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Dual 8.0 mOhm and dual 21 mOhm high-side switch

The 12XS6 is the latest SMARTMOS achievement in automotive lighting drivers. It belongs to an expanding family that helps to control and diagnose incandescent lamps and light-emitting diodes (LEDs) with enhanced precision. It combines flexibility through daisy chainable SPI 5.0 MHz, extended digital and analog feedbacks, safety, and robustness.

Output edge shaping helps to improve electromagnetic performance. To avoid shutting off the device upon inrush current, while still being able to closely track the load current, a dynamic overcurrent threshold profile is featured. Current of each channel can be sensed with a programmable sensing ratio. Whenever communication with the external microcontroller is lost, the device enters a fail operation mode, but remains operational, controllable, and protected.

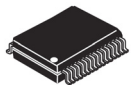
This new generation of high-side switch products family 12XS6 facilitates ECU design due to compatible MCU software and PCB foot prints for each device variant.

Features

- Dual 8.0 mΩ and dual 21 mΩ high-side switches with high transient current capability
- 16-bit 5.0 MHz SPI control of overcurrent profiles, channel control including PWM duty-cycles, output-on and -off open load detections, thermal shut-down and prewarning, and fault reporting
- Output current monitoring with programmable synchronization signal and battery voltage feedback
- Limp home mode
- External smart power switch control
- Operating voltage is 7.08 V to 18 V with sleep current < 5.0 μA, extended mode from 6.08 V to 28 V
- -16 V reverse polarity and ground disconnect protections
- Compatible PCB foot print and SPI software driver among the family

MC08XS6421BEK; MC08XS6421CEK

QUAD HIGH-SIDE SWITCH



BEK SUFFIX (PB-FREE)
98ASA00368D
32-PIN SOICW-EP

CEK SUFFIX (PB-FREE)
98ASA00894D
32-PIN SOICW-EP

Applications

- Low-voltage automotive lighting
- Halogen lamps
- Incandescent bulbs
- Light-emitting diodes (LEDs)
- HID Xenon ballasts

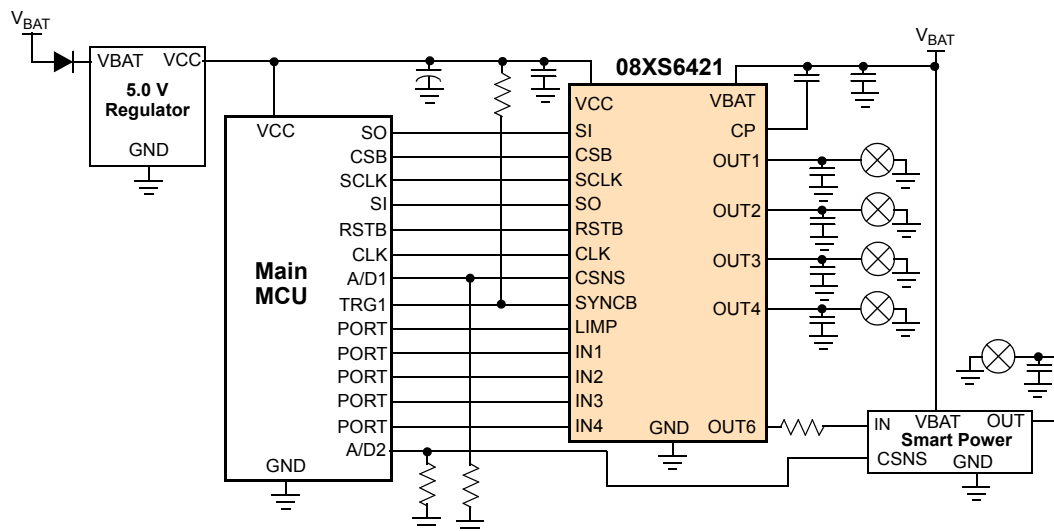


Figure 1. Dual 8.0 mΩ and dual 21 mΩ simplified application diagram

* This document contains certain information on a new product. Specifications and information herein are subject to change without notice.

1 Orderable parts

This section describes the part numbers available to be purchased along with their differences. Valid orderable part numbers are provided on the web. To determine the orderable part numbers for this device, go to <http://www.nxp.com> and perform a part number search for the following device numbers.

Table 1. Orderable part variations

Part number	Notes	Temperature (T _A)	Package	OUT1 R _{DS(on)}	OUT2 R _{DS(on)}	OUT3 R _{DS(on)}	OUT4 R _{DS(on)}	OUT6
MC08XS6421BEK	(1)	-40 °C to 125 °C	SOICW 32-pin exposed pad	21 mΩ	21 mΩ	8.0 mΩ	8.0 mΩ	Yes
MC08XS6421CEK								

Notes

1. To order parts in tape and reel, add the R2 suffix to the part number.

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2 Internal block diagram

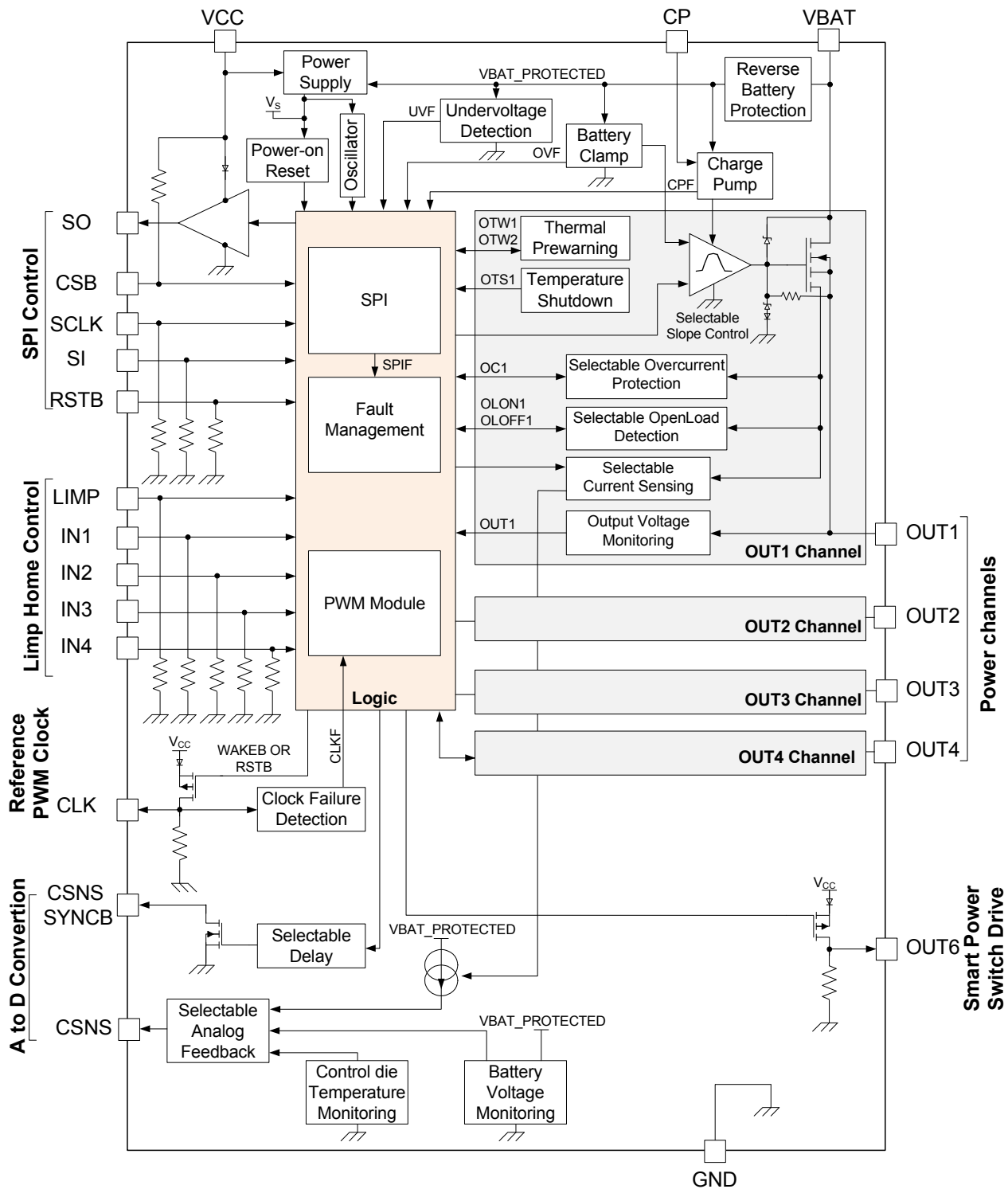


Figure 2. Simplified internal block diagram (quad version)

