



Chipsmall Limited consists of a professional team with an average of over 10 year of expertise in the distribution of electronic components. Based in Hongkong, we have already established firm and mutual-benefit business relationships with customers from,Europe,America and south Asia,supplying obsolete and hard-to-find components to meet their specific needs.

With the principle of “Quality Parts,Customers Priority,Honest Operation,and Considerate Service”,our business mainly focus on the distribution of electronic components. Line cards we deal with include Microchip,ALPS,ROHM,Xilinx,Pulse,ON,Everlight and Freescale. Main products comprise IC,Modules,Potentiometer,IC Socket,Relay,Connector.Our parts cover such applications as commercial,industrial, and automotives areas.

We are looking forward to setting up business relationship with you and hope to provide you with the best service and solution. Let us make a better world for our industry!



Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

Email & Skype: info@chipsmall.com Web: www.chipsmall.com

Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China



MC33932

5.0 A throttle control H-bridge

Rev. 6.0 — 10 September 2018

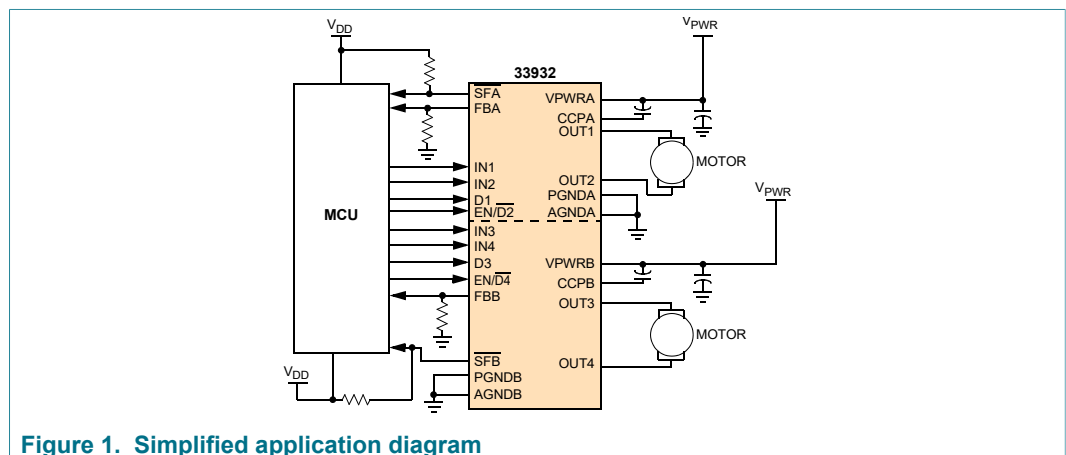
Data sheet: technical data

1 General description

The 33932 is a monolithic H-bridge power IC in a robust thermally enhanced package. The 33932 has two independent monolithic H-bridge power ICs in the same package. They are designed primarily for automotive electronic throttle control, but are applicable to any low voltage DC servo motor control application within the current and voltage limits stated in this specification. It meets the stringent requirements of automotive applications and is fully AEC-Q100 grade 1 qualified.

Each H-bridge in the 33932 is able to control inductive loads with currents up to 5.0 A peak. RMS current capability is subject to the degree of heat sinking provided to the device package. Internal peak-current limiting (regulation) is activated at load currents above $6.5 \text{ A} \pm 1.5 \text{ A}$. Output loads can be pulse-width modulated at frequencies up to 11 kHz. A load current feedback feature provides a proportional (0.24 % of the load current) current output suitable for monitoring by a microcontroller's A/D input. A status flag output reports undervoltage, overcurrent and overtemperature fault conditions.

Two independent inputs provide polarity control of two half-bridge totem pole outputs. Two independent disable inputs are provided to force the H-bridge outputs to 3-state (high-impedance OFF state).



2 Features and benefits

- 8.0 to 28 V continuous operation (transient operation from 5.0 to 40 V)
- 235 mΩ maximum $R_{DS(on)}$ @ $T_J = 150 \text{ }^\circ\text{C}$ (each H-bridge MOSFET)
- 3.0 V and 5.0 V TTL / CMOS logic compatible inputs
- Output short-circuit protection (short to VPWR or GND)
- Overcurrent limiting (regulation) via internal constant-off-time PWM
- Temperature dependent current limit threshold reduction
- All inputs have an internal source/sink to define the default (floating input) state



- Sleep mode with current draw < 50 μ A (each half with inputs floating or set to match default logic states)
- AEC-Q100 grade 1 qualified

3 Applications

- Electronic throttle control (ETC)
- Exhaust gas recirculation (EGR)
- Turbo flap control
- Industrial and medical pumps
- Stepper motor control
- Dual motor drive

4 Ordering information

Table 1. Ordering information

Type number ^[1]	Package			
	Name	Description	Operating temperature	Version
MC33932VW	HSOP44	HSOP44, plastic, thermal enhanced small outline package; 44 terminals; 0.65 mm pitch; 15.9 mm x 11 mm x 3 mm body	T _A = -40 °C to 125 °C	SOT1305-2
MC33932EK	HSOP54	HSOP54, plastic, heat sink small outline package; 54 terminals; 0.65 mm pitch; 17.9 mm x 7.5 mm x 2.65 mm body		SOT1747-4

[1] To order parts in tape and reel, add the R2 suffix to the part number.

