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# MICROCHIP MCP1790/MCP1791

# **70 mA, High Voltage Regulator**

#### **Features**

- $\cdot$  48V (43.5V  $\pm$ 10%) load dump protected for <500 ms with a 30 second repetition rate (FORD Test Pulse G Loaded)
- Wide steady state supply voltage, 6.0V 30.0V
- **Extended Junction Temperature Range:** -40 to +125°C
- Fixed output voltages: 3.0V, 3.3V, 5.0V
- Low quiescent current: 70 µA typical
- Low shutdown quiescent current: 10 µA typical
- Output Voltage Tolerances of ±2.5% over the temperature range
- Maximum output current of 70 mA  $@ + 125°C$ Junction Temperature
- Maximum continuous input voltage of 30V
- Internal thermal overload protection, +157°C (typical) Junction Temperature
- Internal short circuit current limit, 120 mA (typical) for +5V option
- Short Circuit Current Foldback
- Shutdown Input option (MCP1791)
- Power Good Output option (MCP1791)
- High PSRR, -90 dB@100 Hz (typical)
- $\cdot$  Stable with 1 µF to 1000 µF Tantalum and Electrolytic Capacitors
- Stable with 4.7 µF to 1000 µF Ceramic Capacitors

#### **Applications**

- Low Voltage A/C powered (24VAC) Fire Alarms, CO<sup>2</sup> Sensors, HVAC Controls
- Automotive Electronics
- Automotive Accessory Power Adapters
- Electronic Thermostat Controls
- Microcontroller power

#### **General Description**

The MCP1790/MCP1791 regulator provides up to 70 mA of current. The input operating voltage range is specified from 6.0V to 30V continuous, 48V absolute max, making it ideal for automotive and commercial 12/24 VDC systems.

The MCP1790/MCP1791 has a tight tolerance output voltage load regulation of ±0.2% (typical) and a very good line regulation at ±0.0002%/V (typical). The regulator output is stable with ceramic, tantalum and electrolytic capacitors. The MCP1790/MCP1791 regulator incorporates both thermal and short circuit protection.

The MCP1790 is the 3-pin version of the MCP1790/ MCP1791 family. The MCP1791 is the 5-pin version and incorporates a Shutdown input signal and a Power Good output signal.

The regulator is specifically designed to operate in the automotive environment and will survive +48V (43.5V ±10%) load dump transients and double-battery jumps. The device is designed to meet the stringent quiescent current requirements of the automotive industry. The device is also designed for the commercial low voltage fire alarm/detector systems, which use 24 VDC to supply the required alarms throughout buildings. The low ground current, 110 µA (typ.), of the CMOS device will provide a power cost savings to the end users over similar bipolar devices. Typical buildings using hundreds of 24V powered fire and smoke detectors can see substantial savings on energy consumption and wiring gage reduction compared to bipolar regulators.

The MCP1790 device will be offered in the 3-pin DD-PAK, and SOT-223 packages.

The MCP1791 device will be offered in the 5-pin DD-PAK, and SOT-223 packages.

The MCP1790/MCP1791 will have a junction temperature operating range of -40°C to +125°C.

# **Package Types**



### **TYPICAL APPLICATION**



## **1.0 ELECTRICAL CHARACTERISTICS**

#### **Absolute Maximum Ratings †**



**† Notice:** Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

# **AC/DC CHARACTERISTICS**

 ${\sf Electrical\; Specifications:}$  Unless otherwise noted,  $\rm{V_{IN}}$  =  $\rm{V_{OUT(MAX)}}$  +  $\rm{V_{DROPOUT(MAX)}}$ ,  ${\sf (Note~1)},$   $\rm{I_{OUT}}$  = 1 mA,  $\text{C}_{\text{OUT}}$  = 4.7 µF (X7R Ceramic),  $\text{C}_{\text{IN}}$  = 4.7 µF (X7R Ceramic), T<sub>A</sub> = +25°C, SHDN > 2.4V. **Boldface** type applies for junction temperatures, T<sub>J</sub> (Note 5) of -40°C to +125°C.