3 Pin connections

3.1 Pinout diagram

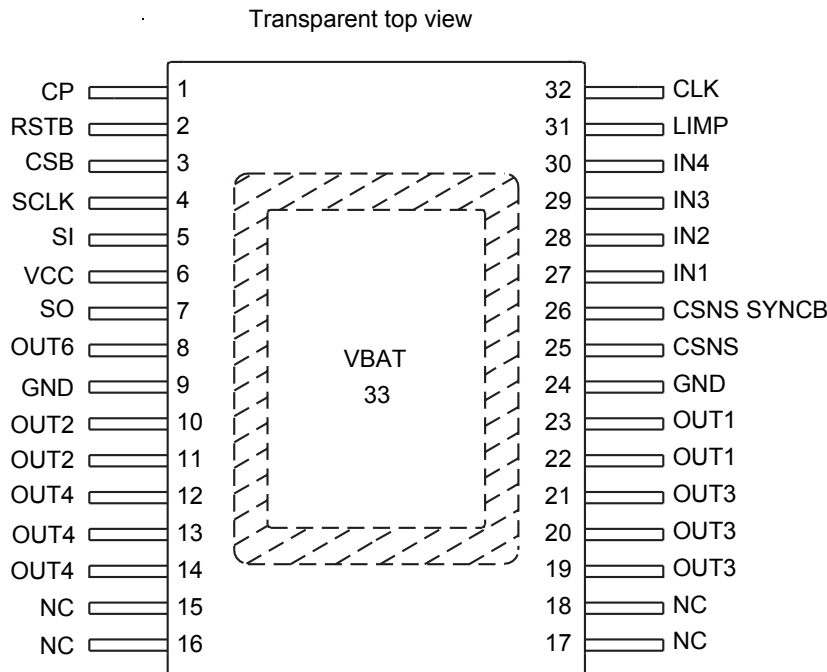


Figure 3. 12XS6 pinout diagram

3.2 Pin definitions

Table 2. 12XS6 pin definitions

Pin number	Pin name	Pin function	Formal name	Definition
1	CP	Internal supply	Charge pump	This pin is the connection for an external capacitor for charge pump use only.
2	RSTB	SPI	Reset	This input pin is used to initialize the device configuration and fault registers, as well as place the device in a low-current sleep mode. This pin has a passive internal pull-down.
3	CSB	SPI	Chip select	This input pin is connected to a chip select output of a master microcontroller (MCU). When this digital signal is high, SPI signals are ignored. Asserting this pin low starts the SPI transaction. The transaction is indicated as completed when this signal returns to a high level. This pin has a passive internal pull-up to V _{CC} through a diode.
4	SCLK	SPI	Serial clock	This input pin is connected to the MCU providing the required bit shift clock for SPI communication. This pin has a passive internal pull-down.
5	SI	SPI	Serial input	This pin is the data input of the SPI communication interface. The data at the input is sampled on the positive edge of the SCLK. This pin has a passive internal pull-down.
6	VCC	Power supply	MCU power supply	This pin is a power supply pin is for internal logic, the SPI I/Os, and the OUT6 driver.
7	SO	SPI	Serial output	This output pin is connected to the SPI serial data input pin of the MCU, or to the SI pin of the next device of a daisy chain of devices. The SPI changes on the negative edge of SCLK. When CSB is high, this pin is high-impedance.

Table 2. 12XS6 pin definitions (continued)

Pin number	Pin name	Pin function	Formal name	Definition
8	OUT6	Output	External solid state	This output pin controls an external smart power switch by logic level. This pin has a passive internal pull-down.
9, 24	GND	Ground	Ground	These pins are the ground for the logic and analog circuitries of the device. For ESD and electrical parameter accuracy purpose, the ground pins must be shorted in the board.
10, 11	OUT2	Output	Channel #2	Protected high-side power output pins to the load.
12, 13, 14	OUT4	Output	Channel #4	Protected high-side power output pins to the load.
15, 16, 17, 18	NC	N/A	Not connected	These pins may not be connected. It is recommended to put these pins to ground.
19, 20, 21	OUT3	Output	Channel #3	Protected high-side power output pins to the load.
22, 23	OUT1	Output	Channel #1	Protected high-side power output pins to the load.
25	CSNS	Feedback	Current sense	This pin reports an analog value proportional to the designated OUT[1:5] output current, or the temperature of the exposed pad, or the battery voltage. It is used externally to generate a ground referenced voltage for the microcontroller (MCU). Current recopy and analog voltage feedbacks are SPI programmable.
26	CSNS SYNCB	Feedback	Current sense synchronization	This open drain output pin allows synchronizing the MCU A/D conversion. This pin requires an external pull-up resistor to V_{CC} .
27	IN1	Input	Direct input #1	This input wakes up the device. This input pin is used to directly control corresponding channel in fail mode. During normal mode, the control of the outputs by the control inputs is SPI programmable. This pin has a passive internal pull-down.
28	IN2	Input	Direct input #2	This input wakes up the device. This input pin is used to directly control corresponding channel in fail mode. During normal mode, the control of the outputs by the control inputs is SPI programmable. This pin has a passive internal pull-down.
29	IN3	Input	Direct input #3	This input wakes up the device. This input pin is used to directly control corresponding channel in fail mode. During normal mode, the control of the outputs by the control inputs is SPI programmable. This pin has a passive internal pull-down.
30	IN4	Input	Direct input #4	This input wakes up the device. This input pin is used to directly control corresponding channel in fail mode. During normal mode the control of the outputs by the control inputs is SPI programmable. This pin has a passive internal pull-down.
31	LIMP	Input	Limp home	The fail mode can be activated by this digital input. This pin has a passive internal pull-down.
32	CLK	Input/Output	Device mode feedback Reference PWM clock	This pin is an input/output pin. It is used to report the device sleep-state information. It is also used to apply the reference PWM clock which is divided by 2^8 in normal operating mode. This pin has a passive internal pull-down.
33	VBAT	Power supply	Battery power supply	This exposed pad connects to the positive power supply and is the source of operational power for the device.

4 General product characteristics

4.1 Relationship between ratings and operating requirements

The analog portion of device is supplied by the voltage applied to the VBAT exposed pad. Thereby the supply of internal circuitry (logic in case of a V_{CC} disconnect, charge pump, gate drive,...) is derived from the VBAT pin.

In case of a reverse battery:

- the internal supply rail is protected (max. -16 V)
- the output drivers (OUT1:OUT4) are switched on, to reduce the power consumption in the drivers when using incandescent bulbs

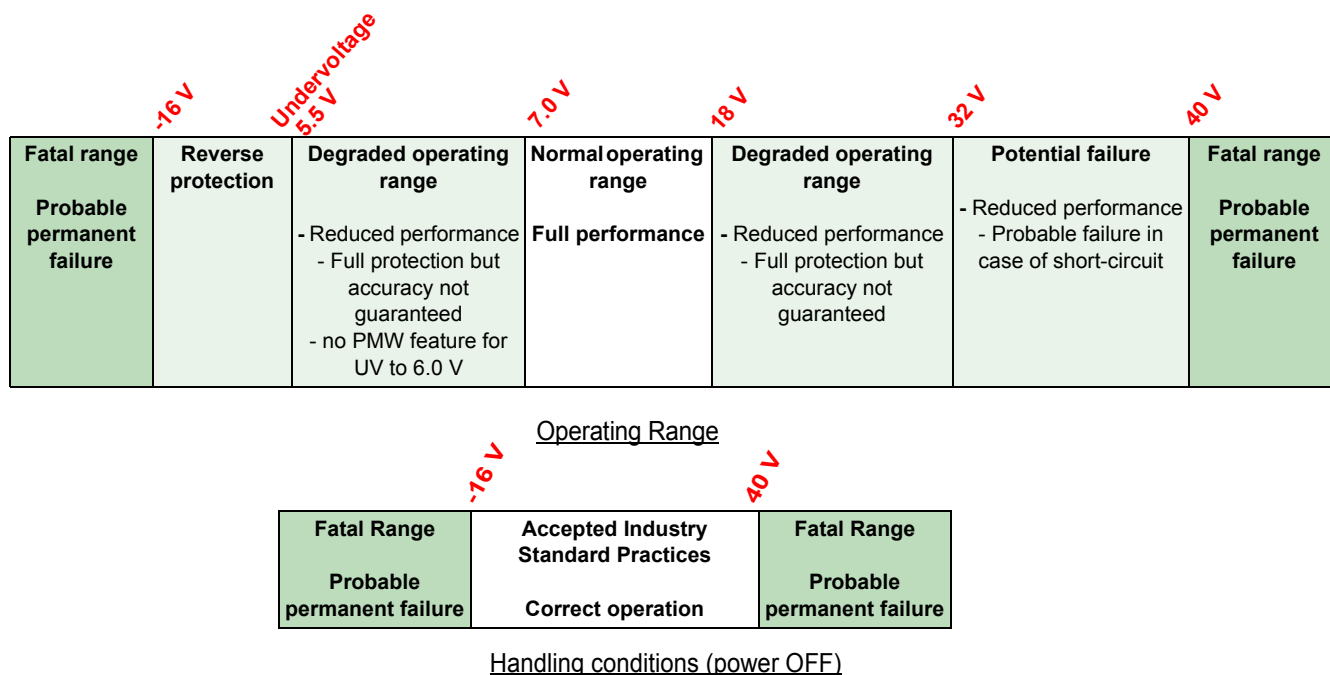


Figure 4. Ratings vs. operating requirements (VBAT pin)

The device's digital circuitry is powered by the voltage applied to the VCC pin. If VCC is disconnected, the logic part is supplied by the VBAT pin.