5 Block diagram

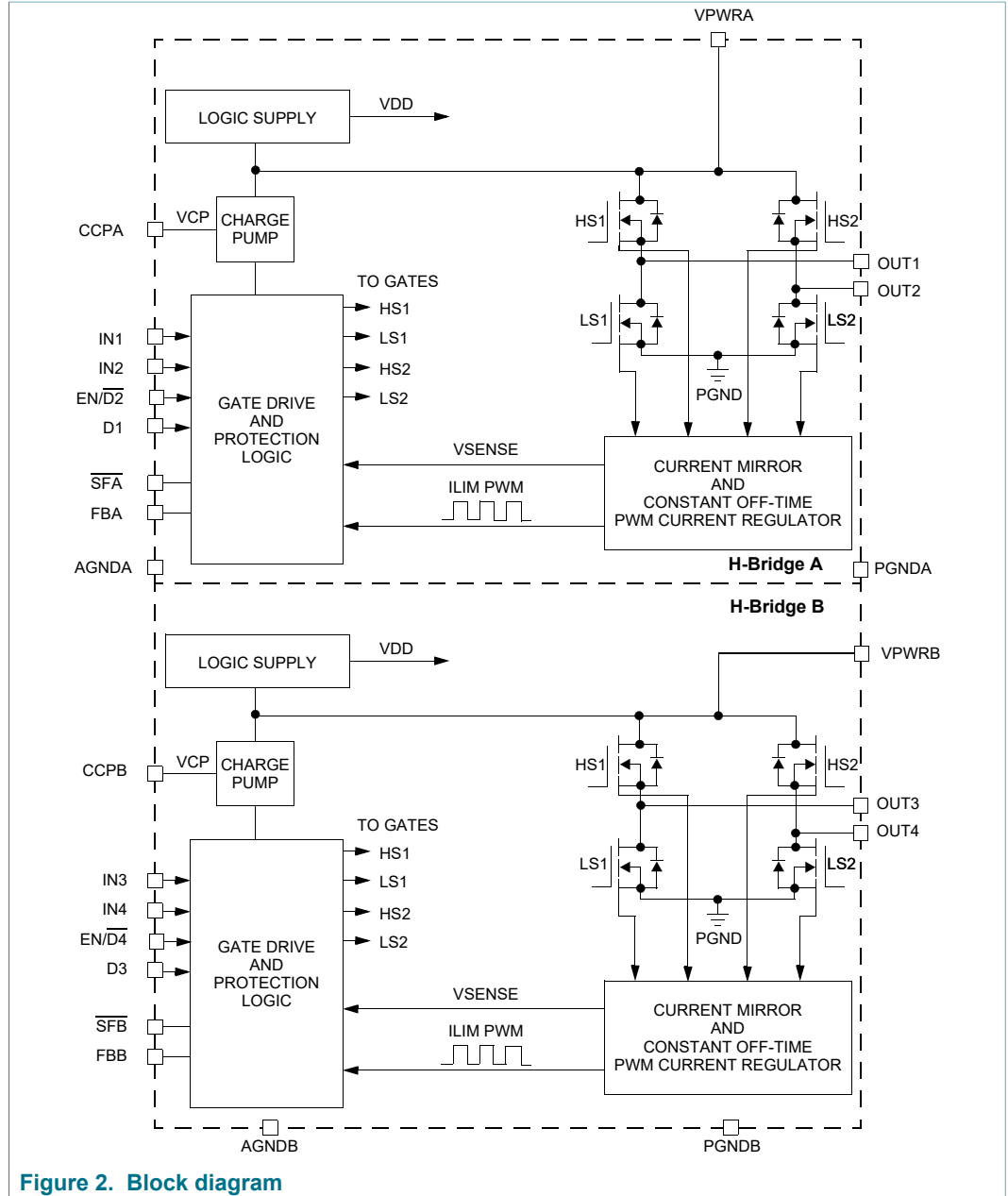


Figure 2. Block diagram

6 Pinning information

6.1 Pinning

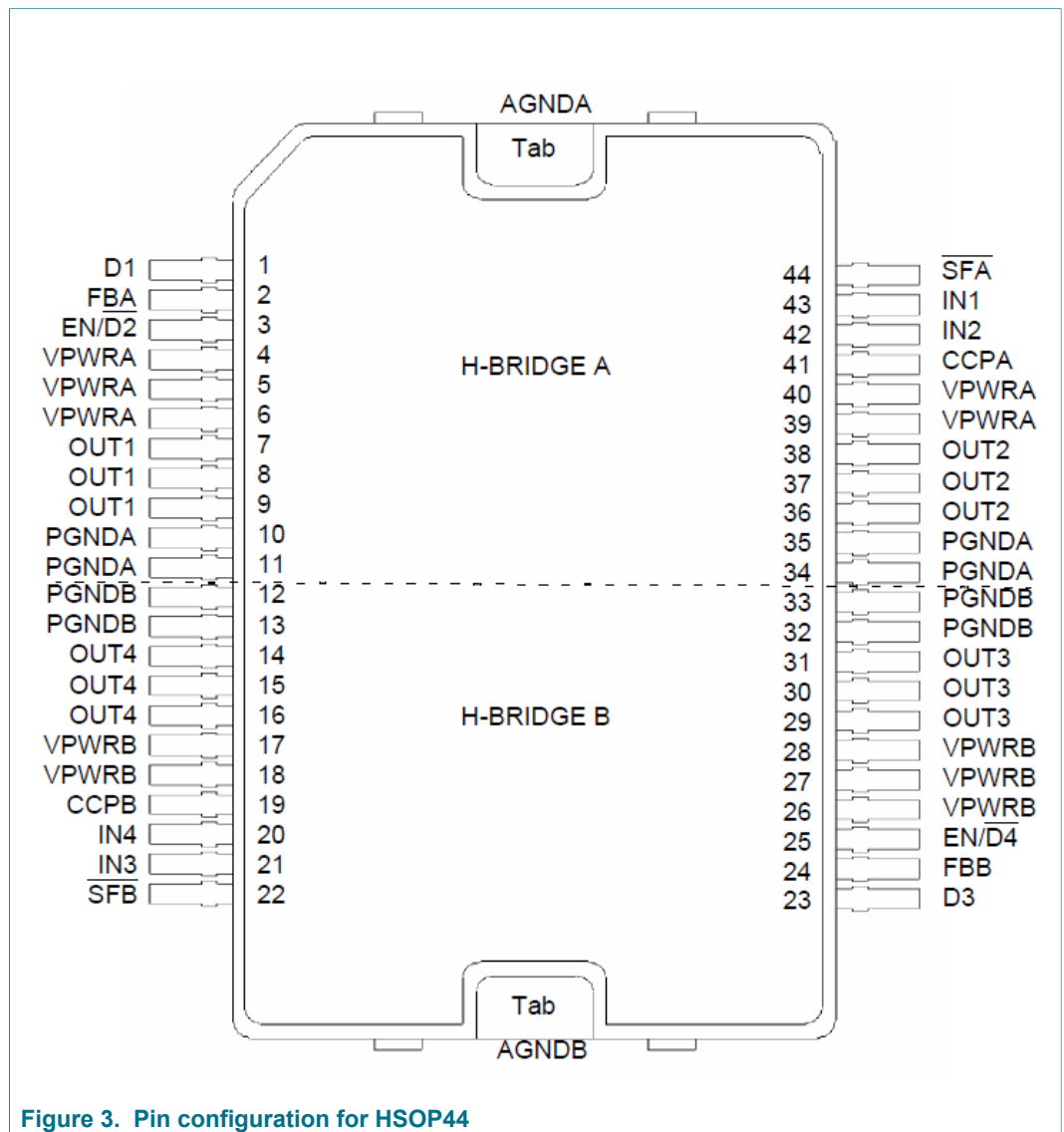


Figure 3. Pin configuration for HSOP44

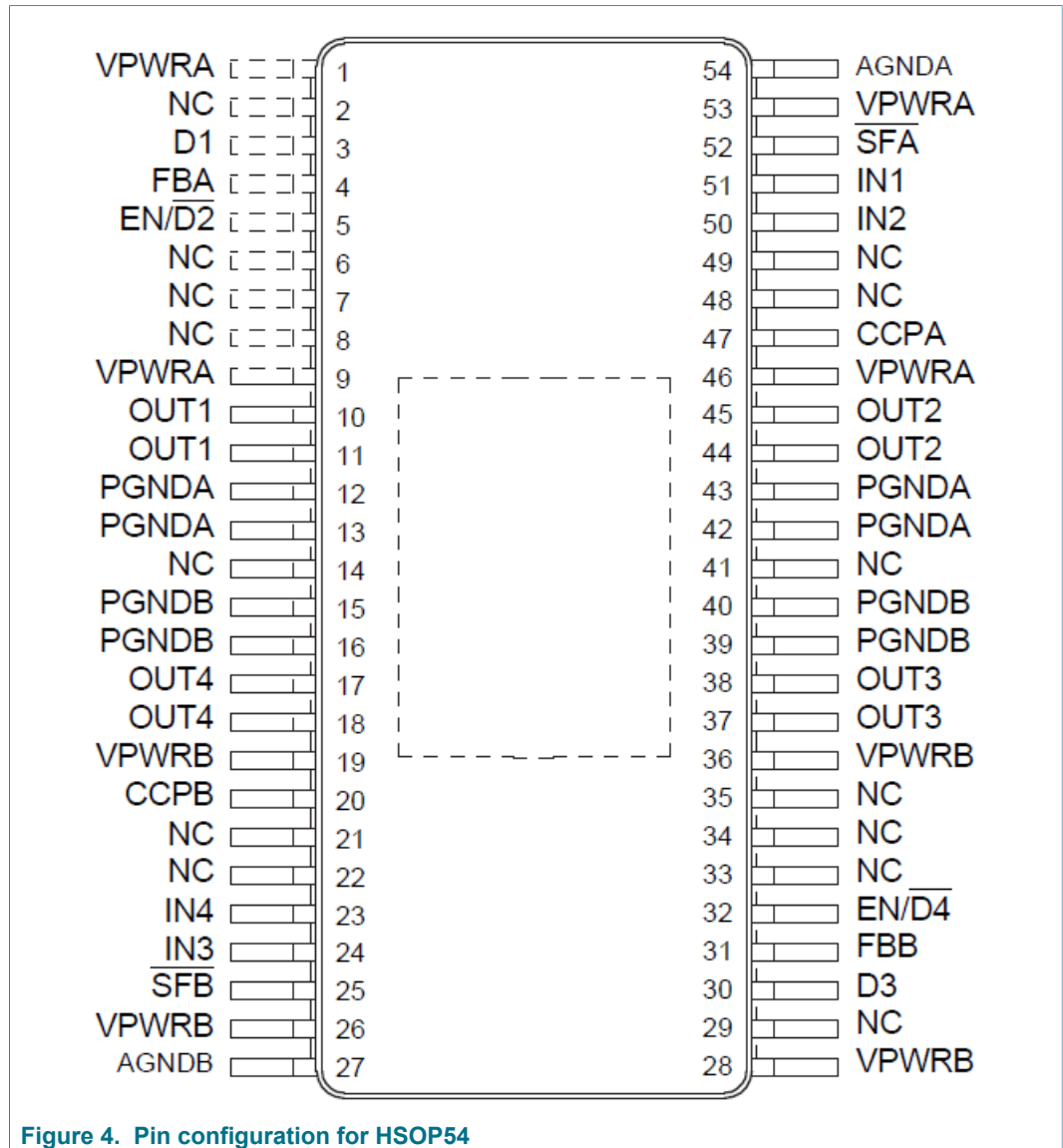


Figure 4. Pin configuration for HSOP54

6.2 Pin description

For functional description of each pin, see [Section 7.2 "Functional pin description"](#).

Table 2. Pin description

Symbol	Pin HSOP (VW)	Pin SOICW-EP (EK)	Function	Name	Description
D1	1	3	Logic input	Disable input 1 (active high)	When D1 is logic high, both OUT1 and OUT2 are 3-stated. Schmitt trigger input with ~80 µA source so default condition = disabled.
FBA	2	4	Analog output	Feedback	H-bridge A load current feedback output provides ground referenced 0.24 % of the high-side output current (tie to GND through a resistor if not used)

Symbol	Pin HSOP (VW)	Pin SOICW-EP (EK)	Function	Name	Description
EN/D $\bar{2}$	3	5	Logic input	Enable input	When EN/D $\bar{2}$ is logic high, the H-bridge A is operational. When EN/D $\bar{2}$ is logic low, the H-bridge A outputs are 3-stated and placed in Sleep mode (logic input with ~80 μ A sink so default condition = Sleep mode).
VPWRA	4, 5, 6, 39, 40	1, 9, 46, 53	Power input	Positive power supply	These pins must be connected together physically as close as possible and directly soldered down to a wide, thick, low resistance supply plane on the PCB.
OUT1	7, 8, 9	10, 11	Power output	H-bridge output 1	H-bridge A source of HS1 and drain LS1
PGNDA	10, 11, 34, 35	12, 13, 42, 43	Power ground	Power ground	High-current power ground pins must be connected together physically as close as possible and directly soldered down to a wide, thick, low resistance ground plane on the PCB. PGNDA should be connected to PGNDB with a low-impedance path.
PGNDB	12, 13, 32, 33	15, 16, 39, 40	Power ground	Power ground	High-current power ground pins must be connected together physically as close as possible and directly soldered down to a wide, thick, low resistance ground plane on the PCB. PGNDB should be connected to PGNDA with a low-impedance path.
OUT4	14, 15, 16	17, 18	Power output	H-bridge output 4	H-bridge B source of HS2 and drain of LS2
VPWRB	17, 18, 26, 27, 28	19, 26, 28, 36	Power input	Positive power supply	These pins must be connected together physically as close as possible and directly soldered down to a wide, thick, low resistance supply plane on the PCB.
CCPB	19	20	Analog output	Charge pump capacitor	External reservoir capacitor connection for H-bridge B internal charge pump; connected to VPWRB. Allowable values are 30 nF to 100 nF ^[1]
IN4	20	23	Logic input	Input 4	Logic input control of OUT4
IN3	21	24	Logic input	Input 3	Logic input control of OUT3
SFB	22	25	Logic output - open drain	Status flag B (active low)	H-bridge B open drain active low Status flag output (requires an external pull-up resistor to V _{DD} . Maximum permissible load current < 0.5 mA. Maximum V _{SFLOW} < 0.4 V at 0.3 mA. Maximum permissible pull-up voltage < 7.0 V.
D3	23	30	Logic input	Disable input 3 (active high)	When D3 is logic high, both OUT3 and OUT4 are 3-stated. Schmitt trigger input with ~80 μ A source so default condition = disabled.

Symbol	Pin HSOP (VW)	Pin SOICW-EP (EK)	Function	Name	Description
FBB	24	31	Analog output	Feedback B	H-bridge B load current feedback output provides ground referenced 0.24 % of the high-side output current (tie to GND through a resistor if not used)
EN/ $\overline{D4}$	25	32	Logic input	Enable input	When EN/ $\overline{D4}$ is logic high, H-bridge B is operational. When EN/ $\overline{D4}$ is logic low, the H-bridge B outputs are 3-stated and H-bridge B is placed in Sleep mode (logic input with ~ 80 μ A sink so default condition = Sleep mode).
OUT3	29, 30, 31	37, 38	Power output	H-bridge output 3	H-bridge B source of HS1 and drain of LS1
OUT2	36, 37, 38	44, 45	Power output	H-bridge output 2	H-bridge A source of HS2 and drain of LS2
CCPA	41	47	Analog output	Charge pump capacitor	External reservoir capacitor connection for H-bridge B internal charge pump; connected to VPWRB. Allowable values are 30 nF to 100 nF. ^[1]
IN2	42	50	Logic input	Input 2	Logic input control of OUT2
IN1	43	51	Logic input	Input 1	Logic input control of OUT1; e.g., when IN1 is logic high, OUT1 is set to VPWRA, and when IN1 is logic low, OUT1 is set to PGNDA. Schmitt trigger input with ~ 80 μ A source so default condition = OUT1 high.
\overline{SFA}	44	52	Logic output - open drain	Status flag (active low)	H-bridge A open drain active low status flag output requires an external pull-up resistor to V_{DD} . Maximum permissible load current < 0.5 mA. Maximum V_{SFLOW} < 0.4 V at 0.3 mA. Maximum permissible pull-up voltage < 7.0 V.
AGNDA AGNDB	TAB	54 27	Analog ground	Analog signal ground	The low-current analog signal ground must be connected to PGND via low-impedance path (<10 m Ω , 0 Hz to 20 kHz)
n.c.	n.a.	2, 6, 7, 8, 14, 21, 22, 29, 33, 31, 32, 33, 34, 35, 41, 48, 49	None	not connected	These pins have no electrical connection or function
EP	n.a.	EP	Thermal pad	Exposed pad	Exposed TAB is also the main heat sinking path for the device and must be connected to ground

[1] This capacitor is required for proper performance of the device.