**Note 1:** The minimum V<sub>IN</sub>, V<sub>IN(MIN)</sub> must meet two conditions: V<sub>IN</sub>  $\geq$  6.0V and V<sub>IN</sub>  $\geq$  V<sub>OUT(MAX)</sub> +V<sub>DROPOUT(MAX)</sub>.

**2:**  $V_R$  is the nominal regulator output voltage.

- **3:** Load regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.
- **4:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air. (i.e.,  $T_A$ ,  $T_J$ ,  $\theta_{JA}$ ). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum 165°C rating. Sustained junction temperatures above 165°C can impact the device reliability.
- **5:** The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired Junction temperature. The test time is small enough such that the rise in the Junction temperature over the ambient temperature is not significant.
- **6:** Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of  $\mathsf{V}_{\mathsf{IN}}$  =  $\mathsf{V}_{\mathsf{R}}$  +  $\mathsf{V}_{\mathsf{D}\mathsf{R}\mathsf{O}\mathsf{P}\mathsf{O}\mathsf{U}\mathsf{T}(\mathsf{MAX})}$ .
- **7:** Sustained junction temperatures above 165°C can impact the device reliability.
- **8:** The Short Circuit Recovery Time test is done by placing the device into a short circuit condition and then removing the short circuit condition before the device die temperature reaches 125 °C. If the device goes into thermal shutdown, then the Short Circuit Recovery Time will depend upon the thermal dissipation properties of the package and circuit board.
- **9:** TCV<sub>OUT</sub> = (V<sub>OUT-HIGH</sub> V<sub>OUT-LOW</sub>) \*10^6/(V<sub>R</sub> \* ΔTemperature), V<sub>OUT-HIGH</sub> = highest voltage measured over the temperature range.  $V_{\text{OUT-LOW}}$  = lowest voltage measured over the temperature range.

# **AC/DC CHARACTERISTICS (CONTINUED)**

**Electrical Specifications:** Unless otherwise noted,  $V_{IN} = V_{OUT(MAX)} + V_{DROPOUT(MAX)}$ , (Note 1),  $I_{OUT} = 1$  mA,  $\text{C}_{\text{OUT}}$  = 4.7 µF (X7R Ceramic),  $\text{C}_{\text{IN}}$  = 4.7 µF (X7R Ceramic), T<sub>A</sub> = +25°C, SHDN > 2.4V.



**Note 1:** The minimum V<sub>IN</sub>, V<sub>IN(MIN)</sub> must meet two conditions: V<sub>IN</sub>  $\geq 6.0$ V and V<sub>IN</sub>  $\geq$  V<sub>OUT(MAX)</sub> + V<sub>DROPOUT(MAX)</sub>.

**2:**  $V_R$  is the nominal regulator output voltage.

- **3:** Load regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.
- **4:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air. (i.e.,  $T_A$ ,  $T_J$ ,  $\theta_{JA}$ ). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum 165°C rating. Sustained junction temperatures above 165°C can impact the device reliability.
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- **6:** Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of  $\mathsf{V}_{\mathsf{IN}}$  =  $\mathsf{V}_{\mathsf{R}}$  +  $\mathsf{V}_{\mathsf{D}\mathsf{R}\mathsf{O}\mathsf{P}\mathsf{O}\mathsf{U}\mathsf{T}(\mathsf{MAX})}$ .
- **7:** Sustained junction temperatures above 165°C can impact the device reliability.
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- **9:** TCV<sub>OUT</sub> = (V<sub>OUT-HIGH</sub> V<sub>OUT-LOW</sub>) \*10^6/(V<sub>R</sub> \* ΔTemperature), V<sub>OUT-HIGH</sub> = highest voltage measured over the temperature range.  $V_{\text{OUT-LOW}}$  = lowest voltage measured over the temperature range.

# **AC/DC CHARACTERISTICS (CONTINUED)**

**Electrical Specifications:** Unless otherwise noted,  $V_{IN} = V_{OUT(MAX)} + V_{DROPOUT(MAX)}$ , (Note 1),  $I_{OUT} = 1$  mA,  $C_{\text{OUT}} = 4.7 \,\mu\text{F}$  (X7R Ceramic),  $C_{\text{IN}} = 4.7 \,\mu\text{F}$  (X7R Ceramic),  $T_A = +25^{\circ}$ C, SHDN > 2.4V. **Boldface** type applies for junction temperatures, T<sub>1</sub> (Note 5) of -40°C to +125°C.