The output driver for SPI signals, CLK pin (wake feedback), and OUT6 are supplied by the VCC pin only. This pin must be protected externally in case of a reverse polarity, and in case of a high-voltage disturbance.

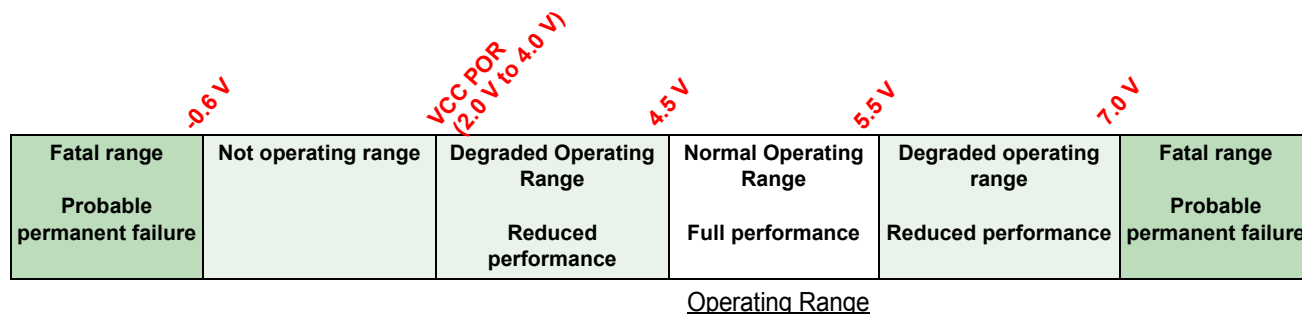


Figure 5. Ratings vs. operating requirements (VCC pin)

4.2 Maximum ratings

Table 3. Maximum ratings

All voltages are with respect to ground unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device.

Symbol	Description (rating)	Min.	Max.	Unit	Notes
Electrical ratings					
V _{BAT}	VBAT voltage range	-16	40	V	
V _{CC}	VCC logic supply voltage	-0.3	7.0	V	
V _{IN}	Digital input voltage <ul style="list-style-type: none"> • IN1:IN4 and LIMP • CLK, SI, SCLK, CSB, and RSTB 	-0.3 -0.3	40 20	V	(2)
V _{OUT}	Digital output voltage <ul style="list-style-type: none"> • SO, CSNS, SYNC, OUT6, CLK 	-0.3	20	V	(2)
I _{CL}	Negative digital input clamp current	–	5.0	mA	(3)
I _{OUT}	Power channel current <ul style="list-style-type: none"> • 8.0 mΩ channel • 21 mΩ channel 	– –	11 5.5	A	(4)
E _{CL}	Power channel clamp energy capability <ul style="list-style-type: none"> • 8.0 mΩ channel - Initial T_J = 25 °C • 8.0 mΩ channel - Initial T_J = 150 °C • 21 mΩ channel - Initial T_J = 25 °C • 21 mΩ channel - Initial T_J = 150 °C 	– – – –	200 100 75 50	mJ	(5)
V _{ESD}	ESD voltage <ul style="list-style-type: none"> • Human body model (HBM) - VBAT, power channel, and GND pins • Human body model (HBM) - All other pins • Charge device model (CDM) - Corner pins • Charge device model (CDM) - All other pins 	-8000 -2000 -750 -500	+8000 +2000 +750 +500	V	(6)

Notes

2. Exceeding voltage limits on those pins may cause a malfunction or permanent damage to the device.
3. Maximum current in negative clamping for IN1:IN4, LIMP, RSTB, CLK, SI, SO, SCLK, and CSB pins.
4. Continuous high-side output current rating so long as maximum junction temperature is not exceeded. Calculation of maximum output current using package thermal resistance is required.
5. Active clamp energy using single-pulse method (L = 2.0 mH, R_L = 0 Ω, V_{BAT} = 14 V). Refer to [6.1.4. Output clamps, page 37](#) section.
6. ESD testing is performed in accordance with the human body model (HBM) (C_{ZAP} = 100 pF, R_{ZAP} = 1500 Ω), and the charge device model.

4.3 Thermal characteristics

Table 4. Thermal ratings

All voltages are with respect to ground unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device.

Symbol	Description (rating)	Min.	Max.	Unit	Notes
Thermal ratings					
T_A T_J	Operating temperature • Ambient • Junction	-40 -40	+125 +150	°C	
T_{STG}	Storage temperature	-55	+ 150	°C	
T_{PPRT}	Peak package reflow temperature during reflow	–	260	°C	(7) (8)

Thermal resistance and package dissipation ratings

$R_{\theta JB}$	Junction-to-board	–	8.0	°C/W	(9)
$R_{\theta JA}$	Junction-to-ambient, natural convection, four-layer board (2s2p)	–	21	°C/W	(10) (11)
$R_{\theta JC}$	Junction-to-case (case top surface)	–	17	°C/W	(12)

Notes

- Pin soldering temperature limit is for 10 seconds maximum duration. Not designed for immersion soldering. Exceeding these limits may cause malfunction or permanent damage to the device.
- NXP's package reflow capability meets Pb-free requirements for JEDEC standard J-STD-020C. For peak package reflow temperature and moisture sensitivity levels (MSL), go to www.nxp.com, search by part number (remove prefixes/suffixes), enter the core ID to view all orderable parts, and review parametrics.
- Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
- Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
- Per JEDEC JESD51-6 with the board (JESD51-7) horizontal.
- Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).

4.4 Operating Conditions

This section describes the operating conditions of the device. Conditions apply to all the following data, unless otherwise noted.

Table 5. Operating conditions

All voltages are with respect to ground unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device.

Symbol	Ratings	Min.	Max.	Unit	Notes
V_{BAT}	Functional operating supply voltage - Device is fully functional. All features are operating.	7.0	18	V	
	Overvoltage range • Jump start • Load dump	– –	28 40	V	
	Reverse battery	-16	–	V	
V_{CC}	Functional operating supply voltage - Device is fully functional. All features are operating.	4.5	5.5	V	

4.5 Supply currents

This section describes the current consumption characteristics of the device.

Table 6. Supply currents

Characteristics noted under conditions $4.5\text{ V} \leq V_{CC} \leq 5.5\text{ V}$, $-40\text{ °C} \leq T_A \leq 125\text{ °C}$, $GND = 0\text{ V}$, unless otherwise noted. Typical values noted reflect the approximate parameter means at $T_A = 25\text{ °C}$ under nominal conditions, unless otherwise noted.

Symbol	Ratings	Min.	Typ.	Max.	Unit	Notes
VBAT current consumptionS						
I_{QVBAT}	Sleep mode measured at $V_{PWR} = 12\text{ V}$ <ul style="list-style-type: none"> • $T_A = 25\text{ °C}$ • $T_A = 125\text{ °C}$ 	–	1.2	5.0	μA	(13) (14)
I_{VBAT}	Operating mode measured at $V_{PWR} = 18\text{ V}$	–	7.0	8.0	mA	(14)
VCC current consumptionS						
I_{QVCC}	Sleep mode measured at $V_{CC} = 5.5\text{ V}$	–	0.05	5.0	μA	
I_{VCC}	Operating mode measured at $V_{PWR} = 5.5\text{ V}$ (SPI frequency 5.0 MHz)	–	2.8	4.0	mA	

Notes

13. With the OUT1:OUT4 power channels grounded.
14. With the OUT1:OUT4 power channels opened.

5 General IC functional description and application information

5.1 Introduction

The 12XS6 is the latest SMARTMOS achievement in automotive drivers for all types of centralized automotive lighting applications. It is an evolution of the successful 12XS3 by providing improved features of a complete family of devices using NXP's latest and unique technologies for the controller and the power stages.