7 Functional description

7.1 Introduction

The 33932 has two identical H-bridge drivers in the same package. The only connection that is shared internally is the analog ground (AGND). This description is given for the H-bridge A half of the total device. However, the H-bridge B half exhibits identical behavior.

Numerous protection and operational features (speed, torque, direction, dynamic braking, PWM control, and closed-loop control) make the 33932 a very attractive, cost-effective solution for controlling a broad range of small DC motors. The 33932 outputs are capable of supporting peak DC load currents of up to 5.0 A from a 28 V V_{PWR} source. An internal charge pump and gate drive circuitry are provided that can support external PWM frequencies up to 11 kHz.

The 33932 has an analog feedback (current mirror) output pin (the FB pin) that provides a constant-current source ratioed to the active high-side MOSFETs' current. This can be used to provide real time monitoring of output current to facilitate closed-loop operation for motor speed/torque control, or for the detection of openload conditions.

Two independent inputs, IN1 and IN2, provide control of the two totem-pole half-bridge outputs. Two independent disable inputs, D1 and EN/D2, provide the means to force the H-bridge outputs to a high-impedance state (all H-bridges switch OFF). The EN/D2 pin also controls an enable function that allows the IC to be placed in a power conserving Sleep mode.

The 33932 has output current limiting (via constant OFF time PWM current regulation), output short-circuit detection with latch-off, and overtemperature detection with latch-off. Once the device is latched-off due to a fault condition, either of the disable inputs (D1 or EN/D2), or V_{PWR} must be toggled to clear the status flag.

Current limiting (load current regulation) is accomplished by a constant OFF time PWM method using current limit threshold triggering. The current limiting scheme is unique in that it incorporates a junction temperature dependent current limit threshold. This means that the current limit threshold is reduced to around 4.2 A as the junction temperature increases above 160 °C. When the temperature is above 175 °C, overtemperature shutdown (latch-off) occurs. This combination of features allows the device to continue operating for short periods of time (< 30 seconds) with unexpected loads, while still retaining adequate protection for both the device and the load.

7.2 Functional pin description

7.2.1 Power ground and analog ground (PGND and AGND)

The power and analog ground pins should be connected together with a very low-impedance connection.

7.2.2 Positive power supply (VPWR)

VPWR pins are the power supply inputs to the device. All VPWR pins must be connected together on the printed circuit board with traces as short as possible, offering as low-impedance as possible between pins.

7.2.3 Status flag (\overline{SF})

This pin is the device fault status output. This output is an active low open drain structure requiring a pull-up resistor to V_{DD} . The maximum V_{DD} is < 7.0 V. See [Table 7](#) for the \overline{SF} output status definition.

7.2.4 Input 1, 2 and Disable input 1 (IN1, IN2, and D1)

These pins are input control pins used to control the outputs. These pins are 3.0 V/ 5.0 V CMOS-compatible inputs with hysteresis. IN1 and IN2 independently control OUT1 and OUT2, respectively. D1 input is used to 3-state disable the H-bridge outputs.

When D1 is set (D1 = logic high) in the disable state, outputs OUT1 and OUT2 are both 3-state disabled; however, the rest of the device circuitry is fully operational and the supply $I_{PWR(STANDBY)}$ current is reduced to a few mA. See [Table 5](#).

7.2.5 H-bridge output (OUT1, OUT2)

These pins are the outputs of the H-bridge with integrated freewheeling diodes. The bridge output is controlled using the IN1, IN2, D1, and $EN/\overline{D2}$ inputs. The outputs have PWM current limiting above the I_{LIM} threshold. The outputs also have thermal shutdown (3-state latch-off) with hysteresis as well as short-circuit latch-off protection.

A disable timer (time t_B) is incorporated to distinguish between load currents higher than the I_{LIM} threshold and short-circuit currents. This timer is activated at each output transition.

7.2.6 Charge pump capacitor (CCP)

This pin is the charge pump output pin and connection for the external charge pump reservoir capacitor. The allowable value is from 30 nF to 100 nF.

This capacitor must be connected from the CCP pin to the VPWR pin. The device cannot operate properly without the external reservoir capacitor.

7.2.7 Enable input/Disable input 2 ($EN/\overline{D2}$)

The $EN/\overline{D2}$ pin performs the same function as D1 pin, when it goes to a logic low, the outputs are immediately 3-stated. It is also used to place the device in a Sleep mode so as to consume very low currents. When the $EN/\overline{D2}$ pin voltage is a logic Low state, the device is in the Sleep mode.

The device is enabled and fully operational when the EN pin voltage is logic high. An internal pull-down resistor maintains the device in Sleep mode in the event EN is driven through a high-impedance I/O or an unpowered microcontroller, or the $EN/\overline{D2}$ input is disconnected.

7.2.8 Feedback (FB)

The 33932 has a feedback output (FB) for real time monitoring of H-bridge high-side output currents to facilitate closed-loop operation for motor speed and torque control.

The FB pin provides current sensing feedback of the H-bridge high-side drivers. When running in the forward or reverse direction, a ground-referenced 0.24 % of load current is output to this pin. Through the use of an external resistor to ground, the proportional feedback current can be converted to a proportional voltage equivalent and

the controlling microcontroller can read the current proportional voltage with its analog-to-digital converter (ADC). This is intended to provide the user with only first-order motor current feedback for motor torque control. The resistance range for the linear operation of the FB pin is $100 \Omega < R_{FB} < 300 \Omega$.

If PWM-ing is implemented using the disable pin input (only D1), a small filter capacitor (~1.0 μ F) may be required in parallel with the R_{FB} resistor to ground for spike suppression.

8 Maximum ratings

Table 3. Maximum ratings

All voltages are with respect to ground unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device. These parameters are not production tested.

Symbol	Parameter	Value	Unit
Electrical ratings			
$V_{PWR(SS)}$ $V_{PWR(T)}$	Power supply voltage • Normal operation (steady-state) • Transient overvoltage	^[1] -0.3 to 28 -0.3 to 40	V
V_{IN}	Logic input voltage	^[2] -0.3 to 7.0	V
V_{SF}	\overline{SFA} , \overline{SFB} output	^[3] -0.3 to 7.0	V
$I_{OUT(CONT)}$	Continuous output current	^[4] 5.0	A
V_{ESD1} V_{ESD2}	ESD voltage • Human body model • Machine model • Charge device model – Corner pins – All other pins	^[5] ± 2000 ± 200 ± 750 ± 500	V
T_{STG}	Storage temperature	-65 to 150	°C
T_A	Operational ambient temperature	^[6] -40 to 125	°C
T_J	Operation junction temperature	^[6] -40 to 150	°C
T_{PPRT}	Peak package reflow temperature during reflow	^[7] ^[8]	°C

- [1] Device will survive repetitive transient overvoltage conditions for durations not to exceed 500 ms at duty cycle not to exceed 10 %. External protection is required to prevent device damage in case of a reverse battery condition.
- [2] Exceeding the maximum input voltage on IN1, IN2, IN3, IN4, EN/D2, EN/D4, D1, or D3 may cause a malfunction or permanent damage to the device.
- [3] Exceeding the pull-up resistor voltage on the open drain \overline{SFA} or \overline{SFB} pin may cause permanent damage to the device.
- [4] Continuous output current capability is dependent on sufficient package heat sinking to keep junction temperature ≤ 150 °C.
- [5] ESD testing is performed in accordance with the Human Body Model ($C_{ZAP} = 100$ pF, $R_{ZAP} = 1500 \Omega$), Machine Model ($C_{ZAP} = 200$ pF, $R_{ZAP} = 0 \Omega$), and the Charge Device Model (CDM), Robotic ($C_{ZAP} = 4.0$ pF).
- [6] The limiting factor is junction temperature, taking into account the power dissipation, thermal resistance, and heat sinking provided. Brief non-repetitive excursions of junction temperature above 150 °C can be tolerated, provided the duration does not exceed 30 seconds maximum. Non-repetitive events are defined as not occurring more than once in 24 hours.
- [7] Pin soldering temperature limit is for 10 seconds maximum duration. Not designed for immersion soldering. Exceeding these limits may cause malfunction or permanent damage to the device.
- [8] NXP's Package Reflow capability meets Pb-free requirements for JEDEC standard J-STD-020C. For Peak Package Reflow Temperature and Moisture Sensitivity Levels (MSL), go to www.nxp.com, search by part number (remove prefixes/suffixes and enter the core ID) to view all orderable parts, and review parametrics.

9 Thermal characteristics

Table 4. Thermal characteristics

Symbol	Parameter	Value	Unit
$R_{\theta JC}$	Approximate junction-to-case thermal resistance	[1] < 1.0	°C/W

[1] Exposed heat sink pad plus the power and ground pins comprise the main heat conduction paths. The actual $R_{\theta JB}$ (junction-to-PC board) values will vary depending on solder thickness and composition and copper trace thickness and area. Maximum current at maximum die temperature represents ~16 W of conduction loss heating in the diagonal pair of output MOSFETs. Therefore, the $R_{\theta JA}$ must be < 5.0 °C/W for maximum current at 70 °C ambient. Module thermal design must be planned accordingly.

10 Static characteristics

Table 5. Static characteristics

Characteristics noted under conditions $5.0\text{ V} \leq V_{PWR} \leq 28\text{ V}$, $-40\text{ °C} \leq T_A \leq 125\text{ °C}$, $GND = 0\text{ V}$, unless otherwise noted. Typical values noted reflect the approximate parameter means at $T_A = 25\text{ °C}$ under nominal conditions, unless otherwise noted.

