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Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

Email & Skype: info@chipsmall.com Web: www.chipsmall.com

Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China







scale Semiconductor Technical Data

Document Number: MC33931

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√RoHS

5.0 A Throttle Control H-bridge

The 33931 is a monolithic H-Bridge Power IC in a robust thermally enhanced package. It is designed primarily for automotive electronic throttle control, but is applicable to any low voltage DC servo motor control application within the current and voltage limits stated in this specification.

The 33931 H-Bridge is able to control inductive loads with currents up to 5.0 A peak. RMS current capability is subject to the degree of heatsinking provided to the device package. Internal peak-current limiting (regulation) is activated at load currents above 6.5 A ± 1.5 A. Output loads can be pulse-width modulated (PWM-ed) at frequencies up to 11 kHz. A load current feedback feature provides a proportional (0.24% of the load current) current output suitable for monitoring by a microcontroller's A/D input. A Status flag output reports under-voltage, over-current, and over-temperature fault conditions.

Two independent inputs provide polarity control of two half-bridge totem-pole outputs. The disable inputs are provided to force the H-bridge outputs to tri-state (high-impedance off-state).

Features

- 5.0 to 28 V continuous operation (transient operation from 5.0 to 40 V)
- 235 mΩ maximum R_{DS(ON)} @ T_J=150 °C (each H-bridge MOSFET)
- 3.0 V and 5.0 V TTL / CMOS logic compatible inputs
- Over-current limiting (regulation) via internal constant-off-time PWM
- Output short-circuit protection (short to VPWR or GND)
- · Temperature-dependant current-limit threshold reduction
- · All inputs have an internal source/sink to define the default (floating input) states
- Sleep mode with current draw < 50 μA

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THROTTLE CONTROL H-BRIDGE



HEAT SINK





EK SUFFIX (PB-FREE) 98ARL10543D 32-PIN SOICW-EP

ORDERING INFORMATION						
Device (Add R2 Suffix for Tape and Reel)	Temperature Range (T _A)	Package				
MC33931VW	-40 to 125°C	44 HSOP				
MC33931EK	-40 to 125 C	32 SOICW-EP				

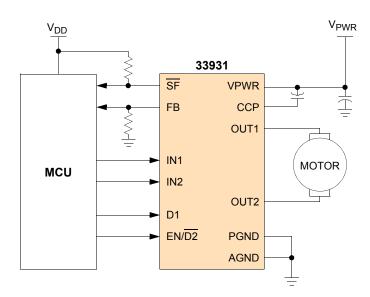


Figure 1. MC33931 Simplified Application Diagram





INTERNAL BLOCK DIAGRAM

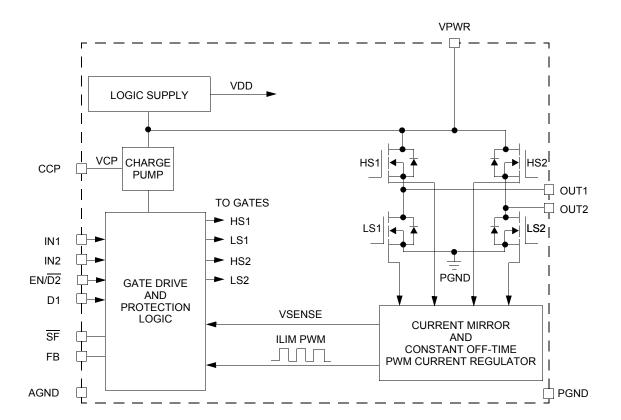


Figure 2. 33931 Simplified Internal Block Diagram



PIN CONNECTIONS

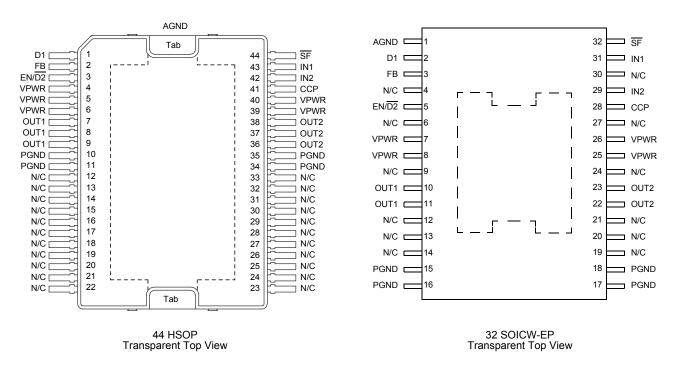


Figure 3. 33931 Pin Connections

Table 1. 33931 Pin Definitions

A functiona	A functional description of each pin can be found in the Functional Description section beginning on page 11.							
Pin HSOP (VW)	Pin SOICW-EP (EK)	Pin Name	Pin Function	Formal Name	Definition			
1	2	D1	Logic Input	Disable Input 1 (Active High)	When D1 is logic HIGH, both OUT1 and OUT2 are tri-stated. Schmitt trigger input with ~80 μ A source so default condition = disabled.			
2	3	FB	Analog Output	Feedback	The load current feedback output provides ground referenced 0.24% of the high side output current. (Tie to GND through a resistor if not used.)			
3	5	EN/D2	Logic Input	Enable Input	When EN/ $\overline{D2}$ is logic HIGH the H-bridge is operational. When EN/ $\overline{D2}$ is logic LOW, the H-bridge outputs are tri-stated and placed in Sleep mode. (logic input with ~80 μ A sink so default condition = Sleep mode.)			
4-6, 40, 39	7, 8, 25, 26	VPWR	Power Input	Positive Power Supply	These pins must be connected together physically as close as possible and directly soldered down to a wide, thick, low resistance supply plane on the PCB.			
7-9	10, 11	OUT1	Power Output	H-bridge Output 1	Source of HS1 and drain of LS1.			
10, 11, 34, 35	15-18	PGND	Power Ground	Power Ground	High-current power ground pins must be connected together physically as close as possible and directly soldered down to a wide, thick, low resistance ground plane on the PCB.			
36-38	22, 23	OUT2	Power Output	H-bridge Output 2	Source of HS2 and drain of LS2.			