It consists of a scalable family of devices compatible in terms of software driver and package footprint. It allows diagnosing the light-emitting diodes (LEDs) with an enhanced current sense precision with synchronization pin. It combines flexibility through daisy chainable SPI 5.0 MHz, extended digital and analog feedbacks, safety, and robustness. It integrates an enhanced PWM module with 8-bit duty cycle capability and PWM frequency prescaler per power channel.

5.2 Features

The main attributes of 12XS6 are:

- Dual, triple, quad, or penta high-side switches with overload, overtemperature, and undervoltage protection
- Control output for one external smart power switch
- 16-Bit SPI communication interface with daisy chain capability
- Dedicated control inputs for use in fail mode
- Analog feedback pin with SPI programmable multiplexer and sync signal
- Channel diagnosis by SPI communication
- Advanced current sense mode for LED usage
- Synchronous PWM module with external clock, prescaler, and multiphase feature
- Excellent EMC behavior
- Power net and reverse polarity protection
- Ultra low-power mode
- Scalable and flexible family concept
- Board layout compatible SOIC54 and SOIC32 package with exposed pad

5.3 Block diagram

The choice of multi-die technology in SOIC exposed pad package including low cost vertical trench FET power die associated with Smart Power control die lead to an optimized solution.

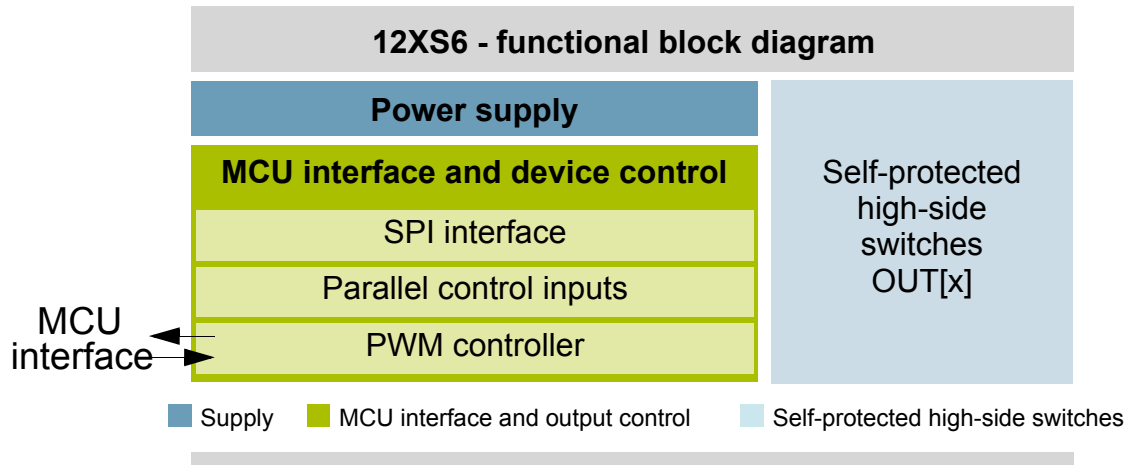


Figure 6. Functional block diagram

5.3.1 Self-protected high-side switches

OUT1:OUT4 are the output pins of the power switches. The power channels are protected against various kinds of short-circuits and have active clamp circuitry that may be activated when switching off inductive loads. Many protective and diagnostic functions are available.

5.3.2 Power supply

The device operates with supply voltages from 5.5 V to 40 V (V_{BAT}), but is full spec. compliant only between 7.0 V and 18 V. The VBAT pin supplies power to the internal regulator, analog, and logic circuit blocks. The VCC pin (5.0 V typ.) supplies the output register of the serial peripheral interface (SPI). Consequently, the SPI registers cannot be read without presence of V_{CC} . The employed IC architecture guarantees a low quiescent current in sleep mode.

5.3.3 MCU interface and device control

In normal mode, the power output channels are controlled by the embedded PWM module, which is configured by the SPI register settings. V_{CC} must be in the authorized range for bidirectional SPI communication. Failure diagnostics and configuration are also performed through the SPI port. The reported failure types are: open load, short-circuit to battery, severe short-circuit to ground, overcurrent, overtemperature, clock-fail, and under and overvoltage.

The device allows driving loads at different frequencies up to 400 Hz.

5.4 Functional description

The device has four fundamental operating modes: sleep, normal, fail, and power off. It possesses multiple high-side switches (power channels) each of which can be controlled independently:

- In normal mode by SPI interface. A second supply voltage (V_{CC}) is required for bidirectional SPI communication
- In fail mode by the corresponding direct inputs IN1:IN4. OUT6 is off in this mode

5.5 Modes of operation

The operating modes are based on the signals:

- wake = (IN1_ON) OR (IN2_ON) OR (IN3_ON) OR (IN4_ON) OR (RSTB). More details in [6.3.3.2. Logic I/O plausibility check, page 55](#) section
- fail = (SPI_fail) OR (LIMP). More details in [6.3.3.1. Loss of communication interface, page 55](#) section

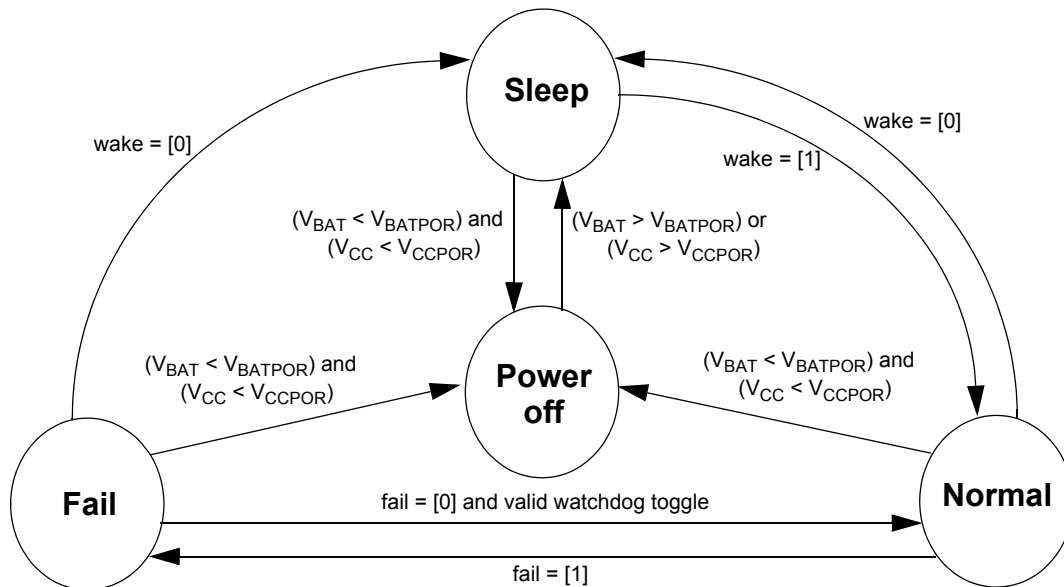


Figure 7. General IC operating modes

5.5.1 Power off mode

The power off mode is applied when V_{BAT} and V_{CC} are below the power on reset threshold ($V_{BAT\ POR}$, $V_{CC\ POR}$). In power off, no functionality is available but the device is protected by the clamping circuits. Refer to [6.2.3. Supply voltages disconnection, page 53](#).

5.5.2 Sleep mode

The sleep mode is used to provide ultra low current consumption. During sleep mode:

- the component is inactive and all outputs are disabled
- the outputs are protected by the clamping circuits
- the pull-up/pull-down resistors are present

The sleep mode is the default mode of the device after applying the supply voltages (V_{BAT} or V_{CC}) prior to any wake-up condition (wake = [0]). The wake-up from sleep mode is provided by the wake signal.

5.5.3 Normal mode

The normal mode is the regular operating mode of the device. The device is in normal mode, when the device is in the wake state (wake = [1]) and no fail condition (fail = [0]) is detected.

During normal mode:

- the power outputs are under control of the SPI
- the power outputs are controlled by the programmable PWM module
- the power outputs are protected by the overload protection circuit
- the control of the power outputs by SPI programming
- the digital diagnostic feature transfers status of the smart switch via the SPI
- the analog feedback output (CSNS and CSNS SYNC) can be controlled by the SPI

The channel control (CHx) can be summarized:

- CH1:4 controlled by ONx or iINx (if it is programmed by the SPI)
- CH6 controlled by ONx
- Rising CHx by definition means starting overcurrent window for OUT1:4

5.5.4 Fail mode

The device enters the fail mode, when:

- the LIMP input pin is high (logic [1])
- or the SPI failure is detected

During fail mode (wake = [1] & fail = [1]):

- the OUT1:OUT4 outputs are directly controlled by the corresponding control inputs (IN1:IN4)
- the OUT6 is turned off
- the PWM module is not available
- while no SPI control is feasible, the SPI diagnosis is functional (depending on the fail mode condition):
 - the SO must report the content of SO register defined by SOA0 to 3 bits
 - the outputs are fully protected in case of an overload, overtemperature, and undervoltage
 - no analog feedback is available
 - the max. output overcurrent profile is activated (OCLO and window times)
 - in case of an overload condition or undervoltage, the autorestart feature controls the OUT1:OUT4 outputs
 - in case of an overtemperature condition, OCHI1 detection, or severe short-circuit detection, the corresponding output is latched OFF until a new wake-up event.