**Note 1:** The minimum V<sub>IN</sub>, V<sub>IN(MIN)</sub> must meet two conditions: V<sub>IN</sub>  $\geq$  6.0V and V<sub>IN</sub>  $\geq$  V<sub>OUT(MAX)</sub> +V<sub>DROPOUT(MAX)</sub>.<br>2. V<sub>o</sub> is the nominal requilator output voltage

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**5:** The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired Junction temperature. The test time is small enough such that the rise in the Junction temperature over the ambient temperature is not significant.

**6:** Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of V<sub>IN</sub> = V<sub>R</sub> + V<sub>DROPOUT(**MAX**)·</sub>

**7:** Sustained junction temperatures above 165°C can impact the device reliability.

**8:** The Short Circuit Recovery Time test is done by placing the device into a short circuit condition and then removing the short circuit condition before the device die temperature reaches 125 °C. If the device goes into thermal shutdown, then the Short Circuit Recovery Time will depend upon the thermal dissipation properties of the package and circuit board.

**9:** TCV<sub>OUT</sub> = (V<sub>OUT-HIGH</sub> - V<sub>OUT-LOW</sub>) \*10^6/(V<sub>R</sub> \* ΔTemperature), V<sub>OUT-HIGH</sub> = highest voltage measured over the temperature range.  $V_{OUT-LOW}$  = lowest voltage measured over the temperature range.

# **TEMPERATURE SPECIFICATIONS**



# **2.0 TYPICAL PERFORMANCE CHARACTERISTICS**

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, C<sub>OUT</sub> = 4.7 uF Ceramic (X7R), C<sub>IN</sub> = 10.0 µF Ceramic (X7R), I<sub>OUT</sub> = 1 mA, Temperature = +25°C,  $V_{\text{IN}}$  = 6.0V, R<sub>PWRGD\_PULLUP</sub> = 10 kΩ To  $V_{\text{OUT}}$ ,  $\overline{V}_{\overline{\text{SHDN}}}$  =  $V_{\text{IN}}$  and device is MCP1790.

**Note:** Junction Temperature (TJ) is approximated by soaking the device under test to an ambient temperature equal to the desired junction temperature. The test time is small enough such that the rise in Junction Temperature over the Ambient temperature is not significant.



*FIGURE 2-1: Power Good Time Delay vs. Temperature (MCP1791).*



*FIGURE 2-2: Quiescent Current vs. Input Voltage.*



*Voltage.*



*FIGURE 2-4: Quiescent Current vs. Junction Temperature.*



*FIGURE 2-5: Ground Current vs. Load Current.*







**Note:** Unless otherwise indicated, C<sub>OUT</sub> = 4.7 uF Ceramic (X7R), C<sub>IN</sub> = 10.0 µF Ceramic (X7R), I<sub>OUT</sub> = 1 mA, Temperature = +25°C,  $V_{\text{IN}}$  = 6.0V, R<sub>PWRGD</sub> <sub>PULLUP</sub> = 10 kΩ To V<sub>OUT</sub>,  $\overline{V}_{\overline{\text{SHDN}}}$  = V<sub>IN,</sub> and device is MCP1790.







*FIGURE 2-8: Line Regulation vs. Temperature.*



*Temperature.*





*Current.*





*FIGURE 2-11: Dropout Voltage vs. Load Current.*



*Temperature.*

*FIGURE 2-12: Dropout Voltage vs.* 

**Note:** Unless otherwise indicated,  $C_{OUT} = 4.7$  uF Ceramic (X7R),  $C_{IN} = 10.0$  µF Ceramic (X7R),  $I_{OUT} = 1$  mA, Temperature = +25°C,

**130 ROUT < 0.1Ω**  $\widehat{\mathbf{f}}$  125 **Short Circuit Current (mA)** Current **120**  $= 5.0V$ **115** Circuit **VOUT = 3.3V 110**  $\overline{\mathbf{r}}$ **105**  $3.0$ င်္တ **100 6 10 14 18 22 26 30 Input Voltage (V)**

 $V_{\text{IN}}$  = 6.0V, R<sub>PWRGD</sub> PULLUP = 10 kΩ To V<sub>OUT</sub>,  $\overline{V}_{\overline{\text{SHDN}}}$  = V<sub>IN</sub> and device is MCP1790.

