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3V Two-Channel Analog Front End

Features:

- Two Synchronous Sampling 24-bit Resolution Delta-Sigma A/D Converters
- 93.5 dB SINAD, -107 dBc Total Harmonic Distortion (THD) (up to 35th harmonic), 112 dB Spurious-Free Dynamic Range (SFDR) for Each Channel
- Flexible Serial Interface that Includes Both SPI and a Simple 2-Wire Interface Ideal for Polyphase Shunt Energy Meters
- Enables 0.1% Typical Active Power Measurement Error over a 10,000:1 Dynamic Range
- Advanced Security Features:
 - 16-bit Cyclic Redundancy Check (CRC) Checksum on All Communications for Secure Data Transfers
 - 16-bit Cyclic Redundancy Check (CRC) Checksum and Interrupt Alert for Register-Map Configuration
 - Register-Map Lock with 8-bit Secure Key
- 2.7V – 3.6V V_{DD} , DV_{DD}
- Programmable Data Rate, up to 125 ksps:
 - 4 MHz Maximum Sampling Frequency
 - 16 MHz Maximum Master Clock
- Oversampling Ratio, up to 4096
- Ultra Low-Power Shutdown Mode with < 10 μ A
- -122 dB Crosstalk between Channels
- Low Drift 1.2V Internal Voltage Reference: 9 ppm/ $^{\circ}$ C
- Differential Voltage Reference Input Pins
- High-Gain Programmable Gain Amplifier (PGA) on Each Channel (up to 32 V/V)
- Phase Delay Compensation with 1 μ s Time Resolution
- Separate Data Ready Pin for Easy Synchronization
- Individual 24-bit Digital Offset and Gain Error Correction for Each Channel
- High-Speed 20 MHz SPI Interface with Mode 0,0 and 1,1 Compatibility
- Continuous Read/Write Modes for Minimum Communication with Dedicated 16-/32-bit Modes
- Available in 20-lead QFN and SSOP Packages
- Extended Temperature Range: -40 $^{\circ}$ C to +125 $^{\circ}$ C (All Specifications are Valid Down to -45 $^{\circ}$ C Operation)

Description:

The MCP3910 is a 3V two-channel Analog Front End (AFE), containing two synchronous sampling delta-sigma, Analog-to-Digital Converters (ADC), two programmable gain amplifiers (PGA), phase delay compensation block, low-drift internal voltage reference, digital offset and gain errors calibration registers, and high-speed 20 MHz SPI-compatible serial interface.

The MCP3910 ADCs are fully configurable with features such as: 16-/24-bit resolution, Oversampling Ratio (OSR) from 32 to 4096, gain from 1x to 32x, independent Shutdown and Reset, dithering and auto-zeroing. Communication is largely simplified with 8-bit commands, including various continuous read/write modes and 16-/24-/32-bit data formats that can be accessed by the Direct Memory Access (DMA) of an 8-/16- or 32-bit MCU, and with the separate data ready pin that can directly be connected to an Interrupt Request (IRQ) input of an MCU.

The MCP3910 includes advanced security features to secure the communications and the configuration settings, such as a CRC-16 checksum on both serial data outputs and on the static register map configuration. It also includes a register-map lock through an 8-bit password to stop unwanted write commands from processing.

For polyphase shunt-based energy meters, the MCP3910 2-Wire serial interface greatly reduces system cost, requiring only a single bidirectional isolator per phase.

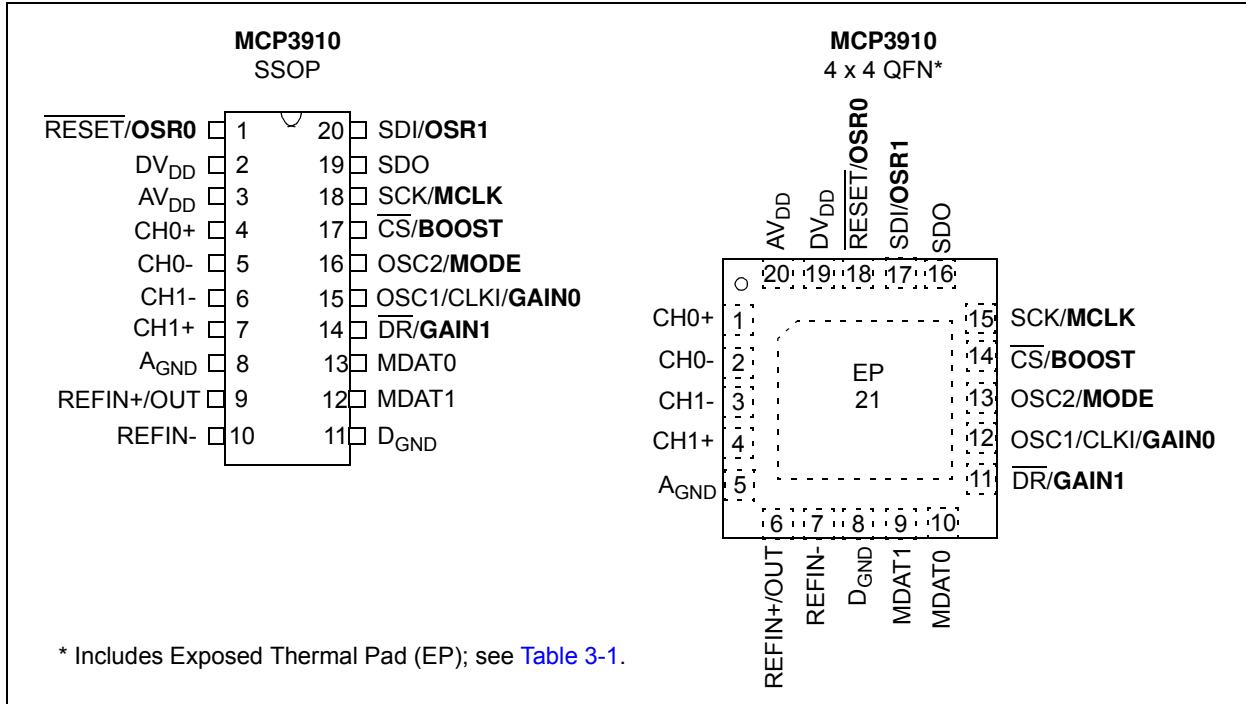
The MCP3910 is capable of interfacing a variety of voltage and current sensors, including shunts, current transformers, Rogowski coils and Hall-effect sensors.

Applications:

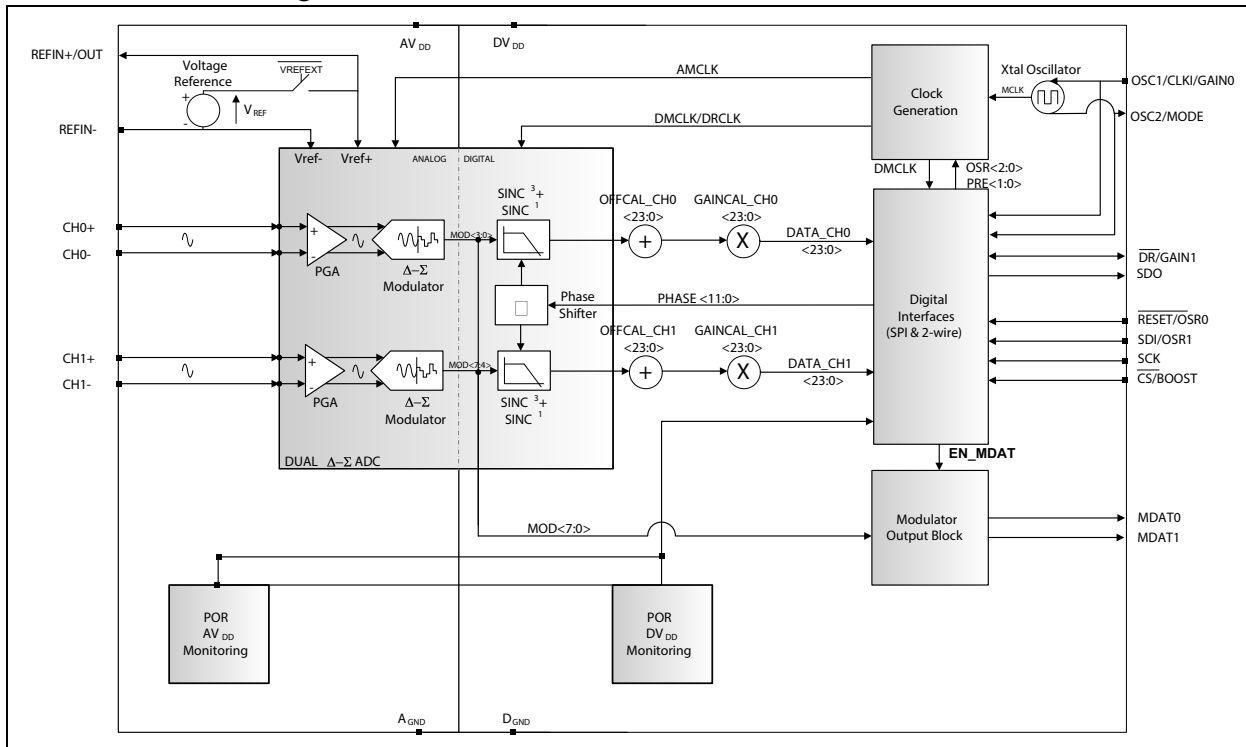
- Single-Phase and Polyphase Energy Meters
- Energy Metering and Power Measurement
- Automotive
- Portable Instrumentation
- Medical and Power Monitoring
- Audio/Voice Recognition
- Isolated Sensor Applications

MCP3910

Package Type



Functional Block Diagram



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

V_{DD}	-0.3V to 4.0V
Digital inputs and outputs w.r.t. A_{GND}	-0.3V to 4.0V
Analog input w.r.t. A_{GND}	-2V to +2V
V_{REF} input w.r.t. A_{GND}	-0.6V to $V_{DD} + 0.6V$
Storage temperature	-65°C to +150°C
Ambient temp. with power applied	-65°C to +125°C
Soldering temperature of leads (10 seconds)	+300°C
ESD on the analog inputs (HBM,MM).....	4.0 kV, 200V
ESD on all other pins (HBM,MM).....	4.0 kV, 200V

† **Notice:** Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions, above those indicated in the operational listings of this specification, is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

1.1 Electrical Specifications

TABLE 1-1: ANALOG SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, all parameters apply at $AV_{DD} = DV_{DD} = 3V$, $MCLK = 4$ MHz; $PRE<1:0> = 00$; $OSR = 256$; $GAIN = 1$; $VREFEXT = 0$, $CLKEXT = 1$, $DITHER<1:0> = 11$; $BOOST<1:0> = 10$, $V_{CM} = 0V$; $T_A = -40^{\circ}C$ to $+125^{\circ}C$ (Note 1); $V_{IN} = 1.2 V_{PP} = -0.5$ dBFS @ 50/60 Hz on all channels.						
Characteristic	Sym.	Min.	Typ.	Max.	Units	Conditions
ADC Performance						
Resolution (No Missing Codes)		24	—	—	bits	$OSR = 256$ or greater
Sampling Frequency	$f_S(DMCLK)$	—	1	4	MHz	For maximum condition, $BOOST<1:0> = 11$
Output Data Rate	$f_D(DRCLK)$	—	4	125	ksps	For maximum condition, $BOOST<1:0> = 11$, $OSR = 32$
Analog Input Absolute Voltage on $CHn+/-$, n between 0 and 1 pins	$CHn+/-$	-1	—	+1	V	All analog input channels, measured to A_{GND}
Analog Input Leakage Current	I_{IN}	—	+/-1	—	nA	$RESET<1:0> = 11$, $MCLK$ running continuously
Differential Input Voltage Range	$(CH_{n+} - CH_{n-})$	-600/GAIN	—	+600/GAIN	mV	$V_{REF} = 1.2V$, proportional to V_{REF}
Offset Error	V_{OS}	-1	0.2	1	mV	Note 5
Offset Error Drift		—	0.5	—	$\mu V/^{\circ}C$	
Gain Error	GE	-4	—	+4	%	Note 5

Note 1: All specifications are valid down to $-45^{\circ}C$ operation.