Symbol	Parameter	Min	Typ	Max	Unit
Power inputs (VPWR)					
$V_{PWR(SS)}$ $V_{PWR(t)}$	Operating voltage range • Steady-state • Transient ($t < 500\text{ ms}$)	[1] [2] 5.0 —	— —	28 40	V
$I_{PWR(SLEEP)}$	Sleep state supply current • $EN/\overline{D2} = \text{Logic } [0]$, $IN1, IN2, D1 = \text{Logic } [1]$, and $I_{OUT} = 0\text{ A}$	[3] —	—	50	µA
$I_{PWR(STANDBY)}$	Standby supply current (part enabled) • $I_{OUT} = 0\text{ A}$, $V_{EN} = 5.0\text{ V}$	—	—	20	mA
$V_{UVLO(ACTIVE)}$ $V_{UVLO(INACTIVE)}$ $V_{UVLO(HYS)}$	Undervoltage lockout thresholds • $V_{PWR(FALLING)}$ • $V_{PWR(RISING)}$ • Hysteresis	4.15 — 150	— — 200	— 5.0 350	V V mV
Charge pump					
$V_{CP} - V_{PWR}$	Charge pump voltage (CP capacitor = 33 nF), no PWM • $V_{PWR} = 5.0\text{ V}$ • $V_{PWR} = 28\text{ V}$	3.5 —	— —	— 12	V
$V_{CP} - V_{PWR}$	Charge pump voltage (CP capacitor = 33 nF), PWM = 11 kHz, • $V_{PWR} = 5.0\text{ V}$ • $V_{PWR} = 28\text{ V}$	3.5 —	— —	— 12	V
Control inputs					
V_I	Operating input voltage ($IN1, IN2, D1, EN/\overline{D2}, IN3, IN4, D3, EN/\overline{D4}$)	—	—	5.5	V
V_{IH} V_{IL} V_{HYS}	Input voltage ($IN1, IN2, D1, EN/\overline{D2}, IN3, IN4, D3, EN/\overline{D4}$) • Logic threshold high • Logic threshold low • Hysteresis	2.0 — 250	— — 400	— 1.0 —	V V mV
I_{IN}	Logic input currents, $V_{PWR} = 8.0\text{ V}$ • Inputs $EN/\overline{D2}, EN/\overline{D4}$ (internal pull-downs), $V_{IH} = 5.0\text{ V}$ • Inputs $IN1, IN2, D1, IN3, IN4, D3$ (internal pull-ups), $V_{IL} = 0\text{ V}$	20 —200	80 -80	200 -20	µA
Power outputs OUT1, OUT2					
$R_{DS(on)}$	Output-on resistance, $I_{LOAD} = 3.0\text{ A}$ [4] • $V_{PWR} = 8.0\text{ V}$, $T_J = 25\text{ °C}$ • $V_{PWR} = 8.0\text{ V}$, $T_J = 150\text{ °C}$ • $V_{PWR} = 5.0\text{ V}$, $T_J = 150\text{ °C}$	— — —	120 — —	— 235 325	mΩ

Symbol	Parameter	Min	Typ	Max	Unit
I _{LIM}	Output current regulation threshold <ul style="list-style-type: none"> T_J < T_{FB} T_J ≥ T_{FB} (foldback region - see Figure 10 and Figure 12) 	5.2 —	6.5 4.2	8.0 —	A
I _{SCH}	High-side short-circuit detection threshold (short-circuit to GND)	11	13	16	A
I _{SCL}	Low-side short-circuit detection threshold (short-circuit to V _{PWR})	9.0	11	14	A
I _{OUTLEAK}	Output leakage current, outputs off, V _{PWR} = 28 V <ul style="list-style-type: none"> V_{OUT} = V_{PWR} V_{OUT} = Ground 	— -60	— —	100 —	μA
V _F	Output MOSFET body diode forward voltage drop <ul style="list-style-type: none"> I_{OUT} = 3.0 A 	—	—	2.0	V
T _{LIM} T _{HYS}	Overtemperature shutdown <ul style="list-style-type: none"> Thermal limit at T_J Hysteresis at T_J 	175 —	— 12	200 —	°C
T _{FB}	Current foldback at T _J	165	—	185	°C
T _{SEP}	Current foldback to thermal shutdown separation	10	—	15	°C
High-side current sense feedback					
I _{FB}	Feedback current (pin FB sourcing current) <ul style="list-style-type: none"> I_{OUT} = 0 mA I_{OUT} = 300 mA I_{OUT} = 500 mA I_{OUT} = 1.5 A I_{OUT} = 3.0 A I_{OUT} = 6.0 A 	0.0 0.0 0.35 2.86 5.71 11.43	— 270 0.775 3.57 7.14 14.29	50 750 1.56 4.28 8.57 17.15	μA μA mA mA mA mA
Status flag ^[8]					
I _{SFLEAK}	Status flag leakage current <ul style="list-style-type: none"> V_{SF} = 5.0 V 	—	—	5.0	μA
V _{SFLOW}	Status flag set voltage <ul style="list-style-type: none"> I_{SF} = 300 μA 	—	—	0.4	V

- [1] Device specifications are characterized over the range of 8.0 V ≤ V_{PWR} ≤ 28 V. Continuous operation above 28 V may degrade device reliability. Device is operational down to 5.0 V, but below 8.0 V the output resistance may increase by 50 percent.
- [2] Device survives the transient overvoltage indicated for a maximum duration of 500 ms. Transient not to be repeated more than once every 10 seconds.
- [3] I_{PWR(SLEEP)} is with Sleep mode activated and EN/D2, = logic [0], and IN1, IN2, D1 = logic [1] or with these inputs left floating.
- [4] Output-on resistance as measured from output to VPWR and from output to GND.
- [5] This parameter is guaranteed by design.
- [6] Outputs switched OFF via D1 or EN/D2.
- [7] Accuracy is better than 20 % from 0.5 A to 6.0 A. Recommended terminating resistor value: R_{FB} = 270 Ω.
- [8] Status flag output is an open drain output requiring a pull-up resistor to logic V_{DD}.
- [9] Status flag leakage current is measured with status flag high and not set.
- [10] Status flag set voltage measured with status flag low and set with I_{SF} = 300 μA. Maximum allowable sink current from this pin is <500 μA. Maximum allowable pull-up voltage < 7.0 V.

11 Dynamic characteristics

Table 6. Dynamic characteristics

Characteristics noted under conditions 5.0 V ≤ V_{PWR} ≤ 28 V, -40 °C ≤ T_A ≤ 125 °C, GND = 0 V, unless otherwise noted. Typical values noted reflect the approximate parameter means at T_A = 25 °C under nominal conditions, unless otherwise noted.

Symbol	Parameter	Min	Typ	Max	Unit
Timing characteristics					
f _{PWM}	PWM frequency ^[1]	—	—	11	kHz
f _{MAX}	Maximum switching frequency during current limit regulation ^[2]	—	—	20	kHz

Symbol	Parameter	Min	Typ	Max	Unit
t_{DON}	Output on delay • $V_{PWR} = 14\text{ V}$	—	—	18	μs
t_{DOFF}	Output off delay • $V_{PWR} = 14\text{ V}$	—	—	12	μs
t_A	I_{LIM} output constant-off time	15	20.5	32	μs
t_B	I_{LIM} blanking time	12	16.5	27	μs
$t_{DDISABLE}$	Disable delay time	—	—	8.0	μs
t_F, t_R	Output rise and fall time	1.5	3.0	8.0	μs
t_{FAULT}	Short-circuit/overtemperature turn-off (latch-off) time	—	—	8.0	μs
t_{POD}	Power-on delay time	—	1.0	5.0	ms
t_{RR}	Output MOSFET body diode reverse recovery time	75	100	150	ns
f_{CP}	Charge pump operating frequency	—	7.0	—	MHz

- [1] The maximum PWM frequency should be limited to frequencies < 11 kHz in order to allow the internal high-side driver circuitry time to fully enhance the high-side MOSFETs.
- [2] The internal current limit circuitry produces a constant-off-time pulse width modulation of the output current. The output load's inductance, capacitance, and resistance characteristics affect the total switching period (OFF time + ON time), and thus the PWM frequency during current limit.
- [3] Output delay is the time duration from 1.5 V on the IN1 or IN2 input signal to the 20 % or 80 % point (dependent on the transition direction) of the OUT1 or OUT2 signal. If the output is transitioning high-to-low, the delay is from 1.5 V on the input signal to the 80 % point of the output response signal. If the output is transitioning low-to-high, the delay is from 1.5 V on the input signal to the 20 % point of the output response signal. See [Figure 5](#).
- [4] The time during which the internal constant-off time PWM current regulation circuit has 3-stated the output bridge.
- [5] Parameter is guaranteed by characterization
- [6] The time during which the current regulation threshold is ignored so that the short-circuit detection threshold comparators may have time to act.
- [7] Disable delay time measurement is defined in [Figure 6](#).
- [8] Rise time is from the 10 % to the 90 % level and fall time is from the 90 % to the 10 % level of the output signal with $V_{PWR} = 14\text{ V}$, $R_{LOAD} = 3.0\ \Omega$. See [Figure 7](#).
- [9] Load currents ramping up to the current regulation threshold become limited at the I_{LIM} value (see [Figure 8](#)). The short-circuit currents possess a di/dt that ramps up to the I_{SCH} or I_{SCL} threshold during the I_{LIM} blanking time, registering as a short-circuit event detection and causing the shutdown circuitry to force the output into an immediate 3-state latch-off (see [Figure 9](#)). Operation in current limit mode may cause junction temperatures to rise. Junction temperatures above $\sim 160\text{ }^\circ\text{C}$ causes the output current limit threshold to "foldback", or decrease, until $\sim 175\text{ }^\circ\text{C}$ is reached, after which the t_{LIM} thermal latch-off occurs. Permissible operation within this foldback region is limited to non-repetitive transient events of duration not to exceed 30 seconds (see [Figure 10](#)).
- [10] Parameter is guaranteed by design.