*FIGURE 2-13: Short Circuit Current vs Input Voltage.*



*FIGURE 2-14: Output Noise Voltage Density vs. Frequency.*



*FIGURE 2-15: Power Supply Ripple Rejection vs. Frequency.*



*FIGURE 2-16: Power Supply Ripple Rejection vs. Frequency.*



*FIGURE 2-17: Startup from VIN (MCP1791).*



*FIGURE 2-18: Startup from Shutdown (MCP1791).*

Note: Unless otherwise indicated, C<sub>OUT</sub> = 4.7 uF Ceramic (X7R), C<sub>IN</sub> = 10.0 µF Ceramic (X7R), I<sub>OUT</sub> = 1 mA, Temperature = +25°C,  $V_{IN}$  = 6.0V, R<sub>PWRGD</sub> PULLUP = 10 kΩ To V<sub>OUT</sub>,  $\overline{V}_{\overline{SHDN}}$  = V<sub>IN,</sub> and device is MCP1790.









*FIGURE 2-20: Dynamic Line Response.*



*FIGURE 2-21: Dynamic Load Response.*



*FIGURE 2-22: Dynamic Load Response.*







*FIGURE 2-24: Short Circuit Response.*

**Note:** Unless otherwise indicated, C<sub>OUT</sub> = 4.7 uF Ceramic (X7R), C<sub>IN</sub> = 10.0 µF Ceramic (X7R), I<sub>OUT</sub> = 1 mA, Temperature = +25°C, V<sub>IN</sub> = 6.0V, R<sub>PWRGD\_PULLUP</sub> = 10 kΩ To V<sub>OUT</sub>, V<sub>SHDN</sub> = V<sub>IN,</sub> and device is MCP1790.



*FIGURE 2-25: Startup Ground Current.*

# **3.0 PIN DESCRIPTIONS**

The descriptions of the pins are listed in Table 3-1 and Table 3-2.









## **3.1 Input Voltage Supply (VIN)**

Connect the unregulated or regulated input voltage source to  $V_{IN}$ . If the input voltage source is located several inches away from the regulator or the input source is a battery, it is recommended that an input capacitor is used. A typical input capacitance value of 1 µF to 10 µF should be sufficient for most applications. The type of capacitor used can be ceramic, tantalum or aluminum electrolytic. The low ESR characteristics of the ceramic will yield better noise and PSRR performance at high-frequency.

## **3.2 Ground (GND)**

Tie GND to the negative side of the output and the negative side of the input capacitor. Only the regulator bias current flows out of this pin; there is no high current. The regulator output regulation is referenced to this pin. Minimize voltage drops between this pin and the negative side of the load.

# **3.3** Regulated Output Voltage (V<sub>OUT</sub>)

The  $V_{IN}$  pin is the regulated output voltage of the regulator. A minimum output capacitance of 1.0 µF tantalum, 1.0  $\mu$ F electrolytic, or 4.7  $\mu$ F ceramic is required for stability. The MCP1790 is stable with ceramic, tantalum, and electrolytic capacitors. See **Section 4.7 "Output Capacitor"** for output capacitor selection guidance.

# **3.4 Shutdown (SHDN)**

The SHDN pin is an active-low input signal that turns the regulator output voltage on and off. When the SHDN input is at a logic-high level, the regulator output voltage is enabled. When the SHDN input is pulled to a logic-low level, the regulator output voltage is disabled. When the SHDN input is pulled low, the PWRGD output signal also goes low and the regulator enters a low quiescent current shutdown state where the typical quiescent current is 10  $\mu$ A. The  $\overline{\text{SHDN}}$  pin is bonded to  $\text{V}_{\text{IN}}$ in the 3-pin versions of the regulator. See Table 4-1.