- This specification implies that the ADC output is valid over this entire differential range and that there is no distortion or instability across this input range. Dynamic Performance specified at -0.5 dB below the maximum signal range, $V_{IN} = 1.2 V_{PP} = 424$ mV_{RMS}, $V_{REF} = 1.2V$ @ 50/60 Hz. See [Section 4.0 “Terminology And Formulas”](#) for definition. This parameter is established by characterization and not 100% tested.
- For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 00$, $RESET<1:0> = 00$, $VREFEXT = 0$, $CLKEXT = 0$.
- For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 11$, $VREFEXT = 1$, $CLKEXT = 1$.
- Applies to all gains. Offset and gain errors depend on PGA gain setting. See [Section 2.0 “Typical Performance Curves”](#) for typical performance.
- Outside this range, the ADC accuracy is not specified. An extended input range of +/-2V can be applied continuously to the part, with no damage.
- For proper operation and for optimizing the ADC accuracy, $AMCLK$ should be limited to the maximum frequency defined in [Table 5-2](#), as a function of the $BOOST$ and PGA setting chosen. $MCLK$ can take larger values as long as the prescaler settings ($PRE<1:0>$) limit $AMCLK = MCLK/PRESCALE$ within the defined range in [Table 5-2](#).

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TABLE 1-1: ANALOG SPECIFICATIONS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all parameters apply at $V_{DD} = DV_{DD} = 3V$, $MCLK = 4\text{ MHz}$; $PRE<1:0> = 00$; $OSR = 256$; $GAIN = 1$; $VREFEXT = 0$, $CLKEXT = 1$, $DITHER<1:0> = 11$; $BOOST<1:0> = 10$, $V_{CM} = 0V$; $T_A = -40^{\circ}C$ to $+125^{\circ}C$ (Note 1); $V_{IN} = 1.2\text{ V}_{PP} = -0.5\text{ dBFS}$ @ 50/60 Hz on all channels.						
Characteristic	Sym.	Min.	Typ.	Max.	Units	Conditions
Gain Error Drift		—	1	—	ppm/ $^{\circ}C$	
Integral Non-Linearity	INL	—	5	—	ppm	
Measurement Error	ME	—	0.1	—	%	Measured with a 10,000:1 dynamic range (from 600 mV _{Peak} to 60 μ V _{Peak}), $V_{DD} = DV_{DD} = 3V$, measurement points averaging time: 20 seconds
Differential Input Impedance	Z_{IN}	232	—	—	k Ω	G = 1, proportional to 1/AMCLK
		142	—	—	k Ω	G = 2, proportional to 1/AMCLK
		72	—	—	k Ω	G = 4, proportional to 1/AMCLK
		38	—	—	k Ω	G = 8, proportional to 1/AMCLK
		36	—	—	k Ω	G = 16, proportional to 1/AMCLK
		33	—	—	k Ω	G = 32, proportional to 1/AMCLK
Signal-to-Noise and Distortion Ratio (Note 2)	SINAD	92	93.5	—	dB	
Total Harmonic Distortion (Note 2)	THD	—	-107	-103	dBc	Includes the first 35 harmonics
Signal-to-Noise Ratio (Note 2)	SNR	92	94	—	dB	
Spurious-Free Dynamic Range (Note 2)	SFDR	—	112	—	dBFS	
Crosstalk (50, 60 Hz)	CTALK	—	-122	—	dB	
AC Power Supply Rejection	AC PSRR	—	-73	—	dB	$V_{DD} = DV_{DD} = 3V + 0.6V_{PP}$ 50/60 Hz, 100/120 Hz
DC Power Supply Rejection	DC PSRR	—	-73	—	dB	$V_{DD} = DV_{DD} = 2.7V$ to $3.6V$
DC Common Mode Rejection	DC CMRR	—	-105	—	dB	V_{CM} from -1V to +1V

Note 1: All specifications are valid down to $-45^{\circ}C$ operation.

- 2:** This specification implies that the ADC output is valid over this entire differential range and that there is no distortion or instability across this input range. Dynamic Performance specified at -0.5 dB below the maximum signal range, $V_{IN} = 1.2\text{ V}_{PP} = 424\text{ mV}_{RMS}$, $V_{REF} = 1.2V$ @ 50/60 Hz. See [Section 4.0 "Terminology And Formulas"](#) for definition. This parameter is established by characterization and not 100% tested.
- 3:** For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 00$, $RESET<1:0> = 00$, $VREFEXT = 0$, $CLKEXT = 0$.
- 4:** For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 11$, $VREFEXT = 1$, $CLKEXT = 1$.
- 5:** Applies to all gains. Offset and gain errors depend on PGA gain setting. See [Section 2.0 "Typical Performance Curves"](#) for typical performance.
- 6:** Outside this range, the ADC accuracy is not specified. An extended input range of $\pm 2V$ can be applied continuously to the part, with no damage.
- 7:** For proper operation and for optimizing the ADC accuracy, AMCLK should be limited to the maximum frequency defined in [Table 5-2](#), as a function of the BOOST and PGA setting chosen. MCLK can take larger values as long as the prescaler settings ($PRE<1:0>$) limit $AMCLK = MCLK/PRESCALE$ within the defined range in [Table 5-2](#).

TABLE 1-1: ANALOG SPECIFICATIONS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all parameters apply at $AV_{DD} = DV_{DD} = 3V$, $MCLK = 4\text{ MHz}$; $PRE<1:0> = 00$; $OSR = 256$; $GAIN = 1$; $VREFEXT = 0$, $CLKEXT = 1$, $DITHER<1:0> = 11$; $BOOST<1:0> = 10$, $V_{CM} = 0V$; $T_A = -40^{\circ}C$ to $+125^{\circ}C$ (Note 1); $V_{IN} = 1.2 V_{PP} = -0.5\text{ dBFS @ } 50/60\text{ Hz}$ on all channels.						
Characteristic	Sym.	Min.	Typ.	Max.	Units	Conditions
Internal Voltage Reference						
Tolerance	V_{REF}	1.176	1.2	1.224	V	$VREFEXT = 0$, $T_A = +25^{\circ}C$ only
Temperature Coefficient	TCV_{REF}	—	9	—	ppm/ $^{\circ}C$	$T_A = -45^{\circ}C$ to $+125^{\circ}C$, $VREFEXT = 0$
Output Impedance	$ZOUTV_{REF}$	—	0.6	—	$k\Omega$	$VREFEXT = 0$
Internal Voltage Reference Operating Current	$AI_{DD}V_{REF}$	—	54	—	μA	$VREFEXT = 0$, $SHUTDOWN<1:0> = 11$
Voltage Reference Input						
Input Capacitance		—	—	10	pF	
Differential Input Voltage Range ($V_{REF+} - V_{REF-}$)	V_{REF}	1.1	—	1.3	V	$VREFEXT = 1$
Absolute Voltage on $REFIN+$ pin	V_{REF+}	$V_{REF-} + 1.1$	—	$V_{REF-} + 1.3$	V	$VREFEXT = 1$
Absolute Voltage $REFIN-$ pin	V_{REF-}	-0.1	—	+0.1	V	$REFIN-$ should be connected to A_{GND} when $VREFEXT = 0$
Master Clock Input						
Master Clock Input Frequency Range	f_{MCLK}	—	—	20	MHz	$CLKEXT = 1$ (Note 7)
Crystal Oscillator Operating Frequency Range	f_{XTAL}	1	—	20	MHz	$CLKEXT = 0$ (Note 7)
Analog Master Clock	AMCLK	—	—	16	MHz	Note 7
Crystal Oscillator Operating Current	DIDDXTAL	—	80	—	μA	$CLKEXT = 0$
Power Supply						
Operating Voltage, Analog	AV_{DD}	2.7	—	3.6	V	
Operating Voltage, Digital	DV_{DD}	2.7	—	3.6	V	
Operating Current, Analog (Note 3)	$I_{DD,A}$	—	1.5	1.95	mA	$BOOST<1:0> = 00$
		—	1.8	2.3	mA	$BOOST<1:0> = 01$
		—	2.5	3.2	mA	$BOOST<1:0> = 10$
		—	4.4	5.5	mA	$BOOST<1:0> = 11$

Note 1: All specifications are valid down to $-45^{\circ}C$ operation.

- 2:** This specification implies that the ADC output is valid over this entire differential range and that there is no distortion or instability across this input range. Dynamic Performance specified at -0.5 dB below the maximum signal range, $V_{IN} = 1.2 V_{PP} = 424\text{ mV}_{RMS}$, $V_{REF} = 1.2V @ 50/60\text{ Hz}$. See **Section 4.0 “Terminology And Formulas”** for definition. This parameter is established by characterization and not 100% tested.
- 3:** For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 00$, $RESET<1:0> = 00$, $VREFEXT = 0$, $CLKEXT = 0$.
- 4:** For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 11$, $VREFEXT = 1$, $CLKEXT = 1$.
- 5:** Applies to all gains. Offset and gain errors depend on PGA gain setting. See **Section 2.0 “Typical Performance Curves”** for typical performance.
- 6:** Outside this range, the ADC accuracy is not specified. An extended input range of $\pm 2V$ can be applied continuously to the part, with no damage.
- 7:** For proper operation and for optimizing the ADC accuracy, AMCLK should be limited to the maximum frequency defined in **Table 5-2**, as a function of the BOOST and PGA setting chosen. MCLK can take larger values as long as the prescaler settings ($PRE<1:0>$) limit $AMCLK = MCLK/PRESCALE$ within the defined range in **Table 5-2**.

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TABLE 1-1: ANALOG SPECIFICATIONS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all parameters apply at $AV_{DD} = DV_{DD} = 3V$, $MCLK = 4\text{ MHz}$; $PRE<1:0> = 00$; $OSR = 256$; $GAIN = 1$; $VREFEXT = 0$, $CLKEXT = 1$, $DITHER<1:0> = 11$; $BOOST<1:0> = 10$, $V_{CM} = 0V$; $T_A = -40^{\circ}C$ to $+125^{\circ}C$ (Note 1); $V_{IN} = 1.2\text{ V}_{PP} = -0.5\text{ dBFS}$ @ 50/60 Hz on all channels.						
Characteristic	Sym.	Min.	Typ.	Max.	Units	Conditions
Operating Current, Digital	$I_{DD,D}$	—	0.2	0.4	mA	$MCLK = 4\text{ MHz}$, proportional to $MCLK$
		—	0.7	—	mA	$MCLK = 16\text{ MHz}$, proportional to $MCLK$
Shutdown Current, Analog	$I_{DDS,A}$	—	—	1	μA	AV_{DD} pin only (Note 4)
Shutdown Current, Digital	$I_{DDS,D}$	—	—	2	μA	DV_{DD} pin only (Note 4)
Pull-down Current on OSC2 Pin (External Clock Mode)	I_{OSC2}	—	35	—	μA	$CLKEXT = 1$

Note 1: All specifications are valid down to $-45^{\circ}C$ operation.