Table 1. 33931 Pin Definitions (continued)

A functional description of each pin can be found in the Functional Description section beginning on page 11. Pin Pin Pin Pin **HSOP** SOICW-EP **Formal Name** Definition **Function** Name (VW) (EK) 41 28 CCP Analog Charge Pump External reservoir capacitor connection for the internal charge pump; connected to VPWR. Allowable values are 30 nF to Output Capacitor 100 nF. **Note:** This capacitor is required for the proper performance of the device. Logic input control of OUT2;e.g., when IN2 is logic HIGH, OUT2 42 29 IN2 Input 2 Logic Input is set to VPWR, and when IN2 is logic LOW, OUT2 is set to PGND. (Schmitt trigger Input with ~80 μA source so default condition = OUT2 HIGH.) Logic input control of OUT1; e.g., when IN1 is logic HIGH, OUT1 43 31 IN1 Logic Input Input 1 is set to VPWR, and when IN1 is logic LOW, OUT1 is set to PGND. (Schmitt trigger Input with ~80 μ A source so default condition = OUT1 HIGH.) Open drain active LOW Status Flag output (requires an external 32 SF 44 Logic Status Flag Output -(Active Low) pull-up resistor to V_{DD}. Maximum permissible load current < 0.5 mA. Maximum V_{SFLOW} < 0.4 V @ 0.3 mA. Maximum Open Drain permissible pull-up voltage < 7.0 V.) TAB 1 AGND The low-current analog signal ground must be connected to Analog Analog Signal PGND via low-impedance path (<10 m Ω , 0 Hz to 20 kHz). Ground Ground 12-33 4, 6, 9, 12-14, N/C No Connect Pin is not used None 19-21, 24, 27, 30 ΕP ΕP Thermal Exposed TAB is also the main heatsinking path for the device and Exposed Pad must be connected to GND. Pad



ELECTRICAL CHARACTERISTICS

MAXIMUM RATINGS

Table 2. Maximum Ratings

All voltages are with respect to ground, unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device. These parameters are not production tested.

Ratings	Symbol	Value	Unit
ELECTRICAL RATINGS	1		
Power Supply Voltage			V
Normal Operation (Steady-state)	V _{PWR(SS)}	-0.3 to 28	
Transient Over-voltage ⁽¹⁾	V _{PWR(T)}	-0.3 to 40	
Logic Input Voltage ⁽²⁾	V _{IN}	-0.3 to 7.0	V
SF Output ⁽³⁾	V _{SF}	-0.3 to 7.0	V
Continuous Output Current ⁽⁴⁾	I _{OUT(CONT)}	5.0	А
ESD Voltage ⁽⁵⁾			V
Human Body Model	V _{ESD1}	±2000	
Machine Model	V _{ESD2}	±200	
Charge Device Model			
Corner Pins		±750	
All Other Pins		±500	
THERMAL RATINGS			•
Storage Temperature	T _{STG}	- 65 to 150	°C
Operating Temperature ⁽⁶⁾			°C
Ambient	T _A	-40 to 125	
Junction	T _J	-40 to 150	
Peak Package Reflow Temperature During Reflow ^{(7),(8)}	T _{PPRT}	Note 8	°C
Approximate Junction-to-Case Thermal Resistance (9)	$R_{ heta JC}$	<1.0	°C/W

Notes

- Device will survive repetitive transient over-voltage conditions for durations not to exceed 500 ms @ duty cycle not to exceed 10%.
 External protection is required to prevent device damage in case of a reverse battery condition.
- 2. Exceeding the maximum input voltage on IN1, IN2, EN/D2 or D1 may cause a malfunction or permanent damage to the device.
- 3. Exceeding the pull-up resistor voltage on the open drain SF pin may cause permanent damage to the device.
- Continuous output current capability is dependent on sufficient package heatsinking to keep junction temperature ≤150°C.
- 5. ESD testing is performed in accordance with the Human Body Model (C_{ZAP} = 100 pF, R_{ZAP} = 1500 Ω), Machine Model (C_{ZAP} = 200 pF, R_{ZAP} = 0 Ω), and the Charge Device Model (CDM), Robotic (C_{ZAP} = 4.0 pF).
- 6. The limiting factor is junction temperature, taking into account the power dissipation, thermal resistance, and heat sinking provided. Brief non-repetitive excursions of junction temperature above 150°C can be tolerated, provided the duration does not exceed 30 seconds maximum. (Non-repetitive events are defined as not occurring more than once in 24 hours.)
- 7. Pin soldering temperature limit is for 10 seconds maximum duration. Not designed for immersion soldering. Exceeding these limits may cause malfunction or permanent damage to the device.
- 8. Freescale's Package Reflow capability meets Pb-free requirements for JEDEC standard J-STD-020C. For Peak Package Reflow Temperature and Moisture Sensitivity Levels (MSL), Go to www.freescale.com, search by part number [e.g. remove prefixes/suffixes and enter the core ID to view all orderable parts. (i.e. MC33xxxD enter 33xxx), and review parametrics.
- 9. Exposed heatsink pad plus the power and ground pins comprise the main heat conduction paths. The actual $R_{\theta JB}$ (junction-to-PC board) values will vary depending on solder thickness and composition and copper trace thickness and area. Maximum current at maximum die temperature represents ~16 W of conduction loss heating in the diagonal pair of output MOSFETs. Therefore, the $R_{\theta JA}$ must be <5.0 °C/W for maximum current at 70 °C ambient. Module thermal design must be planned accordingly.



STATIC ELECTRICAL CHARACTERISTICS

Table 3. Static Electrical Characteristics

Characteristics noted under conditions $5.0 \text{ V} \le \text{V}_{PWR} \le 28 \text{ V}$, $-40^{\circ}\text{C} \le \text{T}_{A} \le 125^{\circ}\text{C}$, GND = 0 V, unless otherwise noted. Typical values noted reflect the approximate parameter means at T_{A} = 25°C under nominal conditions, unless otherwise noted.