## **3.5 Power Good Output (PWRGD)**

The PWRGD pin is an open-drain output signal that is used to indicate when the regulator output voltage is within 90% (typically) of its nominal regulation value. The PWRGD threshold has a typical hysteresis value of 2%. The typical PWRGD delay time due to  $V_{OUT}$ rising above 90% +3% (maximum hysteresis) is 30 µs. The typical PWRGD delay time due to  $V_{\text{OUT}}$  falling below 90% is 235 µs. These delay times are internally fixed.

## **3.6 Exposed Pad (EP)**

The DDPAK package has an exposed tab on the package. A heat sink may be mounted to the tab to aid in the removal of heat from the package during operation.

The exposed tab or pad of all of the available packages is at the ground potential of the regulator.

# **4.0 DEVICE OVERVIEW**

The MCP1790/MCP1791 are high input voltage regulators capable of providing up to 70 mA of current at output voltages up to 5.0V. The devices have an input voltage operating range from 6.0V to 30V, 48V absolute max. The MCP1790 devices are 3-pin devices consisting of  $V_{IN}$ ,  $V_{OUIT}$  and GND. The MCP1791 devices have 5 or more pins which add Shutdown and Power-Good functions to the MCP1790 device.

#### **4.1 Startup**

In the start phase, the device must see at least 6.0V to initiate operation during power up. In the Power-down mode, the  $V_{IN}$  monitor will be turned off.

#### **4.2 Thermal Shutdown**

The regulator has a thermal shutdown. If the thermal protection circuit detects an over temperature condition, typically 157°C junction temperature, the regulator will shut down. The device will recover from the thermal shutdown when the junction temperature drops to 137°C (typical).

### **4.3 Regulator Output Voltage**

The MCP1790/MCP1791 regulators are available in fixed output voltage configurations. The standard output voltages are 3.0V, 3.3V, and 5.0V.

#### **4.4 External Protection**

#### 4.4.1 TRANSIENT VOLTAGE PROTECTION (LOAD DUMP)

The regulator is capable of withstanding a 48V (43.5V ±10) load dump transient for a duration of 500 ms and a repetition rate of 30 seconds (ES-XW7T-1A278-AC, Ford Motor Company, Test Pulse G Loaded).

While not necessary, good design practice dictates adding an external transient suppressor, between VBB and ground, with a low value resistor in series with the battery supply and the  $V_{IN}$  pin, to limit currents and voltages due to electrical transients. Because of the regulator startup current, the resistor value should be less than  $(V<sub>BAT</sub> -6V) / 200$  mA. For a 12V battery voltage, the resistor value should be less than 30 ohms.

#### 4.4.2 REVERSE BATTERY PROTECTION

An external reverse battery blocking diode may be used to provide polarity protection.



*FIGURE 4-1: MCP1790/MCP1791 Block Diagram.*

# **4.5 Shutdown (SHDN)**

The MCP1791 has a Shutdown (SHDN) input signal that enables or disables the regulator output voltage. When the SHDN input signal is greater than 2.40V, the regulator output voltage is enabled. Note that the regulator output may still be disabled by the undervoltage lockout incorporated within the  $V_{1N}$ circuitry.

The value of the SHDN signal to put the regulator into Shutdown mode is  $\leq 0.8V$ . The SHDN pin is pulled low by an internal resistor. If the SHDN pin is left floating, the internal pull-down resistor will put the regulator into shutdown mode.

When the SHDN input signal is pulled to a logic-low, the PWRGD output signal will also go low and the regulator will enter a low quiescent current state where the typical quiescent current is 10 µA. There is a short time delay (approximately 400 ns) when the SHDN input signal transitions from high-to-low to prevent signal noise from disabling the regulator. The SHDN pin will ignore low-going pulses that are up to 400 ns in pulse width. If the SHDN input is pulled low for more than 400 ns, the regulator will enter Shutdown mode. This small bit of filtering helps to reject any system noise spikes on the SHDN input signal.