12 Timing diagrams

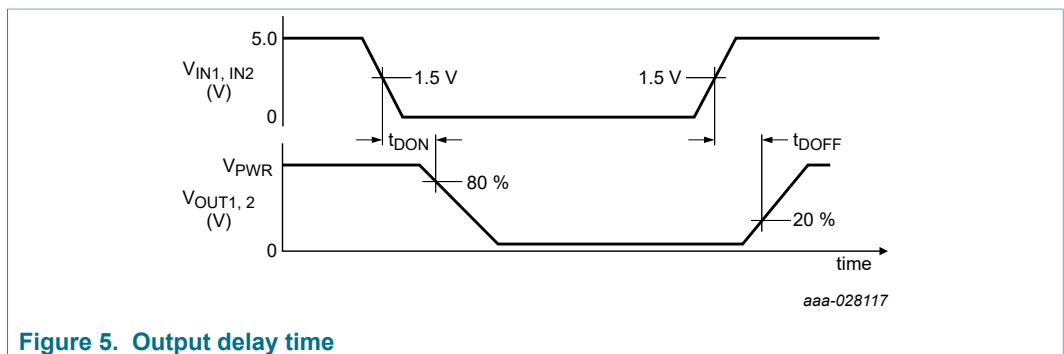


Figure 5. Output delay time

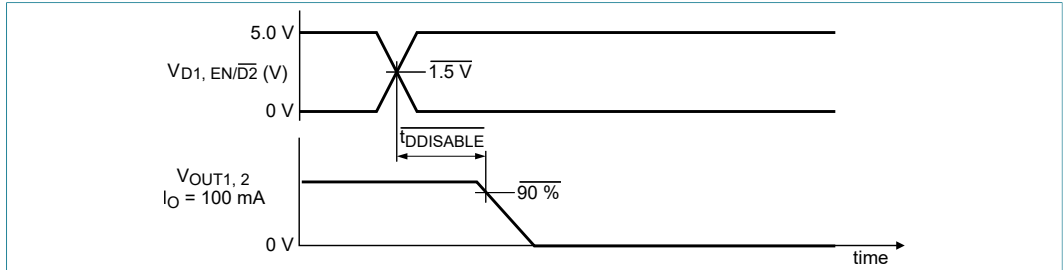


Figure 6. Disable delay time

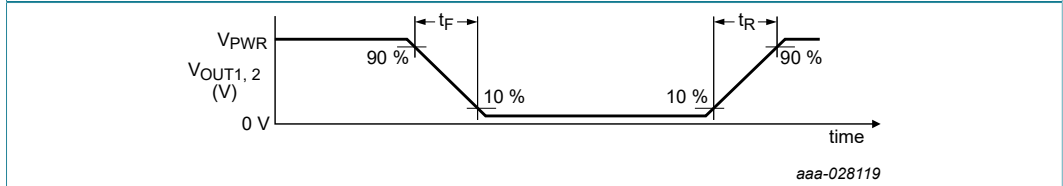


Figure 7. Output switching time

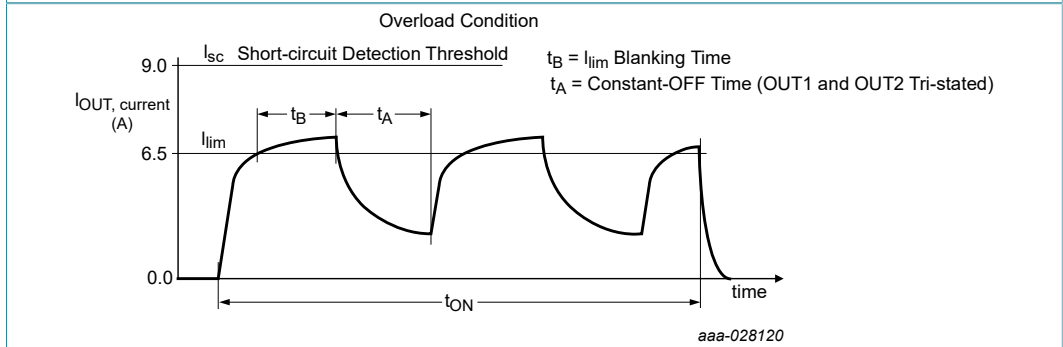


Figure 8. Current limit blanking time and constant-off time

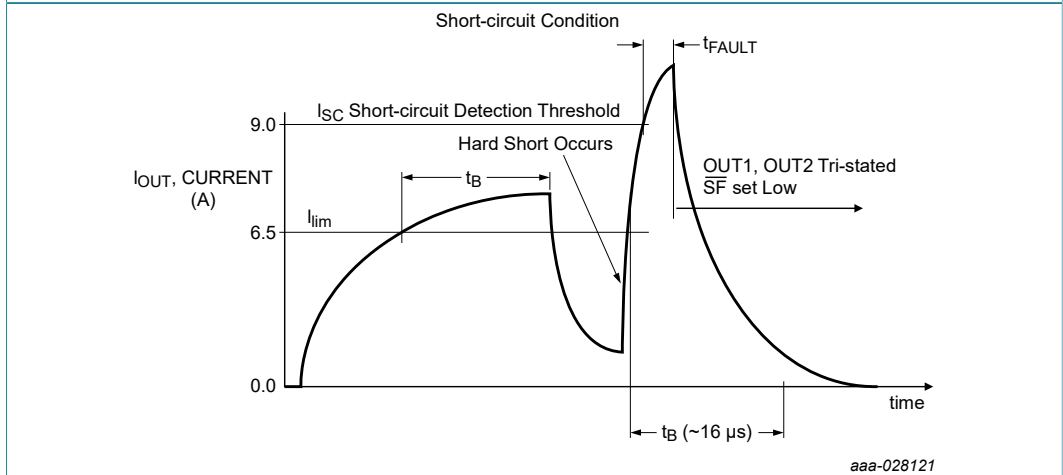


Figure 9. Short-circuit detection turn-off time t_{FAULT}

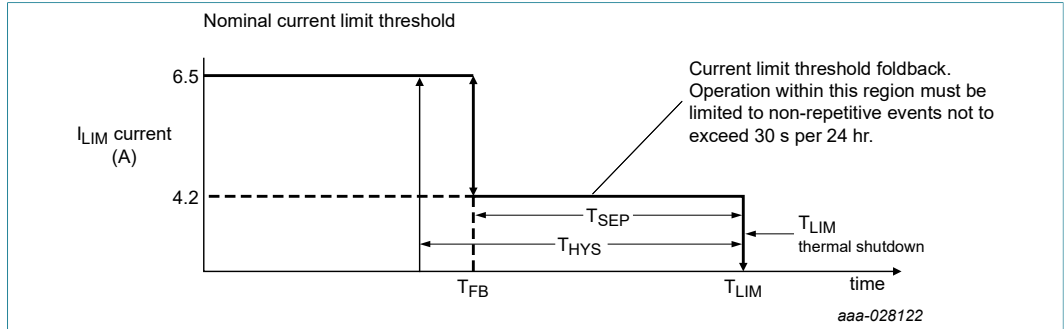


Figure 10. Output current limiting foldback region

13 Functional internal block description

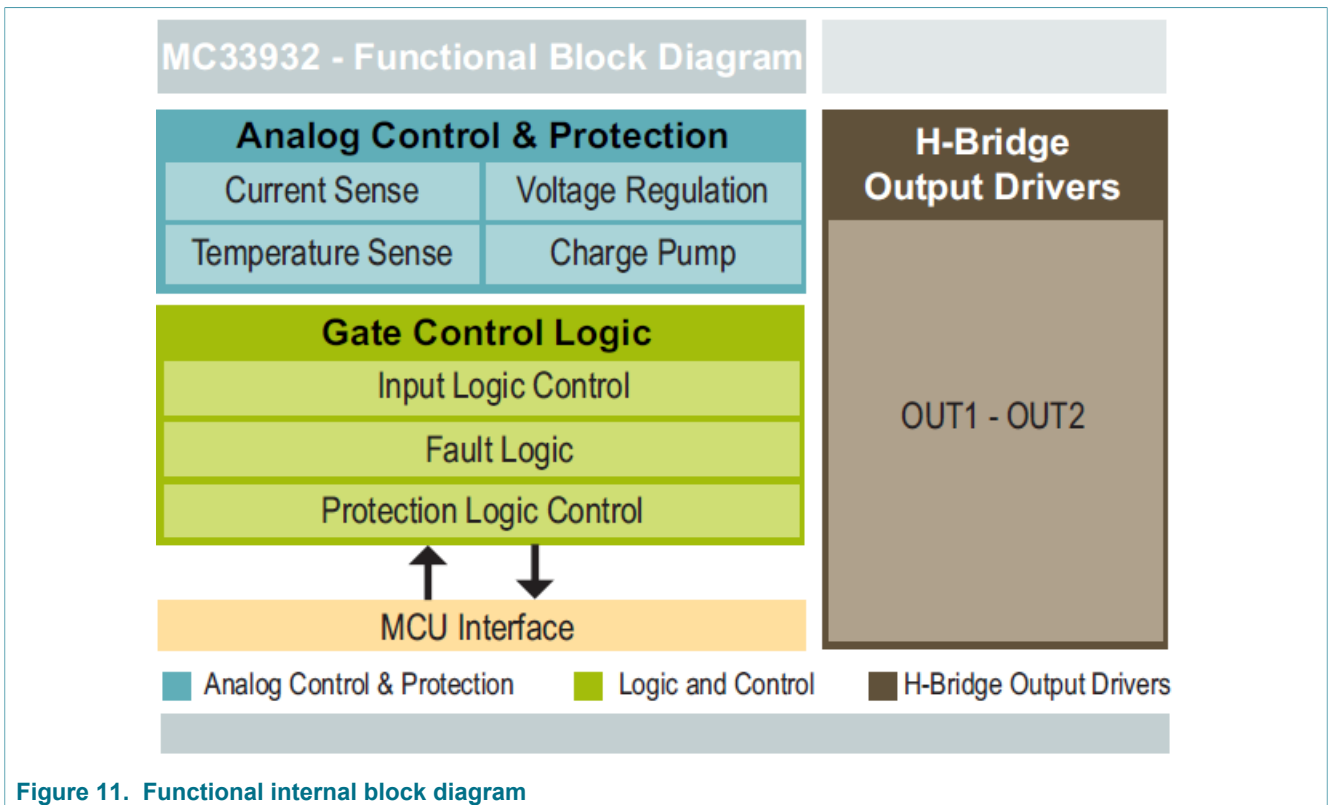


Figure 11. Functional internal block diagram

13.1 Analog control and protection circuitry

An on-chip voltage regulator supplies the internal logic. The charge pump provides gate drive for the H-bridge MOSFETs. The current and temperature sense circuitry provides detection and protection for the output drivers. Output undervoltage protection shuts down the MOSFETs.

13.2 Gate control logic

The 33932 is a monolithic H-bridge power IC designed primarily for any low-voltage DC servo motor control application within the current and voltage limits stated for the device. Two independent inputs provide polarity control of two half-bridge totem-pole outputs.

Two independent disable inputs are provided to force the H-bridge outputs to 3-state (high-impedance OFF state).

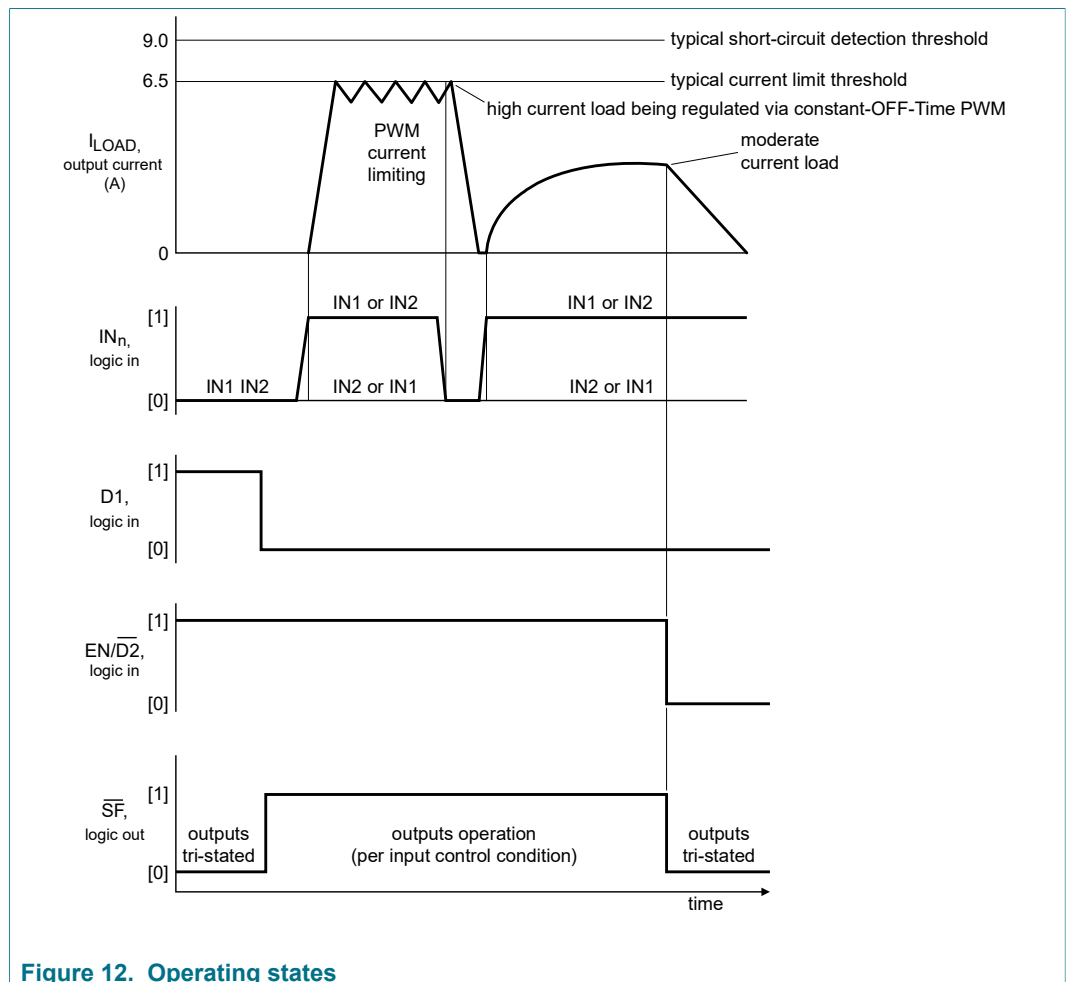
13.3 H-bridge output drivers: OUT1 and OUT2