Characteristic	Symbol	Min	Тур	Max	Unit
POWER INPUTS (VPWR)					
Operating Voltage Range ⁽¹⁰⁾					V
Steady-state	V _{PWR(SS)}	5.0	_	28	
Transient (t < 500 ms) $^{(11)}$	$V_{PWR(t)}$	_	_	40	
Sleep State Supply Current ⁽¹²⁾	I _{PWR(SLEEP)}				μА
$EN/\overline{D2}$ = Logic [0], IN1, IN2, D1 = Logic [1], and I _{OUT} = 0A	· ····(OLLLI)	_	_	50	
Standby Supply Current (Part Enabled)	I _{PWR(STANDBY)}				mA
I _{OUT} = 0 A, V _{EN} = 5.0 V	(_	_	20	
Under-voltage Lockout Thresholds					
V _{PWR} (FALLING)	V _{UVLO(ACTIVE)}	4.15	_	_	V
V _{PWR(RISING)}	V _{UVLO(INACTIVE)}	-	_	5.0	V
Hysteresis	V _{UVLO(HYS)}	150	200	350	mV
CHARGE PUMP			l	l	Į.
Charge Pump Voltage (CP Capacitor = 33 nF), No PWM	$V_{CP} - V_{PWR}$				V
V _{PWR} = 5.0 V		3.5	_	_	
V _{PWR} = 28 V		_	_	12	
Charge Pump Voltage (CP Capacitor = 33 nF), PWM = 11 kHz,	V _{CP} -V _{PWR}				V
V _{PWR} = 5.0 V		3.5	_	_	
V _{PWR} = 28 V		_	_	12	
CONTROL INPUTS			l .	l .	
Operating Input Voltage (IN1, IN2, D1, EN/D2)	VI	_	_	5.5	V
Input Voltage (IN1, IN2, D1, EN/\overline{D2})					
Logic Threshold HIGH	V _{IH}	2.0	_	_	V
Logic Threshold LOW	V_{IL}	_	_	1.0	V
Hysteresis	V _{HYS}	250	400	_	mV
Logic Input Currents, VPWR = 5.0 V	I _{IN}				μА
Input EN/ $\overline{D2}$ (internal pull-downs), $V_{IH} = 5.0 \text{ V}$		20	80	200	
Inputs IN1, IN2, D1 (internal pull-ups), VIL = 0 V		-200	-80	-20	

Notes

- 10. Device specifications are characterized over the range of 8.0 V \leq V_{PWR} \leq 28 V. Continuous operation above 28 V may degrade device reliability. Device is operational down to 5.0 V, but below 8.0 V the output resistance may increase by 50 percent.
- Device will survive the transient over-voltage indicated for a maximum duration of 500 ms. Transient not to be repeated more than once every 10 seconds.
- 12. $I_{PWR(SLEEP)}$ is with Sleep Mode activated and EN/ $\overline{D2}$, = logic [0], and IN1, IN2, D1 = logic [1] or with these inputs left floating.



Table 3. Static Electrical Characteristics (continued)

Characteristics noted under conditions 5.0 V \leq V_{PWR} \leq 28 V, -40°C \leq T_A \leq 125°C, GND = 0 V, unless otherwise noted. Typical values noted reflect the approximate parameter means at T_A = 25°C under nominal conditions, unless otherwise noted.

Characteristic	Symbol	Min	Тур	Max	Unit
POWER OUTPUTS OUT1, OUT2		1	1	1	1
Output-ON Resistance ⁽¹⁴⁾ , I _{LOAD} = 3.0 A	R _{DS(ON)}				mΩ
V_{PWR} = 8.0 V, T_{J} = 25 °C	, ,	_	120	_	
V_{PWR} = 8.0 V, T_{J} = 150 °C		_	_	235	
V _{PWR} = 5.0 V, T _J = 150 °C		_	_	325	
Output Current Regulation Threshold	I _{LIM}				Α
$T_J < T_{FB}$		5.2	6.5	8.0	
$T_J \ge T_{FB}$ (Foldback Region - see Figure 9 and Figure 11) ⁽¹³⁾		_	4.2	-	
High Side Short Circuit Detection Threshold (Short Circuit to Ground) ⁽¹³⁾	I _{SCH}	11	13	16	Α
Low Side Short Circuit Detection Threshold (Short Circuit to V _{PWR}) ⁽¹³⁾	I _{SCL}	9.0	11	14	Α
Output Leakage Current ⁽¹⁵⁾ , Outputs off, V _{PWR} = 28 V	I _{OUTLEAK}				μΑ
$V_{OUT} = V_{PWR}$		_	_	100	
V _{OUT} = Ground		-60	_	_	
Output MOSFET Body Diode Forward Voltage Drop, I _{OUT} = 3.0 A	V _F	_	_	2.0	V
Over-temperature Shutdown ⁽¹³⁾					°C
Thermal Limit @ T _J	T_{LIM}	175	_	200	
Hysteresis @ T _J	T_{HYS}	_	12	_	
Current Foldback at T _J ⁽¹³⁾	T _{FB}	165	_	185	°C
Current Foldback to Thermal Shutdown Separation ⁽¹³⁾	T _{SEP}	10	_	15	°C

HIGH SIDE CURRENT SENSE FEEDBACK

Feedback Current (pin FB sourcing current) ⁽¹⁶⁾	I _{FB}				
$I_{OUT} = 0 \text{ mA}$		0.0	_	50	μΑ
I _{OUT} = 300 mA		0.0	270	750	μА
I _{OUT} = 500 mA		0.35	0.775	1.56	mA
I _{OUT} = 1.5 A		2.86	3.57	4.28	mA
I _{OUT} = 3.0 A		5.71	7.14	8.57	mA
I _{OUT} = 6.0 A		11.43	14.29	17.15	mA

STATUS FLAG⁽¹⁷⁾

Status Flag Leakage Current ⁽¹⁸⁾	ISFLEAK				μА
V _{SF} = 5.0 V		_	_	5.0	
Status Flag SET Voltage ⁽¹⁹⁾	V _{SFLOW}				V
$I_{\overline{SF}} = 300 \mu\text{A}$		_	_	0.4	

Notes

- 13. This parameter is Guaranteed By Design.
- 14. Output-ON resistance as measured from output to VPWR and from output to GND.
- 15. Outputs switched OFF via D1 or EN/D2.
- 16. Accuracy is better than 20% from 0.5 A to 6.0 A. Recommended terminating resistor value: R_{FB} = 270 Ω .
- 17. Status Flag output is an open drain output requiring a pull-up resistor to logic V_{DD} .
- 18. Status Flag Leakage Current is measured with Status Flag HIGH and not SET.
- 19. Status Flag Set Voltage measured with Status Flag LOW and SET with $I_{\overline{SF}}$ = 300 μ A. Maximum allowable sink current from this pin is <500 μ A. Maximum allowable pull-up voltage < 7.0 V.

DYNAMIC ELECTRICAL CHARACTERISTICS

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Table 4. Dynamic Electrical Characteristics

Characteristics noted under conditions $5.0 \text{ V} \leq \text{V}_{\text{PWR}} \leq 28 \text{ V}, -40^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 125^{\circ}\text{C}$, GND = 0 V, unless otherwise noted. Typical values noted reflect the approximate parameter means at T_{A} = 25°C under nominal conditions, unless otherwise noted.