On the rising edge of the SHDN input, the shutdown circuitry will have a 100 µs delay before allowing the regulator output to turn on. This delay helps to reject any false turn-on signals or noise on the SHDN input signal. After the 100 us delay, the regulator will start charging the output capacitor as the regulator output voltage rises from 0V to its final regulated value. The charging current will be limited by the short circuit current value of the device. If the SHDN input signal is pulled low during the 100 µs delay period, the timer will be reset and the delay time will start over again on the next rising edge of the SHDN input. The total time from the SHDN input going high (turn-on) to the regulator output being in regulation shall typically be 200 µs (100  $\mu$ s + 100  $\mu$ s) for a C<sub>LOAD</sub> = 1.0  $\mu$ F.



*FIGURE 4-2: Shutdown Input Timing Diagram.*

#### **4.6 Low Voltage Shutdown**

The MCP1790/MCP1791 incorporates a Low Voltage Shutdown circuit that turns off the output of the regulator whenever the input voltage,  $V_{IN}$ , is below the specified turn off voltage,  $V_{OFF}$ . When the input voltage  $(V_B)$  drops below the differential needed to provide stable regulation, the output voltage ( $V_{REG}$ ) shall track the input down to approximately +4.00V. The regulator will turn off the output at this point.

The output will turn on when  $V_{IN}$  rises above the  $V_{ON}$ value specified in the data sheet. This feature is independent of the Shutdown input signal (SHDN) that is provided for external regulator control. If the SHDN input signal is active (LOW), then the output of the regulator shall be disabled regardless of input voltage.





## **4.7 Output Capacitor**

The MCP1790/MCP1791 requires a minimum output capacitance of 1 µF tantalum or electrolytic capacitance. The minimum value for ceramic capacitors is 4.7 µF. The regulator is stable for all three types of capacitors from  $4.7 \,\mu$ F to 1000  $\mu$ F (see Figure 4-3). The MCP1790/MCP1791 regulator may be used with a 1 µF ceramic output capacitor if a 0.300Ω resistor is placed in series with the capacitor. The low ESR and corresponding pole of the ceramic capacitor causes the instability below 4.7 µF.

The Equivalent Series Resistance (ESR) of the output capacitor must be no greater than 3 ohms. The output capacitor should be located as close to the regulator output as is practical. Ceramic materials X7R and X5R have low temperature coefficients and are recommended because of their size, cost and environmental robustness qualities.

#### **4.8 Output Current and Current Limiting**

The MCP1790/MCP1791 devices are tested and ensured to supply a minimum of 70 mA of output current.

The MCP1790/MCP1791 also incorporate an output current limit foldback. When the regulator is in an overcurrent condition,  $V_{OUT}$  will decrease with increasing load. When  $V_{\text{OUT}}$  falls below 30% (typical) of  $V_R$ , the output current will start to fold back. The output current will fold back to less than 100 mA (typical) when  $V_{OUT}$  is near 0 volts.

The 5.0V regulator has an overload current limiting of approximately 120 mA. If  $V_{REG}$  is lower than 3.5V,  $I_{OUT}$ will start to fold back and decrease along with  $V_{\text{REG}}$ until IOUT is less than 105 mA and  $V_{RFG}$  is near 0 volts.

The 3.3V regulator has an overload current limiting of approximately 130 mA. If  $V_{REG}$  is lower than 2.5V,  $I_{OUT}$ will start to fold back and decrease along with  $V_{REG}$ until IOUT is less than 99 mA and  $V_{REG}$  is near 0 volts.



*FIGURE 4-3: ESR Curves for Load Capacitor Selection.*

#### **4.9 Power Good Output (PWRGD)**

The MCP1791 has an open-drain Power Good (PWRGD) output signal capable of sinking a minimum of 5.0 mA of current while maintaining a PWRGD output voltage of 0.4V or less.

As the output voltage of the LDO rises, the PWRGD output will be held low until the output voltage has exceeded the power good threshold (V<sub>PWRGD\_TH</sub>) level by an amount equal to the power good hysteresis value (( $V_{PWRGD\ HYS}$ ), typically 2% of  $V_R$ . Once this threshold has been exceeded, the power good output signal will be pulled high by an external pull-up resistor, indicating that the output voltage is stable and within regulation **limits** 

If the output voltage of the LDO falls below the power good threshold ( $V_{PWRGBTH}$ ) level, the power good output will transition low. The power good circuitry has a 235 µs delay when detecting a falling output voltage, which helps to increase noise immunity of the power good output and avoid false triggering of the power good output during fast output transients. See Figure 4-4 for power good timing characteristics.

