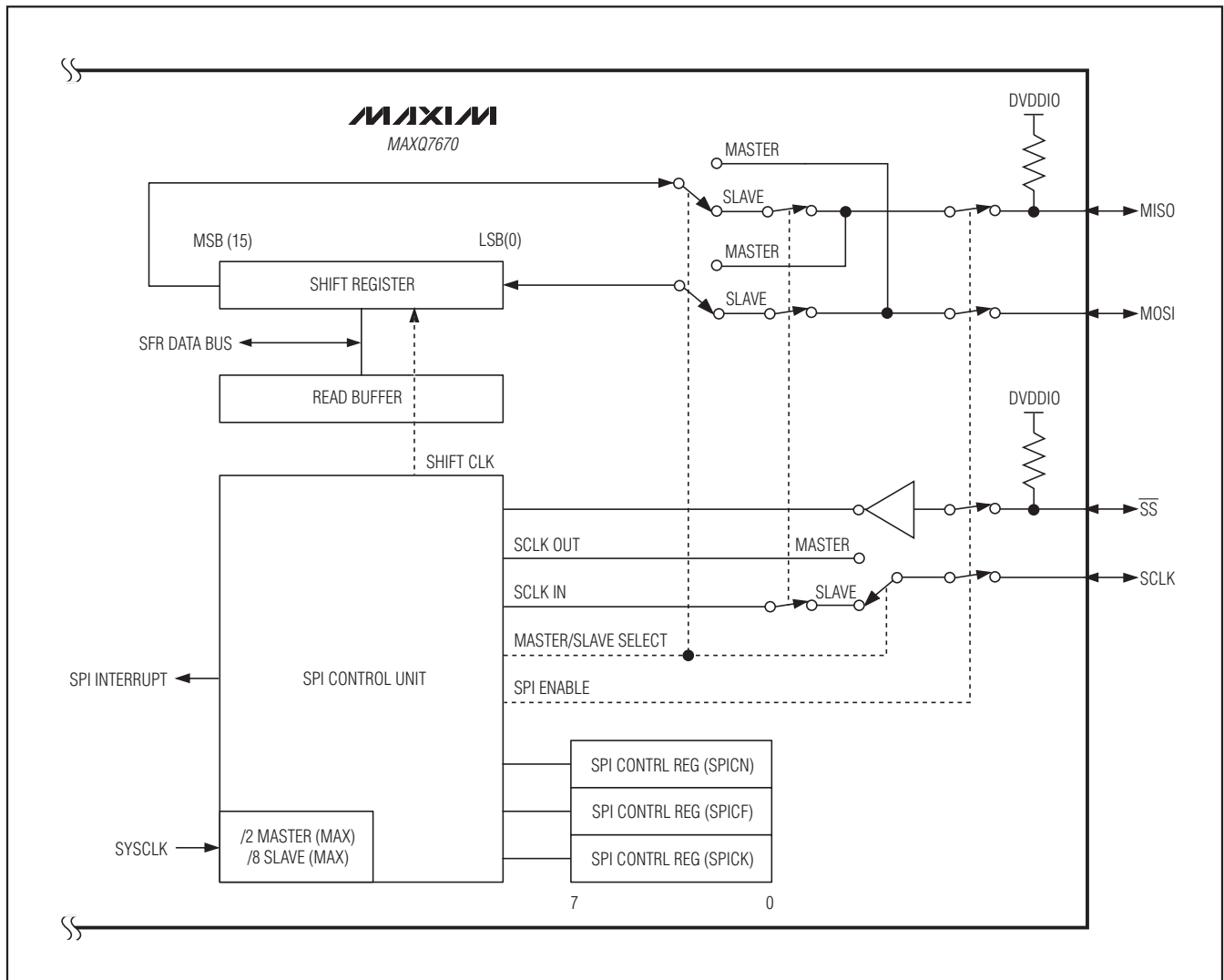


Figure 10. SPI Functional Diagram