- 2:** This specification implies that the ADC output is valid over this entire differential range and that there is no distortion or instability across this input range. Dynamic Performance specified at -0.5 dB below the maximum signal range, $V_{IN} = 1.2\text{ V}_{PP} = 424\text{ mV}_{RMS}$, $V_{REF} = 1.2V$ @ 50/60 Hz. See **Section 4.0 “Terminology And Formulas”** for definition. This parameter is established by characterization and not 100% tested.
- 3:** For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 00$, $RESET<1:0> = 00$, $VREFEXT = 0$, $CLKEXT = 0$.
- 4:** For these operating currents, the following configuration bit settings apply: $SHUTDOWN<1:0> = 11$, $VREFEXT = 1$, $CLKEXT = 1$.
- 5:** Applies to all gains. Offset and gain errors depend on PGA gain setting. See **Section 2.0 “Typical Performance Curves”** for typical performance.
- 6:** Outside this range, the ADC accuracy is not specified. An extended input range of $\pm 2V$ can be applied continuously to the part, with no damage.
- 7:** For proper operation and for optimizing the ADC accuracy, $AMCLK$ should be limited to the maximum frequency defined in **Table 5-2**, as a function of the $BOOST$ and PGA setting chosen. $MCLK$ can take larger values as long as the prescaler settings ($PRE<1:0>$) limit $AMCLK = MCLK/PRESCALE$ within the defined range in **Table 5-2**.

1.2 Serial Interface Characteristics

TABLE 1-2: SERIAL DC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all parameters apply at $DV_{DD} = 2.7$ to 3.6 V , $T_A = -40^{\circ}C$ to $+125^{\circ}C$ (Note 1), $C_{LOAD} = 30\text{ pF}$, applies to all digital I/O.						
Characteristic	Sym.	Min.	Typ.	Max.	Units	Conditions
High-Level Input Voltage	V_{IH}	0.7 DV_{DD}	—	—	V	Schmitt-Triggered
Low-Level Input Voltage	V_{IL}	—	—	0.3 DV_{DD}	V	Schmitt-Triggered
Input Leakage Current	I_{LI}	—	—	± 1	μA	$\overline{CS} = DV_{DD}$, $V_{IN} = D_{GND}$ to DV_{DD}
Output Leakage Current	I_{LO}	—	—	± 1	μA	$\overline{CS} = DV_{DD}$, $V_{OUT} = D_{GND}$ or DV_{DD}
Hysteresis Of Schmitt-Trigger Inputs	V_{HYS}	—	300	—	mV	Note 3 , $DV_{DD} = 3.3V$ only
Low-Level Output Voltage	V_{OL}	—	—	0.4	V	$I_{OL} = +1.7\text{ mA}$, $DV_{DD} = 3.3V$
High-Level Output Voltage	V_{OH}	$DV_{DD} - 0.5$	—	—	V	$I_{OH} = -1.7\text{ mA}$, $DV_{DD} = 3.3V$
Internal Capacitance (All Inputs and Outputs)	C_{INT}	—	—	7	pF	$T_A = +25^{\circ}C$, $SCK = 1.0\text{ MHz}$, $DV_{DD} = 3.3V$ (Note 2)

Note 1: All specifications are valid down to $-45^{\circ}C$ operation.

2: This parameter is periodically sampled and not 100% tested.

3: This parameter is established by characterization and not production tested.

TABLE 1-3: SERIAL AC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all parameters apply at $DV_{DD} = 2.7$ to 3.6 V, $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$, $GAIN = 1$, $C_{LOAD} = 30$ pF.

Characteristic	Sym.	Min.	Typ.	Max.	Units	Conditions
Serial Clock Frequency	f_{SCK}	—	—	20	MHz	
\overline{CS} Setup Time	t_{CSS}	25	—	—	ns	
\overline{CS} Hold Time	t_{CSH}	50	—	—	ns	
\overline{CS} Disable Time	t_{CSD}	50	—	—	ns	
Data Setup Time	t_{SU}	5	—	—	ns	
Data Hold Time	t_{HD}	10	—	—	ns	
Serial Clock High Time	t_{HI}	20	—	—	ns	
Serial Clock Low Time	t_{LO}	20	—	—	ns	
Serial Clock Delay Time	t_{CLD}	50	—	—	ns	
Serial Clock Enable Time	t_{CLE}	50	—	—	ns	
Output Valid from SCK Low	t_{DO}	—	—	25	ns	
Output Hold Time	t_{HO}	0	—	—	ns	
Output Disable Time	t_{DIS}	—	—	25	ns	
Reset Pulse Width (\overline{RESET})	t_{MCLR}	100	—	—	ns	
Data Transfer Time to \overline{DR} (Data Ready)	t_{DODR}	—	—	25	ns	Note 2
Modulator Mode Entry to Modulator Data Present	t_{MODSU}	—	—	100	ns	
Data Ready Pulse Low Time	t_{DRP}	—	$1/(2 \times DMCLK)$	—	μs	
2-Wire Mode Enable Time	t_{MODE}	—	—	50	ns	
2-Wire Mode Watchdog Timer	t_{WATCH}	3.6	—	35	μs	

Note 1: This parameter is periodically sampled and not 100% tested.

2: This parameter is established by characterization and not production tested.

TABLE 1-4: TEMPERATURE SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, all parameters apply at $AV_{DD} = 2.7$ to 3.6 V, $DV_{DD} = 2.7$ to 3.6 V.

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Temperature Ranges						
Operating Temperature Range	T_A	-40	—	+125	$^\circ\text{C}$	Note 1, Note 2
Storage Temperature Range	T_A	-65	—	+150	$^\circ\text{C}$	
Thermal Package Resistances						
Thermal Resistance, 20L 4x4 QFN	θ_{JA}	—	46.2	—	$^\circ\text{C/W}$	
Thermal Resistance, 20L SSOP	θ_{JA}	—	87.3	—	$^\circ\text{C/W}$	

Note 1: The internal junction temperature (T_J) must not exceed the absolute maximum specification of $+150^\circ\text{C}$.

2: All specifications are valid down to -45°C operation.

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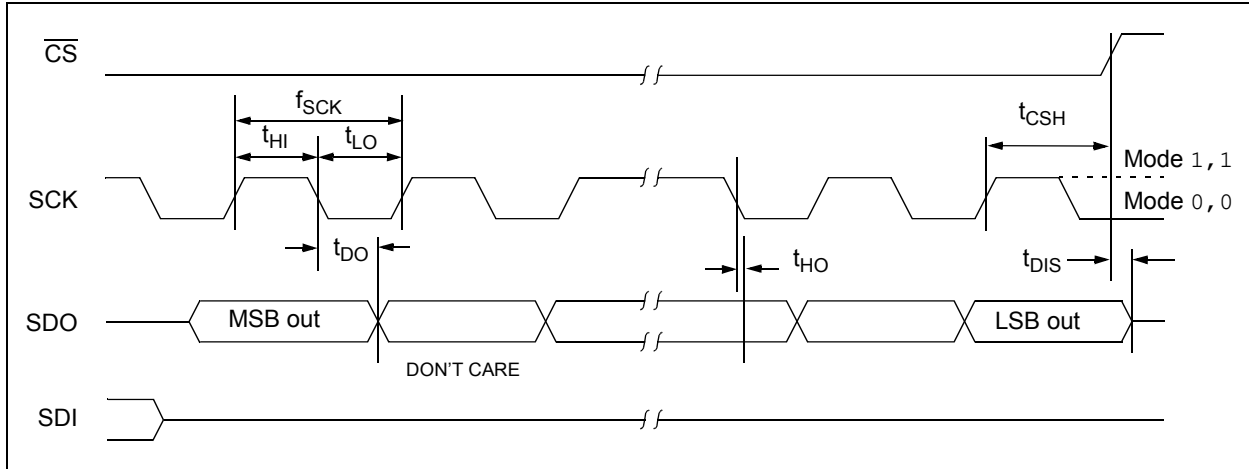


FIGURE 1-1: Serial Output Timing Diagram.

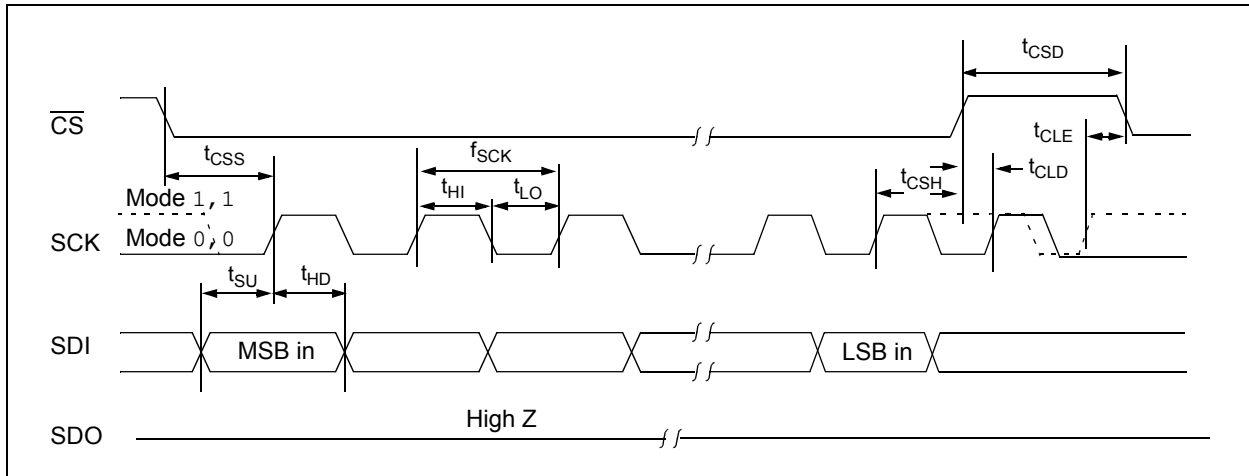


FIGURE 1-2: Serial Input Timing Diagram.

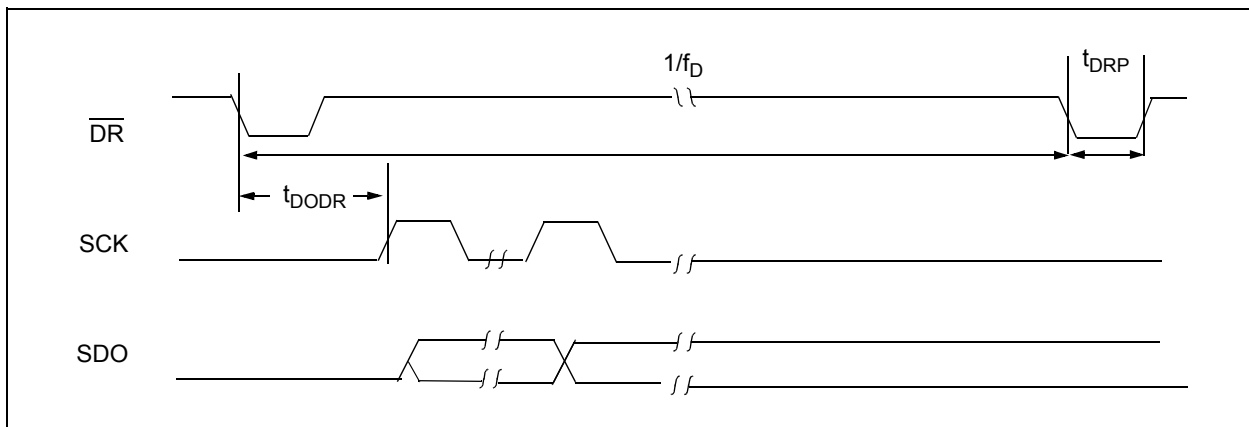


FIGURE 1-3: Data Ready Pulse/Sampling Timing Diagram.