The H-bridge is the power output stage. The current flow from OUT1 to OUT2 is reversible and under full control of the user by way of the input control logic. The output stage is designed to produce full load control under all system conditions.

All protective and control features are integrated into the control and protection blocks. The sensors for current and temperature are integrated directly into the output MOSFET for maximum accuracy and dependability.

14 Functional device operation

14.1 Operational modes



14.2 Logic commands

Table 7. Truth table

The 3-state conditions and the status flag are reset using D1 or EN/D2. The truth table uses the following notations: L = low, H = high, X = high or low, and Z = high-impedance.

Device state	Input conditions				Status	Outputs	
	EN/D2	D1	IN1	IN2		OUT1	OUT2
Forward	H	L	H	L	H	H	L
Reverse	H	L	L	H	H	L	H
Freewheeling low	H	L	L	L	H	L	L
Freewheeling high	H	L	H	H	H	H	H
Disable 1 (D1)	H	H	X	X	L	Z	Z
IN1 disconnected	H	L	Z	X	H	H	X
IN2 disconnected	H	L	X	Z	H	X	H
D1 disconnected	H	Z	X	X	L	Z	Z
Undervoltage lockout ^[1]	H	X	X	X	L	Z	Z
Overtemperature ^[2]	H	X	X	X	L	Z	Z
Short-circuit ^[2]	H	X	X	X	L	Z	Z
Sleep mode EN/D2	L	X	X	X	H	Z	Z
EN/D2 disconnected	Z	X	X	X	H	Z	Z

- [1] In the event of an undervoltage condition, the outputs 3-state and status flag are set to logic low. Upon undervoltage recovery, status flag is reset automatically or automatically cleared and the outputs are restored to their original operating condition.
- [2] When a short-circuit or overtemperature condition is detected, the power outputs are 3-state latched-off, independent of the input signals, and the status flag is latched to logic low. To reset from this condition requires the toggling of either D1, EN/D2, or V_{PWR}.

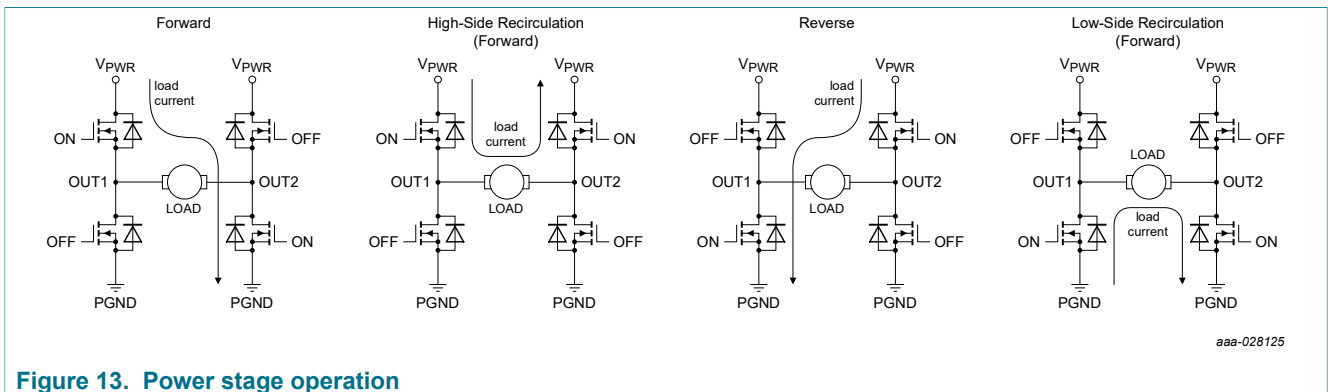


Figure 13. Power stage operation

14.3 Protection and diagnostic features

14.3.1 Short-circuit protection

If an output short-circuit condition is detected, the power outputs 3-state (latch-off) independent of the input (IN1 and IN2) states, and the fault status output flag (SF) is set to logic low. If the D1 input changes from logic high to logic low, or if the EN/D2 input changes from logic low to logic high, the output bridge becomes operational again and the fault status flag resets (cleared) to a logic High state.

The output stage always switches into the mode defined by the input pins (IN1, IN2, D1, and EN/D2), provided the device junction temperature is within the specified operating temperature range.

14.3.2 Internal PWM current limiting

The maximum current flow under normal operating conditions should be less than 5.0 A. The instantaneous load currents will be limited to I_{LIM} via the internal PWM current limiting circuitry. When the I_{LIM} threshold current value is reached, the output stages are 3-stated for a fixed time (t_A) of 20 μ s typical. Depending on the time constant associated with the load characteristics, the output current decreases during the 3-state duration until the next output ON cycle occurs.

The PWM current limit threshold value is dependent on the device junction temperature. When $-40\text{ }^\circ\text{C} < T_J < 160\text{ }^\circ\text{C}$, I_{LIM} is between the specified minimum/maximum values. When T_J exceeds $160\text{ }^\circ\text{C}$, the I_{LIM} threshold decreases to 4.2 A. Shortly above $175\text{ }^\circ\text{C}$, the device overtemperature circuit detects T_{LIM} and an overtemperature shutdown occurs. This feature implements a graceful degradation of operation before thermal shutdown occurs, thus allowing for intermittent unexpected mechanical loads on the motor's gear reduction train to be handled.

Important: Die temperature excursions above $150\text{ }^\circ\text{C}$ are permitted only for non-repetitive durations < 30 seconds. Provision must be made at the system level to prevent prolonged operation in the current-foldback region.

14.3.3 Overtemperature shutdown and hysteresis

If an overtemperature condition occurs, the power outputs are 3-stated (latched-off), and the fault status flag (\overline{SF}) is set to a logic low.

To reset from this condition, D1 must change from a logic high to logic low, or EN/D2 must change from a logic low to logic high. When reset, the output stage switches on again, provided the junction temperature is now below the overtemperature threshold limit minus the hysteresis.

Important: Resetting from the fault condition clears the fault status flag. Powering down and powering up the device also resets the 33932 from the fault condition.

14.3.4 Output avalanche protection

If VPWR becomes an open circuit, the outputs likely 3-state simultaneously due to the disable logic. This could result in an unclamped inductive discharge. The VPWR input to the 33932 should not exceed 40 V during this transient condition, to prevent electrical overstress of the output drivers. This can be accomplished with a zener clamp or MOV, and/or an appropriately valued input capacitor with sufficiently low ESR (see [Figure 14](#)).

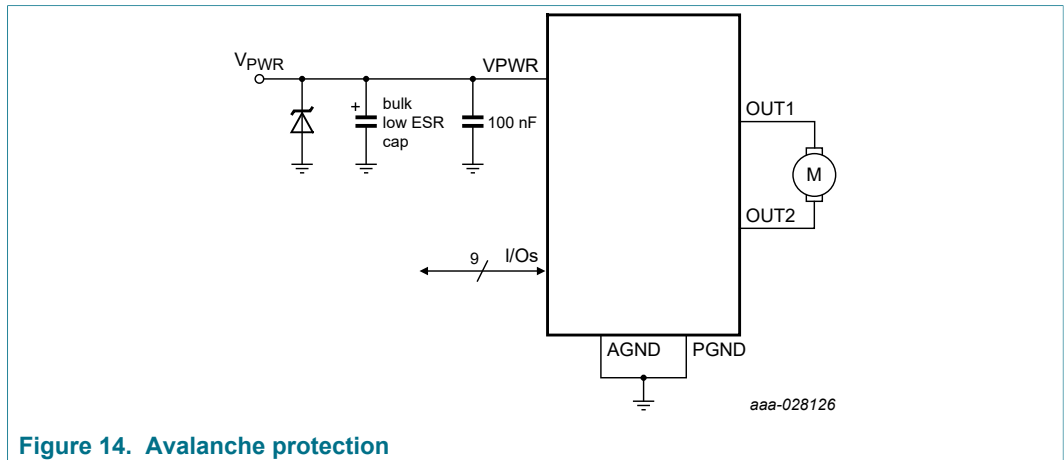


Figure 14. Avalanche protection

15 Application information

A typical application schematic is shown in Figure 15. For precision high current applications in harsh, noisy environments, the V_{PWR} bypass capacitor may need to be substantially larger.