Characteristic	Symbol	Min	Тур	Max	Unit
TIMING CHARACTERISTICS			•	•	•
PWM Frequency ⁽²⁰⁾	f _{PWM}	-	_	11	kHz
Maximum Switching Frequency During Current Limit Regulation (21)	f _{MAX}	_	_	20	kHz
Output ON Delay ⁽²²⁾	t _{DON}				μS
V _{PWR} = 14 V		-	_	18	
Output OFF Delay ⁽²²⁾	t _{DOFF}				μS
V _{PWR} = 14 V		-	_	12	
I _{LIM} Output Constant-OFF Time ⁽²³⁾ (25)	t _A	15	20.5	32	μS
I _{LIM} Blanking Time ⁽²⁴⁾ (25)	t _B	12	16.5	27	μS
Disable Delay Time ⁽²⁶⁾	t _{DDISABLE}	-	_	8.0	μS
Output Rise and Fall Time ⁽²⁷⁾	t _F , t _R	1.5	3.0	8.0	μS
Short-circuit/Over-temperature Turn-OFF (Latch-OFF) Time ^{(28),(29)}	t _{FAULT}	_	_	8.0	μS
Power-ON Delay Time ⁽²⁹⁾	t _{POD}	_	1.0	5.0	ms
Output MOSFET Body Diode Reverse Recovery Time ⁽²⁹⁾	t _{RR}	75	100	150	ns
Charge Pump Operating Frequency ⁽²⁹⁾	f _{CP}	_	7.0	-	MHz

Notes

- 20. The maximum PWM frequency should be limited to frequencies < 11 kHz in order to allow the internal high side driver circuitry time to fully enhance the high side MOSFETs.
- 21. The internal current limit circuitry produces a constant-OFF-time Pulse Width Modulation of the output current. The output load's inductance, capacitance, and resistance characteristics affect the total switching period (OFF-time + ON-time), and thus the PWM frequency during current limit.
- *Output Delay is the time duration from 1.5 V on the IN1 or IN2 input signal to the 20% or 80% point (dependent on the transition direction) of the OUT1 or OUT2 signal. If the output is transitioning HIGH-to-LOW, the delay is from 1.5 V on the input signal to the 80% point of the output response signal. If the output is transitioning LOW-to-HIGH, the delay is from 1.5 V on the input signal to the 20% point of the output response signal. See Figure 4, page 9.
- 23. The time during which the internal constant-OFF time PWM current regulation circuit has tri-stated the output bridge.
- 24. The time during which the current regulation threshold is ignored so that the short-circuit detection threshold comparators may have time to act.
- 25. Parameter is Guaranteed By Characterization.
- 26. * Disable Delay Time measurement is defined in Figure 5, page 9.
- 27. Rise Time is from the 10% to the 90% level and Fall Time is from the 90% to the 10% level of the output signal with V_{PWR} = 14 V, R_{LOAD} = 3.0 ohm. See Figure 6, page 9.
- 28. Load currents ramping up to the current regulation threshold become limited at the I_{LIM} value (see Figure 7). The short-circuit currents possess a di/dt that ramps up to the I_{SCH} or I_{SCL} threshold during the I_{LIM} blanking time, registering as a short-circuit event detection and causing the shutdown circuitry to force the output into an immediate tri-state latch-OFF (see Figure 8). Operation in Current Limit mode may cause junction temperatures to rise. Junction temperatures above ~160 °C will cause the output current limit threshold to "fold back", or decrease, until ~175 °C is reached, after which the T_{LIM} thermal latch-OFF will occur. Permissible operation within this fold back region is limited to non-repetitive transient events of duration not to exceed 30 seconds (see Figure 9).
- 29. Parameter is Guaranteed By Design.



TIMING DIAGRAMS

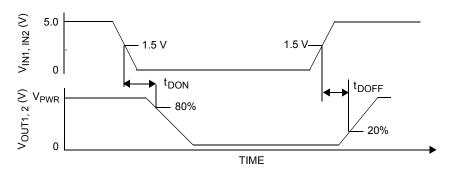


Figure 4. Output Delay Time

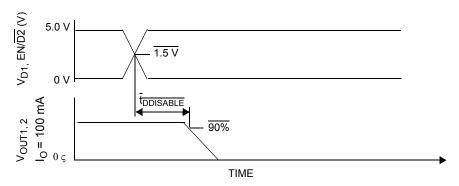


Figure 5. Disable Delay Time

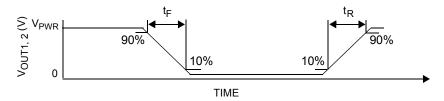


Figure 6. Output Switching Time

Overload Condition

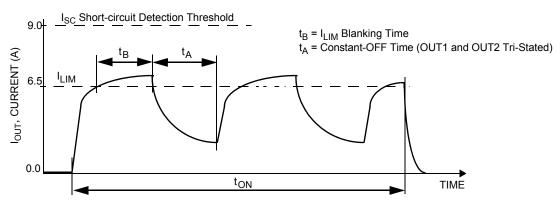
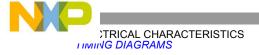


Figure 7. Current Limit Blanking Time and Constant-OFF Time

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9.0 Isc Short-circuit Detection Threshold Hard Short Occurs 6.5 ILIM TIME

Figure 8. Short-circuit Detection Turn-OFF Time $\mathbf{t}_{\text{FAULT}}$

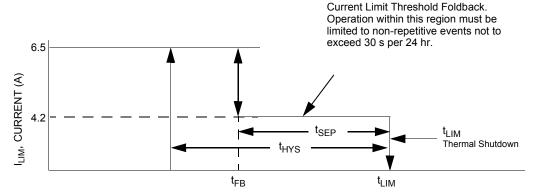


Figure 9. Output Current Limiting Foldback Region



FUNCTIONAL DESCRIPTION

INTRODUCTION

Numerous protection and operational features (speed, torque, direction, dynamic breaking, PWM control, and closed-loop control) make the 33931 a very attractive, cost-effective solution for controlling a broad range of small DC motors. The 33931 outputs are capable of supporting peak DC load currents of up to 5.0 A from a 28 V V_{PWR} source. An internal charge pump and gate drive circuitry are provided that can support external PWM frequencies up to 11 kHz.

The 33931 has an analog feedback (current mirror) output pin (the FB pin) that provides a constant-current source ratioed to the active high side MOSFETs' current. This can be used to provide "real time" monitoring of output current to facilitate closed-loop operation for motor speed/torque control, or for the detection of open load conditions.

Two independent inputs, IN1 and IN2, provide control of the two totem-pole half-bridge outputs. Two independent disable inputs, D1 and EN/D2, provide the means to force the H-bridge outputs to a high-impedance state (all H-Bridge switches OFF). The EN/D2 pin also controls an enable

function that allows the IC to be placed in a power-conserving Sleep mode.