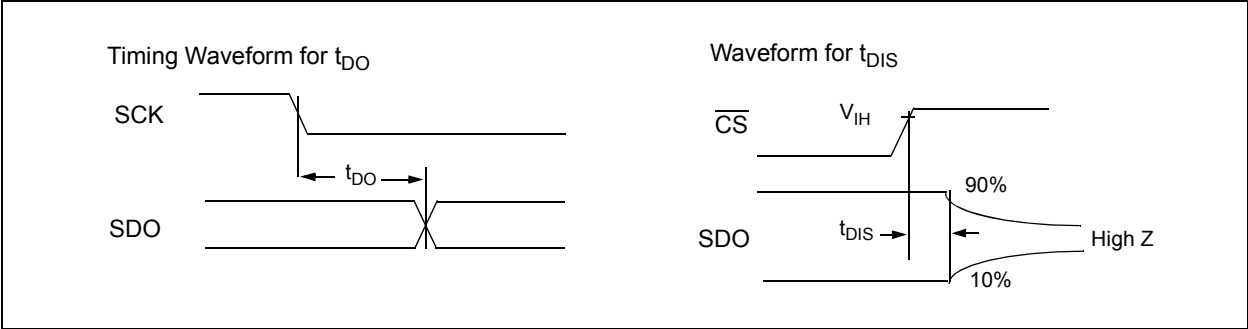


FIGURE 1-4: Timing Diagrams, Continued.

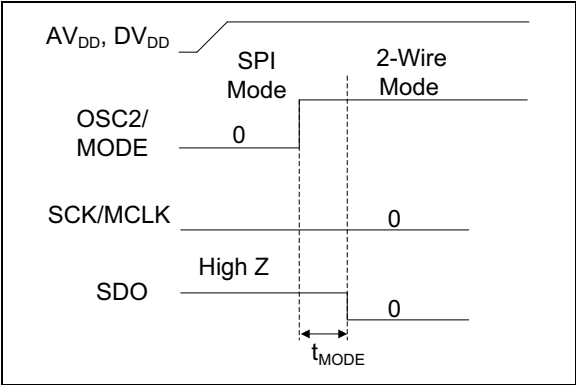


FIGURE 1-5: Entering 2-Wire Interface Mode Timing Diagram.

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NOTES:

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; $Dithering = Maximum$; $V_{IN} = -0.5\text{ dBFS @ }60\text{ Hz}$ on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

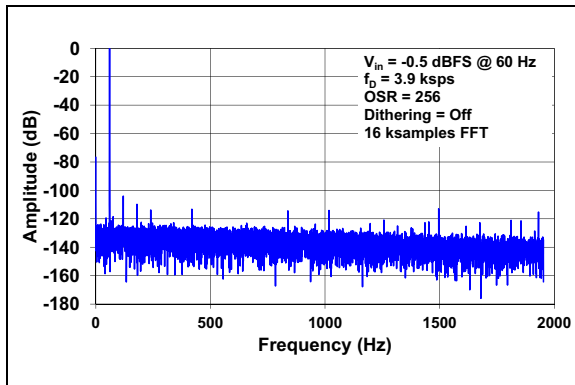


FIGURE 2-1: Spectral Response.

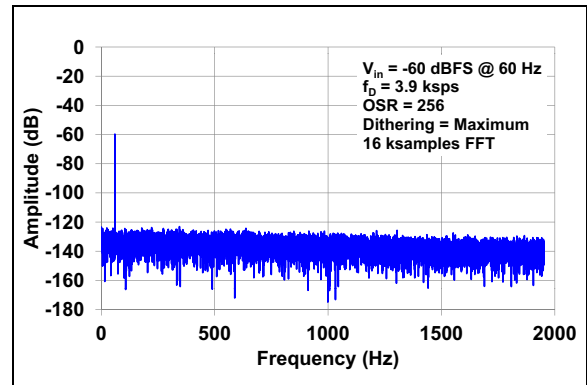


FIGURE 2-4: Spectral Response.

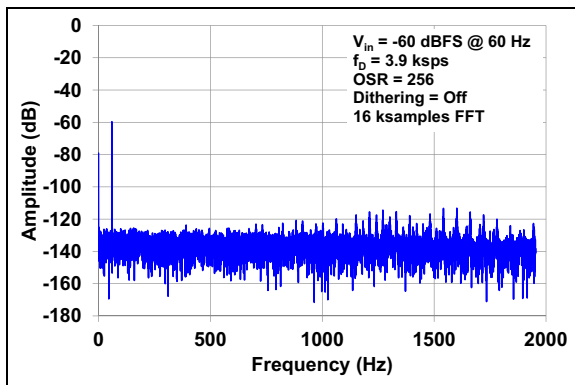


FIGURE 2-2: Spectral Response.

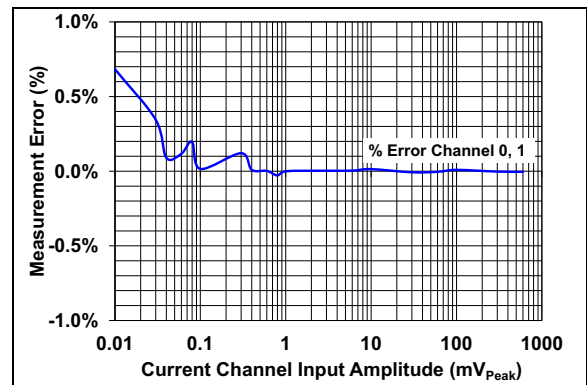


FIGURE 2-5: Measurement Error with 1-Point Calibration.

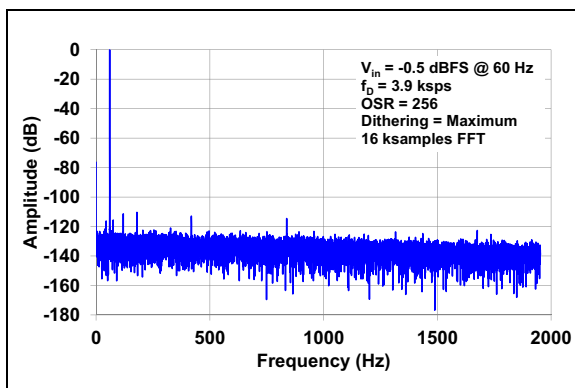


FIGURE 2-3: Spectral Response.

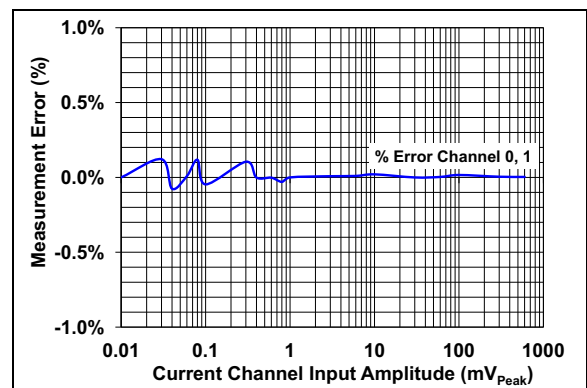


FIGURE 2-6: Measurement Error with 2-Point Calibration.

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Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; $Dithering = \text{Maximum}$; $V_{IN} = -0.5\text{ dBFS}$ @ 60 Hz on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

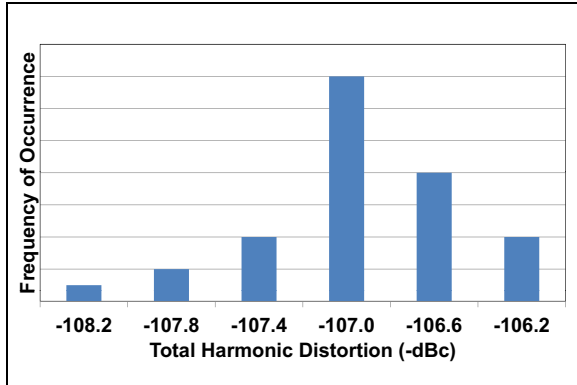


FIGURE 2-7: THD Repeatability Histogram.

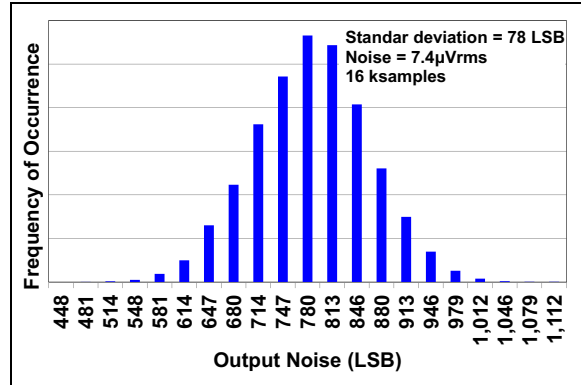


FIGURE 2-10: Output Noise Histogram.

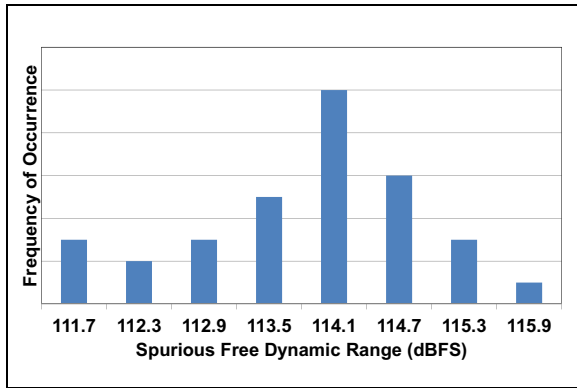


FIGURE 2-8: Spurious-Free Dynamic Range Repeatability Histogram.

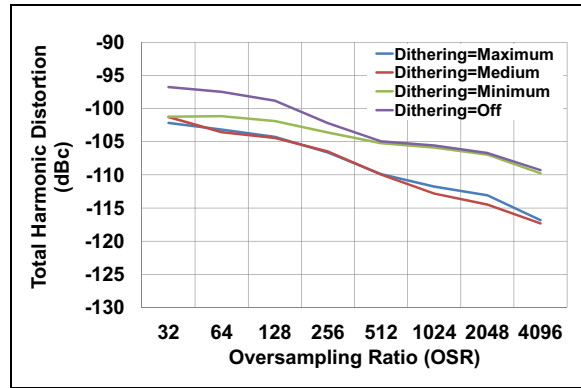


FIGURE 2-11: THD vs. OSR.

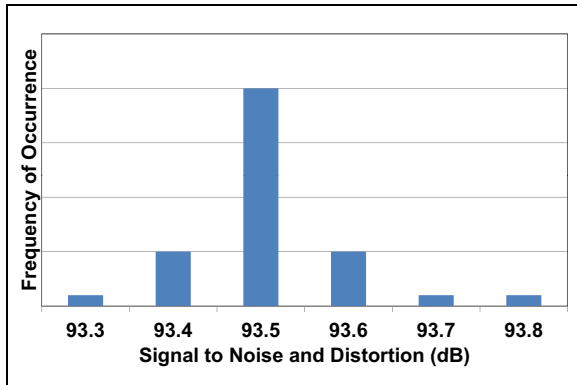


FIGURE 2-9: SINAD Repeatability Histogram.

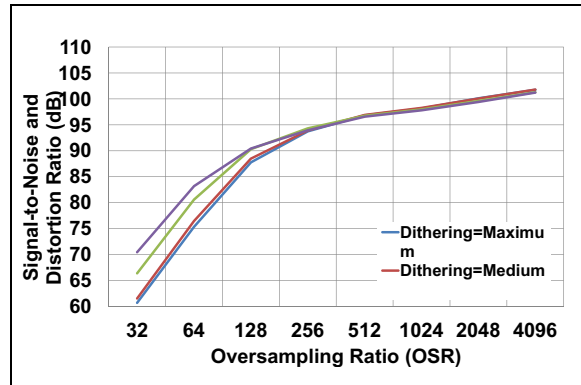


FIGURE 2-12: SINAD vs. OSR.

Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; $Dithering = \text{Maximum}$; $V_{IN} = -0.5\text{ dBFS}$ @ 60 Hz on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

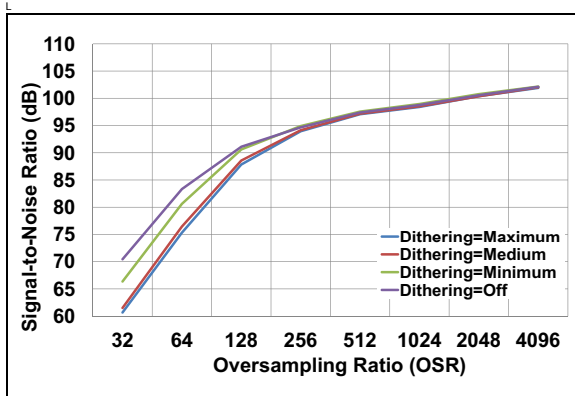


FIGURE 2-13: SNR vs. OSR.

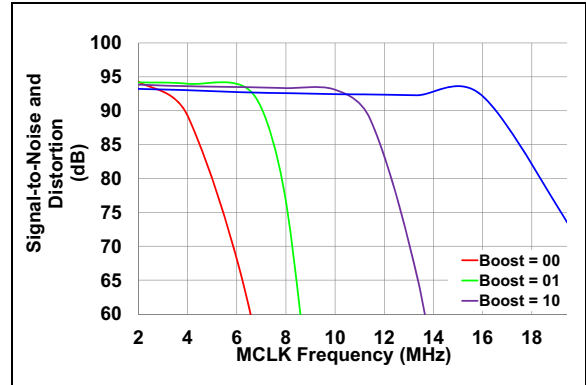


FIGURE 2-16: SINAD vs. MCLK.

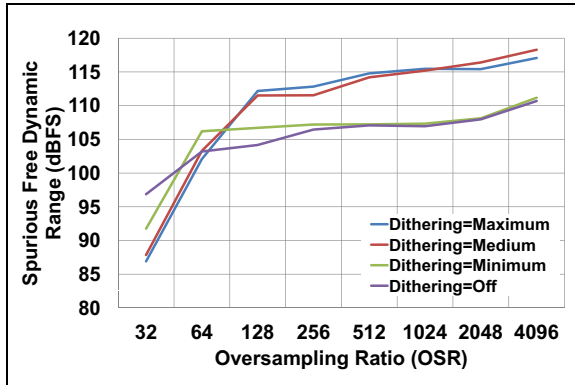


FIGURE 2-14: SFDR vs. OSR.

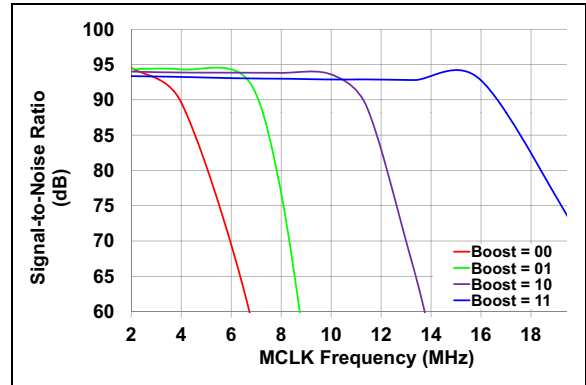


FIGURE 2-17: SNR vs. MCLK.

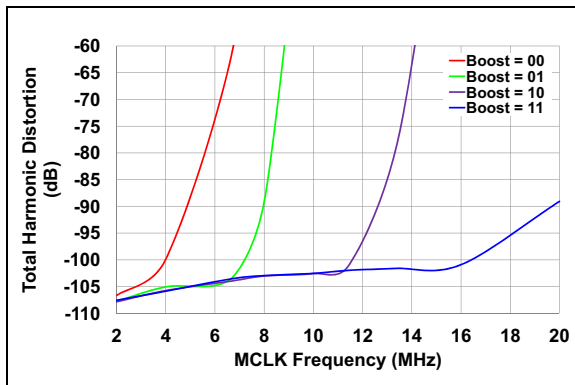


FIGURE 2-15: THD vs. MCLK.

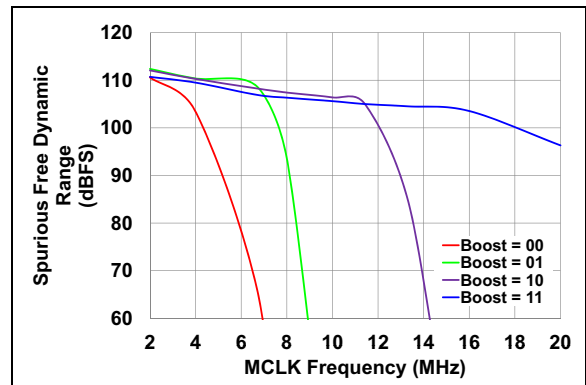


FIGURE 2-18: SFDR vs. MCLK.

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Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; Dithering = Maximum; $V_{IN} = -0.5\text{ dBFS}$ @ 60 Hz on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

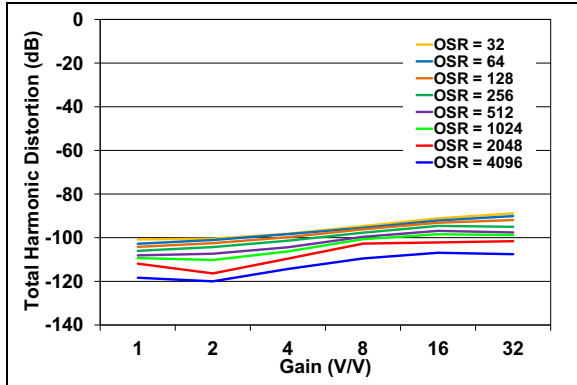


FIGURE 2-19: THD vs. GAIN.

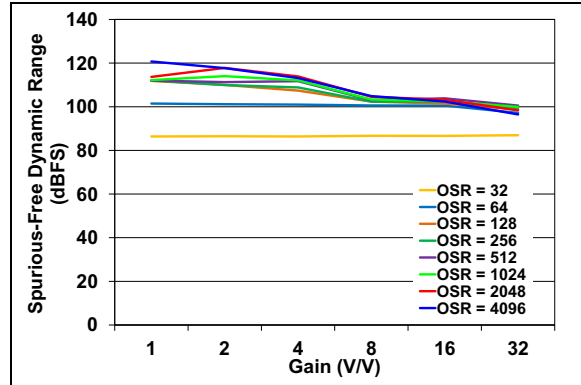


FIGURE 2-22: SFDR vs. GAIN.

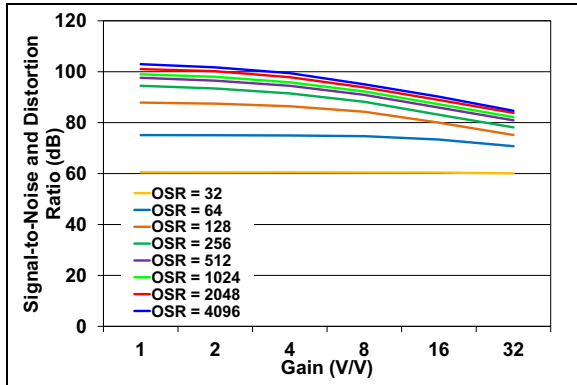


FIGURE 2-20: SINAD vs. GAIN.

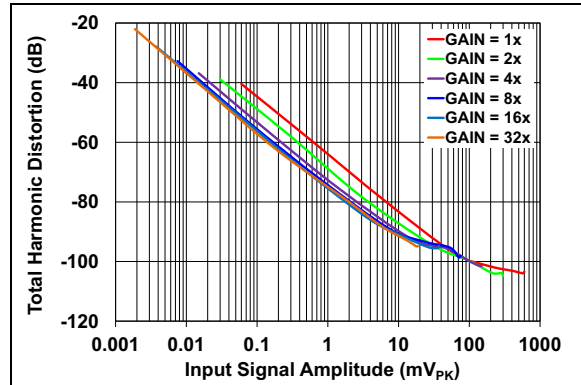


FIGURE 2-23: THD vs. Input Signal Amplitude.

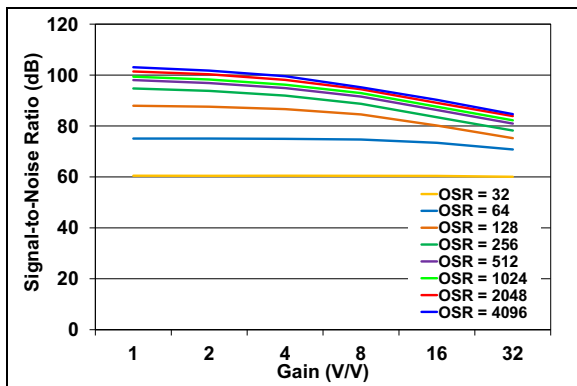


FIGURE 2-21: SNR vs. GAIN.

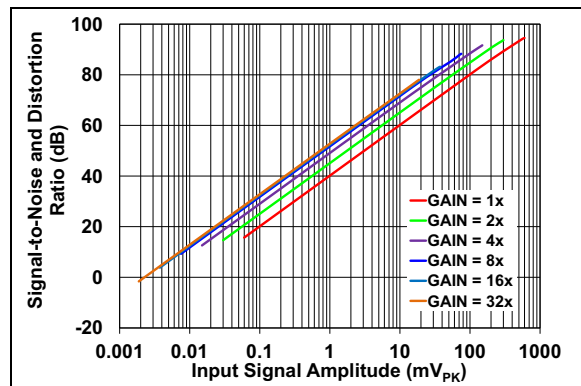


FIGURE 2-24: SINAD vs. Input Signal Amplitude.

Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; $Dithering = \text{Maximum}$; $V_{IN} = -0.5\text{ dBFS}$ @ 60 Hz on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

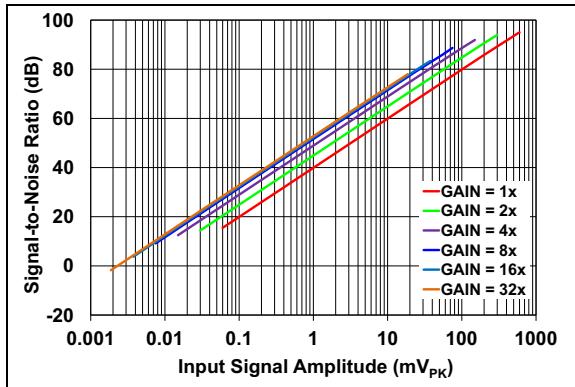


FIGURE 2-25: SNR vs. Input Signal Amplitude.

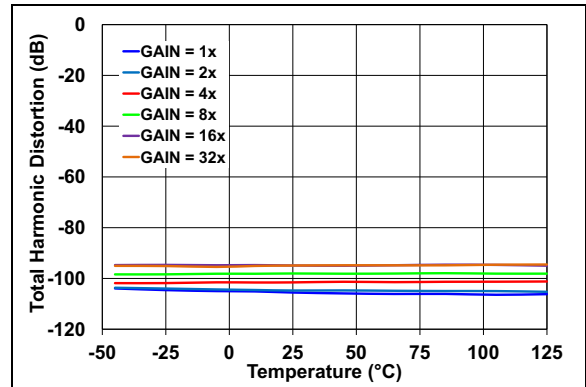


FIGURE 2-28: THD vs. Temperature.

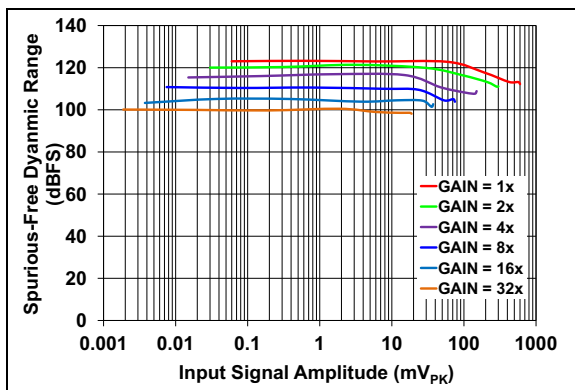


FIGURE 2-26: SFDR vs. Input Signal Amplitude.