When the LDO is put into Shutdown mode using the SHDN input, the power good output is pulled low within 400 ns typical, indicating that the output voltage will be out of regulation. The timing diagram for the power good output when using the shutdown input is shown in Figure 4-5.

The PWRGD output may be pulled up to either  $V_{\text{IN}}$  or  $V<sub>OUT</sub>$ . When pulled to  $V<sub>OUT</sub>$ , the PWRGD output will sink very little current during shutdown. When PWRGD is pulled up to  $V_{\text{IN}}$ , the PWRGD output will sink current during shutdown. That is because  $V_{\text{OUT}}$  is 0 during shutdown while  $V_{\text{IN}}$  is still active. When the PWRGD output is pulled to  $V_{IN}$ , the PWRGD output signal will track  $V_{IN}$  at startup until the threshold of the PWRGD circuitry has been reached and the PWRGD circuitry pulls the signal back low. Therefore, when pulling PWRGD to  $V_{IN}$  instead of  $V_{OUT}$ , the designer must be aware of the PWRGD signal going high while the input voltage is rising at startup. Pulling PWRGD to  $V_{OUT}$ removes the startup pulse.



*FIGURE 4-4: Power Good Timing.*



*Shutdown.*

#### **5.0 APPLICATION CIRCUITS/ ISSUES**

### **5.1 Typical Application**

The MCP1790/MCP1791 is most commonly used as a voltage regulator. It is a high voltage input capability and thermal protection make it ideal for automotive and 24V industrial applications.





### 5.1.1 APPLICATION INPUT CONDITIONS

Package Type = SOT-223-5 Input Voltage Range = 8V to 24V  $V_{IN}$  maximum = 24V  $V_{\text{OUT}}$  typical = 5.0V  $I_{\text{OUT}}$  = 70 mA maximum

#### **5.2 Power Calculations**

#### 5.2.1 POWER DISSIPATION

The internal power dissipation of the MCP1790/ MCP1791 is a function of input voltage, output voltage and output current. The power dissipation, as a result of the quiescent current draw, is so low, it is insignificant (70.0  $\mu$ A x V<sub>IN</sub>). The following equation can be used to calculate the internal power dissipation of the LDO.

#### **EQUATION 5-1:**



The maximum continuous operating junction temperature specified for the MCP1790/MCP1791 is +125°C. To estimate the internal junction temperature of the MCP1790/MCP1791, the total internal power dissipation is multiplied by the thermal resistance from junction to ambient ( $R\theta_{JA}$ ). The thermal resistance from junction to ambient for the SOT-223-5 package is estimated at 62°C/W.

#### **EQUATION 5-2:**

$$
T_{J(MAX)} = P_{TOTAL} \times R\Theta_{JA} + T_{AMAX}
$$



The maximum power dissipation capability for a package can be calculated given the junction-toambient thermal resistance and the maximum ambient temperature for the application. The following equation can be used to determine the package maximum internal power dissipation.

#### **EQUATION 5-3:**



#### **EQUATION 5-4:**



#### **EQUATION 5-5:**

$$
T_J = T_{J(RISE)} + T_A
$$
  
\n
$$
T_J = \text{Junction Temperature}
$$
  
\n
$$
T_{J(RISE)} = \text{Rise in device junction}
$$
  
\ntemperature over the ambient temperature  
\n
$$
T_A = \text{Ambient temperature}
$$

#### **5.3 Power Dissipation Example**

Internal power dissipation, junction temperature rise, junction temperature and maximum power dissipation are calculated in the following example. The power dissipation, as a result of ground current, is small enough to be neglected.