The 33931 has output current limiting (via constant OFF-time PWM current regulation), output short-circuit detection with latch-OFF, and over-temperature detection with latch-OFF. Once the device is latched-OFF due to a fault condition, either of the Disable inputs (D1 or EN/ $\overline{D2}$), or V_{PWR} must be "toggled" to clear the status flag.

Current limiting (Load Current Regulation) is accomplished by a constant-OFF time PWM method using current limit threshold triggering. The current limiting scheme is unique in that it incorporates a junction temperature-dependent current limit threshold. This means that the current limit threshold is "reduced to around 4.2 A" as the junction temperature increases above 160 °C. When the temperature is above 175 °C, over-temperature shutdown (latch-OFF) will occur. This combination of features allows the device to continue operating for short periods of time (<30 seconds) with unexpected loads, while still retaining adequate protection for both the device and the load.

FUNCTIONAL PIN DESCRIPTION

POWER GROUND AND ANALOG GROUND (PGND AND AGND)

The power and analog ground pins should be connected together with a very low-impedance connection.

POSITIVE POWER SUPPLY (VPWR)

VPWR pins are the power supply inputs to the device. All VPWR pins must be connected together on the printed circuit board with as short as possible traces, offering as low an impedance as possible between pins.

STATUS FLAG (SF)

This pin is the device fault status output. This output is an active LOW open drain structure requiring a pull-up resistor $\underline{\text{to V}_{DD}}$. The maximum V_{DD} is <7.0 V. Refer to $\underline{\text{Table 5}}$, for the $\underline{\text{SF}}$ Output status definition.

INPUT 1,2 AND DISABLE INPUT 1 (IN1, IN2, AND D1)

These pins are input control pins used to control the outputs. These pins are 3.0 V/5.0 V CMOS-compatible inputs with hysteresis. IN1 and IN2 independently control OUT1 and OUT2, respectively. D1 input is used to tri-state disable the H-bridge outputs.

When D1 is SET (D1 = logic HIGH) in the disable state, outputs OUT1 and OUT2 are both tri-state disabled; however, the rest of the device circuitry is fully operational and the

supply I_{PWR(STANDBY)} current is reduced to a few mA. Refer to Table 3, Static Electrical Characteristics, page 6.

H-BRIDGE OUTPUT (OUT1, OUT2)

These pins are the outputs of the H-Bridge with integrated free-wheeling diodes. The bridge output is controlled using the IN1, IN2, D1, and $EN/\overline{D2}$ inputs. The outputs have PWM current limiting above the I_{LIM} threshold. The outputs also have thermal shutdown (tri-state latch-OFF) with hysteresis as well as short-circuit latch-OFF protection.

A disable timer (time $t_{\rm B}$) is incorporated to distinguish between load currents that are higher than the $l_{\rm LIM}$ threshold and short-circuit currents. This timer is activated at each output transition.

CHARGE PUMP CAPACITOR (CCP)

This pin is the charge pump output pin and connection for the external charge pump reservoir capacitor. The allowable value is from 30 nF to 100 nF. This capacitor must be connected from the CCP pin to the VPWR pin. The device cannot operate properly without the external reservoir capacitor.

ENABLE INPUT/DISABLE INPUT 2 (EN/D2)

The EN/ $\overline{D2}$ pin performs the same function as D1 pin, when it goes to a logic LOW the outputs are immediately tristated. It is also used to place the device in a Sleep mode so as to consume very low currents. When the EN/ $\overline{D2}$ pin voltage is a logic LOW state, the device is in the Sleep mode.

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The device is enabled and fully operational when the EN pin voltage is logic HIGH. An internal pull-down resistor maintains the device in Sleep mode in the event EN is driven through a high-impedance I/O or an unpowered microcontroller, or the $EN/\overline{D2}$ input becomes disconnected.

FEEDBACK (FB)

The 33931 has a feedback output (FB) for "real time" monitoring of H-Bridge high side output currents to facilitate closed-loop operation for motor speed and torque control.

The FB pin provides current sensing feedback of the H-Bridge high side drivers. When running in the forward or reverse direction, a ground-referenced 0.24% of load current

is output to this pin. Through the use of an external resistor to ground, the proportional feedback current can be converted to a proportional voltage equivalent and the controlling microcontroller can "read" the current proportional voltage with its analog-to-digital converter (ADC). This is intended to provide the user with only first-order motor current feedback for motor torque control. The resistance range for the linear operation of the FB pin is $100~\Omega < R_{FB} < 300~\Omega$.

If PWM-ing is implemented using the disable pin input (only D1), a small filter capacitor (\sim 1.0 μ F) may be required in parallel with the R_{FB} resistor to ground for spike suppression.

VOLTAGE CURRENT SENSE 33931 **REGULATION CHARGE TEMPERATURE** H-BRIDGE **SENSE PUMP ANALOG CONTROL AND PROTECTION OUTPUT DRIVERS** OUT1 - OUT2 INPUT LOGIC CONTROL MCU **FAULT LOGIC INTERFACE** PROTECTION LOGIC CONTROL **GATE CONTROL LOGIC**

FUNCTIONAL INTERNAL BLOCK DESCRIPTION

Figure 10. Functional Internal Block Diagram

ANALOG CONTROL AND PROTECTION CIRCUITRY:

An on-chip voltage regulator supplies the internal logic. The charge pump provides gate drive for the H-Bridge MOSFETs. The current and temperature sense circuitry provides detection and protection for the output drivers. Output under-voltage protection shuts down the MOSFETS.

GATE CONTROL LOGIC:

The 33931 is a monolithic H-Bridge Power IC designed primarily for any low-voltage DC servo motor control application within the current and voltage limits stated for the device. Two independent inputs provide polarity control of

two half-bridge totem-pole outputs. Two independent disable inputs are provided to force the H-Bridge outputs to tri-state (high-impedance off-state).

H-BRIDGE OUTPUT DRIVERS: OUT1 AND OUT2

The H-bridge is the power output stage. The current flow from OUT1 to OUT2 is reversible and under full control of the user by way of the Input Control Logic. The output stage is designed to produce full load control under all system conditions. All protective and control features are integrated into the control and protection blocks. The sensors for current and temperature are integrated directly into the output MOSFET for maximum accuracy and dependability.