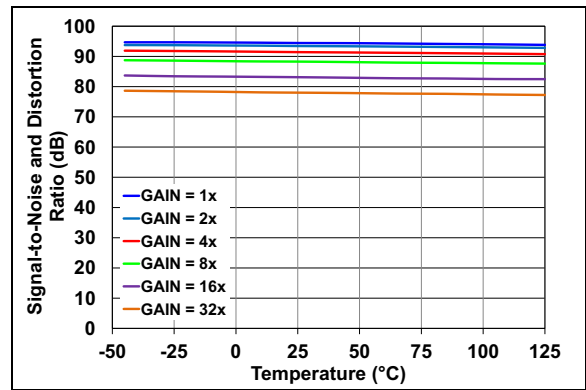


FIGURE 2-29: SINAD vs. Temperature.

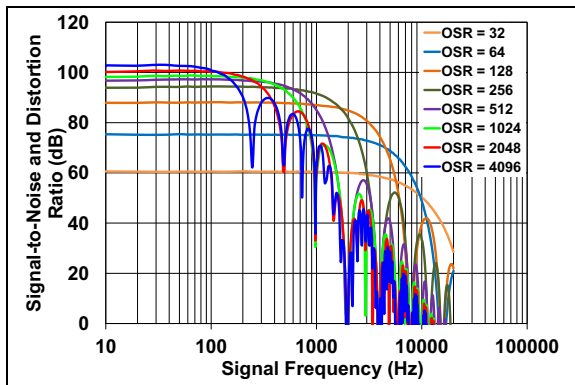


FIGURE 2-27: SINAD vs. Input Frequency.

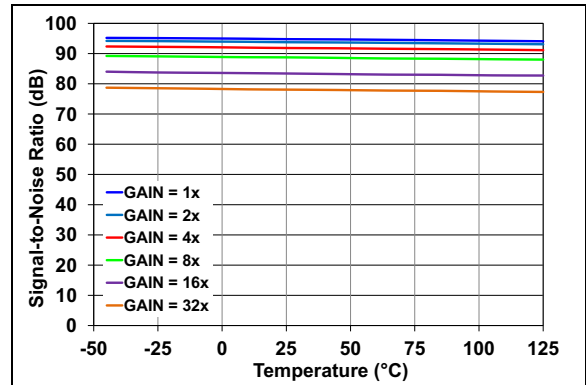


FIGURE 2-30: SNR vs. Temperature.

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Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; $Dithering = \text{Maximum}$; $V_{IN} = -0.5\text{ dBFS}$ @ 60 Hz on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

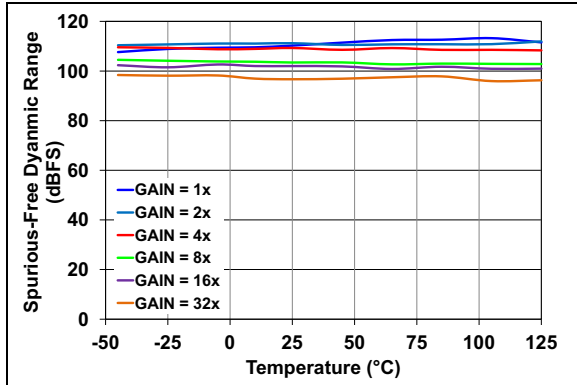


FIGURE 2-31: SFDR vs. Temperature.

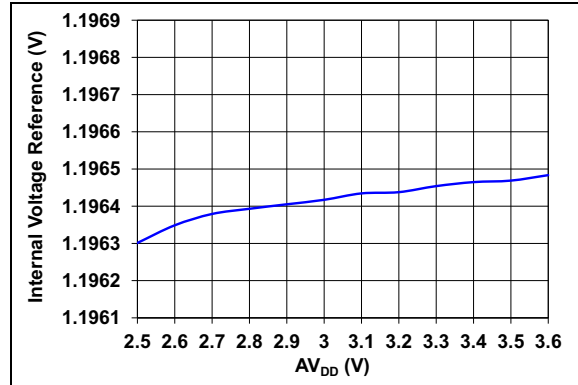


FIGURE 2-34: Internal Voltage Reference vs. Supply Voltage.

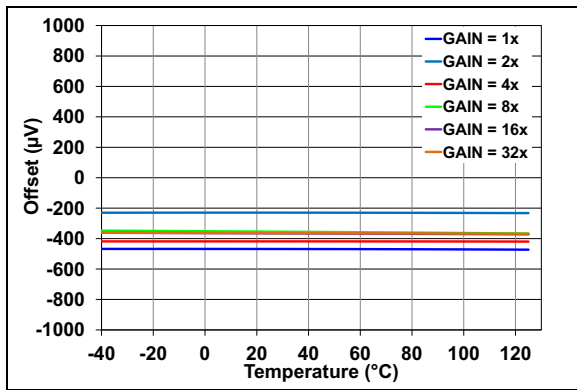


FIGURE 2-32: Offset vs. Temperature vs. Gain.

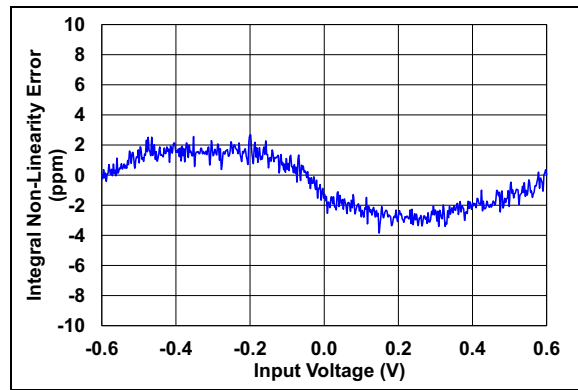


FIGURE 2-35: Integral Non-Linearity (Dithering Maximum).

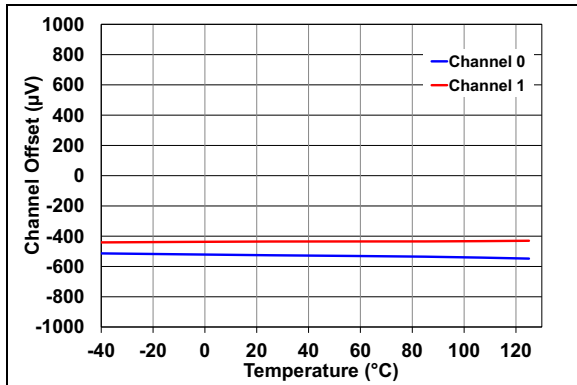


FIGURE 2-33: Channel Offset Matching vs. Temperature.

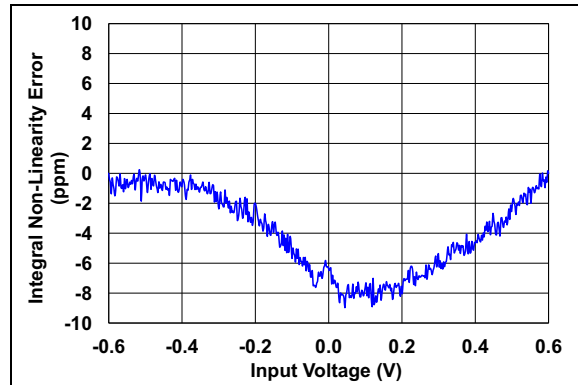


FIGURE 2-36: Integral Non-Linearity (Dithering Off).

Note: Unless otherwise indicated, $AV_{DD} = 3V$, $DV_{DD} = 3V$; $T_A = +25^\circ C$, $MCLK = 4\text{ MHz}$; $PRESCALE = 1$; $OSR = 256$; $GAIN = 1$; $Dithering = \text{Maximum}$; $V_{IN} = -0.5\text{ dBFS @ } 60\text{ Hz}$ on all channels, $VREFEXT = 0$; $CLKEXT = 1$; $BOOST<1:0> = 10$.

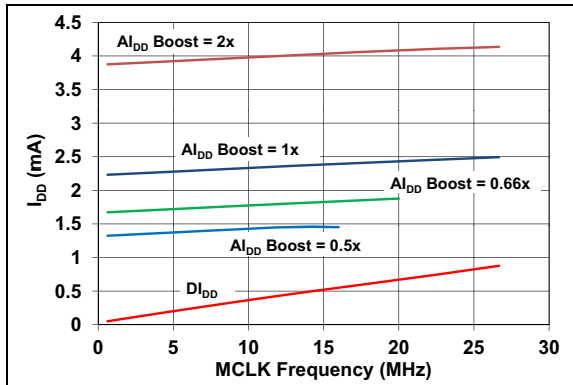


FIGURE 2-37: Operating Current vs. MCLK Frequency vs. Boost, $V_{DD} = 3.0V$.

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NOTES:

3.0 PIN DESCRIPTION

The descriptions of the pins are listed in [Table 3-1](#) for and [Table 2-2](#) for.

TABLE 3-1: TWO CHANNEL MCP3910 PIN FUNCTION TABLE

MCP3910 SSOP	MCP3910 QFN	Symbol	Function
1	18	$\overline{\text{RESET}}/\text{OSR0}$	Master Reset Logic Input Pin or OSR0 Logic Input Pin
2	19	DV_{DD}	Digital Power Supply Pin
3	20	AV_{DD}	Analog Power Supply Pin
4	1	CH0+	Non-Inverting Analog Input Pin for Channel 0
5	2	CH0-	Inverting Analog Input Pin for Channel 0
6	3	CH1-	Inverting Analog Input Pin for Channel 1
7	4	CH1+	Non-Inverting Analog Input Pin for Channel 1
8	5	A_{GND}	Analog Ground Pin, Return Path for internal analog circuitry
9	6	REFIN+/OUT	Non-Inverting Voltage Reference Input and Internal Reference Output Pin
10	7	REFIN-	Inverting Voltage Reference Input Pin
11	8	D_{GND}	Digital Ground Pin, Return Path for internal digital circuitry
12	9	MDAT1	Modulator Data Output Pin for Channel 1
13	10	MDAT0	Modulator Data Output Pin for Channel 0
14	11	$\overline{\text{DR}}/\text{GAIN1}$	Data Ready Signal Output Pin or GAIN1 Logic Input Pin
15	12	OSC1/CLKI/GAIN0	Oscillator Crystal Connection Pin or External Clock Input Pin or GAIN0 Logic Input Pin
16	13	OSC2/MODE	Oscillator Crystal Connection Pin or Serial Interface Mode Logic Input Pin
17	14	$\overline{\text{CS}}/\text{BOOST}$	Serial Interface Chip Select Input Pin or BOOST Logic Input Pin
18	15	SCK/MCLK	Serial Interface Clock Input Pin or Master Clock Input Pin
19	16	SDO	Serial Interface Data Input Pin
20	17	SDI/OSR1	Serial Interface Data Input Pin or OSR1 Logic Input Pin
—	21	EP	Exposed Thermal Pad.

3.1 Master Reset/OSR0 Logic Input (RESET/OSR0)

In SPI mode, this pin is active low and places the entire chip in a Reset state when active.

When $\overline{\text{RESET}}$ is logic low, all registers are reset to their default value, no communication can take place, no clock is distributed inside the part, except in the input structure if MCLK is applied (if MCLK is idle, then no clock is distributed). This state is equivalent to a Power-On Reset (POR) state.

Since the default state of the ADCs is on, the analog power consumption when $\overline{\text{RESET}}$ is logic low is equivalent to when RESET is logic high. Only the digital power consumption is largely reduced because this current consumption is essentially dynamic and is reduced drastically when there is no clock running.