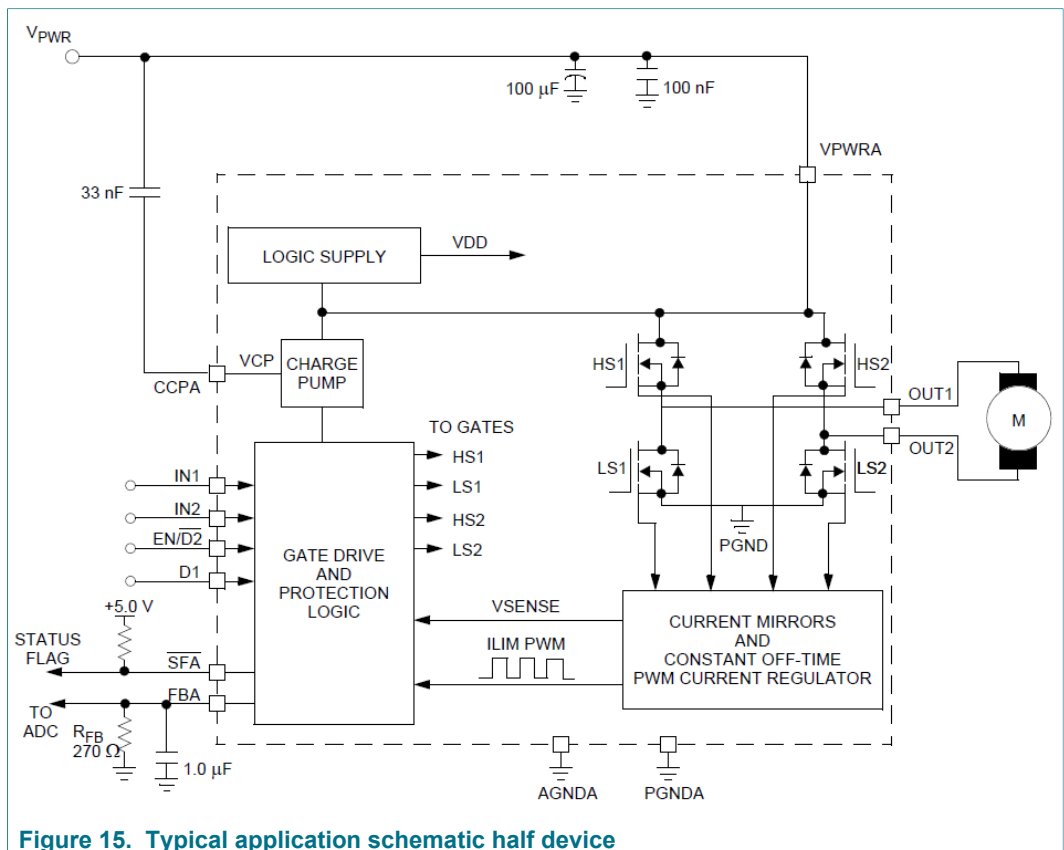
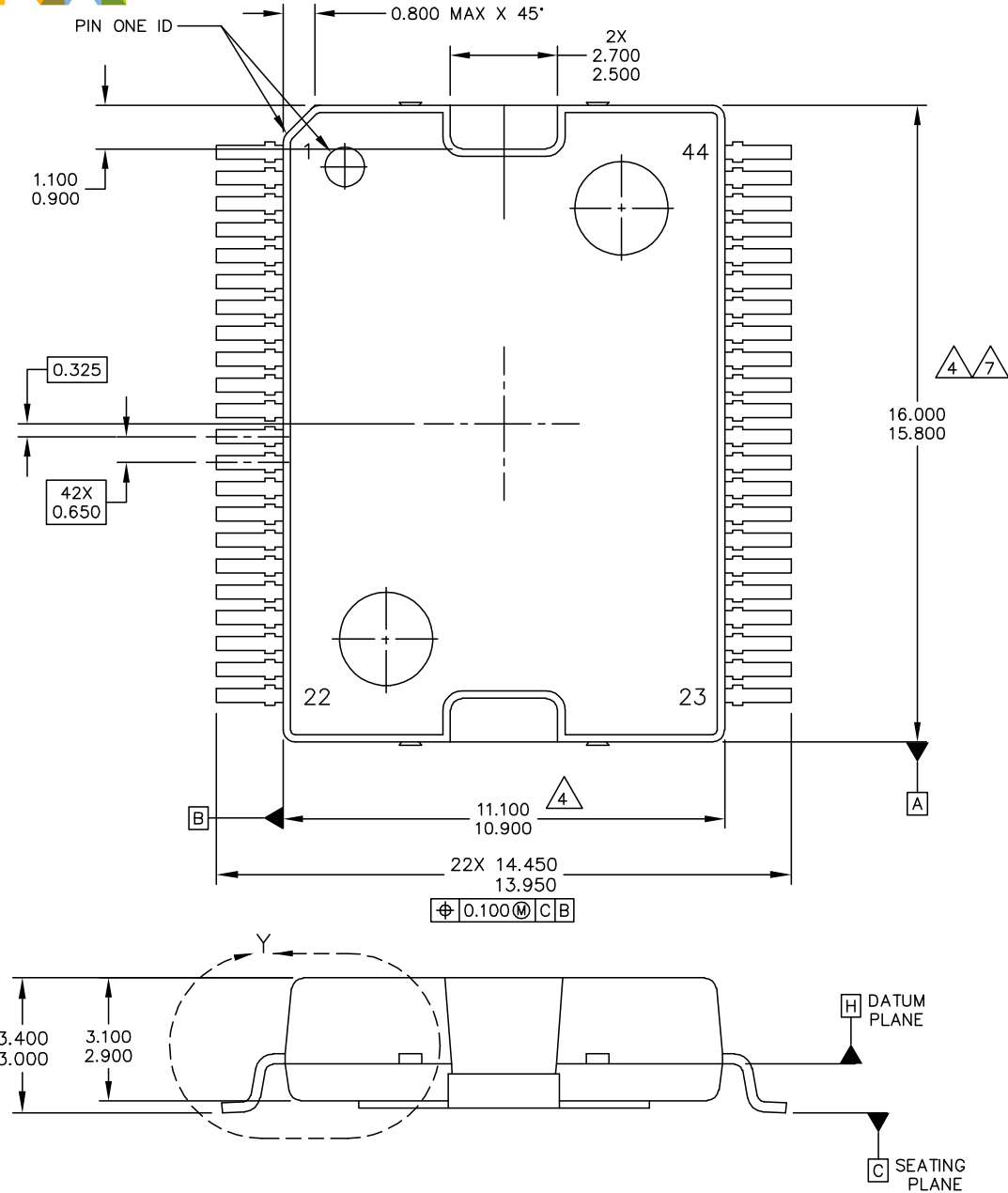


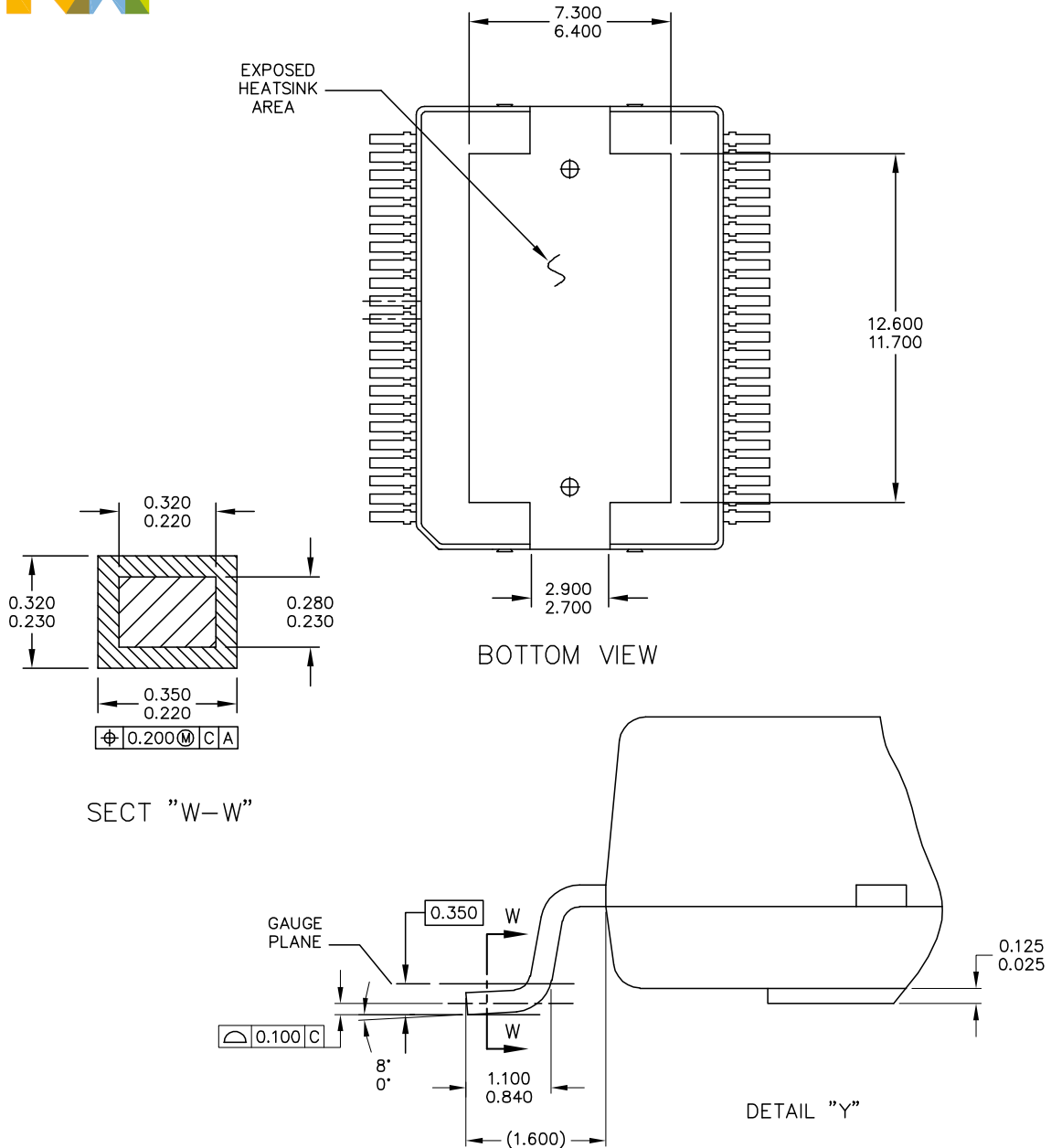
Figure 15. Typical application schematic half device

16 Package outline

Note: The most current package outline is available at www.nxp.com.