The channel control (CHx) can be summarized:

- CH1:4 controlled by iINx, while the overcurrent windows are controlled by IN_ONx
- CH6 is off

5.5.5 Mode transitions

After a wake-up:

- a power on reset is applied and all SPI SI and SO registers are cleared (logic[0])
- the faults are blanked during t_{BLANKING}

The device enters in normal mode after start-up if following sequence is provided:

- V_{BAT} and V_{CC} power supplies must be above their undervoltage thresholds (sleep mode)
- generate wake-up event (wake =1) setting RSTB from 0 to 1

The device initialization is completed after 50 μsec (typ). During this time, the device is robust in case of V_{BAT} interrupts higher than 150 nsec.

The transition from “normal mode” to “fail mode” is executed immediately when a fail condition is detected.

During the transition, the SPI SI settings are cleared and the SPI SO registers are not cleared.

When the fail mode condition was a:

- LIMP input, WD toggle timeout, WD toggle sequence, or the SPI modulo 16 error, the SPI diagnosis is available during fail mode
- SI/SO stuck to static level, the SPI diagnosis is not available during fail mode

The transition from “fail mode” to “normal mode” is enabled, when:

- the fail condition is removed and
- two SPI commands are sent within a valid watchdog cycle (first WD=[0] and then WD=[1])

During this transition:

- all SPI SI and SO registers are cleared (logic[0])
- the DSF (device status flag) in the registers #1:#7 and the RCF (register clearer flag) in the device status register #1 are set (logic[1])

To detach the RCF diagnosis, a read command of the quick status register #1 must be performed.

5.6 SPI interface and configurations

5.6.1 Introduction

The SPI is used to:

- control the device in case of normal mode
- provide diagnostics in case of normal and fail mode

The SPI is a 16 Bit full-duplex synchronous data transfer interface with daisy chain capability.

The interface consists of four I/O lines with 5.0 V CMOS logic levels and termination resistors:

- The SCLK pin clocks the internal shift registers of the device
- The SI pin accepts data into the input shift register on the rising edge of the SCLK signal
- The SO pin changes its state on the rising edge of SCLK and reads out on the falling edge
- The CSB enables the SPI interface
 - with the leading edge of CSB the registers are loaded
 - while CSB is logic [0] SI/SO data are shifted
 - with the trailing edge of the CSB signal, SPI data is latched into the internal registers
 - when CSB is logic [1], the signals at the SCLK and SI pins are ignored and SO is high-impedance

When the RSTB input is:

- low (logic [0]), the SPI and the fault registers are reset. The Wake state then depends on the status of the input pins (IN_ON1:IN_ON4)
- high (logic[1]), the device is in Wake status and the SPI is enabled

The functionality of the SPI is checked by a plausibility check. In case of the SPI failure the device enters the fail mode.

5.6.2 SPI input register and bit descriptions

The first nibble of the 16-bit data word (D15:D12) serves as address bits.

Register	SI address				SI data												
	#	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
name	8	4 Bit address				WD	11 Bit address										

11 bits (D10:D1) are used as data bits.

The D11 bit is the WD toggle bit. This bit has to be toggled with each write command.

When the toggling of the bit is not executed within the WD timeout, the SPI fail is detected.

All register values are logic [0] after a reset. The predefined value is off/inactive unless otherwise noted.

5.6.3 SPI output register and bit descriptions

The first nibble of the 16-bit data word (D12:D15) serves as address bits.

All register values are logic [0] after a reset, except DSF and RCF bits. The predefined value is off/inactive unless otherwise noted.

Register	SO address				SO data													
	#	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	
not used	0	0	0	0	0	X	X	X	X	X	X	X	X	X	X	X	X	
quick status	1	0	0	0	1	FM	DSF	OVLf	OLf	CPF	RCF	CLKF	X	QSF4	QSF3	QSF2	QSF1	
CH1 status	2	0	0	1	0	FM	DSF	OVLf	OLf	res	OTS1	OTW1	OC21	OC11	OC01	OLON1	OLOFF1	
CH2 status	3	0	0	1	1	FM	DSF	OVLf	OLf	res	OTS2	OTW2	OC22	OC12	OC02	OLON2	OLOFF2	
CH3 status	4	0	1	0	0	FM	DSF	OVLf	OLf	res	OTS3	OTW3	OC23	OC13	OC03	OLON3	OLOFF3	
CH4 status	5	0	1	0	1	FM	DSF	OVLf	OLf	res	OTS4	OTW4	OC24	OC14	OC04	OLON4	OLOFF4	
device status	7	0	1	1	1	FM	DSF	OVLf	OLf	res	res	res	TMF	OvF	UVF	SPIf	iLIMP	
I/O status	8	1	0	0	0	FM	res	TOGGLE	iIN4	iIN3	iIN2	iIN1	X	OUT4	OUT3	OUT2	OUT1	
device ID	9	1	0	0	1	FM	res	res	res	DEVID 7	DEVID 6	DEVID 5	DEVID 4	DEVID 3	DEVID 2	DEVID 1	DEVID 0	
not used	10-14	address from 1010 to 1110				X	X	X	X	X	X	X	X	X	X	X	X	X
testmode	15	1	1	1	1	X	X	X	X	X	X	X	X	X	X	X	X	

QSFx #1	= quick status (OC or OTW or OTS or OLON or OLOFF)	#2-#6	OC2x	OC1x	OC0x	over current status
CLKF #1	= PWM clock fail flag		0	0	0	no overcurrent
RCF #1	= registers clear flag		0	0	1	OCH1/1
CPF #1	= charge pump flag		0	1	0	OCH2/2
OLf #1-#7	= open load flag (wired or of all OL signals)		0	1	1	OCH3/3
OVLf #1-#7	= over load flag (wired or of all OC and OTS signals)		1	0	0	OCL0
DSF #1-#7	= device status flag (RCF or UVF or OVF or CPF or CLKF or TMF)		1	0	1	OCHI/QD
FM #1-#8	= fail mode flag		1	1	0	SSC
OLOFFx #2-#6	= open load in off state status bit		1	1	1	not used
OLONx #2-#6	= open load in on state status bit	#9	DEVID2	DEVID1	DEVID0	device type
OTWx #2-#6	= over temperature warning bit		0	0	0	Penta3/2
OTSx #2-#6	= over temperature shutdown bit		0	0	1	Penta0/5
iLIMP #7	= limp input pin status		0	1	0	Quad2/2
SPIf #7	= SPI fail flag		0	1	1	Quad0/4
UVF #7	= under voltage flag		1	0	0	Triple1/2
OVF #7	= over voltage flag		1	0	1	Triple0/3
TMF #7	= testmode activation flag		1	1	0	res
OUTx #8	= status of VBAT/2 comparator (reported in real time)		1	1	1	res
iINx #8	= status of iINx signal (reported in real time)					
TOGGLE #8	= status of iINx_ON signals (iIN1_ON or iIN2_ON or iIN3_ON or iIN4_ON)					
DEVID0 ~ DEVID2 #9	= device type					
DEVID3 ~ DEVID4 #9	= device family					
DEVID5 ~ DEVID7 #9	= design status (incremented number)					