All the analog biases are enabled during a Reset, so that the part is fully operational just after a $\overline{\text{RESET}}$ rising edge, if MCLK is applied when $\overline{\text{RESET}}$ is logic low. If MCLK is not applied, there is a time after a hard reset when the conversion may not accurately correspond to the start-up of the input structure.

This input is Schmitt-triggered.

In 2-Wire Interface mode, this is the OSR0 logic select pin (see [Section 7.0 “2-Wire Serial Interface Description”](#) for the logic input table for OSR0 and OSR1). The pin state is latched when the MODE changes to 2-Wire Interface mode, and is relatched at each watchdog timer reset.

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3.2 Digital V_{DD} (DV_{DD})

DV_{DD} is the power supply voltage for the digital circuitry within the MCP3910. For optimal performance, it is recommended to connect appropriate bypass capacitors (typically a 10 μF in parallel with a 0.1 μF ceramic). DV_{DD} should be maintained between 2.7V and 3.6V for specified operation.

3.3 Analog V_{DD} (AV_{DD})

AV_{DD} is the power supply voltage for the analog circuitry within the MCP3910. For optimal performance, it is recommended to connect appropriate bypass capacitors (typically a 10 μF in parallel with a 0.1 μF ceramic). AV_{DD} should be maintained between 2.7V and 3.6V for specified operation.

3.4 ADC Differential Analog Inputs ($CHn+/CHn-$)

The $CHn\pm$ pins (n comprised between 0 and 1) are the two fully differential analog voltage inputs for the Delta-Sigma ADCs.

The linear and specified region of the channels are dependent on the PGA gain. This region corresponds to a differential voltage range of $\pm 600 \text{ mV}/\text{GAIN}$ with $V_{REF} = 1.2\text{V}$.

The maximum absolute voltage, with respect to A_{GND} , for each $CHn\pm$ input pin is $\pm 1\text{V}$ with no distortion and $\pm 2\text{V}$ with no breaking after continuous voltage. This maximum absolute voltage is not proportional to the V_{REF} voltage.

3.5 Analog Ground (A_{GND})

A_{GND} is the ground reference voltage for the analog circuitry within the MCP3910. For optimal performance, it is recommended to connect it to the same ground node voltage as D_{GND} , again preferable with a star connection.

If an analog ground plane is available, it is recommended that these pins be tied to this plane of the PCB. This plane should also reference all other analog circuitry in the system.

3.6 Non-Inverting Reference Input, Internal Reference Output ($REFIN+/OUT$)

This pin is the non-inverting side of the differential voltage reference input for all ADCs or the internal voltage reference output.

When $VREFEXT = 1$, an external voltage reference source can be used, and the internal voltage reference is disabled. When using an external differential voltage reference, it should be connected to its V_{REF+} pin. When using an external single-ended reference, it should be connected to this pin.

When $VREFEXT = 0$, the internal voltage reference is enabled and connected to this pin through a switch. This voltage reference has minimal drive capability and thus needs proper buffering and bypass capacitances (a 0.1 μF ceramic capacitor is sufficient in most cases), if used as a voltage source.

If the voltage reference is only used as an internal V_{REF} , adding bypass capacitance on $REFIN+/OUT$ is not necessary for keeping ADC accuracy, but a minimal 0.1 μF ceramic capacitance can be connected to avoid EMI/EMC susceptibility issues due to the antenna created by the $REFIN+/OUT$ pin if left floating.

3.7 Inverting Reference Input ($REFIN-$)

This pin is the inverting side of the differential voltage reference input for all ADCs. When using an external differential voltage reference, it should be connected to its V_{REF-} pin. When using an external single-ended voltage reference, or when $VREFEXT = 0$ (default) and using the internal voltage reference, the pin should be directly connected to A_{GND} .

3.8 Digital Ground Connection (D_{GND})

D_{GND} is the ground reference voltage for the digital circuitry within the MCP3910. For optimal performance, it is recommended to connect it to the same ground node voltage as A_{GND} , again preferable with a star connection.

If a digital ground plane is available, it is recommended that these pins be tied to this plane of the Printed Circuit Board (PCB). This plane should also reference all other digital circuitry in the system.

3.9 Modulator Outputs (MDAT0/MDAT1)

MDAT0 and MDAT1 are the output pins for the modulator serial bitstreams of ADC channels 0 and 1, respectively. These pins are high-impedance when the EN_MDAT bit is logic low. When the EN_MDAT bit is enabled, the modulator bitstream of each channel is present on the MDAT0/1 output pins and updated at the AMCLK frequency (see [Section 5.3.5 “Modulator Output Block”](#) for a complete description of the modulator outputs). These pins can be directly connected to an MCU or a DSP when a specific digital filtering is needed. When MDAT output pins are enabled, the \overline{DR} output is disabled. In 2-Wire Interface mode, these pins are automatically inactive. Their state is high-impedance during this 2-Wire Interface mode (therefore these pins can be left grounded in applications using exclusively this mode; this configuration improves the EMI/EMC susceptibility of the device).

3.10 Data Ready Output/GAIN1 Logic Input (DR/GAIN1)

In SPI mode, the data ready output pin indicates if a new conversion result is ready to be read. The default state of this pin is logic high when $\overline{DR_HIZ} = 1$, and is high-impedance when $\overline{DR_HIZ} = 0$ (default). After each conversion is finished, a logic low pulse will take place on the Data Ready pin to indicate the conversion result is ready as an interrupt. This pulse is synchronous with the master clock and has a defined and constant width.

The data ready pin is independent of the SPI interface and acts like an interrupt output. The Data Ready pin state is not latched, and the pulse width (and period) are both determined by the MCLK frequency, over-sampling rate, and internal clock prescale settings. The data ready pulse width is equal to half a DMCLK period and the frequency of the pulses is equal to DRCLK (see [Figure 1-3](#)).

In 2-Wire Interface mode, this is the GAIN1 logic select input pin (see [Section 7.0 “2-Wire Serial Interface Description”](#) for the logic input table for GAIN0 and GAIN1). The pin state is latched when the MODE changes to 2-Wire Interface mode, and is relatched at each watchdog timer reset.

Note: This pin should not be left floating when the DR_HIZ bit is low; a 100 k Ω pull-up resistor connected to DV_{DD} is recommended.

3.11 Crystal Oscillator/Master Clock Input/GAIN0 logic Input Input Pin (OSC1/CLKI/GAIN0)

In SPI mode, OSC1/CLKI and OSC2 provide the master clock for the device. When CLKEXT = 0, a resonant crystal or clock source with a similar sinusoidal waveform must be placed across the OSC1 and OSC2 pins to ensure proper operation.

The typical clock frequency specified is 4 MHz. For proper operation, and for optimizing ADC accuracy, AMCLK should be limited to the maximum frequency defined in [Table 5-2](#) for the function of the BOOST and PGA setting chosen. MCLK can take larger values as long as the prescaler settings (PRE<1:0>) limit AMCLK = MCLK/PRESCALE in the defined range in [Table 5-2](#). Appropriate load capacitance should be connected to these pins for proper operation.

In 2-Wire Interface mode, this is the GAIN0 logic select input pin (see [Section 7.0 “2-Wire Serial Interface Description”](#) for the logic input table for GAIN0 and GAIN1). The pin state is latched when the MODE changes to 2-Wire Interface mode, and is relatched at each watchdog timer reset.

Note: When CLKEXT = 1, the crystal oscillator is disabled. OSC1 becomes the master clock input CLKI, a direct path for an external clock source. One example would be a clock source generated by an MCU.

3.12 Crystal Oscillator Output/Interface MODE Logic Input (OSC2/MODE)

When CLKEXT = 0 (default), a resonant crystal or clock source with a similar sinusoidal waveform must be placed across the OSC1 and OSC2 pins to ensure proper operation. Appropriate load capacitance should be connected to these pins for proper operation.

When CLKEXT = 1 (default condition at POR), this pin is the MODE selection pin for the digital interface. When MODE is logic low, the SPI interface is selected (see [Section 6.0 “SPI Serial Interface Description”](#)) when MODE is logic high, the 2-Wire Interface (see [Section 7.0 “2-Wire Serial Interface Description”](#)) is selected. The MODE input is latched after a POR, a Master Reset and/or a watchdog timer reset.

3.13 Chip Select/ Boost Logic Input (CS/BOOST)

In SPI mode, this pin is the SPI chip select that enables serial communication. When this pin is logic high, no communication can take place. A chip select falling edge initiates serial communication, and a chip select rising edge terminates the communication. No communication can take place even when \overline{CS} is logic low, if RESET is also logic low.

This input is Schmitt-triggered.

In 2-Wire Interface mode, this is the BOOST logic select input pin (see [Section 7.0 “2-Wire Serial Interface Description”](#) for the logic input table for BOOST). The pin state is latched when the MODE changes to 2-Wire Interface mode, and is relatched at each watchdog timer reset.

3.14 Serial Data Clock/ Master Clock Input (SCK/MCLK)

In SPI mode, this is the serial clock pin for SPI communication. Data is clocked into the device on the rising edge of SCK. Data is clocked out of the device on the falling edge of SCK.

The MCP3910 SPI interface is compatible with SPI 0,0 and 1,1 modes. SPI modes can be changed during a CS high time.

The maximum clock speed specified is 20 MHz. SCK and MCLK are two different and asynchronous clocks; SCK is only required when a communication happens, while MCLK is continuously required when the part is converting analog inputs.

This input is Schmitt-triggered.

In the 2-Wire Interface mode, this pin is defining the master clock of the device (MCLK) and simultaneously as the serial clock (SCK) for the interface. In this mode, the clock has to be provided continuously to ensure proper operation. See [Section 7.0 “2-Wire Serial Interface Description”](#) for more information and timing diagrams of the 2-Wire Interface protocol.

3.15 Serial Data Output (SDO)

This is the SPI data output pin. Data is clocked out of the device on the falling edge of SCK.

This pin remains in a high-impedance state during the command byte. It also stays high-impedance during the entire communication for write commands and when the \overline{CS} pin is logic high, or when the RESET pin is logic low. This pin is active only when a Read command is processed. The interface is half-duplex (inputs and outputs do not happen at the same time).

In the 2-Wire Interface mode, this pin is the only digital output pin, and sends synchronous frames at each data ready event with data bits clocked out on the falling edge of SCK.

3.16 Serial Data/OSR1 Logic Input (SDI/OSR1)

This is the SPI data input pin. Data is clocked into the device on the rising edge of SCK. When \overline{CS} is logic low, this pin is used to communicate with a series of 8-bit commands. The interface is half-duplex (inputs and outputs do not happen at the same time).

Each communication starts with a chip select falling edge followed by an 8-bit command word, entered through the SDI pin. Each command is either a Read or a Write command. Toggling SDI after a Read or a Write command when \overline{CS} is logic high has no effect.

This input is Schmitt-triggered.

In 2-Wire Interface mode, this is the OSR1 logic select input pin (see [Section 7.0 “2-Wire Serial Interface Description”](#) for the logic input table for OSR0 and OSR1). The pin state is latched when the MODE changes to 2-Wire Interface mode, and is relatched at each watchdog timer reset.

3.17 Exposed Pad (EP)

This pin is the exposed thermal pad. It must be connected to A_{GND} for optimal accuracy and thermal performance. This pad can also be left floating if necessary. Connecting it to A_{GND} is preferable for the lowest noise performance and best thermal behavior.