#### 5.3.1 POWER DISSIPATION EXAMPLE

#### **Package:**

Package Type = SOT-223-5

#### **Input Voltage:**

 $V_{IN}$  = 8V to 24V

**LDO Output Voltages and Currents:**

 $V_{OUT} = 5.0V$ 

 $I_{\text{OUT}}$  = 50 mA

**Maximum Ambient Temperature:**

 $T_{A(MAX)} = +40^{\circ}C$ 

#### **Internal Power Dissipation:**

Internal Power dissipation is the product of the LDO output current times the voltage across the LDO  $(V_{IN}$  to  $V_{OUT}$ ).

 $P_{LDO(MAX)} = (V_{IN(MAX)} - V_{OUT(MIN))} x I_{OUT(MAX)}$  $P_{LDO}$  = (24V - (0.98 x 5.0V)) x 50 mA  $P_{LDO}$  = 955 milli-Watts

#### 5.3.1.1 Device Junction Temperature Rise

```
The internal junction temperature rise is a function of
internal power dissipation and the thermal resistance
from junction to ambient for the application. The
thermal resistance from junction to ambient ( R \theta_{JA} ) is
derived from an EIA/JEDEC standard for measuring
thermal resistance for small surface mount packages.
The EIA/JEDEC specification is JESD51-7, "High
Effective Thermal Conductivity Test Board for Leaded
Surface Mount Packages". The standard describes the
test method and board specifications for measuring the
thermal resistance from junction to ambient. The actual
thermal resistance for a particular application can vary
depending on many factors, such as copper area and
thickness. Refer to AN792, "A Method to Determine
How Much Power a SOT23 Can Dissipate in an
Application", (DS00792), for more information
regarding this subject.
```
 $T_{\text{J(RISE)}} = P_{\text{TOTAL}} \times Rq_{\text{JA}}$  $T_{JRISF}$  = 955 milli-Watts x 62°C/Watt  $T_{\text{JRISE}} = 59.2^{\circ} \text{C}$ 

#### 5.3.1.2 Junction Temperature Estimate

To estimate the internal junction temperature, the calculated temperature rise is added to the ambient or offset temperature. For this example, the worst-case junction temperature is estimated below.

$$
T_J = T_{JRISE} + T_{A(MAX)}
$$
  

$$
T_J = 99.2^{\circ}C
$$

5.3.1.3 Maximum Package Power Dissipation at +40°C Ambient **Temperature** 

SOT-223-5 (62°C/Watt =  $R\theta_{JA}$ )

 $P_{D(MAX)} = (125^{\circ}C - 40^{\circ}C)/62^{\circ}C/W$ 

 $P_{D(MAX)} = 1.371$  Watts

DDPAK-5 (32°C/Watt =  $R\theta_{JA}$ )

 $P_{D(MAX)} = (125^{\circ}C - 40^{\circ}C)/32^{\circ}C/W$  $P_{D(MAX)} = 2.656$  Watts

# **5.4 Pulsed Load Applications**

For some applications, there are pulsed load current events that may exceed the specified 70 mA maximum specification of the MCP1790/MCP1791. The internal current foldback feature of the MCP1790/MCP1791 will prevent high peak load demands from causing non-recoverable damage. The Current Foldback feature of the device will limit the output voltage and output current during pulsed applications. As the current rises above the foldback current threshold, the output voltage will decrease.

# **6.0 PACKAGING INFORMATION**

### **6.1 Package Marking Information**

#### **3-Lead DDPAK (MCP1790) Example:**



#### **3-Lead SOT-223 (MCP1790)**









# **6.1 Package Marking Information (Continued)**

**5-Lead DDPAK (Fixed) (MCP1791)**



**5-Lead SOT-223 (MCP1791)**











#### 3-Lead Plastic (EB) [DDPAK]

Note: r the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging





Notes:

1. § Significant Characteristic.

2. Dimensions D and E do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" per side.

3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-011B

#### 3-Lead Plastic Small Outline Transistor (DB) [SOT-223]

Note: r the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging











#### Notes:

- 1. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.127 mm per side.
- 2. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-032B

#### 3-Lead Plastic Small Outline Transistor (DB) [SOT-223]

Note: r the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging





#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2032A

#### 5-Lead Plastic (ET) [DDPAK]

Note: r the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging





#### Notes:

- 1. § Significant Characteristic.
- 2. Dimensions D and E do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" per side.
- 3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-012B