5.6.4 Timing diagrams

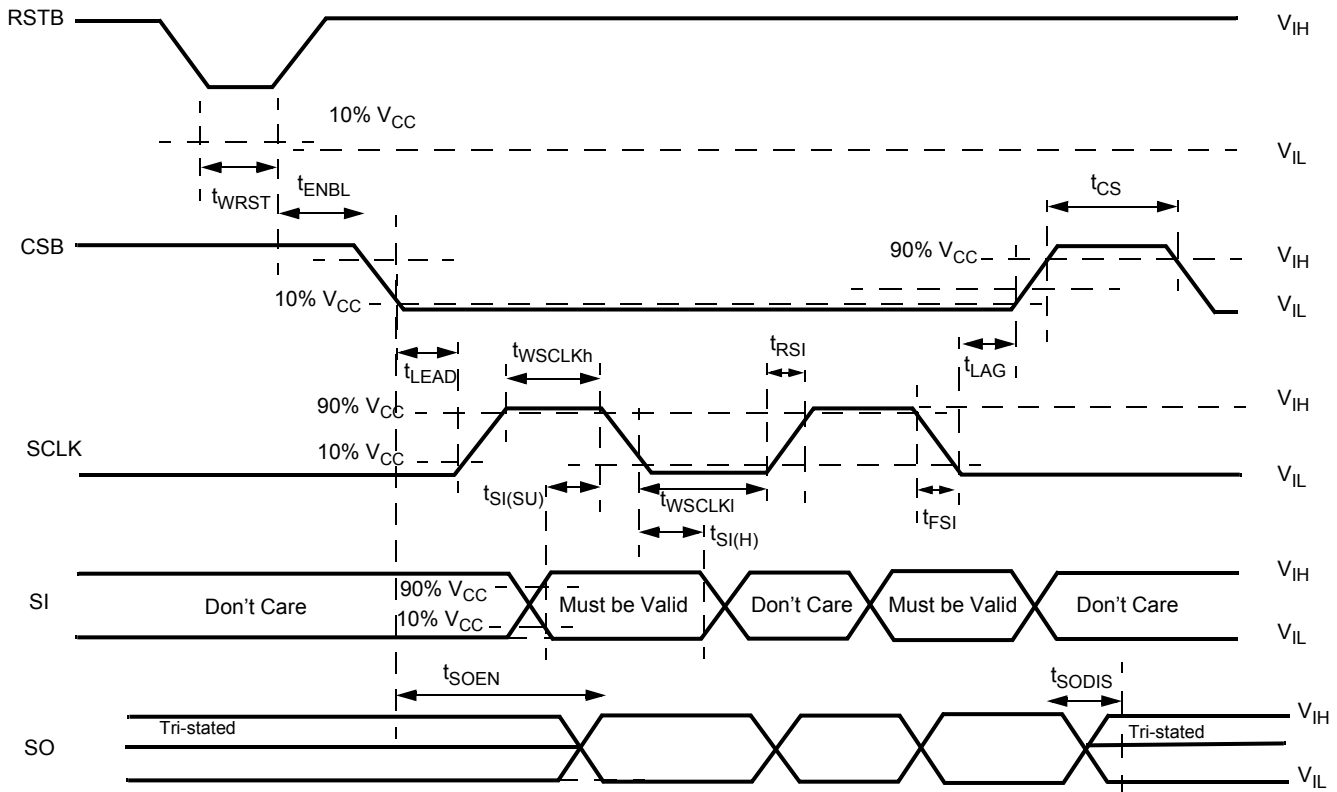


Figure 8. Timing requirements during SPI Communication

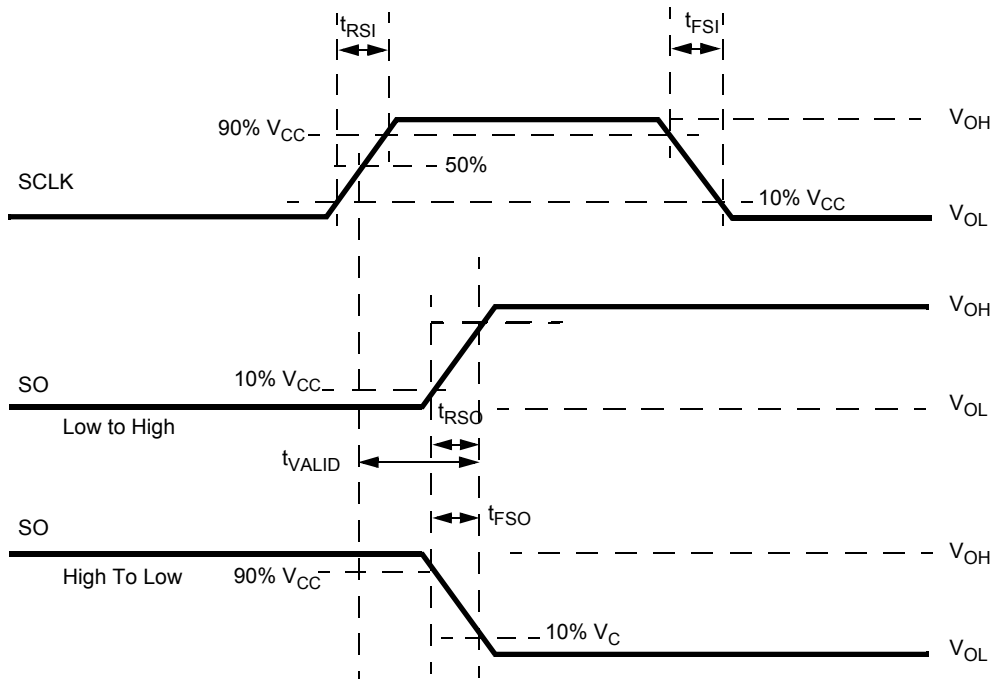


Figure 9. Timing diagram for serial output (SO) data communication

5.6.5 Electrical characterization

Table 7. Electrical characteristics

Characteristics noted under conditions $4.5\text{ V} \leq V_{CC} \leq 5.5\text{ V}$, $-40\text{ }^\circ\text{C} \leq T_A \leq 125\text{ }^\circ\text{C}$, $\text{GND} = 0\text{ V}$, unless otherwise noted. Typical values noted reflect the approximate parameter means at $T_A = 25\text{ }^\circ\text{C}$ under nominal conditions, unless otherwise noted.

Symbol	Characteristic	Min.	Typ.	Max.	Unit	Notes
SPI signals CSB, SI, SO, SCLK, SO						
f_{SPI}	SPI clock frequency	0.5	–	5.0	MHz	
V_{IH}	Logic input high state level (SI, SCLK, CSB, RSTB)	3.5	–	–	V	
$V_{\text{IH(WAKE)}}$	Logic input high state level for wake-up (RSTB)	3.75	–	–	V	
V_{IL}	Logic input low state level (SI, SCLK, CSB, RSTB)	–	–	0.85	V	
V_{OH}	Logic output high state level (SO)	$V_{CC} - 0.4$	–	–	V	
V_{OL}	Logic output low state level (SO)	–	–	0.4	V	
I_{IN}	Logic input leakage current in inactive state (SI = SCLK = RSTB = [0] and CSB = [1])	-0.5	–	+0.5	μA	
I_{OUT}	Logic output tri-state leakage current (SO from 0 V to V_{CC})	-10	–	+1.0	μA	
R_{PULL}	Logic input pull-up/pull-down resistor	25	–	100	$\text{k}\Omega$	
C_{IN}	Logic input capacitance	–	–	20	pF	(15)
$t_{\text{RST_DGL}}$	RSTB deglitch time	7.5	10	12.5	μs	
t_{SO}	SO rising and falling edges with 80 pF	–	–	20	ns	
t_{WCLKh}	Required high state duration of SCLK (required setup time)	80	–	–	ns	
t_{WCLKl}	Required low state duration of SCLK (required setup time)	80	–	–	ns	
t_{CS}	Required duration from the rising to the falling edge of CSB (required setup time)	1.0	–	–	μs	
t_{RST}	Required low state duration for reset RSTB	1.0	–	–	μs	
t_{LEAD}	Falling edge of CSB to rising edge of SCLK (required setup time)	320	–	–	ns	
t_{LAG}	Falling edge of SCLK to rising edge of CSB (required setup lag time)	100	–	–	ns	
$t_{\text{SI(SU)}}$	SI to falling edge of SCLK (required setup time)	20	–	–	ns	
$t_{\text{SI(H)}}$	Falling edge of SCLK to SI (required hold time of the SI signal)	20	–	–	ns	
t_{RSI}	SI, CSB, SCLK, Max. rise time allowing operation at maximum f_{SPI}	–	20	50	ns	
t_{FSI}	SI, CSB, SCLK, Max. fall time allowing operation at maximum f_{SPI}	–	20	50	ns	
$t_{\text{SO(EN)}}$	Time from falling edge of CSB to reach low-impedance on SO (access time)	–	–	60	ns	
$t_{\text{SO(DIS)}}$	Time from rising edge of CSB to reach tri-state on SO	–	–	60	ns	

Notes

15. Parameter is derived from simulations.

6 Functional block requirements and behaviors

6.1 Self-protected high-side switches description and application information

6.1.1 Features

Up to four power outputs are foreseen to drive automotive light applications. The outputs are optimized for driving automotive bulbs, but also HID ballasts, LEDs, and other primarily resistive loads.