FUNCTIONAL DEVICE OPERATION

OPERATIONAL MODES

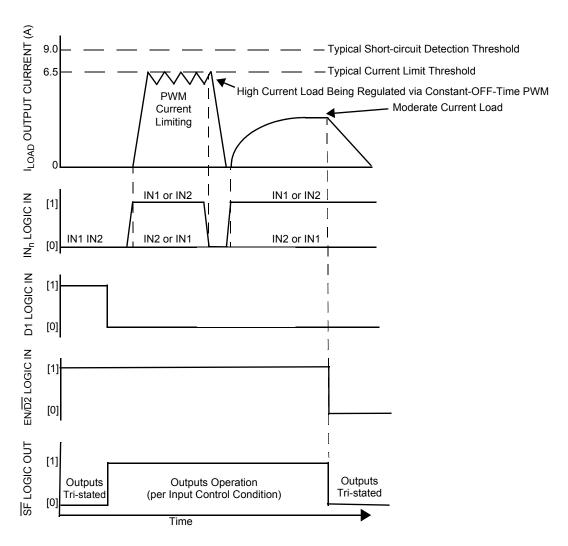


Figure 11. Operating States



LOGIC COMMANDS

Table 5. Truth Table

The tri-state conditions and the status flag are reset using D1 or $EN/\overline{D2}$. The truth table uses the following notations: L = LOW, H = HIGH, X = HIGH or LOW, and Z = High-impedance. All output power transistors are switched off.

Davies State		Input Conditions				Outputs	
Device State	EN/D2	D1	IN1	IN2	SF	OUT1	OUT2
Forward	Н	L	Н	L	Н	Н	L
Reverse	Н	L	L	Н	Н	L	Н
Freewheeling Low	Н	L	L	L	Н	L	L
Freewheeling High	Н	L	Н	Н	Н	Н	Н
Disable 1 (D1)	Н	Н	Х	Х	L	Z	Z
IN1 Disconnected	Н	L	Z	Х	Н	Н	Х
IN2 Disconnected	Н	L	Х	Z	Н	Х	Н
D1 Disconnected	Н	Z	Х	Х	L	Z	Z
Under-voltage Lockout ⁽³⁰⁾	Н	Х	Х	Х	L	Z	Z
Over-temperature ⁽³¹⁾	Н	Х	Х	Х	L	Z	Z
Short-circuit ⁽³¹⁾	Н	Х	Х	Х	L	Z	Z
Sleep Mode EN/D2	L	х	Х	Х	Н	Z	Z
EN/D2 Disconnected	Z	Х	Х	Х	Н	Z	Z

Notes

- 30. In the event of an under-voltage condition, the outputs tri-state and status flag is SET logic LOW. Upon under-voltage recovery, status flag is reset automatically or automatically cleared and the outputs are restored to their original operating condition.
- 31. When a short-circuit or over-temperature condition is detected, the power outputs are tri-state latched-OFF independent of the input signals and the status flag is latched to logic LOW. To reset from this condition requires the toggling of either D1, EN/D2, or V_{PWR}.

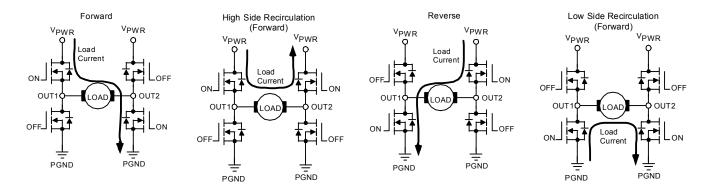


Figure 12. 33931 Power Stage Operation

PROTECTION AND DIAGNOSTIC FEATURES

SHORT-CIRCUIT PROTECTION

If an output short-circuit condition is detected, the power outputs tri-state (latch-OFF) independent of the input (IN1 and IN2) states, and the fault status output flag (\overline{SF}) is SET to logic LOW. If the D1 input changes from logic HIGH to logic

LOW, or if the EN/D2 input changes from logic LOW to logic HIGH, the output bridge will become operational again and the fault status flag will be reset (cleared) to a logic HIGH state.



The output stage will always switch into the mode defined by the input pins (IN1, IN2, D1, and $EN/\overline{D2}$), provided the device junction temperature is within the specified operating temperature range.

INTERNAL PWM CURRENT LIMITING

The maximum current flow under normal operating conditions should be less than 5.0 A. The instantaneous load currents will be limited to I_{LIM} via the internal PWM current limiting circuitry. When the I_{LIM} threshold current value is reached, the output stages are tri-stated for a fixed time (T_A) of 20 μs typical. Depending on the time constant associated with the load characteristics, the output current decreases during the tri-state duration until the next output ON cycle occurs.

The PWM current limit threshold value is dependent on the device junction temperature. When -40 °C < $T_{\rm J}$ <160 °C, $I_{\rm LIM}$ is between the specified minimum/maximum values. When $T_{\rm J}$ exceeds 160 °C, the $I_{\rm LIM}$ threshold decreases to 4.2 A. Shortly above 175 °C the device over-temperature circuit will detect $t_{\rm LIM}$ and an over-temperature shutdown will occur. This feature implements a graceful degradation of operation before thermal shutdown occurs, thus allowing for intermittent unexpected mechanical loads on the motor's gear-reduction train to be handled.

Important Die temperature excursions above 150 °C are permitted only for non-repetitive durations <30 seconds. Provision must be made at the system level to prevent prolonged operation in the current-foldback region.

OVER-TEMPERATURE SHUTDOWN AND HYSTERESIS

If an over-temperature condition occurs, the power <u>outputs</u> are tri-stated (latched-OFF) and the fault status flag (\overline{SF}) is SET to logic LOW.

To reset from this condition, D1 must change from logic HIGH to logic LOW, or EN/D2 must change from logic LOW to logic HIGH. When reset, the output stage switches ON again, provided that the junction temperature is now below the over-temperature threshold limit minus the hysteresis.

Important Resetting from the fault condition will clear the fault status flag. Powering down and powering up the device will also reset the 33931 from the fault condition.

OUTPUT AVALANCHE PROTECTION

If VPWR were to become an open circuit, the outputs would likely tri-state simultaneously due to the disable logic. This could result in an unclamped inductive discharge. The VPWR input to the 33931 should not exceed 40 V during this transient condition, to prevent electrical overstress of the output drivers. This can be accomplished with a zener clamp or MOV, and/or an appropriately valued input capacitor with sufficiently low ESR (see Figure 13).

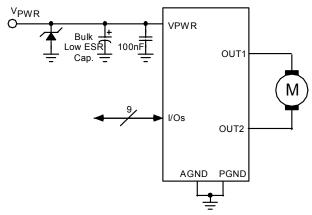


Figure 13. Avalanche Protection



TYPICAL APPLICATIONS

INTRODUCTION

A typical application schematic is shown in Figure 14. For precision high-current applications in harsh, noisy

environments, the V_{PWR} by-pass capacitor may need to be substantially larger.