<p>© NXP SEMICONDUCTORS N.V. ALL RIGHTS RESERVED</p>	<p>MECHANICAL OUTLINE</p>	<p>PRINT VERSION NOT TO SCALE</p>
<p>TITLE: 44 LEAD HSOP W/PROTRUDING HEATSINK</p>	<p>DOCUMENT NO: 98ARH98330A REV: C</p>	<p>STANDARD: NON-JEDEC</p>
	<p>SOT1305-2</p>	<p>05 FEB 2016</p>



© NXP SEMICONDUCTORS N.V. ALL RIGHTS RESERVED	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: 44 LEAD HSOP W/PROTRUDING HEATSINK	DOCUMENT NO: 98ARH98330A REV: C	
	STANDARD: NON-JEDEC	
	SOT1305-2 05 FEB 2016	

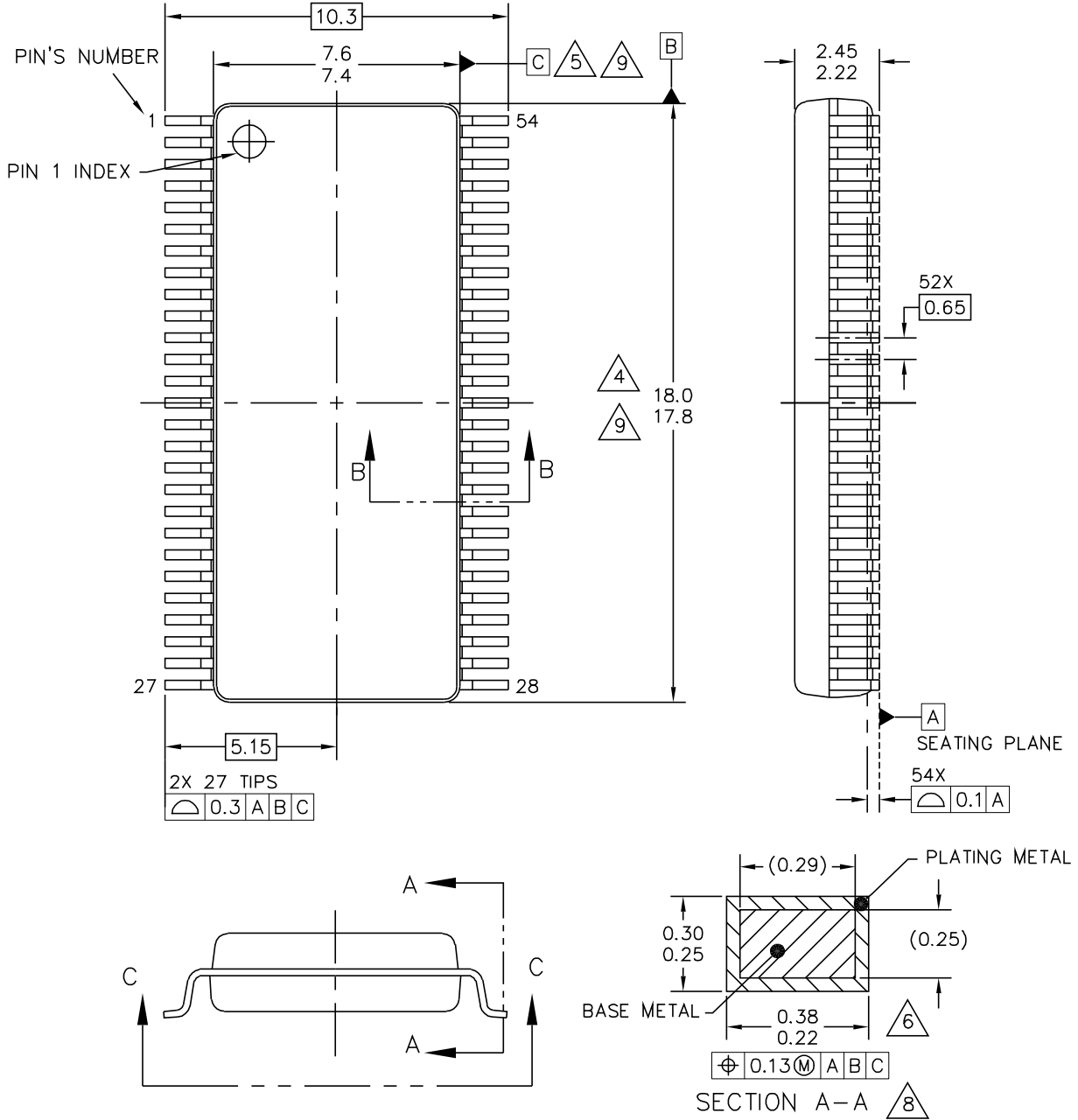


NOTES:

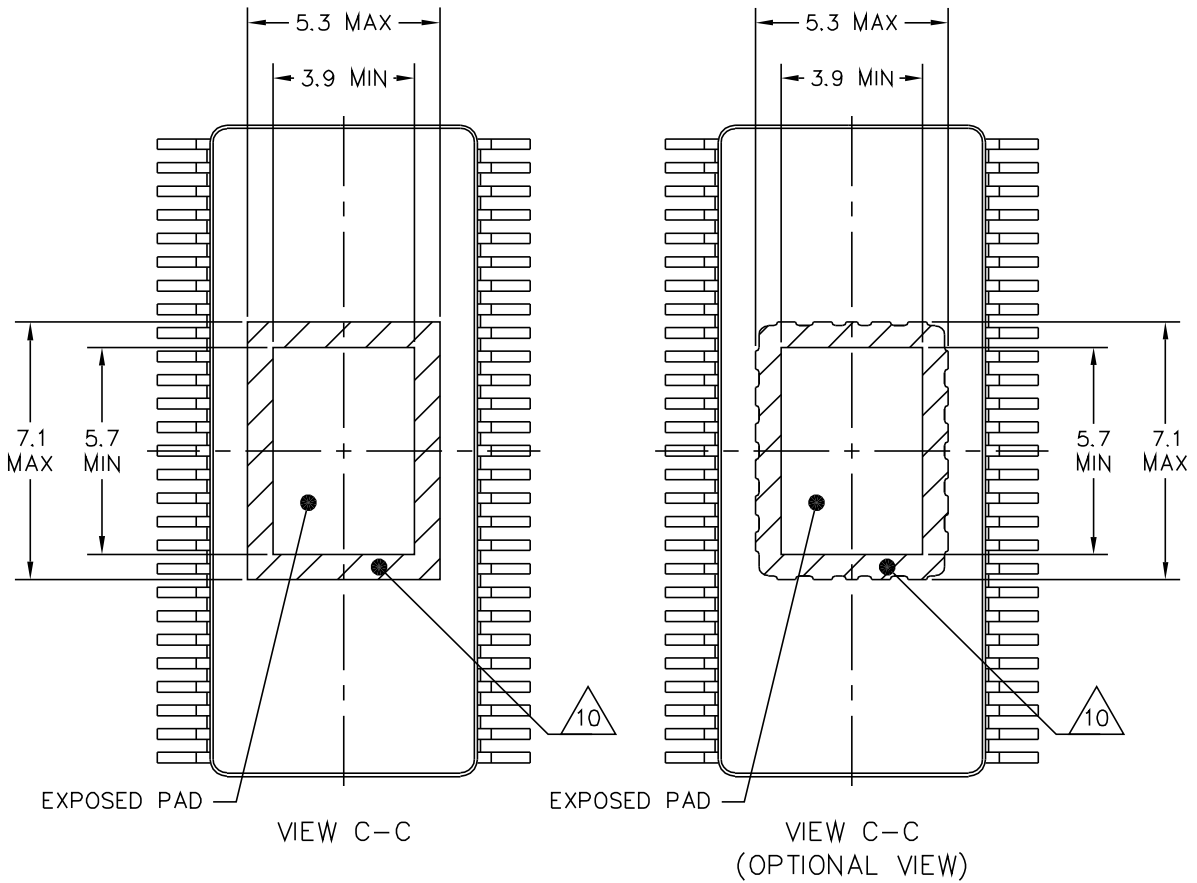
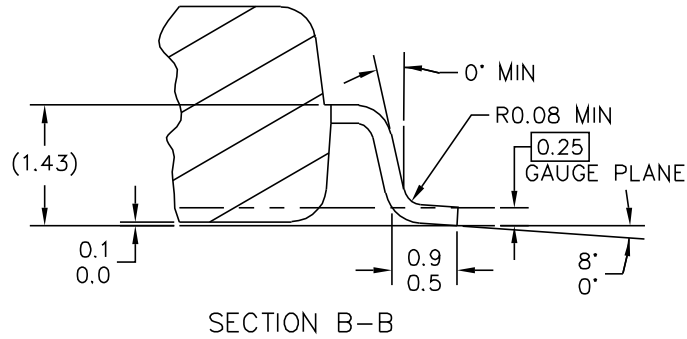
1. CONTROLLING DIMENSION: MILLIMETER
2. DIMENSIONS AND TOLERANCES PER ASME Y14.5M-1994.
3. DATUM PLANE H IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
4. THIS DIMENSIONS DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.15 PER SIDE. THIS DIMENSIONS DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE H.
5. THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 TOTAL IN EXCESS OF THE DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. DATUMS A AND B TO BE DETERMINED AT DATUM PLANE H.
7. THIS DIMENSIONS DOES NOT INCLUDE TIEBAR PROTRUSIONS. ALLOWABLE TIEBAR PROTRUSIONS ARE 0.15 PER SIDE.

© NXP SEMICONDUCTORS N.V. ALL RIGHTS RESERVED	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: 44 LEAD HSOP W/PROTRUDING HEATSINK	DOCUMENT NO: 98ARH98330A	REV: C
	STANDARD: NON-JEDEC	
	SOT1305-2	05 FEB 2016

Figure 16. Package outline SOT1305-2 (HSOP44)



© NXP SEMICONDUCTORS N. V. ALL RIGHTS RESERVED	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: SOIC W/B, 54 TERMINAL, 0.65 PITCH, 4.6 X 6.4 EXPOSED PAD	DOCUMENT NO: 98ASA99334D REV: F	
	STANDARD: NON-JEDEC	
	SOT1747-4 18 JAN 2016	



© NXP SEMICONDUCTORS N. V. ALL RIGHTS RESERVED	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: SOIC W/B, 54 TERMINAL, 0.65 PITCH, 4.6 X 6.4 EXPOSED PAD	DOCUMENT NO: 98ASA99334D	REV: F
	STANDARD: NON-JEDEC	
	SOT1747-4	18 JAN 2016



NOTES:

1. DIMENSIONS ARE IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
3. DATUMS B AND C TO BE DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.
4. THIS DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSION OR GATE BURRS. MOLD FLASH, PROTRUSION OR GATE BURRS SHALL NOT EXCEED 0.15 MM PER SIDE. THIS DIMENSION IS DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.
5. THIS DIMENSION DOES NOT INCLUDE INTER-LEAD FLASH OR PROTRUSIONS. INTER-LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED 0.25 MM PER SIDE. THIS DIMENSION IS DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.
6. THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.46 mm. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND ADJACENT LEAD SHALL NOT BE LESS THAN 0.07 mm.
7. EXACT SHAPE OF EACH CORNER IS OPTIONAL.
8. THESE DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN 0.1 mm AND 0.3 mm FROM THE LEAD TIP.
9. THE PACKAGE TOP MAY BE SMALLER THAN THE PACKAGE BOTTOM. THIS DIMENSION IS DETERMINED AT THE OUTERMOST EXTREMES OF THE PLASTIC BODY EXCLUSIVE OF MOLD FLASH, TIE BAR BURRS, GATE BURRS AND INTER-LEAD FLASH, BUT INCLUDING ANY MISMATCH BETWEEN THE TOP AND BOTTOM OF THE PLASTIC BODY.
10. THESE DIMENSION RANGES DEFINE THE PRIMARY KEEP-OUT AREA. MOLD LOCKING AND RESIN BLEED CONTROL FEATURES MAY BE VISIBLE, AND THEY MAY EXTEND TO 0.5mm FROM MAXIMUM EXPOSED PAD SIZE.

© NXP SEMICONDUCTORS N. V. ALL RIGHTS RESERVED	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: SOIC W/B, 54 TERMINAL, 0.65 PITCH, 4.6 X 6.4 EXPOSED PAD	DOCUMENT NO: 98ASA99334D	REV: F
	STANDARD: NON-JEDEC	
	SOT1747-4	18 JAN 2016

Figure 17. Package outline SOT1747-4 (HSOP54)