4.0 TERMINOLOGY AND FORMULAS

This section defines the terms and formulas used throughout this data sheet. The following terms are defined:

- **MCLK – Master Clock**
- **AMCLK – Analog Master Clock**
- **DMCLK – Digital Master Clock**
- **DRCLK – Data Rate Clock**
- **OSR – Oversampling Ratio**
- **Offset Error**
- **Gain Error**
- **Integral Non-Linearity Error**
- **Signal-to-Noise Ratio (SNR)**
- **Signal-To-Noise Ratio And Distortion (SINAD)**
- **Total Harmonic Distortion (THD)**
- **Spurious-Free Dynamic Range (SFDR)**
- **MCP3910 Delta-Sigma Architecture**
- **Idle Tones**
- **Dithering**
- **Crosstalk**
- **PSRR**
- **CMRR**
- **ADC Reset Mode**
- **Hard Reset Mode (RESET = 0)**
- **ADC Shutdown Mode**
- **Full Shutdown Mode**

4.1 MCLK – Master Clock

This is the fastest clock present on the device. This is the frequency of the crystal placed at the OSC1/OSC2 inputs when CLKEXT = 0 or the frequency of the clock input at the OSC1/CLKI when CLKEXT = 1. In the 2-Wire mode, this is the frequency present at the SCK input pin. See [Figure 4-1](#).

4.2 AMCLK – Analog Master Clock

AMCLK is the clock frequency that is present on the analog portion of the device, after prescaling has occurred via the CONFIG0 PRE<1:0> register bits (see [Equation 4-1](#)). The analog portion includes the PGAs and the two delta-sigma modulators.

EQUATION 4-1:

$$AMCLK = \frac{MCLK}{PRESCALE}$$

TABLE 4-1: MCP3910 OVERSAMPLING RATIO SETTINGS

CONFIG0		Analog Master Clock Prescale
PRE<1:0>		
0	0	AMCLK = MCLK/1 (default)
0	1	AMCLK = MCLK/2
1	0	AMCLK = MCLK/4
1	1	AMCLK = MCLK/8

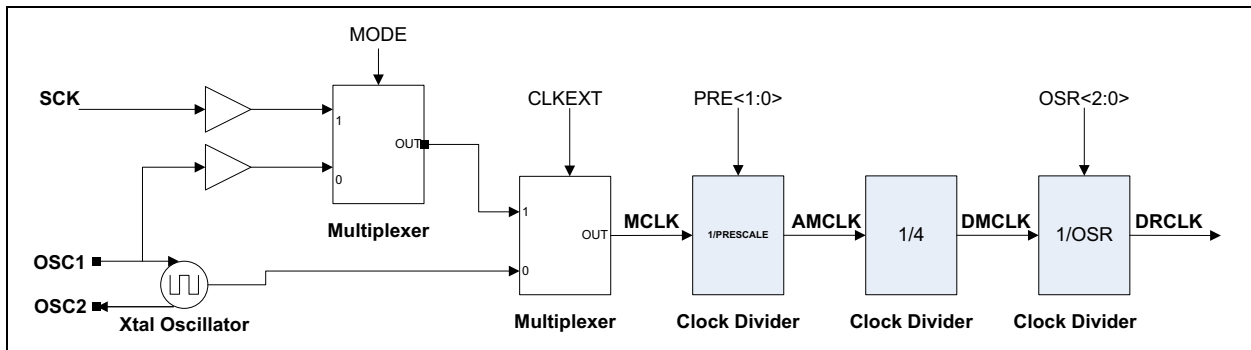


FIGURE 4-1: Clock Sub-Circuitry.

4.3 DMCLK – Digital Master Clock

This is the clock frequency that is present on the digital portion of the device, after prescaling and division by four ([Equation 4-2](#)). This is also the sampling frequency, which is the rate at which the modulator outputs are refreshed. Each period of this clock corresponds to one sample and one modulator output. See [Figure 4-1](#).

EQUATION 4-2:

$$DMCLK = \frac{AMCLK}{4} = \frac{MCLK}{4 \times PRESCALE}$$

4.4 DRCLK – Data Rate Clock

This is the output data rate, i.e. the rate at which the ADCs output new data. Each new data is signaled by a data ready pulse on the \overline{DR} pin.

This data rate is depending on the OSR and the prescaler with the formula in [Equation 4-3](#).

EQUATION 4-3:

$$DRCLK = \frac{DMCLK}{OSR} = \frac{AMCLK}{4 \times OSR} = \frac{MCLK}{4 \times OSR \times PRESCALE}$$

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Since this is the output data rate, and because the decimation filter is a sinc (or notch) filter, there is a notch in the filter transfer function at each integer multiple of this rate.

Table 4-2 describes the various combinations of OSR and PRESCALE, and their associated AMCLK, DMCLK and DRCLK rates.

TABLE 4-2: DEVICE DATA RATES IN FUNCTION OF MCLK, OSR AND PRESCALE, MCLK = 4 MHZ

PRE<1:0>		OSR<2:0>			OSR	AMCLK	DMCLK	DRCLK	DRCLK (ksp/s)	SINAD (dB) Note 1	ENOB from SINAD (bits) Note 1
1	1	1	1	1	4096	MCLK/8	MCLK/32	MCLK/131072	.035	102.5	16.7
1	1	1	1	1	2048	MCLK/8	MCLK/32	MCLK/65536	.061	100	16.3
1	1	1	1	1	1024	MCLK/8	MCLK/32	MCLK/32768	.122	97	15.8
1	1	1	1	1	512	MCLK/8	MCLK/32	MCLK/16384	.244	96	15.6
1	1	0	1	1	256	MCLK/8	MCLK/32	MCLK/8192	0.488	95	15.5
1	1	0	1	0	128	MCLK/8	MCLK/32	MCLK/4096	0.976	90	14.7
1	1	0	0	1	64	MCLK/8	MCLK/32	MCLK/2048	1.95	83	13.5
1	1	0	0	0	32	MCLK/8	MCLK/32	MCLK/1024	3.9	70	11.3
1	0	1	1	1	4096	MCLK/4	MCLK/16	MCLK/65536	.061	102.5	16.7
1	0	1	1	1	2048	MCLK/4	MCLK/16	MCLK/32768	.122	100	16.3
1	0	1	1	1	1024	MCLK/4	MCLK/16	MCLK/16384	.244	97	15.8
1	0	1	1	1	512	MCLK/4	MCLK/16	MCLK/8192	.488	96	15.6
1	0	0	1	1	256	MCLK/4	MCLK/16	MCLK/4096	0.976	95	15.5
1	0	0	1	0	128	MCLK/4	MCLK/16	MCLK/2048	1.95	90	14.7
1	0	0	0	1	64	MCLK/4	MCLK/16	MCLK/1024	3.9	83	13.5
1	0	0	0	0	32	MCLK/4	MCLK/16	MCLK/512	7.8125	70	11.3
0	1	1	1	1	4096	MCLK/2	MCLK/8	MCLK/32768	.122	102.5	16.7
0	1	1	1	1	2048	MCLK/2	MCLK/8	MCLK/16384	.244	100	16.3
0	1	1	1	1	1024	MCLK/2	MCLK/8	MCLK/8192	.488	97	15.8
0	1	1	1	1	512	MCLK/2	MCLK/8	MCLK/4096	.976	96	15.6
0	1	0	1	1	256	MCLK/2	MCLK/8	MCLK/2048	1.95	95	15.5
0	1	0	1	0	128	MCLK/2	MCLK/8	MCLK/1024	3.9	90	14.7
0	1	0	0	1	64	MCLK/2	MCLK/8	MCLK/512	7.8125	83	13.5
0	1	0	0	0	32	MCLK/2	MCLK/8	MCLK/256	15.625	70	11.3
0	0	1	1	1	4096	MCLK	MCLK/4	MCLK/16384	.244	102.5	16.7
0	0	1	1	0	2048	MCLK	MCLK/4	MCLK/8192	.488	100	16.3
0	0	1	0	1	1024	MCLK	MCLK/4	MCLK/4096	.976	97	15.8
0	0	1	0	0	512	MCLK	MCLK/4	MCLK/2048	1.95	96	15.6
0	0	0	1	1	256	MCLK	MCLK/4	MCLK/1024	3.9	95	15.5
0	0	0	1	0	128	MCLK	MCLK/4	MCLK/512	7.8125	90	14.7
0	0	0	0	1	64	MCLK	MCLK/4	MCLK/256	15.625	83	13.5
0	0	0	0	0	32	MCLK	MCLK/4	MCLK/128	31.25	70	11.3

Note 1: For OSR = 32 and 64, DITHER = None. For OSR = 128 and higher, DITHER = Maximum. The SINAD values are given from GAIN = 1.

4.5 OSR – Oversampling Ratio

This is the ratio of the sampling frequency to the output data rate. $OSR = DMCLK/DRCLK$. The default OSR is 256, or with $MCLK = 4\text{ MHz}$, $PRESCALE = 1$, $AMCLK = 4\text{ MHz}$, $f_S = 1\text{ MHz}$, and $f_D = 3.90625\text{ kpsps}$. The bits in [Table 4-3](#) in the CONFIG0 register are used to change the oversampling ratio (OSR).

TABLE 4-3: MCP3910 OVERSAMPLING RATIO SETTINGS

CONFIG0			Oversampling Ratio (OSR)
OSR<2:0>			
0	0	0	32
0	0	1	64
0	1	0	128
0	1	1	256 (Default)
1	0	0	512
1	0	1	1024
1	1	0	2048
1	1	1	4096

4.6 Offset Error

This is the error induced by the ADC when the inputs are shorted together ($V_{IN} = 0V$). The specification incorporates both PGA and ADC offset contributions. This error varies with PGA and OSR settings. The offset is different on each channel and varies from chip to chip. The offset is specified in μV . The offset error can be digitally compensated independently on each channel through the OFFCAL registers with a 24-bit calibration word.

The offset on the MCP3910 has a low-temperature coefficient.

4.7 Gain Error

This is the error induced by the ADC on the slope of the transfer function. It is the deviation expressed in %, compared to the ideal transfer function defined in [Equation 5-3](#). The specification incorporates both PGA and ADC gain error contributions, but not the V_{REF} contribution (it is measured with an external V_{REF}).

This error varies with PGA and OSR settings. The gain error can be digitally compensated independently on each channel through the GAINCAL registers with a 24-bit calibration word.

The gain error on the MCP3910 has a low temperature coefficient.

4.8 Integral Non-Linearity Error

Integral non-linearity error is the maximum deviation of an ADC transition point from the corresponding point of an ideal transfer function, with the offset and gain errors removed, or with the end points equal to zero.

It is the maximum remaining error after calibration of offset and gain errors for a DC input signal.

4.9 Signal-to-Noise Ratio (SNR)

For the MCP3910 ADCs, the signal-to-noise ratio is a ratio of the output fundamental signal power to noise power (not including the harmonics of the signal), when the input is a sine wave at a predetermined frequency (see [Equation 4-4](#)). It is measured in dB. Usually, only the maximum signal-to-noise ratio is specified. The SNR figure depends mainly on the OSR and DITHER settings of the device.

EQUATION 4-4: SIGNAL-TO-NOISE RATIO

$$SNR(dB) = 10\log\left(\frac{SignalPower}{NoisePower}\right)$$

4.10 Signal-To-Noise Ratio And Distortion (SINAD)

The most important Figure of Merit for the analog performance of the ADCs present on the MCP3910 is the Signal-to-Noise and Distortion (SINAD) specification.

The Signal-to-Noise and Distortion ratio is similar to Signal-to-Noise ratio, with the exception that you must include the harmonics power in the noise power calculation (see [Equation 4-5](#)). The SINAD specification depends mainly on the OSR and DITHER settings.

EQUATION 4-5: SINAD EQUATION

$$SINAD(dB) = 10\log\left(\frac{SignalPower}{Noise + HarmonicsPower}\right)$$

The calculated combination of SNR and THD per the following formula also yields SINAD (see [Equation 4-6](#)).

EQUATION 4-6: SINAD, THD AND SNR RELATIONSHIP

$$SINAD(dB) = 10\log\left[10^{\left(\frac{SNR}{10}\right)} + 10^{\left(\frac{-THD}{10}\right)}\right]$$