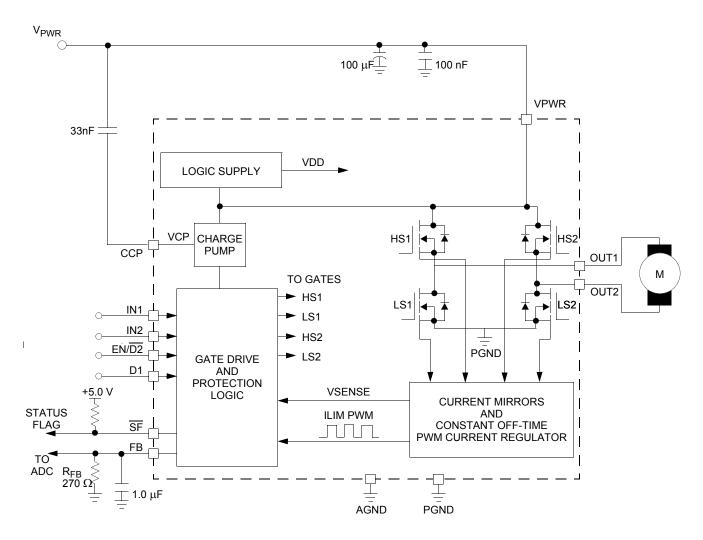


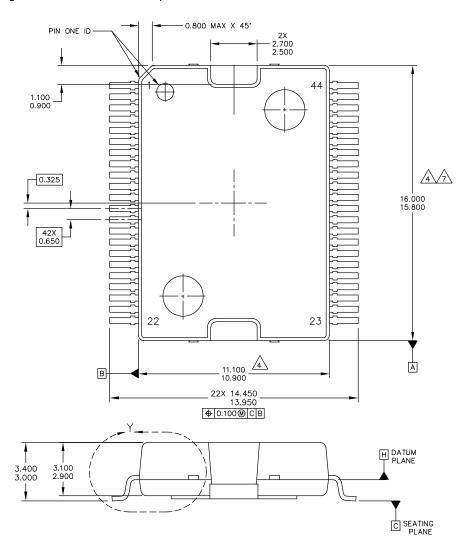
Figure 14. 33931 Typical Application Schematic



PACKAGING

PACKAGE DIMENSIONS

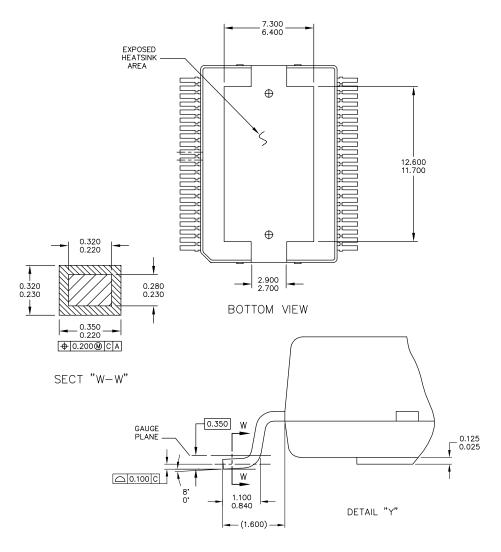
For the most current package revision, visit <u>www.freescale.com</u> and perform a keyword search using the 98Axxxxxxxxx listed on the following pages. Dimensions shown are provided for reference ONLY.



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TITLE: 44 LEAD HSOP		DOCUMENT NO): 98ARH98330A	REV: B
W/PROTRUDING HEAT	CASE NUMBER: 1291-02 16 MAR			
W/TROTHODING TIE/C	OII VI V	STANDARD: NON-JEDEC		

VW SUFFIX 44-PIN 98ARH98330A REVISION B

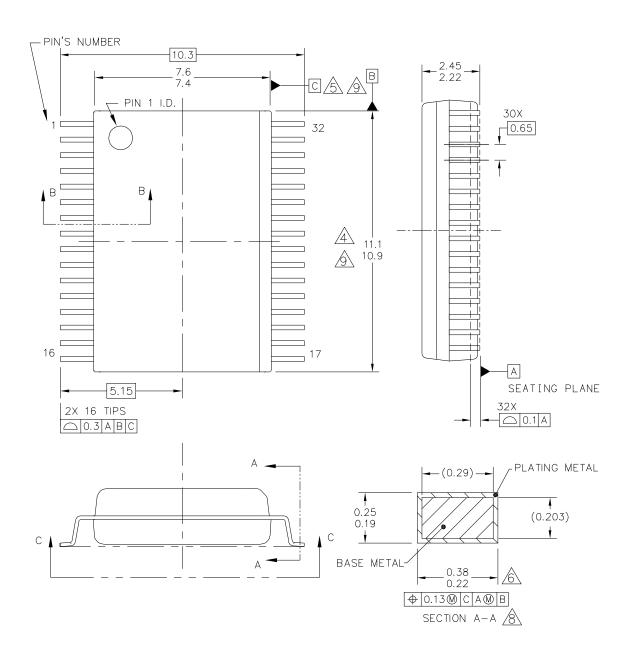




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VW SUFFIX 44-PIN 98ARH98330A REVISION B

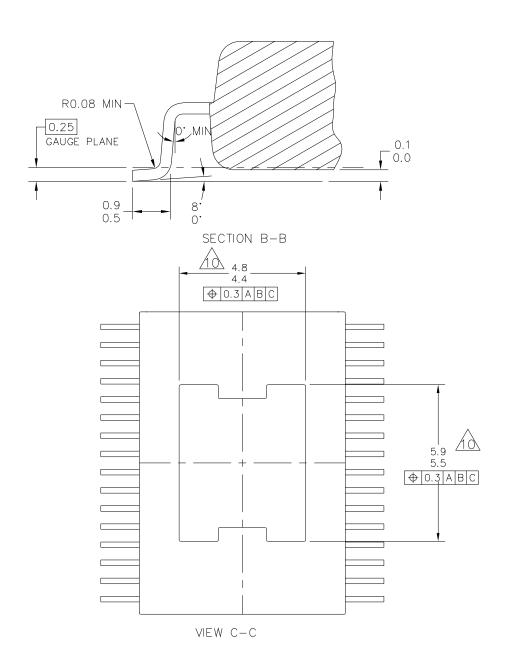




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TITLE: 32LD SOIC W/B, 0.65	PITCH	DOCUMENT NO): 98ARL10543D	REV: D
5.7 X 4.6 EXPOSED		CASE NUMBER	2: 1437-04	23 JUN 2008
CASE OUTLINE		STANDARD: NO	N-JEDEC	

EK SUFFIX 32-PIN 98ARL10543D REVISION D





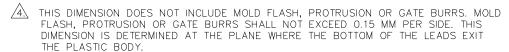
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5.7 X 4.6 EXPOSED PAD		CASE NUMBER	R: 1437–04	23 JUN 2008
CASE OUTLINI	E	STANDARD: NO	N-JEDEC	

EK SUFFIX 32-PIN 98ARL10543D REVISION D



NOTES:

- 1. DIMENSIONS ARE IN MILLIMETERS.
- 2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
- 3. DATUMS B AND C TO BE DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.