The smart switches are controlled by use of high sophisticated gate drivers. The gate drivers provide:

- output pulse shaping
- output protections
- active clamps
- output diagnostics

6.1.2 Output pulse shaping

The outputs are controlled with a closed loop active pulse shaping to provide the best compromise between:

- low switching losses
- low EMC emission performance
- minimum propagation delay time

Depending on the programming of the prescaler setting register #12-1, #12-2, the switching speeds of the outputs are adjusted to the output frequency range of each channel.

The edge shaping must be designed according the following table:

divider factor	PWM freq. (Hz)		PWM period (ms)		d.c. range (hex)		d.c. range (LSB)		min. on/off duty cycle time (µs)
	min.	max.	min.	max.	min.	max.	min.	max	
4	25	100	10	40	03	FB	4	252	156
2	50	200	5	20	07	F7	8	248	156
1	100	400	2.5	10	07	F7	8	248	78

The edge shaping provides full symmetry for rising and falling transition:

- the slopes for the rising and falling edge are matched to provide the best EMC emission performance
- the shaping of the upper edges and the lower edges are matched to provide the best EMC emission performance
- the propagation delay time for the rising edge and the falling edge is matched to provide true duty cycle control of the output duty cycle error, ≤ 1 LSB at max. frequency
- a digital regulation loop is used to minimize the duty cycle error of the output signal

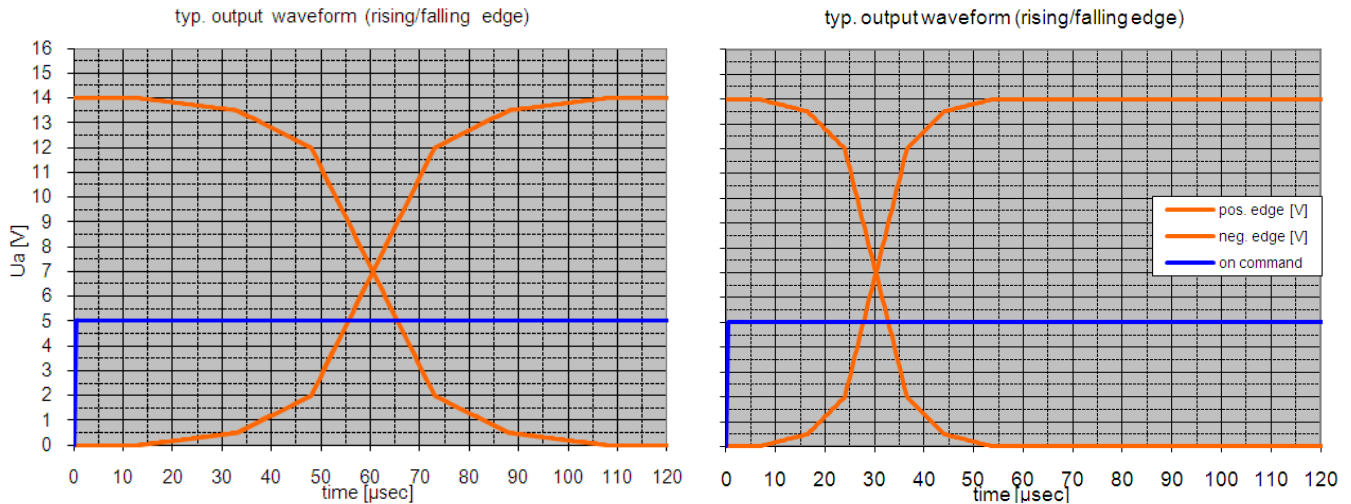


Figure 10. Typical power output switching (slow and fast slew rate)

6.1.2.1 SPI control and configuration

For optimized control of the outputs, a synchronous clock module is integrated. The PWM frequency and output timing during normal mode are generated from the clock input (CLK) by the integrated PWM module. In case of clock fail (very low frequency, very high frequency), the output duty cycle is 100%.

Each output (OUT1:OUT6) can be controlled by an individual channel control register:

Register	SI address				SI data												
	#	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
CHx control	2-7	channel address				WD	PH1x	PH0x	Onx	PWM7 x	PWM6 x	PWM5 x	PWM4 x	PWM3 x	PWM2 x	PWM1 x	PWM0 x

where:

- PH0x:PH1x: phase assignment of the output channel x
- ONx: on/off control including overcurrent window control of the output channel x
- PWM0x:PWM7x: 8-bit PWM value individually for each output channel x

The ONx bits are duplicated in the output control register #8 to control the outputs with either the CHx control register or the output control register.

The PRS1x:PRS0x prescaler settings can be set in the prescaler settings register #12-1 and #12-2.

The following changes of the duty cycle are performed asynchronous (with pos. edge of CSB signal):

- turn-on with 100% duty cycle (CHx = ON)
- change of duty cycle value to 100%
- turn-off (CHx = OFF)
- phase setting (PH0x:PH1x)
- prescaler setting (PRS1x:PRS0x)

A change in phase setting or prescaler setting during CHx = ON may cause an unwanted long on-time. Therefore it is recommended to turn-off the output(s) before execution of this change.

The following changes of the duty cycle are performed synchronous (with the next PWM cycle):

- turn-on with less than 100% duty cycle (OUTx = ONx)
- change of duty cycle value to less than 100%

A change of the duty cycle value can be achieved by a change of the:

- PWM0x:PWM7x bits in individual channel control register #2:#7
- GPWM EN1: GPWM EN6 bits (change between individual PWM and global PWM settings) in global PWM control register #9-1
- incremental/decremental register #14

The synchronization of the switching phases between different devices is provided by the PWM SYNC bit in the initialization 2 register #1.

On the SPI write into initialization 2 register (#1):

- initialization when the bit D1 (PWM SYNC) is logic[1], all counters of the PWM module are reset with the positive edge of the CSB, i.e. the phase synchronization is performed immediately within one SPI frame. It could help to synchronize different 12XS6 devices in the board
- when the bit D1 is logic[0], no action is executed

The switching frequency can be adjusted for the corresponding channel as described in the following table:

CLK freq. (kHz)		Prescaler setting		Divider factor	PWM freq. (Hz)		slew rate	PWM resolution)	
min.	max.	PRS1x	PRS0x		min.	max.		(Bit)	(steps)
25.6	102.4	0	0	4	25	100	slow	8	256
		0	1	2	50	200	slow		
		1	X	1	100	400	fast		

No PWM feature is provided in case of:

- Fail mode
- Clock input signal failure

6.1.2.2 Global PWM control

In addition to the individual PWM register, each channel can be assigned independently to a global PWM register.

The setting is controlled by the GPWM EN bits inside the global PWM control register #9-1. When no control by direct input pin is enabled and the GPWM EN bit is:

- low (logic[0]), the output is assigned to individual PWM (default status)
- high (logic[1]), the output is assigned to global PWM

The PWM value of the global PWM channel is controlled by the global PWM control register #9-2.

Table 8. Global PWM register

ONx	INEN1x	INEN0x	GPWM ENx	INx = 0		INx = 1	
				CHx	PWMx	CHx	PWMx
0	x	x	x	OFF	x	OFF	x
1	0	0	0	ON	individual	ON	individual
			1	ON	global	ON	global
	0	1	0	OFF	individual	ON	individual
			1	OFF	global	ON	global
	1	1	0	ON	individual	ON	global
			1	ON	global	ON	individual

When a channel is assigned to global PWM, the switching phase the prescaler and the pulse skipping are according the corresponding output channel setting.

6.1.2.3 Incremental PWM control

To reduce the control overhead during soft start/stop of bulbs (e.g. theatre dimming), an incremental PWM control feature is implemented.

With the incremental PWM control feature the PWM values of all internal channels OUT1:OUT4 can be incremented or decremented with one SPI frame.

The incremental PWM feature is not available for:

- the global PWM channel
- the external channel OUT6

The control is according the increment/decrement register #14:

- INCR SGN: sign of incremental dimming (valid for all channels)
- INCR 1x, INCR 0x increment/decrement

INCR SGN increment/decrement

0	decrement
1	increment

INCR 1x INCR 0x increment/decrement

0	0	no increment/decrement
0	1	4
1	0	8
1	1	16

This feature limits the duty cycle to the rails (00 resp. FF) to avoid any overflow.

6.1.2.4 Pulse skipping

Due to the output pulse shaping feature and the resulting switching delay time of the smart switches, duty cycles close to 0% resp. 100% can not be generated by the device. Therefore the pulse skipping feature (PSF) is integrated to interpolate this output duty cycle range in normal mode.

The pulse skipping provides a fixed duty cycle pattern with eight states to interpolate the duty cycle values between F7 (Hex) and FF (Hex). The range between 00 (Hex) and 07 (Hex) is not considered to be provided.