THIS DIMENSION DOES NOT INCLUDE INTER-LEAD FLASH OR PROTRUSIONS. INTER-LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED 0.25 MM PER SIDE. THIS DIMENSION IS DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.

THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.4 mm. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND ADJACENT LEAD SHALL NOT LESS THAN 0.07 mm.

 \triangle Exact shape of each corner is optional.

A. THESE DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN 0.10 mm AND 0.3 mm FROM THE LEAD TIP.

THE PACKAGE TOP MAY BE SMALLER THAN THE PACKAGE BOTTOM. THIS DIMENSION IS DETERMINED AT THE OUTERMOST EXTREMES OF THE PLASTIC BODY EXCLUSIVE OF MOLD FLASH, TIE BAR BURRS, GATE BURRS AND INTER-LEAD FLASH, BUT INCLUDING ANY MISMATCH BETWEEN THE TOP AND BOTTOM OF THE PLASTIC BODY.

THESE DIMENSIONS RANGES DEFINE THE PRIMARY KEEP-OUT AREA. MOLD LOCKING AND RESIN BLEED CONTROL FEATURES MAY BE VISIBLE AND THEY MAY EXTEND TO 0.9mm FROM MAXIMUM EXPOSED PAD SIZE

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TITLE: 32LD SOIC W/B, 0.6	65 PITCH	DOCUMENT NO): 98ARL10543D	REV: D
5.7 X 4.6 EXPOSED PAD CASE OUTLINE		CASE NUMBER	2: 1437-04	23 JUN 2008
		STANDARD: NO	N-JEDEC	

EK SUFFIX 32-PIN 98ARL10543D REVISION D



ADDITIONAL DOCUMENTATION

THERMAL ADDENDUM

Introduction

This thermal addendum is provided as a supplement to the MC33931 technical datasheet. The addendum provides thermal performance information that may be critical in the design and development of system applications. All electrical, application, and packaging information is provided in the datasheet.

Package and Thermal Considerations

The MC33931 is offered in a 32-pin SOICW-EP and a 44-pin HSOP single die package. There is a single heat source (P), a single junction temperature (T_J), and thermal resistance ($R_{\theta JA}$). This thermal addendum is specific to the 32-pin SOICW-EP package.

$$\{T_J\} = [R_{\theta JA}] \cdot \{P\}$$

The stated values are solely for a thermal performance comparison of one package to another in a standardized environment. This methodology is not meant to, and will not predict the performance of a package in an application-specific environment. Stated values were obtained by measurement and simulation according to the standards listed below.

Table 6. Table of Thermal Resistance Data

Rating			Value	Unit	Notes
Junction to Ambient Natural Convection	Single Layer board (1s)	$R_{ hetaJA}$	92	°C/W	(32),(33)
Junction to Ambient Natural Convection	Four layer board (2s2p)	$R_{ hetaJA}$	26.6	°C/W	(32),(34)
Junction to Board		$R_{ heta JB}$	7.0	°C/W	(35)
Junction to Case (bottom / flag)		R _{θJC} (bottom)	0.62	°C/W	(38)
Junction to Case (top)		R _{θJC} (top)	23.3	°C/W	(36)
Junction to Package Top	Natural Convection	Ψ_{JT}	2.7	°C/W	(37)

Notes

- 32. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
- 33. Per JEDEC JESD51-2 with the single layer board (JESD51-3) horizontal.
- 34. Per JEDEC JESD51-6 with the board (JESD51-7) horizontal.
- 35. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
- 36. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
- 37. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written as Psi-JT.
- 38. Thermal resistance between the die and the case bottom / flag surface (simulated) (flag bottom side fixed to ambient temperature).



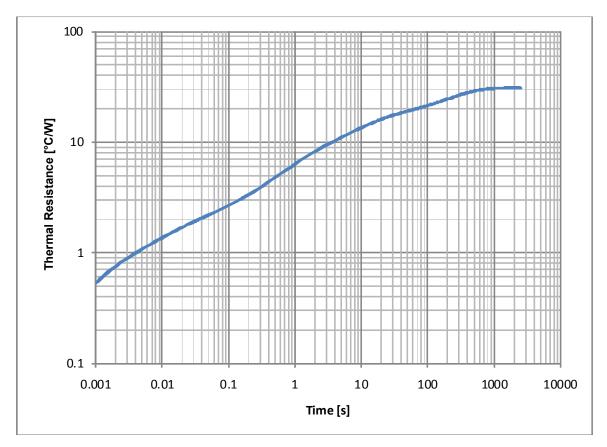


Figure 15. Transient Thermal Resistance $R_{\theta J A}$ MC33931EK on 2s2p Test Board



REFERENCE SECTION

Table 7. Thermal Analysis Reference Documents

Reference	Description	
AN4146	Thermal Modeling and Simulation of 12 V Gen3 eXtreme Switch Devices with SPICE	
BASICTHERMALWP	Basic Principles of Thermal Analysis for Semiconductor Systems	



REVISION HISTORY

REVISION	DATE	DESCRIPTION
1.0	2/2008	Initial Release
2.0	12/2008	Updated Freescale for and style
		Removed PC33931VW/R2 from the ordering information and added MC33931VW/R2
		Changes Max R _{DS(ON)} from 225 to 235 mOhm in the document
		Changed Peak Package Reflow Temperature During Reflow ⁽⁷⁾ , (8)
		Changed Approximate Junction-to-Case Thermal Resistance (9)
		In SHORT-CIRCUIT PROTECTION, changed D2 to EN/D2
3.0	6/2012	Added PC33931EK to the Ordering Information Table
		Added EK ordering and package information
		Added Thermal Addendum and Reference Document sections
		Minor corrections throughout the document.
4.0	10/2012	PC33931EK changed to MC33931EK and released to production
		Document level changed from Advance Information to Technical Data
		Changed SOIC to SOICW-EP