The pulse skipping feature:

- is available individually for the power output channels (OUT1:OUT4)
- is not available for the external channel (OUT6)

The feature is enabled with the PSF bits in the output control register #8.

When the corresponding PSF bit is:

- low (logic[0]), the pulse skipping feature is disabled on this channel (default status)
- high (logic[1]), the pulse skipping feature is enabled on this channel

PWM duty cycle			pulse skipping frame							
hex	dec	[%]	S0	S1	S2	S3	S4	S5	S6	S7
FF	256	100,00%	FF	FF	FF	FF	FF	FF	FF	FF
FE	255	99,61%	F7	FF	FF	FF	FF	FF	FF	FF
FD	254	99,22%	F7	FF	FF	FF	F7	FF	FF	FF
FC	253	98,83%	F7	FF	F7	FF	F7	FF	FF	FF
FB	252	98,44%	F7	FF	F7	FF	F7	FF	F7	FF
FA	251	98,05%	F7	F7	F7	FF	F7	FF	F7	FF
F9	250	97,66%	F7	F7	F7	FF	F7	F7	F7	FF
F8	249	97,27%	F7	F7	F7	F7	F7	F7	F7	FF
F7	248	96,88%								
F6	247	96,48%								
F5	246	96,09%								
F4	245	95,70%								
.	.	.								
.	.	.								
.	.	.								
03	4	1,56%								
02	3	1,17%								
01	2	0,78%								
00	1	0,39%								

6.1.2.5 Input control

Up to four dedicated control inputs (IN1:IN4) are foreseen to:

- wake-up the device
- fully control the corresponding output in case of fail mode
- control the corresponding output in case of normal mode

The control during normal mode is according the INEN0x and INEN1x bits in the input enable register #11. See [Table 8](#).

An input deglitcher is provided at each control input to avoid high frequency control of the outputs. The internal signal is called iINx.

The channel control (CHx) can be summarized:

- Normal mode:
 - CH1:4 controlled by ONx or INx (if it is programmed by the SPI)
 - CH6 controlled by ONx
 - Rising CHx by definition means starting overcurrent window for OUT1:4
- Fail mode:
 - CH1:4 controlled by iINx, while the over current windows are controlled by IN_ONx
 - CH6 are off

The input thresholds are logic level compatible, so the input structure of the pins are able to withstand battery voltage level (max.40 V) without damage. External current limit resistors (i.e. 1.0 kΩ:10 kΩ) can be used to handle reverse current conditions.

The inputs have an integrated pull-down resistor.

6.1.2.6 Electrical characterization

Table 9. Electrical characteristics

Characteristics noted under conditions $7.0\text{ V} \leq V_{\text{BAT}} \leq 18\text{ V}$, $-40\text{ °C} \leq T_{\text{A}} \leq 125\text{ °C}$, $\text{GND} = 0\text{ V}$, unless otherwise noted. Typical values noted reflect the approximate parameter means at $T_{\text{A}} = 25\text{ °C}$ under nominal conditions, unless otherwise noted.

Symbol	Characteristic	Min.	Typ.	Max.	Unit	Notes
Power outputs OUT1:OUT4						
$R_{\text{DS(on)}}$	On-resistance, drain-to-source for 8.0 mΩ power channel <ul style="list-style-type: none"> • $T_{\text{J}} = 25\text{ °C}$ • $T_{\text{J}} = 150\text{ °C}$ • $T_{\text{J}} = 25\text{ °C}$, $V_{\text{BAT}} = -12\text{ V}$ • $T_{\text{J}} = 150\text{ °C}$, $V_{\text{BAT}} = -12\text{ V}$ 	–	7.5	–	mΩ	
$R_{\text{DS(on)}}$	On-resistance, drain-to-source for 21 mΩ power channel <ul style="list-style-type: none"> • $T_{\text{J}} = 25\text{ °C}$ • $T_{\text{J}} = 150\text{ °C}$ • $T_{\text{J}} = 25\text{ °C}$, $V_{\text{BAT}} = -12\text{ V}$ • $T_{\text{J}} = 150\text{ °C}$, $V_{\text{BAT}} = -12\text{ V}$ 	–	19	–	mΩ	
$I_{\text{LEAK SLEEP}}$	Sleep mode output leakage current (output shorted to GND) per channel <ul style="list-style-type: none"> • $T_{\text{J}} = 25\text{ °C}$, $V_{\text{BAT}} = 12\text{ V}$ • $T_{\text{J}} = 125\text{ °C}$, $V_{\text{BAT}} = 12\text{ V}$ • $T_{\text{J}} = 25\text{ °C}$, $V_{\text{BAT}} = 35\text{ V}$ • $T_{\text{J}} = 125\text{ °C}$, $V_{\text{BAT}} = 35\text{ V}$ 	–	–	0.5	μA	
$I_{\text{OUT OFF}}$	Operational output leakage current in off-state per channel <ul style="list-style-type: none"> • $T_{\text{J}} = 25\text{ °C}$, $V_{\text{BAT}} = 18\text{ V}$ • $T_{\text{J}} = 125\text{ °C}$, $V_{\text{BAT}} = 18\text{ V}$ 	–	–	10	μA	
δ_{PWM}	Output PWM duty cycle range (measured at $V_{\text{OUT}} = V_{\text{BAT}/2}$) <ul style="list-style-type: none"> • Low frequency range (25 Hz to 100 Hz) • Medium frequency range (50 Hz to 200 Hz) • High frequency range (100 Hz to 400 Hz) 	4.0	–	252	LSB	
		8.0	–	248		
		8.0	–	248		

Table 9. Electrical characteristics

Characteristics noted under conditions $7.0\text{ V} \leq V_{\text{BAT}} \leq 18\text{ V}$, $-40\text{ }^\circ\text{C} \leq T_{\text{A}} \leq 125\text{ }^\circ\text{C}$, $\text{GND} = 0\text{ V}$, unless otherwise noted. Typical values noted reflect the approximate parameter means at $T_{\text{A}} = 25\text{ }^\circ\text{C}$ under nominal conditions, unless otherwise noted.

Symbol	Characteristic	Min.	Typ.	Max.	Unit	Notes
Power outputs OUT1:OUT4 (continued)						
SR	Rising and falling edges slew-rate at $V_{\text{BAT}} = 14\text{ V}$ (measured from $V_{\text{OUT}} = 2.5\text{ V}$ to $V_{\text{BAT}} - 2.5\text{ V}$) <ul style="list-style-type: none"> Low frequency range Medium frequency range High frequency range 	0.25 0.25 0.5	0.42 0.42 0.84	0.6 0.6 1.25	V/ μs	(16)
ΔSR	Rising and falling edges slew rate matching at $V_{\text{BAT}} = 14\text{ V}$ (SRr/SRf)	0.9	1.0	1.1		(16)
t_{DLY}	Turn-on and turn-off delay times at $V_{\text{BAT}} = 14\text{ V}$ <ul style="list-style-type: none"> Low frequency range Medium frequency range High frequency range 	20 20 10	60 60 30	100 100 50	μs	(16)
Δt_{DLY}	Turn-on and turn-off delay times matching at $V_{\text{BAT}} = 14\text{ V}$ <ul style="list-style-type: none"> Low frequency range Medium frequency range High frequency range 	-20 -20 -10	0.0 0.0 0.0	20 20 10	μs	(16)
$t_{\text{OUTPUT SD}}$	Shutdown delay time in case of fault	0.5	2.5	4.5	μs	
Reference PWM clock						
f_{CLK}	Clock input frequency range	25.6	–	102.4	kHz	

Notes

16. With nominal resistive load: 2.5 Ω and 5.0 Ω respectively for 8.0 m Ω and 21 m Ω channel.

6.1.3 Output protections

The power outputs are protected against fault conditions in normal and fail mode in case of:

- overload conditions
- harness short-circuit
- overcurrent protection against ultra-low resistive short-circuit conditions thanks to smart overcurrent profile and severe short-circuit protection
- overtemperature protection including overtemperature warning
- under and overvoltage protections
- charge pump monitoring
- reverse battery protection

In case a fault condition is detected, the corresponding output is commanded off immediately after the deglitch time $t_{\text{FAULT SD}}$.

The turn-off in case of a fault shutdown (OCHI1, OCHI2, OCHI3, OCLO, OTS, UV, CPF, OLOFF) is provided by the FTO feature (fast turn-off).

The FTO:

- does not use edge shaping
- is provided with high slew rate to minimize the output turn-off time $t_{\text{OUTPUT SD}}$, in regards to the detected fault
- uses a latch which keeps the FTO active during an undervoltage condition ($0 \leq V_{\text{BAT}} \leq V_{\text{BAT UVF}}$)